

## Properties – Continuous time Fourier series (C.T.F.S.)

<b>Definitions:</b> $a_k = \frac{1}{T} \int_T x(t) e^{-jk\omega_0 t} dt \quad x(t) = \sum_{k=-\infty}^{\infty} a_k e^{jk\omega_0 t}$ $a_0 = \frac{1}{T} \int_T x(t) dt$ $x(t)$ periodic with period $T$ sec., Fundam. angular frequency $\omega_0 = 2\pi f_0 = 2\pi/T$ rad./sec. $x(t) \xleftrightarrow{C.T.F.S.} a_{x,k} \quad y(t) \xleftrightarrow{C.T.F.S.} a_{y,k}$ If $x(t) \xrightarrow{LTI} y(t)$ then $a_{y,k} = a_{x,k} H(j\omega) \big _{\omega=k\omega_0}$
<b>Linearity:</b> $Ax(t) + By(t) \xleftrightarrow{C.T.F.S.} A a_{x,k} + B a_{y,k}$
<b>Shifting:</b> $x(t - t_0) \xleftrightarrow{C.T.F.S.} e^{-jk\omega_0 t_0} a_k$
<b>Scaling:</b> $x(\alpha t) \xleftrightarrow{C.T.F.S.} a_k \quad (\alpha > 0, \text{ period } T/\alpha)$
<b>Flipping:</b> $x(-t) \xleftrightarrow{C.T.F.S.} a_{-k}$
<b>Conjugate:</b> $x^*(t) \xleftrightarrow{C.T.F.S.} a_{-k}^*$ $x^*(-t) \xleftrightarrow{C.T.F.S.} a_k^*$
<b>Symmetries:</b> if $x(t)$ is real: $a_k = a_{-k}^*,  a_k  =  a_{-k} , \angle a_k = -\angle a_{-k}$ $x_e(t) = (x(t) + x^*(-t)) / 2 \xleftrightarrow{CTFS} \text{Re}\{a_k\}$ $x_o(t) = (x(t) - x^*(-t)) / 2 \xleftrightarrow{CTFS} j \text{Im}\{a_k\}$
<b>Periodic convolution:</b> $\int_T x(\tau) y(t - \tau) d\tau \xleftrightarrow{C.T.F.S.} T a_k b_k$
<b>Modulation:</b> $x(t)y(t) \xleftrightarrow{C.T.F.S.} a_k * b_k = \sum_{l=-\infty}^{\infty} a_l b_{k-l}$ $e^{jm\omega_0 t} x(t) \xleftrightarrow{C.T.F.S.} a_{k-m}$
<b>Differentiation:</b> $\frac{dx(t)}{dt} \xleftrightarrow{C.T.F.S.} jk\omega_0 a_k$
<b>Parseval:</b> $\frac{1}{T} \int_T  x(t) ^2 dt = \sum_{k=-\infty}^{\infty}  a_k ^2$

## Table of continuous time Fourier series (C.T.F.S.)

$x(t)$ periodic, period $T = \frac{2\pi}{\omega_0} = \frac{1}{f_0}$ sec.	Fourier series coefficients $a_k$
$e^{j\omega_0 t}$	$a_1 = 1$ $a_k = 0$ elsewhere
$\cos(\omega_0 t)$	$a_1, a_{-1} = 1/2$ $a_k = 0$ elsewhere
$\sin(\omega_0 t)$	$a_1 = 1/(2j)$ $a_{-1} = -1/(2j)$ $a_k = 0$ elsewhere
$\begin{cases} 1 &  t  < T_1 \\ 0 & T_1 <  t  < T/2 \end{cases}$ (periodic $T$ )	$a_k = \frac{\sin(k\omega_0 T_1)}{k\pi} \quad k \neq 0$ $a_0 = \frac{T_1 \omega_0}{\pi} = \frac{2T_1}{T}$
1	$a_0 = 1$ $a_k = 0$ elsewhere
$\sum_{n=-\infty}^{\infty} \delta(t - nT)$	$a_k = \frac{1}{T}$

## Properties – Discrete time Fourier series (D.T.F.S.)

<b>Definitions:</b> $a_k = \frac{1}{N} \sum_{n=\langle N \rangle} x[n] e^{-j(k \frac{2\pi}{N})n} \quad x[n] = \sum_{k=\langle N \rangle} a_k e^{j(k \frac{2\pi}{N})n}$ $a_0 = \frac{1}{N} \sum_{n=\langle N \rangle} x[n]$ $x[n]$ periodic with period $N$ samples (fundamental angular frequency $\omega_0 = \frac{2\pi}{N}$ rad./sample) $x[n] \xleftrightarrow{D.T.F.S.} a_{x,k} \quad y[n] \xleftrightarrow{D.T.F.S.} a_{y,k}$ If $x[n] \xrightarrow{LTI} y[n]$ then $a_{y,k} = a_{x,k} H(e^{j\omega}) \Big _{\omega=k \frac{2\pi}{N}}$
<b>Periodicity:</b> $x[n] \xleftrightarrow{D.T.F.S.} a_k = a_{k+N}$
<b>Linearity:</b> $Ax[n] + By[n] \xleftrightarrow{D.T.F.S.} Aa_{x,k} + Ba_{y,k}$
<b>Shifting:</b> $x[n - n_0] \xleftrightarrow{D.T.F.S.} e^{-jk \frac{2\pi}{N} n_0} a_k$
<b>Flipping:</b> $x[-n] \xleftrightarrow{D.T.F.S.} a_{-k}$
<b>Conjugate:</b> $x^*[n] \xleftrightarrow{D.T.F.S.} a_{-k}^*$ $x^*[-n] \xleftrightarrow{D.T.F.S.} a_k^*$
<b>Symmetries:</b> if $x[n]$ is real : $a_k = a_{-k}^*$ , $ a_k  =  a_{-k} $ , $\angle a_k = -\angle a_{-k}$ $x_e[n] = (x[n] + x^*[-n]) / 2 \xleftrightarrow{DTFS} \text{Re}\{a_k\}$ $x_o[n] = (x[n] - x^*[-n]) / 2 \xleftrightarrow{DTFS} j \text{Im}\{a_k\}$
<b>Periodic convolution:</b> $\sum_{m=\langle N \rangle} x[m] y[n - m] \xleftrightarrow{D.T.F.S.} N a_k b_k$
<b>Modulation:</b> $x[n] y[n] \xleftrightarrow{D.T.F.S.} \sum_{l=\langle N \rangle} a_l b_{k-l}$ $e^{jm \frac{2\pi}{N} n} x[n] \xleftrightarrow{D.T.F.S.} a_{k-m}$
<b>Parseval:</b> $\frac{1}{N} \sum_{n=\langle N \rangle}  x[n] ^2 = \sum_{k=\langle N \rangle}  a_k ^2$
<b>Duality :</b> if $x[n] \xleftrightarrow{DTFS} a_k$ then $a[n] \xleftrightarrow{DTFS} \frac{1}{N} x_{-k}$

## Table of discrete time Fourier series (D.T.F.S.)

$x[n]$ periodic, period $N$ samples	Fourier series coefficients $a_k$ (periodic with period $N$ )
$e^{j\omega_0 n}$	if $e^{j\omega_0 n}$ periodic with $\omega_0 = \frac{2\pi m}{N}$ : $a_k = 1 \quad k = m$ $a_k = 0$ elsewhere (except that $a_k$ is also periodic $N$ )
$\cos(\omega_0 n)$	if $\cos(\omega_0 n)$ periodic $\omega_0 = \frac{2\pi m}{N}$ : $a_k = 1/2 \quad k = m, -m$ $a_k = 0$ elsewhere (except that $a_k$ is also periodic $N$ )
$\sin(\omega_0 n)$	if $\sin(\omega_0 n)$ periodic $\omega_0 = \frac{2\pi m}{N}$ : $a_k = 1/(2j) \quad k = m$ $a_k = -1/(2j) \quad k = -m$ $a_k = 0$ elsewhere (except that $a_k$ is also periodic $N$ )
$\begin{cases} 1 &  n  \leq N_1 \\ 0 & N_1 <  n  \leq N/2 \end{cases}$ (periodic $N$ , $N$ even)	$a_k = \frac{\sin\left(\frac{2\pi}{N} k(N_1 + 1/2)\right)}{N \sin\left(\frac{\pi}{N} k\right)}$ $-N/2 \leq k \leq N/2$ , except for $k = 0$ $a_k = (2N_1 + 1)/N \quad k = 0$ (but $a_k$ is also periodic $N$ )
1	$a_k = 1 \quad k = 0$ $a_k = 0$ elsewhere (except that $a_k$ is also periodic $N$ , and for any $N$ value chosen!)
$\sum_{m=-\infty}^{\infty} \delta[n - mN]$	$a_k = \frac{1}{N} \quad \forall k$

## Properties – Continuous time Fourier transform (C.T.F.T.)

<b>Definitions:</b> $X(j\omega) = \int_{-\infty}^{+\infty} x(t)e^{-j\omega t} dt \quad x(t) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(j\omega)e^{j\omega t} d\omega$ $\omega$ in rad./sec. $X(j\omega) = \sum_{k=-\infty}^{+\infty} 2\pi a_k \delta(\omega - k\omega_0) \text{ if } x(t) \text{ periodic}$
<b>Linearity:</b> $ax(t) + by(t) \xrightarrow{CTFT} aX(j\omega) + bY(j\omega)$
<b>Shifting:</b> $x(t - t_0) \xrightarrow{CTFT} e^{-j\omega t_0} X(j\omega)$
<b>Scaling:</b> $x(at) \xrightarrow{CTFT} \frac{1}{ a } X\left(\frac{j\omega}{a}\right)$
<b>Flipping:</b> $x(-t) \xrightarrow{CTFT} X(-j\omega)$
<b>Conjugate:</b> $x^*(t) \xrightarrow{CTFT} X^*(-j\omega)$ $x^*(-t) \xrightarrow{CTFT} X^*(j\omega)$
<b>Symmetries:</b> if $x(t)$ is real-valued: $X(j\omega) = X^*(-j\omega)$ , $ X(j\omega)  =  X(-j\omega) $ , $\angle X(j\omega) = -\angle X(-j\omega)$ $x_e(t) = (x(t) + x^*(-t)) / 2 \xrightarrow{CTFT} \text{Re}\{X(j\omega)\}$ $x_o(t) = (x(t) - x^*(-t)) / 2 \xrightarrow{CTFT} j \text{Im}\{X(j\omega)\}$
<b>Convolution:</b> $x(t) * y(t) = \int_{-\infty}^{\infty} x(\tau)y(t-\tau)d\tau \leftrightarrow X(j\omega)Y(j\omega)$
<b>Modulation:</b> $x(t)y(t) \xrightarrow{CTFT} \frac{1}{2\pi} X(j\omega) * Y(j\omega) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\theta)Y(j(\omega-\theta))d\theta$ $e^{j\omega_0 t} x(t) \xrightarrow{CTFT} X(j(\omega - \omega_0))$ $\cos(\omega_0 t)x(t) \xrightarrow{CTFT} \frac{1}{2} X(j(\omega - \omega_0)) + \frac{1}{2} X(j(\omega + \omega_0))$
<b>Differentiation:</b> $\frac{dx(t)}{dt} \xrightarrow{CTFT} j\omega X(j\omega)$
<b>Integration:</b> $\int_{-\infty}^t x(\tau)d\tau \xrightarrow{CTFT} \frac{1}{j\omega} X(j\omega) + \pi X(j0)\delta(\omega)$
<b>Differentiation in freq.:</b> $tx(t) \xrightarrow{CTFT} j \frac{dX(j\omega)}{d\omega}$
<b>Integration in freq.:</b> $-\frac{1}{jt} x(t) + \pi x(0)\delta(t) \xrightarrow{CTFT} \int_{-\infty}^{\omega} X(j\eta)d\eta$
<b>Parseval:</b> $\int_{-\infty}^{+\infty}  x(t) ^2 dt = \frac{1}{2\pi} \int_{-\infty}^{+\infty}  X(j\omega) ^2 d\omega$
<b>Duality :</b> if $x(t) \xrightarrow{CTFT} X(j\omega)$ then $X(t) \xrightarrow{CTFT} 2\pi x(-j\omega)$

## Table of continuous time Fourier transforms (C.T.F.T.)

signal $x(t)$ typ. aperiodic	$X(j\omega)$ ( $\omega$ in rad./sec.)
if $x(t)$ is periodic, with period $T = \frac{2\pi}{\omega_0} = \frac{1}{f_0}$ sec.	$2\pi \sum_{k=-\infty}^{+\infty} a_k \delta(\omega - k\omega_0)$
$e^{j\omega_0 t}$	$2\pi \delta(\omega - \omega_0)$
$\cos(\omega_0 t)$	$\pi \delta(\omega - \omega_0) + \pi \delta(\omega + \omega_0)$
$\sin(\omega_0 t)$	$\frac{\pi}{j} \delta(\omega - \omega_0) - \frac{\pi}{j} \delta(\omega + \omega_0)$
$\sum_{n=-\infty}^{\infty} \delta(t - nT)$	$\omega_s \sum_{k=-\infty}^{+\infty} \delta(\omega - k\omega_s) \quad \omega_s = \frac{2\pi}{T}$
1	$2\pi \delta(\omega)$
$\begin{cases} 1 &  t  < T_1 \\ 0 &  t  > T_1 \end{cases}$	$\frac{2\sin(\omega T_1)}{\omega}$
$\frac{\sin(\omega_c t)}{\pi t} \quad \omega_c > 0$	$\begin{cases} 1 &  \omega  \leq \omega_c \\ 0 &  \omega  > \omega_c \end{cases}$
$\delta(t)$	1
$u(t)$	$\frac{1}{j\omega} + \pi \delta(\omega)$
$e^{-at} u(t) \quad \text{Re}\{a\} > 0$	$\frac{1}{j\omega + a}$
$-e^{-at} u(-t) \quad \text{Re}\{a\} < 0$	$\frac{1}{j\omega + a}$
$\frac{t^{r-1}}{(r-1)!} e^{-at} u(t) \quad \text{Re}\{a\} > 0$	$\frac{1}{(j\omega + a)^r} \quad r \geq 1$
$-\frac{t^{r-1}}{(r-1)!} e^{-at} u(-t) \quad \text{Re}\{a\} < 0$	$\frac{1}{(j\omega + a)^r} \quad r \geq 1$
$\frac{1}{\text{Im}\{a\}} e^{-\text{Re}\{a\}t} \sin(\text{Im}\{a\}t) u(t) \quad \text{Re}\{a\} > 0, \text{Im}\{a\} \neq 0$ or $-\frac{1}{\text{Im}\{a\}} e^{-\text{Re}\{a\}t} \sin(\text{Im}\{a\}t) u(-t) \quad \text{Re}\{a\} < 0, \text{Im}\{a\} \neq 0$	$\frac{1}{(j\omega + a)(j\omega + a^*)}$ $= \frac{1}{(j\omega)^2 + 2\text{Re}\{a\}(j\omega) +  a ^2}$
$2 A e^{-\text{Re}\{a\}t} \times \cos(\text{Im}\{a\}t - \angle A) u(t) \quad \text{Re}\{a\} > 0$	$\frac{A}{j\omega + a} + \frac{A^*}{j\omega + a^*}$
$-2 A e^{-\text{Re}\{a\}t} \times \cos(\text{Im}\{a\}t - \angle A) u(-t) \quad \text{Re}\{a\} < 0$	$\frac{A}{j\omega + a} + \frac{A^*}{j\omega + a^*}$

**Properties – Discrete time Fourier transform (D.T.F.T.)**

<b>Definitions:</b> $x[n] = x(nT) = x(t) _{t=nT}$ , where $T = 1/f_s = 2\pi/\omega_s$ $X(e^{j\omega})$ is the DTFT of $x[n]$ defined as: $X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n} \quad x[n] = \frac{1}{2\pi} \int_{-\pi}^{+\pi} X(e^{j\omega})e^{j\omega n} d\omega$ Relative to $x(t)$ and its CTFT $X(j\omega)$ , the relation is: $X_p(j\omega) = f_s \sum_{k=-\infty}^{\infty} X(j(\omega - k\omega_s)), \quad X(e^{j\omega}) = X_p(j\omega f_s)$
<b>Periodicity:</b> $x[n] \xleftrightarrow{DTFT} X(e^{j\omega}) = X(e^{j(\omega+2\pi)})$
<b>Linearity:</b> $ax[n] + by[n] \xleftrightarrow{DTFT} aX(e^{j\omega}) + bY(e^{j\omega})$
<b>Shifting:</b> $x[n - n_0] \xleftrightarrow{DTFT} e^{-j\omega n_0} X(e^{j\omega}) \quad n_0 \text{ integer}$
<b>Expansion, insertion of zeros (“upsampling”):</b> $x_{(k)}[n] = x[n/k] \quad \text{if } n \text{ is a multiple of } k$ $x_{(k)}[n] = 0 \quad \text{elsewhere}$ $x_{(k)}[n] \xleftrightarrow{DTFT} X(e^{jk\omega}) \quad \text{where } k \text{ is a positive integer}$
<b>Discrete time sampling (“downsampling”):</b> $x_d[n] = x[Mn] \xleftrightarrow{DTFT} X_d(e^{j\omega}) = \frac{1}{M} \sum_{m=0}^{M-1} X\left(e^{j\left(\frac{\omega}{M} - m\frac{2\pi}{M}\right)}\right)$
<b>Flipping:</b> $x[-n] \xleftrightarrow{DTFT} X(e^{-j\omega})$
<b>Conjugate:</b> $x^*[n] \xleftrightarrow{DTFT} X^*(e^{-j\omega})$ $x^*[-n] \xleftrightarrow{DTFT} X^*(e^{j\omega})$
<b>Symmetries:</b> if $x[n]$ is real : $X(e^{j\omega}) = X^*(e^{-j\omega})$ , $ X(e^{j\omega})  =  X(e^{-j\omega}) , \angle X(e^{j\omega}) = -\angle X(e^{-j\omega})$ $x_e[n] = (x[n] + x^*[-n]) / 2 \xleftrightarrow{DTFT} \text{Re}\{X(e^{j\omega})\}$ $x_o[n] = (x[n] - x^*[-n]) / 2 \xleftrightarrow{DTFT} j \text{Im}\{X(e^{j\omega})\}$
<b>Convolution:</b> $x[n] * y[n] = \sum_{k=-\infty}^{\infty} x[k]y[n-k] \xleftrightarrow{DTFT} X(e^{j\omega})Y(e^{j\omega})$
<b>Modulation:</b> $x[n]y[n] \xleftrightarrow{DTFT} \frac{1}{2\pi} \int_{2\pi} X(e^{j\theta})Y(e^{j(\omega-\theta)})d\theta$ $e^{j\omega_0 n} x[n] \xleftrightarrow{DTFT} X(e^{j(\omega-\omega_0)})$
<b>Accumulation:</b> $\sum_{m=-\infty}^n x[m] \xleftrightarrow{DTFT} \frac{1}{1-e^{-j\omega}} X(e^{j\omega}) + \pi X(e^{j0}) \sum_{m=-\infty}^{+\infty} \delta(\omega - m2\pi)$
<b>Differentiation in freq.:</b> $nx[n] \xleftrightarrow{DTFT} j \frac{dX(e^{j\omega})}{d\omega}$
<b>Parseval:</b> $\sum_{n=-\infty}^{+\infty}  x[n] ^2 = \frac{1}{2\pi} \int_{2\pi}  X(e^{j\omega}) ^2 d\omega$
<b>Duality :</b> If $x[n] \xleftrightarrow{DTFT} X(e^{j\omega})$ then $X(t) \xleftrightarrow{CTFS} x_{-k}$

**Table of discrete time Fourier transforms (D.T.F.T.)**

signal $x[n]$ typ. aperiodic	$X(e^{j\omega})$ (periodic $2\pi$ , $\omega$ in rad./sample)
if $x[n]$ is periodic, with period $N$ samples	$2\pi \sum_{k=-\infty}^{\infty} a_k \delta(\omega - k\frac{2\pi}{N})$ (also periodic $2\pi$ )
$e^{j\omega_0 n}$	$2\pi\delta(\omega - \omega_0)$ (also periodic $2\pi$ )
$\cos(\omega_0 n)$	$\pi\delta(\omega - \omega_0) + \pi\delta(\omega + \omega_0)$ (also periodic $2\pi$ )
$\sin(\omega_0 n)$	$\frac{\pi}{j}\delta(\omega - \omega_0) - \frac{\pi}{j}\delta(\omega + \omega_0)$ (also periodic $2\pi$ )
$\sum_{m=-\infty}^{\infty} \delta[n - mN]$	$\frac{2\pi}{N} \sum_{k=-\infty}^{\infty} \delta(\omega - k\frac{2\pi}{N})$ (also periodic $2\pi$ )
1	$2\pi\delta(\omega)$ (also periodic $2\pi$ )
$\begin{cases} 1 &  n  \leq N_1 \\ 0 &  n  > N_1 \end{cases}$	$\sin(\omega(N_1 + 1/2)) / \sin(\omega/2)$
$\frac{\sin(\omega_c n)}{\pi n} \quad 0 < \omega_c < \pi$	$\begin{cases} 1 & 0 \leq  \omega  \leq \omega_c \\ 0 & \omega_c <  \omega  \leq \pi \end{cases}$ (also periodic $2\pi$ )
$\delta[n]$	1
$u[n]$	$\frac{1}{1 - e^{-j\omega}} + \pi\delta(\omega)$ (also periodic $2\pi$ )
$a^n u[n] \quad  a  < 1$	$1/(1 - ae^{-j\omega})$
$-a^n u[-n-1] \quad  a  > 1$	$1/(1 - ae^{-j\omega})$
$\frac{(n+r-1)!}{n!(r-1)!} a^n u[n] \quad  a  < 1$	$1/(1 - ae^{-j\omega})^r \quad r \geq 1$
$-\frac{(n+r-1)!}{n!(r-1)!} a^n u[-n-1] \quad  a  > 1$	$1/(1 - ae^{-j\omega})^r \quad r \geq 1$
$\frac{r^n}{\sin(\theta)} \sin(\theta(n+1))u[n] \quad 0 \leq r < 1 \quad 0 < \theta < \pi$ or $-\frac{r^n}{\sin(\theta)} \sin(\theta(n+1))u[-n-1] \quad r > 1 \quad 0 < \theta < \pi$	$\frac{1}{(1 - re^{j\theta}e^{-j\omega})(1 - re^{-j\theta}e^{-j\omega})}$ $= \frac{1}{1 - 2r \cos \theta e^{-j\omega} + r^2 e^{-j2\omega}}$
$2 A r^n \cos(\theta n + \angle A)u[n] \quad 0 \leq r < 1 \quad 0 \leq \theta \leq \pi$	$\frac{A}{1 - re^{j\theta}e^{-j\omega}} + \frac{A^*}{1 - re^{-j\theta}e^{-j\omega}}$
$-2 A r^n \cos(\theta n + \angle A)u[-n-1] \quad r > 1 \quad 0 \leq \theta \leq \pi$	$\frac{A}{1 - re^{j\theta}e^{-j\omega}} + \frac{A^*}{1 - re^{-j\theta}e^{-j\omega}}$

## Properties – bilateral (two-sided) Laplace transform

<b>Definitions:</b> $X(s) = \int_{-\infty}^{+\infty} x(t)e^{-st} dt$ $ROC_x$ $\sigma_l < \text{Re}\{s\} < \sigma_r$ $x(t) = \frac{1}{2\pi j} \int_{\sigma-j\infty}^{\sigma+j\infty} X(s)e^{st} ds = \frac{e^{\sigma t}}{2\pi} \int_{\omega=-\infty}^{\omega=\infty} X(\sigma + j\omega)e^{j\omega t} d\omega$ $\sigma \in ROC_x$ $(\sigma=0) \in ROC_x$ $x(t) \xleftrightarrow{LT} X(s)$ $ROC_x$ $y(t) \xleftrightarrow{LT} Y(s)$ $ROC_y$
<b>Linearity:</b> $ax(t) + by(t) \xleftrightarrow{LT} aX(s) + bY(s)$ $ROC_x \cap ROC_y$
<b>Shifting:</b> $x(t-t_0) \xleftrightarrow{LT} e^{-st_0} X(s)$ $ROC_x$ unchanged
<b>Scaling:</b> $x(at) \xleftrightarrow{LT} \frac{1}{ a } X\left(\frac{s}{a}\right)$ $a$ is real valued $ a ROC_x, s_k \rightarrow as_k, a\sigma_l < \text{Re}\{s\} < a\sigma_r$ $a > 0$ $a\sigma_r < \text{Re}\{s\} < a\sigma_l$ $a < 0$
<b>Flipping:</b> $x(-t) \xleftrightarrow{LT} X(-s)$ $ROC_x$ inversed
<b>Conjugate:</b> $x^*(t) \xleftrightarrow{LT} X^*(s^*)$ $ROC_x$ unchanged, $s_k \rightarrow s_k^*$
<b>Symmetry:</b> if $x(t)$ real: $X(s) = X^*(s^*)$ , $ X(s)  =  X(s^*) $ and if $s_k$ is a zero or a pole then $s_k^*$ is also a zero or a pole.
<b>Convolution:</b> $x(t) * y(t) = \int_{-\infty}^{\infty} x(\tau)y(t-\tau)d\tau \xleftrightarrow{LT} X(s)Y(s)$ $ROC_x \cap ROC_y$
<b>Modulation:</b> $e^{s_0 t} x(t) \xleftrightarrow{LT} X(s-s_0)$ $ROC_x$ shifted right by $\text{Re}\{s_0\}$ $s_k \rightarrow s_k + s_0, \sigma_l + \text{Re}\{s_0\} < \text{Re}\{s\} < \sigma_r + \text{Re}\{s_0\}$
<b>Differentiation:</b> $\frac{dx(t)}{dt} \xleftrightarrow{LT} sX(s)$ $ROC_x$ unchanged
<b>Integration:</b> $\int_{-\infty}^t x(\tau)d\tau \xleftrightarrow{LT} \frac{1}{s} X(s)$ $ROC_x \cap (\text{Re}\{s\} > 0)$
<b>Differentiation in freq.:</b> $-tx(t) \xleftrightarrow{LT} \frac{dX(s)}{ds}$ $ROC_x$ unchanged
<b>Initial value theorem:</b> if $x(t) = 0$ $t < 0$ with no impulse $\delta(t)$ or singularity at $t = 0$ : $x(0^+) = \lim_{s \rightarrow \infty} sX(s)$
<b>Final value theorem:</b> if $x(t) = 0$ $t < 0$ has a finite value at $t \rightarrow \infty$ : $\lim_{t \rightarrow \infty} x(t) = \lim_{s \rightarrow 0} sX(s)$

## Table of bilateral (two-sided) Laplace transforms

Signal $x(t)$	Laplace transform $X(s)$	ROC
$\delta(t)$	1	$\forall s$
$u(t)$	$\frac{1}{s}$	$\text{Re}\{s\} > 0$
$-u(-t)$	$\frac{1}{s}$	$\text{Re}\{s\} < 0$
$\begin{cases} 1 &  t  < T_1 \\ 0 &  t  > T_1 \end{cases}$	$\frac{e^{T_1 s} - e^{-T_1 s}}{s}$	$\forall s$
$e^{-at}u(t)$	$1/(s+a)$	$\text{Re}\{s\} > \text{Re}\{-a\}$
$-e^{-at}u(-t)$	$1/(s+a)$	$\text{Re}\{s\} < \text{Re}\{-a\}$
$\frac{t^{r-1}}{(r-1)!} e^{-at}u(t)$	$1/(s+a)^r$ $r \geq 1$	$\text{Re}\{s\} > \text{Re}\{-a\}$
$-\frac{t^{r-1}}{(r-1)!} e^{-at}u(-t)$	$1/(s+a)^r$ $r \geq 1$	$\text{Re}\{s\} < \text{Re}\{-a\}$
$\frac{1}{\text{Im}\{a\}} e^{-\text{Re}\{a\}t} \sin(\text{Im}\{a\}t)u(t)$ $\text{Im}\{a\} \neq 0$	$\frac{1}{(s+a)(s+a^*)}$ $= \frac{1}{s^2 + 2\text{Re}\{a\}s +  a ^2}$	$\text{Re}\{s\} > \text{Re}\{-a\}$
$\frac{-1}{\text{Im}\{a\}} e^{-\text{Re}\{a\}t} \sin(\text{Im}\{a\}t)u(-t)$ $\text{Im}\{a\} \neq 0$	$\frac{1}{(s+a)(s+a^*)}$ $= \frac{1}{s^2 + 2\text{Re}\{a\}s +  a ^2}$	$\text{Re}\{s\} < \text{Re}\{-a\}$
$2 A e^{-\text{Re}\{a\}t} \times \cos(\text{Im}\{a\}t - \angle A)u(t)$	$\frac{A}{s+a} + \frac{A^*}{s+a^*}$	$\text{Re}\{s\} > \text{Re}\{-a\}$
$-2 A e^{-\text{Re}\{a\}t} \times \cos(\text{Im}\{a\}t - \angle A)u(-t)$	$\frac{A}{s+a} + \frac{A^*}{s+a^*}$	$\text{Re}\{s\} < \text{Re}\{-a\}$

## A few properties – unilateral (one-sided) Laplace transform

<b>Definitions:</b> $X(s) = \int_0^{+\infty} x(t)e^{st} dt$ $x(t) = \frac{1}{2\pi j} \int_{\sigma-j\infty}^{\sigma+j\infty} X(s)e^{st} ds$ $x(t) \xleftrightarrow{ULT} X(s)$ $y(t) \xleftrightarrow{ULT} Y(s)$
<b>Linearity:</b> $ax(t) + by(t) \xleftrightarrow{ULT} aX(s) + bY(s)$
<b>Differentiation:</b> $\frac{dx(t)}{dt} \xleftrightarrow{ULT} sX(s) - x(0^-)$ $\frac{d^k x(t)}{dt^k} \xleftrightarrow{ULT} s^k X(s) - s^{k-1}x(0^-) - \dots - s^0 \frac{dx^{k-1}(t)}{dt^{k-1}} \Big _{t=0^-}$
<b>Integration:</b> $\int_0^t x(\tau)d\tau \xleftrightarrow{ULT} \frac{1}{s} X(s)$

## Properties – Bilateral (two-sided) z-transform

<b>Definitions:</b> $X(z) = \sum_{n=-\infty}^{\infty} x[n]z^{-n}$ $ROC_x$ $r_i <  z  < r_o$ $x[n] = \frac{1}{2\pi j} \oint_{ROC_x} X(z)z^{n-1}dz = \frac{r^n}{2\pi} \int_{\omega=-\pi}^{\pi} X(re^{j\omega})e^{j\omega n}d\omega$ $z \in ROC_x \quad (r=1) \in ROC_x$
<b>Linearity:</b> $ax[n] + by[n] \xleftrightarrow{z.t.} aX(z) + bY(z)$ $ROC_x \cap ROC_y$
<b>Shifting:</b> $x[n - n_0] \xleftrightarrow{z.t.} z^{-n_0} X(z)$ $ROC_x$ unchanged
<b>Conjugate:</b> $x^*[n] \xleftrightarrow{z.t.} X^*(z^*)$ $ROC_x$ unchanged, $z_k \rightarrow z_k^*$
<b>Symmetry:</b> if $x[n]$ real: $X(z) = X^*(z^*)$ , $ X(z)  =  X(z^*) $ if $z_k$ is a zero or a pole then $z_k^*$ is also a zero or a pole
<b>Flipping:</b> $x[-n] \xleftrightarrow{z.t.} X(z^{-1})$ $1/ROC_x$ , $z_k \rightarrow 1/z_k$ , $1/r_o <  z  < 1/r_i$
<b>Upsampling, expansion, insertion of zeros:</b> $x_{(M)}[n] = x[n/M]$ if $n$ is a multiple of $M$ , else $x_{(M)}[n] = 0$ $x_{(M)}[n] \xleftrightarrow{z.t.} X(z^M)$ where $M$ is a positive integer $ROC_x^{1/M}$ , $z_k \rightarrow z_k^{1/M}$ , $r_i^{1/M} <  z  < r_o^{1/M}$
<b>Downsampling:</b> $x_p[n] = x[n]$ if $n$ is a multiple of $M$ , else $x_p[n] = 0$ $x_p[n] \xleftrightarrow{z.t.} X_p(z) = \frac{1}{M} \sum_{m=0}^{M-1} X(e^{-jk\frac{2\pi}{M}}z)$ , $M$ positive integer $ROC_x$ unchanged, $z_i \rightarrow e^{jk\frac{2\pi}{M}}z_i$ $x_d[n] = x_p[Mn] = x[Mn] \xleftrightarrow{z.t.} X_d(z) = \frac{1}{M} \sum_{m=0}^{M-1} X(e^{-jk\frac{2\pi}{M}}z^{1/M})$ $ROC_x^M$ , $z_k \rightarrow z_k^M$ , $r_i^M <  z  < r_o^M$
<b>Convolution:</b> $\sum_{k=-\infty}^{\infty} x[k]y[n-k] \xleftrightarrow{z.t.} X(z)Y(z)$ $ROC_x \cap ROC_y$
<b>Modulation:</b> $a^n x[n] \xleftrightarrow{z.t.} X\left(\frac{z}{a}\right)$ $ a ROC_x$ , $z_k \rightarrow az_k$ , $ a r_i <  z  <  a r_o$ $e^{j\omega_0 n} x[n] \xleftrightarrow{z.t.} X(e^{-j\omega_0}z)$ $ROC_x$ unchanged
<b>Accumulation:</b> $\sum_{m=-\infty}^n x[m] \xleftrightarrow{z.t.} \frac{1}{1-z^{-1}} X(z)$ $ROC_x \cap ( z  > 1)$
<b>Differentiation in freq.:</b> $nx[n] \xleftrightarrow{z.t.} -z \frac{dX(z)}{dz}$ same $ROC_x$
<b>Initial value theorem:</b> if $x[n] = 0$ $n < 0$ $x[0] = \lim_{z \rightarrow \infty} X(z)$
<b>Final value theorem:</b> if $x[n] = 0$ $n < 0$ , and poles of $(z-1)X(z)$ inside unit circle: $\lim_{n \rightarrow \infty} x[n] = \lim_{z \rightarrow 1} (z-1)X(z)$

## Table of bilateral (two-sided) z-transforms

Signal $x[n]$	z-transform $X(z)$	ROC
$\delta[n]$	1	$\forall z$
$u[n]$	$1/(1-z^{-1})$	$ z  > 1$
$-u[-n-1]$	$1/(1-z^{-1})$	$ z  < 1$
$a^n u[n]$	$1/(1-az^{-1})$	$ z  >  a $
$-a^n u[-n-1]$	$1/(1-az^{-1})$	$ z  <  a $
$\frac{(n+r-1)!}{n!(r-1)!} a^n u[n]$	$1/(1-az^{-1})^r$ $r \geq 1$	$ z  >  a $
$\frac{-(n+r-1)!}{n!(r-1)!} a^n u[-n-1]$	$1/(1-az^{-1})^r$ $r \geq 1$	$ z  <  a $
$\frac{r^n}{\sin(\theta)} \sin(\theta(n+1))u[n]$ $0 < \theta < \pi$	$\frac{1}{(1-re^{j\theta}z^{-1})(1-re^{-j\theta}z^{-1})}$ $= \frac{1}{1-2r\cos(\theta)z^{-1}+r^2z^{-2}}$	$ z  > r$
$-\frac{r^n}{\sin(\theta)} \sin(\theta(n+1))u[-n-1]$ $0 < \theta < \pi$	$\frac{1}{(1-re^{j\theta}z^{-1})(1-re^{-j\theta}z^{-1})}$ $= \frac{1}{1-2r\cos(\theta)z^{-1}+r^2z^{-2}}$	$ z  < r$
$2 A r^n \cos(\theta n + \angle A)u[n]$ $0 \leq \theta \leq \pi$	$\frac{A}{1-re^{j\theta}z^{-1}} + \frac{A^*}{1-re^{-j\theta}z^{-1}}$ $= \frac{2\operatorname{Re}\{A\} - 2r A \cos(\angle A - \theta)z^{-1}}{1-2r\cos(\theta)z^{-1}+r^2z^{-2}}$	$ z  > r$
$-2 A r^n \cos(\theta n + \angle A)u[-n-1]$ $0 \leq \theta \leq \pi$	$\frac{A}{1-re^{j\theta}z^{-1}} + \frac{A^*}{1-re^{-j\theta}z^{-1}}$ $= \frac{2\operatorname{Re}\{A\} - 2r A \cos(\angle A - \theta)z^{-1}}{1-2r\cos(\theta)z^{-1}+r^2z^{-2}}$	$ z  < r$
$a^n$ $0 \leq n \leq N-1$ 0 elsewhere	$\frac{1-a^N z^{-N}}{1-az^{-1}}$	$ z  > 0$

## A few properties – unilateral (one-sided) z transform

<b>Definitions:</b> $X(z) = \sum_{n=0}^{\infty} x[n]z^{-n}$ $x[n] = \frac{1}{2\pi j} \oint X(z)z^{n-1}dz$
<b>Linearity:</b> $ax[n] + by[n] \xleftrightarrow{u.z.t.} aX(z) + bY(z)$
<b>Shifting:</b> $x[n-1] \xleftrightarrow{u.z.t.} z^{-1}[X(z) + x[-1]z] = z^{-1}X(z) + x[-1]$ $x[n-k] \xleftrightarrow{u.z.t.} z^{-k} \left[ X(z) + \sum_{l=1}^k x[-l]z^l \right]$ $k > 0$ $x[n+1] \xleftrightarrow{u.z.t.} z[X(z) - x[0]] = zX(z) - zx[0]$ $x[n+k] \xleftrightarrow{u.z.t.} z^k \left[ X(z) - \sum_{l=0}^{k-1} x[l]z^{-l} \right]$ $k > 0$