

LTI systems and the convolution operation

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Index

- 1 Introduction to LTI systems
- 2 Discrete-time LTI systems
- 3 Continuous-time LTI systems
- 4 Properties of the Convolution operation

- The theory of LTI systems has direct applications in a wide set of technical areas:
 - ▶ Nuclear magnetic resonance spectroscopy
 - ▶ Seismology
 - ▶ Electric circuit design
 - ▶ Control Theory
 - ▶ Any application that involves Signal Processing
- Our goal in Systems and Circuits: predict the output of a given LTI system for a given input.
- In future courses you will face the design of LTI systems according to certain specifications.

Time Invariance

A system is time-invariant if a time shift in the input signal causes a time shift in the output signal.

- Given $y[n] = f(x[n])$, the system is time-invariant if $f(x[n - n_0]) = y[n - n_0]$
 $\forall n_0$.
- Given $y(t) = f(x(t))$, the system is time-invariant if $f(x(t - t_0)) = y(t - t_0)$
 $\forall t_0$.

Linearity

Linear system possesses the important property of superposition.

For any system, consider two arbitrary inputs and their respective outputs:

$$x_1(t) \rightarrow y_1(t)$$

$$x_2(t) \rightarrow y_2(t),$$

the system is linear if

$$ax_1(t) + bx_2(t) \rightarrow ay_1(t) + by_2(t)$$

for any two complex constant $a, b \in \mathbb{C}$.

Linear discrete-time signals

$$ax_1[n] + bx_2[n] \rightarrow ay_1[n] + by_2[n]$$

Index

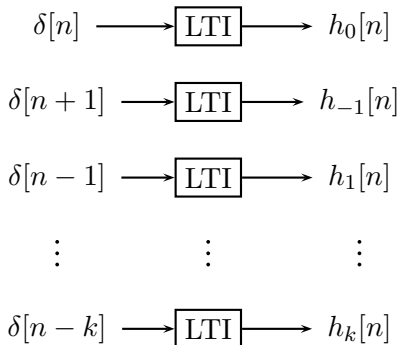
- 1 Introduction to LTI systems
- 2 Discrete-time LTI systems**
- 3 Continuous-time LTI systems
- 4 Properties of the Convolution operation



Remember that any discrete-time sequence $x[n]$ can be decomposed as a linear combination of unit impulses:

$$\begin{aligned} x[n] &= \sum_{k=-\infty}^{\infty} x[k]\delta[n-k] \\ &= \dots + x[-20]\delta[n+20] + x[-19]\delta[n+19] + \dots \\ &\quad + x[-1]\delta[n+1] + x[0]\delta[n] + x[1]\delta[n-1] + \dots \end{aligned}$$

If the system is **linear** and we are able to compute



then we can make use of the superposition property!

$$x[n] = \sum_{k=-\infty}^{\infty} x[k] \delta[n - k] \longrightarrow \boxed{\text{LTI}} \longrightarrow y[n]$$

$$y[n] = \sum_{k=-\infty}^{\infty} x[k] h_k[n]$$

- This is just a consequence of the system linearity.
- The problem is reduced to evaluate the system's output for any $\delta[n - k]$.
- The problem can be further reduced by exploiting that the system is **time-invariant**.

$$\delta[n] \longrightarrow \boxed{\text{LTI}} \longrightarrow h_0[n] = h[n]$$

$$\delta[n+1] \longrightarrow \boxed{\text{LTI}} \longrightarrow h_{-1}[n] = h[n+1]$$

$$\delta[n-1] \longrightarrow \boxed{\text{LTI}} \longrightarrow h_1[n] = h[n-1]$$

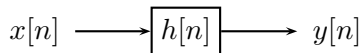
$$\vdots \qquad \qquad \vdots \qquad \qquad \vdots$$

$$\delta[n-k] \longrightarrow \boxed{\text{LTI}} \longrightarrow h_k[n] = h[n-k]$$

Therefore, for any input $x[n]$, if $h[n]$ is known then the system output is given by

$$y[n] = \sum_{k=-\infty}^{\infty} x[k]h_k[n] = \sum_{k=-\infty}^{\infty} x[k]h[n-k]$$

- $h[n]$ is the system's **impulse response**.
- Any LTI system is **completely defined** by $h[n]$!



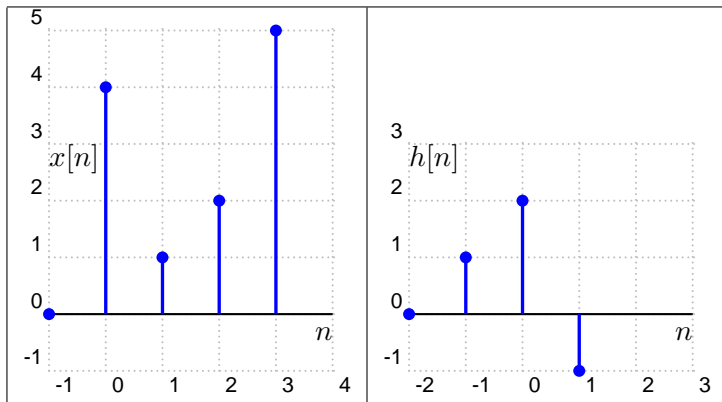
Convolution

Given two discrete-time signals $x[n]$ and $h[n]$:

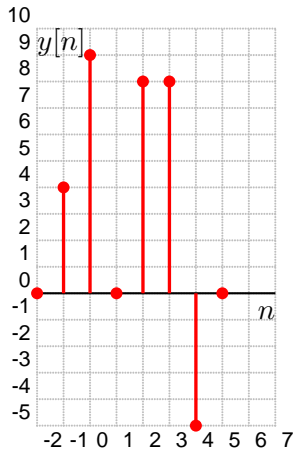
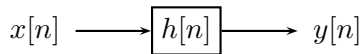
$$\sum_{k=-\infty}^{\infty} x[k]h[n-k] = x[n] * h[n]$$

is the convolution operation between them.

Example

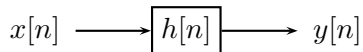


Example



Problem 39

- $x[n] = \alpha^n u[n]$ with $\alpha \in (0, 1)$
- $h[n] = u[n]$



Sol.

$$y[n] = \begin{cases} 0 & n < 0 \\ \frac{1-\alpha^{n+1}}{1-\alpha} & n \geq 0 \end{cases}$$

Index

- 1 Introduction to LTI systems
- 2 Discrete-time LTI systems
- 3 Continuous-time LTI systems**
- 4 Properties of the Convolution operation

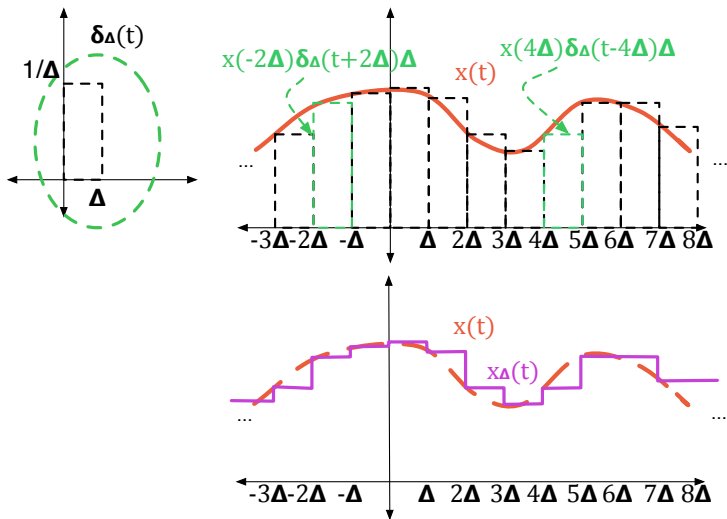
The discussion for continuous-time LTI systems is just a generalization of the discrete-time case.



Remember that any continuous-time signal $x(t)$ can be decomposed as a linear combination of an infinite number of impulses:

$$x(t) = \int_{-\infty}^{\infty} x(\tau) \delta(t - \tau) d\tau$$

Approximate $x(t)$ by a combination of scaled and equally spaced versions of $\delta_{\Delta}(t)$



If the system is also **time-invariant**:

$$\delta_{\Delta}(t - k\Delta) \rightarrow \boxed{\text{LTI}} \rightarrow h_{k\Delta}(t) = h_0(t - k\Delta)$$

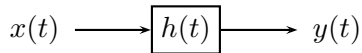
$$x_{\Delta}(t) \rightarrow \boxed{\text{LTI}} \rightarrow y_{\Delta}(t) = \sum_{k=-\infty}^{\infty} x(k\Delta) h_0(t - k\Delta) \Delta$$

Taking the limit $\Delta \rightarrow 0$:

- $k\Delta \rightarrow \tau$
- $\sum \rightarrow \int$
- $\Delta \rightarrow d\tau$
- $h_0(t) = h(t)$

$$y(t) = \lim_{\Delta \rightarrow 0} y_{\Delta}(t) = \int_{-\infty}^{\infty} x(\tau) h(t - \tau) d\tau$$

- $h(t)$ is the **system's response to the impulse** $\delta(t)$.
- Any LTI system is **completely defined** by $h(t)$!



Continuous-time Convolution

Given two signals $x(t)$ and $h(t)$:

$$\int_{-\infty}^{\infty} x(\tau)h(t - \tau)d\tau = x(t) * h(t)$$

is the continuous-time convolution operation between them.

Problem 43

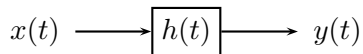
Consider the two signals $x(t)$ and $h(t)$ given by:

$$x(t) = \begin{cases} 1 & 0 < t < T \\ 0 & \text{otherwise} \end{cases},$$

$$h(t) = \begin{cases} t & 0 < t < 2T \\ 0 & \text{otherwise} \end{cases}.$$

If $h(t)$ is the impulse response of an LTI system, compute the system's output when $x(t)$ is the input signal.

Sol.



$$y(t) = \begin{cases} 0 & t < 0 \\ \frac{t^2}{2} & 0 \leq t < T \\ tT - \frac{T^2}{2} & T \leq t < 2T \\ Tt + \frac{3}{2}T^2 - \frac{t^2}{2} & 2T \leq t < 3T \\ 0 & t \geq 3T \end{cases},$$

Index

- 1 Introduction to LTI systems
- 2 Discrete-time LTI systems
- 3 Continuous-time LTI systems
- 4 Properties of the Convolution operation**

Commutative property

$$x[n] * h[n] = h[n] * x[n]$$

Proof.:

Given

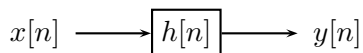
$$x[n] * h[n] = \sum_{k=-\infty}^{\infty} x[k]h[n-k]$$

define $v = n - k$, thus

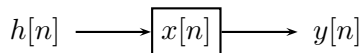
$$x[n] * h[n] = \sum_{v=-\infty}^{-\infty} x[n-v]h[v] = \sum_{v=-\infty}^{+\infty} h[v]x[n-v] = h[n] * x[n]$$

Commutative property

Therefore,



is equivalent to



Commutative property for the continuous-time convolution

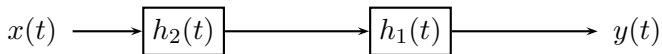
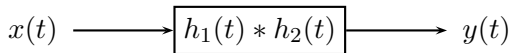
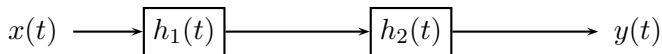
$$x(t) * h(t) = h(t) * x(t) \Rightarrow \int_{-\infty}^{\infty} x(\tau)h(t - \tau)d\tau = \int_{-\infty}^{\infty} h(\tau)x(t - \tau)d\tau$$

Associative property

$$x(t) * (h(t) * z(t)) = (x(t) * h(t)) * z(t)$$

$$x[n] * (h[n] * z[n]) = (x[n] * h[n]) * z[n]$$

Therefore, the following configurations are equivalent:

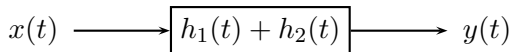
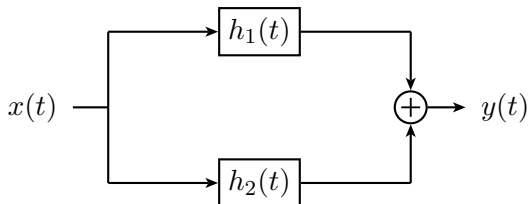


Distributive property with respect to the sum

$$x[n] * (y[n] + z[n]) = x[n] * y[n] + x[n] * z[n]$$

$$x(t) * (y(t) + z(t)) = x(t) * y(t) + x(t) * z(t)$$

The following configurations are equivalent:



Convolution with an impulse signal

Remember that any signal can be decomposed as a linear combination of an infinite number of unit impulses:

$$x[n] = \sum_{k=-\infty}^{\infty} x[k]\delta[n-k],$$
$$x(t) = \int_{-\infty}^{\infty} x(\tau)\delta(t-\tau)d\tau$$

Therefore

$$x[n] * \delta[n] = x[n],$$
$$x(t) * \delta(t) = x(t).$$

Convolution with a delayed impulse (discrete-time)

$$x[n] * \delta[n - n_0] = \sum_{k=-\infty}^{\infty} x[k] \delta[n - k - n_0]$$

Define $v = k + n_0$, thus

$$\begin{aligned} x[n] * \delta[n - n_0] &= \sum_{v=-\infty}^{\infty} x[v - n_0] \delta[n - v] \\ &= x[n - n_0] * \delta[n] = x[n - n_0]. \end{aligned}$$

Therefore,

$$x[n] * \delta[n - n_0] = x[n - n_0],$$

Convolution with a delayed impulse (continuous-time)

$$x(t) * \delta(t - t_0) = \int_{\tau=-\infty}^{\infty} x(\tau) \delta(t - \tau - t_0) d\tau$$

Define $v = \tau + t_0$ and $x'(t) = x(t - t_0)$, thus

$$\begin{aligned} x(t) * \delta(t - t_0) &= \int_{v=-\infty}^{\infty} x(v - t_0) \delta(t - v) dv \\ &= \int_{v=-\infty}^{\infty} x'(v) \delta(t - v) dv = x'(t) * \delta(t) = x'(t) \end{aligned}$$

Therefore,

$$x(t) * \delta(t - t_0) = x(t - t_0),$$