

# Homework #2 Solutions

## Problem 2-1

$$y[n] = \sum_{k=-\infty}^{\infty} h[k] x[n-k]$$

Let  $h[n]$  be non zero for  $0 \leq n \leq M-1$

Let  $x[n]$  be non zero for  $0 \leq n \leq N-1$

$$y[n] = \sum_{k=0}^{M-1} h[k] x[n-k] \quad \text{with } k=0$$

$y[n]$  is non zero when  $0 \leq n-k \leq N-1$ .

This  $n-k$  equals  $\phi$  when  $n=0$  and  $k=0$ .

It is maximum when  $n-k=N-1$ . This happens when  $k=M-1$ . So,

$$n - (M-1) = N-1$$

$$n = M + N - 2$$

Hence, the total number of output samples is

$$M + N - 2$$

Problem 2-2

$$y[n] = \sum_{m=-\infty}^{\infty} h[m] x[n-m]$$

where  $x[n]$  is a periodic sequence  
 $\text{so, } x[n-m+kN] = x[n-m]$

$$\begin{aligned} y[n+kN] &= \sum_{m=-\infty}^{\infty} h[m] x[n-m+kN] \\ &= \sum_{m=-\infty}^{\infty} h[m] x[n-m] \end{aligned}$$

Thus,  $y[n]$  is also periodic with period  $N$ .

Problem 2-3

$$\begin{aligned} y[n] &= x[n] * h[n] \\ &= (3\delta[n-2] - 2\delta[n+1]) * (-\delta[n+2] + 4\delta[n] - 2\delta[n-1]) \\ &= -3\delta[n-2] * \delta[n+2] + 12\delta[n-2] * \delta[n] - 6\delta[n-2] * \delta[n-1] \\ &\quad + 2\delta[n+1] * \delta[n+2] - 8\delta[n+1] * \delta[n] + 4\delta[n+1] * \delta[n-1] \\ &= -3\delta[n] + 12\delta[n-2] - 6\delta[n-3] + 2\delta[n+3] - 8\delta[n+1] + 4\delta[n] \\ &= 2\delta[n+3] - 8\delta[n+1] + 8\delta[n] + 12\delta[n-2] - 6\delta[n-3] \end{aligned}$$

Problem 2-4  $h[n]$  is non-negative for  $n \geq 0$

$$y[n] = \sum_{k=-\infty}^{\infty} h[k] x[n-k] = \sum_{k=0}^n h[k], n \geq 0$$

when  $x[n] = u[n]$

$$y[n] = 0 \text{ for } n < 0$$

Since  $h[n]$  is non-negative for  $n \geq 0$ ,  $y[n]$  is monotonically increasing for  $n \geq 0$ , and is not oscillatory. Hence, it can have no overshoot.

Problem 2-5

$$x_a(t) = \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega$$

$$= \int_{-\infty}^{\infty} (1) e^{j\omega t} d\omega = \frac{1}{j2\pi t} (e^{-jt} + e^{jt})$$

$$= \frac{1}{j2\pi t} (e^{jt} - e^{-jt}) = \frac{\sin(t)}{\pi t} =$$

Problem 2-6

$$\begin{aligned}
 (a) \quad V_a(j\omega) &= \int_{-\infty}^{\infty} \cos(\omega_0 t) e^{-j\omega t} dt \\
 &= \frac{1}{2} \int_{-\infty}^{\infty} (e^{j\omega_0 t} + e^{-j\omega_0 t}) e^{-j\omega t} dt \\
 &= \frac{1}{2} \int_{-\infty}^{\infty} \left( e^{-j(\omega - \omega_0)t} + e^{-j(\omega + \omega_0)t} \right) dt \\
 &= \frac{1}{2} \delta(\omega - \omega_0) + \frac{1}{2} \delta(\omega + \omega_0)
 \end{aligned}$$

$$\begin{aligned}
 (b) \quad V_a(j\omega) &= \int_{-\infty}^{\infty} e^{j\omega t} e^{-j\omega t} dt = \int_{-\infty}^{\infty} e^{-j(\omega - \omega_0)t} dt \\
 &= \delta(\omega - \omega_0)
 \end{aligned}$$

Notice: Complex spectrum of complex function  
is one-sided. Spectrum of real signal  
is two-sided.

Problem 2-7

$$X_a(j\omega) = \int_{-\infty}^{\infty} x_a(t) e^{-j\omega t} dt$$

$$= |x_a(j\omega)| e^{j\theta_a(\omega)}$$

where  $\theta_a(\omega) = \tan^{-1} \left\{ \frac{\text{Im}(x(j\omega))}{\text{Re}(x(j\omega))} \right\}$

Thus,

$$X_a(-j\omega) = \int_{-\infty}^{\infty} x_a(t) e^{j\omega t} dt$$

$$= |x_a(-j\omega)| e^{+j\theta_a(\omega)}$$

Since  $x_a(t)$  is real-valued,  $X(j\omega)$  and  $X(-j\omega)$  are complex conjugates. Therefore,

$$|X_a(-j\omega)| = |X_a(j\omega)|$$

and  $\theta_a(-\omega) = -\theta_a(\omega)$

So, for real-valued function, the magnitude spectrum is an even function and the phase spectrum is an odd function.

Problem 2-3

The impulse response is found by using a delta function as the input and then computing the output.

$$\delta[n] \longrightarrow [LTI] \longrightarrow h[n]$$

$$y[n] = \sum_{k=-\infty}^{\infty} s[k] h[n-k] = h[n]$$

The frequency response is the spectrum of the impulse response.

$$H(e^{j\omega}) = \sum_{k=-\infty}^{\infty} h[k] e^{-jk\omega}$$