

Implementation of Variable Precision Floating Point Fast Square Root Architecture

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References

- ▶ Advanced Components in the Variable Precision Floating Point Library: *Xiaojun Wang, Sherman Braganza, Miriam Lesser*
- ▶ Reciprocation, Square Root, Inverse Square Root, and Some Elementary Functions using Small Multipliers: *Milos Ercegovac, Jean-Michel Muller*

Floating-Point Square Root Architecture

- ▶ **Overview:** Uses small table lookup and takes advantage of Taylor expansion
- ▶ **Key Steps:**
 - ▶ **Reduction:** Reduces input size through initial lookup
 - ▶ **Evaluation:** Computes Taylor series terms for accuracy
 - ▶ **Post-Processing:** Multiplies result by corrective factor
- ▶ **Pipelining:** Linear/Non-Iterative, FPGA-friendly. Not implemented in the paper.
- ▶ **Utility:** Reduces hardware and time complexity by using only lookup tables and multipliers.

Input and Output

- ▶ 1 sign bit, n -Mantissa Bits and m -Exponent Bits.
- ▶ The current implementation is designed for IEEE Single Precision Floating Point Format ($n = 23$, $m = 8$). Can be generalized.
- ▶ Bottom-up approach for designing; Top-down approach for thinking.
- ▶ Let exponent = E , mantissa = M , and sign = S

Square Root Module

- ▶ Check for special cases. Else, compute in parallel:
- ▶ $S_{out} = 0$
- ▶ $E_{out} = \text{floor}[\frac{E_{in} - \text{Bias}}{2}] + \text{Bias}$, where $\text{Bias} = 2^{n-1} - 1$.
- ▶ $M_{out} = \text{FixedPointSquareRoot}(M_{in})$. If $[E_{in} - \text{Bias}]$ is odd, subtract 1 and assign $M_{out} := \sqrt{2}M_{out}$
- ▶ Special Cases of Input:

Input	Output
\pm Zero ($E_{in} = 0, M_{in} = 0$)	Zero (00000000)
Negative Non-zero ($S_{in} = 1$)	NaN (7FC00000)
+Inf ($S_{in} = 0, E_{in} = FF$ and $M_{in} = 0$)	+Inf (7F800000)
-Inf ($S_{in} = 1, E_{in} = FF$ and $M_{in} = 0$)	NaN (7FC00000)
NaN ($E_{in} = FF$ and $M_{in} \neq 0$)	NaN (7FC00000)

FixedPointSquareRoot Module

- ▶ **Input:** Y is an n -bit fixed point number corresponding to the fractional part (mantissa). Hidden bit assumed to be 1. Thus, the range is $[1, 2)$. Here,
- ▶ **Idea:** Let $k = \lceil \frac{n+1}{4} \rceil$. Then, we write

$$\sqrt{Y} = \frac{\sqrt{Y\hat{R}}}{\sqrt{\hat{R}}} \quad (1)$$

- ▶ We want \hat{R} such that $\sqrt{Y\hat{R}}$ is very close to 1, and thus Taylor Series can be used.
- ▶ \hat{R} is a $(k+1)$ -bit number whose square root is computed using a small LUT as $(k+1)$ is much smaller than n .
- ▶ Using an LUT, we calculate $\hat{R} = \frac{1}{Y^{(k)}}$, rounded down, where $Y^{(k)}$ is Y truncated to k bits.

Dealing with Numerator

- ▶ **Goal:** To calculate $\sqrt{Y\hat{R}}$
- ▶ As $Y\hat{R} \approx 1$, we define the error term, a $4k$ -bit number

$$A = Y\hat{R} - 1 \quad (2)$$

- ▶ Now, as $Y^{(k)}$ is Y truncated to k^{th} bit, we have

$$\begin{aligned} Y^{(k)} &\leq Y \leq Y^{(k)} + 2^{-k} \\ \implies 1 &\leq \frac{Y}{Y^{(k)}} \leq 1 + 2^{-k} \quad \text{as } Y^{(k)} \geq 1 \end{aligned}$$

- ▶ Also, as $\frac{1}{Y^{(k)}} - 2^{-k-1} < \hat{R} \leq \frac{1}{Y^{(k)}}$,

$$\begin{aligned} \implies 1 - 2^{-k} &< \hat{R}Y^{(k)} < 1 \\ \implies \text{Combining, } 1 - 2^{-k} &< \hat{R}Y^{(k)} < 1 + 2^{-k} \\ \implies -2^{-k} &< A < +2^{-k} \end{aligned}$$

Dealing with Numerator - Taylor Series

- Therefore, we can write

$$A = 0 + A_2 2^{-2k} + A_3 2^{-3k} + A_4 2^{-4k} \dots \quad (3)$$

where each $A_k \in \mathbb{Z}$, $|A_k| \leq 2^k - 1$

- Using Taylor Series, we get the first few terms as:

$$\sqrt{1+A} = 1 + \frac{A}{2} - \frac{1}{8} A_2^2 \cdot 2^{-4k} - \frac{1}{4} A_2 A_3 \cdot 2^{-5k} + \frac{1}{16} A_2^3 \cdot 2^{-6k} \quad (4)$$

- If we store A_2, A_3 as k -bit integers, their squares, products and cubes as n -bit integers, and the result $B = \sqrt{1+A}$ as n -bit fractional number (excluding hidden bit), then the required right shifts effectively are reduced by n :
 - Square: $4k - n$ (here, $28 - 23 = 5$)
 - Cube: $6k - n$ (here, $42 - 23 = 19$)
 - Product: $7k - n$ (here, $35 - 23 = 12$)

Negative value of A

- ▶ If A is negative, then $A_2, A_3, A_4 \dots$ are all negative.

$$\begin{aligned}\sqrt{1 - |A|} &= 1 - \frac{|A|}{2} - \frac{1}{8}|A_2|^2 \cdot 2^{-4k} \\ &\quad - \frac{1}{4}|A_2||A_3| \cdot 2^{-5k} - \frac{1}{16}|A_2|^3 \cdot 2^{-6k}\end{aligned}$$

- ▶ Two's complement of A is taken after (2) to avoid overflow in (3) when using negative numbers directly. This automatically gives the values of $|A_2|, |A_3|$ etc
- ▶ In both positive and negative cases, let $B = \sqrt{1 + A}$

Dealing with Denominator

- ▶ Denominator $\sqrt{\hat{R}}$ is a $(k + 1)$ -bit number whose square root can be computed using a small LUT as $(k + 1)$ is much smaller than n .
- ▶ However, as we do not want to use dividers which increase complexity, we use a LUT to calculate $\frac{1}{\sqrt{\hat{R}}}$ (rounded down), storing n -bit result M .
- ▶ These LUTs are generated using Python code.
- ▶ Final Output $\sqrt{Y} = \frac{B}{\sqrt{\hat{R}}} = B \cdot M$.
- ▶ If we use a minimal-sized multiplier, we need to add 0.5×2^{-4k} to final output.

Limitations and Errors

- ▶ Error $< 2.39 \times 2^{-4k}$
- ▶ One bit of error may be present at the least significant bit of mantissa. Happens sometimes when exponent is odd.
- ▶ Denormal Numbers ($< 2^{-126}$), where hidden bit is zero, are not supported by this paper, as the fixed point module support inputs in $[1, 2)$ only.