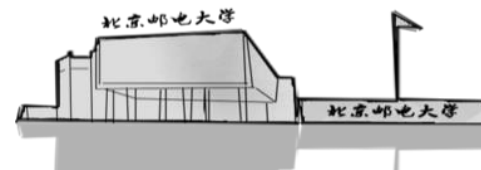


Chapter 5

Baseband Transmission of Digital Signals

**School of Information and Communication
Engineering**

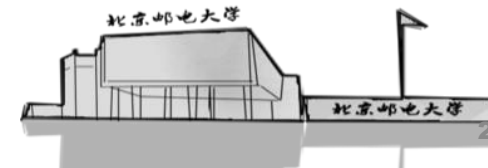
**Beijing University of Posts and
Telecommunications**





Baseband Transmission of Digital Signals

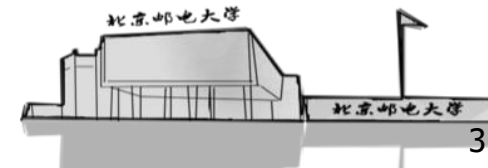
- Introduction
- Baseband signal and pulse modulation
- Digital PAM signal transmission through AWGN channel
- Digital PAM signal transmission through baseband channel
- Channel Equalization
- Eye Pattern
- Partial Response System
- Symbol Synchronization
- Summary





Digital PAM Signal Transmission through Baseband channel

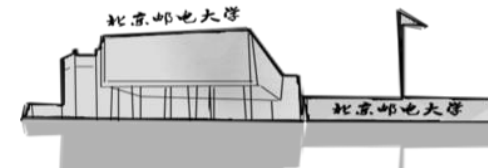
- ❑ Baseband Channel
- ❑ Inter Symbol Interference (ISI)
- ❑ Nyquist Criterion





! The reasons for modeling PAM signal transmission through baseband channel

- ❑ Band limited channel will lead to dispersion in time domain, which might cause **Inter Symbol Interference (ISI)**.
- ❑ Baseband transmission is adopted in practical applications, e.g., **PCM system**.
- ❑ Bandpass systems are usually discussed **through their equivalent baseband systems**.

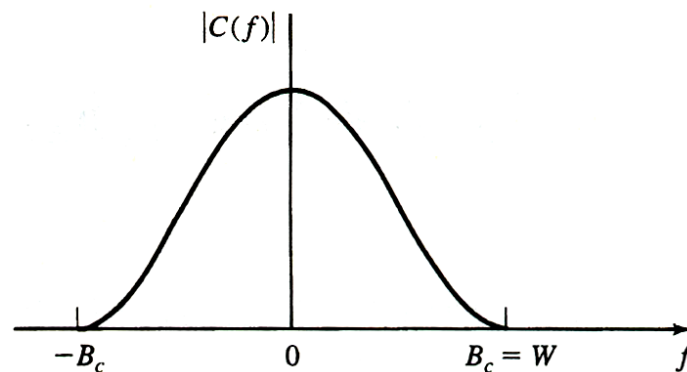


Baseband Channel Model

$$c(t) \Leftrightarrow C(f)$$

□ For baseband channel

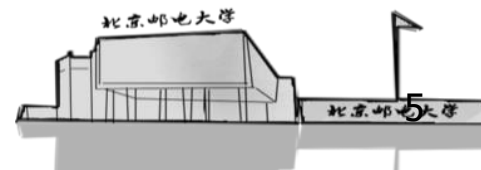
$$C(f) = |C(f)| e^{j\theta(f)}, \quad |f| \leq W$$



$|C(f)|$ ~ **Amplitude-frequency characteristic**

$\theta(f)$ ~ **Phase-frequency characteristic**

$\tau_G(f) = -\frac{1}{2\pi} \frac{d\theta(f)}{df}$ ~ **Group delay characteristic**



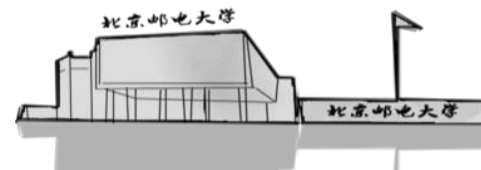
Ideal Baseband Channel Model

□ For ideal baseband channel

$|C(f)|$ and $\tau_G(f)$ are both constant.

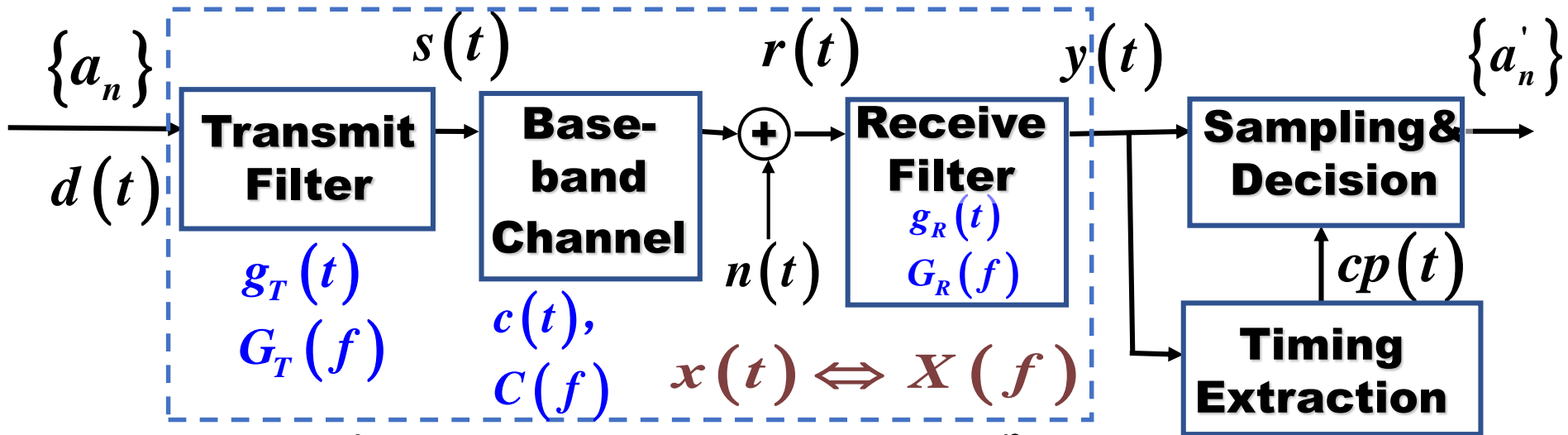
□ Since the frequency spectrum of the transmitted signal is filtered by the channel, its time domain pulse might be dispersed.

□ Thus, ISI might be introduced.





Inter-symbol Interference

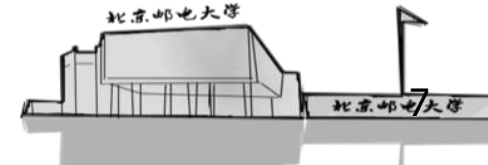


$$\blacksquare d(t) = \sum_{n=-\infty}^{\infty} a_n \delta(t - nT_s) \quad \blacksquare s(t) = \sum_{n=-\infty}^{\infty} a_n g_T(t - nT_s),$$

$$\blacksquare r(t) = \sum_{n=-\infty}^{\infty} a_n h(t - nT_s) + n(t), \quad h(t) = g_T(t) * c(t)$$

$$\blacksquare y(t) = \sum_{n=-\infty}^{\infty} a_n x(t - nT_s) + \gamma(t), \quad \gamma(t) = n(t) * g_R(t)$$

$$x(t) = g_T(t) * c(t) * g_R(t) \iff X(f) = G_T(f) \cdot C(f) \cdot G_R(f)$$





□ Output signal of receive filter

$$y(t) = \sum_{n=-\infty}^{\infty} a_n x(t - nT_s) + \gamma(t)$$

At the sampling moment, $t = mT_s + t_0$ ($t_0 = 0$),

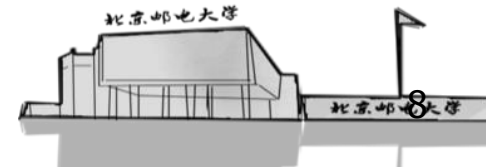
$$y(mT_s) = \sum_{n=-\infty}^{\infty} a_n x(mT_s - nT_s) + \gamma(mT_s)$$

Or

$$y_m = \sum_{n=-\infty}^{\infty} a_n x_{m-n} + \gamma_m = x_0 a_m + \underbrace{\sum_{n \neq m} a_n x_{m-n}}_{\text{Inter-symbol Interference(ISI)}} + \underbrace{\gamma_m}_{\text{Random noise}}$$

**Inter-symbol
Interference(ISI)**

**Random
noise**



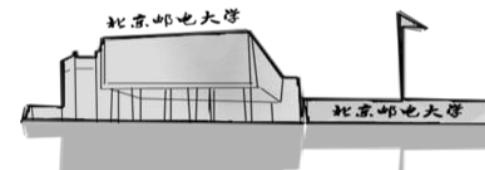
Nyquist Criterion

- ❑ It is possible to eliminate ISI at the sampling moments by ensuring that the system satisfies the Nyquist criterion.
- ❑ At the moment $t=mT_s$,

$$y_m = x_0 a_m + \sum_{\substack{n=-\infty \\ m \neq n}}^{\infty} a_n x_{m-n} + \gamma_m$$

- ❑ If there is no ISI

$$x(nT_s) = \begin{cases} 1 & \text{for } n = 0 \\ 0 & \text{for } n \neq 0 \end{cases}$$





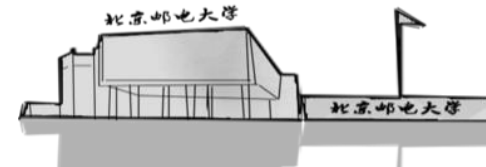
Nyquist Criterion

□ Then

$$y_m = a_m + \gamma_m$$

□ In frequency domain, the Nyquist criterion becomes

$$\sum_{m=-\infty}^{\infty} X\left(f + \frac{m}{T_s}\right) = T_s, \quad |f| \leq \frac{1}{2T_s}$$

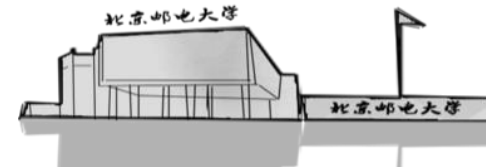


□ In time domain

$$\begin{aligned}x(t) &= \sum_{n=-\infty}^{\infty} \delta(t - nT_s) \\&= \sum_{n=-\infty}^{\infty} x(nT_s) \delta(t - nT_s) \\&= \delta(t)\end{aligned}$$

□ In frequency domain

$$X(f) * \frac{1}{T_s} \sum_{n=-\infty}^{\infty} \delta\left(f - \frac{n}{T_s}\right) = \frac{1}{T_s} \sum_{n=-\infty}^{\infty} X\left(f - \frac{n}{T_s}\right) = 1 \Rightarrow \sum_{n=-\infty}^{\infty} X\left(f - \frac{n}{T_s}\right) = T_s$$

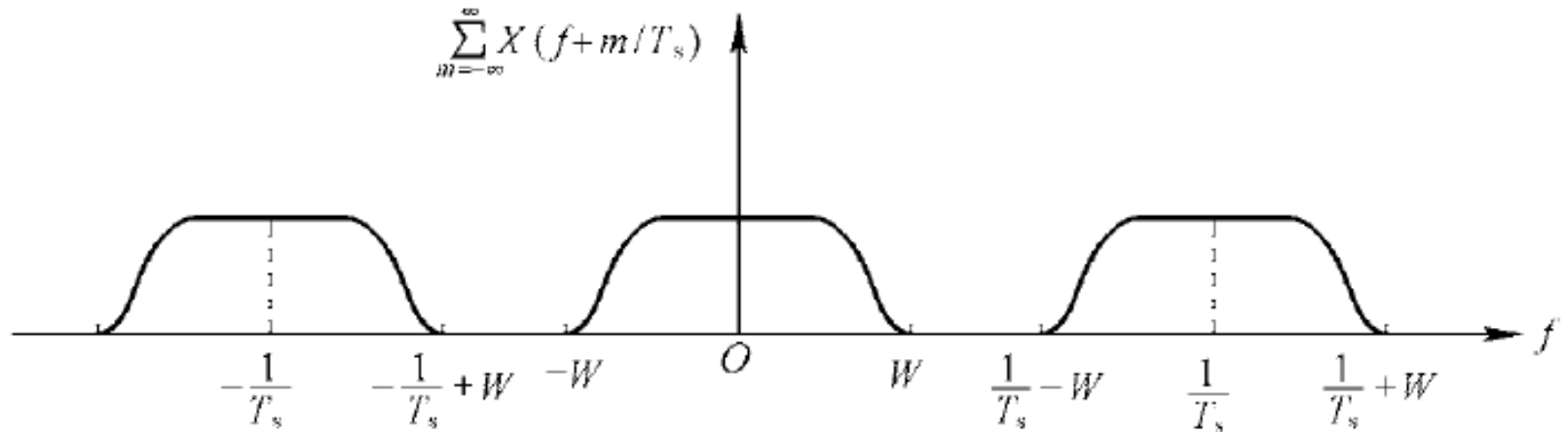




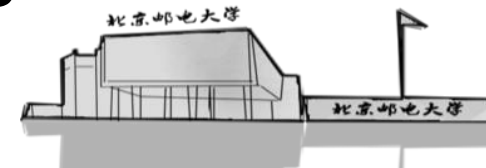
Nyquist Criterion

□ Observation 1:

- **When $W < 1/2T_s$, the Nyquist Criterion can not be satisfied.**



- **In the example illustrated above, it is very clear that gaps between spectra make a constant overlapping impossible.**

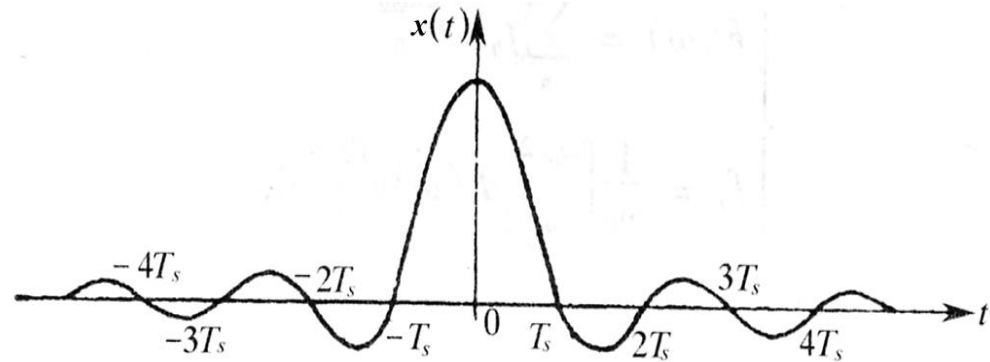
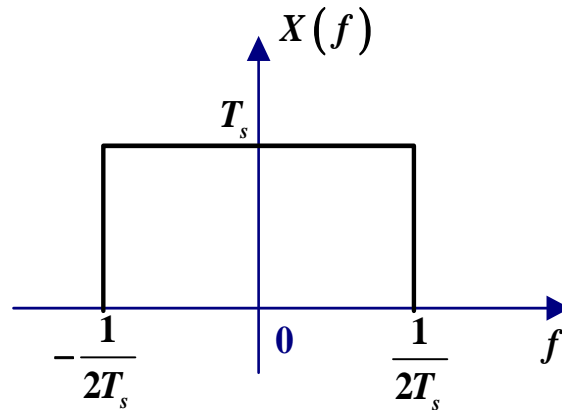




Nyquist Criterion

□ Observation 2:

- **The minimum bandwidth of $X(f)$ for no ISI transmission: $W=1/2T_s$**



$$X(f) = \begin{cases} T_s, & |f| \leq \frac{1}{2T_s} \\ 0, & \text{otherwise} \end{cases}$$

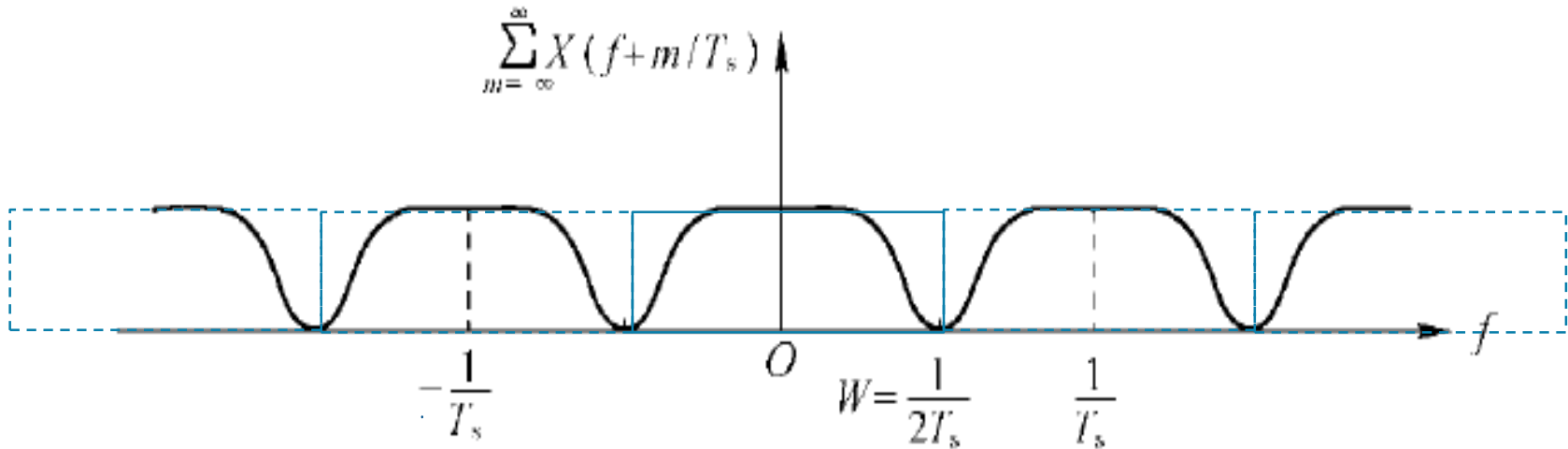
$$x(t) = \text{sinc}\left(\frac{t}{T_s}\right)$$

- **But, this ideal filter is very difficult to be produced.**



Nyquist Criterion

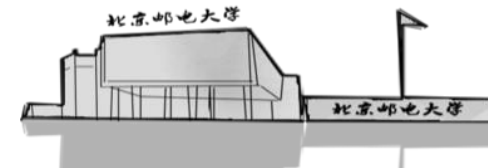
- **The minimum bandwidth of $X(f)$ for no ISI transmission: $W=1/2T_s$**



- **Maximum frequency efficiency:**

$$\eta = \frac{R_s}{W} = 2(\text{Baud/Hz})$$

- **also** $\eta = \frac{R_b}{W} = 2\log_2 M = 2k(\text{bit/s/Hz})$





Discussion

For a given baseband transmission system with bandwidth W

- Its maximum transmission rate without ISI: $2W$
- Its maximum frequency efficiency: 2Baud/Hz



For a given transmitted baseband digital signal with symbol rate R_s

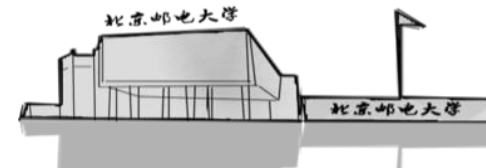
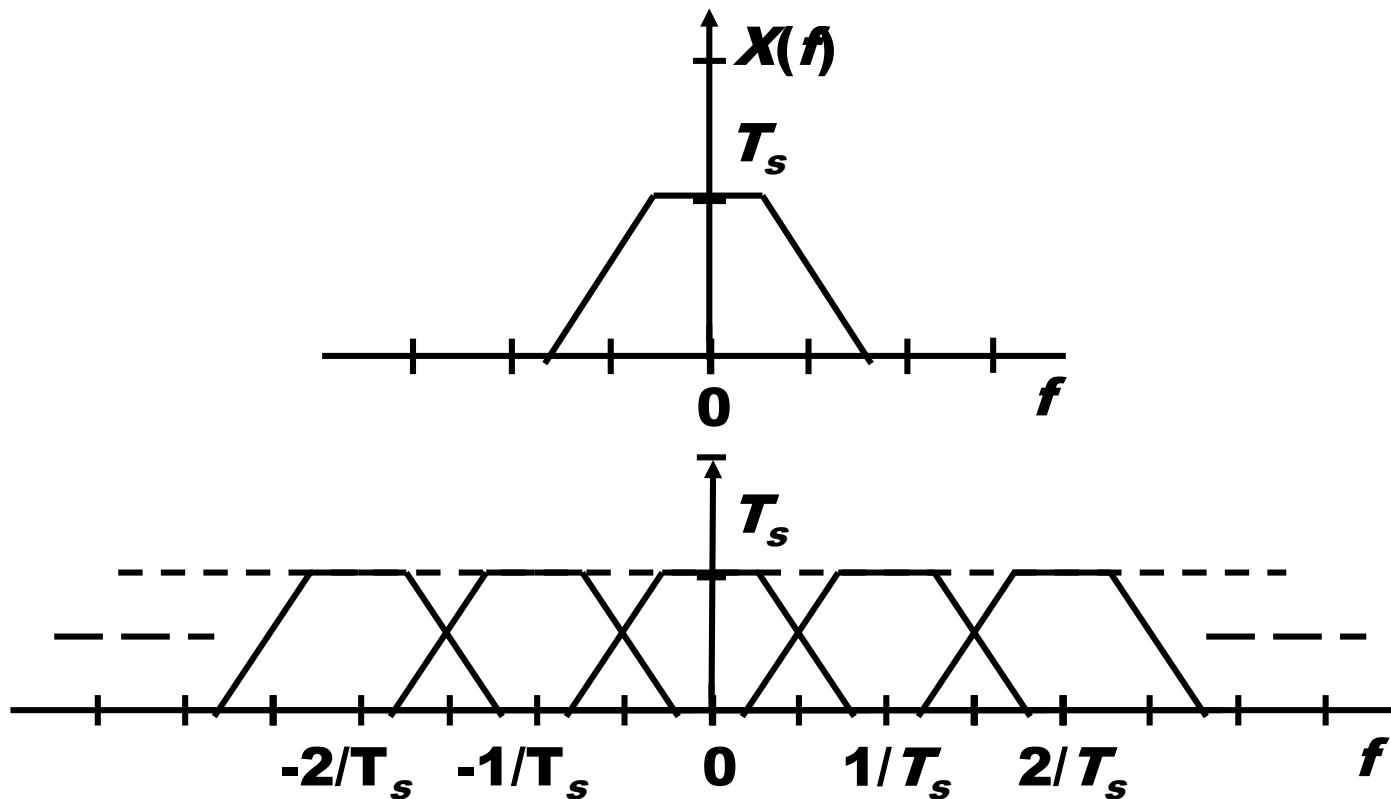
- For no ISI transmission, the minimum required bandwidth of system : $R_s/2$

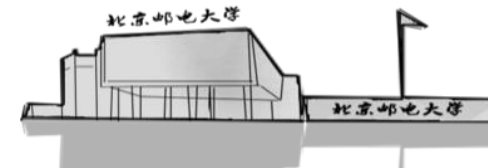
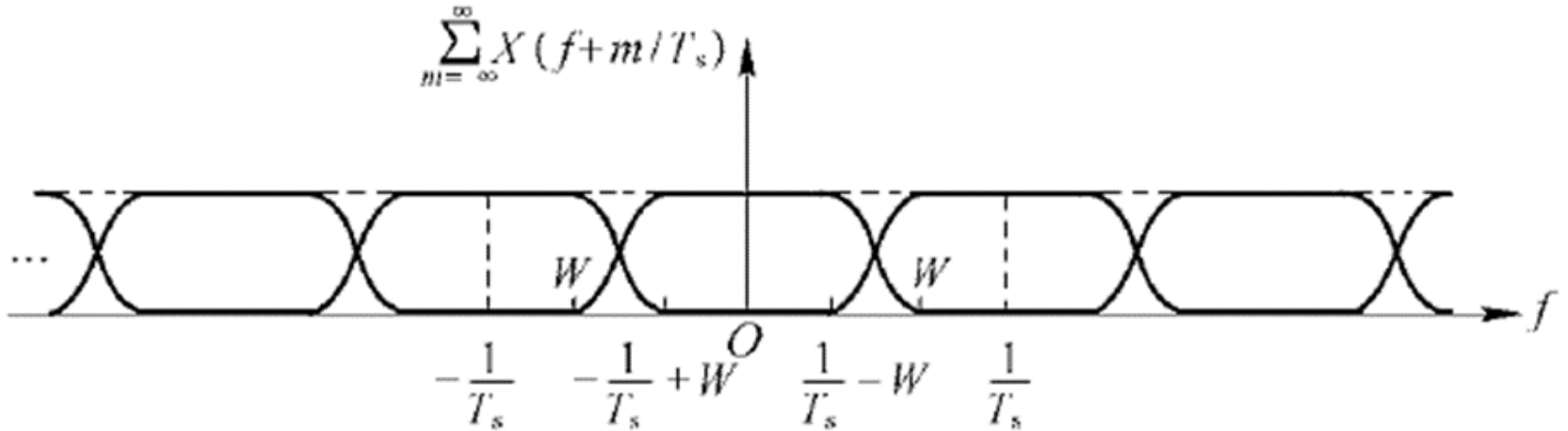
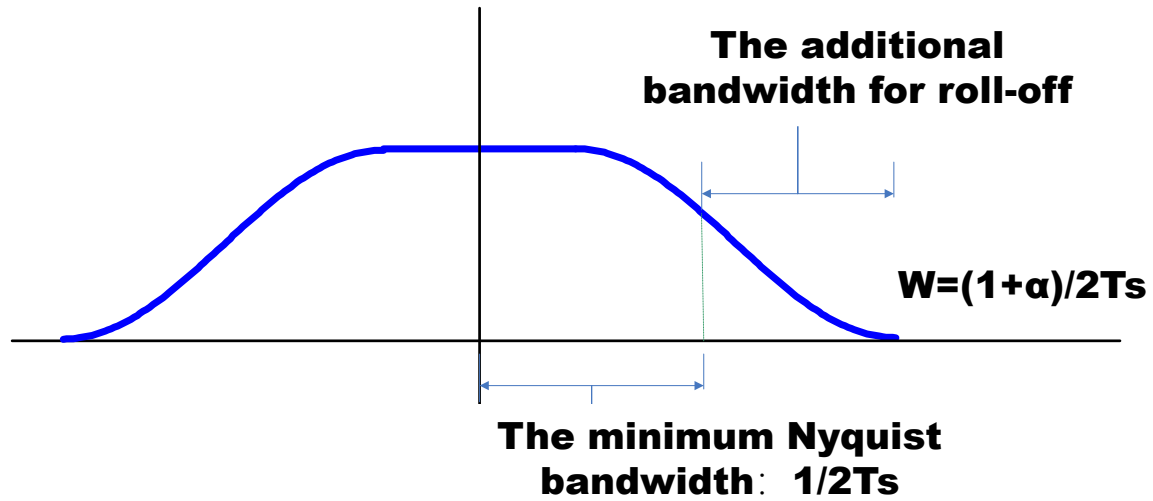


Nyquist Criterion

□ Observation 3:

- Shape of $X(f)$ when $W > 1/2T_s$

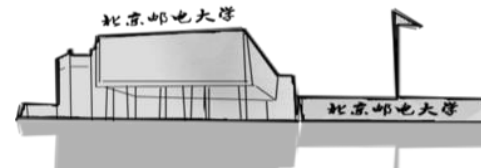




□ A reasonable scheme: **Raised Cosine Filter**

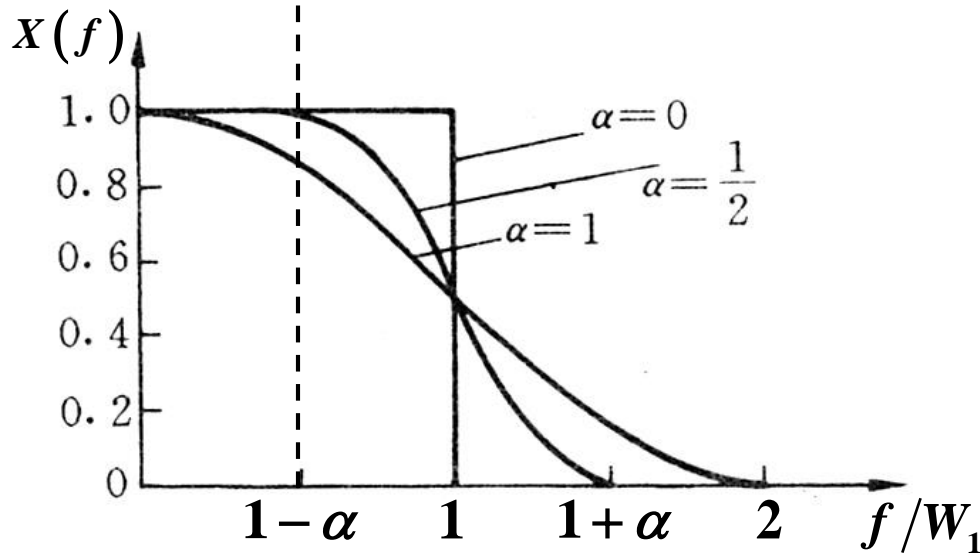
$$X(f) = \begin{cases} T_s, & 0 \leq |f| \leq \frac{1-\alpha}{2T_s} \\ \frac{T_s}{2} \left[1 + \cos \frac{\pi T_s}{\alpha} \left(|f| - \frac{1-\alpha}{2T_s} \right) \right], & \frac{1-\alpha}{2T_s} < |f| \leq \frac{1+\alpha}{2T_s} \\ 0, & |f| > \frac{1+\alpha}{2T_s} \end{cases}$$

With Roll-off factor: $0 \leq \alpha \leq 1$





RC Filter



- **Bandwidth and symbol rate:**

$$W_1 = R_s / 2$$

$$W = W_1 + W_2 \\ = \frac{R_s}{2}(1 + \alpha)$$

- **Spectrum efficiency:**

$$\eta = \frac{2}{1 + \alpha} \text{ (Baud/Hz)}$$

$\alpha = 0$

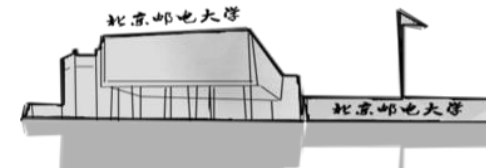


Ideal LPF

$\alpha = 1$



$$X(f) = \begin{cases} \frac{T_s}{2} [1 + \cos \pi f T_s], & |f| \leq \frac{1}{T_s} \\ 0, & |f| > \frac{1}{T_s} \end{cases}$$



RC Filter

□ Impulse response of RC Filter

$$x(t) = \text{sinc}\left(\frac{t}{T_s}\right) \cdot \frac{\cos(\alpha\pi t f_s)}{1 - 4\alpha^2 t^2 f_s^2} \quad x(t)$$

$$\alpha = 1$$

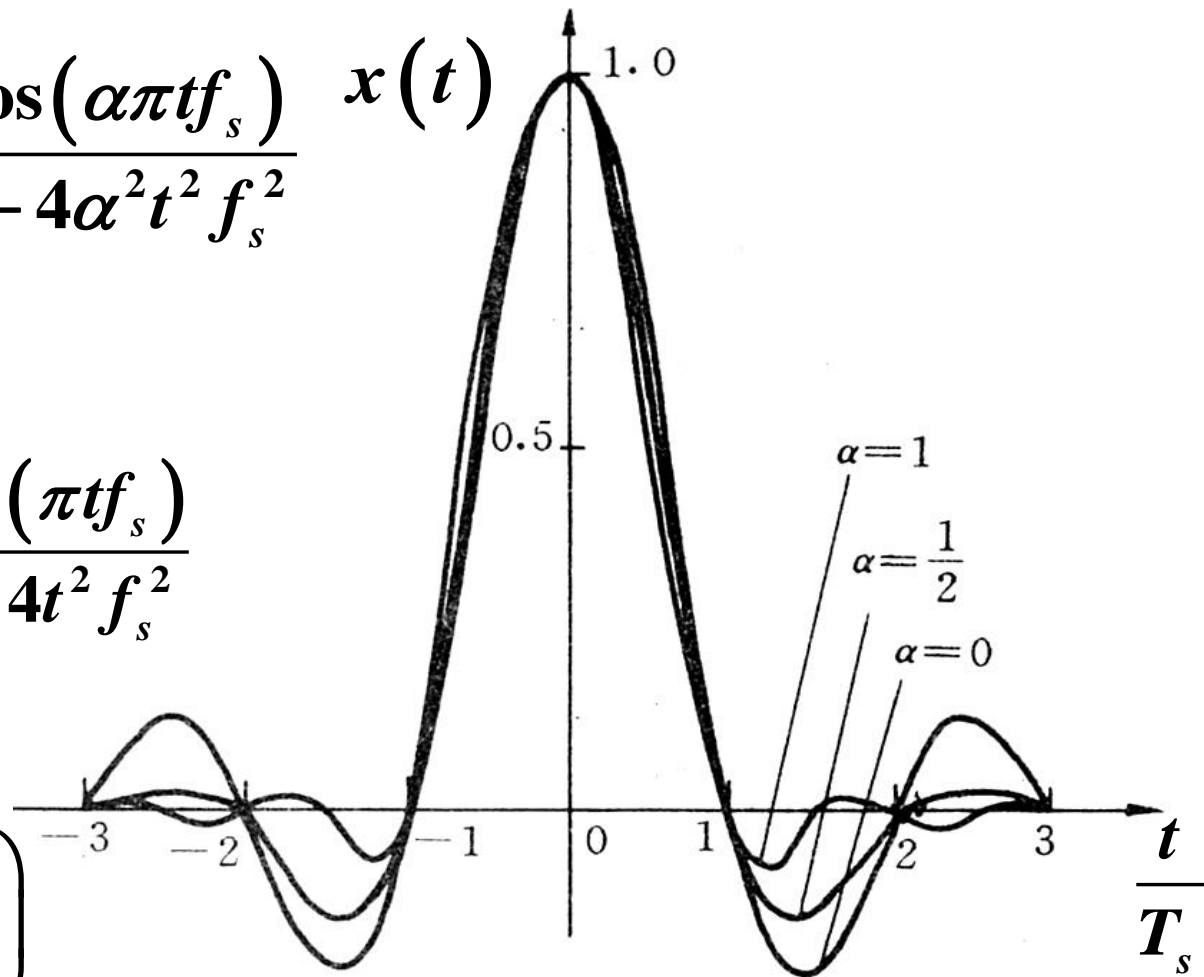


$$x(t) = \text{sinc}\left(\frac{t}{T_s}\right) \cdot \frac{\cos(\pi t f_s)}{1 - 4t^2 f_s^2}$$

$$\alpha = 0$$



$$x(t) = \text{sinc}\left(\frac{t}{T_s}\right)$$





Summary

- ❑ **Digital Baseband signal transmission through Baseband channel might encounter ISI.**
- ❑ **To eliminate ISI, we should properly design the impulse response and transfer function of the system, i.e., $x(t)$ and $X(f)$.**
- ❑ **The Nyquist Criterion must be satisfied for no ISI transmission.**

