



**Department of Electrical & Electronic Engineering**

**Brac University**

**Semester- Fall 2021**

**Course Number: EEE 342**

**Course Title: Introduction to Communication Engineering  
Laboratory**

## **Lab Report**

**Section: 1**

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**Final Project**

***Group Number: 06***

***Other Group members:***

<b><i>SL</i></b>	<b><i>ID</i></b>	<b><i>Name</i></b>
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## **Task No-1**

### **Objective:**

The primary objective of this analysis in the initial part (task-1) of the project was to design a communication link for AM audio transmission. Firstly, we will trigger a message signal, then modulate it with a carrier, and then add noise to the modulated signal using an AWGN (Additive White Gaussian Noise) channel. Then we will need to demodulate this modulated signal in order to recover our original message signal.

### **Required Blocks & Components:**

- Sine Wave
- Analog Filter Design
- Band Limited White Noise
- Zero Order Hold
- Spectrum Analyzer
- Scope
- Adder
- Product
- Gain
- Constant

### **Methodology:**

- We used a Sine wave as message signal, sampled at  $1.8^{-6}$  with an amplitude of 1 and a frequency of 1000 Hz. Then, with the modulation index set to 0.5, we inserted a gain block. Afterwards, there's a carrier signal, which is a Sine wave with an amplitude of 17.88 and a frequency of 1350000 Hz sampled at  $1.8^{-6}$ . This particular amplitude was chosen to reduce SNR in this case.
- We demodulated our modulated signal by multiplying a sync carrier with our modulated signal and then reconstructing our basic message signal with a Lowpass bandpass filter. We've also added a spectrum analyzer scope to see how the message signal, carrier signal, and zero order hold affect each other. The ability to visualize the message signal, analog filter, carrier signal, and reconstructed signal has been visualized through scope.

01

$$SNR = 40 \text{ dB}$$

$$\mu = 0.5$$

$$f_m = 1 \text{ kHz}$$

$$f_c = 1.35 \text{ MHz}$$

$$RF \text{ Band width} = 9 \text{ kHz}$$

channel noise power spectral density,  $N_0 =$

$$-150 \text{ dBm/Hz}$$

$$= 1 \times 10^{-18}$$

$$\text{watt/Hz}$$

Let's assume,

$$\text{channel Noise} = 120$$

$$f_m = 1 \text{ kHz}$$

$$= 1000 \text{ Hz}$$

We know,

$$SNR = 10 \log_{10} \left( \frac{P_s}{P_n} \right)$$

$$40 = 10 \log \left( \frac{P_s}{P_n} \right)$$

$$10^4 = \frac{(\mu^2 A c_o^2)/4}{2 N_0 f_m}$$

$$P_s = \frac{\mu^2 A c^2}{4}$$

$$P_n = N_0 B_m$$

$$= N_0 \times 2 \times f_m$$

$$A c_o = \sqrt{\frac{2 \times N_0 \times f_m \times 10^4 \times 4}{\mu^2}}$$

$$= 1.789 \times 10^{-5}$$

$$\text{Power loss, } P_{\text{loss}} = \frac{n^2 A_{\text{eff}}^2}{(0.5)^2 \times (1.789 \times 10^{-5})^2}$$

$$= 2 \times 10^{-11} \text{ W}$$

from channel loss

$$120 = 10 \log_{10} \frac{P_{\text{in}}}{P_{\text{out}}}$$

$$10^{12} = \frac{P_{\text{in}}}{P_{\text{out}}}$$

$$\therefore P_{\text{in}} = P_{\text{out}} \times 10^{12}$$

$$= 2 \times 10^{11} \times 10^{12}$$

$$= 20$$

$$\therefore A_{\text{eff}} = \sqrt{\frac{P_{\text{in}} \times 4}{n^2}}$$

$$= \sqrt{\frac{20 \times 4}{(0.5)^2}}$$

$$= \cancel{44.7} 17.88 \text{ m}^2$$

channel Bandwidth =  $2^{\infty}$ -fm

$$\therefore g = 2^{\infty} \text{ fm}$$

$$f_m = \frac{g}{2}$$

$$f_m = 4.5 \text{ kHz}$$

if  $f_m = 200 \text{ Hz}$

$$\therefore S/N = 10 \log_{10} \left( \frac{P_s}{P_N} \right)$$

$$= 10 \log_{10} \left\{ \frac{(0.5)^2 \times (1.789 \times 10^{-5})^2}{2 \times 10^{-18} \times 200} \right\}$$

$$= 10 \log_{10} 56 \times 10^3$$

$$= 46.99$$

if  $\mu_0 = 1$ ,  $f_m = 1 \text{ kHz}$

$$\text{SNR} = 10 \log_{10} \left( \frac{\frac{\mu^2 A c^2}{4}}{2 N_0 f_m} \right)$$

$$= 10 \log_{10} \left( \frac{(1)^2 \times (1.789 \times 10^{-5})^2}{4 \times 10^{-18} \times 10^3} \right)$$

$$= 10 \log_{10} 40 \times 10^3$$

$$= 46.02$$

$\mu = 1.5$ ,  $f_m = 1 \text{ kHz}$

$$\text{SNR} = 10 \log_{10} \left( \frac{\frac{\mu^2 A c^2 / 4}{2 \times N_0 f_m}}{1} \right)$$

$$= 10 \log_{10} \left( \frac{(1.5)^2 \times (1.78 \times 10^{-5})^2}{4 \times 10^{-18} \times 10^3} \right)$$

$$= 49.49$$

if  $f_m = 4.5 \text{ kHz}$

$$= 4500 \text{ Hz}$$

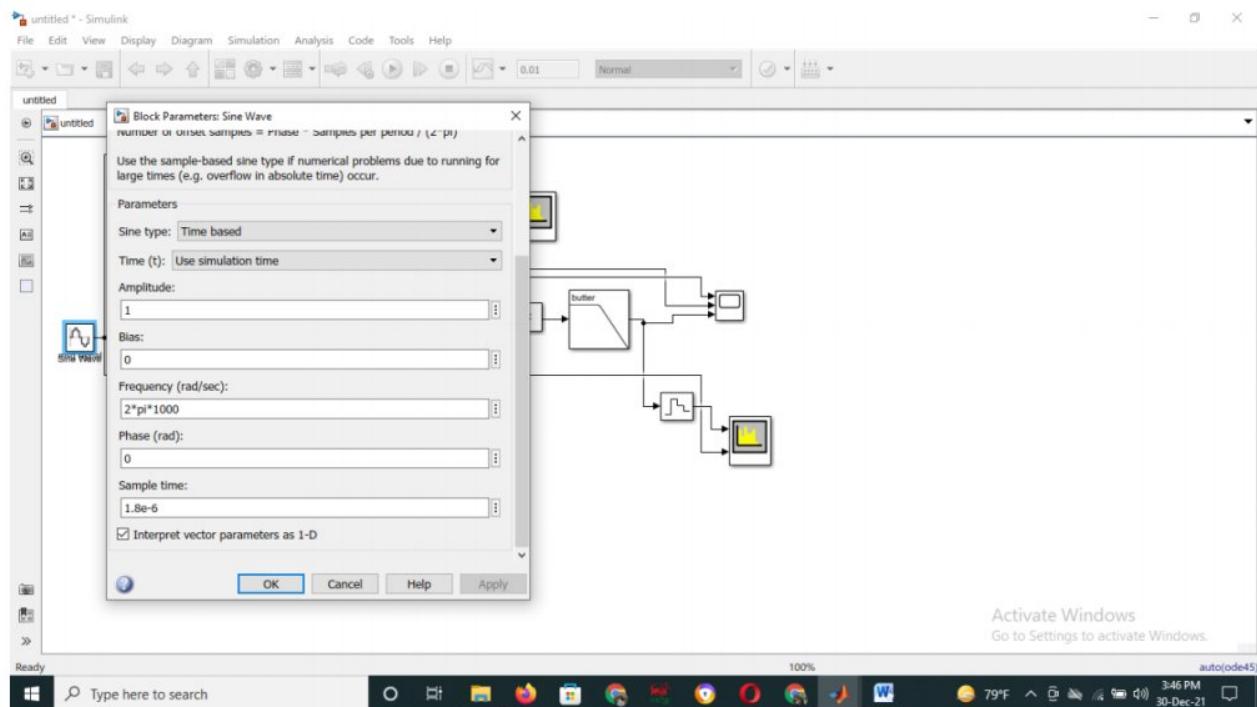
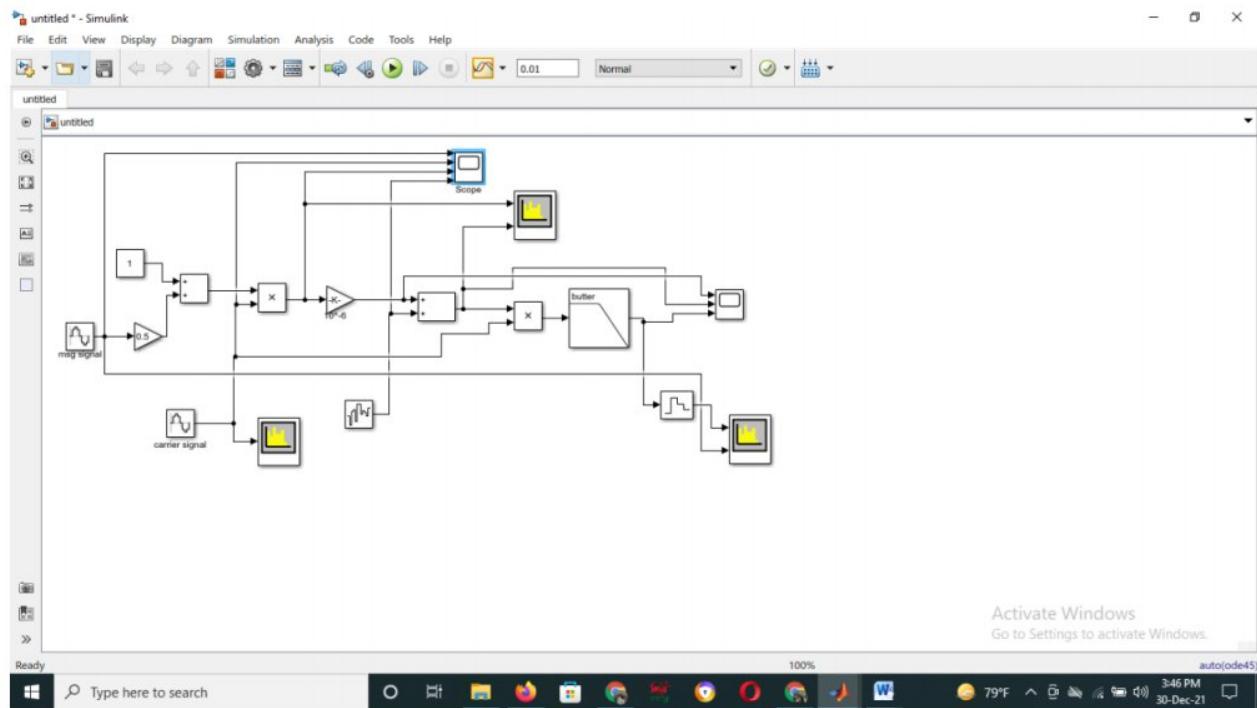
$$\text{SNR} = 10 \log_{10} \left( \frac{P_s}{P_d} \right)$$

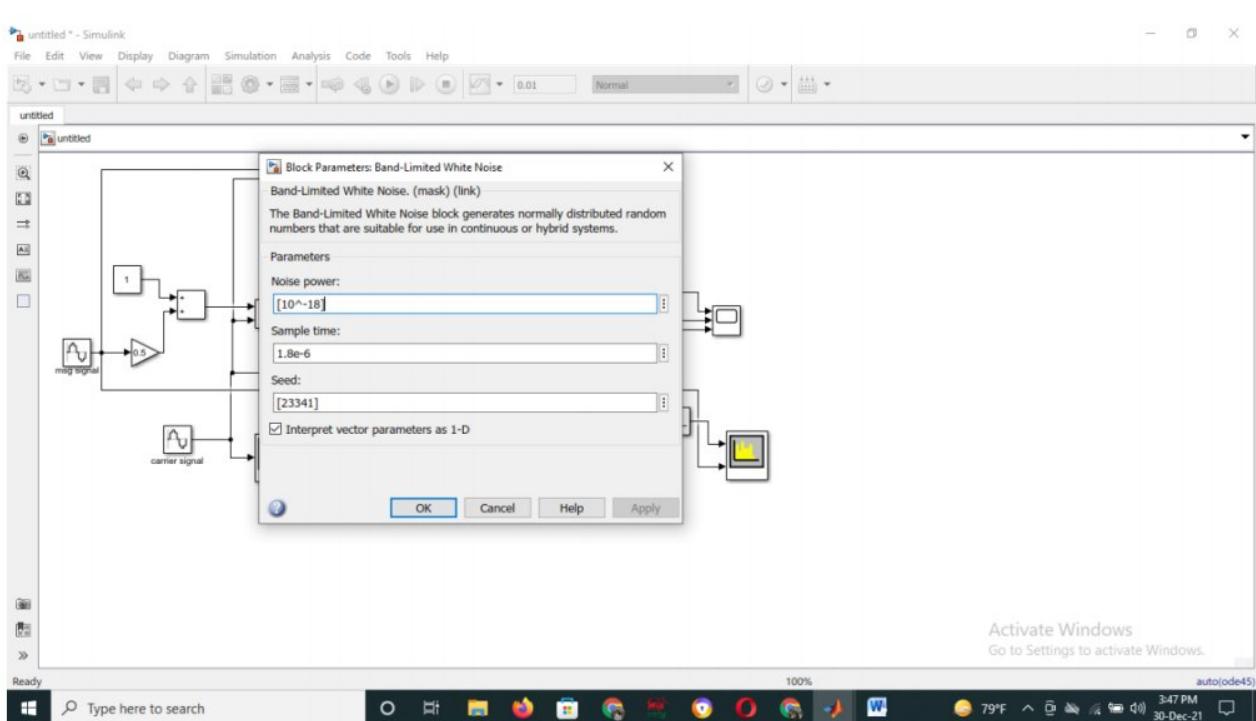
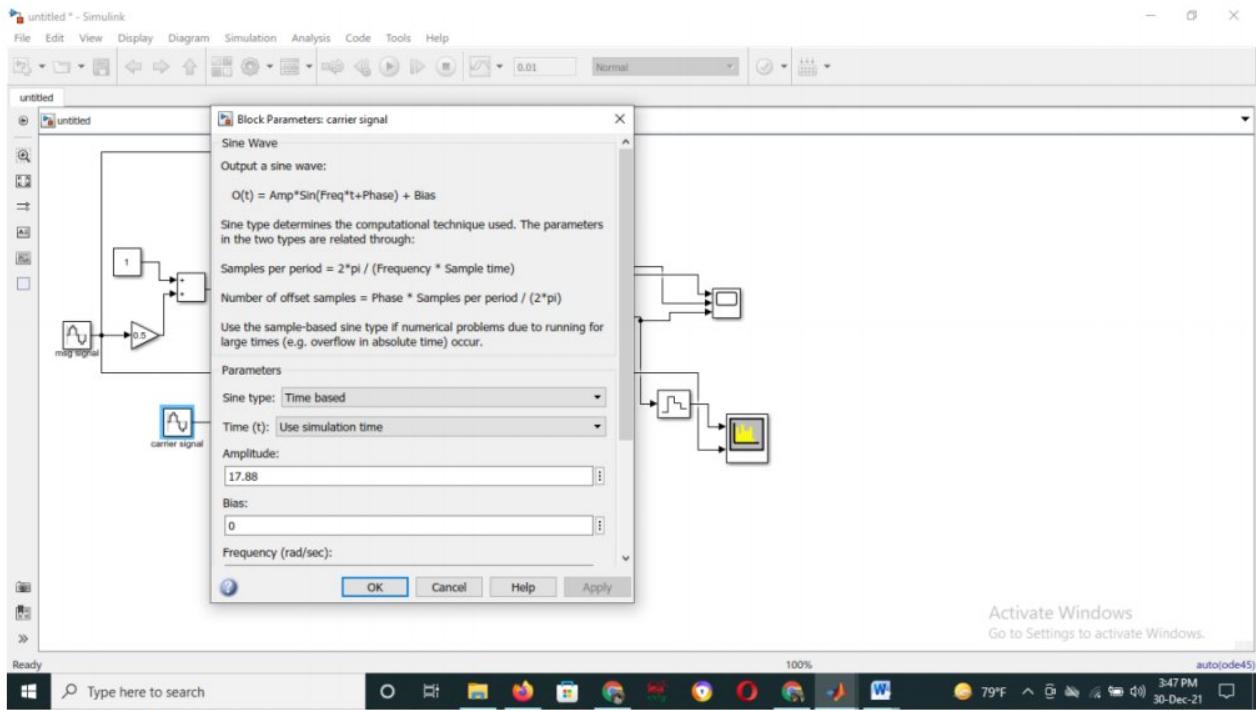
$$= 10 \log_{10} \left\{ \frac{(0.5)^e \times (1.789 \times 10^{-5})^2}{4 \times 10^{-18} \times 4500} \right\}$$

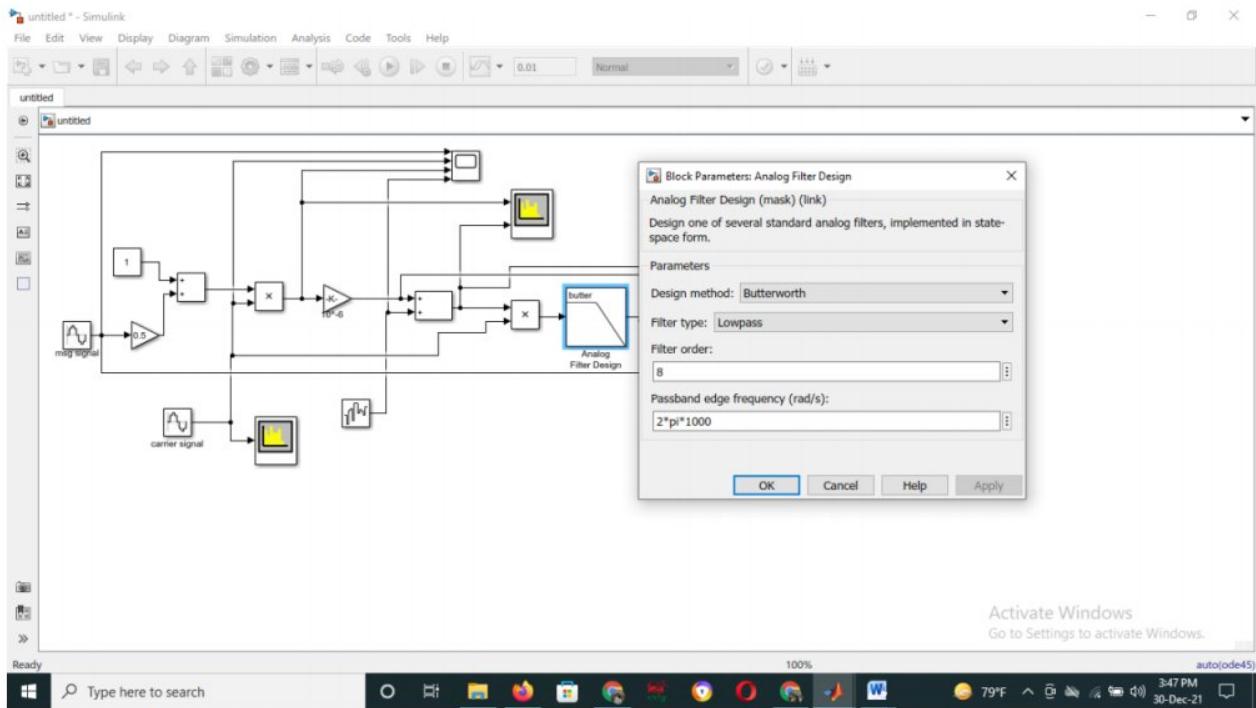
$$= 10 \log_{10} 2222.58$$

$$= 33.47$$

## Block Diagram :

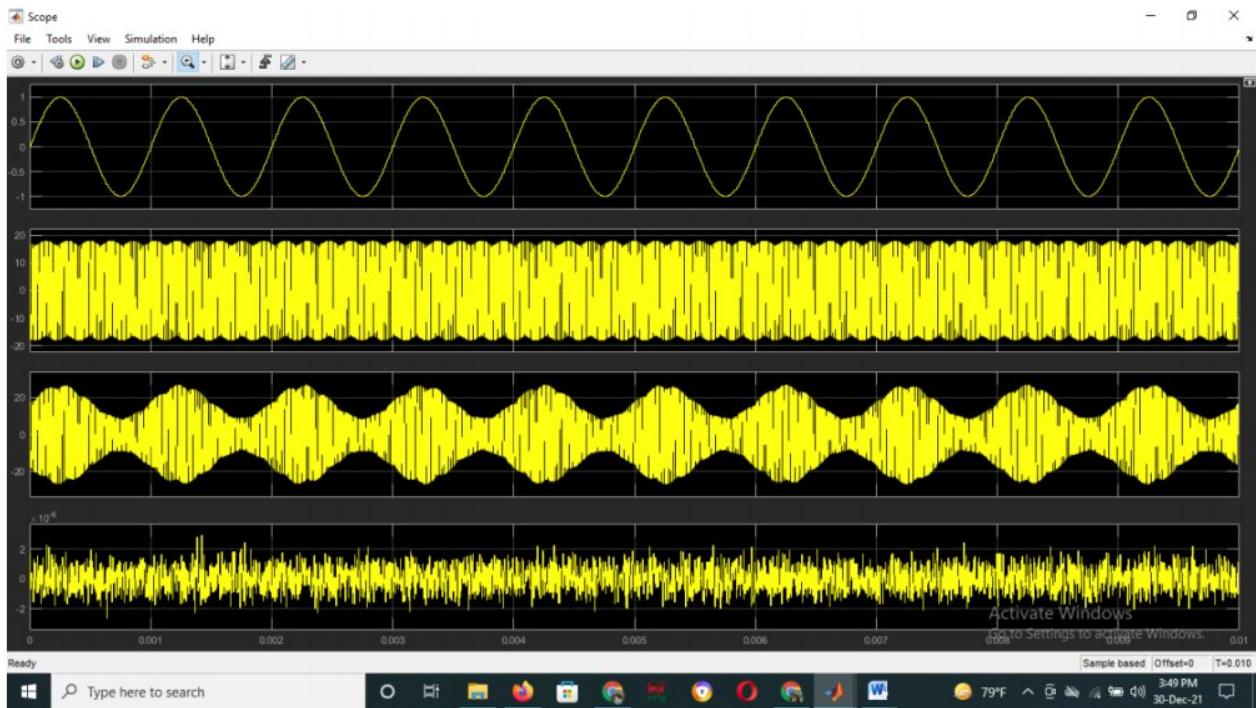




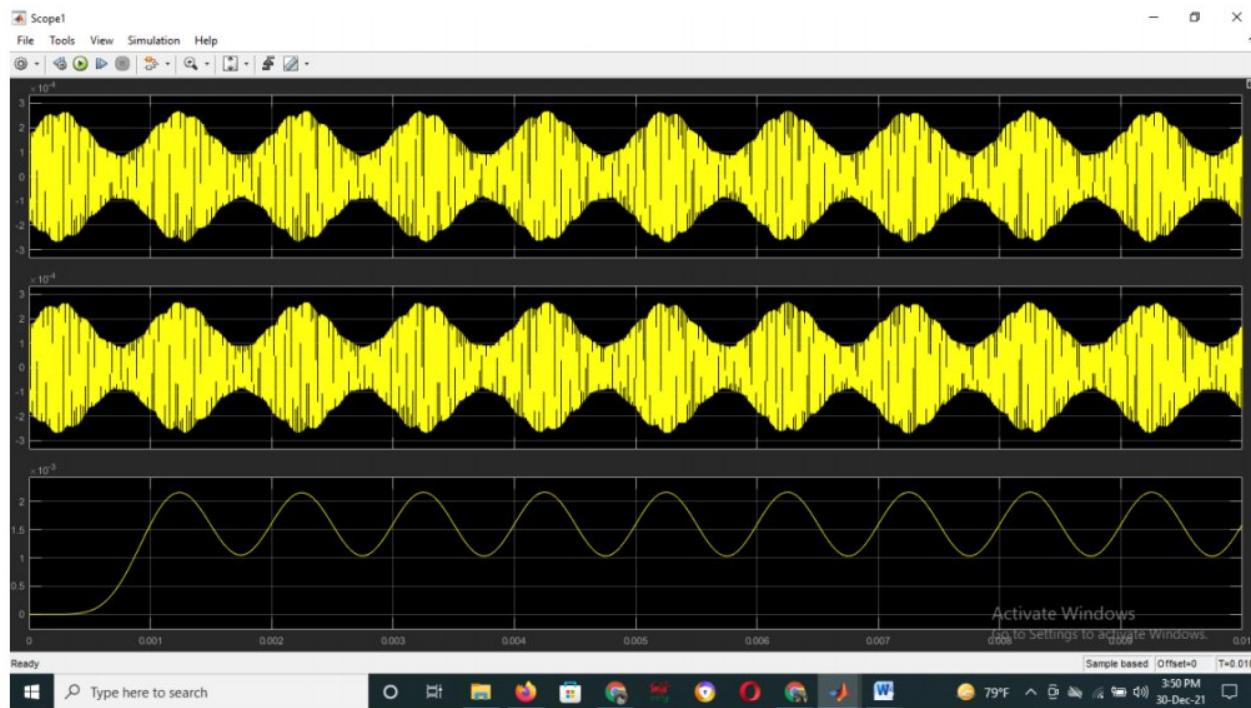


## Results:

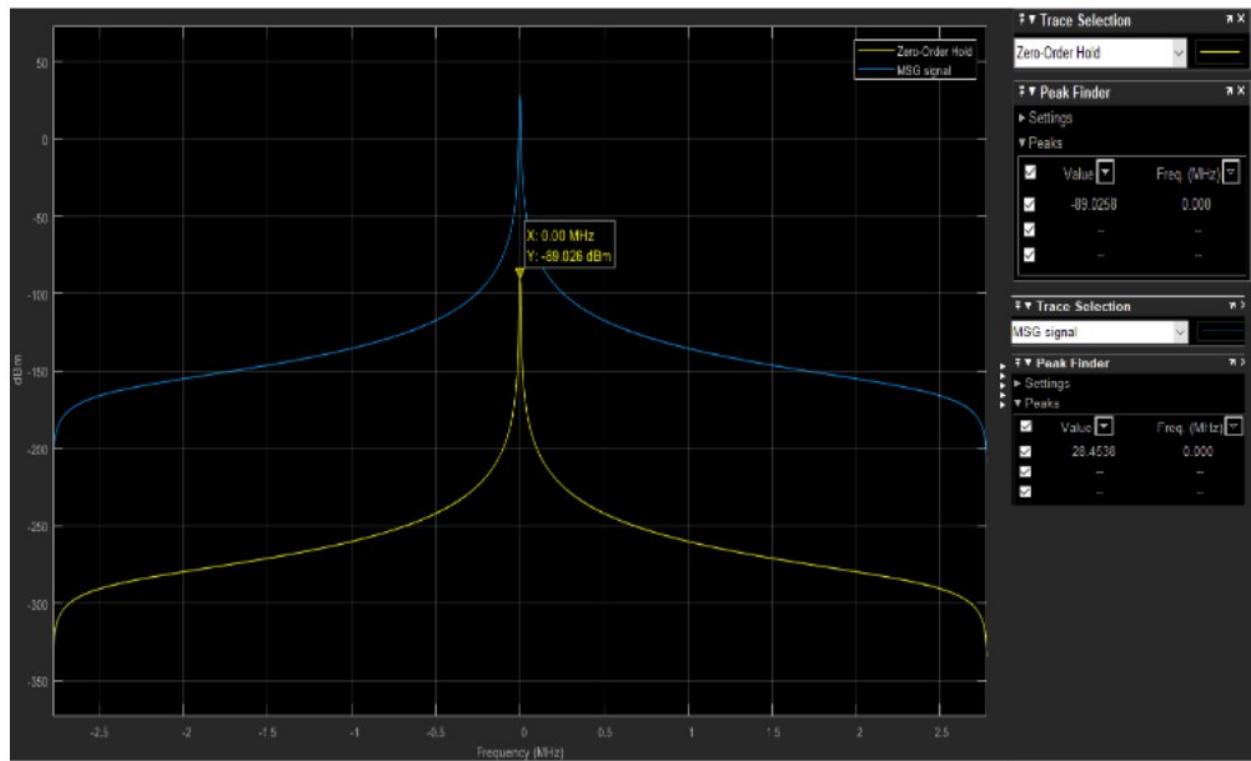
waveform of message, carrier, Modulated signal and noise:



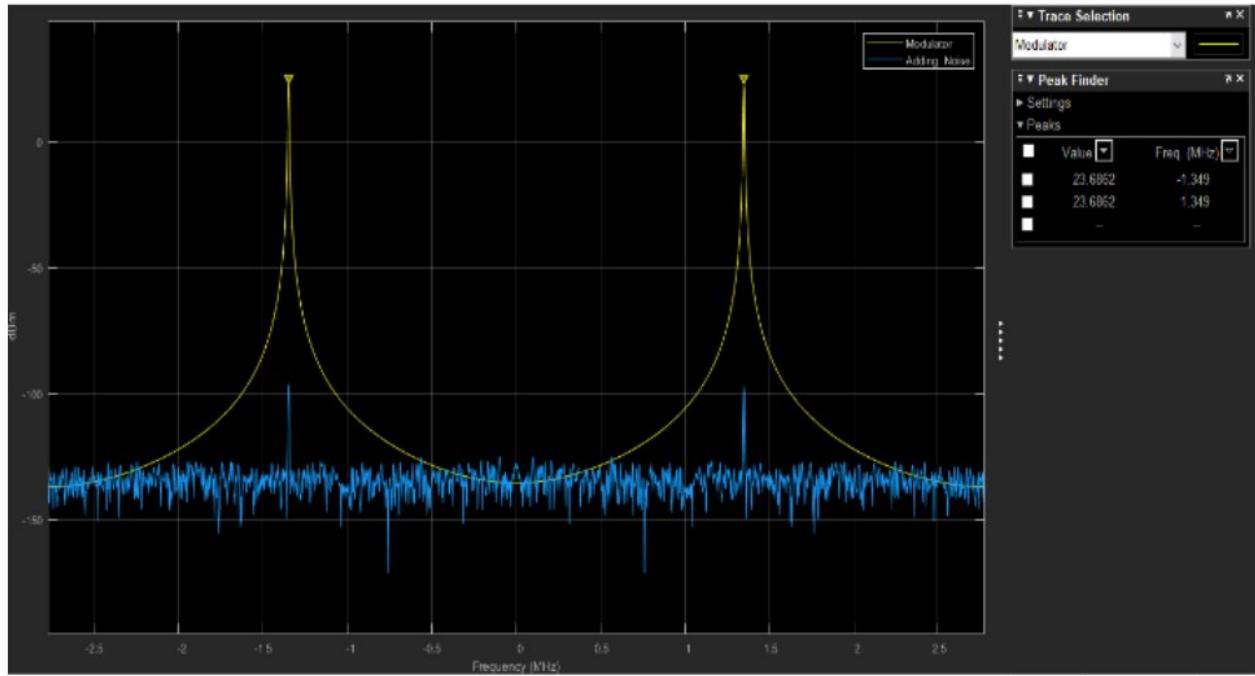
waveform of channel loss, added noise and demodulated filtered signal



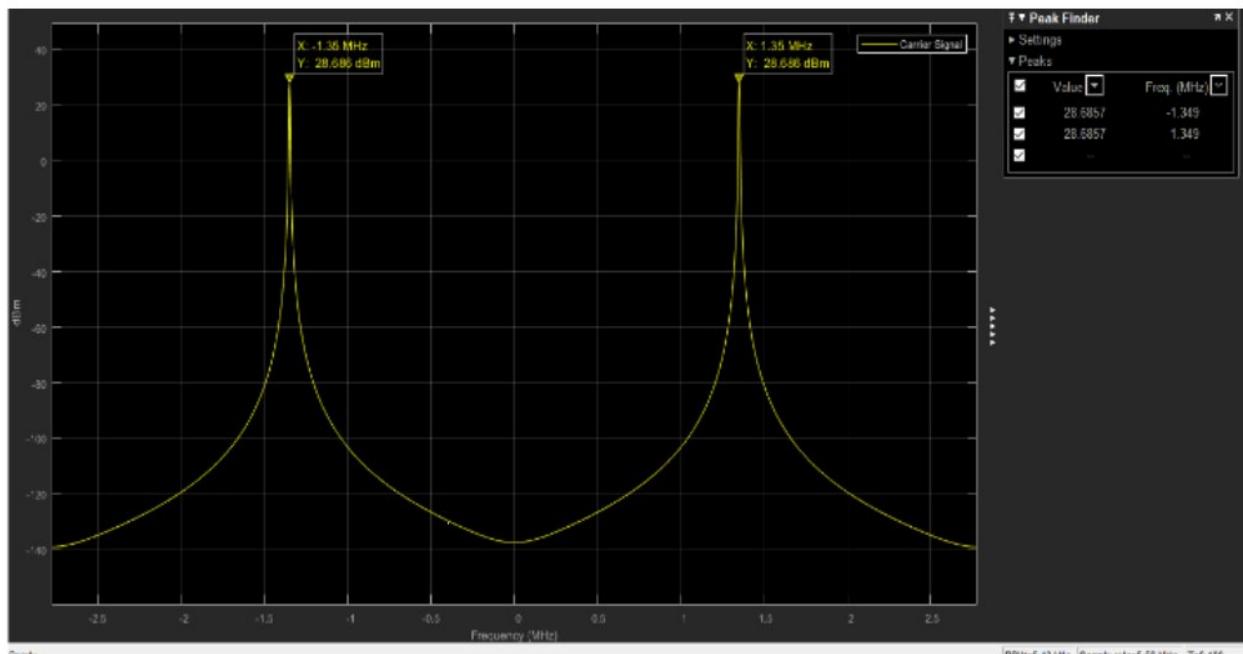
Spectrum of massage signal and output signal



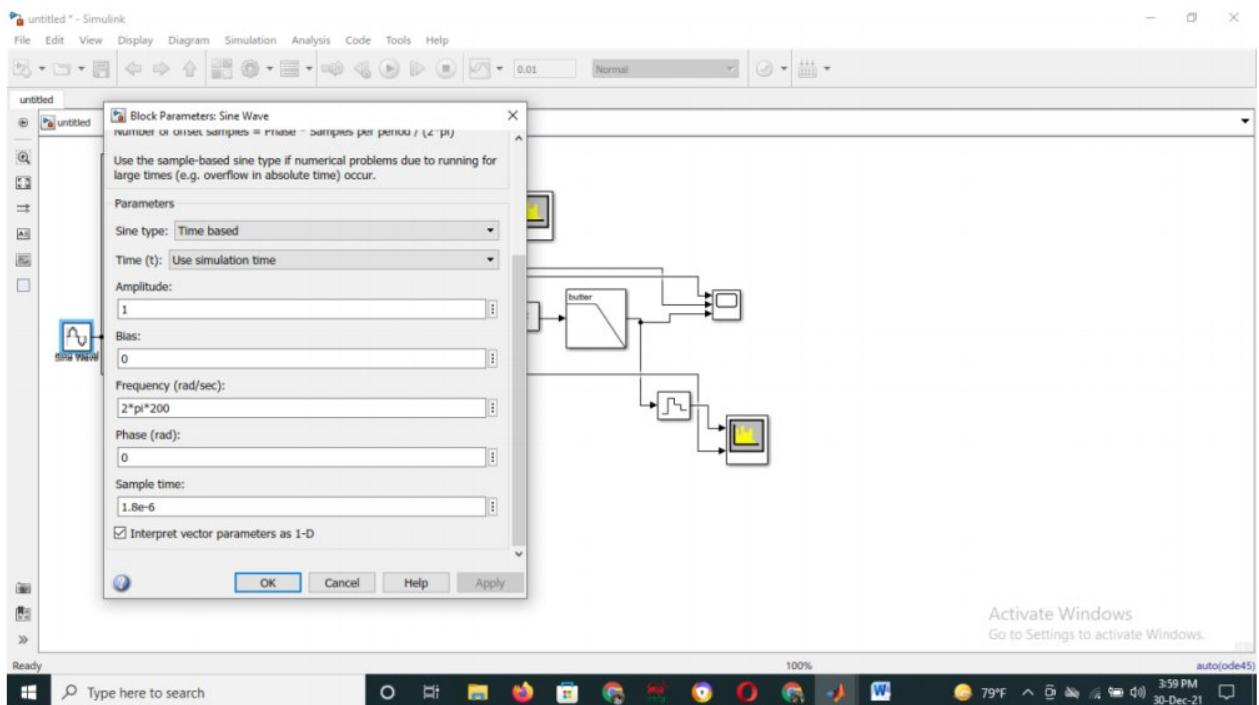
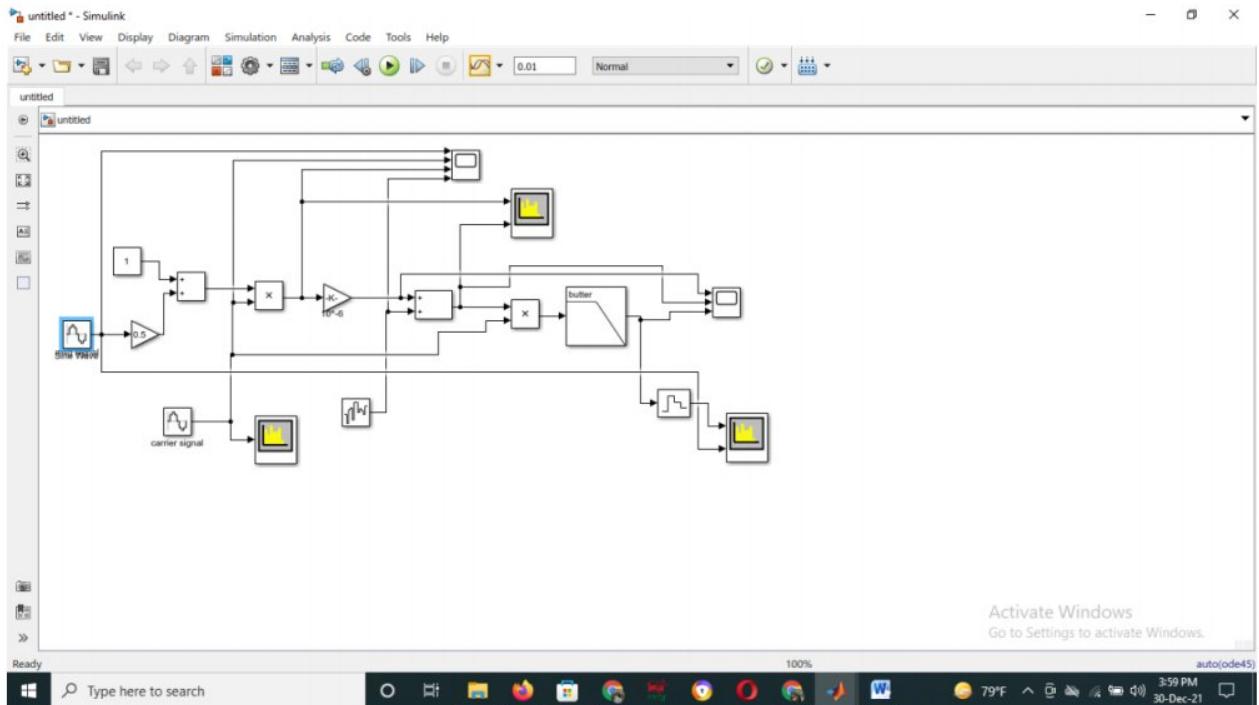
## Spectrum of modulated signal and received signal

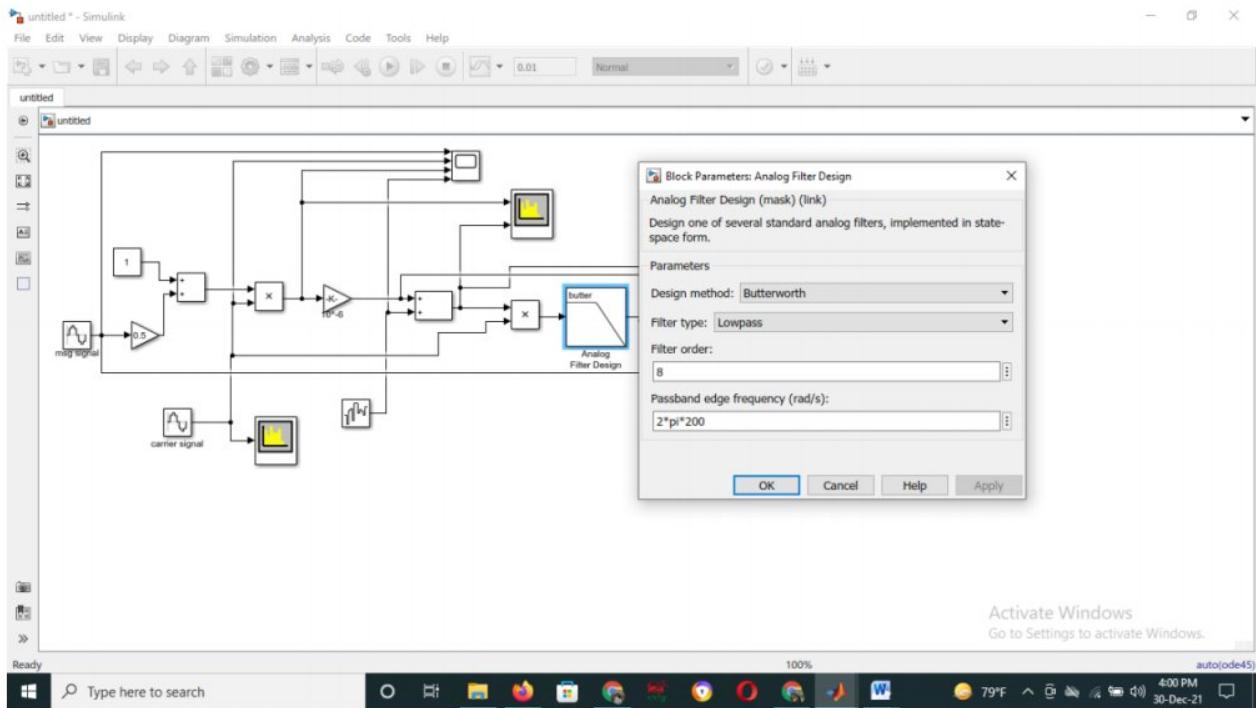


## Carrier power from the graph



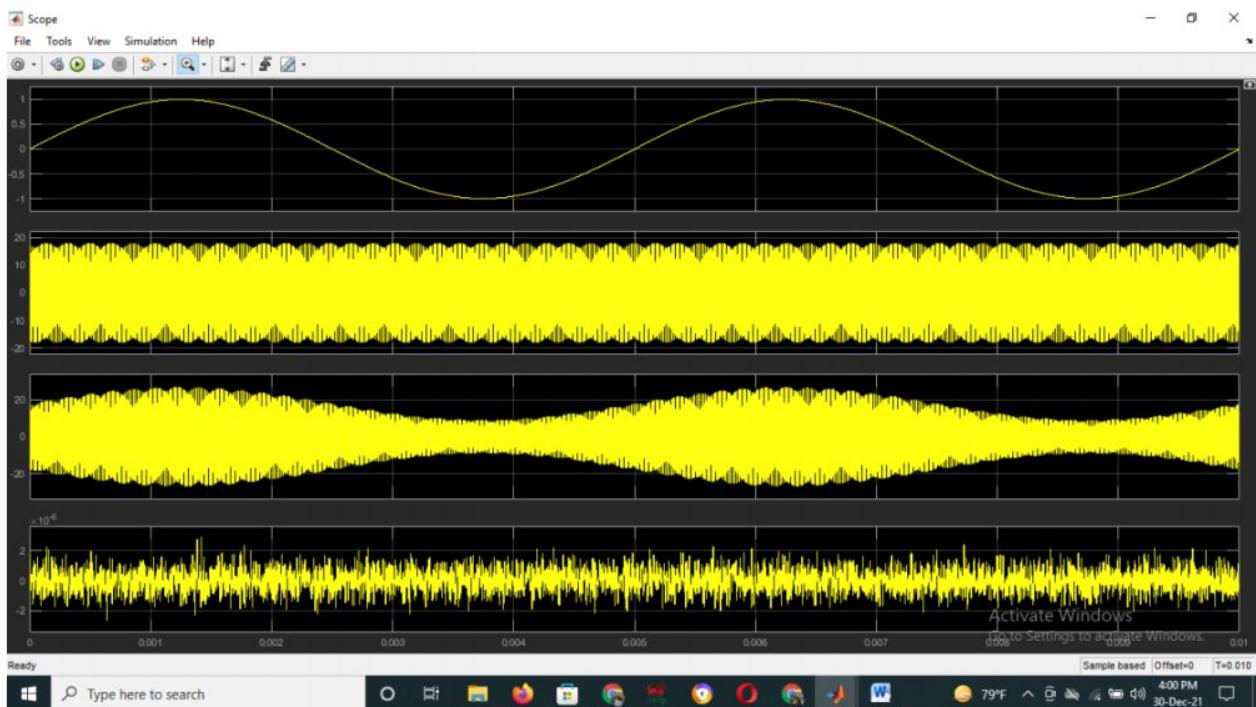
Changed, message frequency=200Hz & filter cut off frequency=200Hz



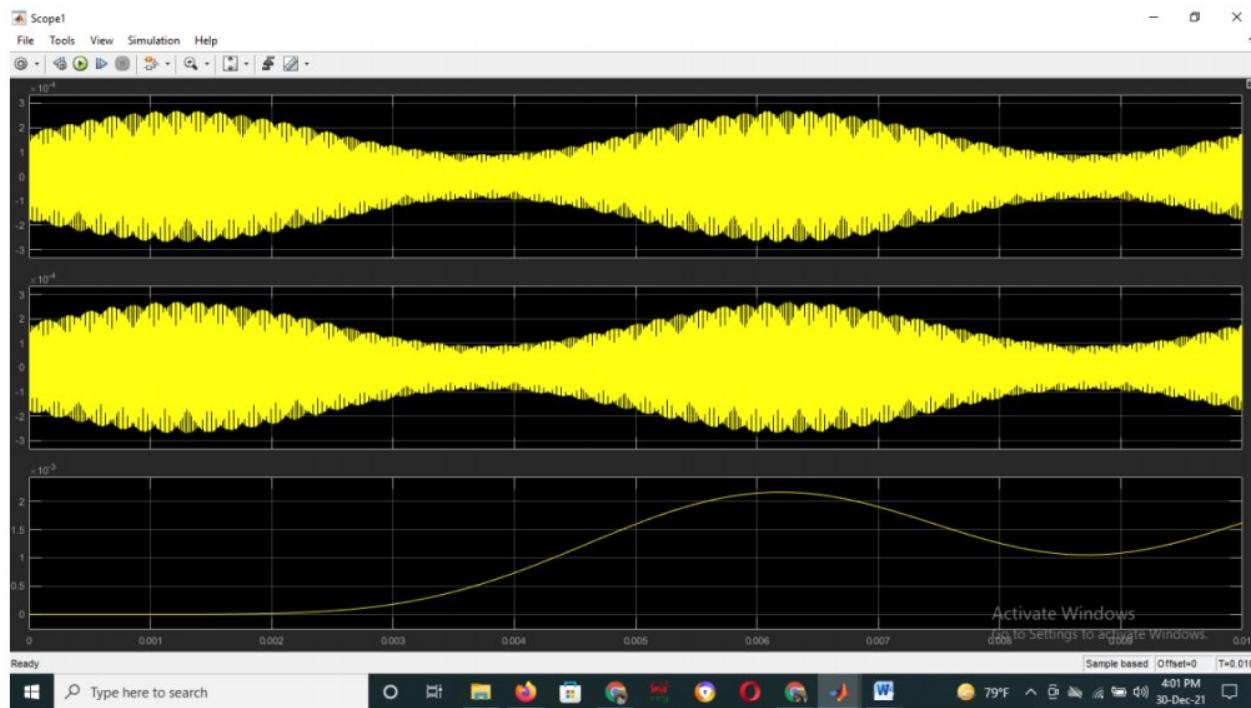


## Results:

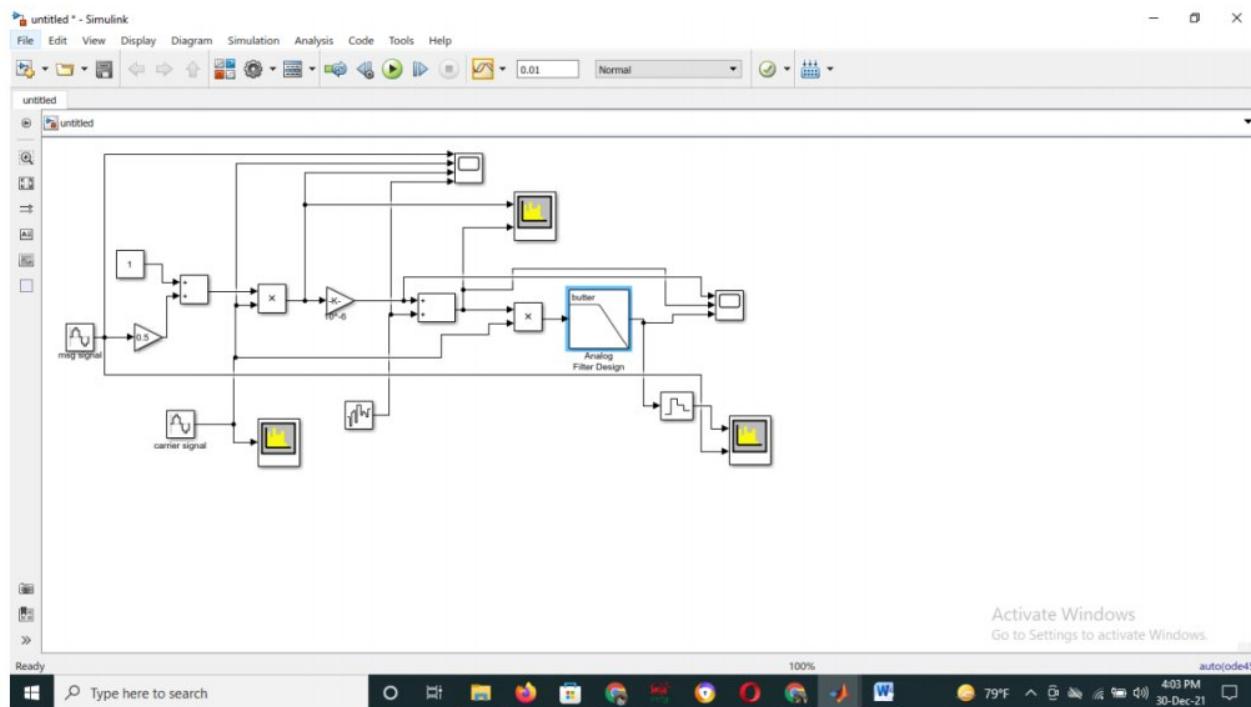
waveform of message, carrier, Modulated signal and noise

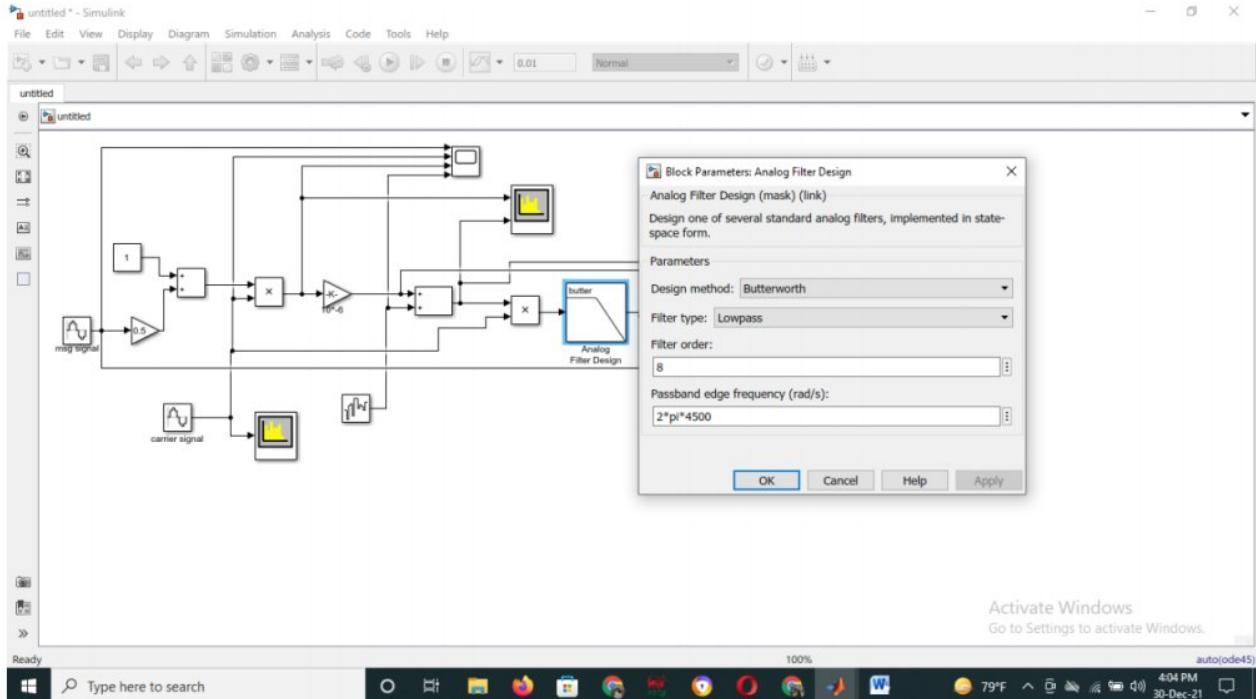
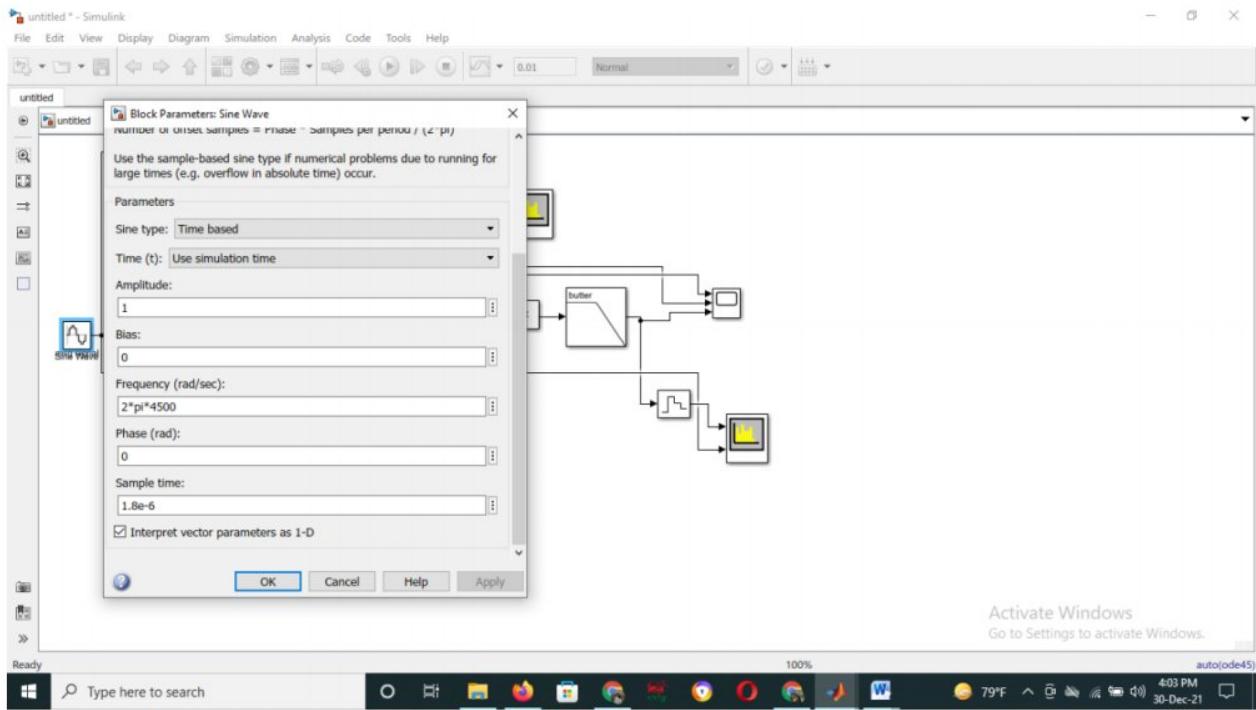


waveform of channel loss, added noise and demodulated filtered signal



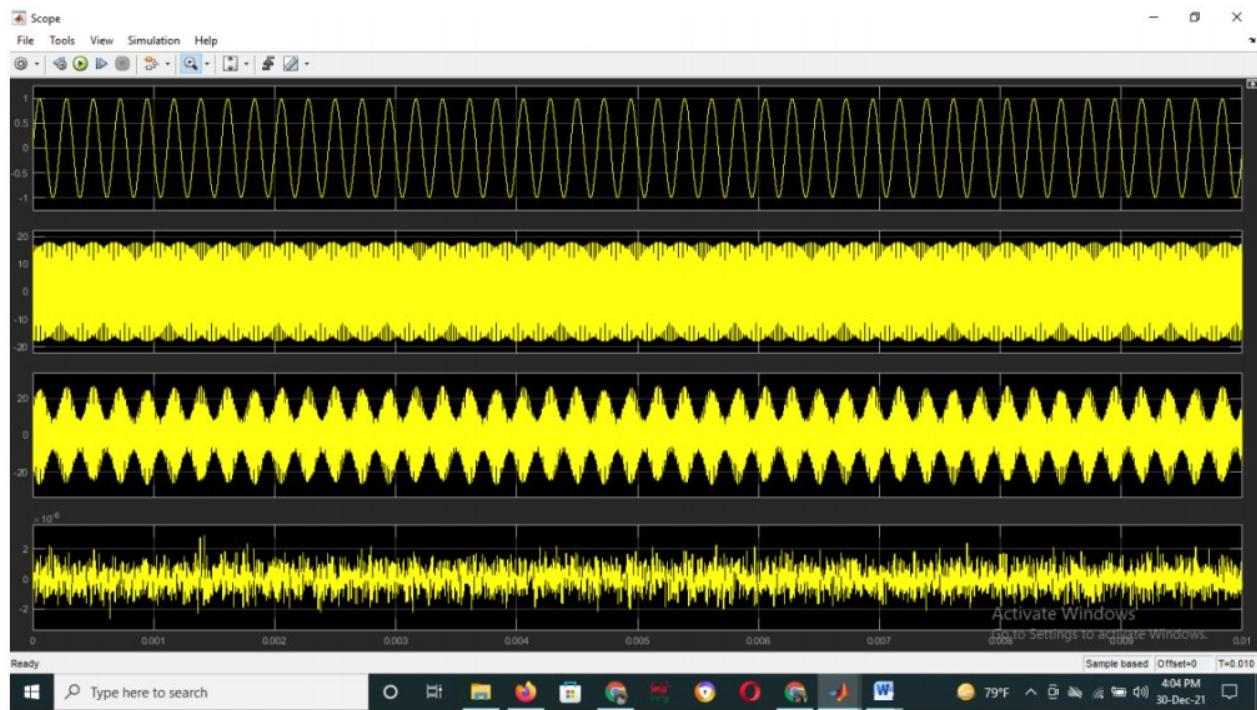
Changed, message frequency=4500Hz & filter cut off frequency=4500Hz



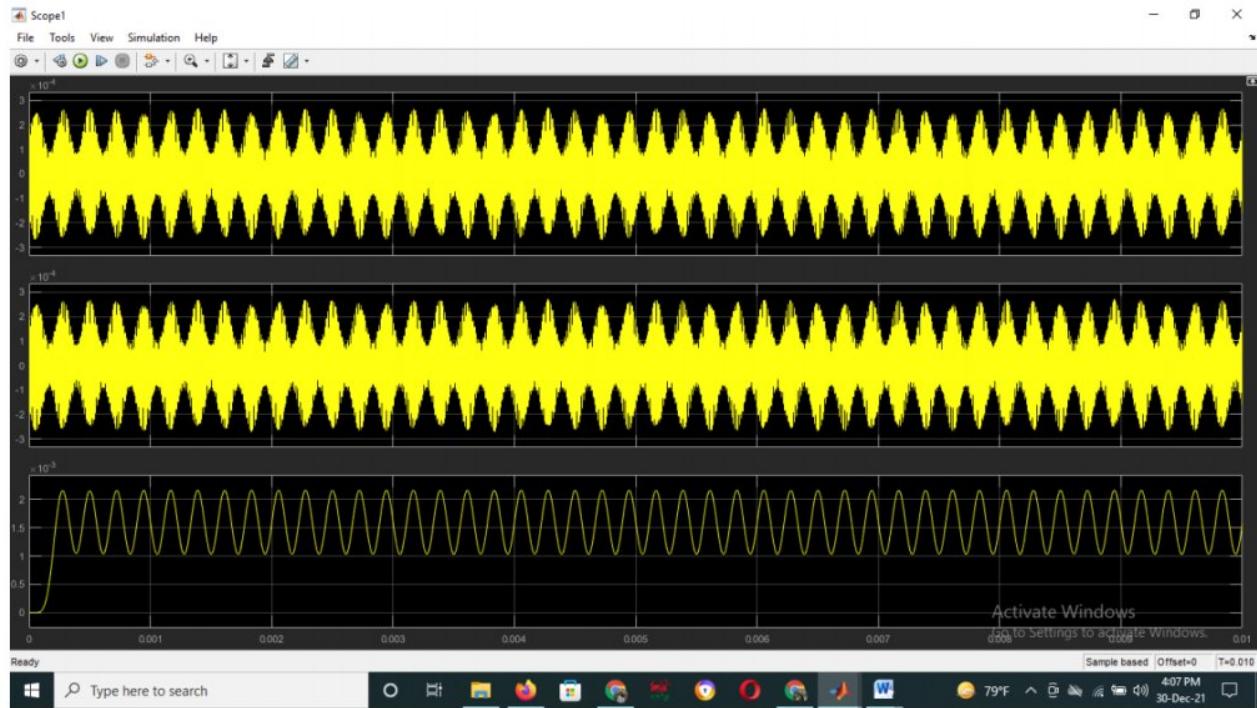


## Results:

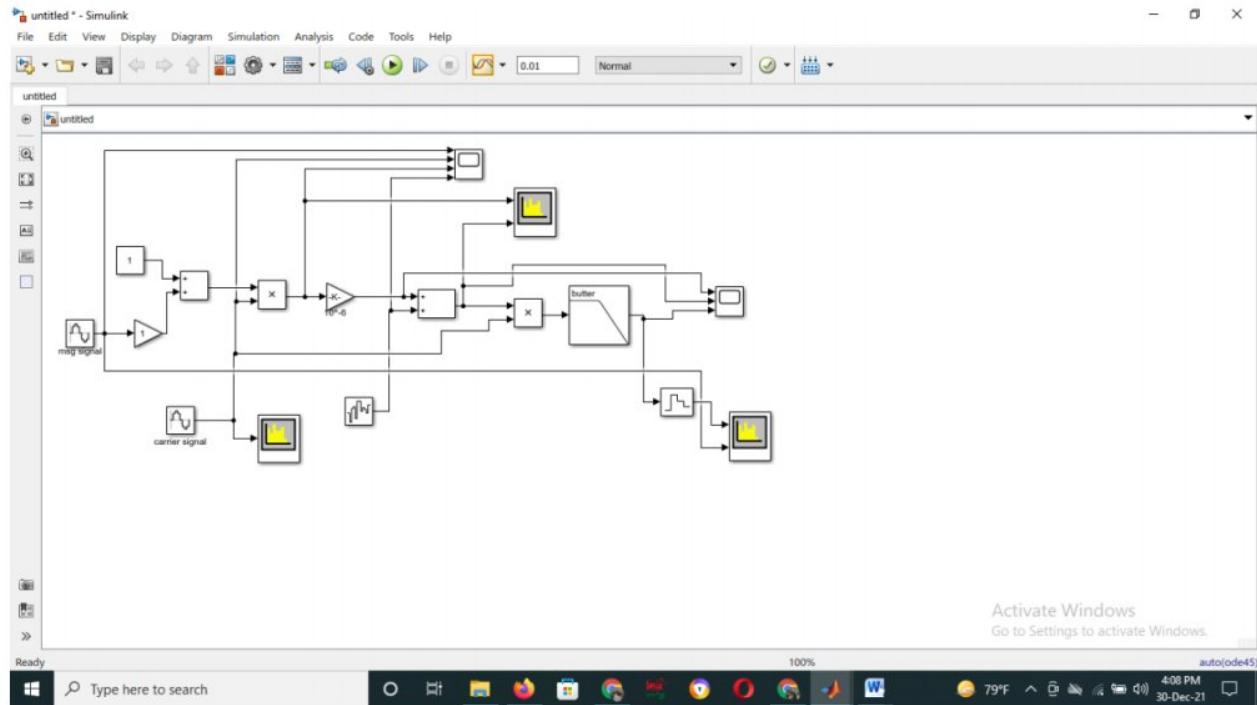
waveform of message, carrier, Modulated signal and noise



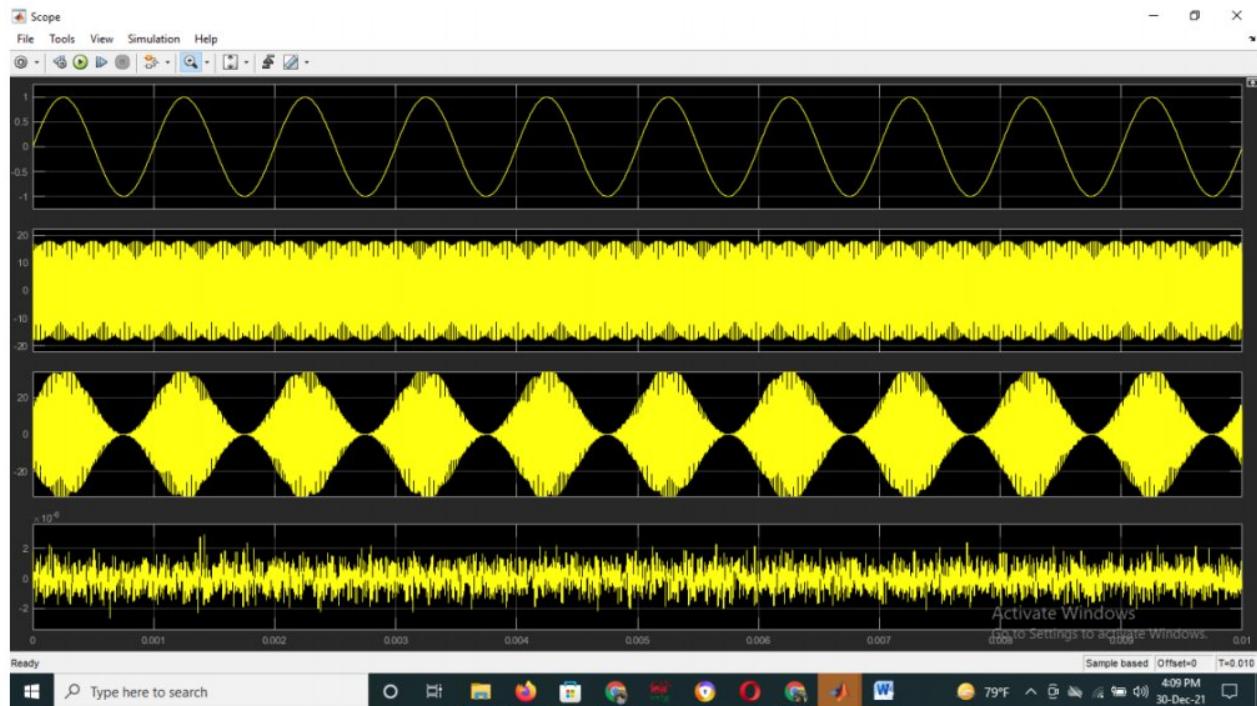
waveform of channel loss, added noise and demodulated filtered signal



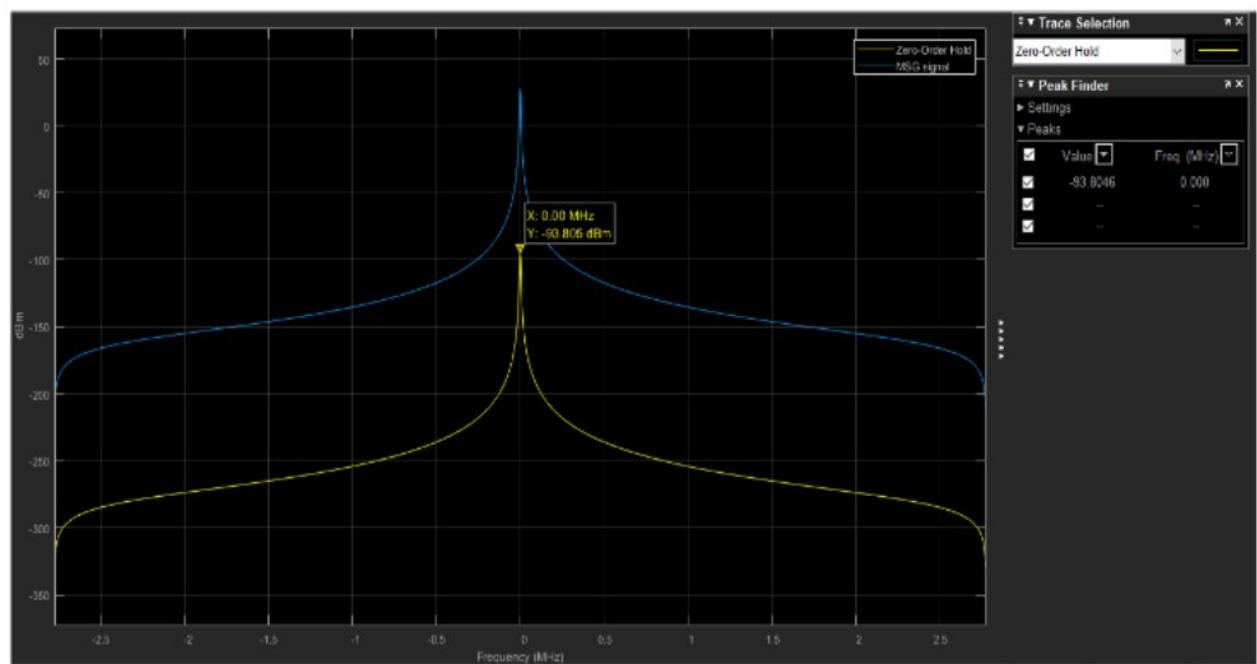
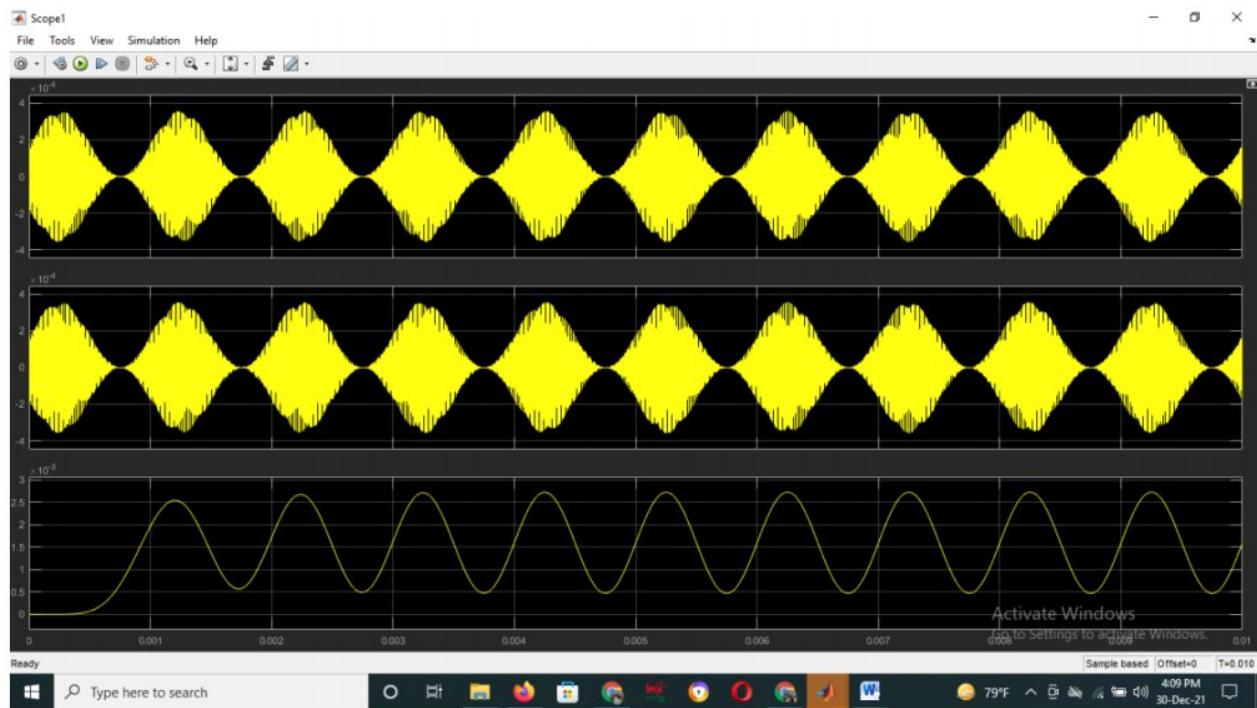
For modulation index = 1



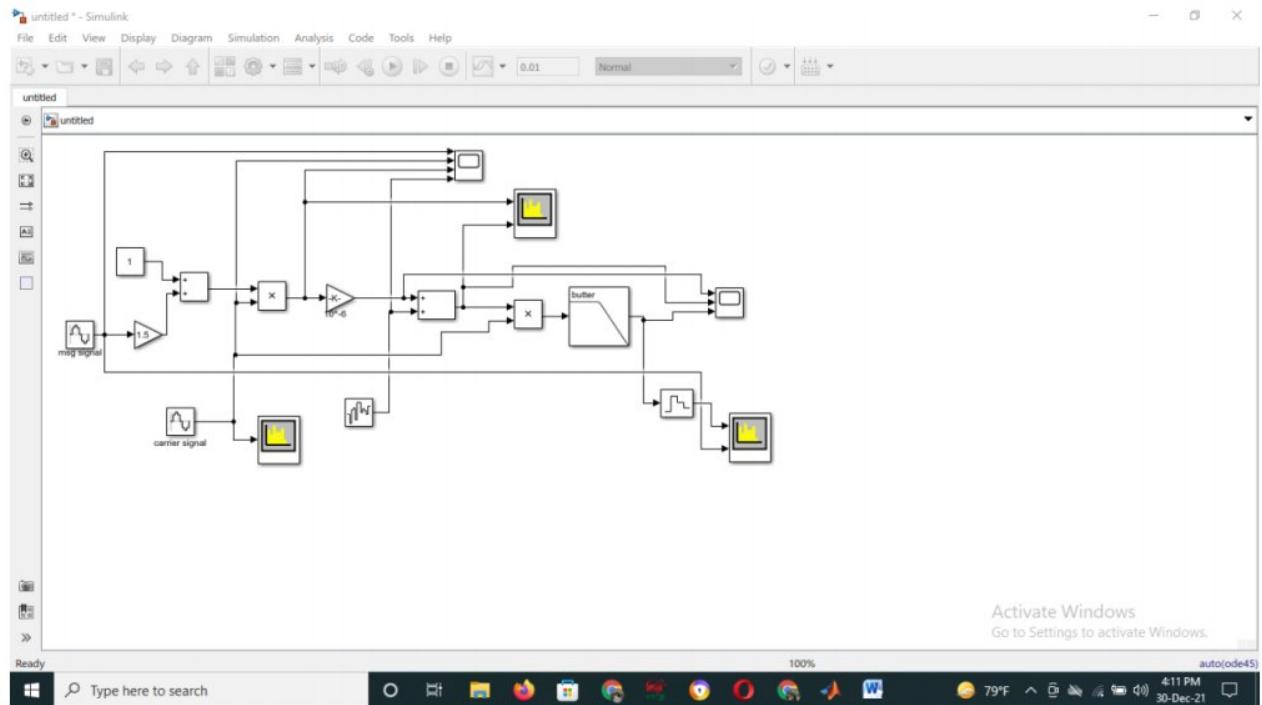
waveform of message, carrier, Modulated signal and noise



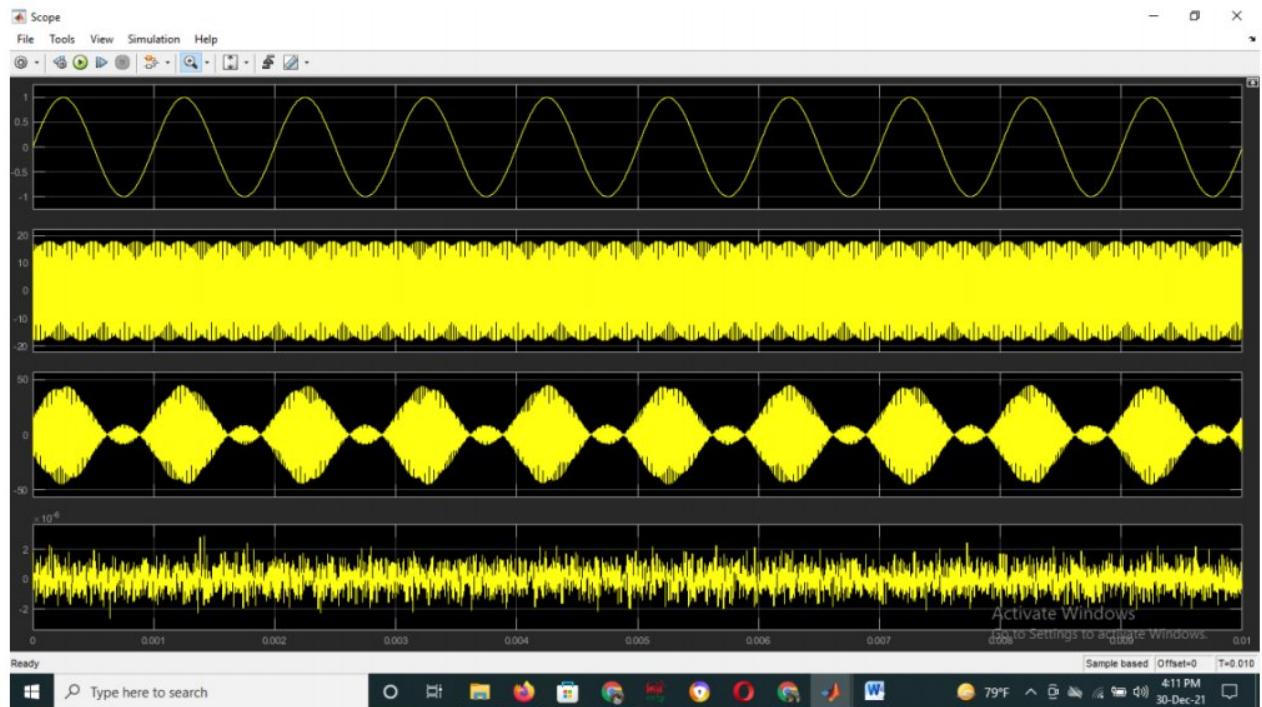
waveform of channel loss, added noise and demodulated filtered signal



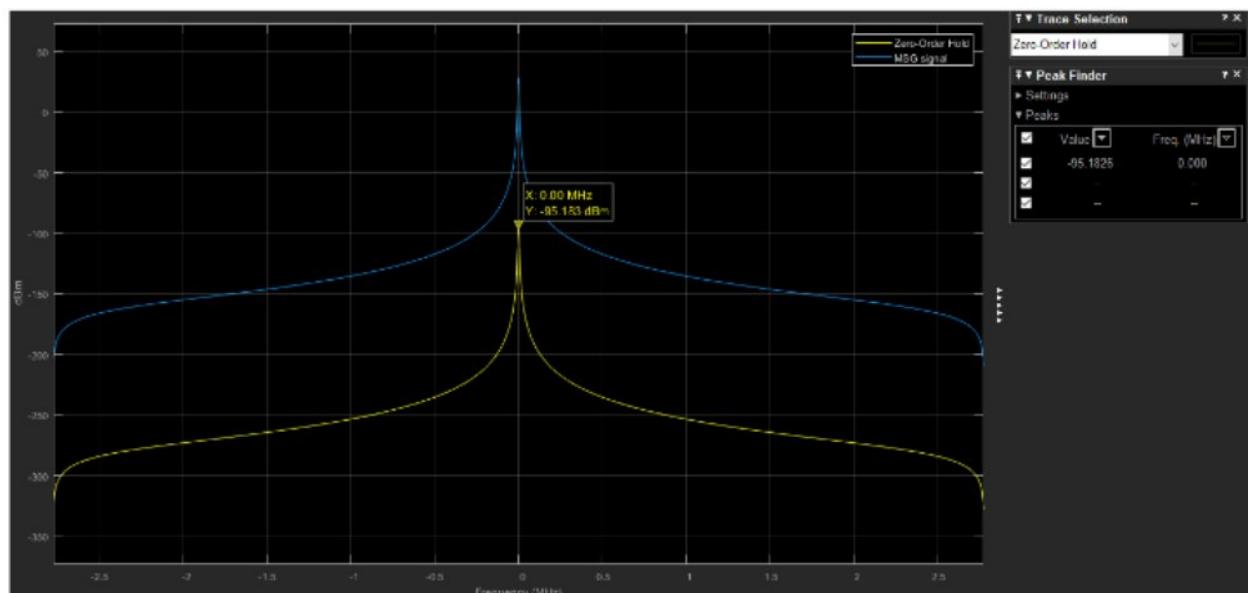
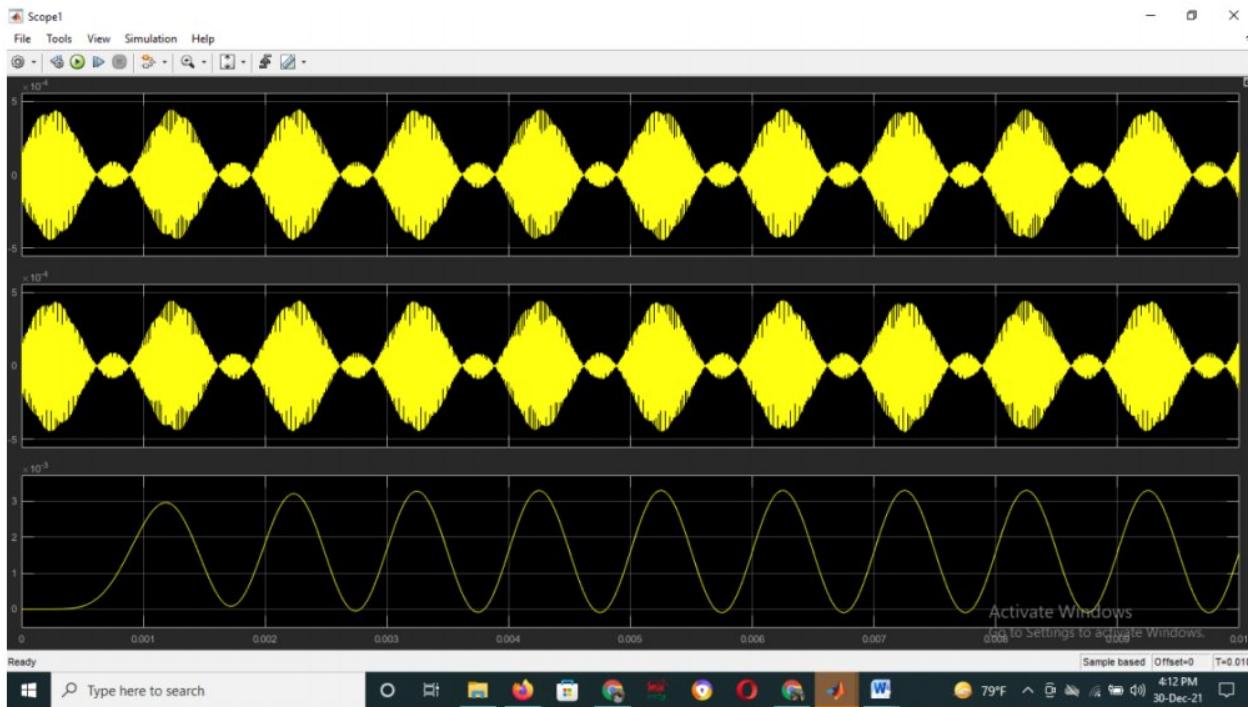
For modulation index = 1.5



waveform of message, carrier, Modulated signal and noise



waveform of channel loss, added noise and demodulated filtered signal



### Discussion & Conclusion:

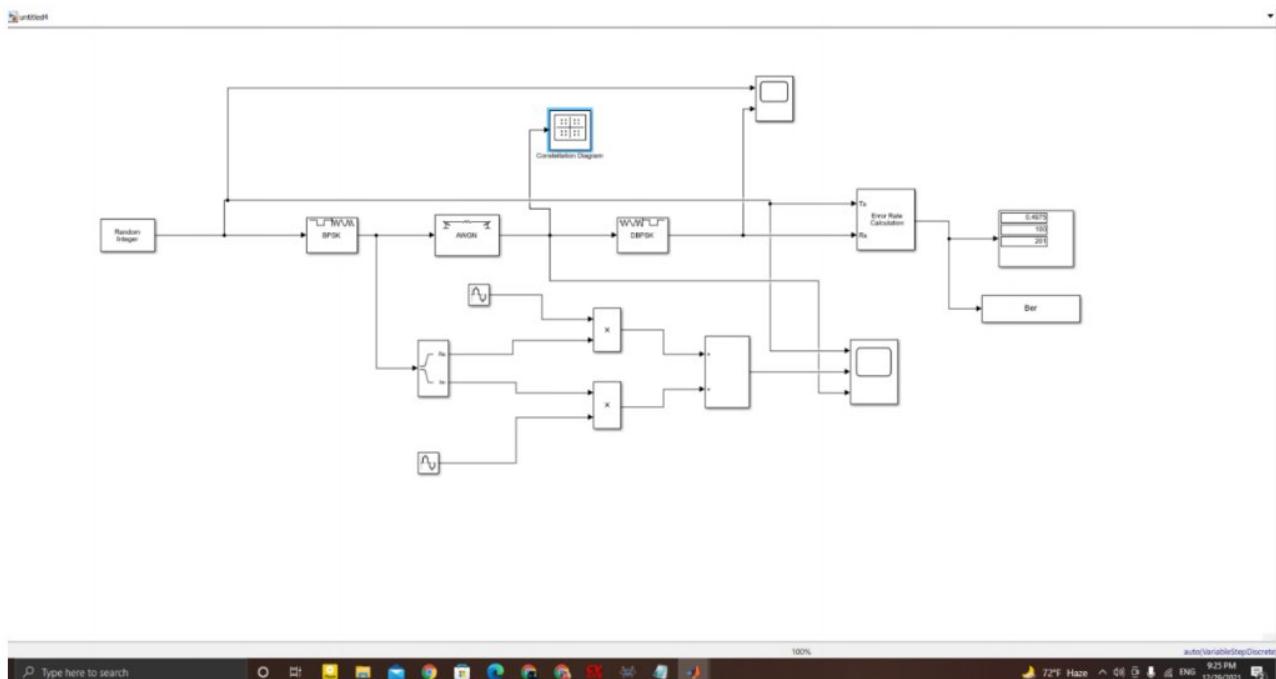
Modulation Index =1 gives perfect modulation (100%). Whenever MI<1 we get under modulation. The demodulated component became more distorted when the gain was increased, suggesting an increase in the modulation index to 1 and then 1.5. We could see there was already an over modulation as soon as we examined at the time domain graph of the modulation index 1.5. Modulation exceeding 1 ( $MI>1$ ) results in over modulation and distortion, which is why it is strongly discouraged. When, MI value is more than 1, the carrier experiences a 180 degree phase reversal, which causes additional sidebands, hence the wave shape gets distorted. Over modulated wave causes interference which cannot be eliminated. SNR also changed while changing MI. For MI= 0.5 SNR is 40 ; when MI= 1 our SNR is 46.02 and for MI=1.5, SNR=49.49 and fm=4500hz then SNR =33.47

When we altered the message signal to 200 Hz, we can visualize from the figures given above

that our output wave forms changed radically. We had trouble creating the entire process and computing the necessary queries. We had to conduct extensive research in this topic and watch our theory lesson videos to gain a general comprehension of the fundamentals.

### **Task 2-3**

#### **Circuit Diagram for Binary Phase Shift keying**

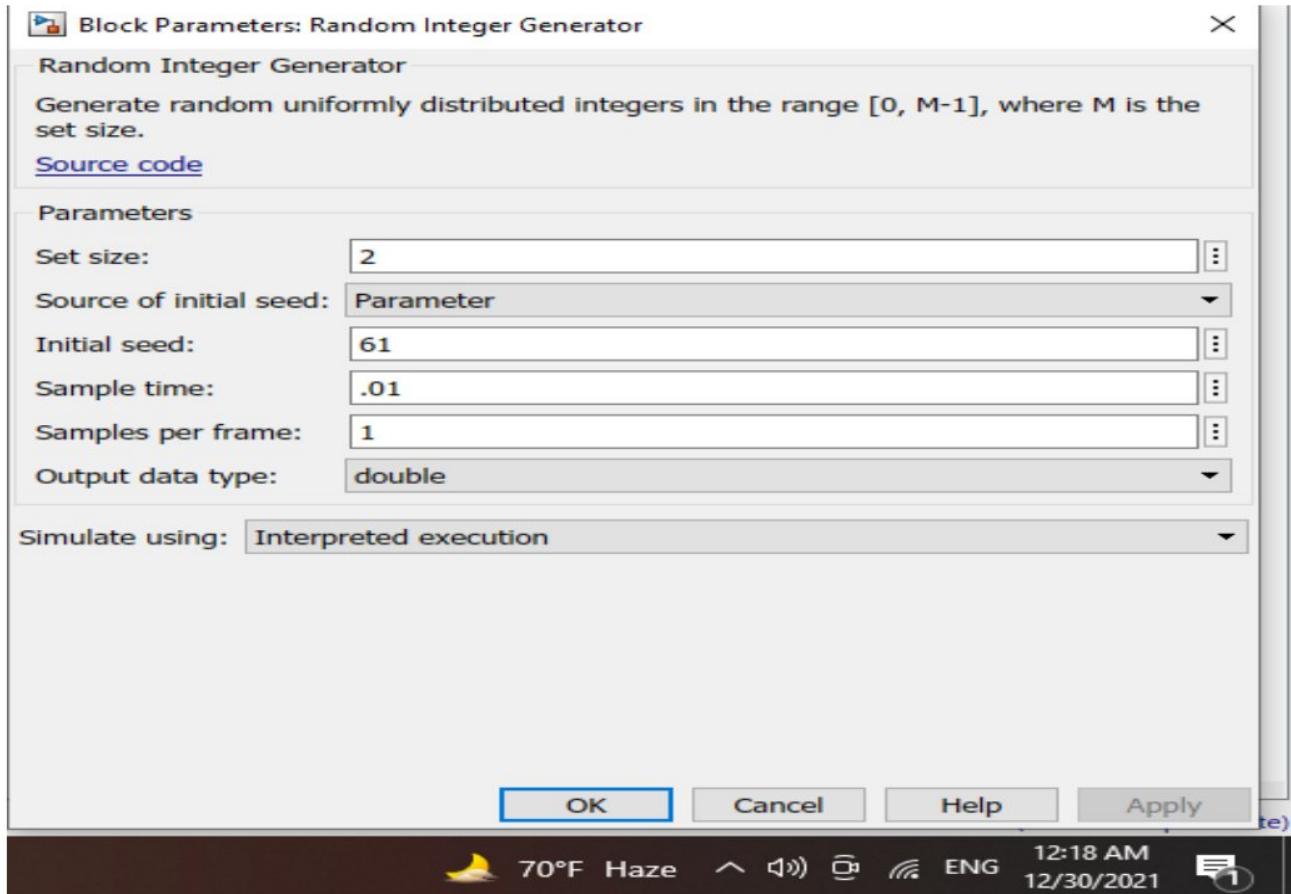


#### **Simulink Blocks**

- 1.Random Integer Generator
- 2.BPSK modulator and demodulator
- 3.AWGN channel
- 4.Complex to Real Image block
- 5.Constellation block
- 6.Sine Wave
- 7.Scope
- 8.Error rate calculation
- 9.Display and To Workspace block
- 10.Product and adder block.

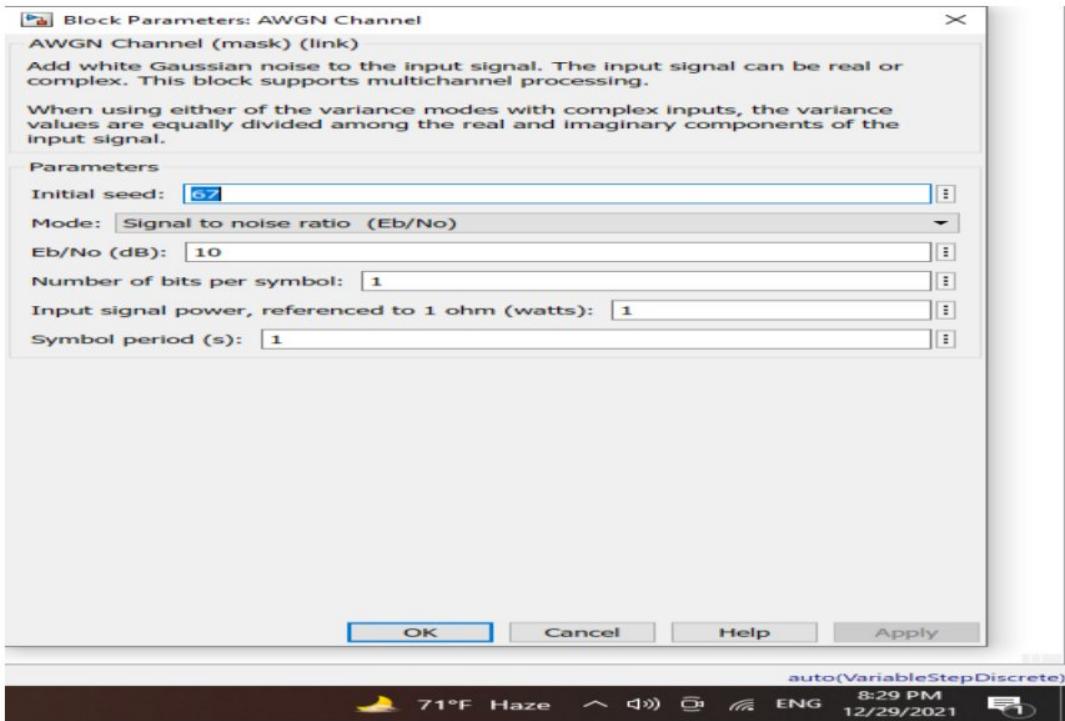
## Parameters for the BPSK

1.Random Generator: Set size 2 defines that it generate one bit at a time which is either 0 or 1.

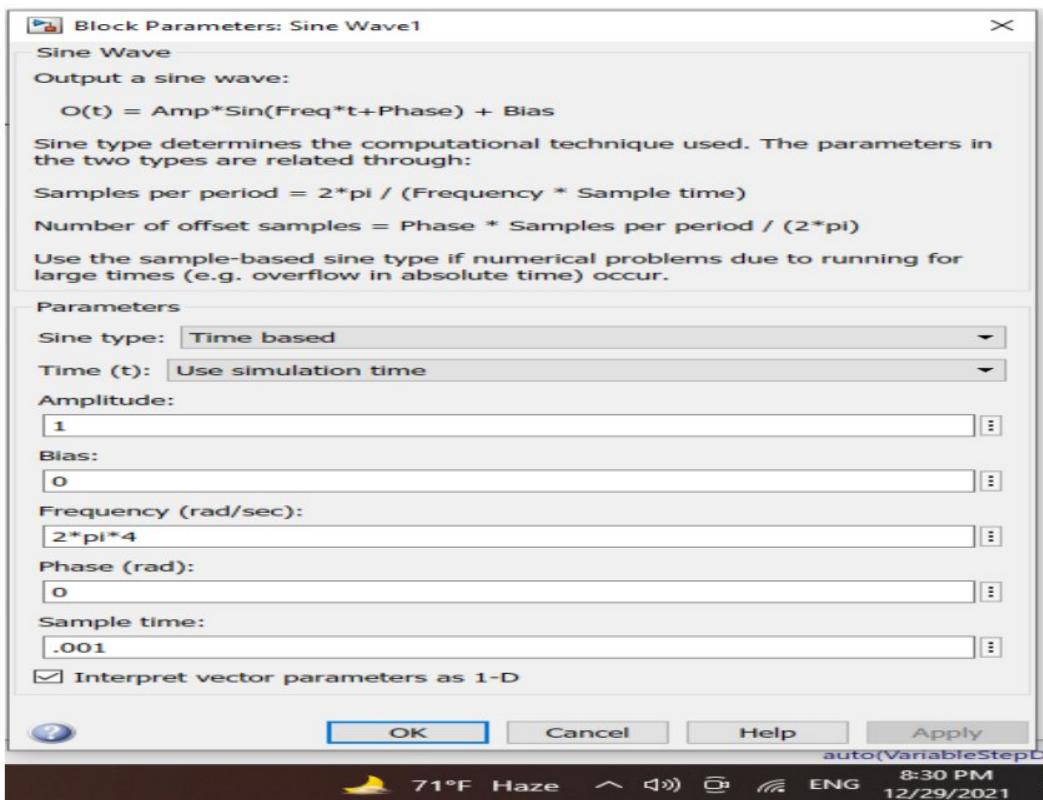


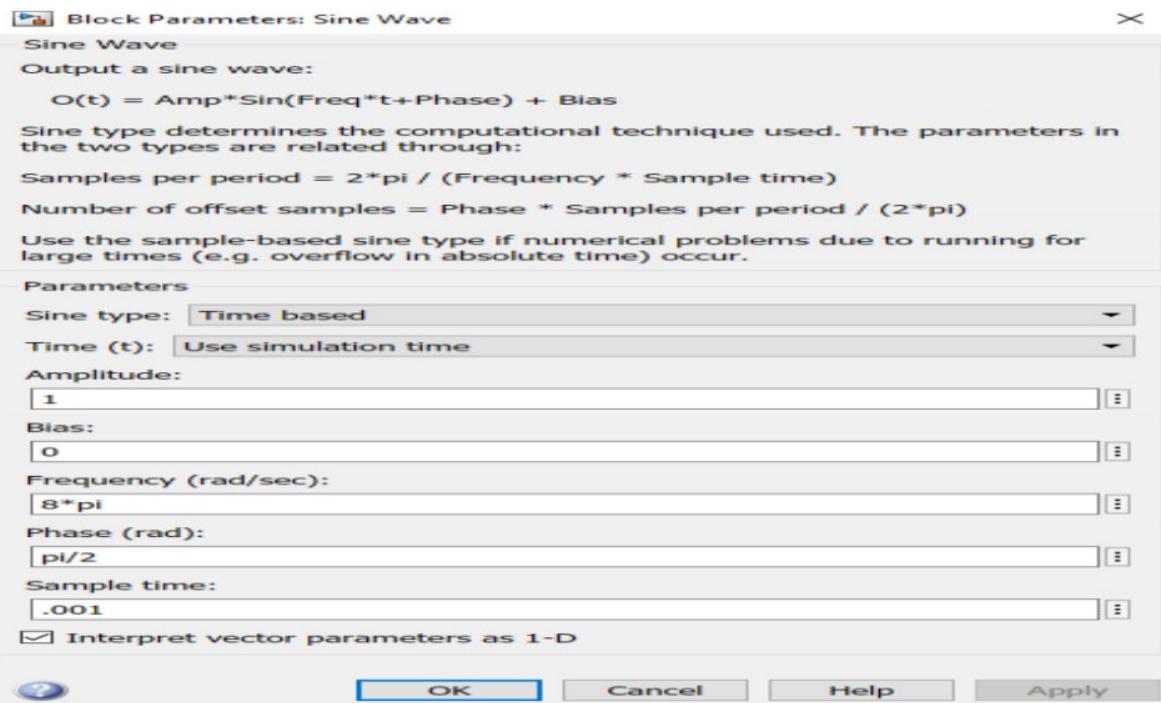
## 2.Additive White Gaussian Noise Channel

We vary Eb/No from 0 to: 20 with an increment of 5 to observe the bit error rate for both BPSK and QPSK communication channel.

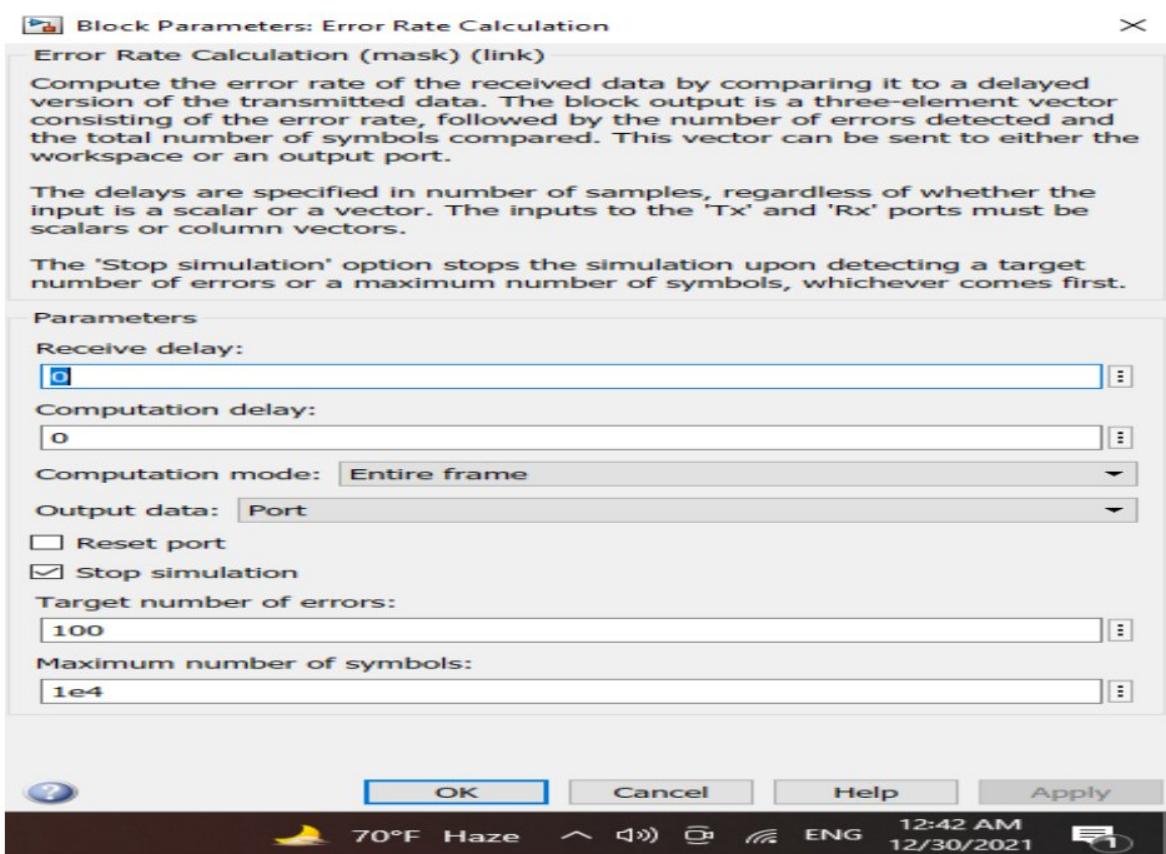


**3. Sine wave:** These signals are required to expand the complex modulated signal into real and imaginary parts.





4. Error rate calculation is accommodated of maximum le4 symbol and count up to 100 number of errors as target.



EB/No results

Eb/No=5db

0.4975
100
201

Eb/No = 15db

→

0.4902
100
204

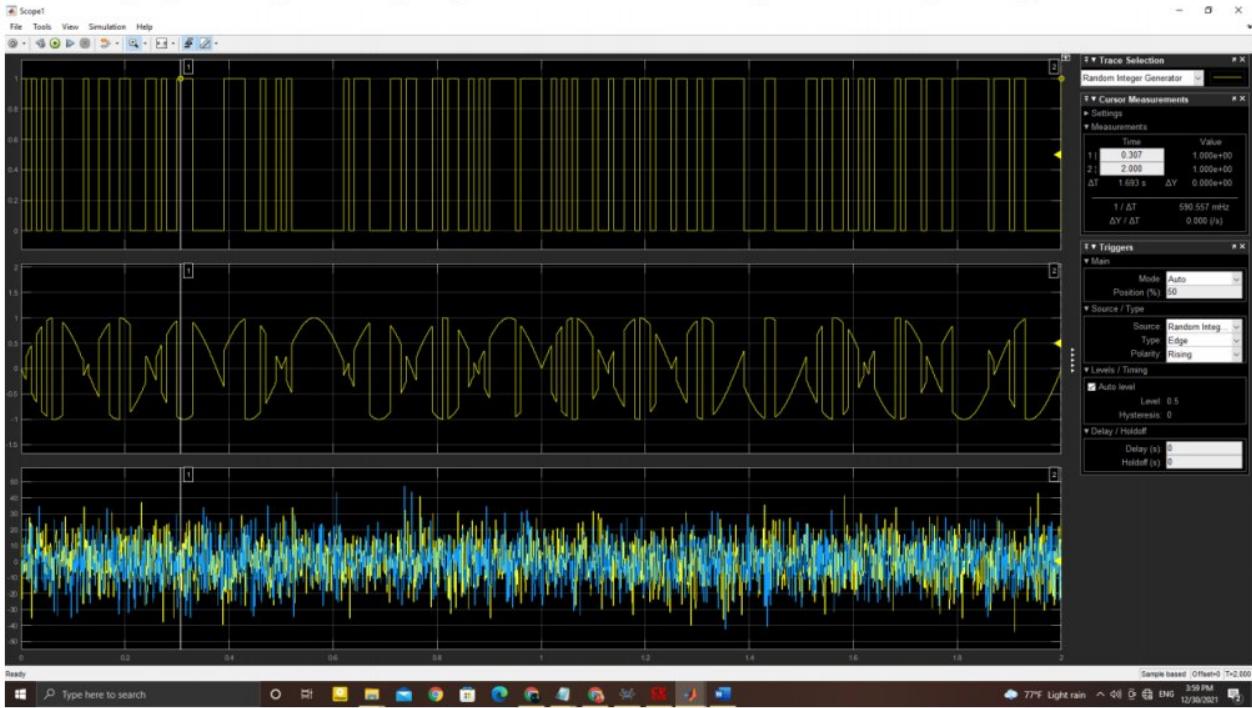
Eb/No =20

→

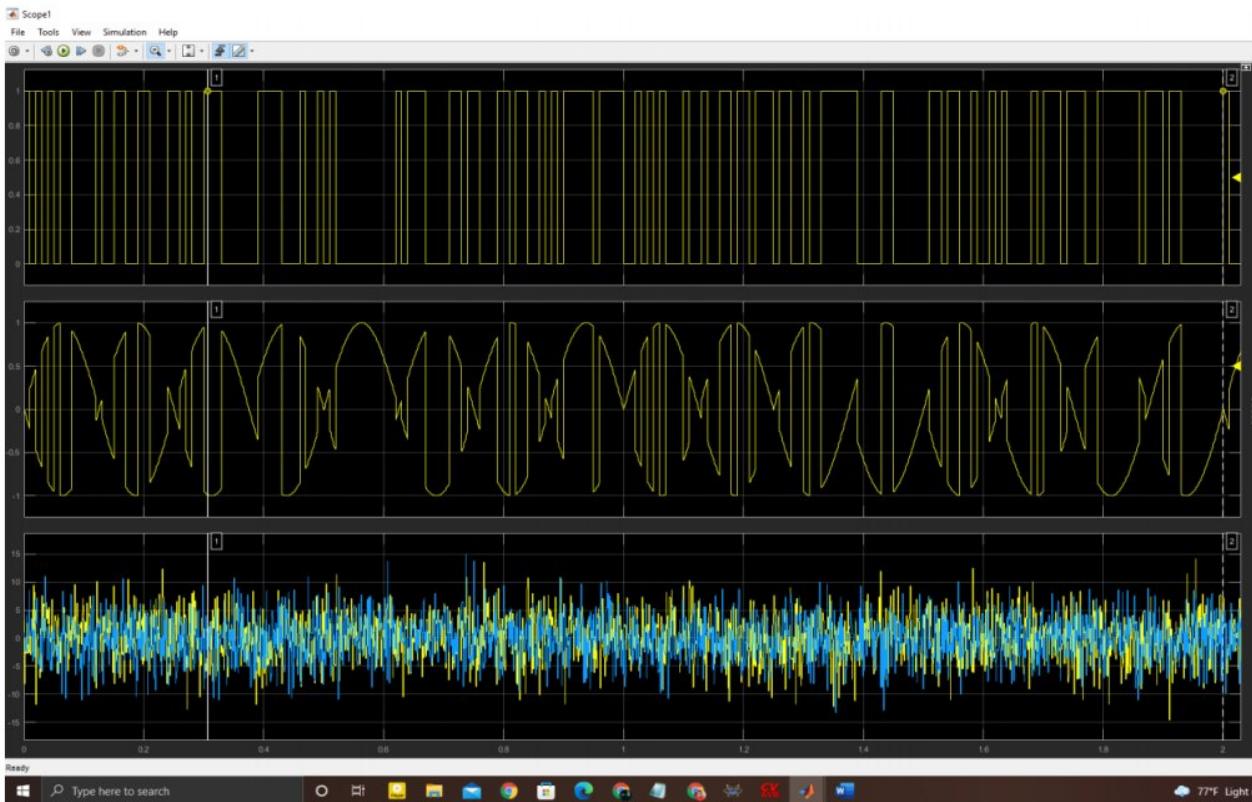
0.5102
100
196

From the result we can observe that the error rate decrease with an escalation of signal energy and Noise power ratio.

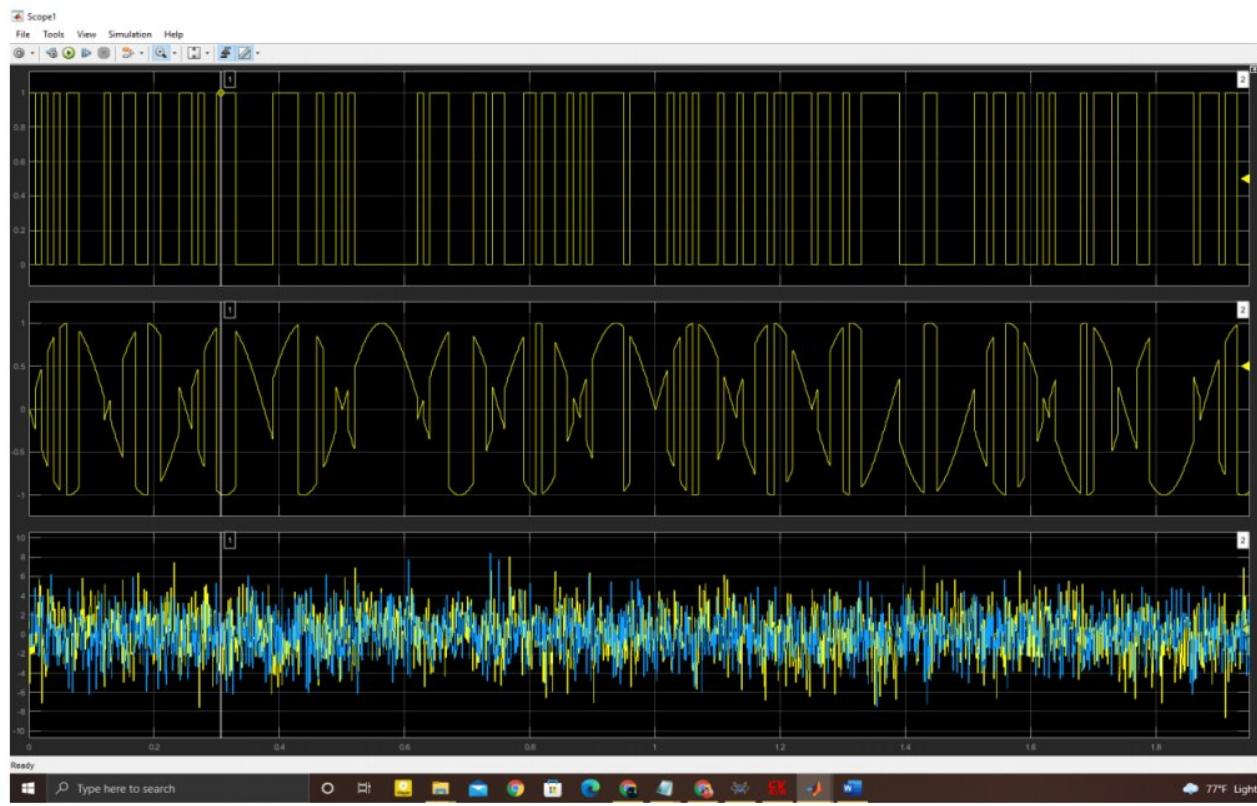
Here the graph for Input Binary sequence vs Modulated Signal vs Corrupted Signal at Eb/No=5db



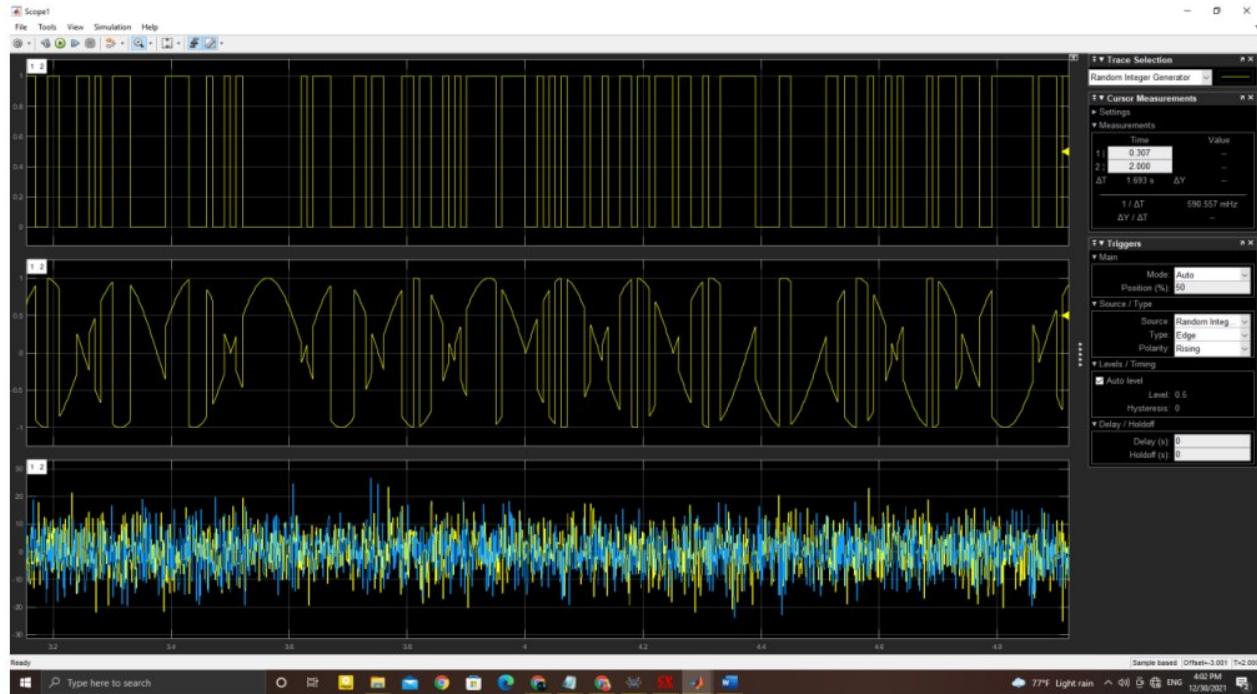
Input Binary sequence vs Modulated Signal vs Corrupted Signal at Eb/No=15db



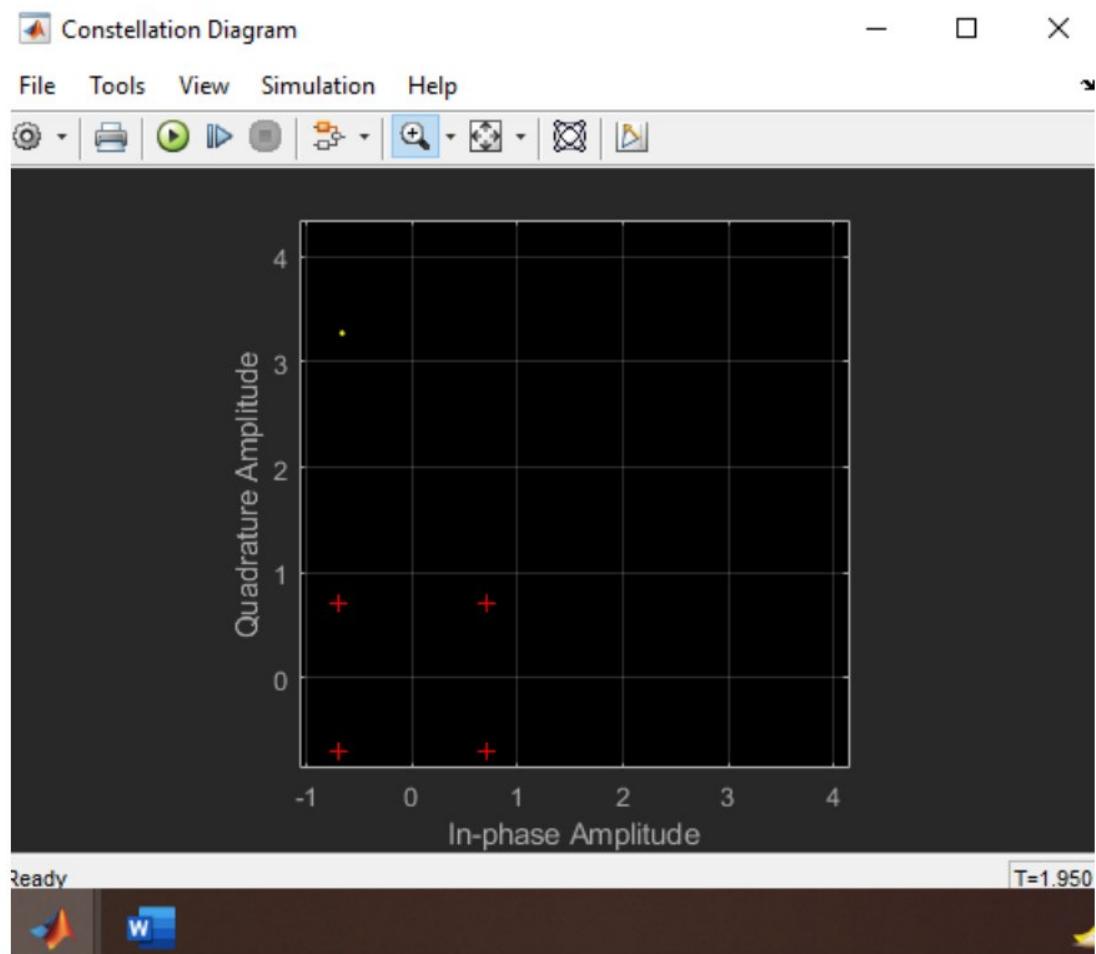
### Input Binary sequence vs Modulated Signal vs Corrupted Signal at Eb/No=20db



### Input Binary sequence vs Modulated Signal vs Corrupted Signal at Eb/No=10db



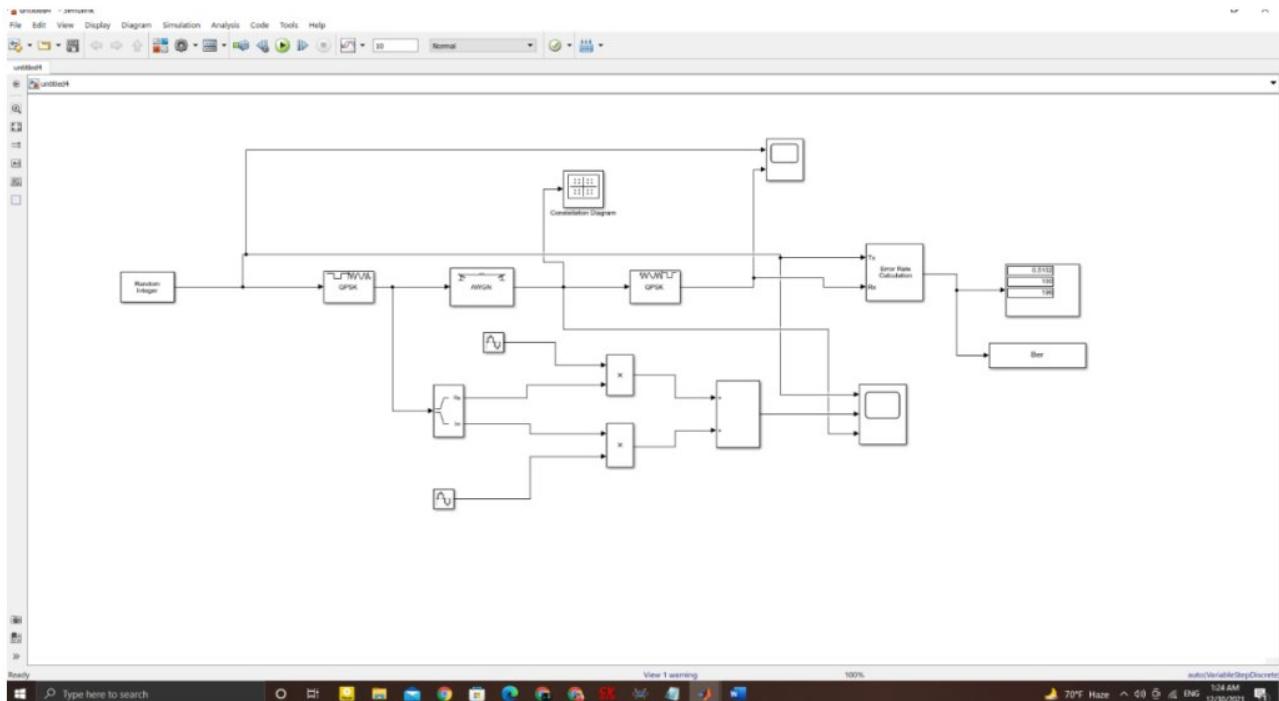
## Constellation Diagram of BPSK Modulator



## **Task2:**

Here we asked to show Signals generate link and modulator diagrams, simulation results including waveforms, evaluation of results, contrasting between BPSK and QPSK.

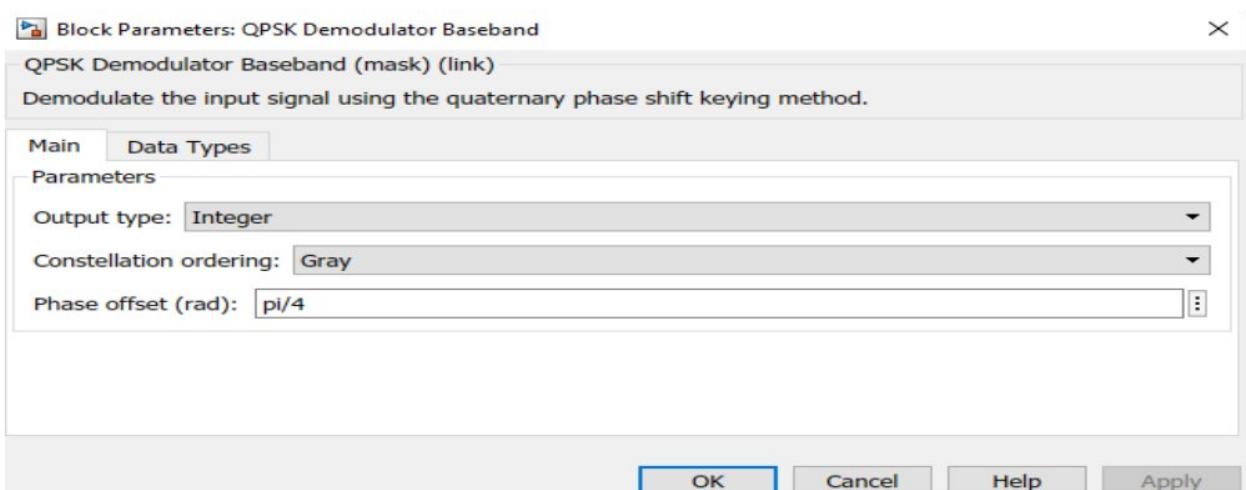
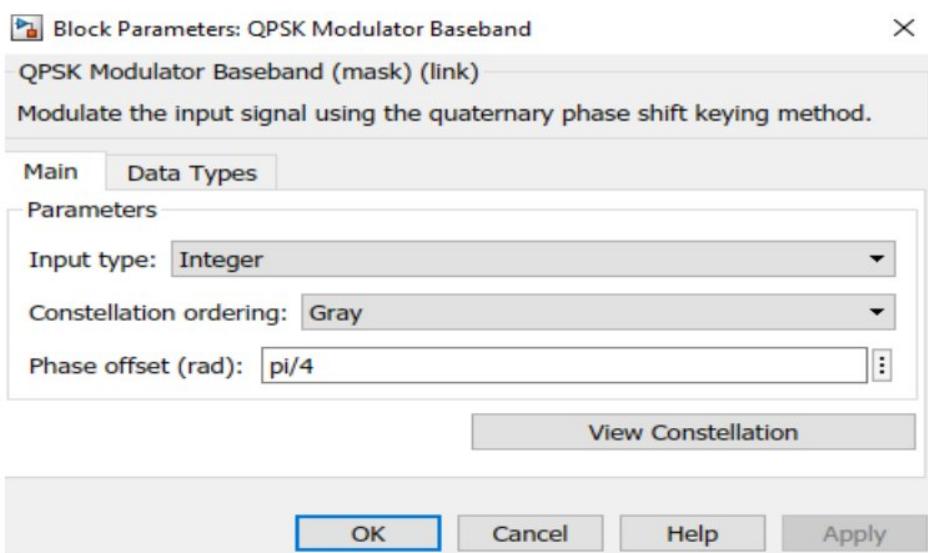
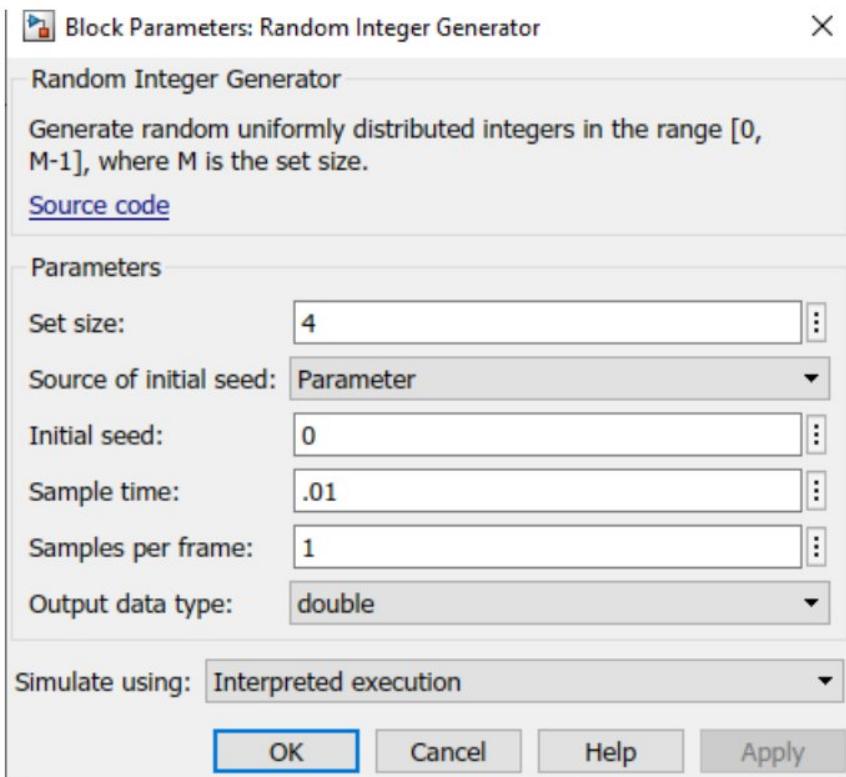
Circuit Diagram for Quadrature phase shift keying



### **Simulink Blocks**

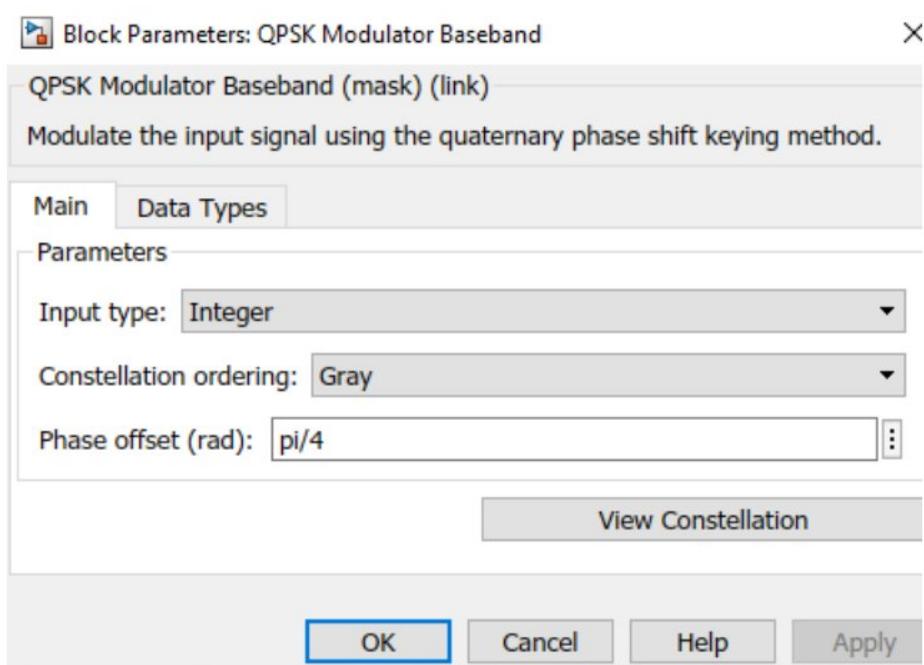
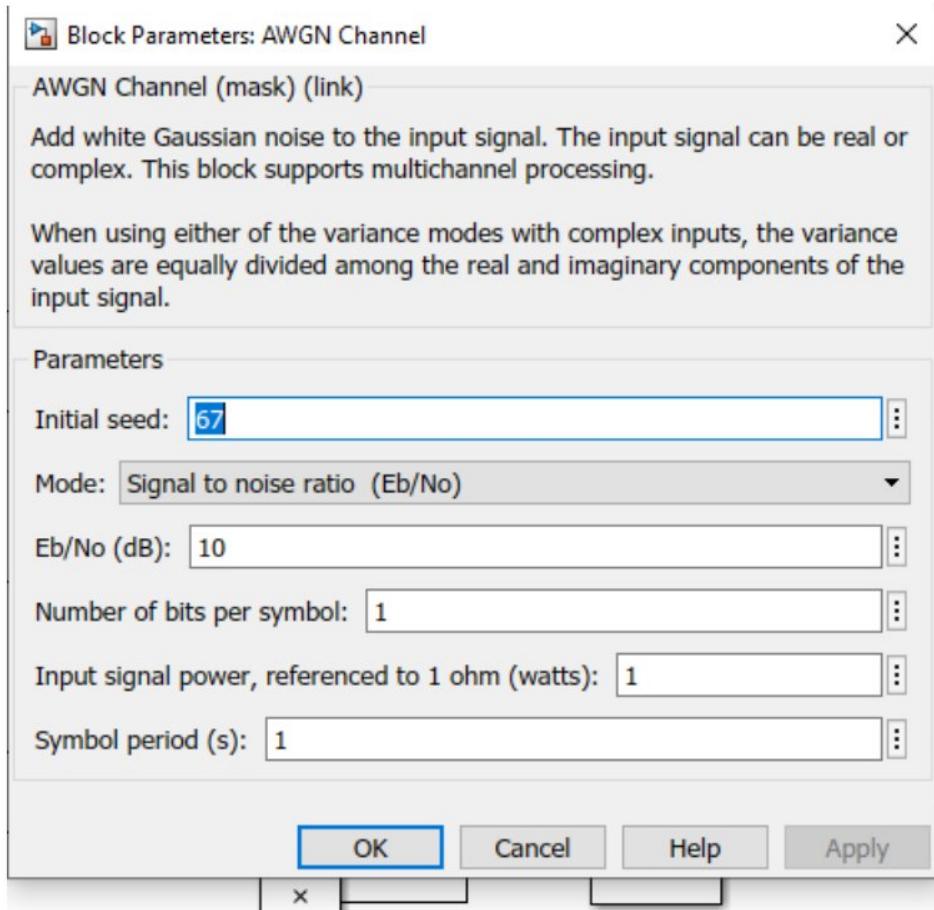
- 1.Random Integer Generator
- 2.QBPSK modulator and demodulator
- 3.AWGN channel
- 4.Complex to Real Image block
- 5.Constellation block
- 6.Sine Wave
- 7.Scope
- 8.Error rate calculation
- 9.Display and To Workspace block
- 10.Product and adder block.

## Parameters:

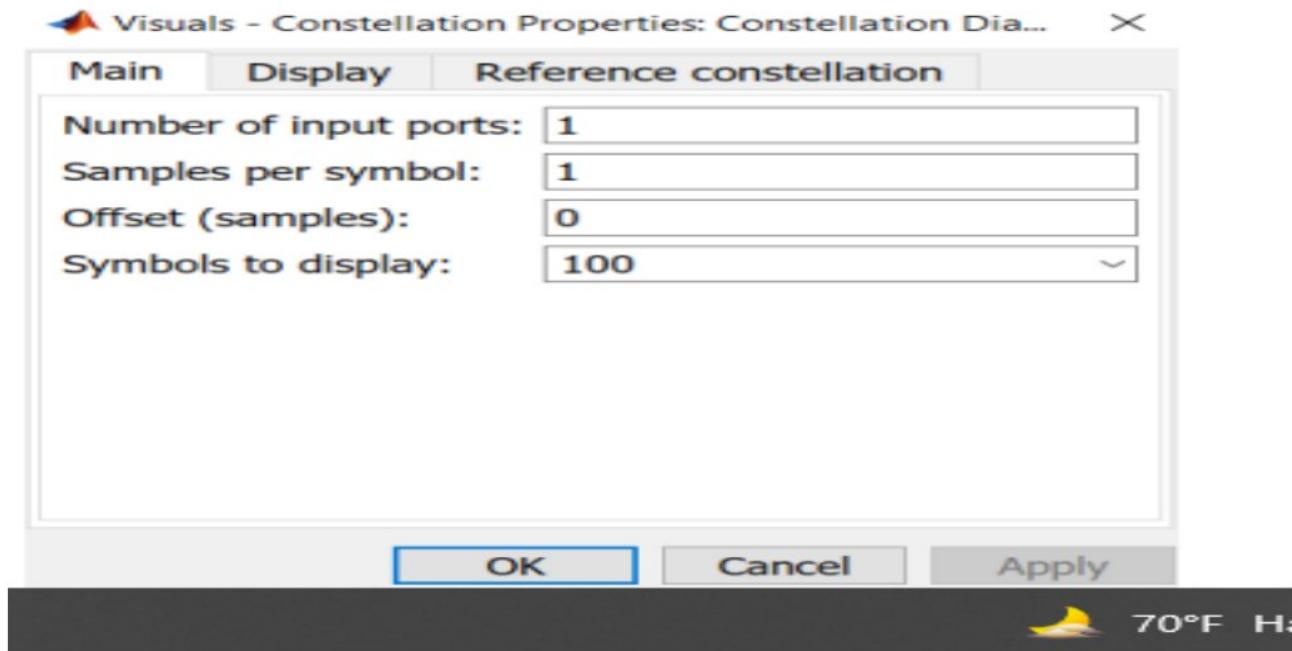


## Additive White Gaussian Noise Channel

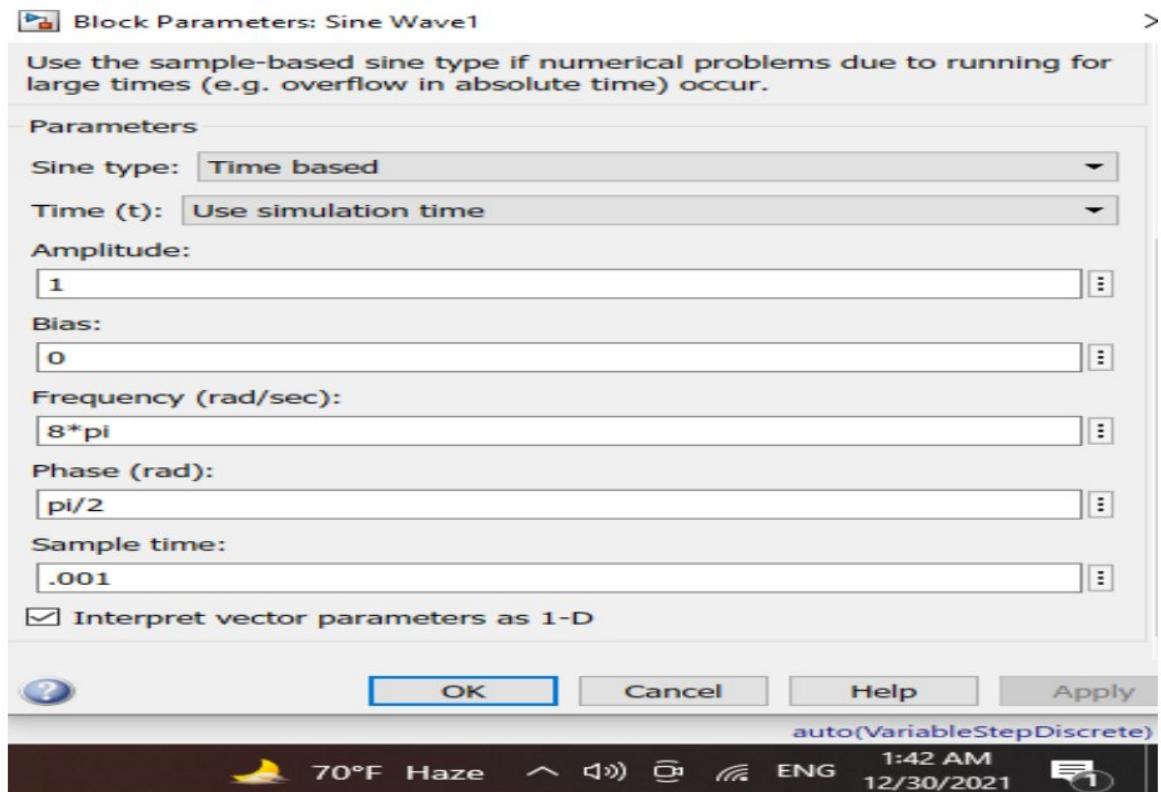
We vary Eb/No from 1 to: 20 with an increment of .5 to observe the bit error rate for both BPSK and QPSK communication channel.

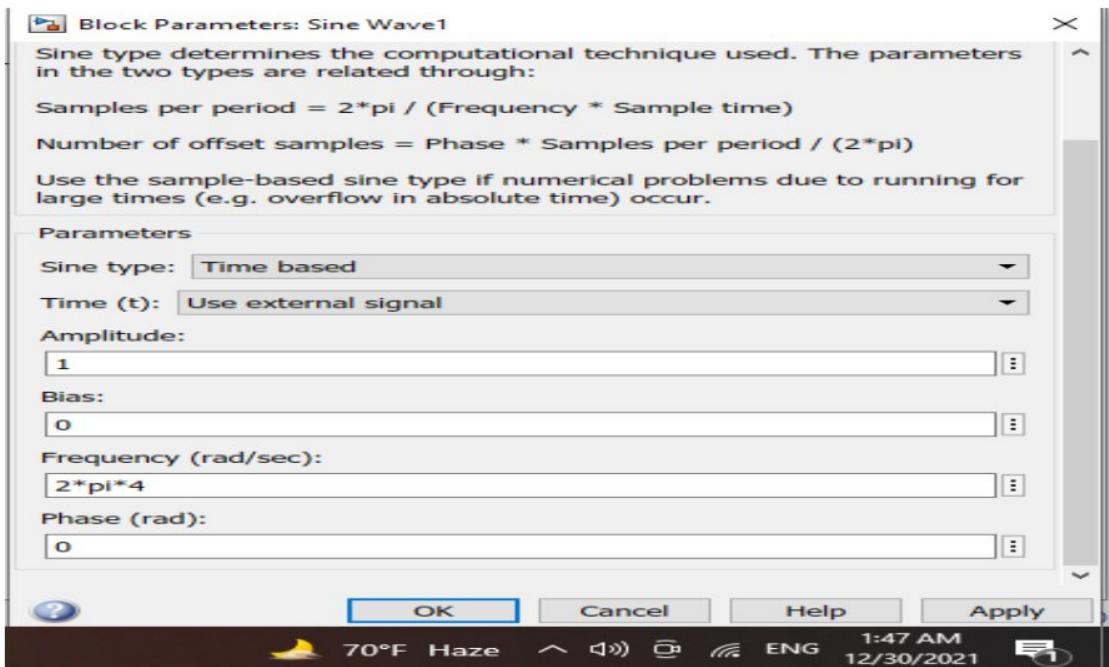


Constellation Diagram of modulator it defines the possible symbol that can be transmitted by the system as a collection of points.

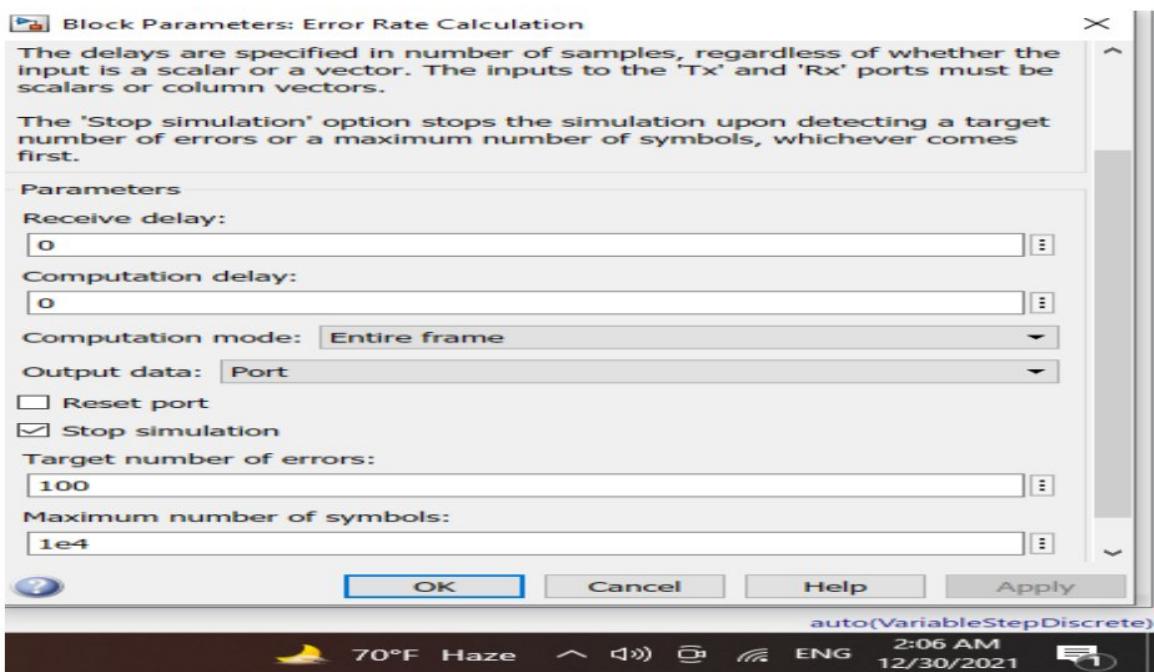


**Sine wave :** Sine wave: These signals are required to expand the complex modulated signal into real and imaginary parts.

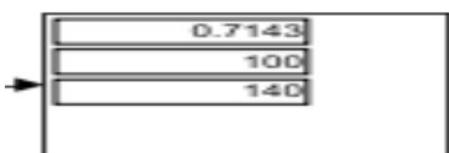


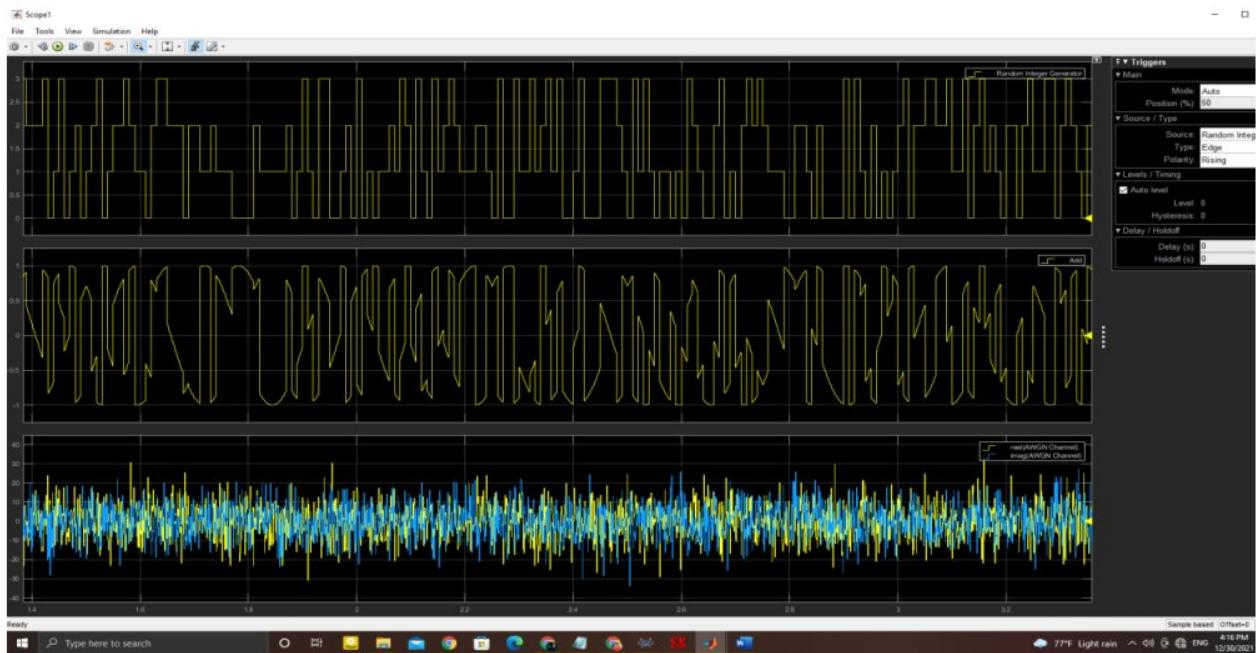


Error rate calculation is accommodated of maximum  $1e4$  symbol and count up to 100 number of errors as target.

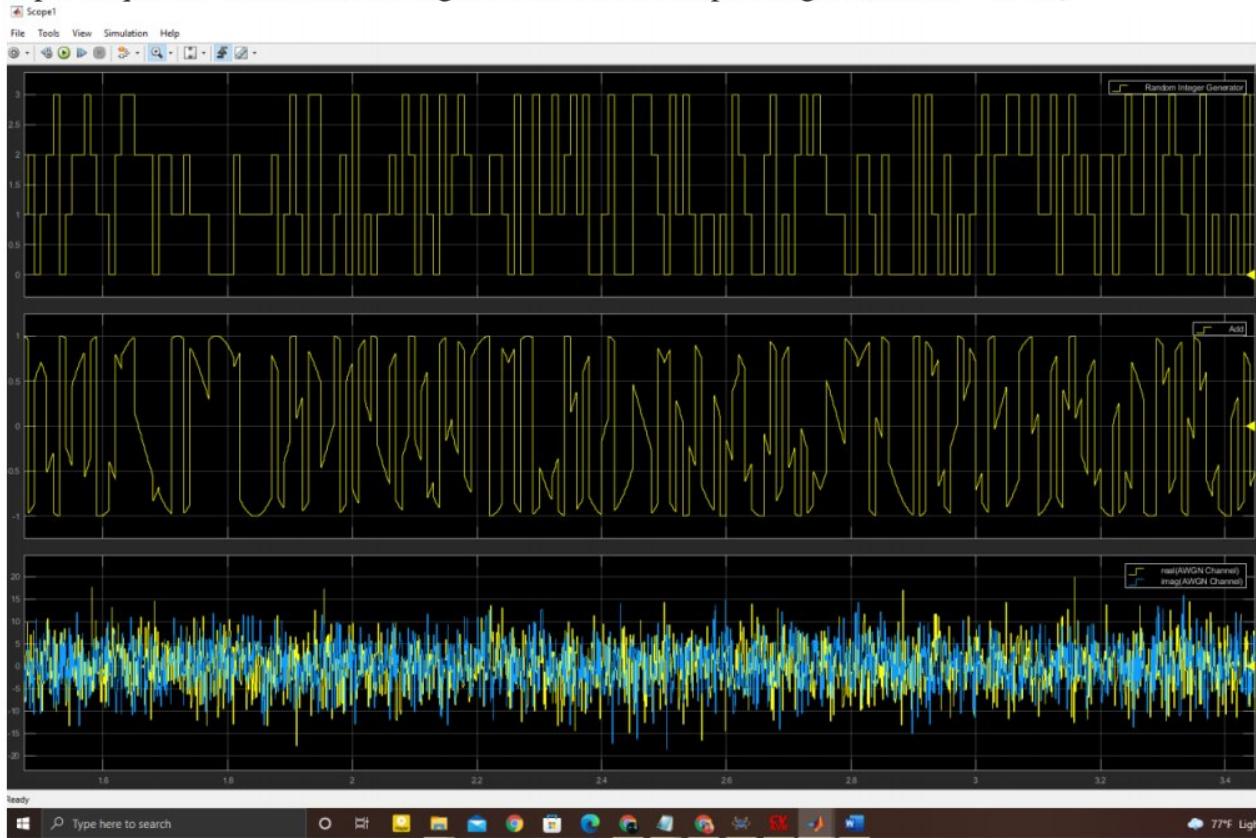


**Results:** i. Error Rate for  $E_b/N_0 = 5\text{dB}$

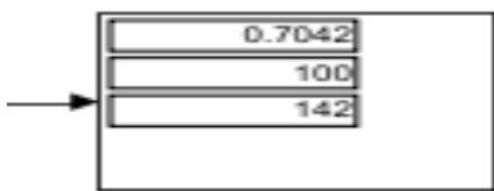




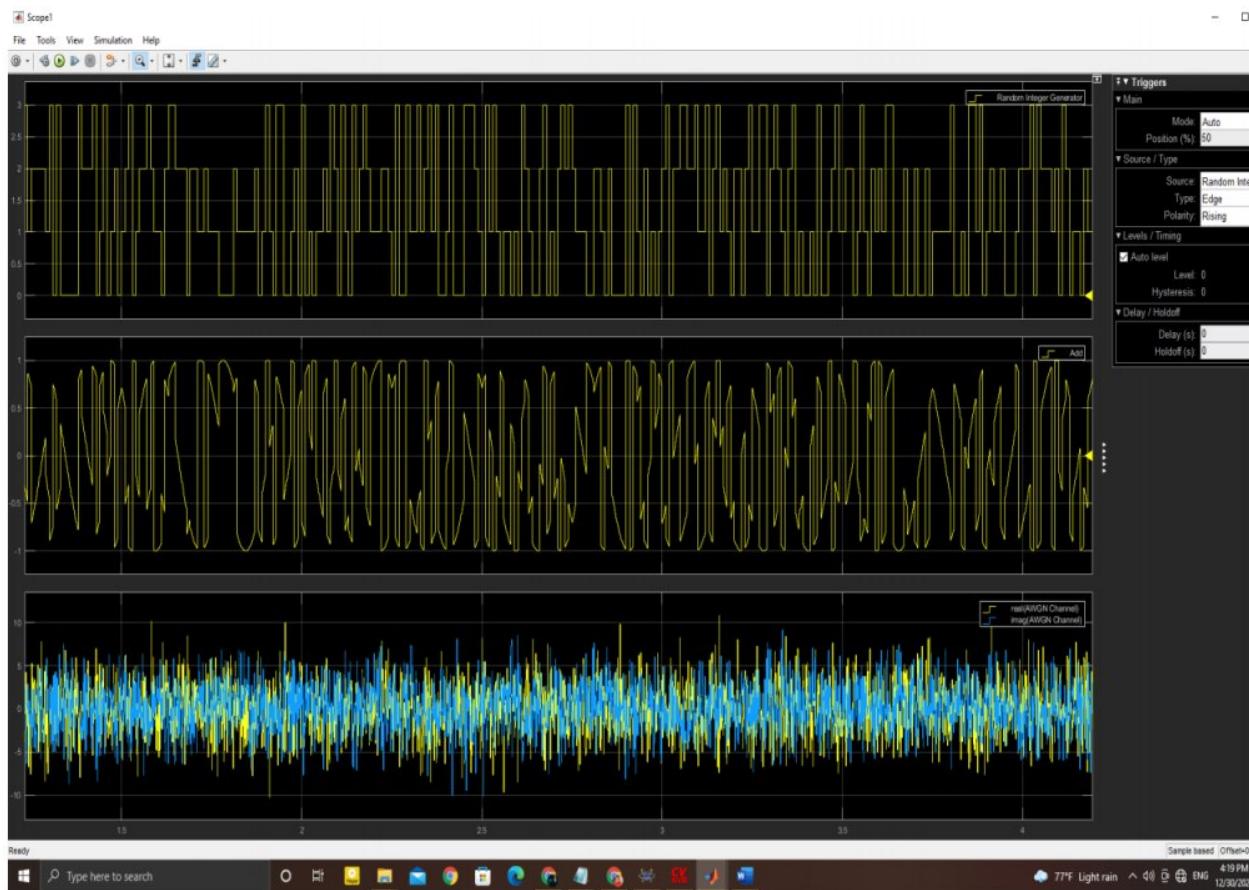
Input Sequence Vs. Modulated Signal Vs. Noise Corrupted Signal ( $E_b/N_0 = 10\text{dB}$ )



Error Rate For  $E_b/N_0 = 10\text{dB}$

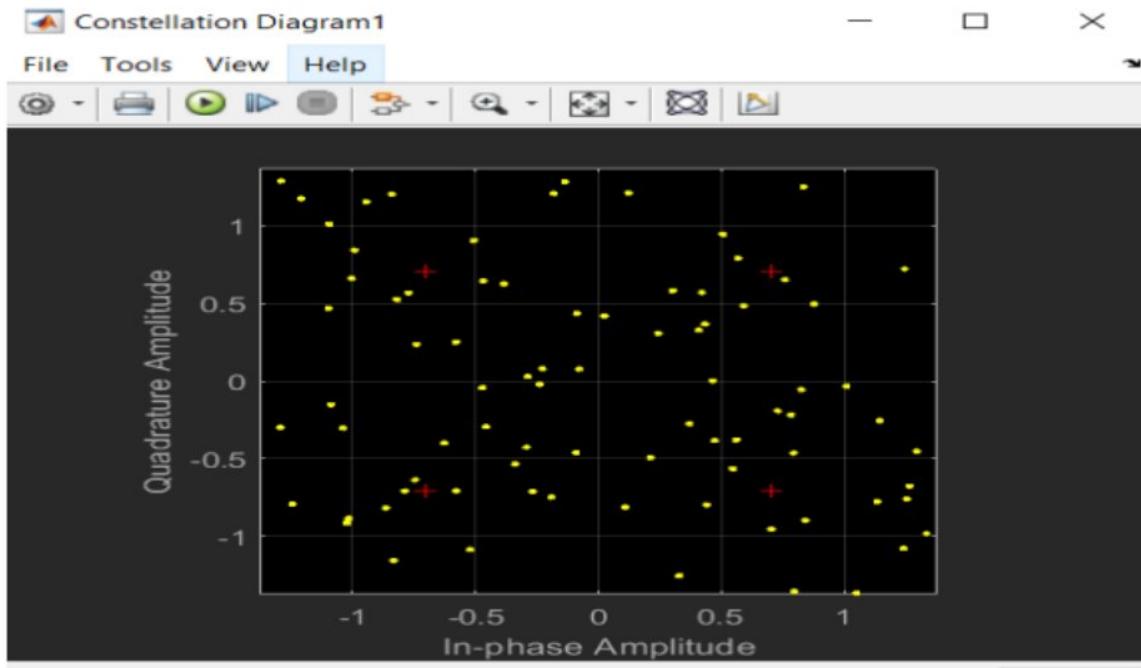


## Input Sequence Vs. Modulated Signal Vs. Noise Corrupted Signal (Eb/No = 15dB)

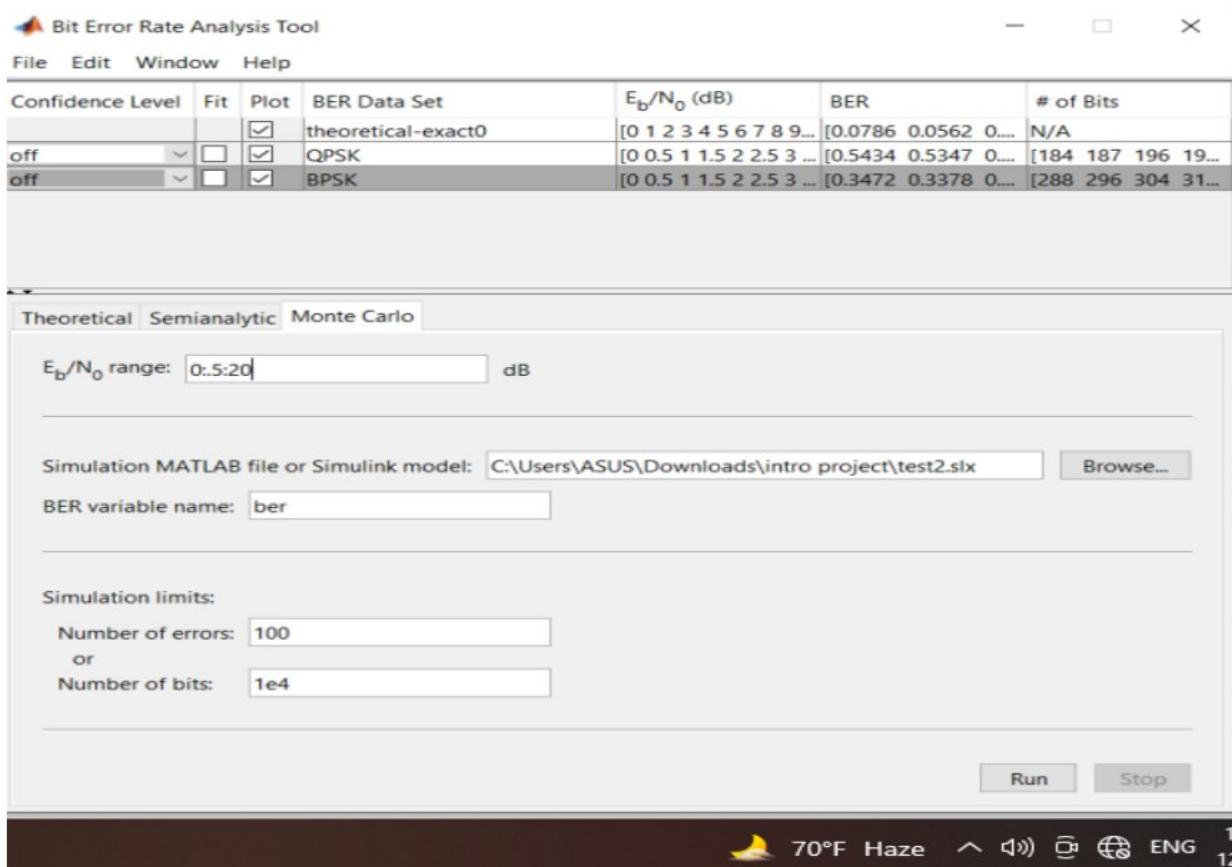


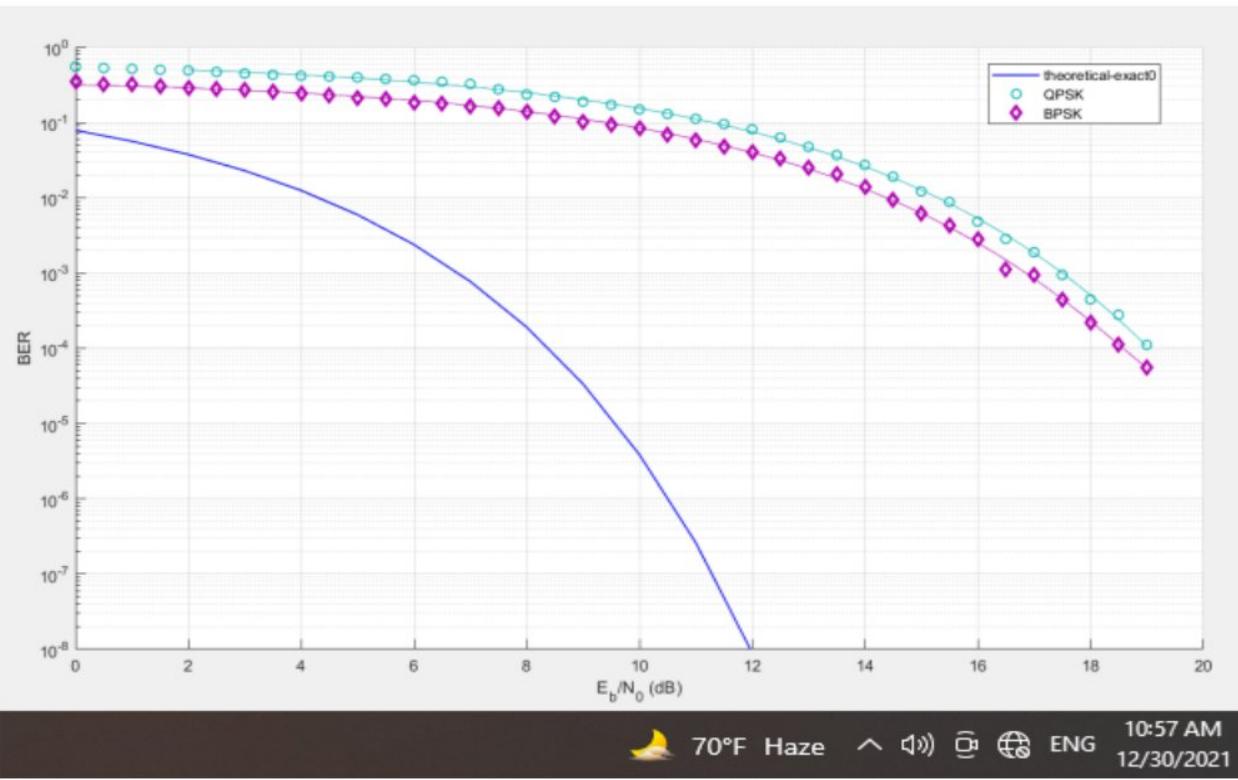
we observe a decreasing bit error rate with an increasing Eb/No in QPSK; But in case of BPSK, the error rate is even lower for the same level of Eb/No ratio. From our results, it is evident that BPSK modulation is preferred in cases where we need to consider small amounts of transmitting energy. The main reason is that the BPSK offers acceptable BER while transmitting signals of relatively low energy.

## Constellation Diagram of QPSK Modulator



BerTools and performance curve





70°F Haze ⌂ ENG 10:57 AM  
12/30/2021

We know QPSK uses higher power to transmit two bits simultaneously like BPSK. So the Eb/No ratio would be higher in case of QPSK compare to BPSK to achieve the same bit error probability. to achieve ber  $10^{-5}$

Eb/No for BPSK = 18.5 from curve

Eb/No=19 these are approx. value

So  $(E_b/N_0)_{BPSK}/(E_b/N_0)_{QPSK} = 18.5/19 = .97368$ =so its can be said that QPSK need higher signal power than BPSK to achieve same BER.

#### Comment on BPSK

One bit per symbol is sent in BPSK. It is capable of transmitting 1 bit per second. QPSK has a bit rate of  $2T$  and transmits two bits per symbol. QPSK can transmit at a rate of 2 bits per Hz in baseband. According to the performance curve, Bpsk requires a signal energy to noise level ratio of 18.5 to produce a BER of  $10^{-5}$ , but QPSK requires almost 19. Because each carrier step has a spectrum efficiency of 2 bits/Hz, BPSK is exceptionally spectrally efficient. At the same bandwidth as BPSK, it is feasible to produce double the data rate. Qpsk is therefore more efficient, although it consumes twice as much power as BPSK.

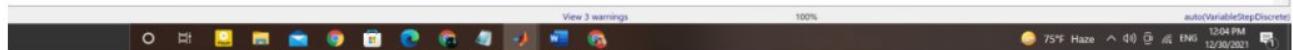
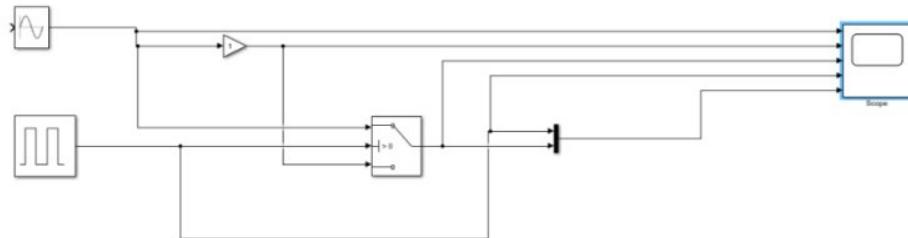
### **Task3:**

Here we asked to construct a BPSK modulator with operational function blocks on the time domains BPSK and evaluate the modulation.

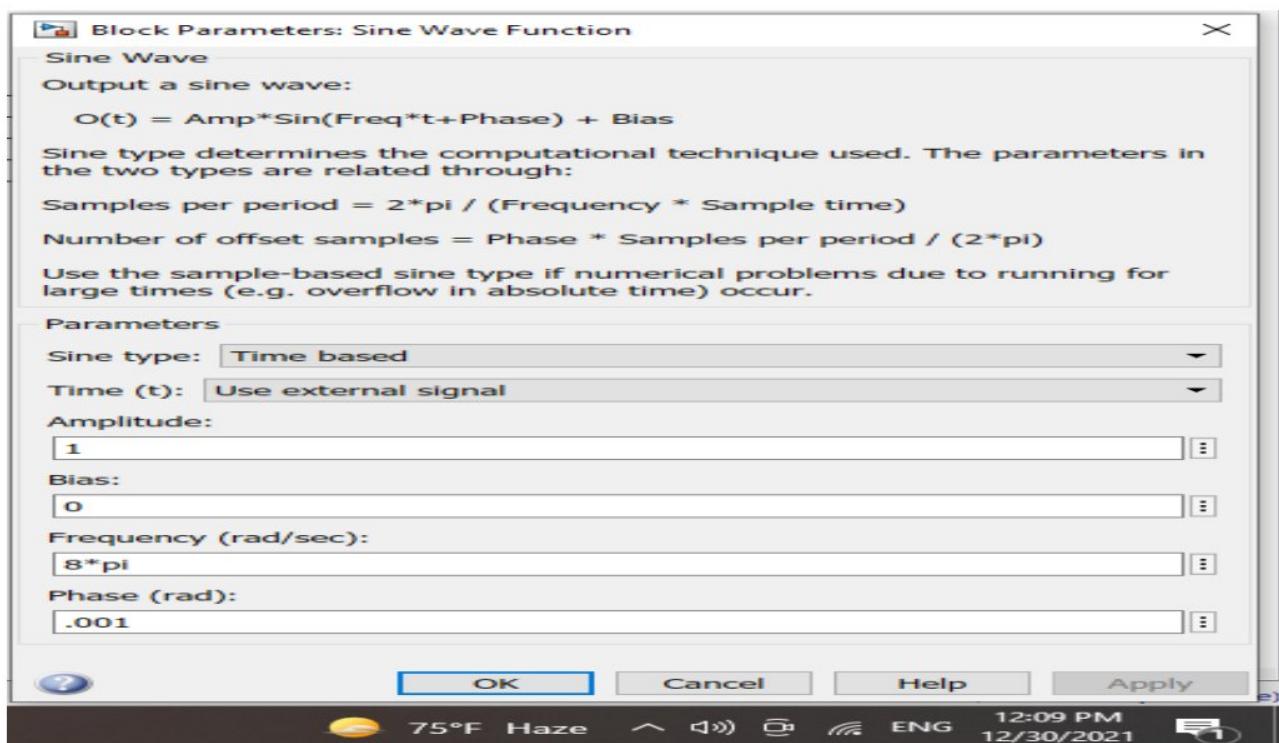
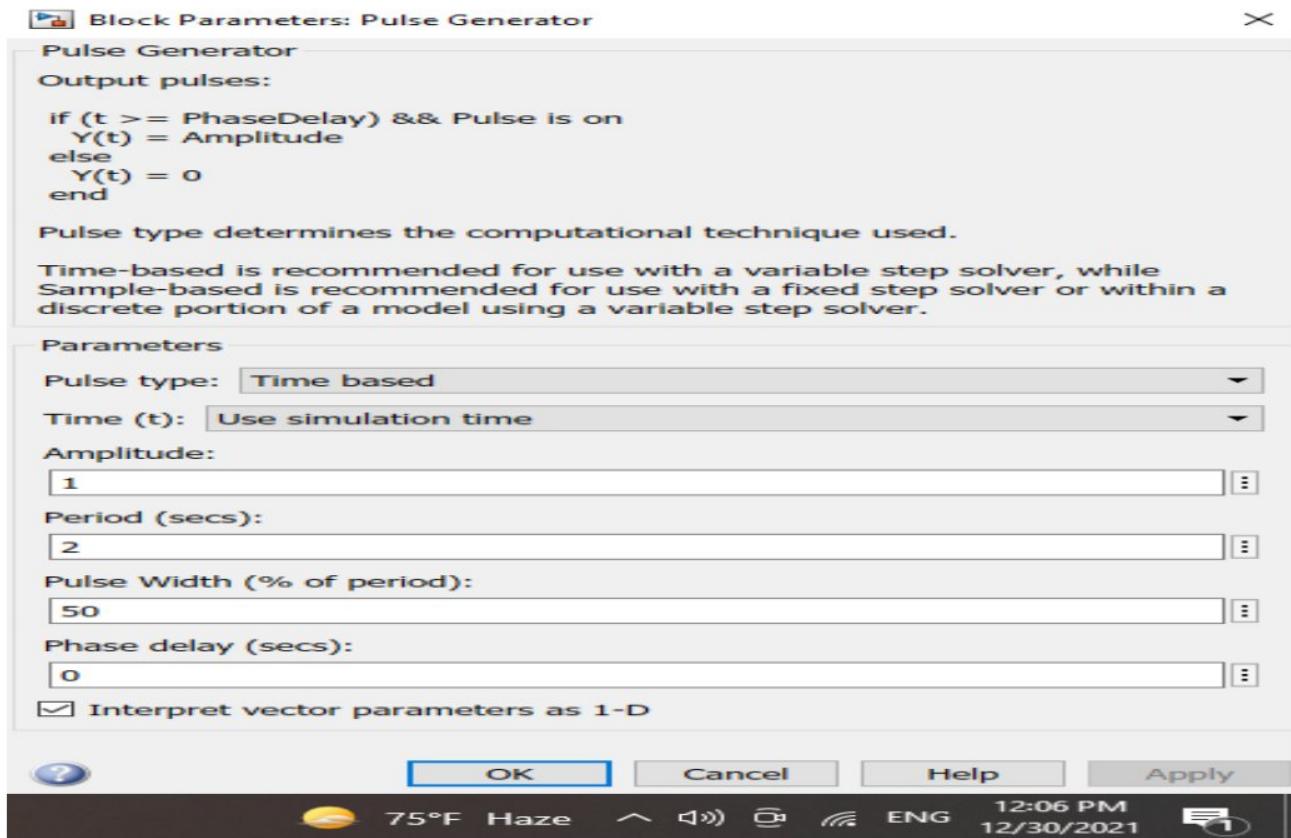
#### Simulink Blocks

1. pulse Generator
- 2.sine wave
- 3.Switch
4. Mux
- 5.Scope

The BPSK modulation is accomplished in Matlab simulink using a bit sequence generator as the input signal of a bit stream supplied to a switch that compares the threshold level of a signal with two sine wave carriers in opposing phase and frequency. The scope receives the switch's output, which displays the time-based modulated signal.



## Parameters:





### Discussion:

We have to use the built-in mode in order to modulate the input signal in terms of BPSK and QPSK. However, the modulated signal has a complicated phrase that is difficult to comprehend with scope. As a result, we employed complicated to create a Real-image. Only show the true part of the block. Furthermore, using the bertool function from MATLAB was the only way to observe BER versus Eb/No in Simulink. To attain the BER level 10-5, we had to produce roughly 2000 sequences of input samples, which took a long time. For a better perspective, we have to increase the amount of symbols displayed in the constellation block. For BPSK and the Eb/No versus Ber curve, the number of bits per symbol in the AWGN channel was 1. In the main AWGN block, we have to write variable term Eb/No as Eb/No value and real value in Bertool.

### Conclusion:

In this project we tried to plot circuit in Simulink and generate and observe results. As Eb/No grows, the bit error rate in QPSK and BPSK decreases. Although, given the same Eb/No ratio, the error rate in BPSK is far lower. Our findings suggest that BPSK modulation is favored when just a little amount of transmitting energy is needed. We established a communication link utilizing BPSK and QPSK and observed the waveforms in this section of our study. We experimented with several Eb/No values (5,10,15) to observe how they interact with the BER error rate.