# ${\rm DS100M06VL}$

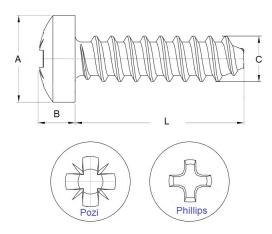
12. oktobar 2024.

# Sadržaj

1	Mounting holes
2	Input connectors 2.1 PCB mount connector
3	Charging connector 3.1 Panel mount USB C
4	Input divider       4         4.1 Resistance divider       4         4.2 Capacitance divider       4         4.2.1 Trim potenciometers       4         4.2.2 Varicap trimming       4
5	Offset results       5         5.1 10x gain enabled       5         5.2 1x gain enabled       5         5.3 solving       5
6	Sources of offset       8         6.1 Input buffer       8         6.1.1 Input bias current offset       8         6.1.2 Input voltage offset       8         6.2 Output buffer       8         6.2.1 Input bias current offset       8         6.2.2 Input voltage offset       8
7	Output diff amp
8	Noise       6         8.1 Resistor noise       6         8.2 Input buffer noise       6         8.3 Output buffer noise       6

# 1 Mounting holes

4 mounting posts are spaced in a 80 mm by 40 mm rectangle. Plastic post hole width is about 2.5 mm, but to allow for a self tapping screw, the PCB hole must be M3.



Apropriate screw is has a B of 2 mm, L of 6 mm, A of 5 mm and C of 3 mm.

# 2 Input connectors

#### 2.1 PCB mount connector



Pomona connector cost around 8 EUR.

### 2.2 Panel mount

# 3 Charging connector

## 3.1 Panel mount USB C

Aliexpress panel mount USB C connector cost around 1.2 EUR per 10 qty.

This connector may not have the required resistors in order to negociate current demand Kupujemprodajem konektor

# 4 Input divider

#### 4.1 Resistance divider

With single sided input impedance of 4 Meg and a required attenuation of 1:400, total impedance from opamp input to gnd must be:

$$R_{in} = \frac{4 \text{ M}\Omega}{399} = 10.025 \text{ k}\Omega \tag{1}$$

Total resistance to ground is a parallel connection of the dividing resistor, 10 M $\Omega$  input offset trim pot and 10 M $\Omega$  opamp input resistance.

Dividing resistor value should then equal:

$$R_{div} = \frac{5\mathrm{M}\Omega \cdot 10.025\mathrm{k}\Omega}{5\mathrm{M}\Omega - 10.025\mathrm{k}\Omega} = 10.045\Omega \tag{2}$$

Placing a 0.1% resistor (10.010 k $\Omega$  max), trim pot should be around 50  $\Omega$ .

#### 4.2 Capacitance divider

4 series 10 pF capacitors yield total 2.5 pF with a tolerance of 5%. In order to achieve 1:400 attenuation, total capacitance on the buffers input pin should be 997.5 pF. If we account for the 5% tolerance, total capacitance should be trimmable in the range for 947.62 pF to 1047.4 pF.

AD8039 features a typical input capacitance of 2 pF and thus should not significantly impact the total capacitance.

#### 4.2.1 Trim potenciometers

Available trimmable capacitor has a capacitance range from 8 pF to 50 pF, not enough to account for the 5% capacitor tolerance.

These trim capacitors are expensive (5 EUR per 1 qty) and are only single turn, limiting trim accuracy.

#### 4.2.2 Varicap trimming

In order to allow for auto-calibration, input capacitance needs to be control voltage dependant. To achieve this, a varicap diode may be used.

This component can be trimmed manually, using a trim potentiometer.

Since varicap add a voltage dependant capacitance value, total capacitance before the adding the varicap needs to be less then the total capacitance required.

As 5% capacitors are used, fixed capacitors used should equal:

$$C_{placed} = \frac{C_{req}}{1.05} - C_{varicap\ min} \tag{3}$$

Parallel connection of 2 470 pF yield capacitance in the range of 893 pF to 987 pF.

Since the capacitance change required is 10% of required capacitance, varicap capacitance change in the available control voltage range (0 V to 5V) must be around 100 pF.

Diode selection:

• Cheaper varicap diode is toshibas at 0.26 EUR per 10 qty.

• Better documented varicap diode is skyworks at 0.51 EUR per 10 qty.

PCB needs to support both of these diode footprints.

### 5 Offset results

#### 5.1 10x gain enabled

-8.5 mV

## 5.2 1x gain enabled

 $9.2~\mathrm{mV}$ 

## 5.3 solving

$$V_{off\_10} = 10 \cdot (2V_{in\_off} + V_{curr\_off}) + V_{out\_off}$$

$$\tag{4}$$

$$V_{off\_1} = 2V_{in\_off} + V_{curr\_off} + V_{out\_off}$$

$$\tag{5}$$

$$V_{off\_10} - V_{off\_1} = 9 \cdot (2V_{in\_off} + V_{curr\_off}) = -17.7 \text{mV}$$
 (6)

Total input offset is -2 mV. Total output offset is 11.2 mV.

## 6 Sources of offset

# 6.1 Input buffer

#### 6.1.1 Input bias current offset

AD8039 has an input bias current offset of 25nA, across the 20 k $\Omega$  input impedance, generates 0.5 mV of offset. Taking into account the second input buffer, maximum offset is 1 mV.

#### 6.1.2 Input voltage offset

AD8039 has an max input voltage offset of 3mV. Worst case total offset is 6mV.

## 6.2 Output buffer

#### 6.2.1 Input bias current offset

AD8009 has an max input current offset of 150 $\mu$ A. With an input impedance of 100  $\Omega$ , total offset is 15 mV. If both inputs have opposing offsets, the total offset is 30 mV

#### 6.2.2 Input voltage offset

AD8009 has an max input voltage offset of 5mV.

# 7 Output diff amp

AD8039 load resistor is it's outputs series resistor.

# 8 Noise

### 8.1 Resistor noise

$$V_{rms} = \sqrt{4k_B T R \Delta f} \tag{7}$$

At the bandwidth of 100 MHz and temperature of 25  $^{\circ}\mathrm{C}$ , noise is

$$V_{rms} = 3.63 \text{mV}_{RMS} \tag{8}$$

## 8.2 Input buffer noise

AD8039 has an input noise level of  $8nV/\sqrt{Hz}$  meaning that total input noise rms at 100 MHz bandwidth is 80  $\mu V_{RMS}.$ 

## 8.3 Output buffer noise

AD8009 has an input noise level of 1.9 nV/ $\sqrt{\rm Hz}$  meaning that total input noise rms at 100 MHz bandwidth is 19  $\mu\rm V_{RMS}.$