

# TTK4190 Guidance and Control of Vehicles

## Draft Solution of Assignment 2

## Guidelines

- This presentation is only intended as a draft of a possible way to solve the assignment.
- There may exist better ways to solve the problems, so don't (necessarily) be alarmed if your solution deviates from this.

# Outline

## 1. Open-Loop Analysis

## 2. Autopilot for Course Hold using Aileron and Successive Loop Closure

# Problem 1 - Open-loop Analysis

## Problem 1a)

*What is the ground speed of the aircraft (numerical value) in the absence of wind?*

### What we know

- The relationship between ground speed ( $V_g$ ), air speed ( $V_a$ ), and wind ( $V_w$ ) is given by the wind triangle
- The wind speed is zero

### Solution

$$V_g = V_a + V_w = 580\text{km/h} + 0\text{km/h} = 580\text{km/h} \approx 161.11\text{m/s}$$

# Problem 1 - Open-loop Analysis

## Problem 1b)

*Write down two expressions for the sideslip angle  $\beta$  in the absence of wind. One expression should depend on aircraft velocity and the other on aircraft heading.*

### Notes

- The book is somewhat messy when it comes to the relationship between sideslip, wind and air velocity.
- The sideslip and crab angle are equal when there is no wind (and not necessarily zero).

### Solution

The first expression (Eq. (28) p.20 in Beard & McLain) is based on the air velocity:

$$\beta = \sin^{-1} \left( \frac{v_r}{\sqrt{u_r^2 + v_r^2 + w_r^2}} \right) = \sin^{-1} \left( \frac{v}{\sqrt{u^2 + v^2 + w^2}} \right) = \sin^{-1} \left( \frac{v}{V_g} \right)$$

The second expression is based on the course and heading angle (crab equal to sideslip without wind):

$$\beta = \chi - \psi$$

## Problem 1 - Open-loop Analysis

### Problem 1c)

Compute the Dutch-roll natural frequency and relative damping ratio for the aircraft.

$$\begin{bmatrix} \dot{\beta} \\ \dot{\phi} \\ \dot{p} \\ \dot{r} \\ \dot{\delta}_a \end{bmatrix} = \begin{bmatrix} -0.322 & 0.052 & 0.028 & -1.12 & 0.002 \\ 0 & 0 & 1 & -0.001 & 0 \\ -10.6 & 0 & -2.87 & 0.46 & -0.65 \\ 6.87 & 0 & -0.04 & -0.32 & -0.02 \\ 0 & 0 & 0 & 0 & -7.5 \end{bmatrix} \begin{bmatrix} \beta \\ \phi \\ p \\ r \\ \delta_a \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ 0 \\ 0 \\ 7.5 \end{bmatrix} \delta_a^c$$

### Solution - First Part

We are only looking at the equations for the yaw rate and the sideslip (neglect other terms):

$$\dot{\beta} = -0.322\beta - 1.12r$$

$$\dot{r} = 6.87\beta - 0.32r$$

- Using the damp function on the reduced system yields  $\zeta = 0.115$  and  $\omega_n = 2.79$ .
- Using the damp function directly on **A** yields  $\zeta = 0.111$  and  $\omega_n = 2.85$ .

# Problem 1 - Open-loop Analysis

## Problem 1c)

*Describe how the Dutch roll mode affect the yaw and roll motion? How would the motion change with increased relative damping ratio?*

### Solution - Second Part

- [▶ Dutch Roll Animation](#)
- Since the damping ratio is low and under-damped, the tail will be wagging from side-to-side.
- Increased damping ratio would reduce the oscillations.
- A yaw damper can be used to limit the effect of the dutch-roll mode.

# Problem 1 - Open-loop Analysis

## Problem 1d)

Compute the spiral-divergence mode. Is the mode unstable?

$$\begin{bmatrix} \dot{\beta} \\ \dot{\phi} \\ \dot{p} \\ \dot{r} \\ \dot{\delta}_a \end{bmatrix} = \begin{bmatrix} -0.322 & 0.052 & 0.028 & -1.12 & 0.002 \\ 0 & 0 & 1 & -0.001 & 0 \\ -10.6 & 0 & -2.87 & 0.46 & -0.65 \\ 6.87 & 0 & -0.04 & -0.32 & -0.02 \\ 0 & 0 & 0 & 0 & -7.5 \end{bmatrix} \begin{bmatrix} \beta \\ \phi \\ p \\ r \\ \delta_a \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ 0 \\ 0 \\ 7.5 \end{bmatrix} \delta_a^c$$

## Assumptions

- $\dot{p} = p = 0$
- Rudder command is negligible

## Solution

We are only looking at the equations for roll rate and yaw rate:

$$\dot{p} = -10.6\beta + 0.46r - 0.65\delta_a = 0$$

$$\dot{r} = 6.87\beta - 0.32r - 0.02\delta_a$$



# Problem 1 - Open-loop Analysis

## Problem 1d)

### Solution

By solving the first equation for  $\beta$  and inserting it into the second equation:

$$\dot{r} = -0.0218r - 0.442\delta_a$$

- The pole (mode) is -0.0218, which means that the mode is very slow and stable.
- "damp(A)" gives a pole of -0.00056.

## Problem 1 - Open-loop Analysis

### Problem 1e)

Compute the roll mode. Is the roll mode faster or slower than the spiral-divergence mode?

$$\begin{bmatrix} \dot{\beta} \\ \dot{\phi} \\ \dot{p} \\ \dot{r} \\ \dot{\delta}_a \end{bmatrix} = \begin{bmatrix} -0.322 & 0.052 & 0.028 & -1.12 & 0.002 \\ 0 & 0 & 1 & -0.001 & 0 \\ -10.6 & 0 & -2.87 & 0.46 & -0.65 \\ 6.87 & 0 & -0.04 & -0.32 & -0.02 \\ 0 & 0 & 0 & 0 & -7.5 \end{bmatrix} \begin{bmatrix} \beta \\ \phi \\ p \\ r \\ \delta_a \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ 0 \\ 0 \\ 7.5 \end{bmatrix} \delta_a^c$$

### Assumptions

- We are only concerned with the equation for the roll rate  $p$ .
- $\beta = r = \delta_r = 0$

### Solution

The equation for the roll rate can (with the assumptions above) be written as

$$\dot{p} = -2.87p - 0.65\delta_a$$

- The pole (mode) is -2.87 and it is much faster than the spiral-divergence mode.
- "damp(A)" yields -2.88.

# Outline

1. Open-Loop Analysis

2. Autopilot for Course Hold using Aileron and Successive Loop Closure

## Problem 2 - Autopilot Design

### Problem 2 a)

*Explain why the smallest signed angle function  $ssa(.)$  should be applied to the angle errors  $e_\chi$  and  $e_\phi$*

#### Smallest Signed Angle (SSA)

The operator  $ssa : \mathbb{R} \rightarrow [-\pi, \pi)$  maps the unconstrained angle  $\tilde{x} = x - x_0 \in \mathbb{R}$  representing the difference between the two angles  $x$  and  $x_0$  to the smallest difference between the angles

$$\tilde{x}_s = ssa(\tilde{x})$$

where  $\tilde{x}_s \in \mathbb{S}^1$ .

Since the angle errors  $e_\chi$  and  $e_\phi$  can lie outside  $[-\pi, \pi)$  it is necessary to confide them within this interval using the  $ssa$  function.

## Problem 2 - Autopilot Design

### Problem 2b)

Find numerical values for  $a_{\phi_1}$  and  $a_{\phi_2}$  based on the state-space model.

$$\begin{bmatrix} \dot{\beta} \\ \dot{\phi} \\ \dot{p} \\ \dot{r} \\ \dot{\delta_a} \end{bmatrix} = \begin{bmatrix} -0.322 & 0.052 & 0.028 & -1.12 & 0.002 \\ 0 & 0 & 1 & -0.001 & 0 \\ -10.6 & 0 & -2.87 & 0.46 & -0.65 \\ 6.87 & 0 & -0.04 & -0.32 & -0.02 \\ 0 & 0 & 0 & 0 & -7.5 \end{bmatrix} \begin{bmatrix} \beta \\ \phi \\ p \\ r \\ \delta_a \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ 0 \\ 0 \\ 7.5 \end{bmatrix} \delta_a^c$$

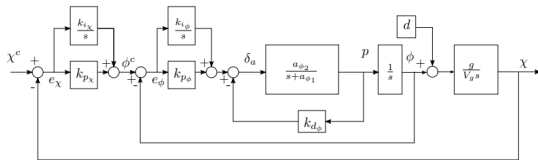


Figure: Block diagram of simplified system

### Assumptions

- We are considering a simplified system between roll, roll rate, and aileron
- Ignore sideslip and yaw rate, i.e.,  $\beta = r = 0$ .

## Problem 2 - Autopilot Design

### Problem 2b)

#### Solution

With the assumptions outlined above, the simplified roll rate equation becomes:

$$\dot{p} = -2.87p - 0.65\delta_a$$

Laplace transform:

$$H(s) = \frac{p(s)}{\delta_a(s)} = \frac{-0.65}{s + 2.87} = \frac{a_{\phi_2}}{s + a_{\phi_1}}$$

Therefore,

$$a_{\phi_1} = 2.87$$

$$a_{\phi_2} = -0.65$$

- The answer can be extracted directly from the roll mode.

## Problem 2c) - Successive Loop Closure

### Main Considerations

I will not go through the exact solution in detail. Some of the main concerns are:

- The inner loop must have a much higher bandwidth than the outer loop.
- It is necessary to start with the inner most-loop → Gains in roll loop identified first.
- Gains in the course loop are identified in the end.
- Using the Evans form is necessary in order to find the interval for  $k_{i_\phi}$

### Roll Loop (inner loop)

By following the procedure outlined in Sec. 6.3.1 in Beard & McLain, we get:

$$k_{p_\phi} = \frac{\delta_a^{max}}{e_\phi^{max}} \text{sign}(a_{\phi_2}) = \frac{30^\circ}{15^\circ} (-1) = -2$$

$$\omega_{n_\phi} = \sqrt{|a_{\phi_2}| \frac{\delta_a^{max}}{e_\phi^{max}}} \approx 1.14$$

$$k_{d_\phi} = \frac{2\zeta_\phi \omega_{n_\phi} - a_{\phi_1}}{a_{\phi_2}} \approx 1.94$$

where the damping ratio (design parameter) is chosen as  $\zeta_\phi = 0.707$ .

## Problem 2c) - Successive Loop Closure

### Integral Gain in Roll loop

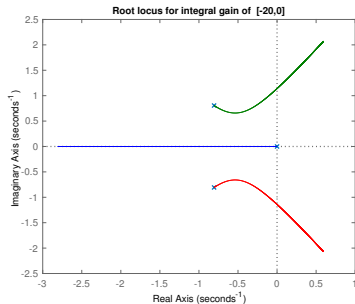


Figure: Root-Locus For Negative Gain

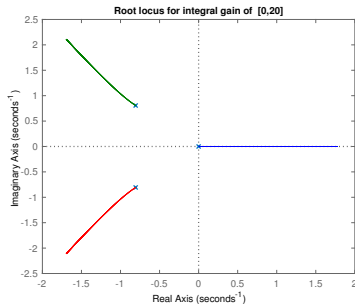


Figure: Root-Locus For Positive Gain

- The feasible interval for the gain is  $k_{i\phi} \in [-3.21, 0]$ .



## Problem 2c) - Successive Loop Closure

### Course Loop (outer loop)

- The inner loop is now approximated as a DC-gain of 1.
- The bandwidth should be at least 5 times lower than that of the inner loop.
- Choosing a bandwidth ten times slower than in the roll loop and a damping ratio of 1 yields:

$$\omega_{n_\chi} = \frac{1}{W_\chi} \omega_{n_\phi} = \frac{1}{10} \cdot 1.14 = 0.114$$

$$k_{p_\chi} = 2\zeta_\chi \omega_{n_\chi} \frac{V_g}{g} = 2 \cdot 1 \cdot \frac{1.14}{10} \cdot \frac{161.11}{9.81} \approx 3.74$$

$$k_{i_\chi} = \frac{\omega_{n_\chi}^2 V_g}{g} = \left( \frac{1.14}{10} \right)^2 \cdot \frac{161.11}{9.81} \approx 0.21$$

## Problem 2d) - Integral gain in the roll loop

*Can you come up with a reason for why it in some cases may be beneficial to remove the integral action from the roll loop?*

- The first thing you should do is ask yourself if integral action is necessary or not.
- In general, you don't want to add integral action unless you need it - can lead to delay and instability in your system.

## Integrator in roll loop

The transfer function from  $\phi^c$  to  $\phi$  without integral action can be shown to be:

$$\frac{\phi}{\phi^c}(s) = \frac{k_{p\phi} a_{\phi_2}}{s^2 + (a_{\phi_1} + k_{d\phi} a_{\phi_2})s + k_{p\phi} a_{\phi_2}}$$

For a constant set point  $\phi^c(s) = \frac{1}{s}\bar{\phi}$ , it can be shown using the final value theorem that

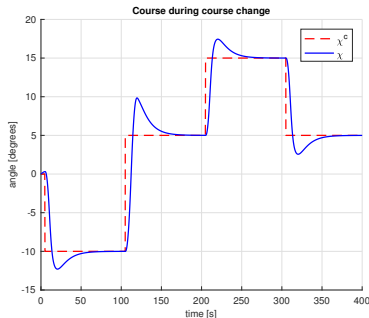
$$\lim_{t \rightarrow \infty} \phi(t) = \lim_{s \rightarrow 0} s\phi(s) = \bar{\phi}$$

⇒ Integral action is not necessary as we don't have any disturbance in the inner roll loop.

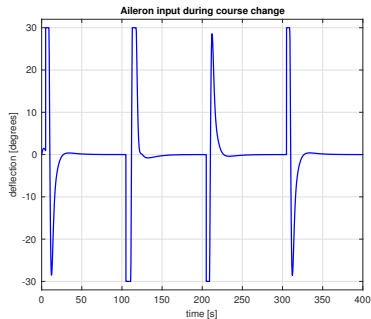
- Also worth mentioning that we don't really "care" in this case if the roll angle deviates from the setpoint as long as the course angle converges.

## Problem 2e) - Simulation of Simplified System

*Present simulation results for the system in Figure 1 with course changing maneuvers: choose a series of steps (you may choose yourself the desired values).*



**Figure:** Course during course changing maneuvers.



**Figure:** Aileron input during course changing maneuvers.

## Problem 2f) - Simulation of the State-Space Model

Present simulation results for course changing maneuvers (choose the same desired values as in Problem 2d) with the complete state-space model.

- We replace the simplified roll dynamics with the full state-space model.
- The real coordinated-turn equation should be used instead of the simplified version.
- The lowpass filter is already a part of the state-space model (no need to add this in Simulink).

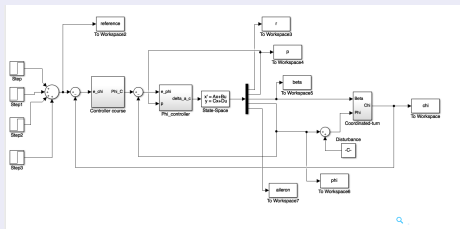


Figure: Overall system

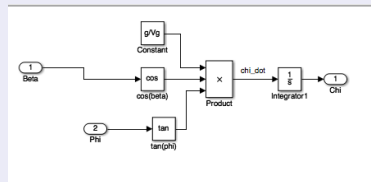
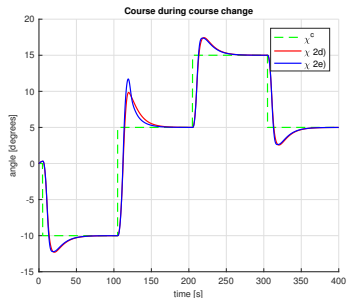
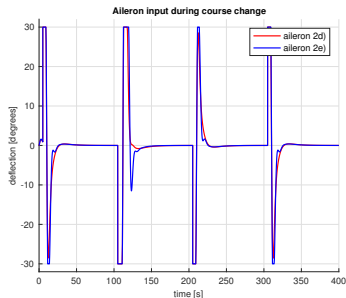


Figure: Coordinated-turn

## Problem 2f) - Simulation of State Space Model and Real Coordinated-Turn Equation



**Figure:** Course during course changing maneuvers.



**Figure:** Aileron during course changing maneuvers.

- The simplified model is, in this case, an accurate approximation of the "real" dynamics.
- Remember that the relationship between heading and roll actually is nonlinear.

## Problem 2g) - Integrator windup

*Would you claim that integrator windup is a problem in your simulations? If that is the case, can you propose a solution that would limit the problem?*

- When the input is saturated, the integrator will continue to accumulate error which can lead to large overshoots and consequently large delay
- There exists a number of different anti-windup schemes
- The easiest one to implement is to turn off the integrator by setting  $k_{i_x} = 0$  when the input goes into saturation. This can be achieved with a simple switching mechanism in Simulink

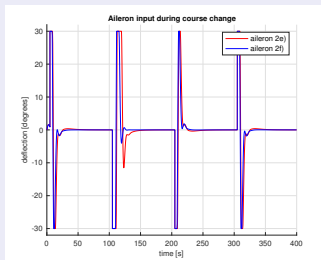
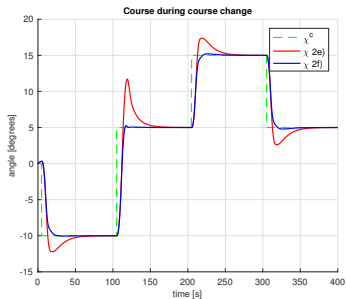
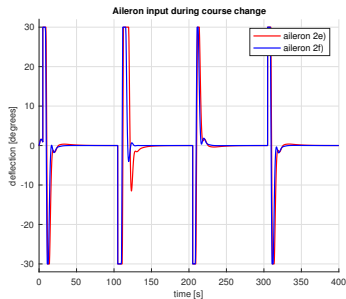


Figure: Aileron during course changing maneuvers.

## Problem 2g) - Integrator windup



**Figure:** Course during course changing maneuvers.



**Figure:** Aileron during course changing maneuvers.

- How well you are able to control windup will also depend on the size of the step error.
- By using a simple switching mechanism it is possible to avoid overshoot.
- If you have overshoot and the input is saturated, you should have some sort of proposal/action to handle windup.