

# **ROCO218: Control Engineering**

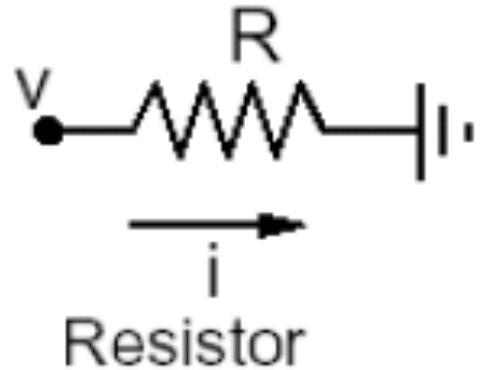
## **Dr Ian Howard**

### Lecture 2

Modelling electrical circuits  
with differential equations

# Electrical resistance

Resists the flow of current



$$v = iR$$

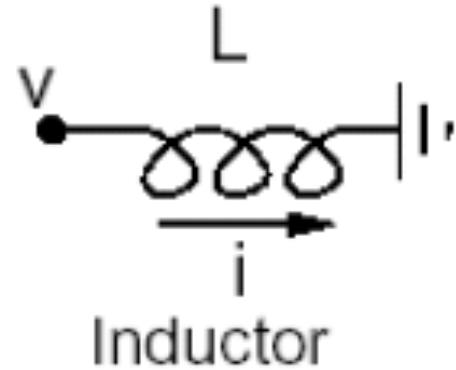
Where

i is the current in A

v is voltage in V

R is the resistance in  $\Omega$

# Electrical inductance



$$v = L \frac{di}{dt}$$

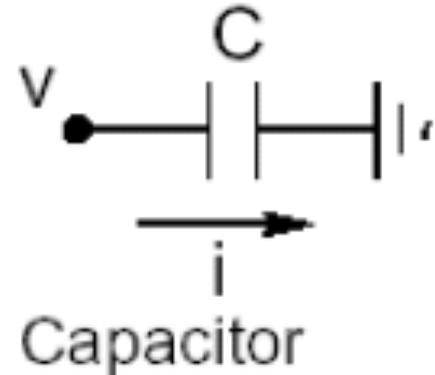
Where

i is the current in A

v is voltage in V

L is the inductance in H

# Electrical capacitance



$$i = C \frac{dv}{dt}$$

Where

i is the current in A

v is voltage in V

C is the capacitance in F

Therefore

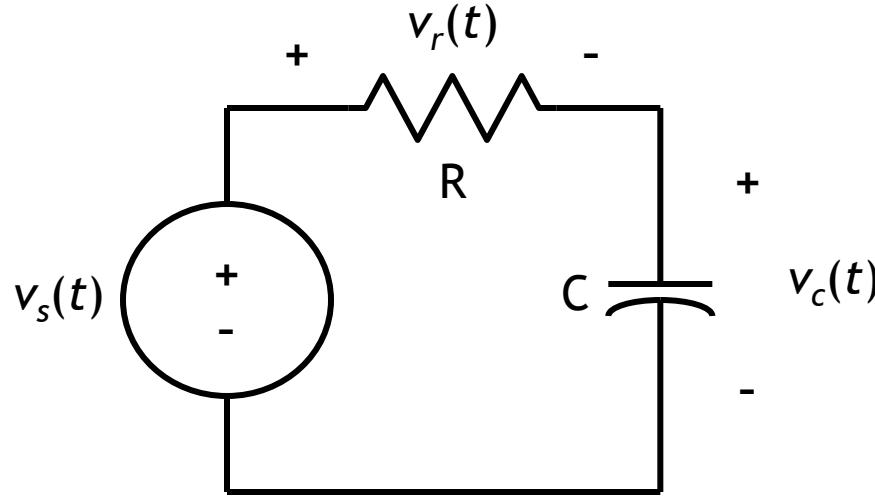
$$v = \frac{1}{C} \int i(t).dt$$

Also

$$Q = Cv$$

Where Q is charge

# A First-order RC Circuit



- One capacitor and one resistor in series
- The source and resistor may be equivalent to a circuit with many resistors and sources

# Voltages in an RC circuit

$$V_s = V_R + V_C$$

Adding up voltages

$$V_R = IR$$

Ohms law

$$I = \frac{dQ}{dt} = C \frac{dV_C}{dt}$$

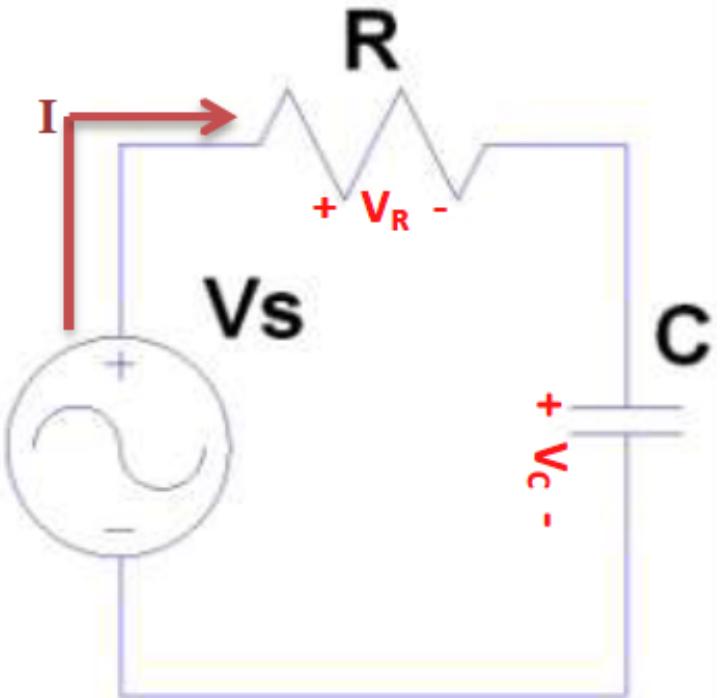
Current proportional to rate of change of charge

$$V_s = RC \frac{dV_C}{dt} + V_C$$

Substitute I from capacitor expression

$$\frac{dV_C}{dt} = \frac{1}{RC} (V_s - V_C)$$

Rearranged



This is a first order linear differential equation

# Current in an LR circuit

$$V_S = V_R + V_L$$

Adding up voltages

$$V_R = IR$$

Ohms law

$$V_L = L \frac{dI}{dt}$$

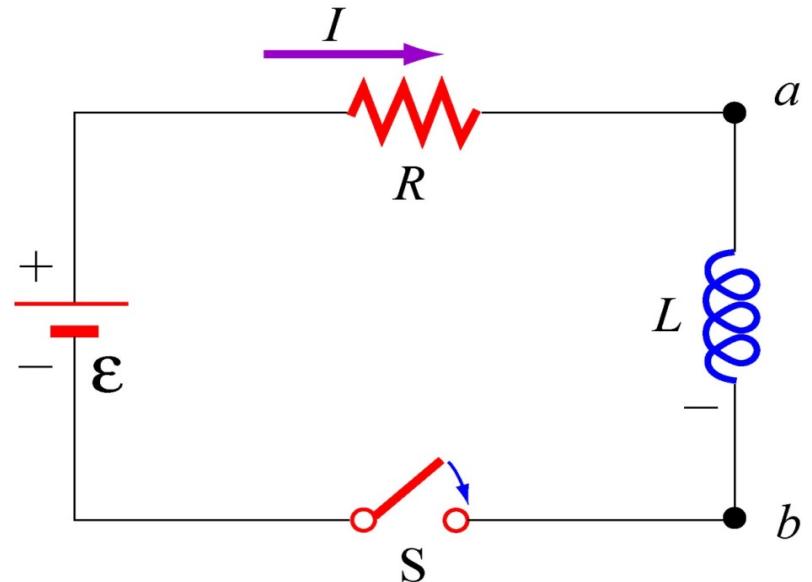
Voltage proportional to  
rate of change of current

$$V_S = IR + L \frac{dI}{dt}$$

Substituted values

$$\frac{dI}{dt} = \frac{R}{L} \left( \frac{V_S}{R} - I \right)$$

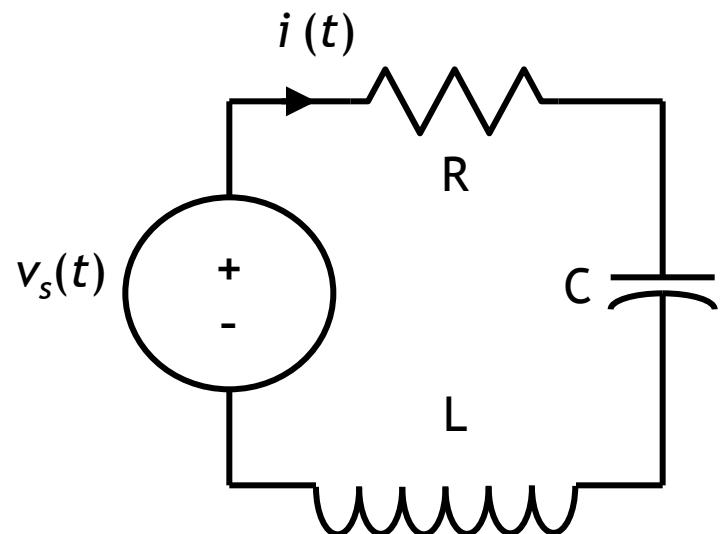
Rearranged



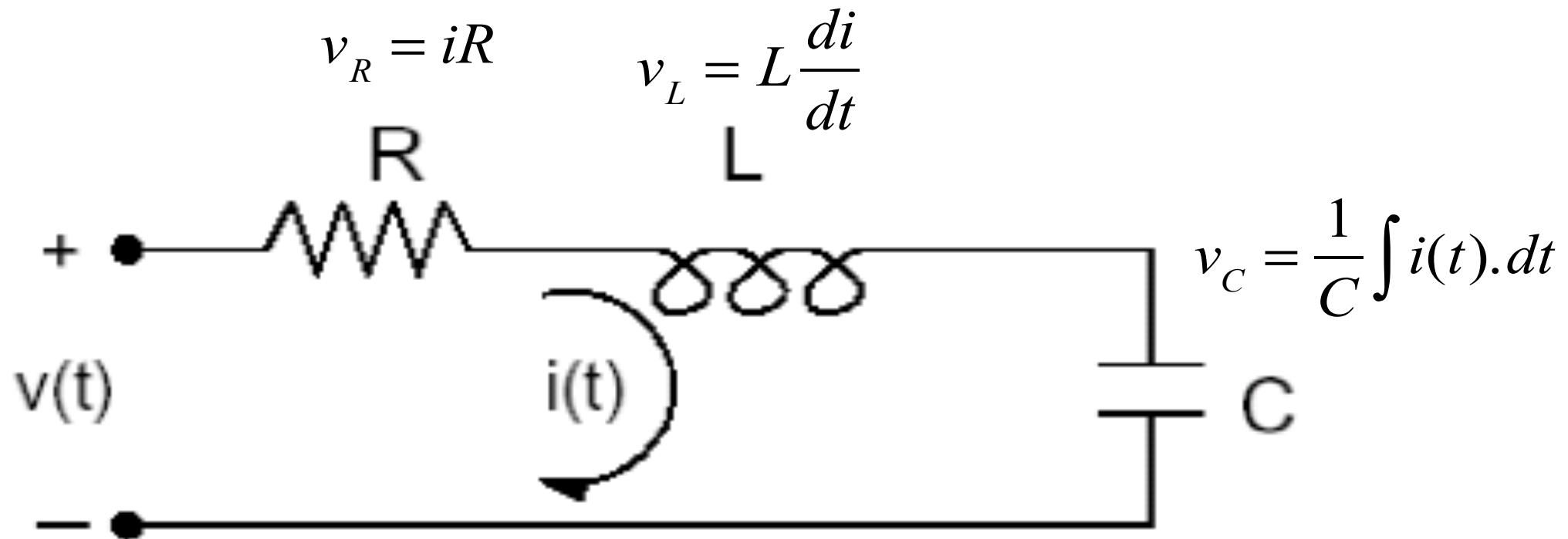
This is also a first order linear differential equation

# 2nd order LCR circuits

- Any circuit with a single capacitor, a single inductor, an arbitrary number of sources, and an arbitrary number of resistors is a circuit of order 2
- The source and resistor may be equivalent to a circuit with many resistors and sources
- Any voltage or current in such a circuit is the solution to a 2<sup>nd</sup> order differential equation



# Series RLC circuit

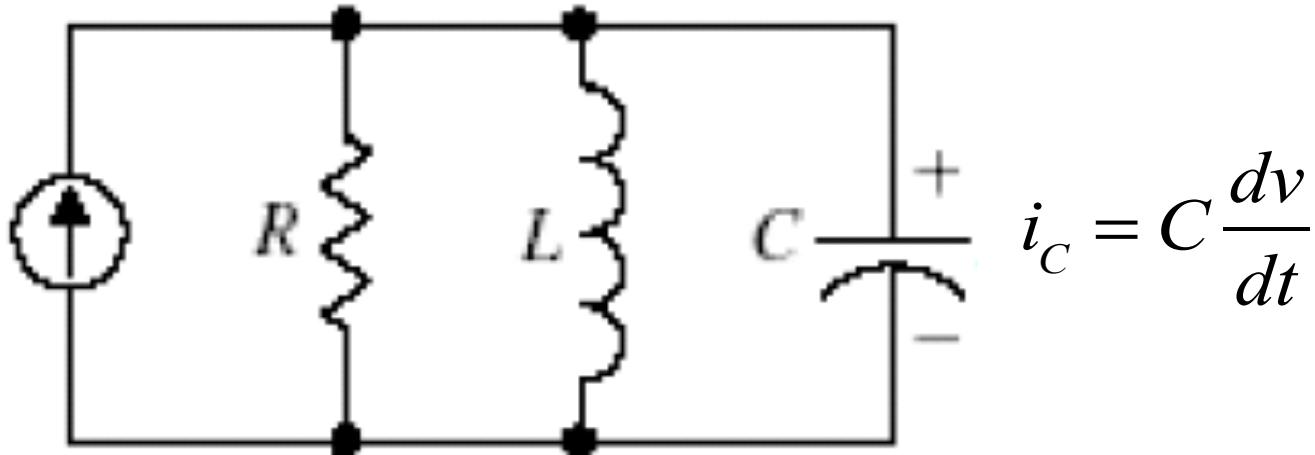


Adding up the contributions to overall applied voltage

$$v(t) = iR + L \frac{di}{dt} + \frac{1}{C} \int i(t).dt$$

# Parallel RLC circuit

$$i_R = \frac{v}{R} \quad i_L = \frac{1}{L} \int v dt$$



Adding up the contributions to overall applied current

$$i(t) = \frac{v}{R} + \frac{1}{L} \int v(t) dt + C \frac{dv}{dt}$$

# Summary: RLC Characteristics

Element	V/I Relation	Comments
Resistor	$v_R(t) = R i_R(t)$	$V = I R$ R has SI unit Ohms
Capacitor	$i_C(t) = C \frac{d v_C(t)}{dt}$	$Q = CV$ C has SI unit is Farads
Inductor	$v_L(t) = L \frac{d i_L(t)}{dt}$	L has SI unit Henrys

# **ROCO218: Control Engineering**

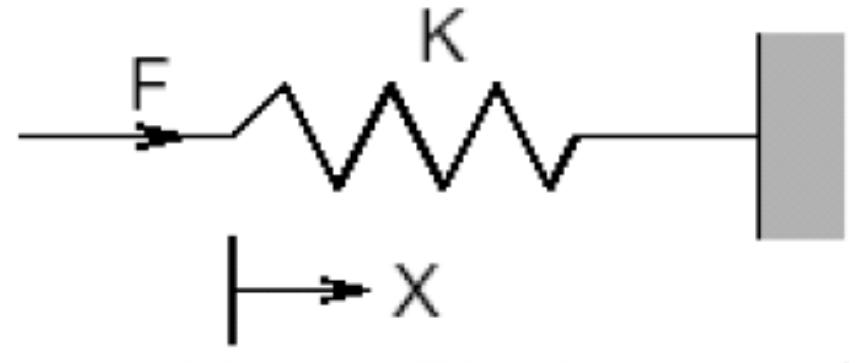
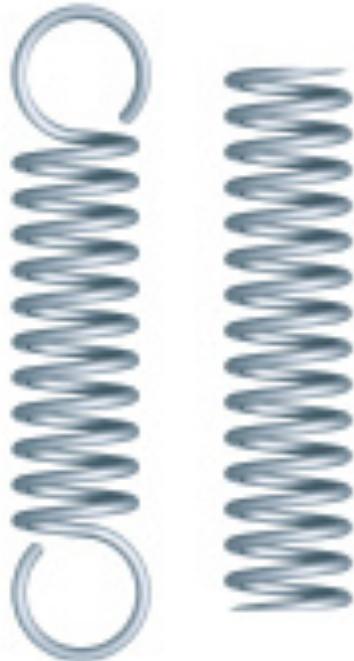
## **Dr Ian Howard**

### Lecture 2

Modelling mechanical systems  
using differential equations

# Compression & extension springs

Resists with opposing force proportional to extension of movement



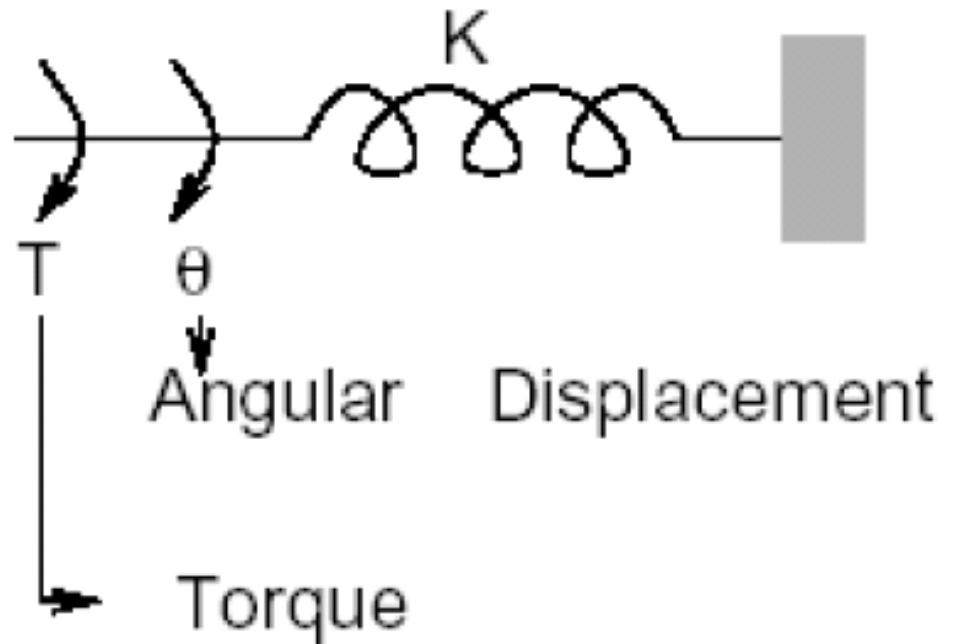
$$f = kx$$

where

$f$  is force in N

$k$  is the spring constant in N/m

# Torsional springs



$$T = k\theta$$

where

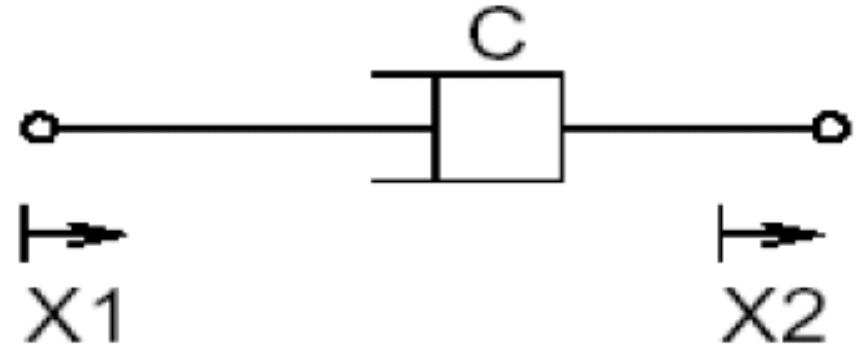
T is torque in Nm

k is the spring constant in Nm/rad

# Dampers and dashpots

Resists with opposing force proportional to velocity of movement

Like moving your hand in water – resistance proportional to how fast you move



Linear Displacement

$$f = C \frac{dx}{dt}$$

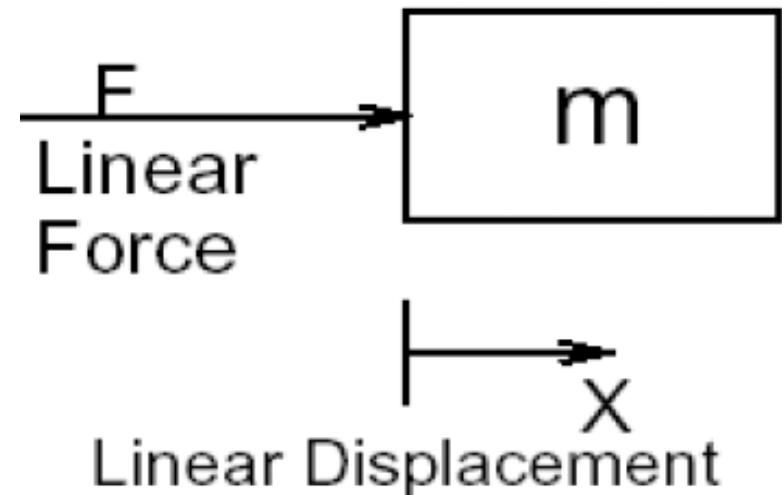
Where

f is force in N

C is viscosity in Ns/m

# Inertial Mass

Resists with opposing force proportional to acceleration of movement



Where

f is force in N

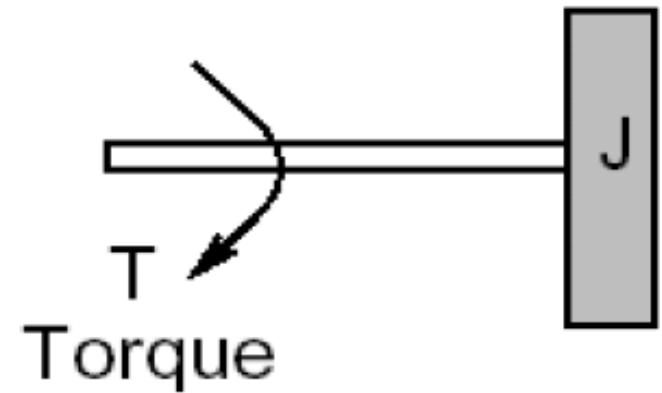
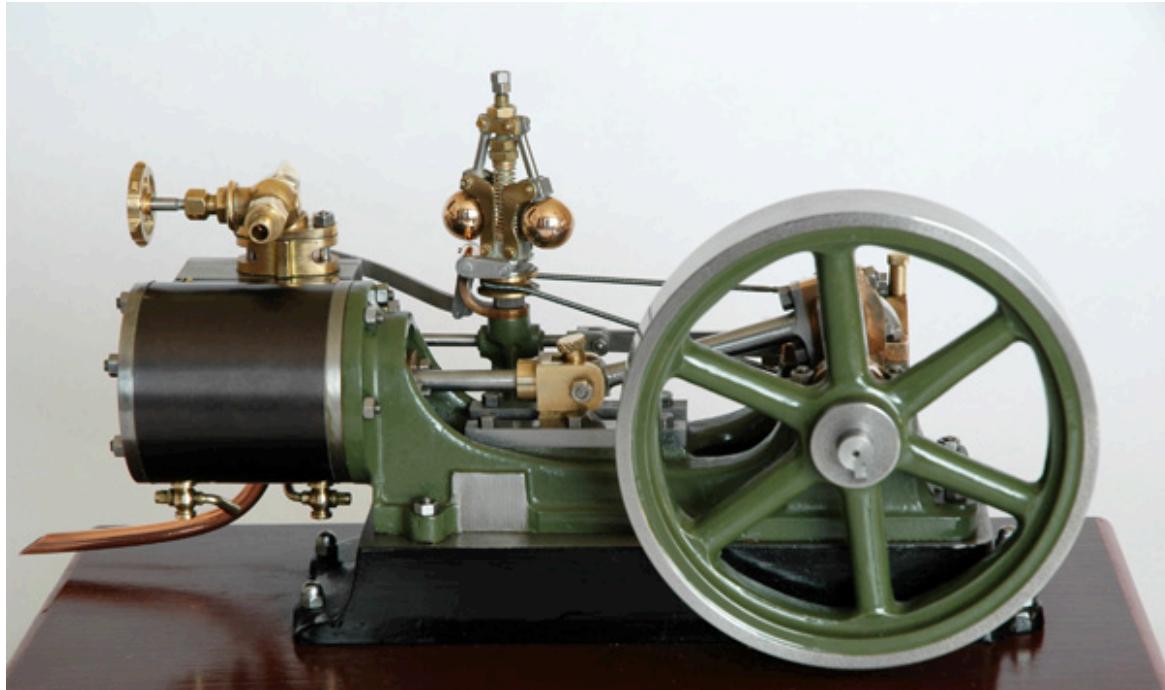
m is mass in Kg

a is linear acceleration in m/s<sup>2</sup>

$$f = m \frac{d^2 x}{dt^2} = ma$$

# Moment of Inertia

Resists with opposing torque proportional to angular acceleration



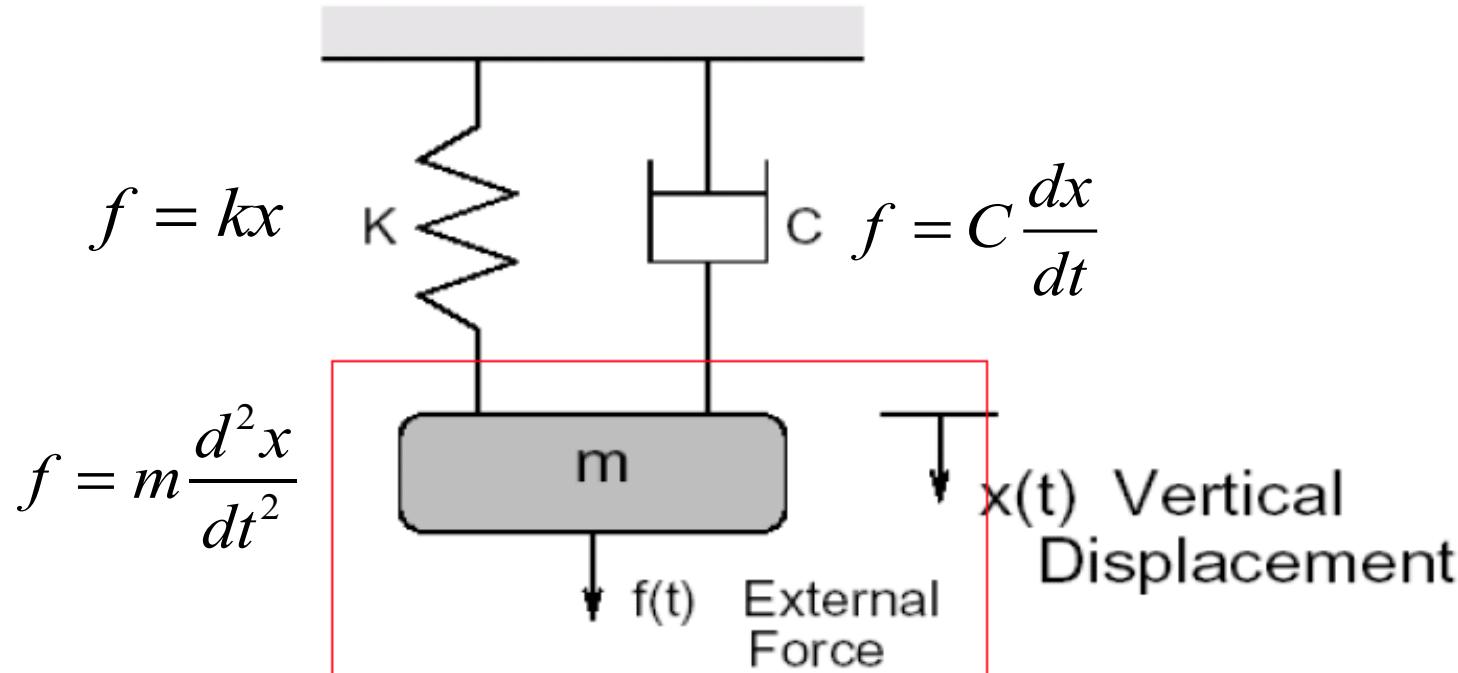
Where

T is torque in Nm

J is moment of inertial in Kgm<sup>2</sup>

$$T = J \frac{d^2\theta}{dt^2}$$

# Spring mass damper system



Adding up the contributions to overall force

$$f(t) = m \frac{d^2x}{dt^2} + C \frac{dx}{dt} + kx$$

# **ROCO218: Control Engineering**

## **Dr Ian Howard**

### Lecture 2

### Circular movement

# Constant speed circular movement

Consider a point mass travelling around central point in a circle at a constant speed with rotational period

Therefore in  $T$  seconds it rotates through an angle of  $2\pi$  radians

Thus angular velocity  $\omega$  is given by

$$\omega = \frac{2\pi}{T} \Rightarrow T = \frac{2\pi}{\omega}$$

The distance travelled by the mass in time  $T$  seconds is given by

$$d_{circ} = 2\pi r$$

The average speed at which it does so is therefore

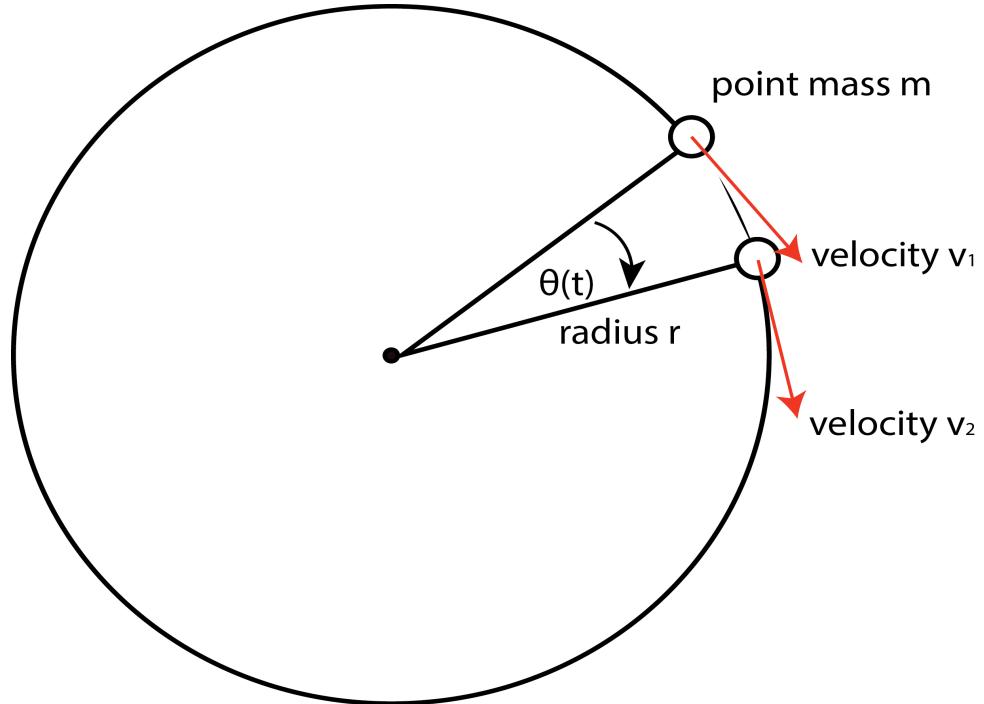
$$v = \frac{d_{circ}}{T} = \frac{2\pi r}{T}$$

Substituting in

$$T = \frac{2\pi}{\omega}$$

$$\Rightarrow v = \frac{2\pi r}{\left(\frac{2\pi}{\omega}\right)}$$

$$\Rightarrow v = \omega r$$



Note also

The rotational frequency can be related to rotational period  $T$

$$f = \frac{1}{T}$$

$$\Rightarrow \omega = 2\pi f$$

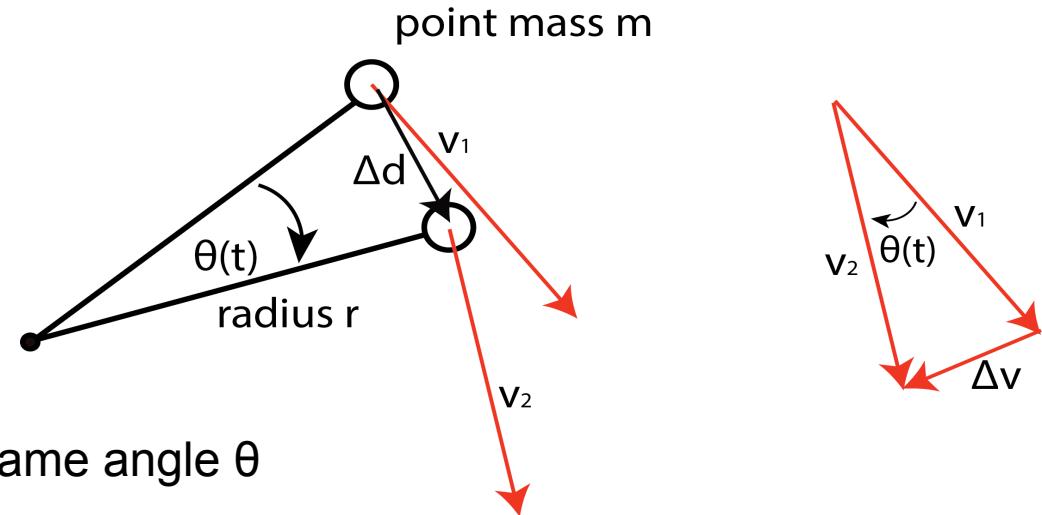
# Simple derivation of centrifugal acceleration

Consider the moving mass

As the mass rotates through an angle of  $\theta$  the tangential velocity of the mass changes from  $v_1$  to  $v_2$   
Mass moves a distance  $\Delta d$

Velocity magnitude stays constant but its direction changes so velocity changes by  $\Delta v$

By consideration of similar triangles with the same angle  $\theta$



$$\frac{\Delta d}{r} = \frac{\Delta v}{v}$$

if  $\Delta d$  is very very small is can be approximated by  $\Delta d \approx v\Delta t$

Substituting in this value

$$\Rightarrow \frac{v\Delta t}{r} = \frac{\Delta v}{v} \Rightarrow \frac{\Delta v}{\Delta t} = \frac{v^2}{r}$$

As  $\Delta t$  tends to 0, term  $\Delta v/\Delta t$  represents acceleration

$$\Rightarrow a = \frac{v^2}{r}$$

Since from before

$$v = \omega r \Rightarrow a = \omega^2 r$$

# Chain rule for single variable differentiation

- Consider the one variable equation corresponding to a function of a function

$$z = f(g(x))$$

- Writing

$$y = g(x)$$

- Chain rule gives

$$\frac{dz}{dx} = \frac{dz}{dy} \cdot \frac{dy}{dx}$$

- For example if

$$z = e^{x^2}$$

- In this case let

$$y = x^2 \quad \Rightarrow \frac{dy}{dx} = 2x$$

$$z = e^y \quad \Rightarrow \frac{dz}{dy} = e^y$$

$$\Rightarrow \frac{dz}{dx} = e^y \cdot 2x \quad = 2xe^{x^2}$$

# Calculus derivation of centrifugal acceleration

We can write down the projections of the mass position along the x and y axes

$$x(t) = r \cos(\theta(t))$$

$$y(t) = r \sin(\theta(t))$$

We can write mass position vector  $P(t)$  as sum of its base vector components

$$P(t) = r \cos(\theta(t)) \bar{x} + r \sin(\theta(t)) \bar{y}$$

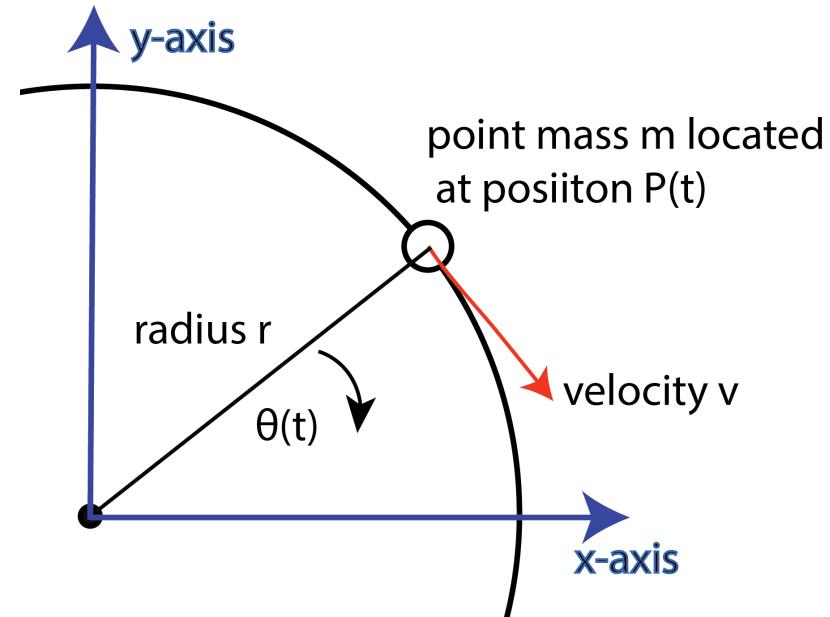
$$P(t) = r [\cos(\theta(t)) \bar{x} + \sin(\theta(t)) \bar{y}]$$

Differentiating both sides wrt time t gives

$$\frac{dP(t)}{dt} = V(t) = r \frac{d}{dt} [\cos(\theta(t)) \bar{x} + \sin(\theta(t)) \bar{y}]$$

$$\Rightarrow V(t) = r \left[ -\sin(\theta(t)) \frac{d\theta}{dt} \bar{x} + \cos(\theta(t)) \frac{d\theta}{dt} \bar{y} \right]$$

$$\Rightarrow V(t) = -r \frac{d\theta}{dt} [\sin(\theta(t)) \bar{x} - \cos(\theta(t)) \bar{y}]$$



# Calculus derivation of centrifugal acceleration

Now angular velocity is given by

$$\frac{d\theta}{dt} = \omega$$

Which is a constant if speed is constant. Substituting into

$$V(t) = -r \frac{d\theta}{dt} [\sin(\theta(t))\bar{x} - \cos(\theta(t))\bar{y}]$$

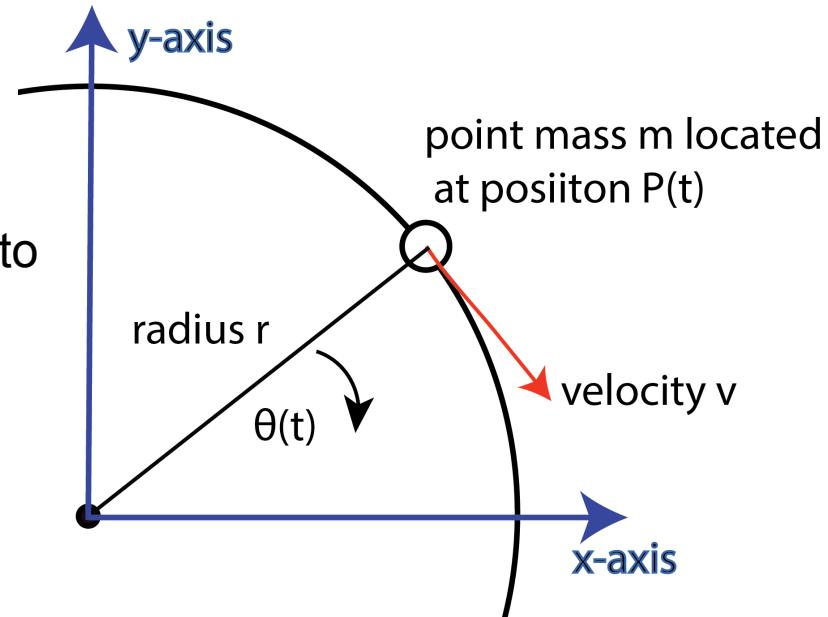
$$\Rightarrow V(t) = -r\omega [\sin(\theta(t))\bar{x} - \cos(\theta(t))\bar{y}]$$

Differentiating the expression again

$$\Rightarrow \frac{dV(t)}{dt} = -r\omega \frac{d}{dt} [\sin(\theta(t))\bar{x} - \cos(\theta(t))\bar{y}]$$

$$\Rightarrow \frac{dV(t)}{dt} = -r\omega \left[ \cos(\theta(t)) \frac{d\theta}{dt} \bar{x} + \sin(\theta(t)) \frac{d\theta}{dt} \bar{y} \right] = -r\omega^2 [\cos(\theta(t))\bar{x} + \sin(\theta(t))\bar{y}]$$

$$\Rightarrow \frac{dV(t)}{dt} = -\omega^2 P(t)$$



# Calculus derivation of centrifugal acceleration

Now acceleration  $A(t)$  is given by

$$A(t) = \frac{dV(t)}{dt}$$

$$\Rightarrow A(t) = -\omega^2 P(t)$$

Taking the modulus of both sides

$$\Rightarrow |A(t)| = |\omega^2 P(t)|$$

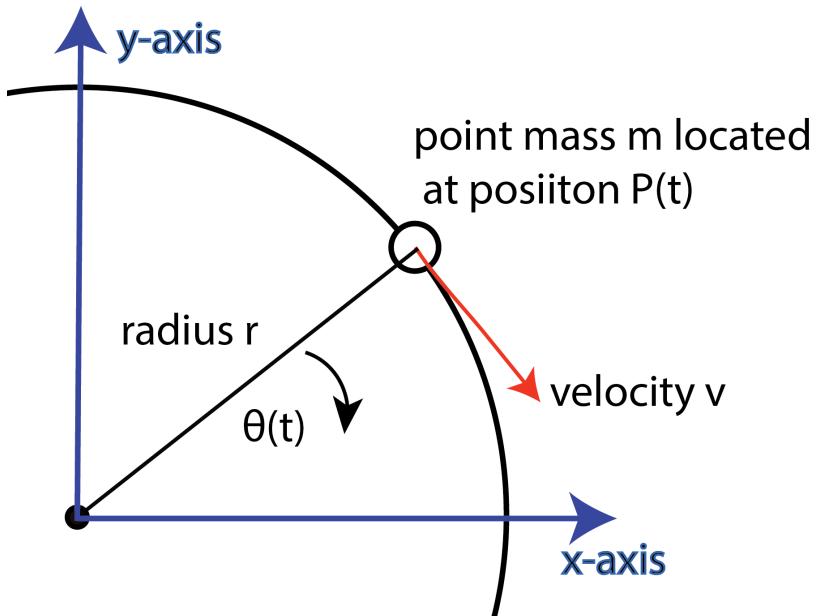
Note that  $|P(t)|$  is just the radius of the circle

$$\Rightarrow a = \omega^2 r$$

Or as from before

$$v = \omega r$$

$$\Rightarrow a = \frac{v^2}{r}$$



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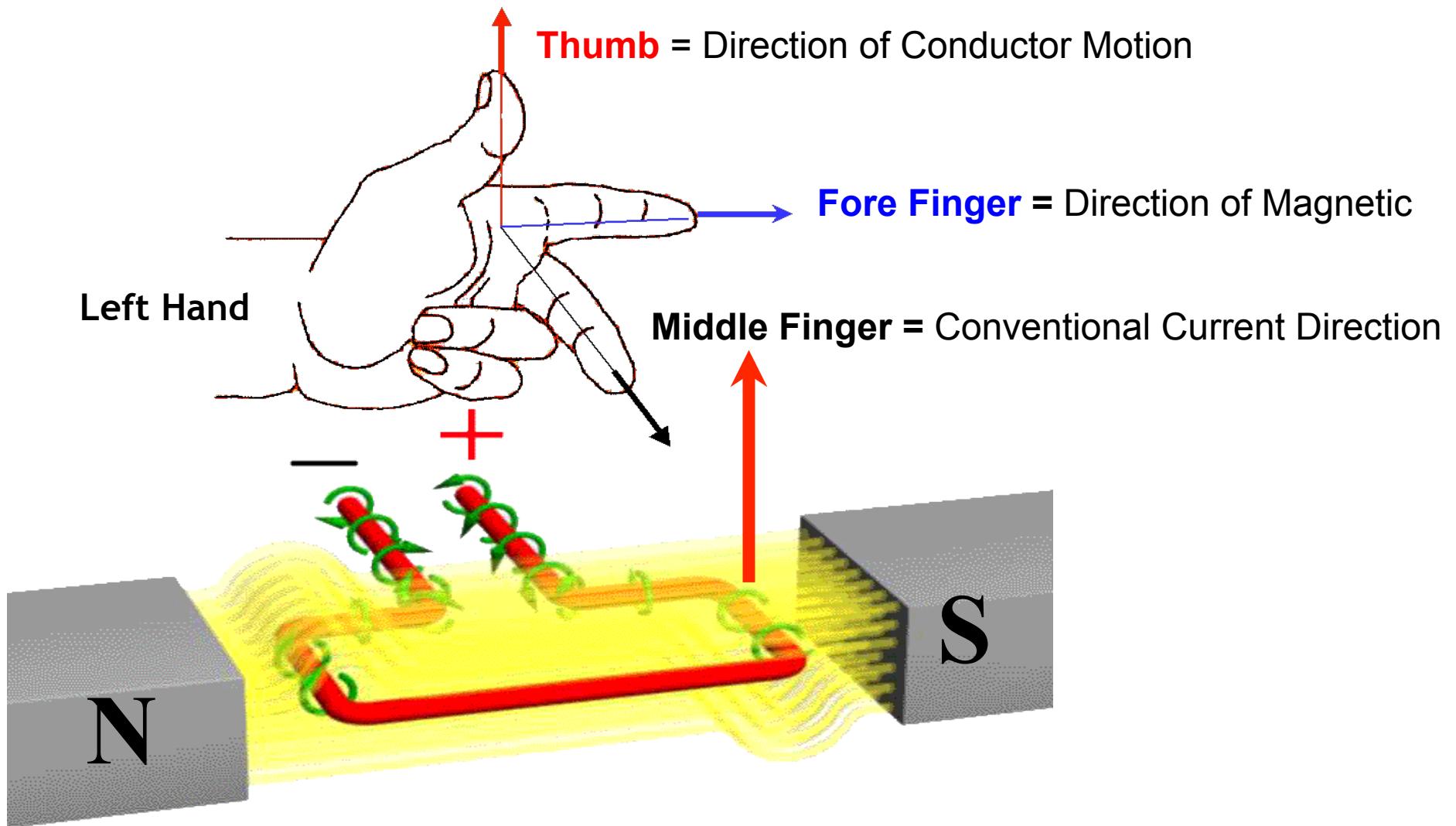
## **Dr Ian Howard**

### Lecture 2

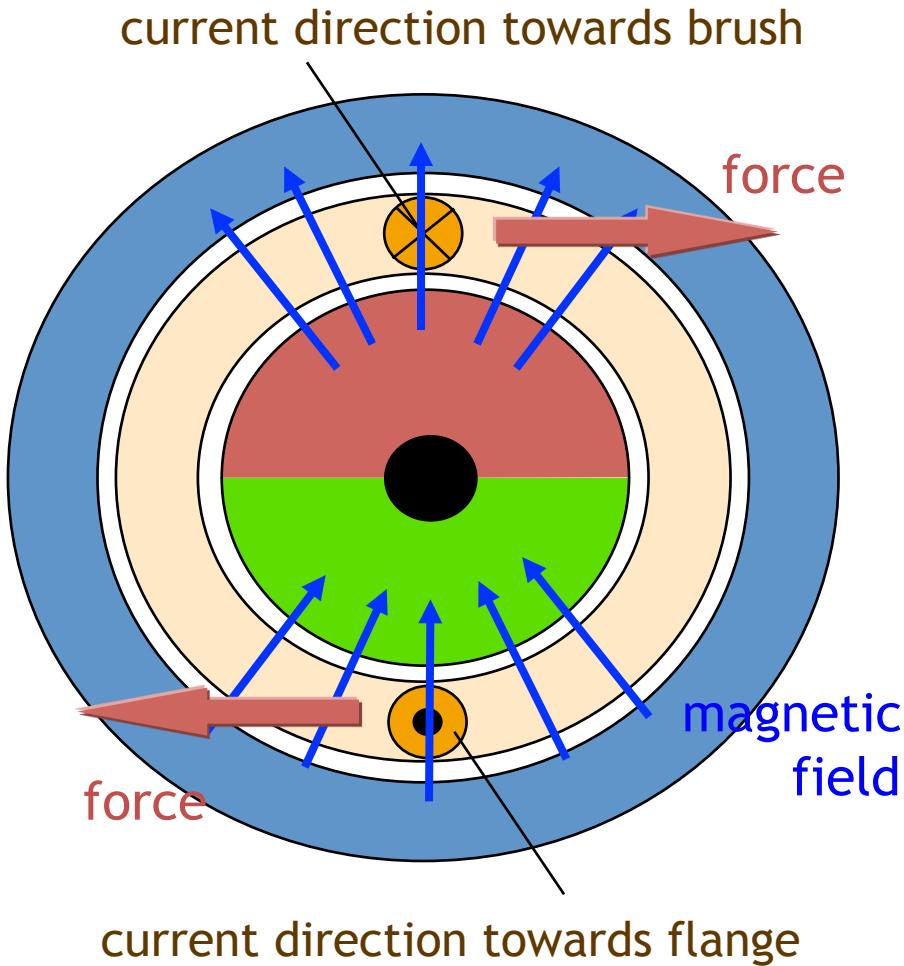
### Modelling a DC motor

# Fleming's Left Hand (Motor) Rule

The Left Hand Rule determines the movement direction of conductor carrying current



# Motor Torque constant $K_t$



**forces:**

force on current leading conductor in a magnetic field

**torque:**

sum of all forces at the distance to the rotating axis

**influencing parameters:**

geometry

field density

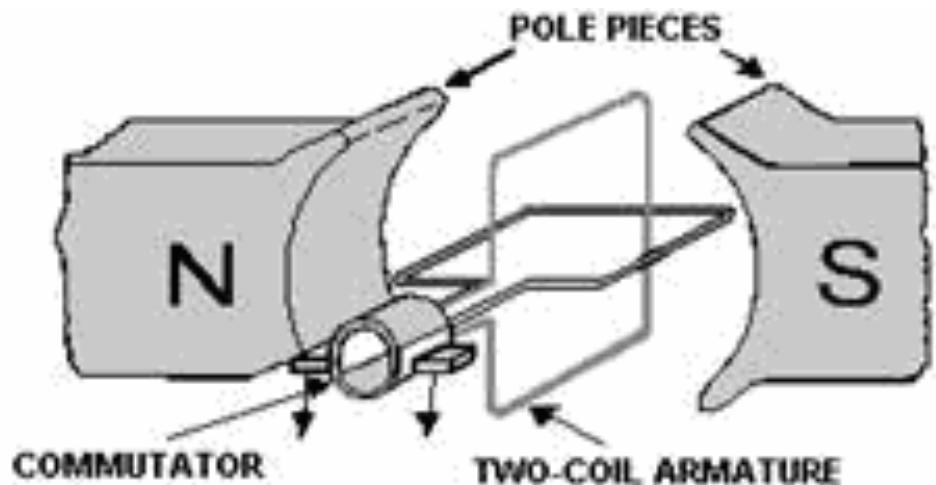
winding number

} Depend on the design

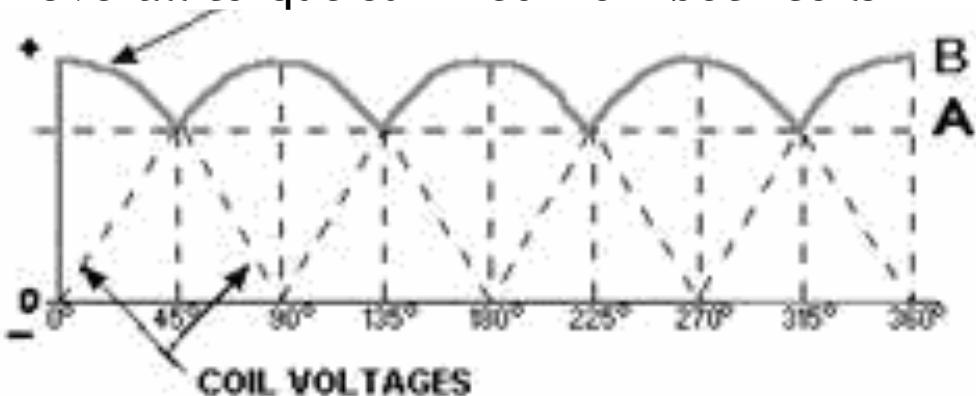
Motor torque  $T_m$  is given by

$$T_m = K_t i(t)$$

# Multiple coils reduces torque ripple



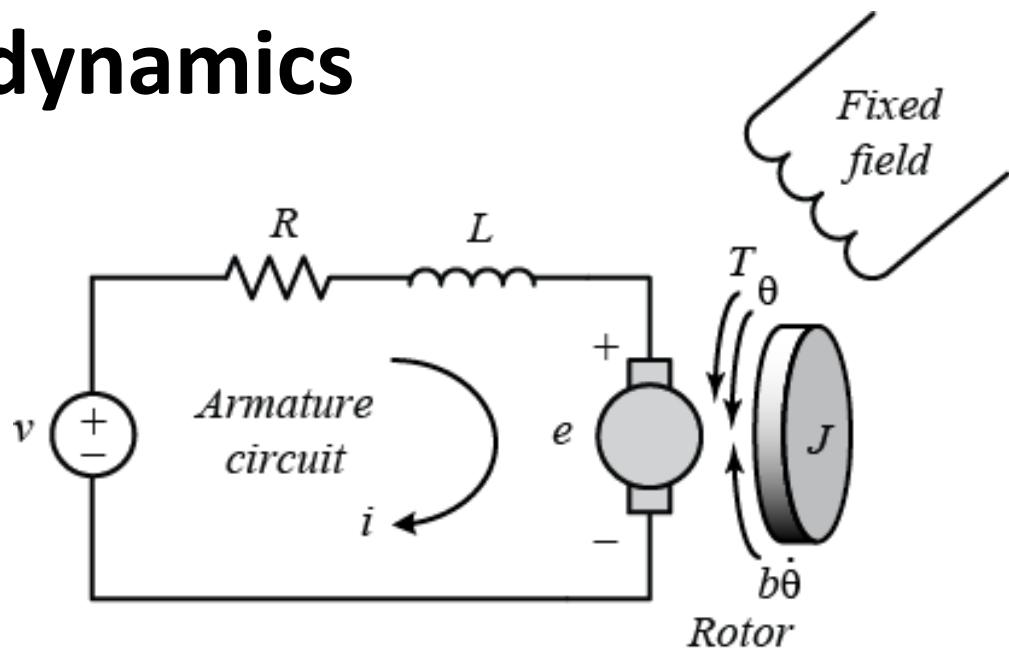
Overall torque summed from both coils



Typical small armature  
with multiple multi-turn coils

# DC motor dynamics

- $V$  applied voltage,  $I$  is the current drawn
- $R$  resistance of coil and brush contact
- $L$  is the coil inductance
- EMF is back emf of motor
- Lets relate output speed of motor to the applied voltage!



- $R$  = Armature resistance (in ohms)
- $L$  = Armature inductance (in Henrys)
- $J$  = Moment of inertia for the motor rotor (kg.m<sup>2</sup>)
- $b$  = Motor viscous friction constant (in N.m.s)
- $K_t$  = Motor torque constant (in N.m/A)
- $K_e$  = Electromotive force constant (in V/rad/sec)

# DC motor dynamics

Motor torque  $T_m$  is given by the current multiplied by the torque constant of the motor

$$T_m = K_t i(t)$$

Mechanical resisting torque  $T_r$  is given by inertia and viscous friction

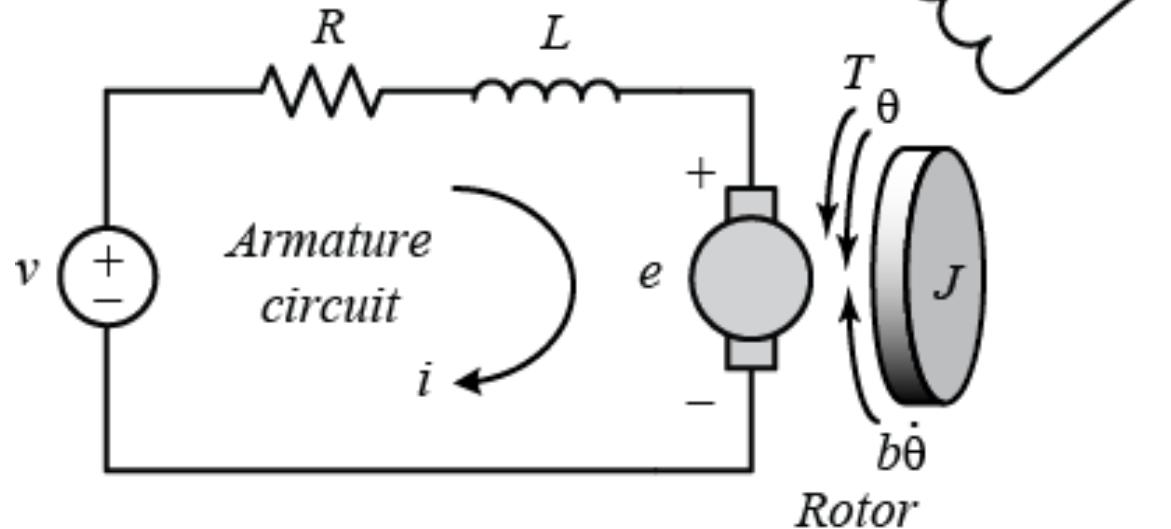
$$T_r = b \frac{d\theta}{dt} + J \frac{d^2\theta}{dt^2}$$

Therefore equating the two terms gives torque equation

$$K_t i(t) = b \frac{d\theta}{dt} + J \frac{d^2\theta}{dt^2}$$

Rearranging torque equation

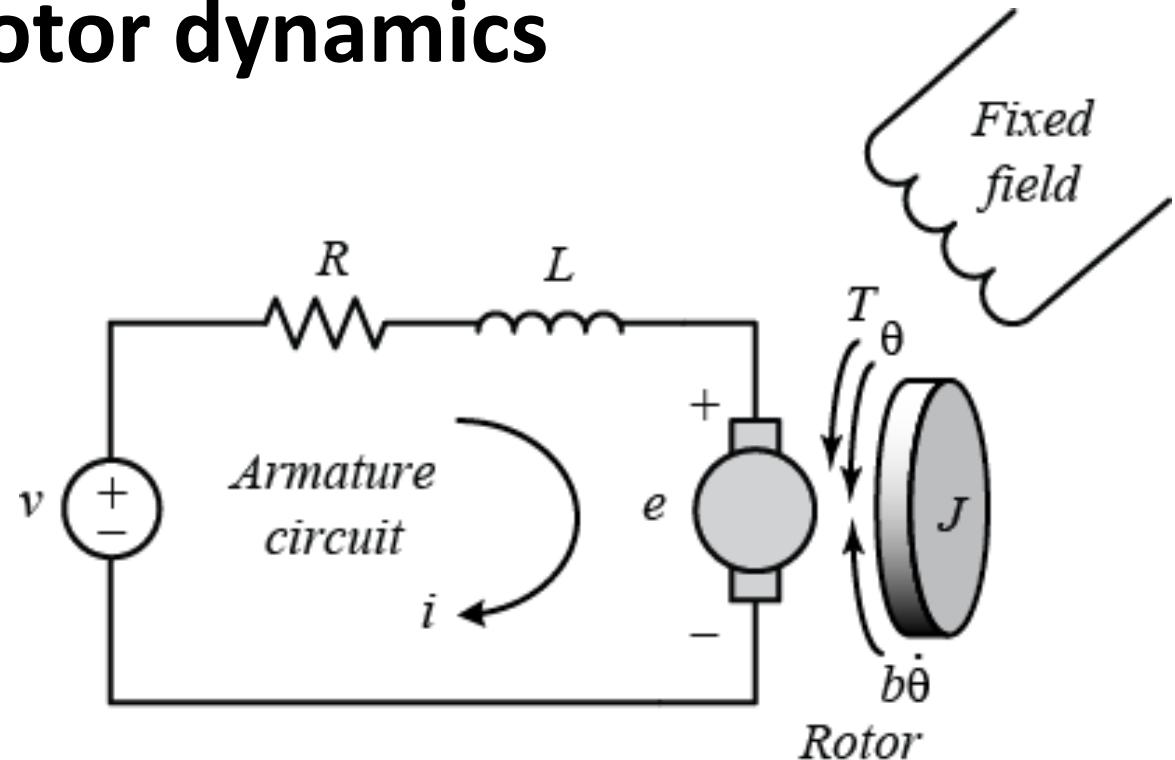
$$\Rightarrow J \frac{d^2\theta}{dt^2} = -b \frac{d\theta}{dt} + K_t i(t) \quad \Rightarrow \frac{d^2\theta}{dt^2} = -\frac{b}{J} \frac{d\theta}{dt} + \frac{K_t}{J} i(t)$$



# DC motor dynamics

Summing voltages around the circuit gives voltage equation

$$v(t) = i(t)R + L \frac{di}{dt} + K_e \frac{d\theta}{dt}$$



$$\Rightarrow L \frac{di}{dt} = -K_e \frac{d\theta}{dt} - i(t)R + v(t)$$

$$\Rightarrow \frac{di}{dt} = -\frac{K_e}{L} \frac{d\theta}{dt} - \frac{R}{L} i(t) + \frac{1}{L} v(t)$$

# State space DC motor dynamics

Now writing

$$\frac{d^2\theta}{dt^2} = \frac{d}{dt} \left( \frac{d\theta}{dt} \right) \quad \text{from} \quad \frac{d^2\theta}{dt^2} = -\frac{b}{J} \frac{d\theta}{dt} + \frac{K_t}{J} i(t)$$

$$\Rightarrow \frac{d}{dt} \left( \frac{d\theta}{dt} \right) = -\frac{b}{J} \frac{d\theta}{dt} + \frac{K_t}{J} i(t)$$

Now writing

$$\frac{di}{dt} = \frac{d}{dt}(i(t)) \quad \text{from} \quad \frac{di}{dt} = -\frac{K_e}{L} \frac{d\theta}{dt} - \frac{R}{L} i(t) + \frac{1}{L} v(t)$$

$$\Rightarrow \frac{d}{dt}(i(t)) = -\frac{K_e}{L} \frac{d\theta}{dt} - \frac{R}{L} i(t) + \frac{1}{L} v(t)$$

# State space DC motor dynamics

So we have the two equations

$$\frac{d}{dt} \left( \frac{d\theta}{dt} \right) = -\frac{b}{J} \frac{d\theta}{dt} + \frac{K_t}{J} i(t)$$

$$\frac{d}{dt}(i(t)) = -\frac{K_e}{L} \frac{d\theta}{dt} - \frac{R}{L} i(t) + \frac{1}{L} v(t)$$

We can write these two equations in matrix form

$$\frac{d}{dt} \begin{bmatrix} \frac{d\theta}{dt} \\ i(t) \end{bmatrix} = \begin{bmatrix} -\frac{b}{J} & \frac{K_t}{J} \\ -\frac{K_e}{L} & -\frac{R}{L} \end{bmatrix} \begin{bmatrix} \frac{d\theta}{dt} \\ i(t) \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{L} \end{bmatrix} v(t)$$

If we want to observe the speed of the motor as the output  $y$

$$y = \begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} \frac{d\theta}{dt} \\ i(t) \end{bmatrix}$$

# State space DC motor dynamics

As we will later see the two equations describe the motor in state space form

$$\frac{d}{dt} \begin{bmatrix} \frac{d\theta}{dt} \\ i(t) \end{bmatrix} = \begin{bmatrix} -\frac{b}{J} & \frac{K_t}{J} \\ -\frac{K_e}{L} & -\frac{R}{L} \end{bmatrix} \begin{bmatrix} \frac{d\theta}{dt} \\ i(t) \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{L} \end{bmatrix} v(t) \quad y = \begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} \frac{d\theta}{dt} \\ i(t) \end{bmatrix}$$

By observation of standard state space form

$$\dot{X} = AX + BU$$

$$Y = CX + DU$$

We see that

$$A = \begin{bmatrix} -\frac{b}{J} & \frac{K_t}{J} \\ -\frac{K_e}{L} & -\frac{R}{L} \end{bmatrix} \quad B = \begin{bmatrix} 0 \\ \frac{1}{L} \end{bmatrix} \quad C = \begin{bmatrix} 1 & 0 \end{bmatrix} \quad D = 0$$

# **ROCO218: Control Engineering**

## **Dr Ian Howard**

### Lecture 2

Numerical solutions of 1<sup>st</sup> order differential equation

# Differential equations

Ordinary differential equation (ODE)

- contains one or more functions of one independent variable and its derivatives
- For example

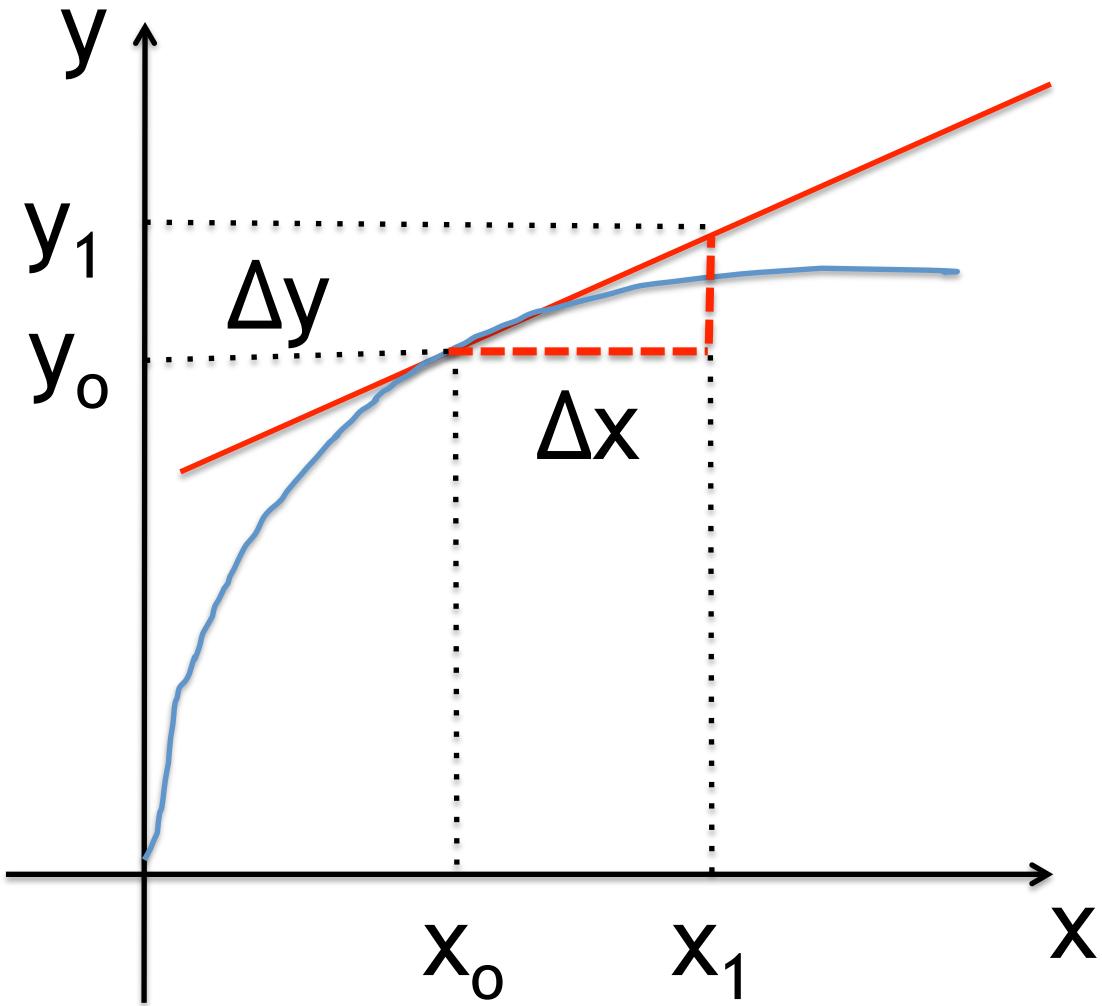
$$\frac{\partial u}{\partial x}(x, y) = 0.$$

Partial differential equation

- may compare more than one independent variable
- contains unknown multivariable functions and their partial derivatives.
- For example

$$\frac{\partial^2 u}{\partial x^2} + \frac{\partial^2 u}{\partial y^2} = 0,$$

# Local linear approximation of gradient



Consider a function  $y=f(x)$

When  $x=x_0$  then  $y=y_0$

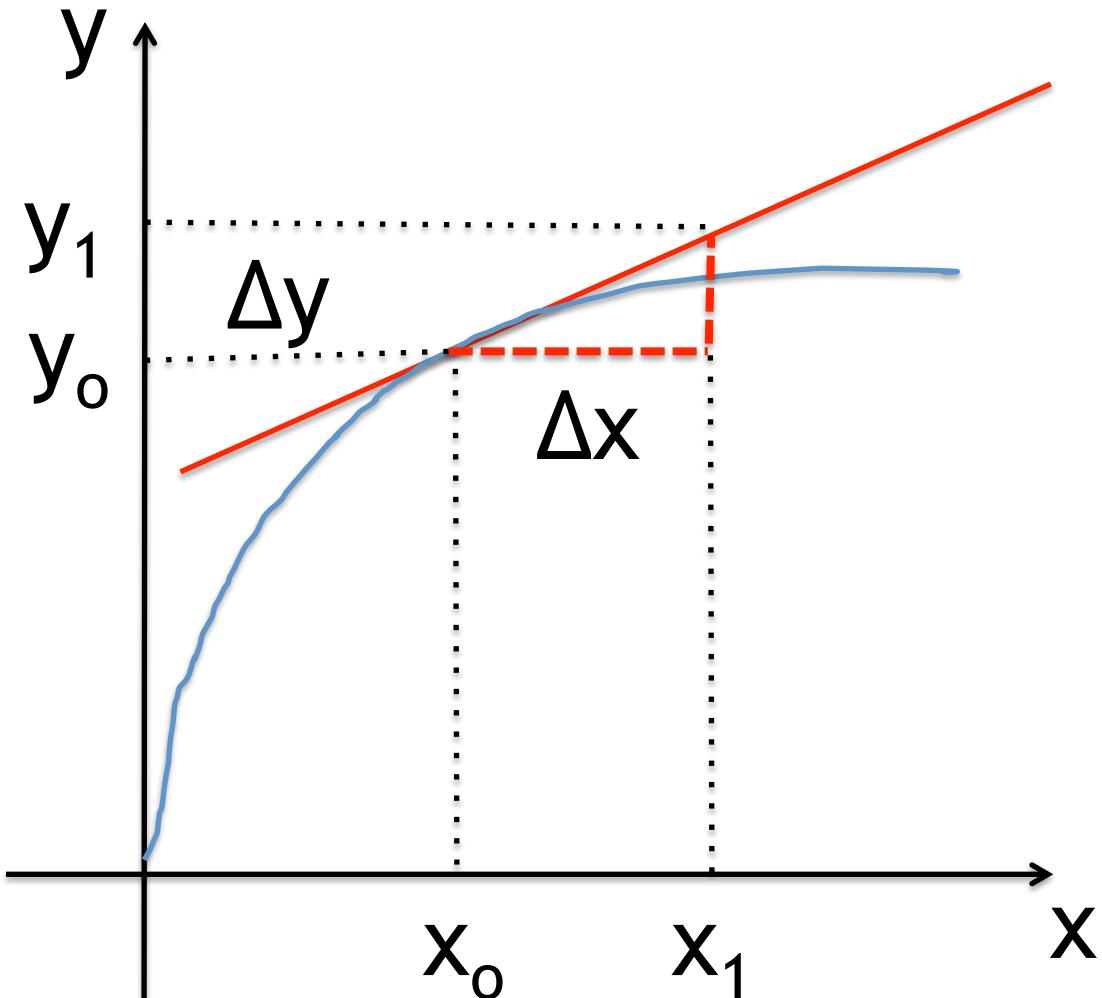
As we increase the value of  $x$  by  $\Delta x$   
we reach a point where  $x_1=x_0+\Delta x$   
Similarly this increases  $y$  by  $\Delta y$   
reaching the value  $y_1=y_0+\Delta y$

Thus

$$(x_1, y_1) = (x_0 + \Delta x, y_0 + \Delta y)$$

The gradient of the curve at  $(x_0, y_0)$   
is the tangent to the curve at this  
point

# Local linear approximation of gradient



In general the differential of a function is given by

$$f'(x) = \lim_{dx \rightarrow 0} \frac{f(x + dx) - f(x)}{dx}$$

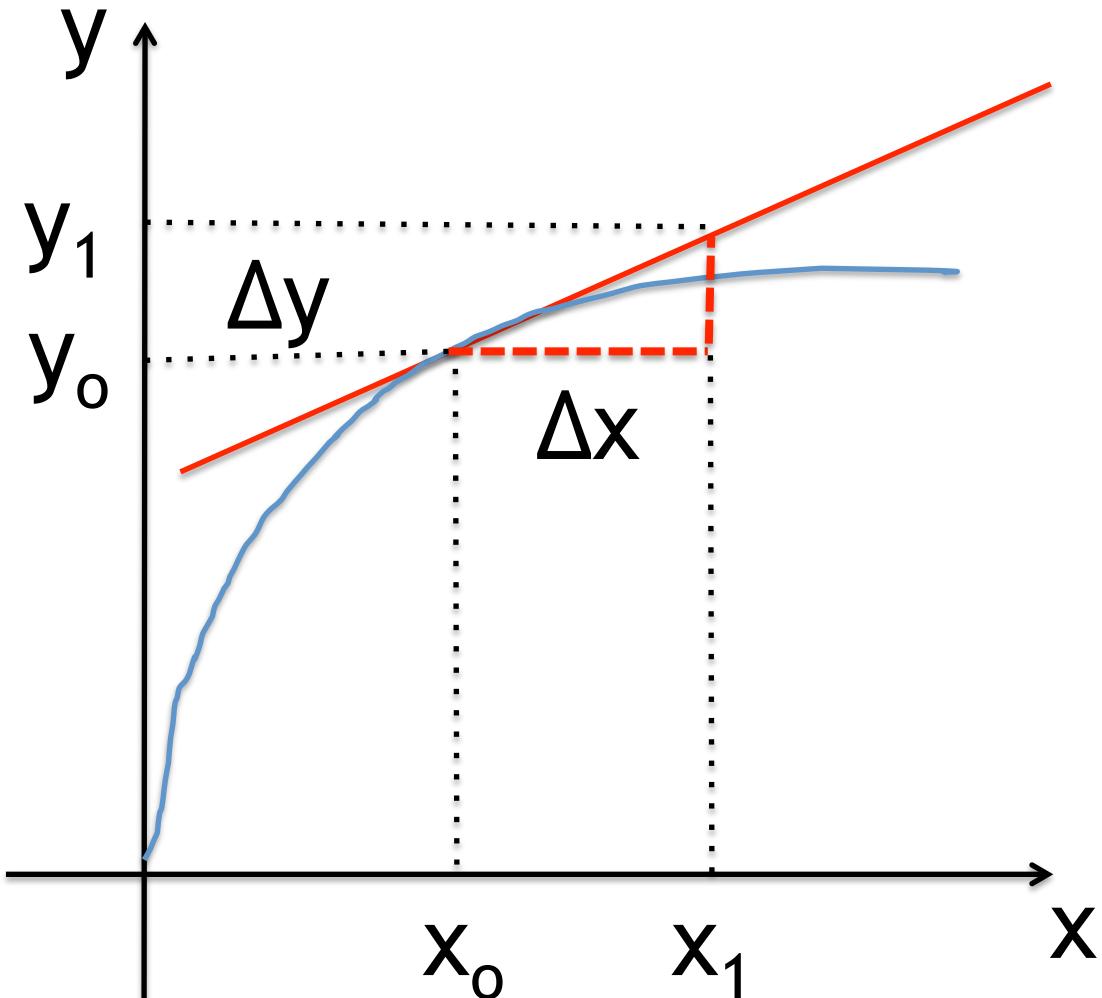
Which is only strictly true in the limit where  $dx$  tends to zero  
In our case

$$\frac{dy}{dx} \approx \frac{\Delta y}{\Delta x}$$

So the gradient at  $(x_0, y_0)$  can be approximated by

$$\left. \frac{dy}{dx} \right|_{(x_0, y_0)} \approx \frac{\Delta y}{\Delta x}$$

# Local linear approximation of gradient



We can use the relationship to iteratively estimate  $y_1$  from  $y_0$  since

$$\frac{dy}{dx} \Big|_{(x_0, y_0)} \approx \frac{\Delta y}{\Delta x} \approx \frac{(y_1 - y_0)}{\Delta x}$$

Writing  $\Delta x$  as step size  $h$

$$\Rightarrow y_1 - y_0 = h \cdot \frac{dy}{dx} \Big|_{(x_0, y_0)}$$

$$\Rightarrow y_1 = y_0 + h \cdot \frac{dy}{dx} \Big|_{(x_0, y_0)}$$

We can use this to give next estimate of  $y_1$  if we know  $x_0$  and the gradient at  $(x_0, y_0)$

# Euler's method of integration

- This is Euler's method to solve a 1<sup>st</sup> order differential equation!
- It uses a local linear approximation to the gradient
- We can iteratively estimate the value of the function at the next step
- Given the relationship

$$y' = f(t, y), \quad y(t_0) = y_0$$

$$y_{n+1} = y_n + h \cdot f(t_n, y_n)$$

Where  
t is time

y is the output value of the function  
At initial time  $t_0$  the output is  $y_0$   
At future time  $t_{(n+1)}$  the output is  $y_{(n+1)}$   
h is the step size

- Notice the simulation error will rise as the step size increases!

# Remember: Voltages in an RC circuit

$$V_s = V_R + V_C$$

Adding up voltages

$$V_R = IR$$

Ohms law

$$I = \frac{dQ}{dt} = C \frac{dV_C}{dt}$$

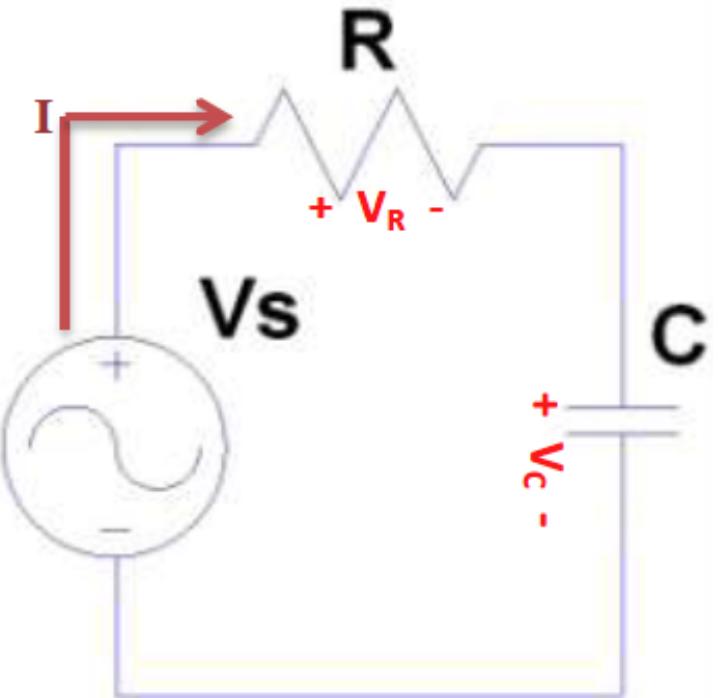
Current proportional to  
rate of change of charge

$$V_s = RC \frac{dV_C}{dt} + V_C$$

Substitute I from  
capacitor expression

$$\frac{dV_C}{dt} = \frac{1}{RC} (V_s - V_C)$$

Rearranged



This is a first order linear differential equation

# Simulation of voltages in an RC circuit

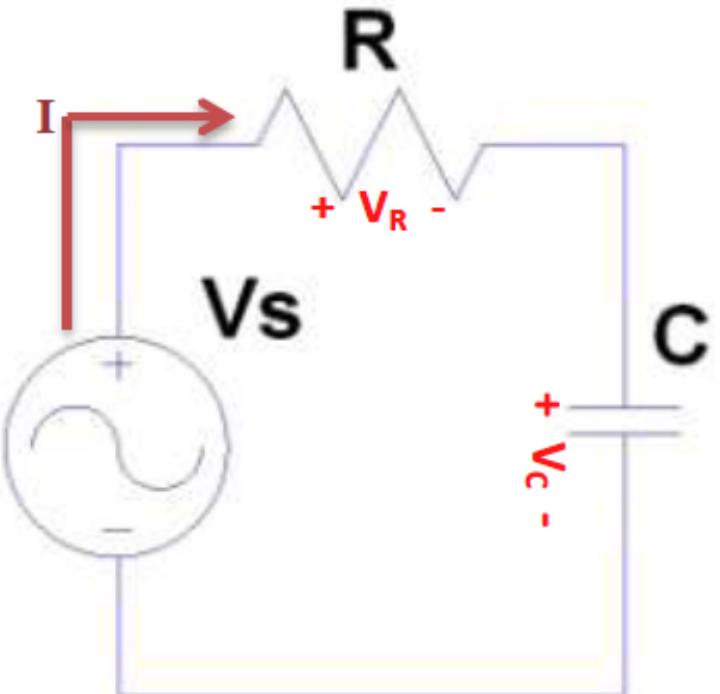
Here we wish to compute the evolution of  $V_c$  with time. We know that

$$\frac{dV_C}{dt} = \frac{1}{RC} (V_s - V_C)$$

From the iterative relationship we derived previously

$$y_1 = y_0 + h \cdot \frac{dy}{dx} \Big|_{(x_0, y_0)}$$

$$\Rightarrow V_C(t+1) = V_C(t) + h \frac{1}{RC} (V_s - V_C(t))$$



- One of your lab assignments is to implement this simulation in Matlab!
- Matlab also has more sophisticated numerical methods to solve differential equations- e.g. ode45

# Implementing Euler integration in Matlab

We can write simple script using the update equation to estimate capacitor voltage is given by the equation

$$\Rightarrow V_C(t+1) = V_C(t) + h \frac{1}{RC} (V_s - V_C(t))$$

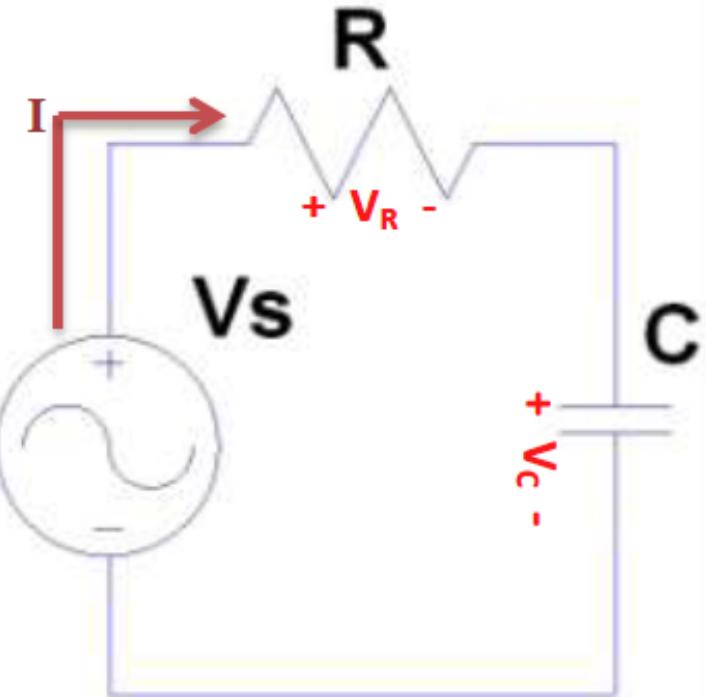
```
function Vc = GetCapVoltage(R, C, Vs, h)
% calling function to get Vc
% given resistance R, capacitance C, driving voltage Vs
% and time step h

% got input sample count
len = length(Vs)-1;

% set initial votage to zero
Vc(1) = 0;

% loop over all samples -1
for n = 1:len

    % Euler update equation
    Vc(n+1) = Vc(n) + h * (1/(R*C)) * (Vs(n) - Vc(n));
end
end
```



# Implementing Euler integration in Matlab

```
% reset the workspace and display
close all
clear all
clc

% parameters
timeLen = 0.4; % time duration of seconds
samplingRate = 1000; %sampling frequency in Hz
freq = 5; % signal frequency in Hz

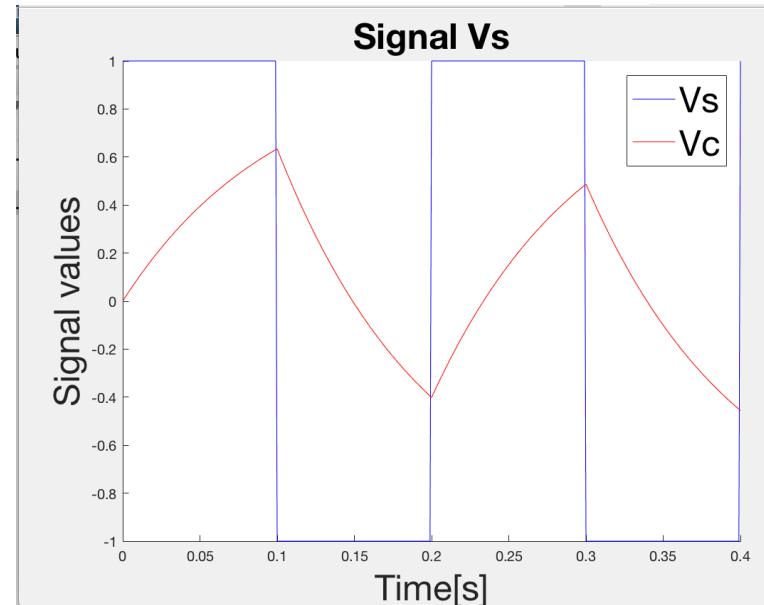
% resistance 1000 ohms
R = 1000;
% capacitor value 100 microFarads
C = 100e-6;
% step in secoinds
h = 1/samplingRate;
% time sampling points
T = 0:h:timeLen;

% number of databaseamples
sampleCnt = length(T);

% build square wave driving voltage
Vs = square(2 * pi * freq * T);

% calling function to get Vc
Vc = GetCapVoltage(R, C, Vs, h);
```

```
% plot the figure
figure
% want all plots to remain on figure
hold on
% write a title
h = title('Signal Vs');
set(h, 'FontSize', 25)
% plot driving voltage Vs
plot(T, Vs, 'b');
% plot capacitor voltage Vx
plot(T, Vc, 'r');
% label plots
h=legend('Vs', 'Vc');
set(h, 'FontSize', 25)
% label the axes
h=xlabel('Time[s]')
set(h, 'FontSize', 25)
h = ylabel('Signal values');
set(h, 'FontSize', 25)
```



# Using ode45 to perform integration

[help ode45](#)

- Solve non-stiff differential equations, medium order method.
- $[TOUT,YOUT] = \text{ode45}(ODEFUN,TSPAN,Y0)$
- with  $TSPAN = [T0 TFINAL]$  integrates the system of differential equations  $y' = f(t,y)$  from time  $T0$  to  $TFINAL$  with initial conditions  $Y0$ .
- $ODEFUN$  is a function handle.
- For a scalar  $T$  and a vector  $Y$ ,  $ODEFUN(T,Y)$  must return a column vector corresponding to  $f(t,y)$ .
- Each row in the solution array  $YOUT$  corresponds to a time returned in the column vector  $TOUT$ .
- To obtain solutions at specific times  $T0, T1, \dots, TFINAL$  (all increasing or all decreasing), use  $TSPAN = [T0 T1 \dots TFINAL]$ .

# Using ode45 to perform integration

```
function VcDot = RCDE(Vc, R, C, Vs)
% Vc is capatotor voltage
% R resistance
% C capacitance
% Vs input control applied voltage
% xdot is the returned 1st time derivative of capacitor voltage

% differential equation for capacitor voltage
VcDot = (Vs - Vc)/(R*C);

% parameters
timeLen = 0.4; % time duration of seconds
samplingRate = 1000; %sampling frequency in Hz
freq = 5; % signal frequency in Hz

% resistance 1000 ohms
R = 1000;
% capacitor value 100 microFarads
C = 100e-6;
% step in secoinds
h = 1/samplingRate;
% time sampling points
T = 0:h:timeLen;

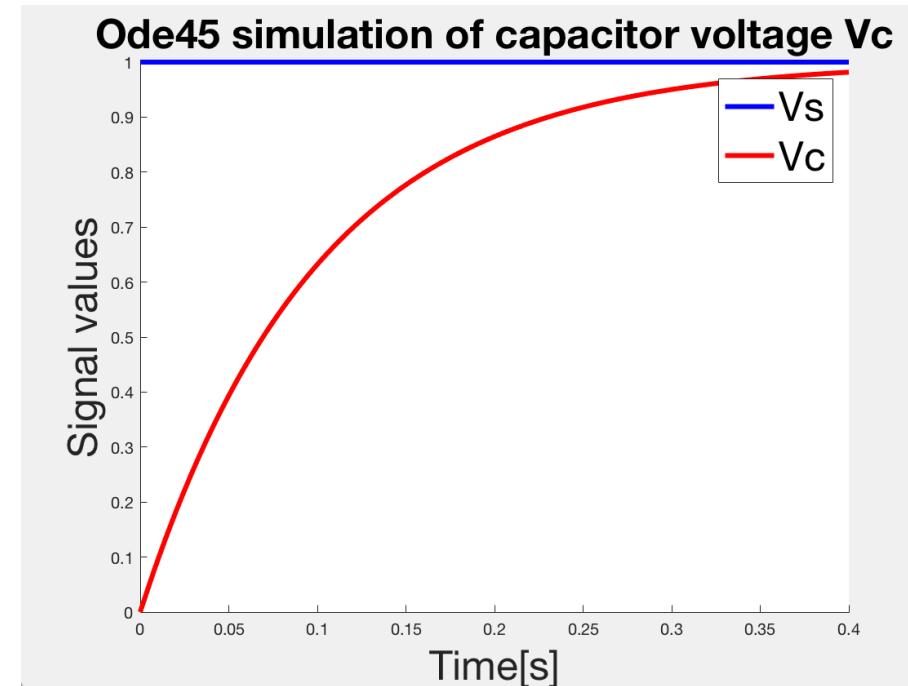
% number of data samples
sampleCnt = length(T);

% build target unity driving voltage
Vs = ones(size(T));

% initial conditions - no capacotr votlage
V0 = 0;

% run simulation
[tSS, Vc] = ode45(@(t,x)RCDE(x, R, C, Vs(1)), T, V0);
```

```
% plot the figure
figure
% want all plots to remain on figure
hold on
% write a title
h = title('Ode45 simulation of capacitor voltage Vc');
set(h, 'FontSize', 25)
% plot driving voltage Vs
h = plot(T, Vs, 'b');
set(h,'LineWidth', 3);
% plot capacitor voltage Vx
h = plot(T, Vc, 'r');
set(h,'LineWidth', 3);
% label plots
h=legend('Vs','Vc');
set(h, 'FontSize', 25)
% label the axes
h=xlabel('Time[s]')
set(h, 'FontSize', 25)
h = ylabel('Signal values');
set(h, 'FontSize', 25)
```



# Interlude

## 10 minute break

# **ROCO218: Control Engineering**

## Lecture 2

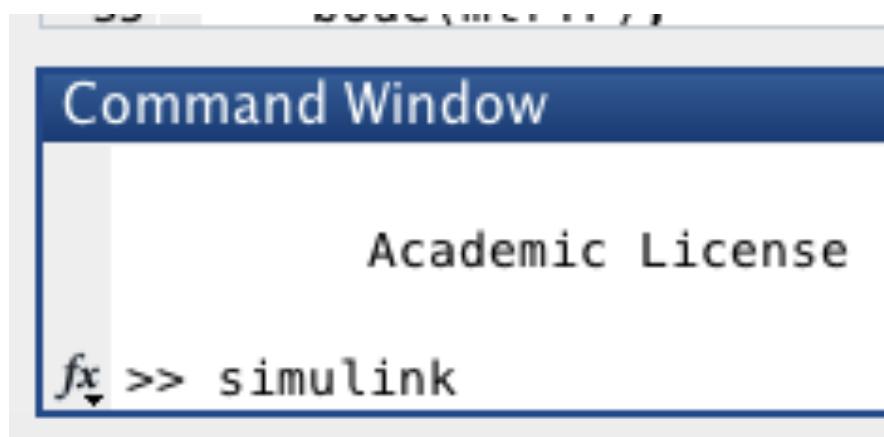
### Introduction to Simulink

# Simulink

- Used to model, analyze and simulate dynamic systems using block diagrams.
- Provides a graphical user interface for constructing block diagram of a system – therefore is easy to use.
- Can be used for simulation of various systems:
  - Linear, nonlinear
- Input signals can be arbitrarily generated:
  - Standard: sinusoidal, polynomial, square, impulse
  - Customized: from a function, look-up table
- Output signals can be stored or demonstrated in different ways.
  - However modeling a system is not necessarily easy !

# Launching Simulink

- First launch Matlab
- Then in the Matlab command window launch simulink
- Do this at the >> prompt by typing **simulink**
- and press ↵Enter

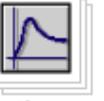
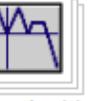
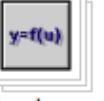
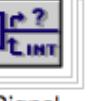
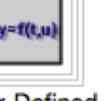
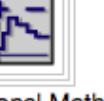


# Simulink window

Simulink

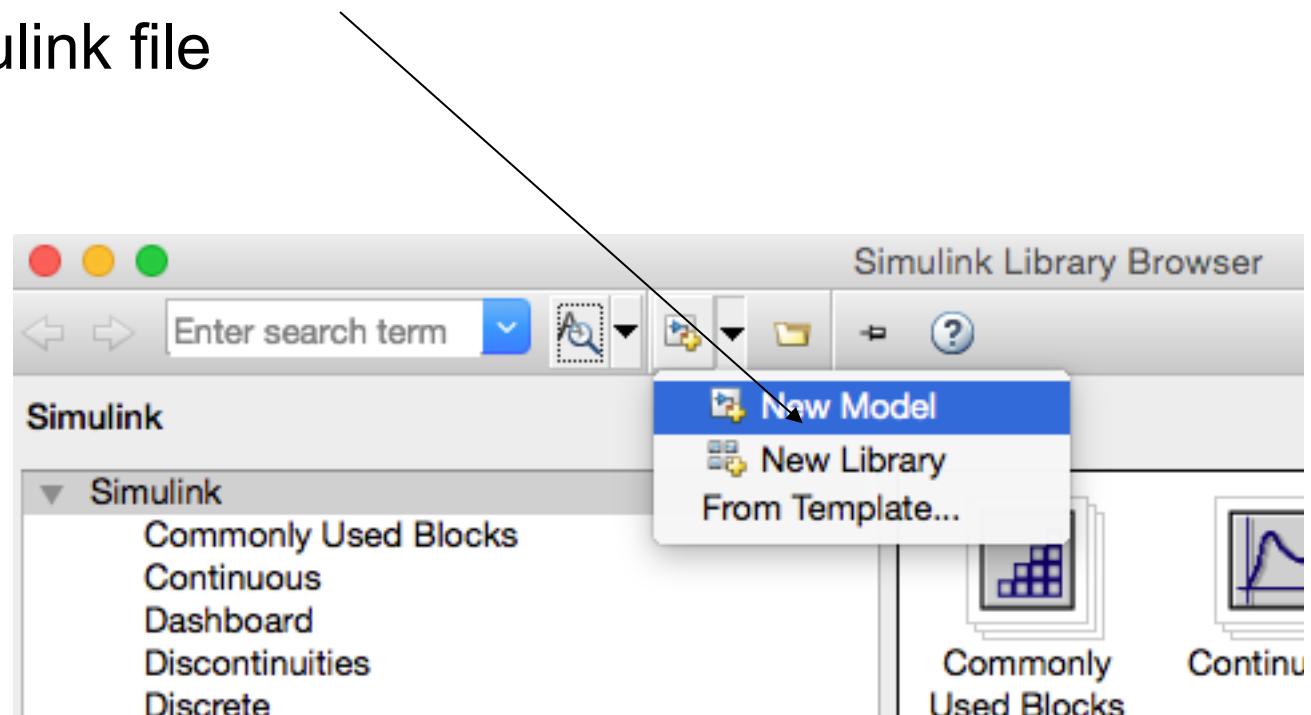
▼ Simulink

- Commonly Used Blocks
- Continuous
- Dashboard
- Discontinuities
- Discrete
- Logic and Bit Operations
- Lookup Tables
- Math Operations
- Model Verification
- Model-Wide Utilities
- Ports & Subsystems
- Signal Attributes
- Signal Routing
- Sinks
- Sources
- User-Defined Functions
  - Additional Math & Discrete
  - Communications System Toolbox
  - Communications System Toolbox HDL Support
  - Computer Vision System Toolbox
  - Control System Toolbox
  - DSP System Toolbox
  - DSP System Toolbox HDL Support
  - Fuzzy Logic Toolbox
  - HDL Coder
  - Image Acquisition Toolbox
  - Instrument Control Toolbox
  - Model Predictive Control Toolbox
  - Neural Network Toolbox
  - Robust Control Toolbox
  - Simscape
  - Simulink 3D Animation
  - Simulink Coder
  - Simulink Control Design
  - Simulink Design Optimization

 Commonly Used Blocks	 Continuous	 Discontinuities	 Discrete
 Logic and Bit Operations	 Lookup Tables	 Math Operations	 Model-Wide Utilities
 Model Verification	 Ports & Subsystems	 Signal Attributes	 Signal Routing
 Sinks	 Sources	 User-Defined Functions	 Additional Math & Discrete

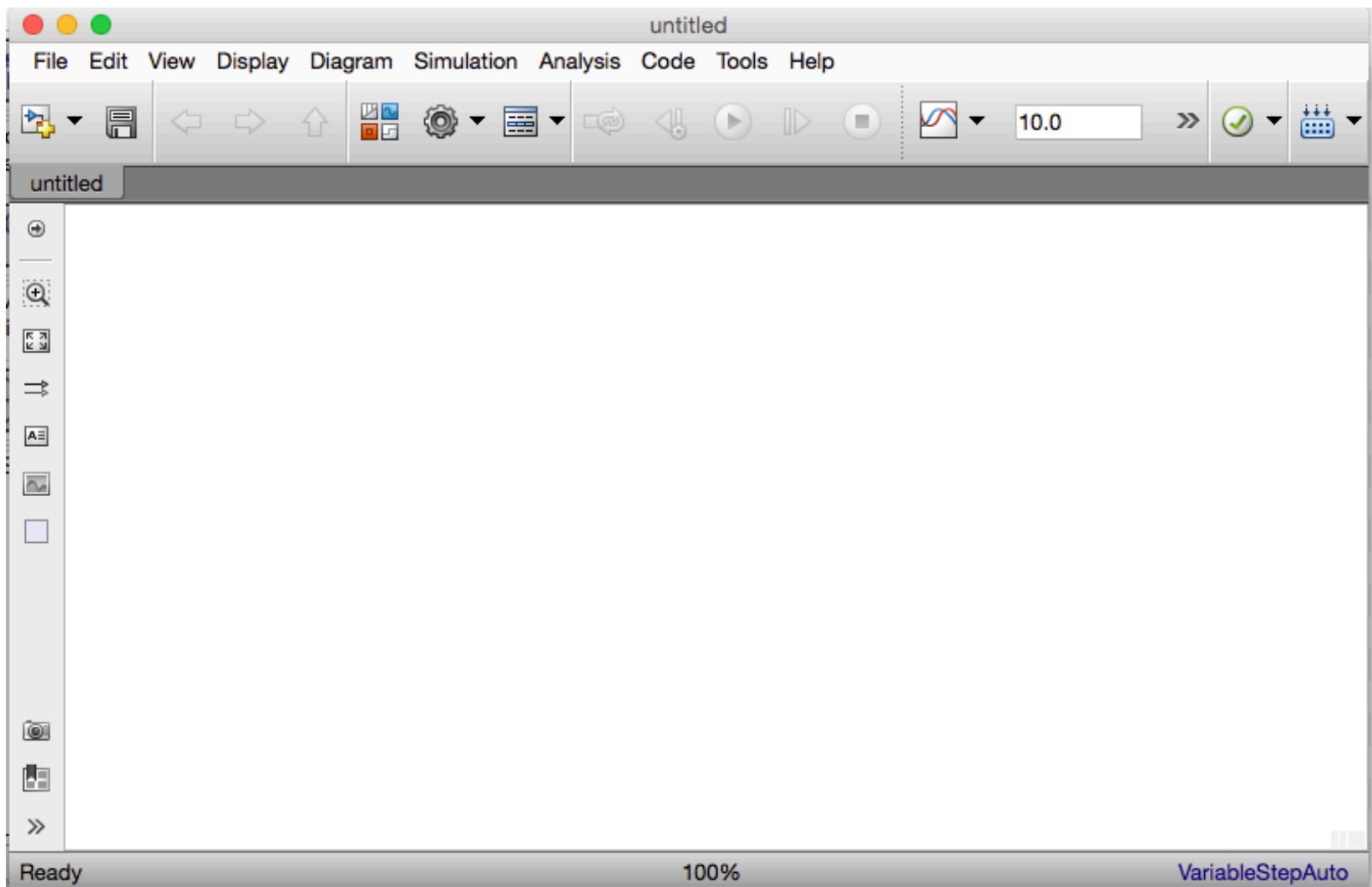
# Create a new model

- Next click the new-model icon in the upper left corner to start a new Simulink file



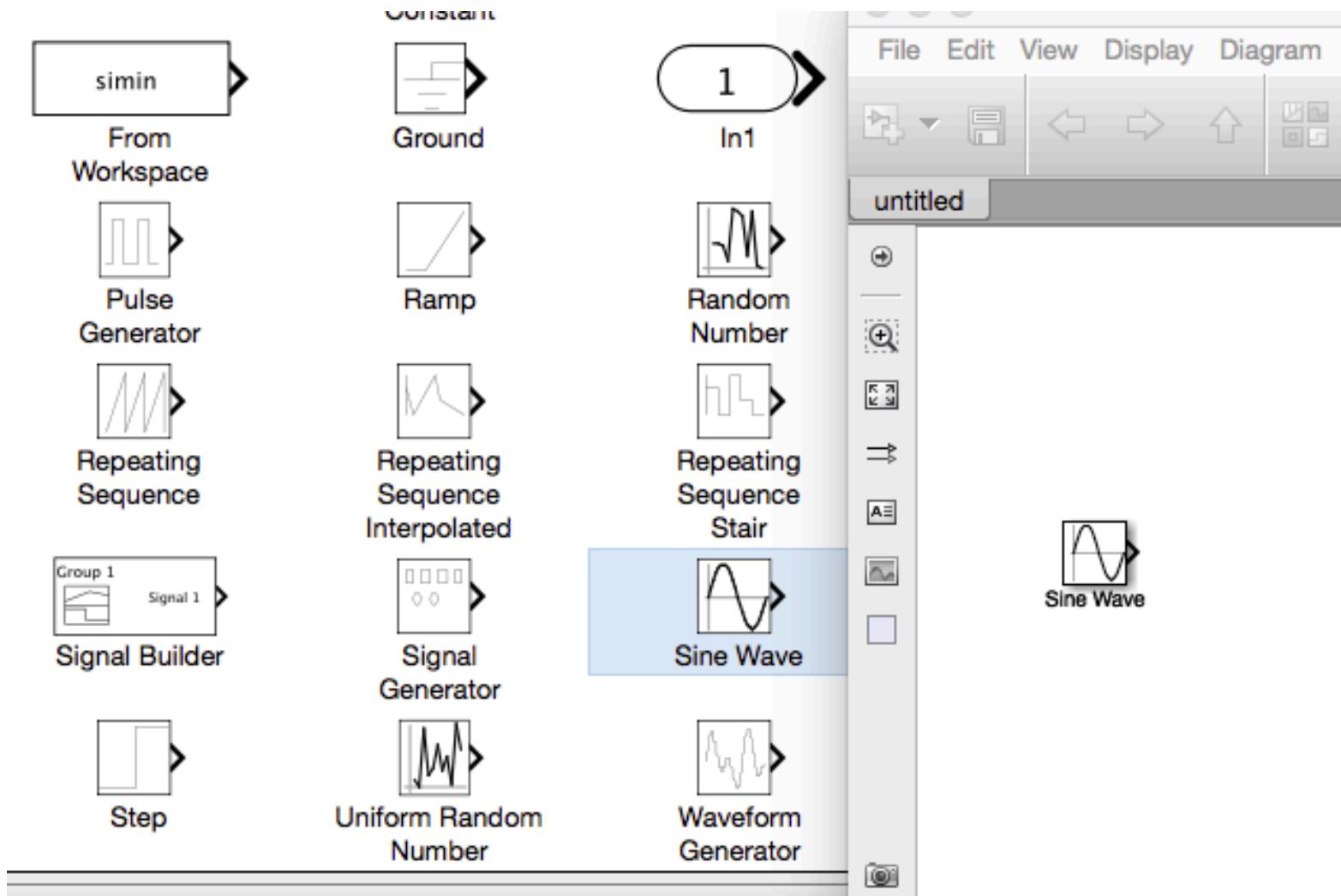
# Your workspace

- This opens a window



# Add components

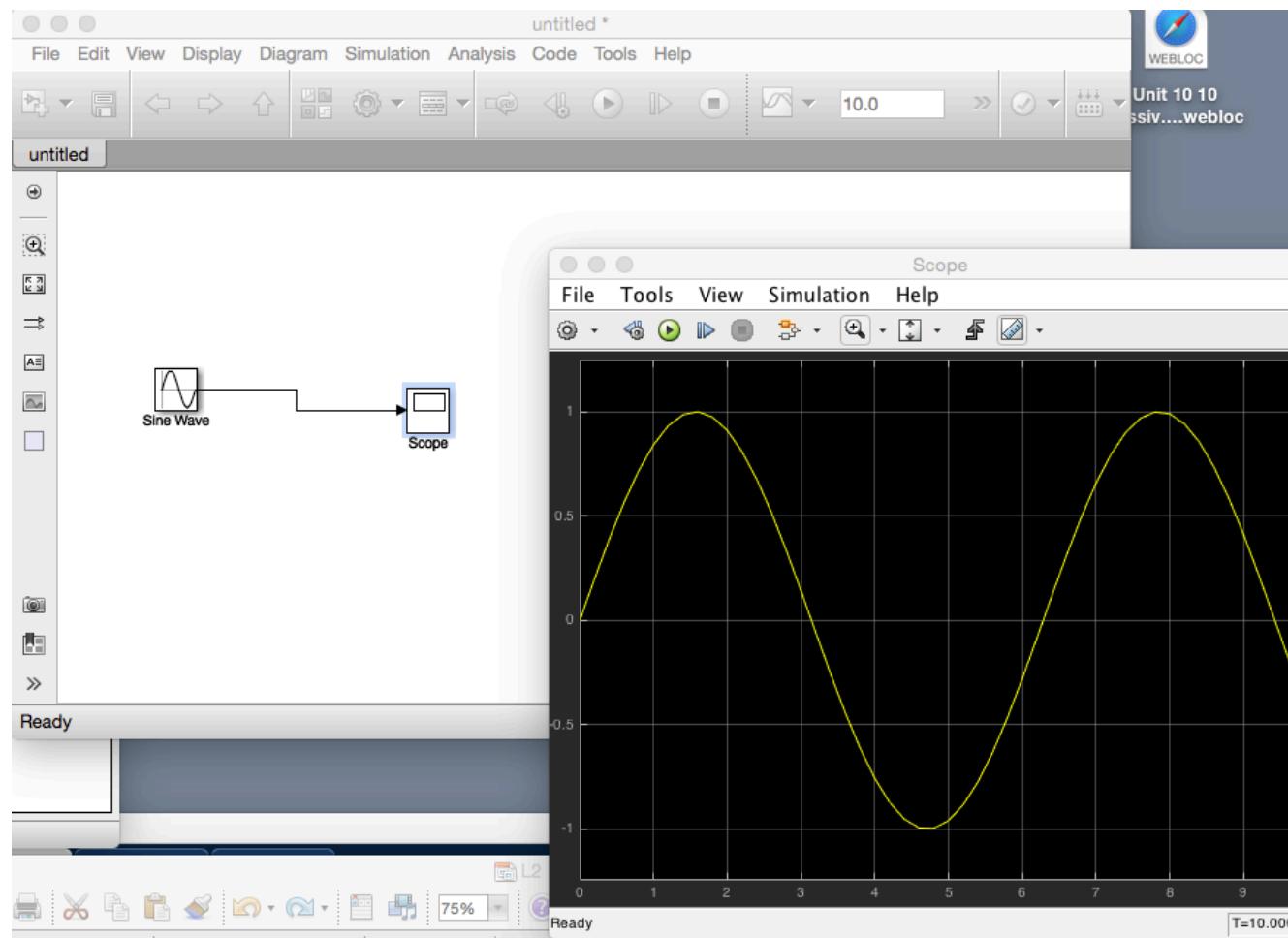
- Drag selected components into workspace



# Connect components and run model



- Connect components



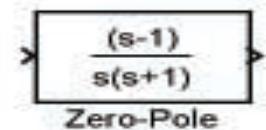
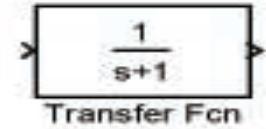
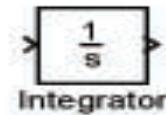
- Click run button
- Click on scope to open display

# Available components blocks

Main components exhibit dynamics:

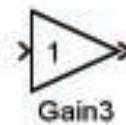
- integrators,
- transfer function
- zero-pole description

Parameters can be set for various components



Also math components:

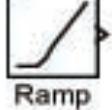
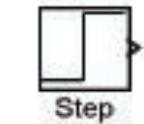
- gain (amplifier)  $kx$  :  $x$  a scalar
- addition ( $a+b+c$ )
- product ( $abc$ )
- number of i/o points and the sign of each are adjustable



# Many components blocks available

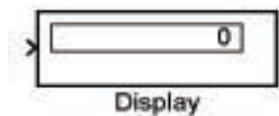
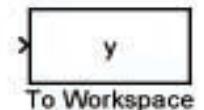
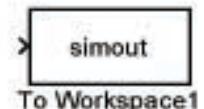
Sources: input signals

- constant, step, ramp
- pulse, sine wave, square wave
- from data file
- signal generator
- a clock to record time

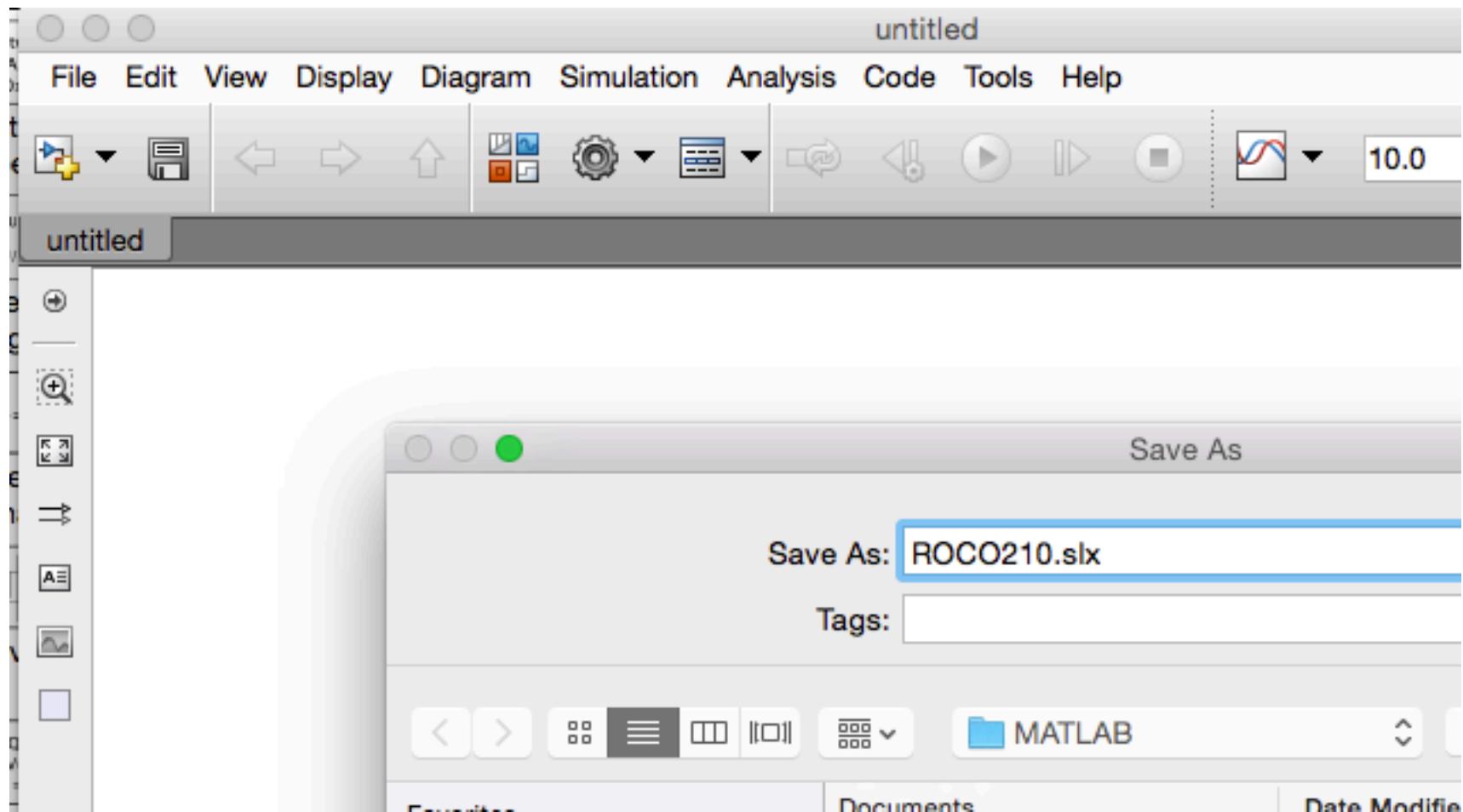


Sinks: for output display or storage

- export to workspace; you can give a name to the variable, such as u, y, x, etc.
- scope
- digital display



# Save your model



# **ROCO218: Control Engineering**

## **Dr Ian Howard**

### Lecture 1

Simulink 2<sup>nd</sup> order differential equation

# Mass-spring-damper system

- Build a Simulink model that solves the following differential equation
  - 2nd-order mass-spring-damper system
  - zero initial conditions
  - input  $f(t)$  is a step with magnitude 3
  - parameters:  $m = 0.25$ ,  $c = 0.5$ ,  $k = 1$

$$m\ddot{x} + c\dot{x} + kx = f(t)$$

# Create the simulation diagram

The simulation diagram for solving the ODE is created step by step.

After each step, elements are added to the Simulink model.

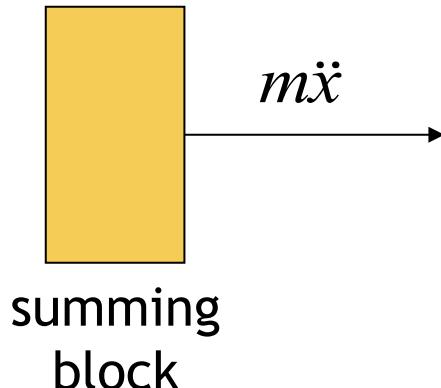
$$m\ddot{x} + c\dot{x} + kx = f(t)$$

# Create the simulation diagram

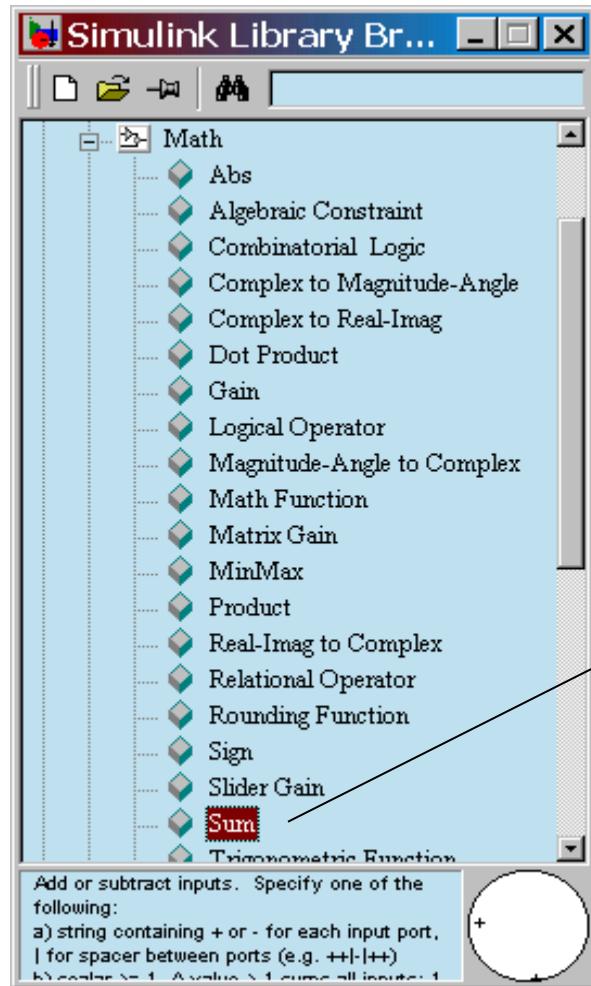
- First, solve for the term with highest-order derivative

$$m\ddot{x} = f(t) - c\dot{x} - kx$$

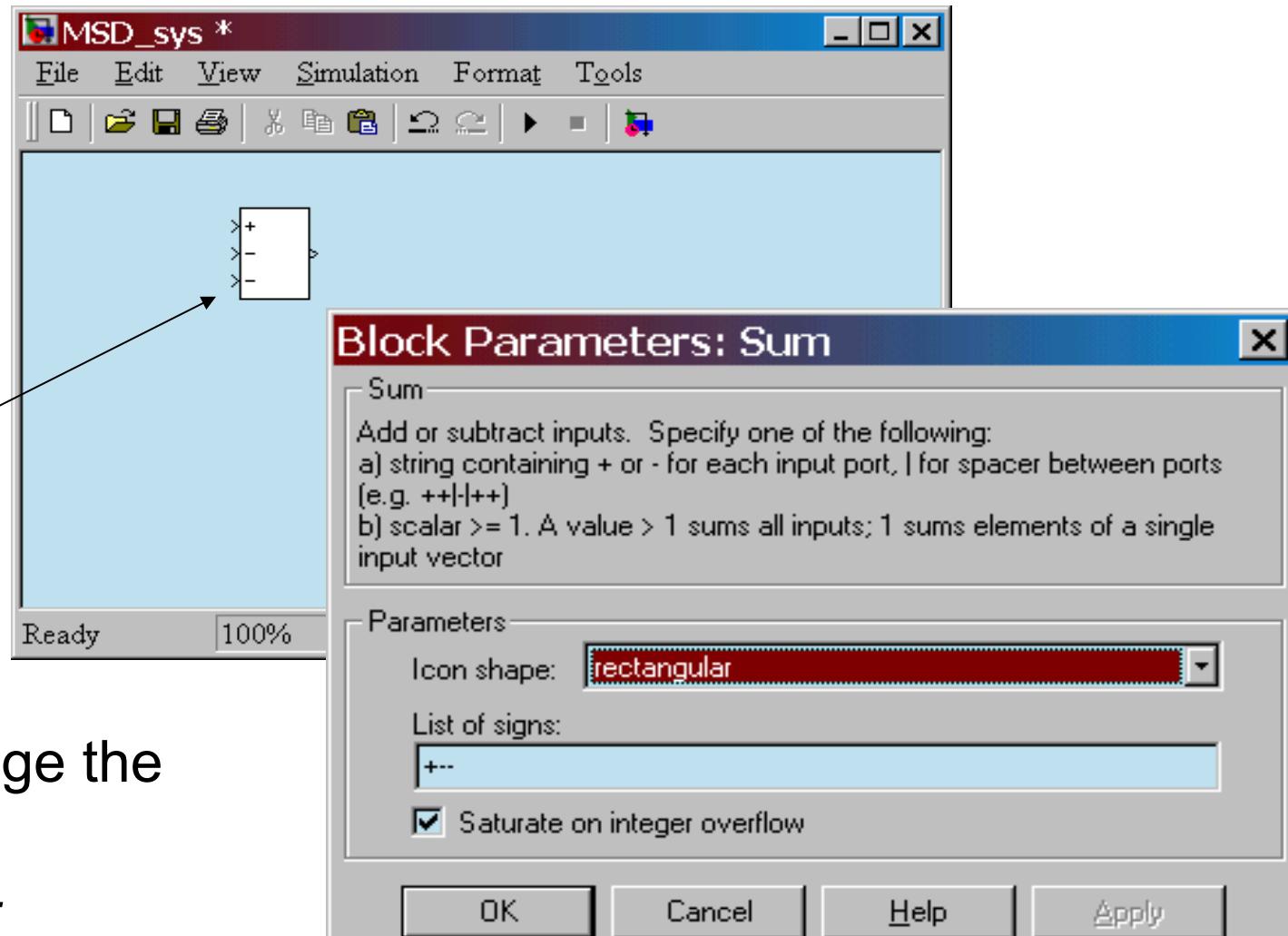
- Make the left-hand side of this equation the output of a summing block



# Create the simulation diagram



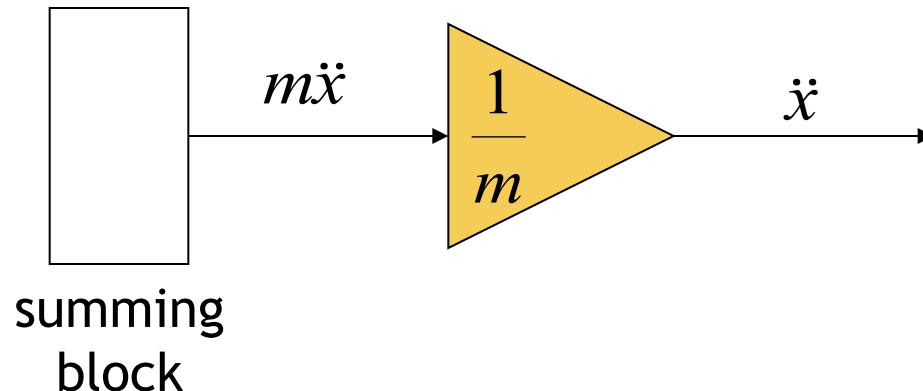
Drag a *Sum* block from the *Math* library



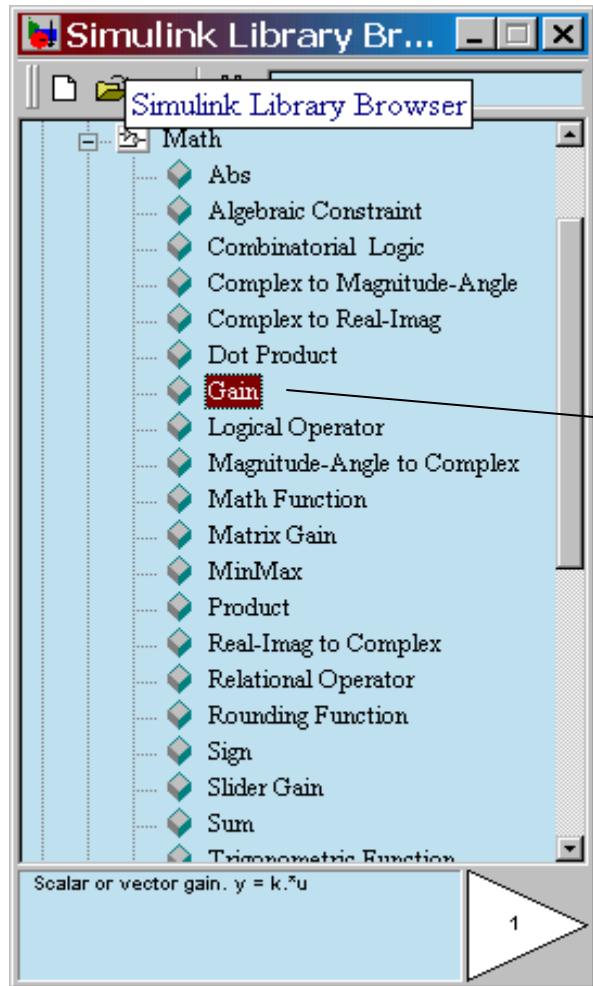
Double-click to change the  
block parameters to  
*rectangular* and *+ - -*

# Create the simulation diagram

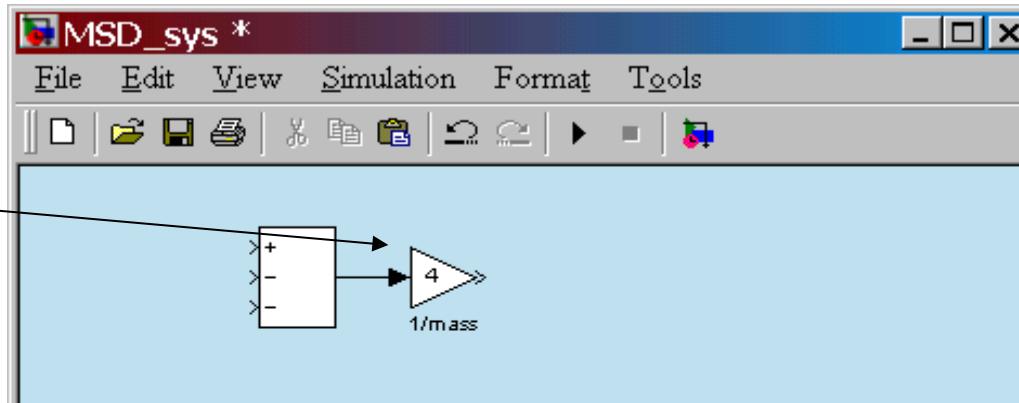
- Add a gain (multiplier) block to eliminate the coefficient and produce the highest-derivative alone



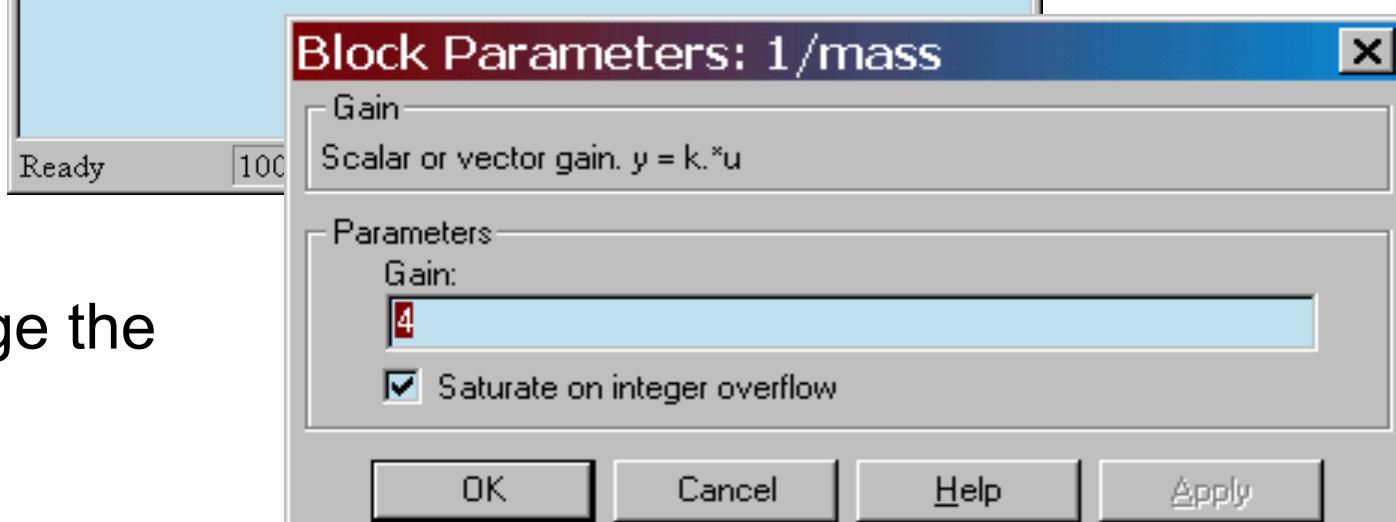
# Create the simulation diagram



Drag a *Gain* block from the *Math* library



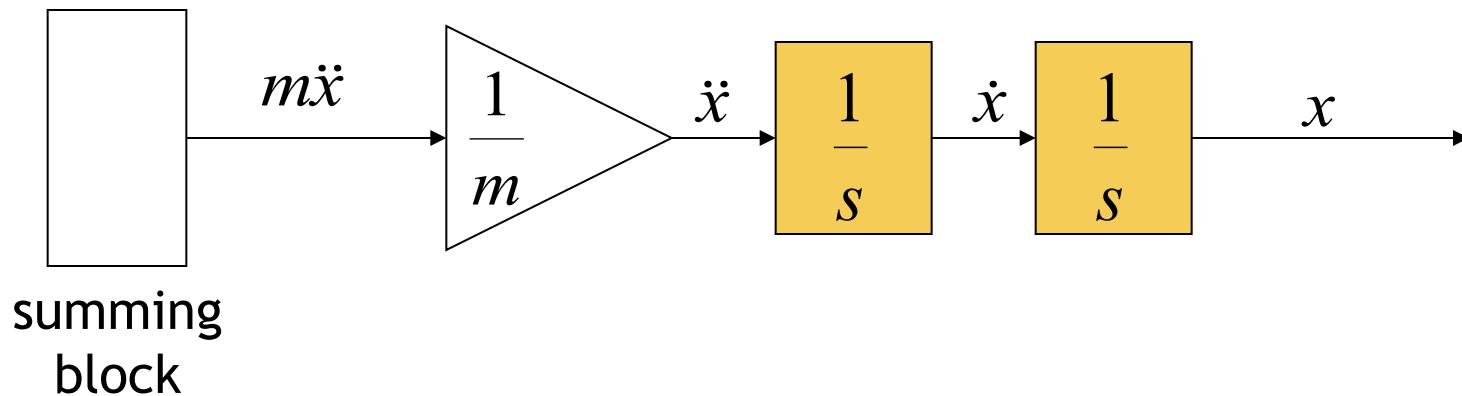
The gain is 4 since  $1/m=4$ .



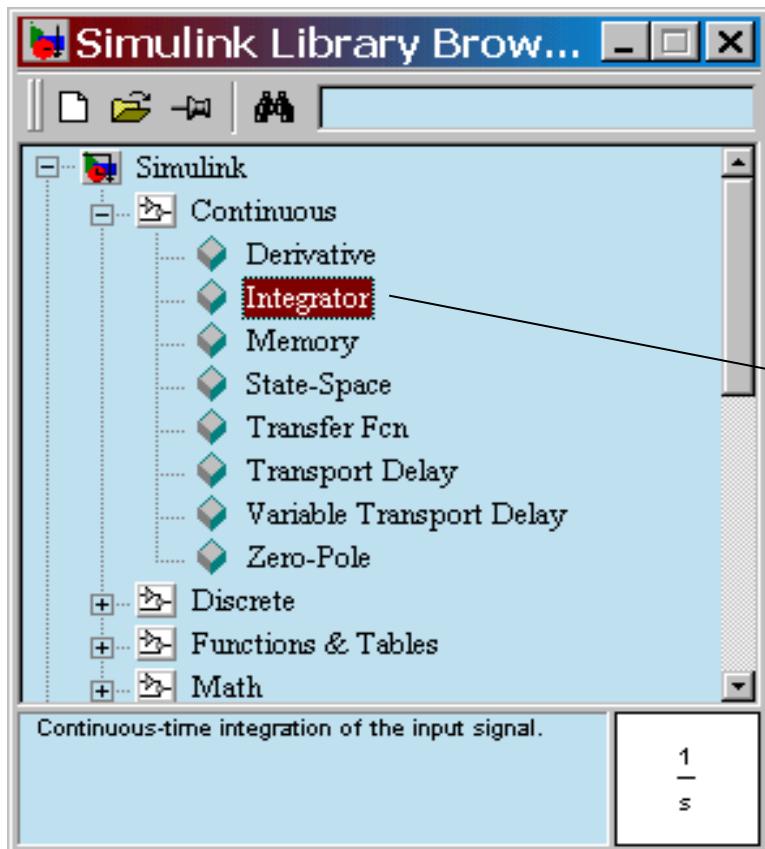
Double-click to change the block parameters.  
Add a title.

# Create the simulation diagram

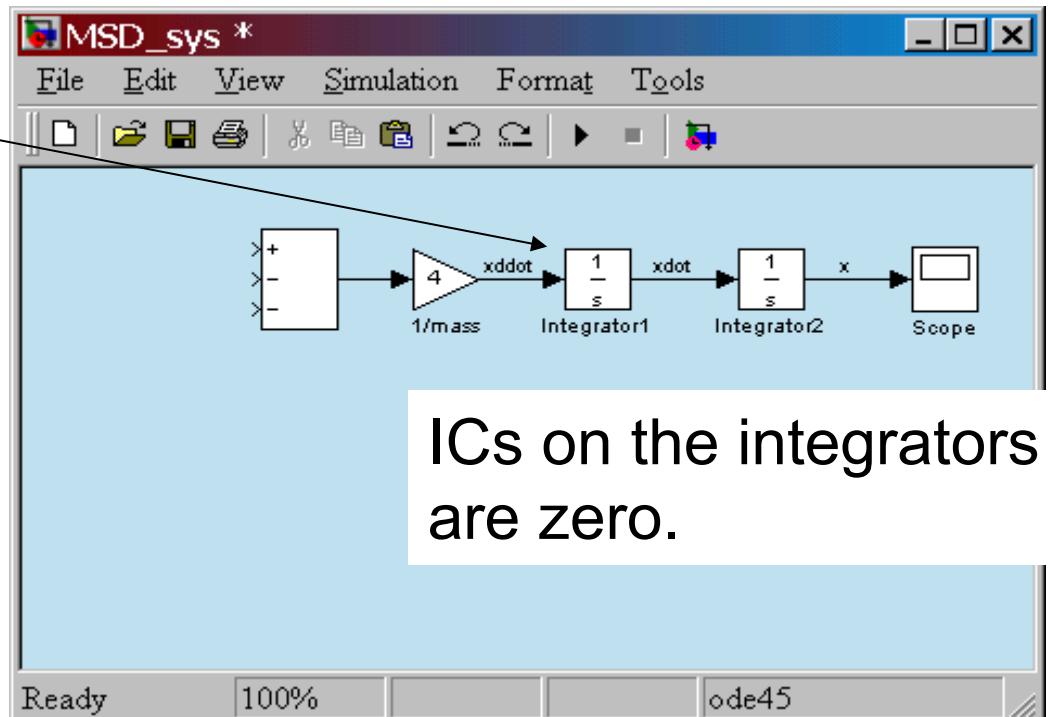
- Add integrators to obtain the desired output variable



# Create the simulation diagram



Drag *Integrator* blocks from the *Continuous* library



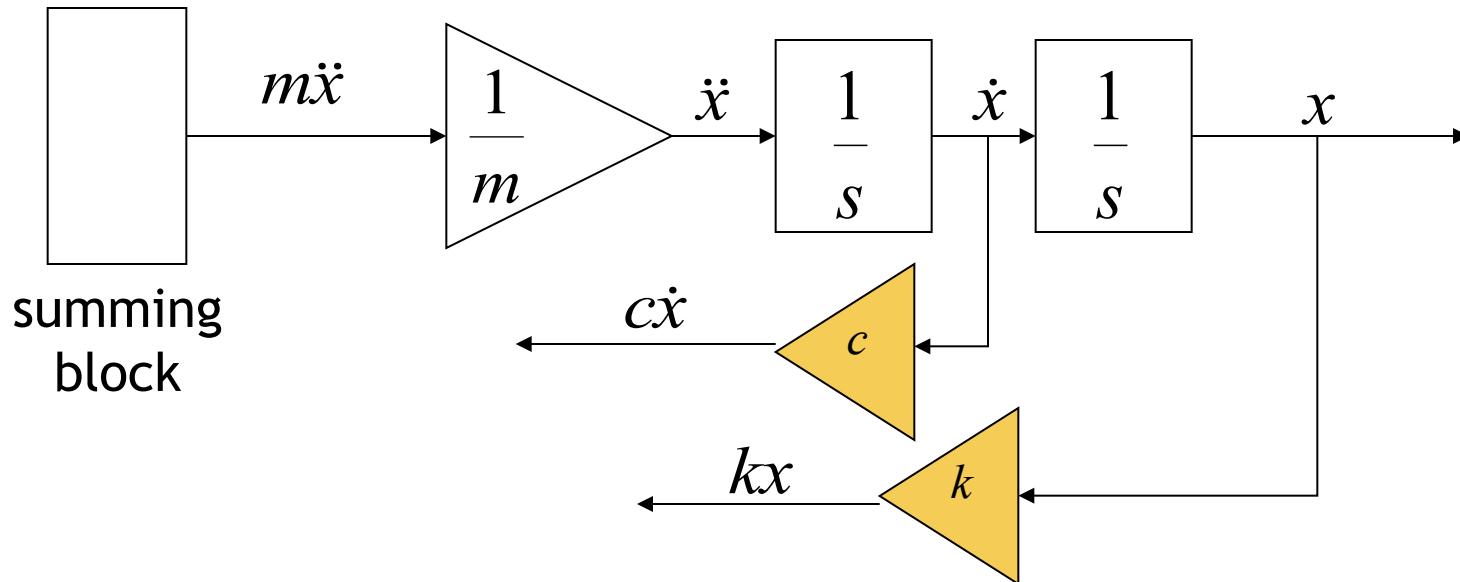
ICs on the integrators are zero.

Add a scope from the *Sinks* library.  
Connect output ports to input ports  
Label the signals by double-clicking on the leader line

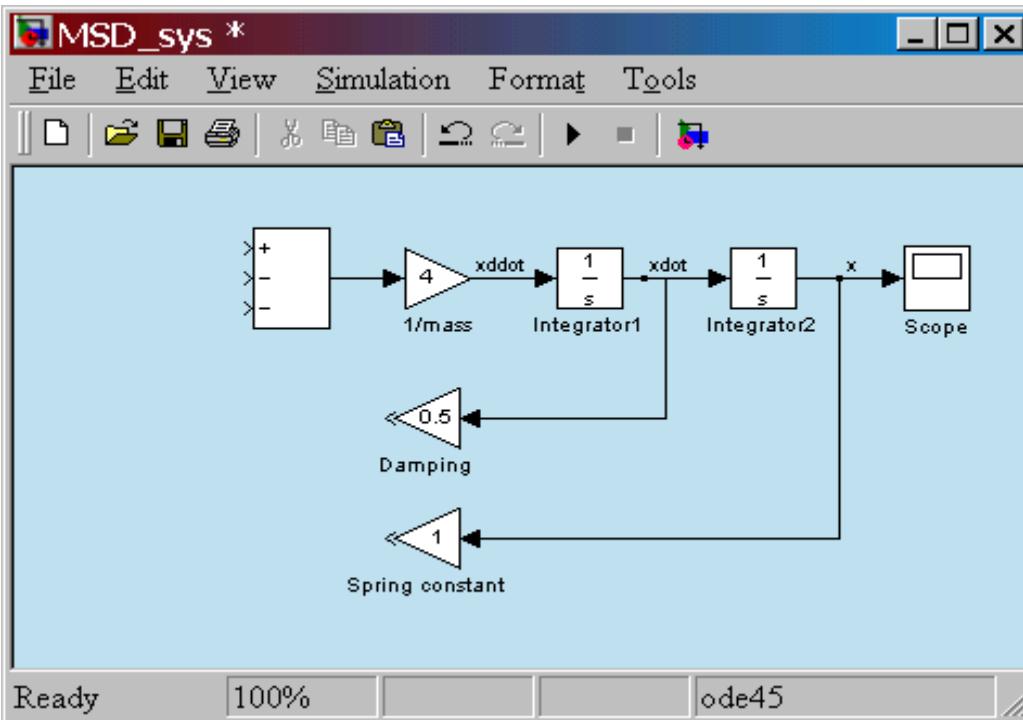
# Create the simulation diagram

$$m\ddot{x} = f(t) - c\dot{x} - kx$$

- Connect to the integrated signals with gain blocks to create the terms on the right-hand side of the equation of motion

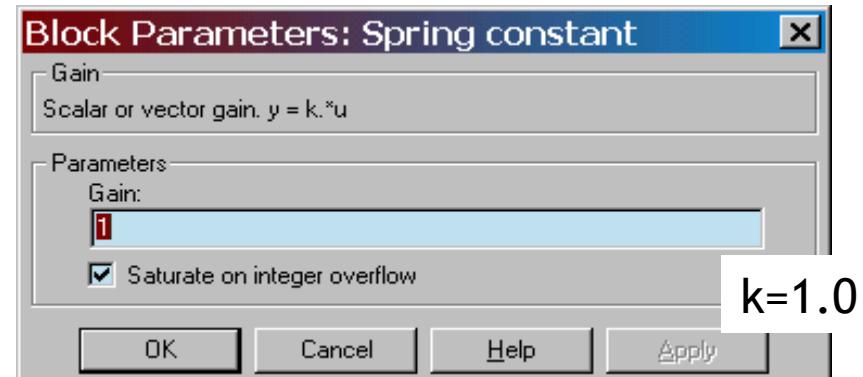
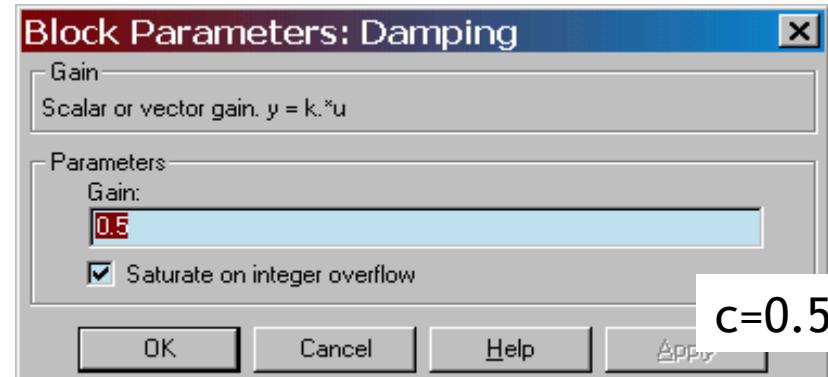


# Create the simulation diagram



Drag new *Gain* blocks from the *Math* library

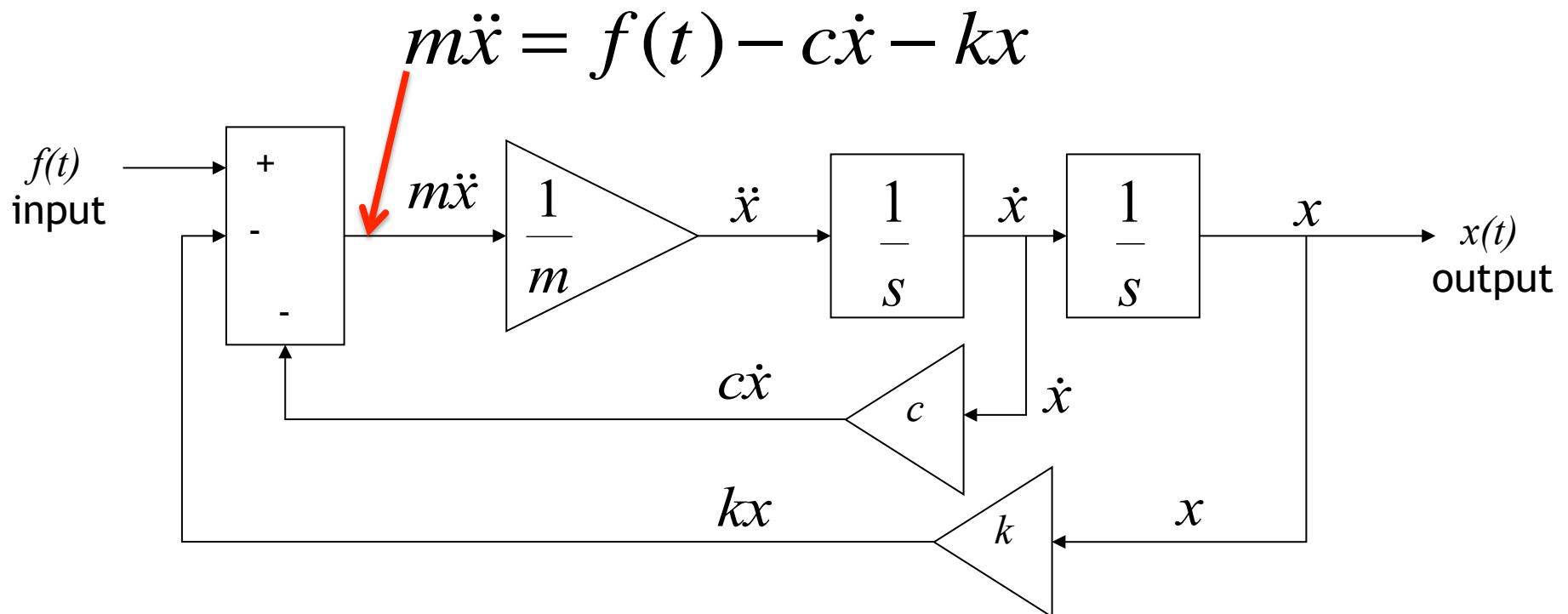
To flip the gain block, select it and choose *Flip Block* in the *Format* pull-down menu.



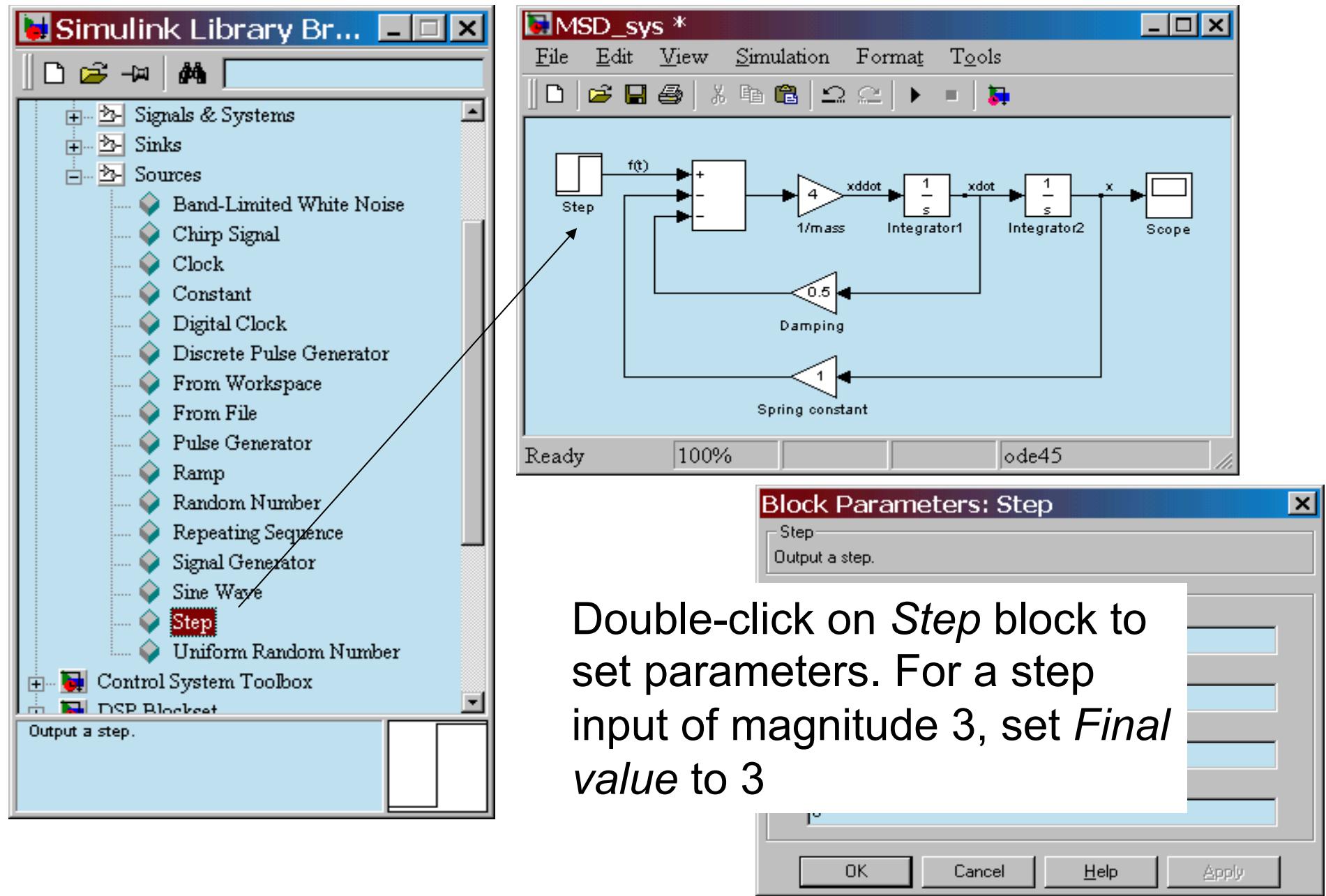
- Double-click on gain blocks to set parameters
- Connect from the gain block input backwards up to the branch point.
- Re-title the gain blocks.

# Complete the model

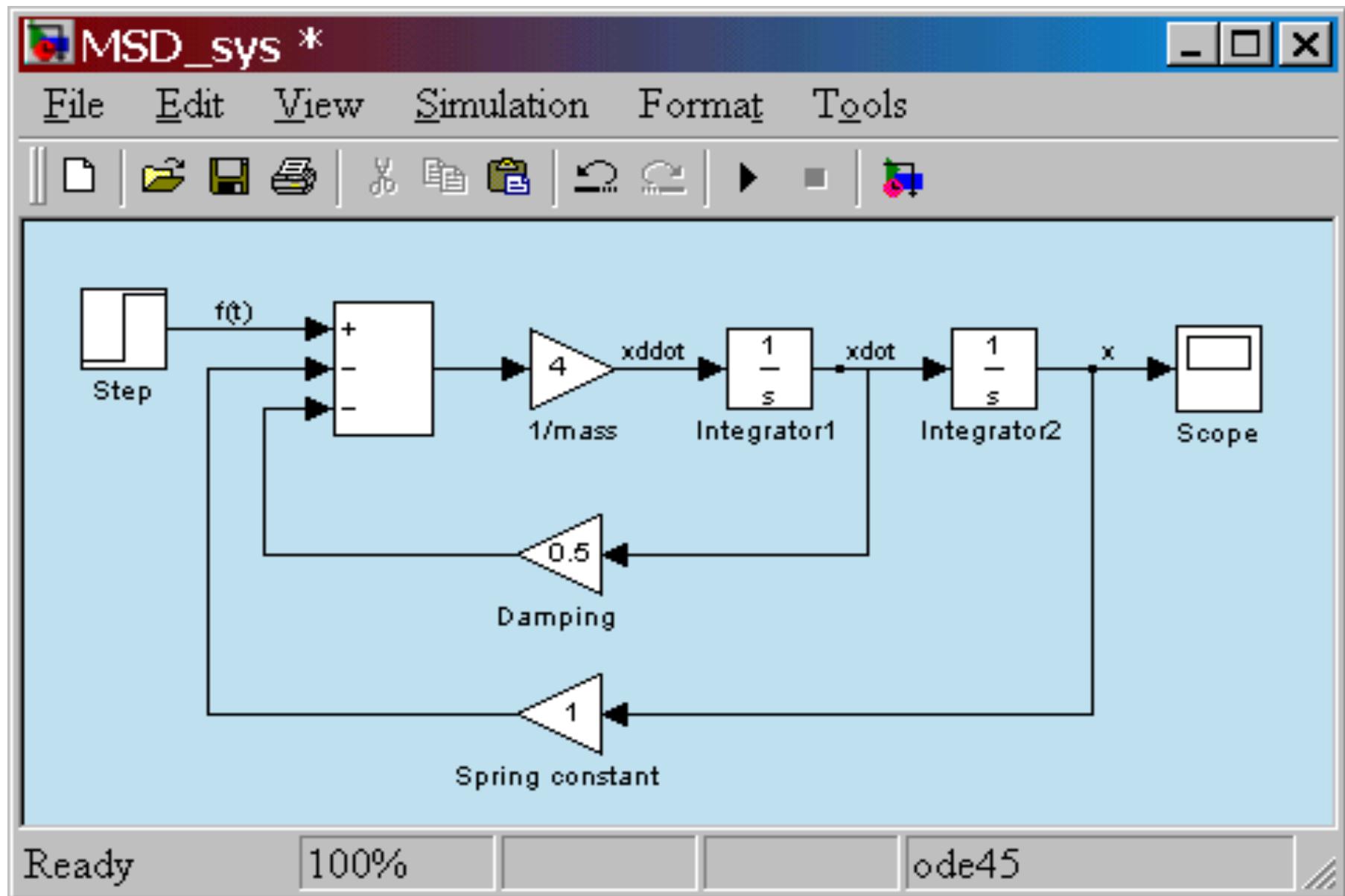
- Bring all the signals and inputs to the summing block
- Check signs on the summer



# Drive with a step function



# Final Simulink model



# Run the simulation

