

23

MONDAY
DECEMBER

(357-008) • WK 52

M	T	W	T	F	S	S	
				1	2	3	N
4	5	6	7	8	9	10	O
11	12	13	14	15	16	17	V
18	19	20	21	22	23	24	'19
25	26	27	28	29	30		

DESIGN OF PID

•9 Goal → To design a PID controller with active feedback control loop for angular movement of a dc motor

•10 Problem → No specs given

•11 Assumptions → small general purpose dc motor and typical values.

•12

Armature resistance (R) = 1Ω

•1 Armature inductance (L) = $0.5 H$

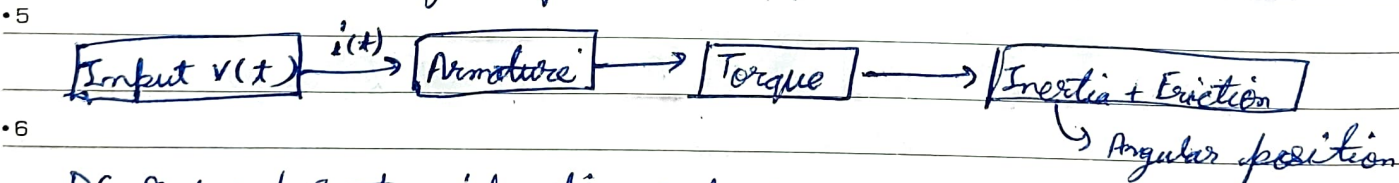
Torque constant (k_t) = $0.01 N \cdot m / A$

•2 Back EMF constant (k_e) = $0.01 V_s / \text{rad}$

Rotor inertia (J) = $0.01 \text{ kg} \cdot \text{m}^2$

•3 Viscous friction coeff. (B) = 0.1 Nms/rad

•4 We are trying to model how the motor responds to voltage input $V(t)$ and produces angular position as output $\theta(t)$



DC Motor has two interacting systems:

- Electrical current (through armature coil)
- Mechanical rotation of the shaft.

ELECTRICAL DYNAMICS [Behaves as RL circuit with Back EMF] ($R \neq L$)

$$V(t) = L \frac{di(t)}{dt} + R i(t) + e_b(t)$$

2019

$$\left\{ e_b(t) = k_e \frac{d\theta(t)}{dt} \right\}$$

J	M	T	W	T	F	S	S
		1	2	3	4	5	
A	6	7	8	9	10	11	12
N	13	14	15	16	17	18	19
20	20	21	22	23	24	25	26
	27	28	29	30	31		

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DECEMBER
SATURDAY

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$$P(s) = \frac{\theta(s)}{V(s)} \quad \hookrightarrow \quad \frac{O/P}{I/P}$$

MECHANICAL DYNAMICS:-

$$\text{Torque} = \text{Inertia} \times \text{Ang. Acc.}$$

$$T(t) = J \frac{d^2 \theta(t)}{dt^2} + B \frac{d\theta(t)}{dt}$$

$$T(t) = k_i i(t)$$

Combining the two equations by eliminating $i(t)$

Laplace transform ($IC=0$)

$$V(s) = L s I(s) + R I(s) + k_b \omega(s)$$

$$k_t I(s) = J s \omega(s) + B \omega(s)$$

$$I(s) = \frac{J s + B}{k_t} \omega(s)$$

$$V(s) = [L s + R] \frac{J s + B}{k_t} \omega(s) + k_b \omega(s)$$

$$\frac{\omega(s)}{V(s)} = \frac{k_t}{(L s + R)(J s + B) + k_b k_t}$$

SUNDAY 22

$$\frac{\omega(s)}{s} = \frac{0.1}{s}$$

$$\frac{\theta(s)}{V(s)} = \frac{k_t}{s[(L s + R)(J s + B) + k_b k_t]}$$

=

$$\frac{0.1}{s(0.005s^2 + 0.06s + 0.100)}$$

2019

20

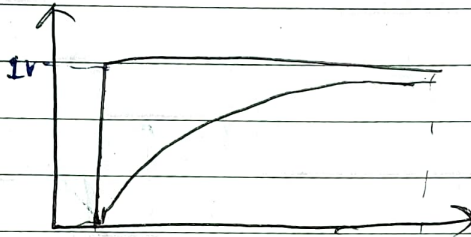
FRIDAY
DECEMBER

(354-011) • WK 51

M	T	W	T	F	S	S
				1	2	3
4	5	6	7	8	9	10
11	12	13	14	15	16	17
18	19	20	21	22	23	24
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NOV
'19In Simulink :-

- 9 • DC motor model (plant)
- PID block
- 10 • Setpoint (reference input)
- Feedback loop
- 11 • Scope to Visualise Results.

• 12 Sim I $T = 50s$

Obs: overdamped

• 2 Probable causes :-

- K_p too low
- K_d high or zero
- K_i not contributing enough

 $K_p = 1$ $K_i = 1$ $K_d = 0$

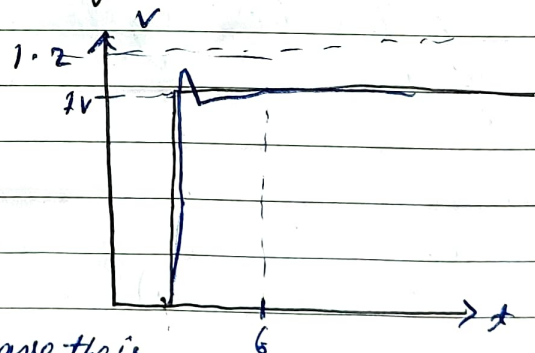
• 4

SII # After 1 hour of mindnumbing trial and error

$$K_p = 200$$

$$K_i = 100$$

$$K_d = .8$$



But such high values have their own cost and tend to be impractical

J	M	T	W	T	F	S	S
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'20	27	28	29	30	31		

WK 51 • (353-012)

DECEMBER
THURSDAY

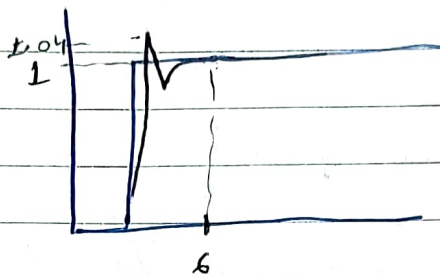
19

STII

$$K_p = 100$$

$$K_i = 100$$

$$K_d = 4$$



Peak 1.04 v

Settling time = 6s

STIV

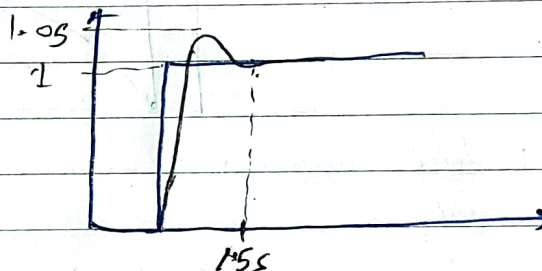
Finally!!

$$K_p = 180$$

$$K_i = 180$$

$$K_d = 5$$

$$N (\text{filter coefficient}) = 100$$



Settling time = 1.5s

Peak overshoot = 1.05v

One of the best system designed by me but there are tradeoffs.

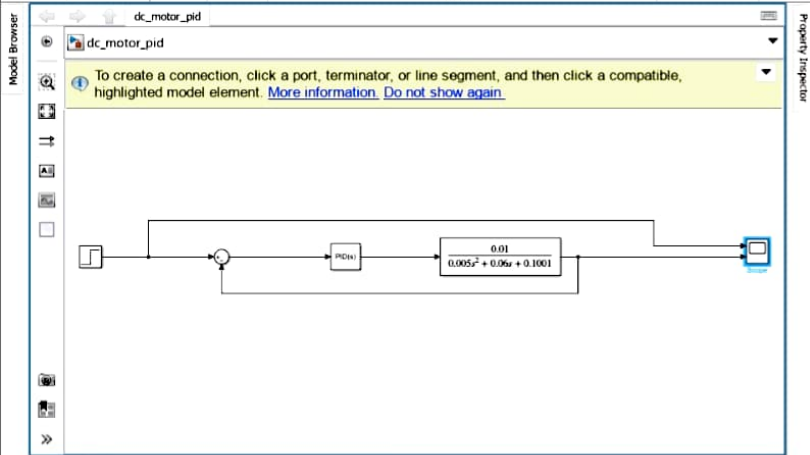
PROS:

- Fast Settling without being erratic or aggressive
- Minimal overshoot ($\approx 5\%$)
- Does not rely solely on massive brute force (high K values)

SIMULATION DEBUG MODELING FORMAT APPS SCOPE

+ Open Stop Time: 20
 New Save LIBRARY PREPARE Normal
 Print Fast Restart Step Back Run Step Forward Stop REVIEW RESULTS

FILE SIMULATE



Diagnostic Viewer

4:13 PM: Simulation

! 0

⚠ 0

i 0

⋮

