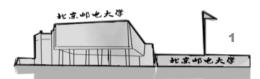


Chapter 5

Baseband Transmission of Digital Signals

School of Information and Communication Engineering

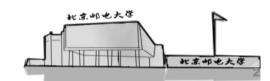
Beijing University of Posts and Telecommunications



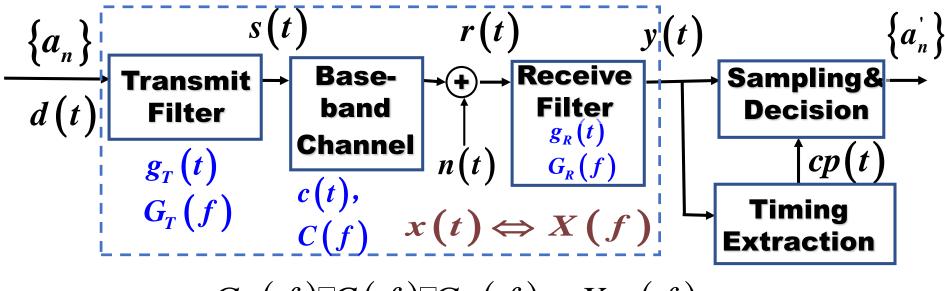


Baseband Transmission of Digital Signals

- ■Introduction
- Baseband signal and pulse modulation
- □ Digital PAM signal transmission through AWGN channel
- □ Digital PAM signal transmission through baseband channel
- □ Optimal transmission of digital PAM signal through ideal baseband channel and under AWGN condition
- ■Eye Diagram
- □ Channel Equalization
- □Partial Response System
- ■Symbol Synchronization
- **□Summary**



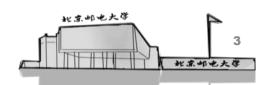




$$G_{T}(f)\square C(f)\square G_{R}(f) = X_{RC}(f)$$

If Matched Filter was adopted at the receiver for optimal reception

$$G_R(f) = G_T^*(f)C^*(f)$$



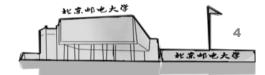
For ideal baseband channel, without time delay

$$G_{T}(f) = G_{R}(f) = \sqrt{X_{RC}(f)}$$

For ideal baseband channel, with time delay

$$G_{T}(f) = \sqrt{X_{RC}(f)}e^{-j2\pi f t_{T}}$$

$$G_R(f) = \sqrt{X_{RC}(f)}e^{-j2\pi f t_R}$$



BER analysis

For equal probability transmission of bipolar PAM signal

$$s_1(t) = -s_2(t) \Rightarrow E_b = E_1 = E_2, \quad \rho_{12} = -1$$

For optimal reception with minimum average BER

$$P_{b} = \frac{1}{2} \operatorname{erfc}\left(\sqrt{\frac{E_{b}}{N_{0}}}\right) = Q\left(\sqrt{\frac{2E_{b}}{N_{0}}}\right)$$

where, N_0 is the single side power spectral density of AWGN, E_b is the average bit energy.

$${}^{2023/11/21}E_{b} = E_{1} = E_{2} = \int_{-\infty}^{\infty} s_{1or2}^{2}(t) dt = \int_{-\infty}^{\infty} \left| C(f) G_{T}(f) \right|^{2} df = \int_{-\infty}^{\infty} \left| G_{T}(f) \right|^{2} df$$

The PSD of the transmitted signal

$$P_{T}(f) = \frac{1}{T_{b}} |G_{T}(f)|^{2} = \frac{1}{T_{b}} X_{RC}(f)$$

The average bit energy of the transmitted signal

$$E_{b} = \int_{-\infty}^{\infty} \left| G_{T}(f) \right|^{2} df = \int_{-\infty}^{\infty} X_{RC}(f) df = 1$$

also:
$$E_b = T_b \int_{-\infty}^{\infty} P_T(f) df$$





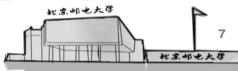
At the receiver, the equivalent noise bandwidth

$$B_{n} = \frac{1}{2|G_{R}(f)|_{\text{max}}^{2}} \int_{-\infty}^{\infty} |G_{R}(f)|^{2} df = \frac{1}{2X_{RC}(0)} \int_{-\infty}^{\infty} X_{RC}(f) df = \frac{1}{2T_{b}} = \frac{R_{b}}{2}$$
(Note: $B_{n} = R_{s} / 2$ for MPAM)

• The output noise of the receive filter $G_R(f)$

Average noise power

$$\sigma_n^2 = \frac{N_0}{2} \int_{-\infty}^{\infty} |G_R(f)|^2 df = \frac{N_0}{2} \int_{-\infty}^{\infty} X_{RC}(f) df = \frac{N_0 E_b}{2} = \frac{N_0}{2}$$





The PSD of the output signal from the receive filter

$$P_{s}(f) = \frac{\sigma_{a}^{2}}{T_{s}} |G_{T}(f)|^{2} + \frac{m_{a}^{2}}{T_{s}^{2}} \sum_{m=-\infty}^{\infty} \left| G_{T}\left(\frac{m}{T_{s}}\right) \right|^{2} \delta\left(f - \frac{m}{T_{s}}\right)$$

 For equal probability transmission of bipolar sequence {a_n}E{A,-A}

$$m_{a} = 0, \delta_{a}^{2} = A^{2}, |G_{T}(f)|^{2} = |G_{R}(f)|^{2} = X_{RC}(f)$$

$$P_{s}(f) = \frac{A^{2}}{T_{s}} X_{RC}(f)$$

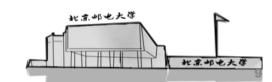
$$P_{b} = \frac{1}{2} erfc \left(\sqrt{\frac{E_{b}}{N_{0}}}\right) = \frac{1}{2} erfc \left(\sqrt{\frac{A^{2}}{N_{0}}}\right), E_{b} = A^{2}$$

$$E_{b} = A^{2}$$



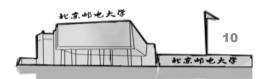
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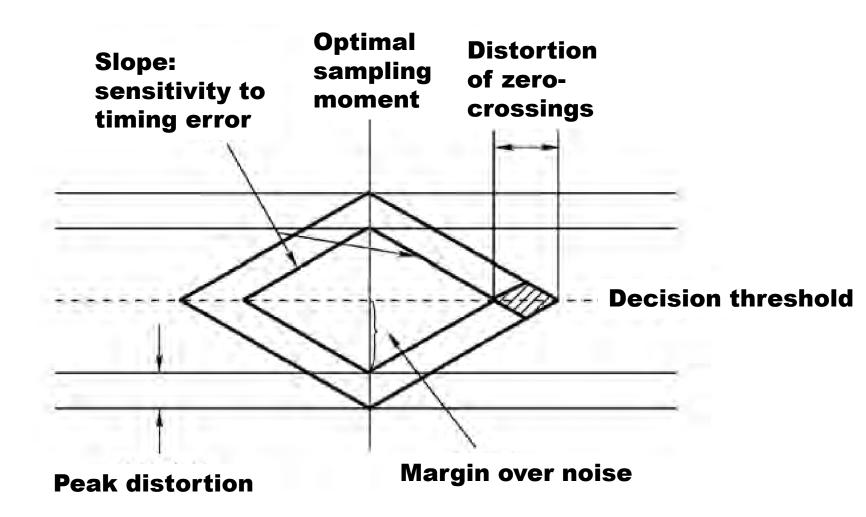




- Eye diagram is a method of practical observation of the received signal by using an oscilloscope.
- It requires that the scanning period of the oscilloscope is times of the symbol period T_{s} .
- It is the overlapping of the output signal waves, which reflects the variations of the signals.
- It looks like eyes.

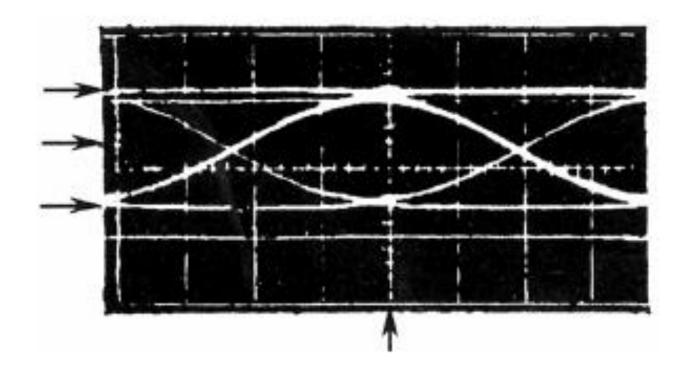


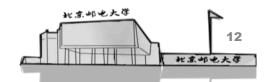






Eye diagram of 2PAM







Eye diagram of 4PAM

