

Reed Muller Codes Achieve Capacity on Erasure Channels¹

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The basic problem



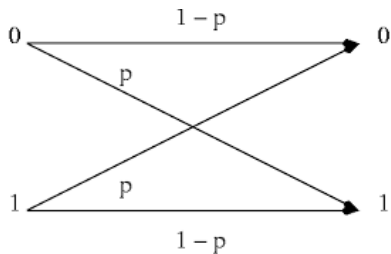
- Goal : Recover the **correct codeword** from a **noisy received codeword**
- Errors?

Communication Channel

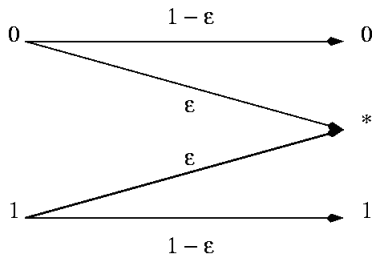
Abstraction of a physical transmission medium

- Input alphabet
- Output alphabet
- Conditional probability distribution $P(Y|X)$

Binary Symmetric Channel



Binary Erasure Channel



Symmetric, why?

A channel for which there exists a permutation π of the output alphabet Υ such that :

- $\pi^{-1} = \pi$
- $W(y|1) = W(\pi(y)|0) \quad \forall y \in \Upsilon$

We shall talk about BDMCs from here on.

Information theoretic parameters

Entropy of a discrete random variable X is defined to be

$$H(X) = - \sum_{x \in \mathcal{X}} p(x) \log p(x)$$

Information theoretic parameters

Mutual information between 2 discrete RVs X and Y is defined to be

$$I(X; Y) = \sum_{x \in \mathcal{X}} \sum_{y \in \mathcal{Y}} p(x, y) \log \frac{p(x, y)}{p(x)p(y)} = H(X) - H(X|Y)$$

A Coding Scheme

- Encoding
- Decoding

Parameters of a code

Say, for every k bits of information the code generates n bits.

- Rate of a code (R) = $\frac{k}{n}$
- Distance of code (d) = $\min_{C_1 \neq C_2 \in \mathcal{C}} \Delta(C_1, C_2)$

Capacity of a channel

$$C = \text{Max}_{p_X(x)} I(X; Y)$$

Shannon's random coding approach

Given a noisy channel with channel capacity C and information transmitted at a rate R , then if $R < C$ **there exist** codes that allow the probability of error at the receiver to be made arbitrarily small. The converse is also true. For $R > C$, an arbitrarily small probability of error is not achievable.

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Reed Solomon Codes

$$RS_{F,S,n,k}(m) = (f(\alpha_1), f(\alpha_2), f(\alpha_3), \dots, f(\alpha_n)),$$

where $f(X) = m_0 + m_1X + \dots + m_kX^k$

Vandermonde Matrix

$$\begin{bmatrix} x_1 & x_2 & x_3 & \dots & \dots & x_n \end{bmatrix} =$$

$$\begin{bmatrix} m_1 & m_2 & m_3 & \dots & \dots & m_n \end{bmatrix} * \begin{bmatrix} 1 & 1 & 1 & \dots & 1 \\ \alpha_1 & \alpha_2 & \alpha_3 & \dots & \alpha_n \\ \alpha_1^2 & \alpha_2^2 & \alpha_3^2 & \dots & \alpha_n^2 \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ \alpha_1^{k-1} & \alpha_2^{k-1} & \alpha_3^{k-1} & \dots & \alpha_n^{k-1} \end{bmatrix}$$

$$x_1^N = m_1^N G$$

Reed Muller Codes

Given a field size q and a number m of variables, and a total degree bound r , the $RM_{q,m,r}$ code is the code over F_q defined by the encoding map

$$f(X_1, X_2, \dots, X_m) \mapsto \langle f(\alpha) \rangle_{\alpha \in F_q^m}$$

Transitivity of Codes

1-transitivity

A code \mathcal{C} is said to be 1-transitive if for any j_1 and $j_2 \in [N]$ satisfying $j_1 \neq j_2$, there exists a permutation $\pi : [N] \rightarrow [N]$ such that :

- $\pi(j_1) = j_2$
- $y_{\pi(1)}, y_{\pi(2)} \dots y_{\pi(n)} \in \mathcal{C}$ for every $y_1, y_2 \dots y_n \in \mathcal{C}$

Transitivity of Codes

2-transitivity

A code \mathcal{C} is said to be 2-transitive if for any $j_1, j_2, j_3, j_4 \in [N]$ satisfying $j_1 \neq j_2$ and $j_3 \neq j_4$, there exists a permutation $\pi : [N] \rightarrow [N]$ such that :

- $\pi(j_1) = j_3$
- $\pi(j_2) = j_4$
- $y_{\pi(1)}, y_{\pi(2)} \dots y_{\pi(n)} \in \mathcal{C}$ for every $y_1, y_2 \dots y_n \in \mathcal{C}$

RS codes are 2-transitive

Proof :

RS codes are generated by the Vandermonde matrix G .

$$x_1^N = m_1^N G$$

Pick any permutation satisfying the constraints.

$$\text{Given : } x_1^N = m_1^N * G$$

$$\pi x_1^N = \pi m_1^N * G$$

$$x_{\pi(1)}^N = m_{\pi(1)}^N * G$$

Thus, $x_{\pi(1)}, x_{\pi(2)}, x_{\pi(3)} \dots x_{\pi(n)}$ is a codeword too.

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The channel model

- The i -th bit is transmitted through a binary erasure channel with erasure probability p . Denote this channel by $\text{BEC}(\underline{p})$.
- Uniform codeword assumption : A uniform distribution exists on space of codewords.

ML/MAP decoding

$$\hat{x}_{ML}(y) = \arg \max_x f(y|x)$$

$$\hat{x}_{MAP}(y) = \arg \max_x f(y|x)g(x)$$

MAP EXIT functions

The vector extrinsic information transfer function associated with the i th bit is defined to be

$$h_i(\underline{p}) \triangleq H(X_i | \underline{Y}_{\sim i}(\underline{p}_{\sim i}))$$

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The area theorem

The average EXIT function h satisfies the following :

$$\int_0^1 h(p) dp = \frac{K}{N}.$$

A characterization for the 'bad' patterns

Ω_i is the set of $w \in \{0, 1\}^{N-1}$ for which $\exists c \in \mathcal{C}$ such that $c_i = 1$ and $c_{\sim i} \preceq w$

The 'bad' set Ω_i

- Ω_i is monotone
- $\mu_{\underline{p}}(\Omega_i) = h_i(\underline{p}) = \sum_{A \in \Omega_i} \prod_{l \in A} p_l \prod_{l \in A^c \setminus \{i\}} (1 - p_l)$
- Ω_i is 1-transitive if \mathcal{C} is 2-transitive.

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Inching towards the goal, are we?

Let's quickly look at what we know until now.

- $h_i(p)$ captures the probability of error of MAP decoder.
- Ω_i encodes $h_i(p)$.
- All EXIT functions are equal to the average exist function h .
- The area under the h vs p curve is the rate.

Sharp thresholds on graphs

"Every **monotone** graph property has a sharp threshold"

The pivotal patterns

$$\partial_j \Omega \triangleq \{\underline{x} \in \{0, 1\}^N \mid \mathbb{1}_\Omega(\underline{x}) \neq \mathbb{1}_\Omega(\underline{x}^{(j)})\},$$

$$\text{where } \underline{x}^{(j)} = \begin{cases} x_l & \text{for } l \neq j \\ 1 - x_j & \text{otherwise} \end{cases}$$

Influences of the variables

The influence of bit $j \in [N]$ is defined by $I_j^{(p)}(\Omega) \triangleq \mu_p(\partial_j \Omega)$

$$\text{Total influence } I^{(p)}(\Omega) = \sum_{l=1}^N I_l^{(p)}(\Omega)$$

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Russo-Margulis lemma

If Ω be a monotone set, then $\frac{d\mu_p(\Omega)}{dp} = I^{(p)}(\Omega)$

All influences are equal

There exists a bijection between $(\partial_j \Omega_i)$ and $(\partial_k \Omega_i)$ for distinct $i, j, k \in [N]$

Modified Russo-Margulis

Let Ω be a montone set and suppose that, for all $0 \leq p \leq 1$, the influences of all bits are equal $I_1^{(p)}(\Omega) = I_2^{(p)}(\Omega) = \dots = I_N^{(p)}(\Omega)$. The following is true :

- 1 There exists an universal constant $C \geq 1$ which is independent of p , Ω and N , such that

$$\frac{d\mu_p(\Omega)}{dp} \geq C(\log N)(\mu_p(\Omega))(1 - \mu_p(\Omega))$$

- 2 For any $0 < \epsilon \leq \frac{1}{2}$,

$$p_{1-\epsilon} - p_\epsilon \leq \frac{2}{C} \frac{\log \frac{1-\epsilon}{\epsilon}}{\log M}$$

where $p_t \triangleq h^{-1} = \inf\{p \in [0, 1] | h(p) \geq t\}$ is the inverse function for the average EXIT function².

² $h(p)$ is a strictly increasing continuous polynomial function and hence inverse is well-defined on $[0, 1]$

Dictator functions

$$f(x_1, x_2, x_3, \dots, x_n) = f(x_i) \text{ for some } i \in [N] .$$

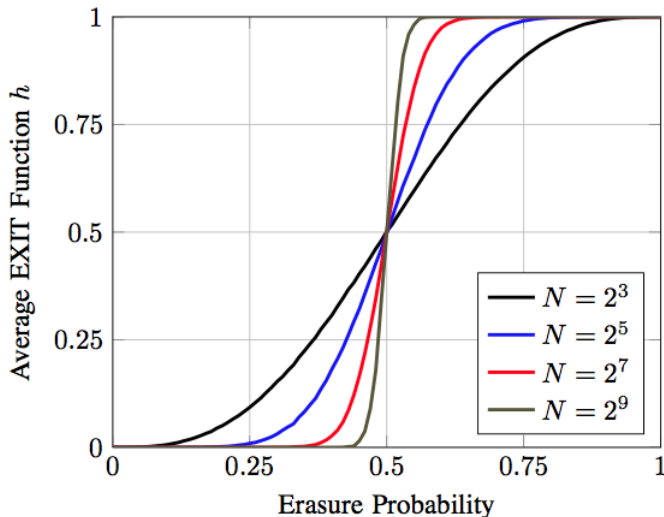
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Reed Muller and polar codes achieve capacity on BECs



Further Reading

-  Shrinivas Kudekar, Marco Mondelli, Eren Sasoglu, Rudiger Urbanke "*Reed-Muller Codes Achieve Capacity on the Binary Erasure Channel under MAP Decoding.*", arXiv:1505.0583, 2015.
-  Santhosh Kumar and Henry D. Pfister, "*Reed-Muller Codes Achieve Capacity on Erasure Channels*", arXiv:1505.05123, 2015.
-  Venkatesan Guruswami, Atri Rudra, "*Error-correction up to the information-theoretic limit*" Communications of the ACM, Volume 52 Issue 3, March 2009.
-  C. E. Shannon, "*A mathematical theory of communication*" ACM SIGMOBILE Mobile Computing and Communications Review, Volume 5 Issue 1, January 2001.

Thank You.

Proposition: (i) A generator matrix of $\mathcal{R}(1,1)$ is

$$\begin{pmatrix} 1 & 1 \\ 0 & 1 \end{pmatrix}$$

(ii) If G_m is a generator matrix for $\mathcal{R}(1,m)$, then a generator matrix for $\mathcal{R}(1,m+1)$ is

$$G_{m+1} = \begin{pmatrix} G_m & G_m \\ 0 \cdots 0 & 1 \cdots 1 \end{pmatrix}$$

$$\Omega_i \triangleq \{A \subseteq [N] \setminus i \mid \exists B \subseteq [N] \setminus i, B \cup \{i\} \in \mathcal{C}, B \subseteq A\}$$