

DELFT UNIVERSITY OF TECHNOLOGY

MASTER THESIS

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Author:

Serwan Asaad

Supervisor:

Dr. Alessandro Bruno

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Contents

I	Deep-reactive ion etched resonators	4
II	Muxmon experiment	6
1	Muxmon chip architecture	9
2	Qubit characterization	11
2.1	Continuous-wave measurements	11
2.1.1	Scanning for resonators	12
2.1.2	Powersweeping the resonators	13
2.1.3	Scan for qubit sweet-spots	14
2.1.4	Scanning for qubits	15
2.1.4.1	Spectroscopy	16
2.1.4.2	Avoided crossing	17
2.1.5	Tracking the qubits	18
2.1.6	Flux matrix	19
2.2	Time-domain measurements	20
2.2.1	Qubit control	20
2.2.2	Calibrating drive amplitude	22
2.2.3	Tuning qubit frequency	22
2.2.4	T1 measurement	22
2.2.5	Echo measurement	22
2.2.6	Motzoi parameter	22
2.2.7	AllXY measurement	22
2.2.8	Measuring single shots	22
3	Calibration routines	23
3.1	Mixer calibration	23
4	Randomized benchmarking	24
4.1	Second excited state	24
	Appendices	25

A	Additional notes	26
A.1	Qubit characterization	26
A.1.1	Part I: Continuous-wave measurements	26
A.1.1.1	Powersweep	26
A.1.1.2	Spectroscopy	26
A.1.2	Part I: Time-domain measurements	26
A.1.2.1	AllXY	26
B	Algorithms	27
B.1	Tracked spectroscopy	27
B.2	Peak finding	27
B.2.1	Normal peak finder	27
B.2.2	Derivative peak finder	27
B.3	Compiled RB?	27

INTRODUCTION

Part I

Deep-reactive ion etched resonators

TODO:

- Explain heterodyne detection
- Explain VNA
- Explain that resonance frequency f_r is not necessarily at the transmission minimum when the resonator exhibits asymmetry.

Part II

Muxmon experiment

large TODO:

- Explain surface code architecture
- Explain the general structure of a chip:
 - Feedline through which signal is sent and measured
 - CPW resonators, capacitively coupled to feedline
 - Single qubit coupled to resonators (coupling location?)
 - Resonator buses

small TODO:

- Rename DAC voltage to flux, and mention early on that this renaming will be used
- flux-bias line or flux bias line?

Introduction

At this moment circuit QED is at the stage where multi-qubit experiments are being realized.

Chapter 1

Muxmon chip architecture

Topics that should be explained in this section:

- The Muxmon0 and Muxmon1 chip are designed with two purposes
 1. Testing multiplexing using the Duplexer
 2. Explore qubit frequency re-use
- Explain similarities of chips
 - Three qubits per chip
 - All three qubits have individual flux tuning
 - Air bridges are used, not only for connect the ground planes, but also such that the feed line can pass over other coplanar waveguides without contact
- Explain differences between Muxmon0 and Muxmon1.
 - The Muxmon0 chip has a driving line connected to each of the qubits. It has two resonator buses at 4.9 GHz and 5.0 GHz. These could also be used for two-qubit gates.

Advantage Able to fully control each qubit individually, even when multiple qubits share the same frequency.

Advantage Less coupling between data qubits.

Disadvantage Requires more driving lines.

Disadvantage Adds extra source of dissipation for the qubits.
 - The Muxmon1 chip has two driving lines, each capacitively coupled to one of the two data qubits, and to the ancilla qubit.

Advantage Less driving lines required

Advantage Less dissipation due to capacitive coupling

Disadvantage Cannot individually control data qubit and ancilla qubit when they share the same frequency

Disadvantage More coupling between qubits

– Simplified model of the surface code

- Explain concepts of cross-coupling and readout cross-talk

Cross-coupling The coupling between qubits.

Cross-coupling leads to transfer of excitation.

An associated coupling strength g can be associated to cross-coupling.

Can be determined by driving one qubit extremely hard, and measuring signal from other qubit.

TODO: Show values of cross-coupling found, or do this in characterization section

TODO: Leads to coherent errors?

TODO: Two types of cross-coupling? Direct leakage of pulse pulse, and transfer of excitation? cross-driving?

Readout cross-talk Coupling between a qubit and a resonator that are not directly coupled.

A part of the signal measured from one resonator is then due to the state of another qubit

TODO: Understand more behind readout cross-talk

Left to think about:

- Should I already include items such as coherence times, the fact that Muxmon0 performs better than Muxmon 1?
- Where should I include coherence times versus frequency?
- Should the part on cross-coupling and readout cross-talk not be in characterization section?
- Should the section on the Duplexer go in here?

Figures that need to be included:

- Muxmon0 and Muxmon1 chip, preferably optical microscopy
- SEM image of air-bridges such that coplanar wave-guides cross without intersecting
- schematic of cross-coupling and readout cross-talk
It could be good to create this using the actual Muxmon chip as background, with arrows indicating how the different effects operate

Chapter 2

Qubit characterization

Description This chapter gives a step-by-step description of how to find a resonator and qubit, and subsequently how to tune the qubit's parameters.

In the design of cQED chips, the parameters of the qubits and resonators are always targeted which are ideal for the experiment. For coplanar waveguide resonators one can already obtain relatively good parameters for the required dimensions from simple formulae **TODO:** refer to formula. For superconducting transmon qubits, however, finding the right dimensions that correspond to the desired parameters is a much more complicated process. The qubit's frequency, for instance, depends on the qubit's coupling energy E_c and Josephson energy E_J . The coupling energy E_c can be reasonably estimated from classical simulations. Finding the right dimensions for the Josephson junction that result in the desired Josephson energy E_J , however, is difficult, and usually physically testing different junctions is necessary to determine an accurate conversion from the desired E_J to the Josephson junction dimensions.

Nevertheless, the actual parameters of the resonators and qubits are almost never where one expects them to be. Once the sample is cooled in the dilution refrigerator, an inevitable game of hide-and-seek follows with the goal of finding the frequency of the resonators and qubits, and subsequently determining their properties. This chapter describes the measurements that were performed to characterize the MuxMon samples.

2.1 CONTINUOUS-WAVE MEASUREMENTS

Once a sample is properly cooled down it is ready to be measured. At this stage the sample is still an unknown terrain, where the experimenter only has a rough map, containing the sample's targeted parameters, and the specific properties of the resonators and qubits.

The first step is to look for signs of life. These manifest themselves as resonance frequencies of the resonators and the qubits that are coupled to them. As we are not yet interested in the properties of the resonators and qubits which can only be obtained through measurements with accurately timed pulses, we send continuous tones through the feedline, and measure deviations in the transmission. These measurements are known as continuous-wave measurements

2.1.1 Scanning for resonators

Since communication with the qubits is mediated through their coupling to resonators, the first step is to find these resonators. This is done using a transmission measurement, in combination with heterodyne detection, and has been explained in section **TODO**: Create section in Resonator chapter.

There is one difference in measuring a resonator when there is a qubit coupled to it. When considering the qubit as a two-level system, the behaviour of the coupled resonator-qubit system is governed by the Jaynes-Cummings Hamiltonian [1]:

$$\hat{H} = \hbar\omega_r \left(\hat{a}^\dagger \hat{a} + \frac{1}{2} \right) + \frac{\hbar\omega_q}{2} \hat{\sigma}_z + \hbar g \left(\hat{a}^\dagger \sigma_- + \hat{a} \sigma_+ \right) \quad (2.1)$$

where ω_r is the bare resonance frequency of the resonator, ω_q is the resonance frequency of the qubit's ground to excited state transition, and the qubit's two states are in the spin-representation. This Hamiltonian consists of three terms. The first term corresponds to the energy level of the resonator, the second to the energy level of the transmon, and the third is a coupling term between the two with coupling strength g .

The difference between the resonator's frequency ω_r and the qubit's frequency ω_q is given by the detuning $\Delta = \omega_q - \omega_r$. If the magnitude of the detuning is large compared to the coupling strength g , the system is in the dispersive regime. In this case the Hamiltonian can be approximated by the dispersive Jaynes-Cummings Hamiltonian:

$$\hat{H} = \frac{\hbar\omega'_q}{2} \hat{\sigma}_z + \left(\hbar\omega'_r + \hbar\chi \hat{\sigma}_z \right) \hat{a}^\dagger \hat{a} \quad (2.2)$$

The coupling between the qubit and resonator causes both qubit's frequency and the resonator's frequency to shift: $\omega'_q = \omega_q + \chi_{01}$, $\omega'_r = \omega_r - \chi_{12}/2$.

Aside from experiencing a frequency shift dependent on the amount of detuning, Equation 2.2 shows that the resonator also experiences a shift depending on the state of the qubit. The resonator's frequency is decreased by an amount 2χ when the qubit is in the excited state. The parameter χ is the dispersive shift, and is given by:

$$\chi = \chi_{01} - \chi_{12}/2 \approx \frac{g^2}{\Delta} \frac{E_c}{\hbar\Delta - E_c} \quad (2.3)$$

where $\chi_{ij} = \frac{g_{ij}^2}{\omega_{ij} - \omega_c}$ are the partial dispersive shifts.

Due to this coupling between resonator and qubit, it is important to choose the right RF power. When the amount of photons in the resonator reaches a certain point, this coupling will result in the resonator experiencing nonlinear effects. The resonator will thereby lose its Lorentzian lineshape. Therefore the RF power should be kept sufficiently low to avoid these nonlinear effects, while still maintaining a good signal-to-noise ratio.

TODO:

- In the strong coupling regime (Leads to hybridization of qubit and resonator states: quton and fobit)

- Explain that the quality factor is low because the resonator is coupled strongly to the qubit (equation including coupling to qubit?)
- Doesn't χ diverge when $\Delta \rightarrow E_c$?
- explain concept of anharmonicity, and that $\alpha \approx E_c$
- Maybe explain concept of number splitting in dispersive Jaynes-Cummings. Number splitting is the phenomenon that the qubit's frequency shifts by an amount 2χ for every photon in the resonator.
Alternatively mention this in another section.
- Mention that $g_{12} = \sqrt{g}$, and that other coupling strengths are exponentially suppressed in the transmon

Figures:

- Figure of transmission showing all three Muxmon0 resonators

2.1.2 Powersweeping the resonators

Once the resonators have been located, the next stage is to find the qubit that is capacitively coupled to each of the resonators. Instead of directly scanning the entire frequency spectrum in search of the qubit, it is relatively straightforward to perform some initial measurements aimed at gaining information about our resonator and qubit, which will allow us to search for our qubit with much greater accuracy.

As explained previously, the capacitive coupling between the resonator and qubit shifts the resonator frequency ω_r from its bare frequency. When the amount of photons in the resonator reaches a certain point, the resonator experiences nonlinearity, thereby losing its Lorentzian lineshape. When increasing the RF power even further, at a certain point the resonator regains its Lorentzian lineshape. In doing so its resonance frequency has shifted to its bare frequency ω_{bare} . If this frequency shift is observed, it indicates that the resonator's frequency was shifted, and hence that the qubit is alive. Measuring this frequency shift is commonly done in a powersweep. A powersweep is a measurement in which a resonator scan is performed for a range of powers.

A powersweep additionally provides information about at what power the resonator enters the nonlinear regime. For measurements involving the qubit the readout power must be below this threshold power. Furthermore, from the frequency shift between the dressed cavity frequency and the bare cavity frequency, the amount of detuning between the qubit and the resonator can be estimated using Equation 2.3.

If no shift is observed, it could mean that the qubit is dead (e.g. because the Josephson junction is shorted). However, this is not necessarily the case. An alternative possibility is that the detuning between qubit and resonator is very large, and as a result the frequency shift cannot be discerned. At this point it is too early to draw conclusions, and we may almost draw the analogy with Schrödinger's cat in a box.

TODO:

- Explain theory behind transition to bare cavity frequency (Reed's thesis has some information)

Figures:

- Powersweep of ancilla qubit

2.1.3 Scan for qubit sweet-spots

Some of the qubits have a tunable resonance frequency. This is done through a superconducting quantum interference device (SQUID). In this case the two islands that compose the transmon qubit are connected by two Josephson junctions instead of one, effectively forming a loop. The SQUID loop is sensitive to the amount of flux passing through the loop. The amount of flux going through the SQUID loop can be changed by changing the surrounding magnetic field. This is commonly done by having a flux bias line in close proximity to the SQUID loop. Current flowing through the flux-bias line alters the magnetic field in the vicinity of the SQUID loop, and hence changes the amount of flux through the SQUID loop. A digital-to-analog converter is used to specify the amount of current that is sent through the flux bias line. Depending on the amount of flux through the SQUID loop, the resonance frequency of the qubit changes accordingly. These qubits are therefore called flux-tunable.

For qubits that are flux-tunable, finding the sweet-spot of the qubit can be done without knowledge of the qubit's frequency. This can be done by sweeping the DAC voltage and measuring the shift in the resonance frequency. Because the frequency of the qubit varies as the amount of current through the flux-bias line changes, the detuning between the qubit and the resonator consequently changes. As a result the dispersive shift χ , and therefore the resonator's frequency, also varies. At the sweet-spot of the qubit, the resonator's frequency ω_r is at a maximum. This is irrespective of whether the qubit's frequency ω_q is above or below the resonator's frequency.

The accurate way to measure the sweet-spot is to perform resonator scans as the DAC voltage is varied. The result is a 2D scan shown in **TODO: Figure**.

A faster second approach for finding the qubit sweet-spot, at the cost of providing less information, is by choosing a fixed frequency close to the resonator's frequency ω_r (preferably slightly below, where the transmission slope is steepest). By measuring the amount of transmission as the DAC voltage is being varied, one obtains essentially a line-cut of **TODO: Figure**. The idea this measurement is that if the qubit's frequency f_q decreases, the resonator's frequency also decreases, resulting in a decrease in transmission (closer to ω_r). Likewise, if the qubit's frequency ω_q increases, the resonator's frequency increases **TODO: Why?**, resulting in an increase in transmission (further away from ω_r).

At the qubit's sweet-spot, the resonator's frequency ω_r is at a maximum, and so the transmission should also be at a maximum. Furthermore, because the amount of detuning only depends on the deviation from the flux sweet-spot **TODO: improve**, the transmission should be symmetric with respect to the DAC voltage sweet-spot. If the resonator's frequency ω_r shifts by a large amount in the course of this measurement, it becomes harder to determine where the sweet-spot is (although even then often it can still be discerned). Nevertheless,

this method is considerably faster than performing a full two-dimensional scan of frequency versus DAC voltage, and in most cases it works like a charm.

In the case where the powersweep showed no measurable frequency shift, these two measurements are also useful in discerning whether or not the qubit is actually dead, or whether it was simply far detuned from the resonator.

TODO:

- Explain how qubit's frequency has a cosine dependence on DAC
- Give detailed information on SQUID loop
 - Why does the qubit frequency change in a SQUID loop
 - Sweet-spot
- Mention flux-noise?
 - $1/f$ noise
 - usually not limiting, as it is very slow
 - This noise can be seen as occasional jumps (every few hours?) It would mean that every few hours the frequency must be recalibrated.

Figures:

- SEM picture of SQUID loop, including flux-bias line
- Figure of 2D resonator scan vs DAC voltage
- Figure of 1D resonator fixed frequency DAC voltage scan

2.1.4 Scanning for qubits

Once the preliminary measurements have been performed that characterize the resonators and provide hints about the whereabouts of the qubit it is time to actually find the qubit.

The measurement to perform in order to find the qubit depends on the amount of detuning between the resonator and qubit, which can be estimated from powersweep measurements. If the amount of detuning is large compared to the coupling strength ($\Delta \gg g$), the system is in the dispersive regime. In this case one commonly performs a two-tone spectroscopy to find the qubit's frequency. If, on the other hand, the detuning is comparable to the coupling strength ($\Delta \sim g$), the qubit and resonator are hybridized, and experience an avoided crossing. In such cases a normal transmission measurement suffices.

If the qubit is flux-tunable, one important question to ask is: at what DAC voltage should the qubit be searched?

One option is to choose zero DAC voltage. In most cases this is alright, but there are cases when this is a bad choice. For instance, trapped magnetic fields during cool-down could position the qubit near the anti-sweet-spot for zero DAC voltage, in which case finding the qubit will be next to impossible.

A better option is to choose a DAC voltage that results in a detuning Δ that is large compared to coupling strength g , such that the system is in the dispersive regime. However, the detuning Δ should not be too high, as it would otherwise result in a negligible dispersive shift χ . The optimum value is usually around a few hundred megahertz, although this is dependent on the device-specific parameters. The detuning can be approximated using powersweeps.

2.1.4.1 Spectroscopy

As was explained in section 2.1.1, in the dispersive regime the resonator experiences a 2χ frequency shift dependent on the state of the qubit. In the case of a resonator capacitively coupled to the feedline, the transmission experiences a dip at the resonator's frequency. This dip will correspondingly shift when the qubit's state is switched. The transmission at the resonator's dip when the qubit is in the ground state will therefore be dependent on the state of the qubit (low when the qubit is in the ground state, high when the qubit is in the excited state). This is the property exploited in a two-tone spectroscopy measurement.

In a two-tone spectroscopy measurement two tones are sent through the feedline.

1. A drive tone with varying frequency ω_d .
2. A measurement tone at the resonator's frequency when the qubit is in the ground state.

The transmission through the feedline is measured while the frequency of the drive tone ω_d is swept. When the drive frequency ω_d is detuned from the qubit's frequency ω_q , the drive is off-resonant with respect to the qubit, and so we measure a low transmission due to a dip being present. However, when the drive frequency ω_d approaches the qubit's frequency ω_q , then the qubit will start to oscillate between its ground and excited state, with a rate dependent on the detuning between the drive frequency ω_d and the qubit's frequency ω_q . The qubit will therefore have a partial population in the excited state, resulting in a shift of the resonator frequency, dependent on the population in the excited state. This is measured as an increase in transmission.

The minimum linewidth of the qubit using spectroscopy is set by its dephasing time T^2 , which can be seen as the uncertainty in its frequency. However, the power of the drive tone causes an additional increase in the linewidth, due to stimulated emission of the qubit. This effect is known as power broadening. This effect can be quite useful for finding qubits, especially when designing high-quality qubits with a very narrow intrinsic linewidth. The optimal power for the drive strength is further dependent on the amount of detuning Δ between the qubit and resonator. The resonator effectively acts as a bandpass filter, centered around the resonator frequency. Therefore, the stronger the detuning, the more the drive tone is suppressed.

The dispersive shift 2χ is also dependent on the amount of detuning Δ between the qubit and resonator. If the detuning Δ is large, the dispersive shift will be very small, and so the difference in transmission will also decrease. Since the resonator has a Lorentzian lineshape, at the resonance frequency this Lorentzian is flat, and so is insensitive to small deviations. To increase the contrast between the on-resonant and off-resonant transmission, it is usually

advantageous to measure at a slight detuning δ away from the resonance frequency, where the transmission slope is high. Since the resonator frequency shifts down when the qubit is excited, it is better to have a positive detuning δ to ensure that the transmission increases as the drive frequency ω_d approaches the qubit frequency ω_q and decreases as it leaves the qubit frequency.

TODO: Spectroscopy section

- Is power broadening really due to stimulated emission?
- Explain optimal RF power
- Is the gradual increase really due to partial population? Is it then an average of the partial population? Or do you get other effects as well?

2.1.4.2 Avoided crossing

When the detuning Δ between the qubit and resonator is not large compared to their coupling strength g , the qubit and resonator experience an avoided crossing. Because all of the qubits in the Muxmon chips are considerably lower in frequency compared to their resonators, the system is always in the dispersive regime, and so the system never approaches the avoided crossing. Nevertheless it is worth mentioning the avoided crossing briefly, as it is a breeding ground for interesting physics.

At the avoided crossing the system is not in the dispersive regime, and so the dispersive Jaynes-Cummings Hamiltonian no longer applies. In this case the resonator and qubit are hybridized, and one can no longer speak of a qubit and resonator as separate entities. Instead, this hybridization results in a so-called quton and phobit.

Avoided crossing info:

- The coupling strength g is the minimum distance between the splitting
- From Reed's thesis p.63 explanation of this avoided crossing is given

$$E_0 = -\frac{\hbar\Delta}{2} \quad (2.4)$$

$$E_1 = n\hbar\omega_r \pm \frac{\hbar}{2}\sqrt{4g^2n + \Delta^2} \quad (2.5)$$

Joining $E_0 + E_1$ results in a qubit approaching a resonator from the top.

When the frequency of the qubit equals that of the resonator, the energy difference reaches a minimum, and is equal to $2g$.

TODO: Avoided crossing

- Explain the quton phobit behaviour near the avoided crossing
- Explain how one can extract coupling from avoided crossing

- Explain that for the Muxmon qubit this effect is not present
- Explain multiple lines near avoided crossing

figures:

- Figure of avoided crossing

TODO:

- Use relatively high source power, results in power broadening
- Mention what to do if qubit is close to or far away from resonator
- Mention the term transmission measurement earlier on.
- Pulsed spectroscopy

2.1.5 Tracking the qubits

For flux-tunable qubits one is usually not interested in finding the frequency at one specific flux value. More important is finding how this frequency changes with varying flux. A naive approach would be to perform a two-dimensional scan of a fixed frequency range versus flux. At a certain flux the qubit crosses the boundary of the chosen frequency window. Every time this happens a new two-dimensional scan has to be performed with an updated frequency window. The larger the chosen frequency window, the less often this has to be updated. On the other hand, choosing a large window also means that a larger portion of the scan is spent not measuring the qubit. There is therefore a trade-off between measurement time and degree of human intervention. Furthermore, the resonator frequency also depends on the amount of detuning. Therefore the frequency of the measurement tone must also be updated every now and then.

As an alternative approach I have created a modified version of the spectroscopy versus flux scan, known as tracked spectroscopy. In this approach after every spectroscopy scan the qubit frequency is extracted through fitting. From the qubit frequencies measured in previous scans the expected frequency at the next flux value is extrapolated. The frequency window of the next spectroscopy scan are then centered around the expected qubit frequency. The qubit is therefore tracked as its frequency changes. The same method is applied to update the resonator frequency.

Tracked spectroscopy has several advantages over the two-dimensional spectroscopy versus flux scans. First of all, since the qubit's frequency follows a smooth curve (see Equation **TODO**: Refer to eqn), the qubit's frequency can be estimated quite accurately. The frequency windows can therefore be narrow. This greatly reduces the measurement time. Secondly, tracked spectroscopy does not need any human intervention, as the frequency window is automatically updated after every spectroscopy scan. Therefore measuring the flux-dependence of the qubit frequencies is an automatized process (provided that the qubit frequency can correctly be extracted).

For more information on the tracked spectroscopy algorithm see appendix B.1.

Improvements

- Avoided crossing
- Variable RF and drive power

TODO:

- Maybe mention possible improvements

Figures:

- Tracked spectroscopy of qubit and resonator

2.1.6 Flux matrix

When a current is passed through a flux-bias line, it is not only the flux through the SQUID of the qubit directly connected to it that is affected. Instead, the flux through SQUIDS of neighbouring qubits are also affected, albeit to a lesser extent. Therefore changing the frequency of one qubit by changing its corresponding flux-bias line current also affects the frequencies of the other flux-tunable qubits on the chip. The effect that the current passing through flux-bias lines has on the frequencies of the different qubits can be decoupled by correcting the flux cross-coupling effects. This is done by implementing a flux matrix.

A flux matrix F is an $n \times n$ matrix, where n is the number of flux-tunable qubits. Each row corresponds to a virtual flux, and can be interpreted as follows: in a row $F_i = [f_{i1}, \dots, f_{in}]$ each element F_{ij} corresponds to the relative amount by which the flux-bias line of qubit j should be changed, such that only the frequency of qubit i changes, whilst the frequencies of the other qubits remain fixed. This results in n decoupled virtual flux parameters.

Creating a flux matrix is done by first determining for each qubit separately how much every flux-bias line affects its frequency. For a given qubit i this can be found by tuning the qubit away from its sweet-spot using its main flux-bias line, to a point where the slope $\partial f_i / \partial V_i$ of the qubit frequency f_i as a function of DAC voltage V_i is steep. This is the region where we are sensitive to changes in qubit frequency. For the main flux-bias line we already know from tracked spectroscopy what the slope $\partial f_i / \partial V_i$ is at this point. For each of the remaining flux-bias lines we measure the slope $\partial f_i / \partial V_j$ as we vary the DAC voltage V_j and measure the response of the qubit frequency f_i . The slope $\partial f_i / \partial V_j$ is a measure for the amount of flux cross-coupling, which should be considerably less than for the main flux-bias line. Finally dividing all the slopes by the slope $\partial f_i / \partial V_i$ of our main flux-bias line, results in the ratio's of the frequency response of qubit i for each of the flux-bias lines. Performing this measurement for all the qubits results in a matrix M , where each element $M_{ij} = \partial f_i / \partial V_j$ is the normalized frequency response of qubit i when varying the DAC voltage of the flux-bias line corresponding to qubit j . The diagonal elements of matrix M should be equal to one.

Tracked spectroscopy provides accurate information on the DAC voltage corresponding to the sweet-spot of a qubit. The sweet-spot may not lie exactly at zero DAC voltage. It may shift due to magnetic flux being trapped during the cooldown of the fridge. These sweet-spots are usually found in the case where all the remaining flux-bias lines are set to zero DAC voltage. There is a certain combination of DAC voltages \vec{V}^{ss} for which all the qubits are at

their simultaneous sweet-spot. This simultaneous sweet-spot can be found using matrix \mathbf{M} . We know for each qubit i the DAC voltage V_i^0 corresponding to its sweet-spot. Since \mathbf{M} stores the amount by which each flux-bias line affects the flux of any qubit, the qubit i remains at its sweet-spot as long as $\mathbf{M}_i \vec{V} = V_i^0$ holds, where \vec{V} is the vector containing the DAC voltages. Therefore the simultaneous sweet-spot \vec{V}^{ss} can be found by solving the set of equations $\mathbf{M} \vec{V}^{ss} = \vec{V}^0$, where \vec{V}^0 contains the DAC voltages of the individual sweet-spots of the qubits.

The flux matrix \mathbf{F} can be found by inverting matrix \mathbf{M} . As mentioned earlier, each row of matrix \mathbf{F} corresponds to the ratio by which the DAC voltages of the flux-bias lines need to be varied, such that the frequency of only one qubit is changed. We may therefore multiply each row by a different constant, as long as the ratio between the elements in each row stays constant. One can normalizing each row such that the main flux-bias line is equal to one. This has the advantage that there is a one-to-one correspondence between DAC voltage and flux, and as a result the tracked spectroscopy will look identical. Additionally, after the normalization each row can be multiplied by a factor such that the virtual flux is equal to the flux quanta present in the SQUID loop. This conversion factor can be obtained by fitting the qubit frequency curve obtained from tracked spectroscopy.

For a given virtual flux vector $\vec{\Phi} = [\phi_1, \dots, \phi_n]$, the corresponding DAC voltages \vec{V} are given by:

$$\vec{V} = \mathbf{F} \vec{\Phi} + \vec{V}^{ss} \quad (2.6)$$

By adding the simultaneous sweet-spot DAC voltages \vec{V}^{ss} , we obtain the additional property that the sweet-spot of the virtual fluxes is set to zero.

After determining the flux matrix \mathbf{F} , there will still be some small remaining cross-coupling, which depends on the accuracy of the measurements. The process of creating a flux matrix can then be repeated, but instead of using DAC voltages as the varying parameters to construct matrix \mathbf{M} , the virtual fluxes should be used. Furthermore, as the cross-coupling is small compared to before, the flux range can be much greater, such that small slopes can be accurately measured. The resulting flux matrix \mathbf{F}_2 can then simply be multiplied with the first flux matrix \mathbf{F} to obtain a more accurate final flux matrix.

2.2 TIME-DOMAIN MEASUREMENTS

2.2.1 Qubit control

So far the measurements described have all dealt with continuous tones being applied and measuring the response in the transmission. These are crude measurements, that are nevertheless able to determine some properties of the qubit and resonator, such as their frequencies. However, to find more detailed properties of the qubits, such as their coherence times, one cannot apply a continuous tone, that drives some incoherent qubit population. Instead well-calibrated pulses are required, which modify the state of the qubit in a precise manner, corresponding to gates being applied to the qubit. These types of measurements are called time-domain measurements, as they require precise pulse timing.

A simple time-domain measurement usually consists of two parts. The first part consists of controlling the qubit. Here pulses are sent which modify the state of the qubit. The second part consists of measuring the state of the qubit. This is done through a readout tone, similar to spectroscopy measurements. This readout tone projects the state of the qubit onto the Z-axis. From the response in the transmission the state of the qubit can be inferred. More complicated experiments may involve some sort of feedback loop, where additional qubit control can be applied depending on the measurement outcome. These measurements require the qubit readout to be quantum non-demolition, i.e. the state of the qubit is not destroyed in the process of qubit readout. This usually involves extremely low-noise amplification, such as through a Josephson parametric amplifier. Feedback measurements are not performed in this experiment, and are therefore beyond the scope of this thesis.

Qubit pulses usually have a Gaussian shape. These pulses can be generated by modulating a carrier signal from an RF generator, and is commonly done using an Arbitrary Waveform Generator (AWG). For the Muxmon experiment the Tektronix AWG5014 is used, which has four channels, each of which can control its voltage output at the nanosecond scale. It further has eight marker channels, which can be used to trigger devices, such as RF generators and the Duplexer. The carrier signal is sent through the LO port of the mixer, and an AWG channel is connected to the IF port of the mixer. The output at the RF port of the mixer is the modulated signal, the amplitude of which depends on the amplitude of the AWG channel output.

For IQ mixers it is also possible to modulate both quadratures of the carrier signal independently, using two AWG channels. This allows for single sideband modulation, which is done by convoluting the pulses of the I and Q quadrature channels with a sine and cosine respectively. The sideband modulation frequency ω_{sb} of these sinusoidal functions is the amount by which the carrier signal is shifted. Single sideband modulation has the advantage that the carrier frequency ω_c is shifted away from mixer leakage, which would otherwise cause a slight continuous rotation of the qubit. For more information on mixer leakage, see section 3.1.

The length of the pulse is an important consideration. Shorter pulses allow for more qubit operations within its coherence time. On the other hand, the shorter a pulse is in time, the larger its frequency spectrum will be. At a certain point the spectrum will be so broad that there will be a nonnegligible signal at the qubit's excited-state to second excited-state transition frequency. This will cause leakage to the second excited-state, thereby leaving the two-state Hilbert space. Furthermore, the finite resolution of the AWG causes the pulses to be discretized, resulting in the pulses being slightly distorted. The shorter the length of the pulse, the more discretization, and therefore distortion, will occur. This can cause further leakage to the second excited-state. It is therefore important that the inverse of the width σ of the Gaussian pulse is small compared to the anharmonicity of the qubit.

Nevertheless, pulses

TODO:

- Explain more on Z projection
- Explain more on non-demolition.
- Signal quadrature determines the rotation angle

- Mention that AWG channel amplitude must not be too high, and that attenuation should be used to avoid spurious resonant modes
- Refer to Noise chapter for more information on single sideband modulation?
- Mention that amount of rotation depends on time duration, refer to Reed's thesis

Questions:

- Are both quadratures modulated independently in IQ mixers?
- Why use sideband modulation?
- Why are Gaussian pulses used?

Figures:

- Gaussian pulse with derivative

2.2.2 Calibrating drive amplitude

- time variation or amplitude variation

2.2.3 Tuning qubit frequency

2.2.4 T1 measurement

2.2.5 Echo measurement

2.2.6 Motzoi parameter

2.2.7 AllXY measurement

2.2.8 Measuring single shots

- * []Explanation of gates
- * []Motzoi parameter
- * []Calibrating drive amplitude using Rabi
- * []Calibrating frequency using Ramsey
- ** Conflict with next section on accurate tune-up?
- * []Measuring decoherence
- ** Relaxation
- ** T1 measurement
- * Dephasing
- ** Ramsey measurement (all noise)
- * Echo measurement (fast noise)
- * Change with frequency
- ** Ideally T_2 is $2 \cdot T_1$
- * T_2 rapidly decreases as moving away from sweet-spot
- * []Single-shot measurements
- ** Able to see relaxation and thermal excitation
- * []Second excited state
- ** Similar to three-tone spectroscopy
- * Nearly identical position in IQ-plane as first excited state, but less relaxation into ground state
- * Since time-trace is averaged, the final signal is therefore further away from the ground state in the IQ-plane
- * []Crosstalk / cross-coupling?

Chapter 3

Calibration routines

3.1 MIXER CALIBRATION

Chapter 4

Randomized benchmarking

4.1 SECOND EXCITED STATE

Appendices

Appendix A

Additional notes

A.1 QUBIT CHARACTERIZATION

A.1.1 Part I: Continuous-wave measurements

A.1.1.1 Powersweep

- If the resonator seems to disappear after the transition from dressed frequency to bare frequency (or vice versa), this is likely due to the qubit being extremely close to the resonator, and so the frequency shift is large.

A.1.1.2 Spectroscopy

- If the deviation in transmission becomes less due to more detuning, increasing the power can also increase the contrast.

A.1.2 Part I: Time-domain measurements

A.1.2.1 AllXY

TODO:

- Table of AllXY pulses

Appendix B

Algorithms

B.1 TRACKED SPECTROSCOPY

- Information may lie in amplitude or in phase, or in a combination of the two. For most accurate measure, calculate distance from ground state.

B.2 PEAK FINDING

B.2.1 Normal peak finder

B.2.2 Derivative peak finder

B.3 COMPILED RB?

Bibliography

- [1] Matthew D. Reed. *Entanglement and Quantum Error Correction with Superconducting Qubits*, year =. PhD thesis.