

## **A THEOREM IN RELATIVISTIC ELECTRONICS**

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### **Abstract**

This paper presents a theorem that connects the dispersion relation of the Electron Cyclotron Maser and the oscillation equation of the Gyromonotron. This theorem gives us a simple way of obtaining the oscillating characteristics of the Gyromonotron provided that dispersion relation of the ECRM is given. Though the theorem is proved only with the case of ECRM and Gyromonotron, it holds for other kinds of Electron Masers, FEL etc. and corresponding oscillators.

### **I. Introduction**

Consider an oscillator composed by an open uniform waveguide cavity and a helix electron beam as shown in figure 1. We want to decide the normal oscillating angular frequencies and gains of the exciting electromagnetic mode in the cavity.

Suppose that the normalized longitudinal field distribution function of the cavity is  $g(z)$  with  $\int |g(z)|^2 dz = 1$  and its Fourier transform is given by

$$\tilde{g}(k) = \int_{-\infty}^{\infty} g(z) \exp(-jkz) dz$$

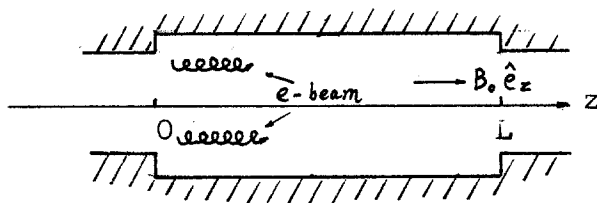


Fig.1 Configuration of Gyromonotron

And suppose that the dispersion relation for  $TE_{mn}$  mode s-harmonic electron cyclotron maser is

$$D(k, jp) = 0$$

in which

$$D(k, jp) = p^2 + (k_z^2 + k^2) c^2 - \frac{\omega_p^2}{\omega} \left[ \frac{p^2 + k^2 c^2}{(jp - kv_z - s\omega_c)^2} \beta_z^2 H_{ms} + \frac{jp - kv_z}{jp - kv_z - s\omega_c} Q_{ms} + U_{ms} \right]$$

Theorem: The normal oscillating frequencies and gains of the stimulated electromagnetic mode of the gyromonotron is given by

$$\langle D(k, jp) \rangle = 0$$

here  $\langle (\dots) \rangle$  is defined by

$$\langle \dots \rangle = \frac{1}{2\pi} \int \tilde{g}(k) \tilde{g}^*(k) (\dots) dk$$

## II. Proof of the theorem

Considering a uniform cylindrical waveguide cavity which is being stimulated by tenuous perturbing current  $\mathbf{J}(\mathbf{x}, t)$ , and assuming that only a  $TE_{mn}$  mode be excited to be oscillating, we can express the radiation fields in the cavity as

$$\begin{aligned}
E_R(\mathbf{x}, \omega) &= -j\omega\mu_0 \frac{j\mathfrak{m}}{R} J_m(k_c R) e^{jm\varphi} g_\alpha(z) D(\omega) \\
E_\varphi(\mathbf{x}, \omega) &= j\omega\mu_0 k_c J'_m(k_c R) e^{jm\varphi} g_\alpha(z) D(\omega) \\
E_z(\mathbf{x}, \omega) &= 0 \\
H_R(\mathbf{x}, \omega) &= k_c J'_m(k_c R) e^{jm\varphi} g'_\alpha(z) D(\omega) \\
H_\varphi(\mathbf{x}, \omega) &= \frac{j\mathfrak{m}}{R} J_m(k_c R) e^{jm\varphi} g'_\alpha(z) D(\omega) \\
H_z(\mathbf{x}, \omega) &= k_c^2 J_m(k_c R) e^{jm\varphi} g_\alpha(z) D(\omega)
\end{aligned} \tag{1}$$

in which  $k_c = \chi'_{mn}/a_w$ ,  $\chi'_{mn}$  is the  $n$ th root of equation  $J'(x)=0$ ,  $a_w$  is the radius of the waveguide,  $g_\alpha(z)$  is the  $\alpha$ th longitudinal mode distribution function of the cavity, and  $D(\omega)$  is the Fourier transform of the amplitude of the oscillation.

Transforming (1) back into the version of time domain, we have that

$$\begin{aligned}
E_R(\mathbf{x}, t) &= -\mu_0 \frac{j\mathfrak{m}}{R} J_m(k_c R) e^{jm\varphi} g_\alpha(z) \dot{D}(t) \\
E_\varphi(\mathbf{x}, t) &= k_c \mu_0 J'_m(k_c R) e^{jm\varphi} g_\alpha(z) \dot{D}(t) \\
E_z(\mathbf{x}, t) &= 0 \\
H_R(\mathbf{x}, t) &= k_c J'_m(k_c R) e^{jm\varphi} g'_\alpha(z) D(t) \\
H_\varphi(\mathbf{x}, t) &= \frac{j\mathfrak{m}}{R} J_m(k_c R) e^{jm\varphi} g'_\alpha(z) D(t) \\
H_z(\mathbf{x}, t) &= k_c J_m(k_c R) e^{jm\varphi} g_\alpha(z) D(t)
\end{aligned} \tag{2}$$

in which  $\dot{D}(t) = dD(t)/dt$ .

Substituting (2) into the Maxwell equation

$$\nabla \times \mathbf{H} = \epsilon_0 \frac{\partial \mathbf{E}}{\partial t} + \mathbf{J}(\mathbf{x}, t) \tag{3}$$

and applying the normalization technique, we find the equation for  $D(t)$ :

$$\begin{aligned}
\ddot{D}(t) g_\alpha(z) + c^2 \left[ k_c^2 g_\alpha(z) - g''_\alpha(z) \right] D(t) = \\
\frac{-1}{\epsilon_0 N_{mn}^2} \int_{S_w} \mathbf{J}_\perp(\mathbf{x}, t) \cdot \mathbf{E}^*(R, \varphi) d\sigma
\end{aligned} \tag{4}$$

in which

$$\mathbf{E}_\perp(R, \varphi) = -\mu_0 \frac{j\omega}{R} J_m(k_c R) e^{jm\varphi} \hat{\mathbf{e}}_R + k_c \mu_0 J'_m(k_c R) e^{jm\varphi} \hat{\mathbf{e}}_\varphi$$

$$N_{mn} = \pi k_c^2 \mu_0 a_w^2 \left(1 - \frac{m}{k_c^2 a_w^2}\right) J_m^2(k_c a_w)$$

$$\mathbf{J}_\perp(\mathbf{x}, t) = -|e| \int_{s_w} \left[ v_\perp \cos(\Phi - \varphi) \hat{\mathbf{e}}_R + v_\perp \sin(\Phi - \varphi) \hat{\mathbf{e}}_\varphi \right] f_1 d\mathbf{P}$$

with  $d\sigma$  denoting  $RdRd\varphi$ ,  $s_w$  representing the cross section of the waveguide, and  $f_1$  standing for the perturbed distribution function of the e-beam.  $(R, \varphi, z)$  are the cylindrical coordinates with axis  $z$  along the axis of the cavity.

Eq. (4) describes the action of the beam upon the radiation field in the cavity; the influence of the radiation field upon the beam is governed by the linearized Vlasov equation:

$$\frac{\partial f_1}{\partial t} + \mathbf{v} \cdot \nabla f_1 - |e| \mathbf{v} \times \mathbf{B}_0 \hat{\mathbf{e}}_z \cdot \nabla_{\mathbf{p}} f_1 = |e| [\mathbf{E}(\mathbf{x}, t) + \mathbf{v} \times \mathbf{B}(\mathbf{x}, t)] \cdot \nabla_{\mathbf{p}} f_0 \quad (5)$$

For simplicity, we introduce a six-dimensional phase coordinates  $(R_g, \varphi_g, r_L, \Theta, p_\parallel, z)$  to replace the general coordinates  $(x, y, z, p_x, p_y, p_z)$ , which are defined by

$$\begin{cases} R_g = [(x - p_y/a)^2 + (y + p_x/a)^2]^{1/2} \\ \varphi_g = \arctg(-\frac{y + p_x/a}{x - p_y/a}) \\ r_L = (p_x^2 + p_y^2)^{1/2}/a \\ \Theta = \arctg(p_y/p_x) - az/p_z - \pi/2 \\ p_\parallel = p_z \end{cases} \quad (6)$$

where  $a = |e| B_0$ .

In terms of the introduced coordinates  $(R_g, \varphi_g, r_L, \Theta, p_\parallel, z)$ , Eq. (5) can be reduced to the following form

$$\frac{\partial f_1}{\partial t} + v_\parallel \frac{\partial f_1}{\partial z} = T_1 \frac{\partial f_0}{\partial R_g} + T_2 \frac{\partial f_0}{\partial \varphi_g} + T_3 \frac{\partial f_0}{\partial r_L} + T_4 \frac{\partial f_0}{\partial \Theta} + T_5 \frac{\partial f_0}{\partial p_\parallel} \quad (7)$$

in which

$$\begin{aligned} T_1 &= \frac{|e|}{a} [(E_R - v_\parallel B_\varphi) \sin(\varphi_g - \varphi) - (E_\varphi + v_\parallel B_R) \cos(\varphi_g - \varphi) + \\ &\quad v_\perp B_z \sin(\varphi_g - \Theta)] \\ T_2 &= \frac{|e|}{a R_g} [(E_R - v_\parallel B_\varphi) \cos(\varphi_g - \varphi) + (E_\varphi + v_\parallel B_R) \sin(\varphi_g - \varphi) + \end{aligned}$$

$$\begin{aligned}
T_3 &= \frac{|e|}{a} \left[ (E_R - v_{\parallel} B_{\varphi}) \sin(\varphi - \theta) + (E_{\varphi} + v_{\parallel} B_R) \cos(\varphi - \theta) \right] \\
T_4 &= \frac{|e|}{ar_L} \left\{ -(E_R - v_{\parallel} B_{\varphi}) \cos(\varphi - \theta) + (E_{\varphi} + v_{\parallel} B_R) \sin(\varphi - \theta) \right. \\
&\quad \left. - v_{\perp} B_z - z \frac{p_z^2}{P_z^2} \omega_c [B_R \cos(\varphi - \theta) - B_{\varphi} \sin(\varphi - \theta)] \right\} \\
T_5 &= |e| \left\{ E_z - v_{\perp} [B_R \cos(\varphi - \theta) - B_{\varphi} \sin(\varphi - \theta)] \right\}
\end{aligned}$$

with  $\theta = \theta + az/p$ .

It should be noted that, in deriving (7), we have utilized the geometry relations of (6) as shown in figure 2.

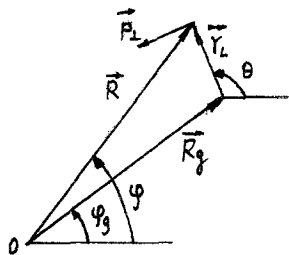


Fig.2 Geometry of coordinate transformation

Fourier-Laplace transforming Eqs.(4) and (7), assuming that the electron beam is axially symmetric yielding that  $\partial f_0 / \partial \varphi_3 = \partial f_0 / \partial \theta = 0$ , and then combining the two resulting equations, we have that

$$(p^2 + \omega_{mn}^2) \tilde{g}_\alpha(k) \tilde{D}(p) = D_I(k, p) + D_{II}(k, p) \quad (8)$$

where

$$\begin{aligned}
D_I(k, p) &= \left[ p D(t=0) + \dot{D}(t) \Big|_{t=0} \right] \tilde{g}_\alpha(k) + \frac{k_c \mu_0 |e| a}{\epsilon_0 N_{mn}} \sum_{s=-\infty}^{\infty} \\
&\quad \left( \int \int \int \int R_g r_L dR_g d\varphi_g dr_L d\theta dp_{\parallel} v_{\perp} J_{m-s}(k_c R_g) \cdot \right. \\
&\quad J'_1(k_c r_L) \frac{f_1(k + s\omega_c/v_{\parallel}, t=0)}{p + jkv_{\parallel} + js\omega_c} e^{-j(m-s)\varphi_g - js\theta} \\
&\quad - \frac{8\pi^3 k_c^2 \mu_0^2 e^2}{\epsilon_0 N_{mn}} D(t=0) \sum_s \left( \int \int \int R_g r_L dR_g dr_L dp_{\parallel} v_{\perp} \right. \\
&\quad J_{m-s}(k_c R_g) J'_1(k_c r_L) \tilde{g}_\alpha(k) \left[ J'_{m-s}(k_c R_g) J_s(k_c r_L) \right. \\
&\quad \left. \left. \cdot \frac{\partial f_0}{\partial R_g} + J_{m-s}(k_c R_g) J'_1(k_c r_L) \frac{\partial f_0}{\partial r_L} \right] (p + jkv_{\parallel} + js\omega_c)^{-1} \right)
\end{aligned}$$

$$D_{\mathbb{I}}(k, p) = -8\pi^3 \frac{k_c^2 \mu_0^2 e^2}{\epsilon_0 N_{mn}^2} \tilde{D}(p) \tilde{g}_\alpha(k) \sum_{\mathbb{I}} \left( \int_0^\infty R_g r_L dR_g dr_L dp_{\parallel} \cdot \right. \\ \left. v_L J_{m-s}(k_c R_g) J_s'(k_c r_L) \left\{ \left[ -J_{m-s}'(k_c R_g) J_s'(k_c r_L) \right. \right. \right. \\ \left. \left. + \frac{m-s}{R_g} \frac{j v_L}{p + j k v_{\parallel} + j s \omega_c} J_{m-s}(k_c R_g) J_s'(k_c r_L) \right] \cdot \right. \\ \left. \frac{\partial f_0}{\partial R_g} + \left[ (p + j k v_{\parallel}) \frac{\partial f_0}{\partial r_L} - j a v_L k \frac{\partial f_0}{\partial p_{\parallel}} \right] J_{m-s}(k_c R_g) \cdot \right. \\ \left. \left. J_s'(k_c r_L) \frac{1}{p + j k v_{\parallel} + j s \omega_c} \right\} \right)$$

in which  $\tilde{D}(p) = \int_0^\infty D(t) \exp(-pt) dt$ ,  $\tilde{g}_\alpha(k) = \int_0^\infty g_\alpha(z) \exp(-jkz) dz$ ,  $f_i(k, t) = \int_0^\infty f_i(z, t) \exp(-jkz) dz$ ,  $\omega_{mn}^2 = (k^2 + k_c^2) c^2$ .

Choosing such a  $f_i(k, t=0)$  that  $D_{\mathbb{I}}(k, p) = [pD(t=0) + \dot{D}(t)|_{t=0}] \tilde{g}_\alpha(k)$  and taking the equilibrium distribution function as

$$f_0 = \frac{N_e}{4\pi R_g r_L a} \delta(R_g - R_{g0}) \delta(r_L - r_{L0}) \delta(p_{\parallel} - p_{\parallel 0}),$$

we can simplify (8) to the following form

$$(p^2 + \omega_{mn}^2) \tilde{g}_\alpha(k) \tilde{D}(p) = [pD(t=0) + \dot{D}(t)|_{t=0}] \tilde{g}_\alpha(k) + \\ \frac{k_c^2 \mu_0^2 e^2 N_e}{\epsilon_0 N_{mn}^2 m_s} \sum_{\mathbb{I}} \left[ \frac{p^2 + k^2 c^2}{(j p - k v_{\parallel 0} - s \omega_{c0})^2} \beta_{\omega}^2 H_{ms} + \right. \\ \left. \frac{j p - k v_{\parallel 0}}{j p - k v_{\parallel 0} - s \omega_{c0}} Q_{ms} + U_{ms} \right] \tilde{g}_\alpha(k) \tilde{D}(p) \quad (9)$$

in which

$$H_{ms} = J_{m-s}^2(k_c R_{g0}) J_s'^2(k_c r_{L0})$$

$$Q_{ms} = 2H_{ms} + 2k_c r_{L0} J_{m-s}(k_c R_{g0}) J_s''(k_c r_{L0}).$$

$$\left\{ J_{m-s}(k_c R_{g0}) J_s''(k_c r_{L0}) + \frac{m-s}{s} \frac{r_L}{R_g} J_{m-s}'(k_c R_{g0}) J_s'(k_c r_{L0}) \right\} \\ U_{ms} = k_c r_{L0} J_s'(k_c r_{L0}) \left\{ J_s(k_c r_{L0}) \left[ \left( 1 - \left( \frac{m-s}{k_c R_{g0}} \right)^2 \right) J_{m-s}^2(k_c R_{g0}) - \right. \right. \\ \left. \left. J_{m-s}'(k_c R_{g0}) \right] - 2 \frac{m-s}{s} J_s'(k_c r_{L0}) J_{m-s}(k_c R_{g0}) J_{m-s}'(k_c R_{g0}) \right\}$$

and  $N_e$  is the electron beam density.

Multiplying (9) by  $\tilde{g}^*(k)$ , and then integrating it with respect to  $k$ , we can arrive at that

$$\tilde{D}(p) = \frac{pD(t=0) + \dot{D}(t)|_{t=0}}{\langle D(k, jp) \rangle} \quad (10)$$

in which

$$D(k, jp) = p^2 + \omega_{mn}^2 - \frac{\omega_p^2}{j\omega} \sum_s \left[ \frac{p^2 + k^2 c^2}{((jp - kv_{so} - s\omega_{cs})^2 + \beta_{so}^2 H_{ms}^2)} + \frac{jp - kv_{so}}{jp - kv_{so} - s\omega_{cs}} Q_{ms} + U_{ms} \right],$$

$\langle (\dots) \rangle$  is defined by

$$\langle (\dots) \rangle = \frac{1}{2\pi} \int \tilde{g}_\alpha(k) g_\alpha^*(k) (\dots) dk$$

and  $\omega_p^2 = k^2 \mu_0^2 e^2 N_0 / (\epsilon_0 m_0 N_{mn})$ .

From Eq.(10), we know that the long term characteristic of oscillation is determined by solving equation

$$\langle D(k, jp) \rangle = 0 \quad (11)$$

for  $jp$  or  $\omega = jp$ . Therefore Eq.(11) can be referred as the characteristic equation of the gyrotron based on ECRM interaction.

Thus we have proved the theorem in the case of gyrotron. In view of above proof procedure, it is evident that the theorem holds for other kinds of HF oscillators based upon corresponding maser and FEL interactions.

According to the theorem, to determine the characteristic frequencies of an oscillator based upon some electron beam instability, one just need to

1) find the dispersion relation of the beam instability  $D(k, \omega) = 0$ ,

2) write down the normalized longitudinal cavity field distribution function  $g(z)$  and its Fourier transformation  $\tilde{g}(k)$ ,

3) solve  $\langle D(k, \omega) \rangle = 0$  for  $\omega$ .

It should be pointed out that, other than Eq.(11), the theorem can take different form. Let us come back to expression (9). Multiplying (9) by  $\prod_s (jp - kv_{se} - s\omega_{ce})^2$  before performing the operator  $\langle \dots \rangle$  on it, we have that

$$D(p) = \frac{\langle \prod_s (jp - kv_{se} - s\omega_{ce})^2 \rangle [pD(t=0) + \dot{D}(t)|_{t=0}]}{\langle \prod_s (jp - kv_{se} - s\omega_{ce})^2 D(k, jp) \rangle} \quad (12)$$

From (12), we know that the normal oscillation characteristic is governed by

$$\langle \prod_s (jp - kv_{se} - s\omega_{ce})^2 D(k, jp) \rangle = 0 \quad (13)$$

We shall see in next section that Eq.(13) is more convenient and easier to solve than (11).

In practical situation, the cold' cavity resonance frequency  $jp^{(0)}$  is chosen to be close to one of the cyclotron frequency harmonic, i.e.,  $jp^{(0)} \approx l\omega_{ce}$ ; and because the electron beam density is sufficiently low, the solution to (13) is in fact near  $jp^{(0)}$ . So we can drop non-resonance terms, i.e., terms with  $s \neq l$ , from  $D(k, jp)$ , and then Eq.(13) can be approximated by

$$\langle (jp - kv_{se} - l\omega_{ce})^2 D_0(k, jp) \rangle = 0 \quad (13')$$

in which  $D_0(k, jp)$  is the simplified  $D(k, jp)$ .

### III. Application of theorem

To illustrate how to use the theorem, let us consider an example of open cavity gyrotron with its axial field distribution function given by

$$g(z) = \sqrt{\frac{2}{L}} \sin(k_{\parallel} z) \text{rect}\left(\frac{z-L/2}{L}\right) \quad (14)$$

here  $k_{\parallel} = \pi/L$ ,  $L$  is the cavity length.

It is easy to obtain the Fourier transform of (14):

$$\tilde{g}(k) = \sqrt{\frac{2}{L}} \frac{2k_{\parallel}}{k_{\parallel}^2 - k^2} \exp(-jkL/2) \quad (15)$$

Substituting (15) and its complex conjugate into Eq.(11), we get that



$$\omega^2 - \Omega^2 - \frac{\omega_p^2}{\gamma_0} \sum_s \left( \beta_{1s}^2 H_{ms} (\omega^2 - k_{\parallel}^2 c^2) I_s^{(1)} + (1 - j s \omega \omega_p I_s^{(2)}) Q_{ms} + U_{ms} \right) = 0 \quad (16)$$

in which  $\Omega^2 = (k_{\perp}^2 + k_{\parallel}^2) c^2$ ,

$$\begin{aligned} I_s^{(1)} = & \frac{2k_{\parallel}}{v_{\parallel 0}^2} \left\{ \left( L + \frac{1}{jk_{\parallel}} \right) \left[ -\frac{v_{\parallel 0}}{j\Omega_-} L e^{-j\Omega_- L / v_{\parallel 0}} + \frac{v_{\parallel 0}^2}{\Omega_-^2} (e^{-j\Omega_- L / v_{\parallel 0}} - 1) \right] - \left\{ \frac{jv_{\parallel 0} L^2}{\Omega_-^2} e^{-j\Omega_- L / v_{\parallel 0}} - j \frac{2v_{\parallel 0}}{\Omega_-} \left( j \frac{v_{\parallel 0} L}{\Omega_-} \exp(-j\Omega_- L / v_{\parallel 0}) + \frac{v_{\parallel 0}^2}{\Omega_-^2} (e^{-j\Omega_- L / v_{\parallel 0}} - 1) \right) \right\} + \left( L + \frac{j}{k_{\parallel}} \right) \left\{ \frac{jv_{\parallel 0} L}{\Omega_+} e^{-j\Omega_+ L / v_{\parallel 0}} + \frac{v_{\parallel 0}^2}{\Omega_+^2} (e^{j\Omega_+ L / v_{\parallel 0}} - 1) \right\} - \left\{ \frac{jv_{\parallel 0} L^2}{\Omega_+^2} e^{-j\Omega_+ L / v_{\parallel 0}} - j \frac{2v_{\parallel 0}}{\Omega_+} \left\{ \frac{jv_{\parallel 0} L}{\Omega_+} e^{-j\Omega_+ L / v_{\parallel 0}} + \frac{v_{\parallel 0}^2}{\Omega_+^2} (e^{j\Omega_+ L / v_{\parallel 0}} - 1) \right\} \right\} \right\} \\ I_s^{(2)} = & -j \frac{2k_{\parallel}}{v_{\parallel 0}} \left\{ \left( L - \frac{j}{k_{\parallel}} \right) \frac{jv_{\parallel 0}}{\Omega_-} (e^{-j\Omega_- L / v_{\parallel 0}} - 1) - \left\{ \frac{jv_{\parallel 0} L}{\Omega_-} e^{-j\Omega_- L / v_{\parallel 0}} + \frac{v_{\parallel 0}^2}{\Omega_-^2} (e^{-j\Omega_- L / v_{\parallel 0}} - 1) \right\} + \left( L + \frac{j}{k_{\parallel}} \right) \frac{jv_{\parallel 0}}{\Omega_+} (e^{-j\Omega_+ L / v_{\parallel 0}} - 1) - \left\{ \frac{jv_{\parallel 0} L}{\Omega_+} e^{-j\Omega_+ L / v_{\parallel 0}} + \frac{v_{\parallel 0}^2}{\Omega_+^2} (e^{-j\Omega_+ L / v_{\parallel 0}} - 1) \right\} \right\} \end{aligned}$$

here  $\Omega_{\pm} = \omega \pm k_{\parallel} v_{\parallel 0} - s \omega \omega_p$ , and  $\omega = jp$ .

Solution to (16) of desired order of approximation can be obtained by either numerical calculation method or iteration method. With iteration method, the zero order approximation solution of (16) can be chosen reasonably as the 'cold' resonance frequency of the cavity, i.e.,  $\Omega = (k_{\perp}^2 + k_{\parallel}^2)^{1/2} c$ .

If substitute (15) and its conjugate complex into (13'), we obtain that

$$\begin{aligned} \omega^4 - 2\langle B \rangle \omega^3 - [\langle B^2 \rangle - \langle A \rangle c^2 - \frac{\omega_p^2}{\gamma_0} (Q_{m1} - \beta_{10}^2 H_{m1})] \omega^2 \\ + [2\langle AB \rangle c^2 + \frac{\omega_p^2}{\gamma_0} Q_{m1} (1 \omega_{\omega})] \omega - \langle AB^2 \rangle c^2 - \\ - \frac{\omega_p^2}{\gamma_0} [Q_{m1} k_{\parallel}^2 v_{\parallel 0}^2 + \beta_{10}^2 H_{m1} k_{\parallel}^2 c^2] = 0 \end{aligned} \quad (17)$$

where  $\omega = jp$ ,

$$\begin{aligned} \langle A \rangle &= k_{\perp}^2 + k_{\parallel}^2, \\ \langle B \rangle &= 1 \omega_{\omega}, \\ \langle B^2 \rangle &= k_{\perp}^2 v_{\parallel 0}^2 + (1 \omega_{\omega})^2, \\ \langle AB \rangle &= (k_{\perp}^2 + k_{\parallel}^2) 1 \omega_{\omega}, \\ \langle AB^2 \rangle &= k_{\perp}^2 k_{\parallel}^2 c^2 + (k_{\perp}^2 + k_{\parallel}^2) (1 \omega_{\omega})^2 + k_{\parallel}^4 v_{\parallel 0}^2. \end{aligned}$$

Since Eq.(17) is an algebrical equation of the fourth order, its roots are much easier to find than that of Eq.(16), making (17) of more practical value in engineering design than (16).

Eqs.(16) and (17) are two forms of linear theory for gyromonotron. Owing to the theorem, not only is their derivation simple but also their formulae are compact and easy to solve in contrast to the complicated derivation procedure and the overelaborate formula of the general gyrotron theory<sup>2,3</sup>.

#### IV. Conclusions

A useful theorem in the field of relativistic electronics has been proposed and proved in this paper. According to the theorem, the oscillating characteristic of any HF oscillator, under the condition of small signal, is connected directly with the dispersion relation of the electron beam instability on which the oscillator is based. Besides, the theorem provides a simple way of analysing the linear oscillating characteristic of the oscillator, making the theorem valuable in engineering.

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