

EBU6018

Advanced Transform Methods

Fourier Transform_1
Fourier Series

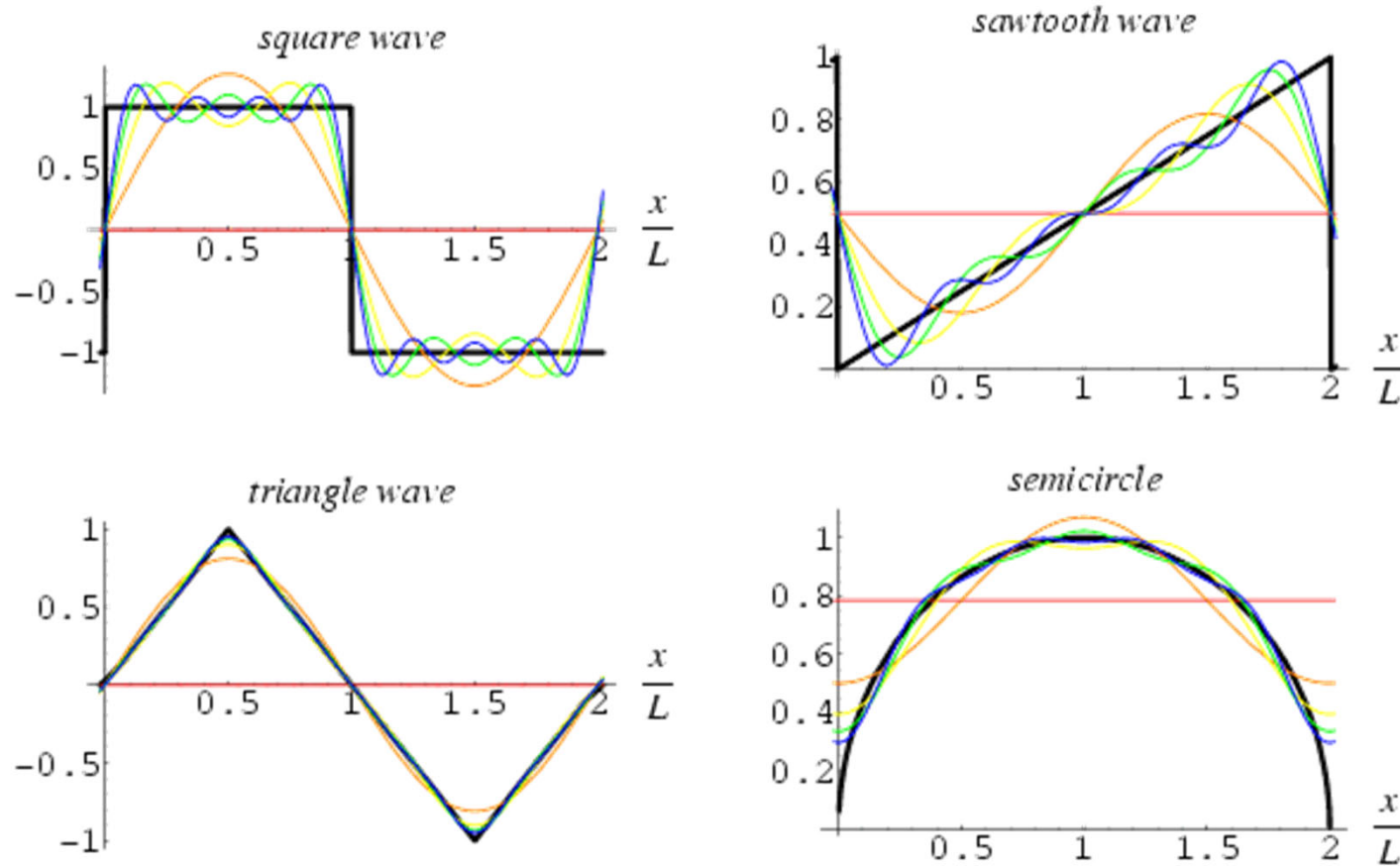
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Fourier Series (FS)

- Periodic signals can be expressed as a sum of sinusoids. The frequency spectrum can be generated by computation of the *Fourier series*.
- The Fourier series is named after the French physicist Jean Baptiste Fourier (1768-1830), who was the first one to propose that periodic waveforms could be represented by a sum of sinusoids (or complex exponentials).
- Obtaining the Fourier Series of a **function in the time domain** means that the same function can be represented in **the frequency domain**.



Fourier Representation



Functions of time as the sum of sinusoids



Trigonometric FS - 1

A periodic signal, $x(t)$, whose period is T , can be represented by the appropriate sum of sin and cos components:

$$x(t) = a_0 + \sum_{n=1}^{\infty} a_n \cos(n\omega t) + \sum_{n=1}^{\infty} b_n \sin(n\omega t) \quad (1)$$

a_0 is the *mean value*, or *zero frequency* term.

Integrating both sides of eqn (1), between $-T/2$ and $T/2$:

$$\int_{-T/2}^{T/2} x(t) dt = \int_{-T/2}^{T/2} a_0 + \int_{-T/2}^{T/2} \left[\sum_{n=1}^{\infty} a_n \cos(n\omega t) + \sum_{n=1}^{\infty} b_n \sin(n\omega t) \right] dt$$

Trigonometric FS - 2

$$\int_{-T/2}^{T/2} x(t) dt = \int_{-T/2}^{T/2} a_0 + \int_{-T/2}^{T/2} \left[\sum_{n=1}^{\infty} a_n \cos(n\omega t) + \sum_{n=1}^{\infty} b_n \sin(n\omega t) \right] dt$$

$$\int_{-T/2}^{T/2} x(t) dt = \int_{-T/2}^{T/2} a_0 dt = a_0.T$$

$$a_0 = 1/T \int_{-T/2}^{T/2} x(t) dt$$



Trigonometric FS - 3

To find a formula for an it is necessary to multiply both sides of eqn(1) by $\cos(m.\omega.t)$ and then integrate over the same limits:

$$\int_{-T/2}^{T/2} x(t) \cos(m.\omega.t) dt = \int_{-T/2}^{T/2} a_0 \cos(m.\omega.t) dt + \int_{-T/2}^{T/2} \left[\sum_{n=1}^{\infty} \cos(m.\omega.t).a_n.\cos(n.\omega.t) + \sum_{n=1}^{\infty} \cos(m.\omega.t) b_n.\sin(n.\omega.t) \right] dt$$

the “cos.cos” terms

the “cos.sin” terms

- Using the appropriate trig identities we see that the cos.sin terms produce $\cos(A).\sin(B) = \frac{1}{2} (\sin(A+B) +/\!-\sin(A-B))$ odd waveforms which all disappear under integration.
- The cos.cos terms produce:
 $\cos(A).\cos(B) = \frac{1}{2} (\cos(A+B) +/\!-\cos(A-B))$
 which will not necessarily disappear under integration:

Trigonometric FS - 4

$$\int_{-T/2}^{T/2} \sum_{n=1}^{\infty} \frac{\cos(m.\omega.t) \cdot a_n \cdot \cos(n.\omega.t)}{a_n \cdot \frac{1}{2} (\cos((m+n).\omega.t) + \cos((m-n).\omega.t))} dt$$

HOWEVER, we are integrating over $-T/2 \rightarrow +T/2$ and this represents an integer number of cycles of the sinusoid, whatever the value of 'm' and 'n'.
BUT when $m=n$, we have a non-zero term after integration:

$$\begin{aligned} \int_{-T/2}^{T/2} x(t) \cdot \cos(m.\omega.t) dt &= \int_{-T/2}^{T/2} a_0 \cdot \cos(m.\omega.t) dt + \int_{-T/2}^{T/2} a_n \cdot \frac{1}{2} \cos((0).\omega.t) dt \\ &+ \int_{-T/2}^{T/2} \left[\sum_{n=1}^{\infty} \cos(m.\omega.t) \cdot a_n \cdot \cos(n.\omega.t) + \sum_{n=1}^{\infty} \cos(m.\omega.t) \cdot b_n \cdot \sin(n.\omega.t) \right] dt \\ \int_{-T/2}^{T/2} x(t) \cos(m.\omega.t) dt &= (a_n/2) \left| t \right|_{-T/2}^{T/2} = a_n \cdot T/2 \end{aligned}$$



Trigonometric FS - 5

BUT $m=n$, so:

$$\int_{-T/2}^{T/2} x(t) \cos(n\omega t) dt = a_n/2 \left| t \right|_{-T/2}^{T/2} = a_n \cdot T/2$$

$$a_n = 2/T \int_{-T/2}^{T/2} x(t) \cos(n\omega t) dt$$

And by similar reasoning:

$$b_n = 2/T \int_{-T/2}^{T/2} x(t) \sin(n\omega t) dt$$



Trigonometric Fourier Series – Cosine-with-phase form

The trigonometric Fourier series given by equ (1) can also be written in the cosine-with-phase form:

$$x(t) = a_0 + \sum_{n=1}^{\infty} A_n \cos(n\omega_0 t + \theta_n) \quad -\infty < t < \infty$$

$$A_n = \sqrt{a_n^2 + b_n^2}, n=1,2,\dots$$

$$\theta_n = \begin{cases} \tan^{-1}\left(-\frac{b_n}{a_n}\right), & n=1,2,\dots, \text{when } a_n \geq 0 \\ \pi + \tan^{-1}\left(-\frac{b_n}{a_n}\right), & n=1,2,\dots, \text{when } a_n < 0 \end{cases}$$



Trigonometric Fourier Series – Dirichlet conditions

Fourier believed that any periodic signal could be expressed as a sum of sinusoids. However, this turned out not to be the case, although virtually all periodic signals arising in engineering do have a Fourier series representation. In particular, a periodic signal $x(t)$ has a Fourier series if it satisfies the following *Dirichlet conditions*:

1. $x(t)$ is absolutely integrable over any period; that is

$$\int_a^{a+T} |x(t)| dt < \infty \quad \text{for any } a$$

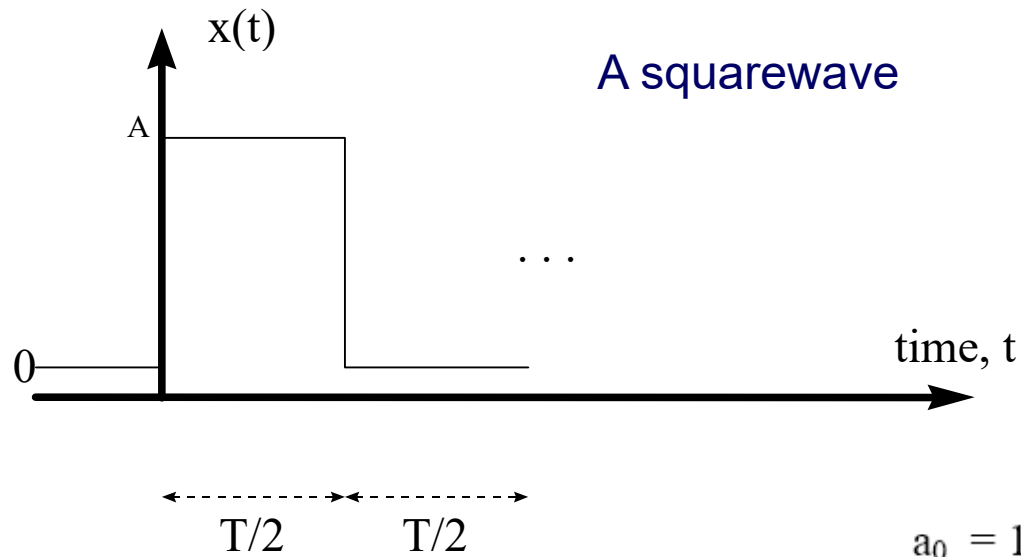
2. $x(t)$ has only a finite number of maxima and minima over any period.

3. $x(t)$ has only a finite number of discontinuities over any period.



Application of the FS

Example 1: An ODD function



$$\begin{aligned} a_0 &= \frac{1}{T} \int_{-T/2}^{T/2} x(t) dt \\ &= \frac{1}{T} \int_0^{T/2} A dt + \int_{T/2}^T 0 dt \\ &= \frac{1}{T} [A \cdot t]_0^{T/2} \\ &= A/2 \end{aligned}$$

Application of the FS Example 1

$$a_n = 2/T \int_{-T/2}^{T/2} x(t) \cdot \cos(n \cdot \omega \cdot t) dt = 2/T \int_0^{T/2} A \cdot \cos(n \cdot \omega \cdot t) dt + \int_{T/2}^T 0 dt$$

$$= 2A/T \left| \sin(n \cdot \omega \cdot t) / (n \cdot \omega) \right|_0^{T/2}$$

$$= A/n\pi [\sin(n \cdot \pi)] = 0$$

$$b_n = 2/T \int_{-T/2}^{T/2} x(t) \cdot \sin(n \cdot \omega \cdot t) dt$$

$$= 2/T \int_0^{T/2} A \cdot \sin(n \cdot \omega \cdot t) dt + \int_{T/2}^T 0 dt$$

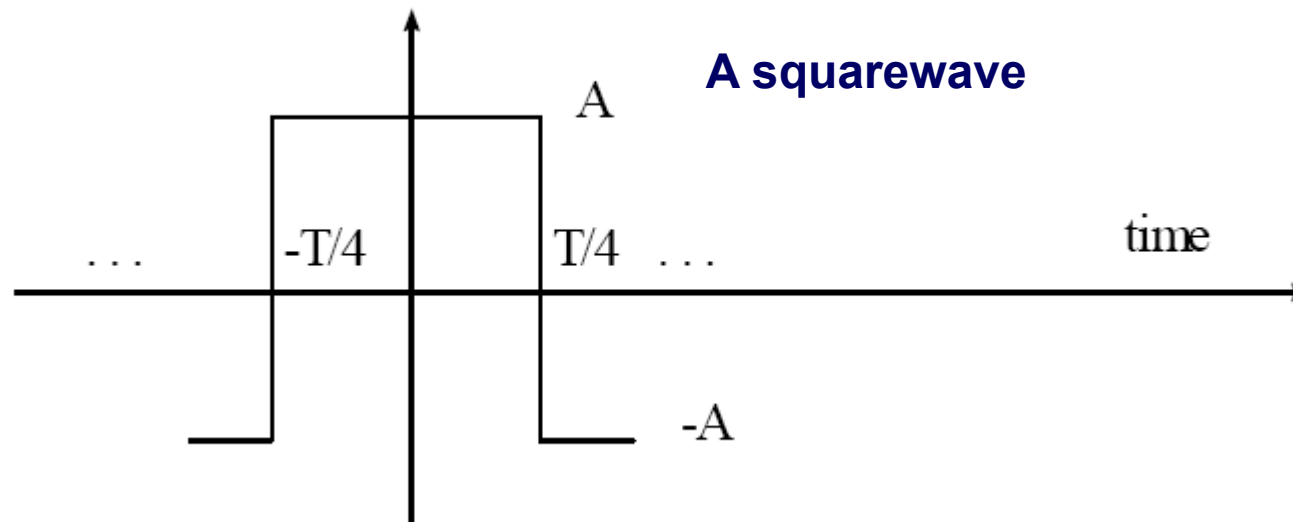
$$= 2A/T \left| -\cos(n \cdot \omega \cdot t) / (n \cdot \omega) \right|_0^{T/2}$$

$$= A/n\pi [1 - \cos(n \cdot \pi)]$$



Application of the FS

Example 2: An EVEN Function



$a_0 = 0$ by inspection

$$a_n = \frac{2}{T} \int_{-T/4}^{3T/4} x(t) \cdot \cos(n \cdot \omega \cdot t) dt$$

Application of the FS

Example 2

$$= 2/T \int_{-T/4}^{T/4} A \cdot \cos(n\omega t) dt + \int_{T/4}^{3T/4} -A \cdot \cos(n\omega t) dt$$

$$= 2A/T \left[\sin(n\omega t) / n\omega \right]_{-T/4}^{T/4} - 2A/T \left[\sin(n\omega t) / n\omega \right]_{T/4}^{3T/4}$$

$$= 2A/nT\omega \left[\sin(n\omega T/4) - \sin(n\omega(-T)/4) - \sin(3n\omega T/4) + \sin(n\omega T/4) \right]$$

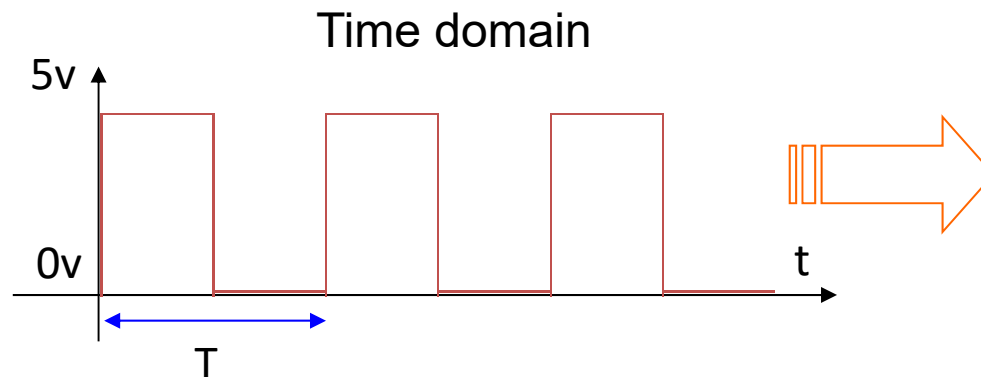
but $\omega T = (2\pi f) \cdot (1/f) = 2\pi$, $\sin(-A) = -\sin(A)$ and $\sin(3n2\pi/4) = -\sin(n\pi/2)$ therefore:

$$= 2A/n2\pi \left[\sin(n2\pi/4) - \sin(-n2\pi/4) - \sin(3n2\pi/4) + \sin(n2\pi/4) \right]$$

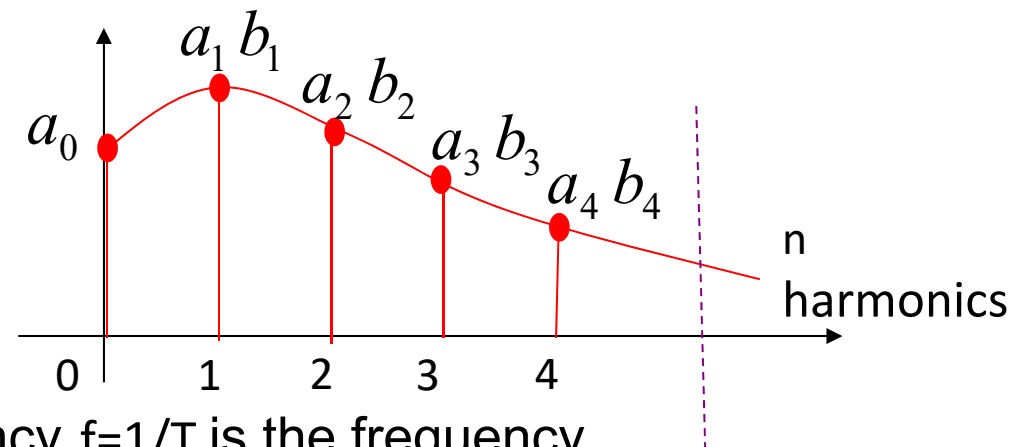
$$= 4A/n\pi \left[\sin(n\pi/2) \right]$$

$b_n = 0$ by inspection

Example: transform the signal to line spectra

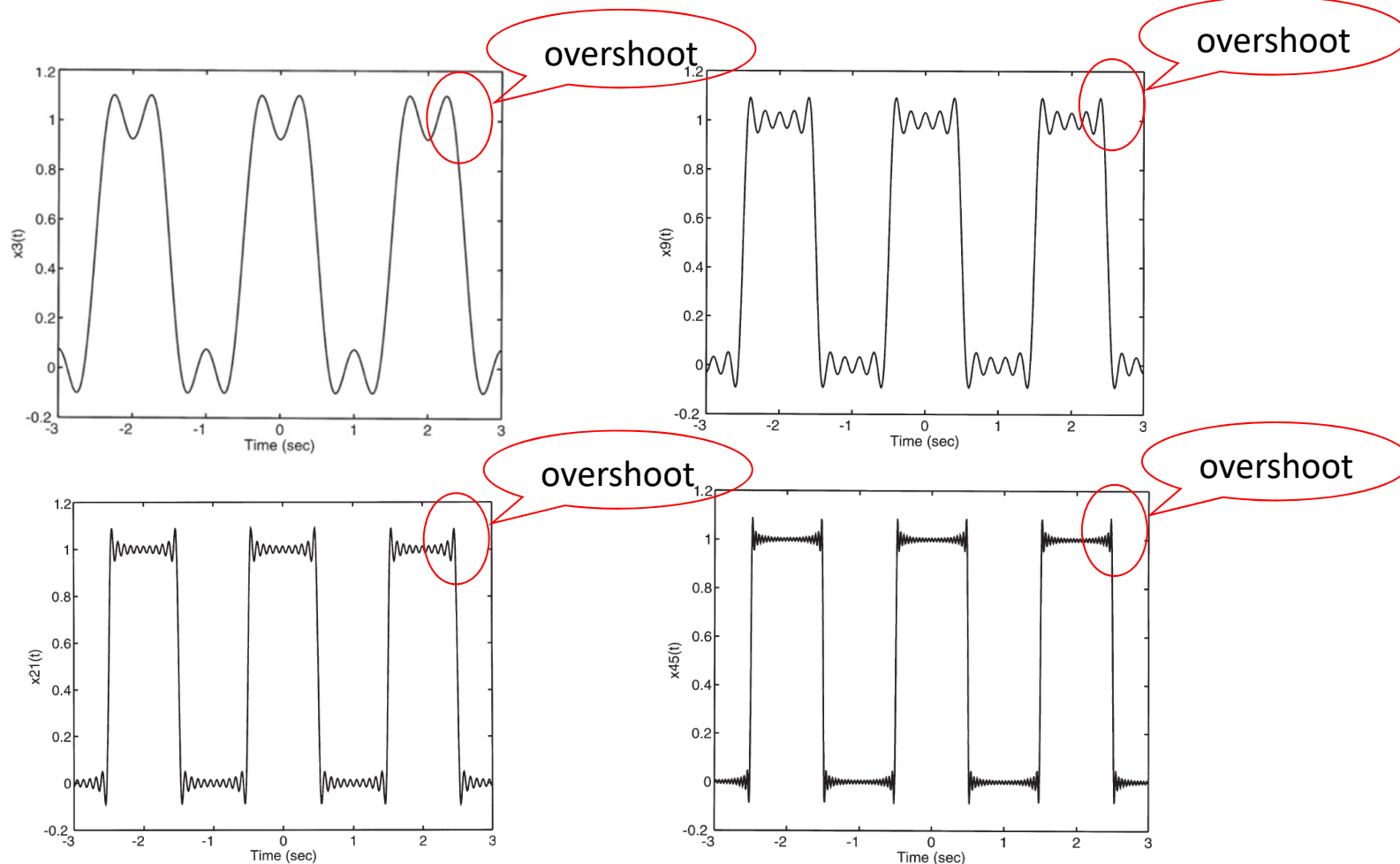


In general, in the frequency domain, there are a set of sine waves and a set of cosine waves. The a values and b values will differ.



The Fundamental Frequency $f=1/T$ is the frequency of the periodic function of time. The harmonics are integer multiples of this (but not necessarily every integer).

Gibbs Phenomenon



Adding the cisoids together will not give a sudden transition in the original function of time.

Gibbs Phenomenon

- The overshoot at the corners is still present even in the limit as N approaches to infinity. This characteristic was first discovered by Josiah Willard Gibbs (1893-1903), and this overshoot is referred to as the *Gibbs phenomenon*.
- Now let $x(t)$ be an arbitrary periodic signal. As a consequence of the Gibbs phenomenon, the Fourier series representation of $x(t)$ is not actually equal to the true value of $x(t)$ at any points where $x(t)$ is discontinuous.
- If $x(t)$ is discontinuous at $t = t_1$, the Fourier series representation is off by approximately 9% at t_1^+ and t_1^- .



The exponential form of the Fourier Series

- Let's recall the original form of Fourier series:

$$x(t) = a_0 + \sum_{n=1}^{\infty} a_n \cos(n\omega t) + \sum_{n=1}^{\infty} b_n \sin(n\omega t)$$

- In order to reduce the amount of 'writing out' the Fourier series, an exponential form can be expressed as:

$$a_n \cos(n\omega t) = (a_n/2) \cdot [e^{jn\omega t} + e^{-jn\omega t}]$$

$$b_n \sin(n\omega t) = (b_n/2j) \cdot [e^{jn\omega t} - e^{-jn\omega t}]$$

$$\begin{aligned} a_n \cos(n\omega t) + b_n \sin(n\omega t) &= (a_n/2) [e^{jn\omega t} + e^{-jn\omega t}] + (b_n/2j) [e^{jn\omega t} - e^{-jn\omega t}] \\ &= X_n \cdot e^{jn\omega t} + X_{-n} \cdot e^{-jn\omega t} \end{aligned}$$

$$\begin{aligned} \text{where: } X_n &= \frac{1}{2} (a_n - j b_n) \quad n \neq 0 \\ X_{-n} &= \frac{1}{2} (a_n + j b_n) \quad n \neq 0 \end{aligned}$$

So the original Fourier Series can be written out as:

$$x(t) = \sum_{n=-\infty}^{\infty} X_n \cdot e^{jn\omega t}$$

Where we have defined: $X_0 = a_0$

Summary of the Fourier Series

- Three forms

- Original (sine and cosine components)

$$x(t) = a_0 + \sum_{n=1}^{\infty} a_n \cos(n\omega t) + \sum_{n=1}^{\infty} b_n \sin(n\omega t)$$

- Cosine-with-phase form

$$x(t) = a_0 + \sum_{n=1}^{\infty} A_n \cos(n\omega_0 t + \theta_k) \quad -\infty < t < \infty$$

- Exponential form

$$x(t) = \sum_{n=-\infty}^{\infty} X_n e^{jn\omega t}$$

- Dirichlet conditions
- Gibbs Phenomenon

