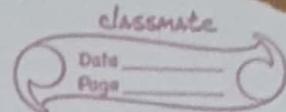


Single Phase AC Circuit Analysis



Comparison of AC and DC

(in this note)

1) AC:

- Voltage and current change periodically.
- Low cost production.
- A.C. voltage can be lowered or raised by using capacitor or choke.
- AC can be converted into D.C. by using converter.
- Generally frequency 50Hz.
- more dangerous than DC.

2) D.C.

- Voltage and current remain constant.
- Higher cost production.
- D.C. current can be decreased by using the resistance.
- DC can be converted into AC by inverter
- Frequency is zero '0'.
- It is less dangerous than AC.

1) # AC Voltage

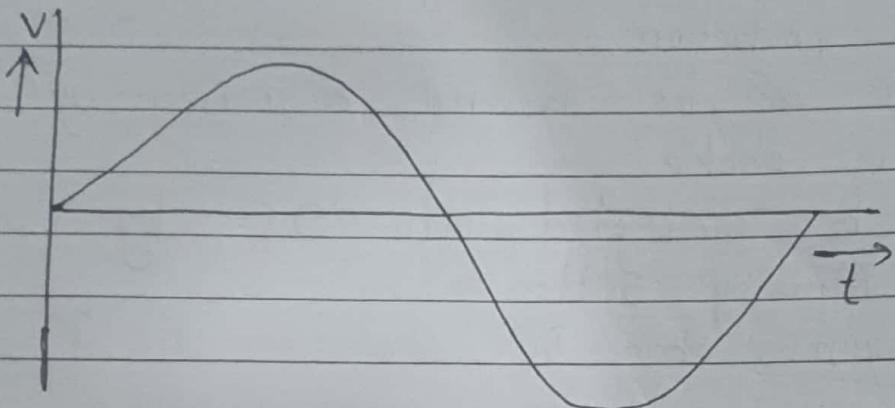
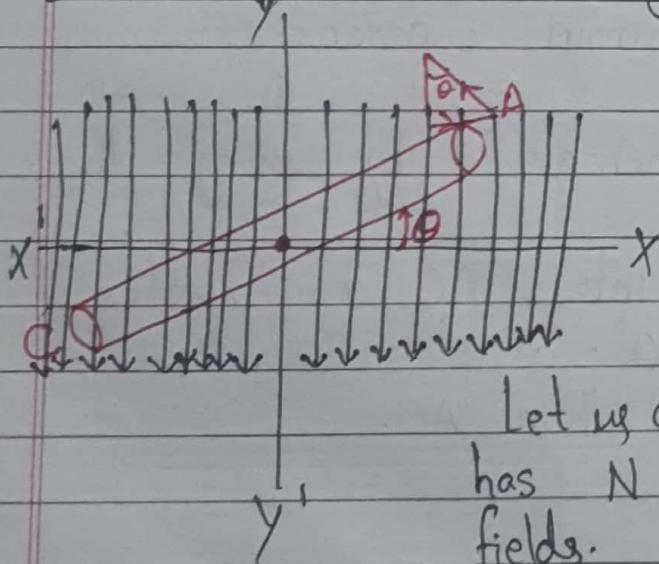
- An alternating voltage is any voltage that varies both in magnitude and polarity with respect to time.

2) # AC Current

- Any current that varies both in magnitude & direction w.r.t. time.

Eqn of alternating voltage & current# Sinusoidal Waveform:

The waveform traced out in graph of V versus t has a shape termed sinusoidal so, the wave is referred as sine - wave.

Generation of alternating voltage and current:-

Let us consider rectangular coil AC which has N turn's and rotating in magnetic fields.

Let the angle be measured from Horizontal position.

where,

B = flux density (Wb/m^2)

l = length of the coil (m)

v = linear velocity (m/sec)

The two sides A and C move parallel to line of magnetic flux. So, no flux ~~is~~ cut the line,

that hence, no emf is generated in the line when the coil is turn with angle θ , then its velocity is resolved into two parts:

- ① $V \cos\theta$ component \rightarrow parallel to direction of the magnetic flux.
- ② $V \sin\theta$ component \rightarrow perpendicular to direction of the magnetic flux.

③ The emf is generated ~~not~~ entirely due to the perpendicular component i.e. $v \sin\theta$.

Now,

emf is generated in one side of coil which contains N conductor,

$$e = N \times Blv \sin\theta \quad \text{--- (i)}$$

Total emf generated in both sides of coil is

$$e = 2NBlv \sin\theta \quad \text{--- (ii)}$$

$$e = E_m \quad [\text{when } E_m = e \text{ and } \theta = 90^\circ]$$

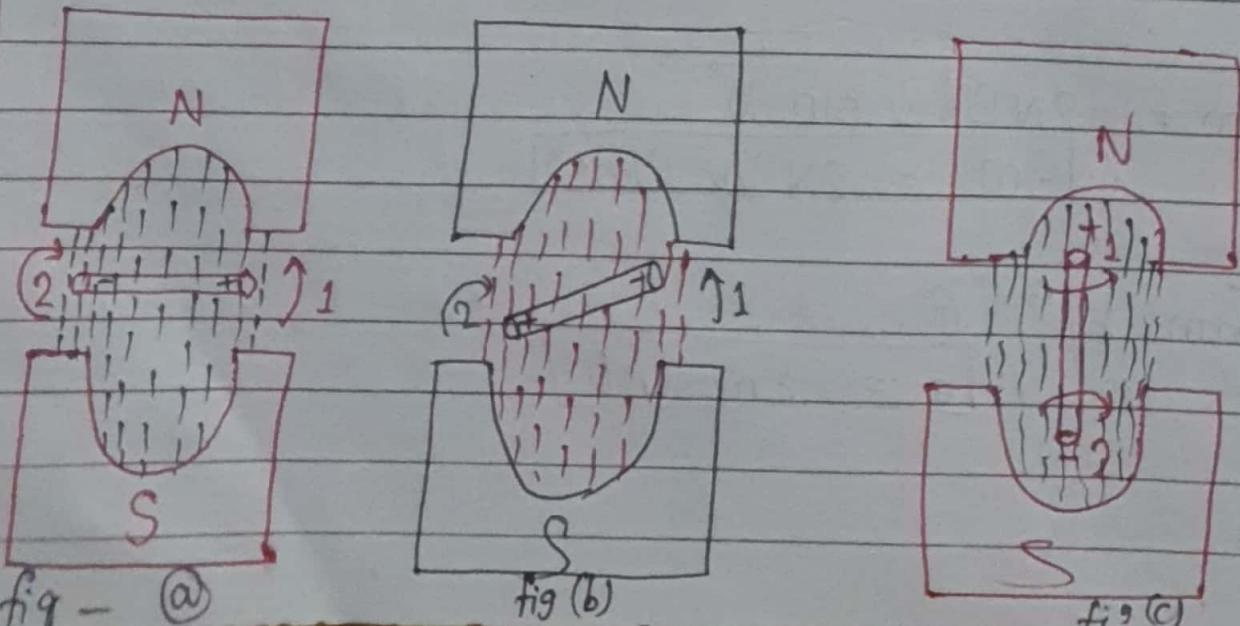
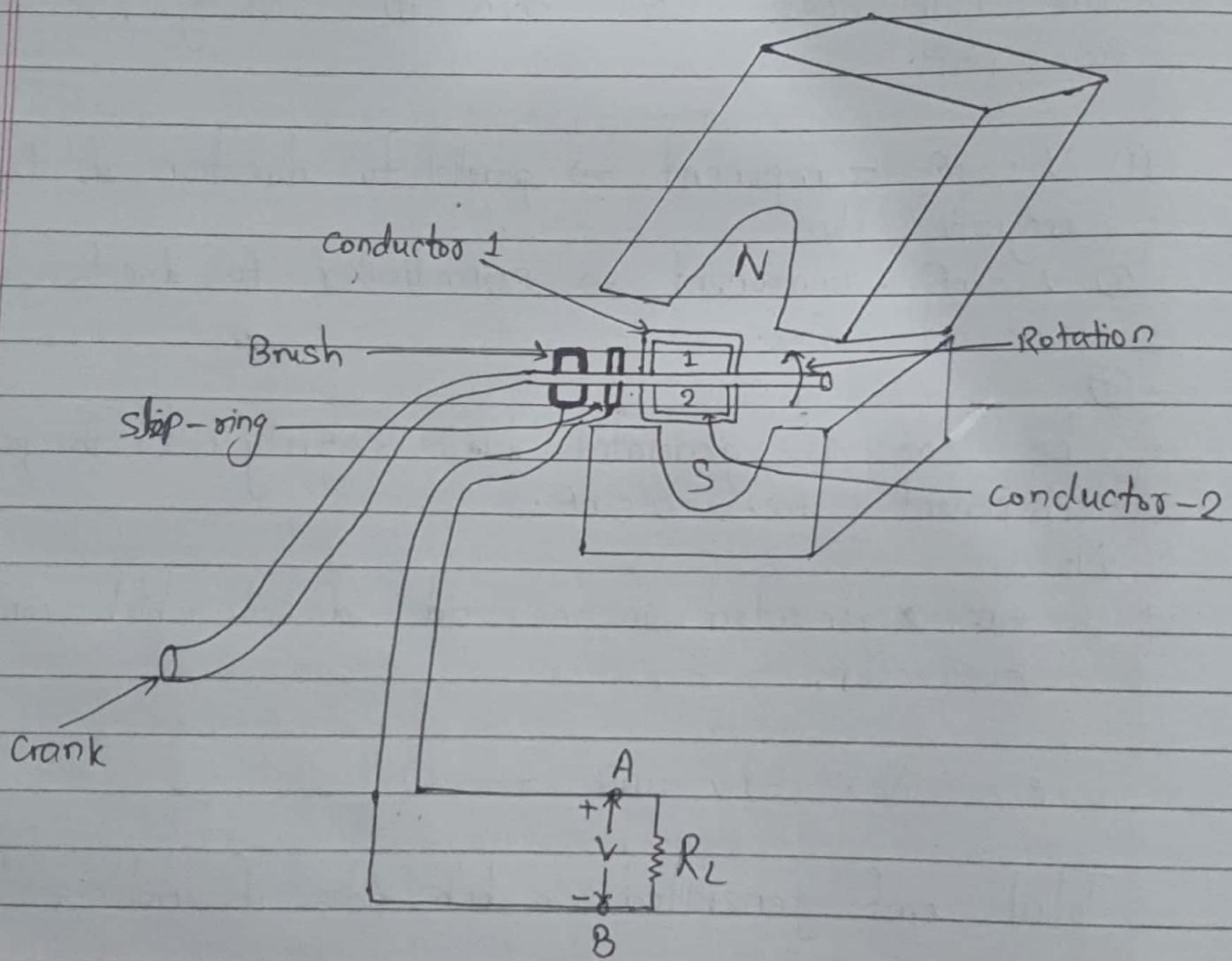
$$E_m = 2NBl v \sin 90^\circ$$

$$\therefore E_m = 2NBlv \text{ (Volts)}$$

from eqn (ii),

$$e = E_m \sin\theta$$

Generation of AC Voltage:



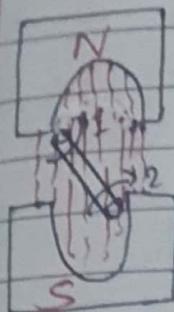
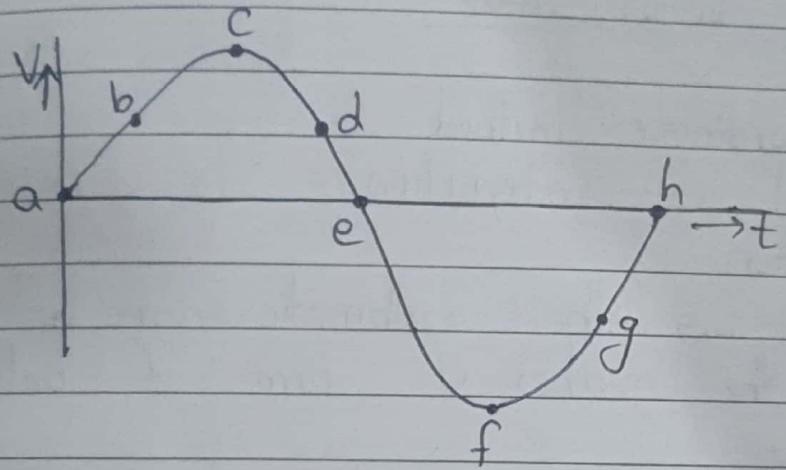


fig (d)



2014 # Sinusoidal function terminology:

- 1) Wave form.
- 2) Instantaneous value
- 3) Amplitude or peak value
- 4) Peak to peak value .
- 5) Frequency
- 6) Phase angle
- 7) Phase angle

• Waveform

→ path traced by a quantity.

• Average value or mean value

R.M.S. value or effective value of virtual value
 ↳ root mean square value

1) Average value or mean value

formula
is reqd



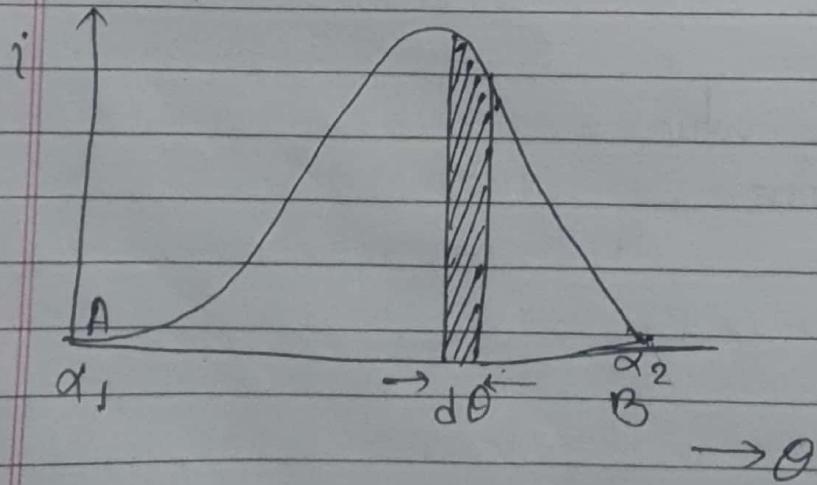
a) Mid-ordinate method

b) Method of Integration

* Method of Integration

Average value is the arithmetic sum of all the values divided by the number of sum of values.

The average value of any cycle of waveform is area under the wave divided by the time period (length of cycle).



Area of the strips = $i d\theta$

The area under the curve and horizontal axis will be the sum of all such strip.

Curve limits from

$$\theta = \alpha_1 \text{ to } \theta = \alpha_2$$

The length of base of curve = $\alpha_2 - \alpha_1$

$$\text{Total area of curve} = \int_{\alpha_1}^{\alpha_2} i d\theta$$

The average height of the curve = Area under the curve

$$= \frac{1}{\alpha_2 - \alpha_1} \int_{\alpha_1}^{\alpha_2} i d\theta$$

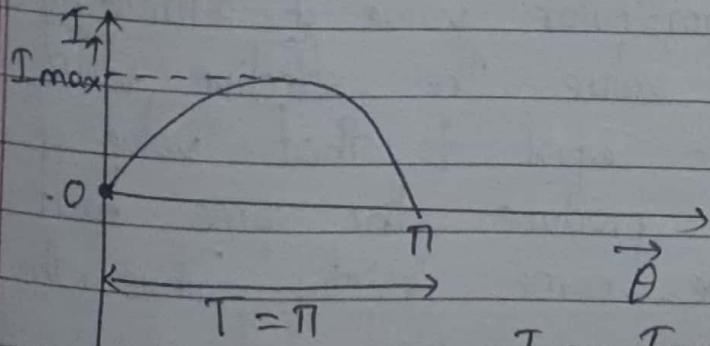
The average value of alternating current i over period T is given by

$$I_{av} = \frac{1}{T} \int_0^T i d\theta \rightarrow \text{in terms of current}$$

$$V_{av} = \frac{1}{T} \int_0^T v d\theta \rightarrow \text{in terms of voltage}$$

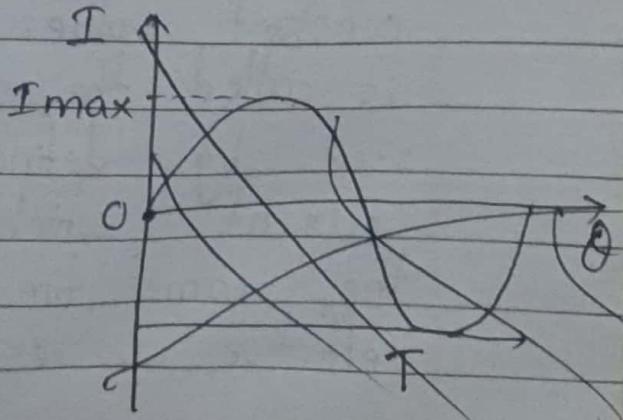
$$P_{av} = \frac{1}{T} \int_0^T p d\theta \rightarrow \text{in terms of power}$$

Average Value of half-wave Sinusoidal wave form:



$$I = I_{max} \cdot \sin\theta$$

$$V = V_{max} \cdot \sin\theta$$



$$\begin{aligned}
 I_{avg} &= \frac{1}{\pi} \int_0^{\pi} i d\theta \\
 &= \frac{1}{\pi} \int_0^{\pi} I_{max} \sin \theta d\theta \\
 &= \frac{I_{max}}{\pi} \int_0^{\pi} \sin \theta d\theta \\
 &= \frac{I_{max}}{\pi} \left[-\cos \theta \right]_0^{\pi} \\
 &= \frac{I_{max}}{\pi} [-\cos \pi + \cos 0] \\
 &= \frac{I_{max}}{\pi} [-(-1) + 1] = \frac{I_{max}}{\pi} [1+1] \\
 &= \frac{2 I_{max}}{\pi} \\
 \therefore I_{avg} &= 0.637 I_{max}
 \end{aligned}$$

~~by formula~~ Effective or R.M.S. Value or Virtual Value:
 Root Mean Square Value

An alternating current varies from instant to instant if it is necessary to determine an equivalent direct current that will be produced the same heat in the same interval of time, in the same resistor as a alternating currents. In other words, we measure the a.c. in terms of d.c. which has constant value. This particular value of alternating current is called the effective value. The effective value of a.c. by definition is equal to that value of direct current which will produce the same heat in the same time in the same resistor. Thus, the effective value is d.c. equivalent to a.c.

Two methods:

- ① Mid - ordinate method
- ② By Integration method

② By Integration method

Heat produced by an alternating current of Instantaneous value i in resistor R in time dt is $= i^2 R dt$.

Total heat produced in one cycle [i.e. in time T] is given by

$$H_{a.c.} = \int_0^T i^2 R dt$$

Heat produced by an equivalent direct current I in resistor R in time T is given:

$$\text{or, } I^2 RT = \int_0^T i^2 R dt$$

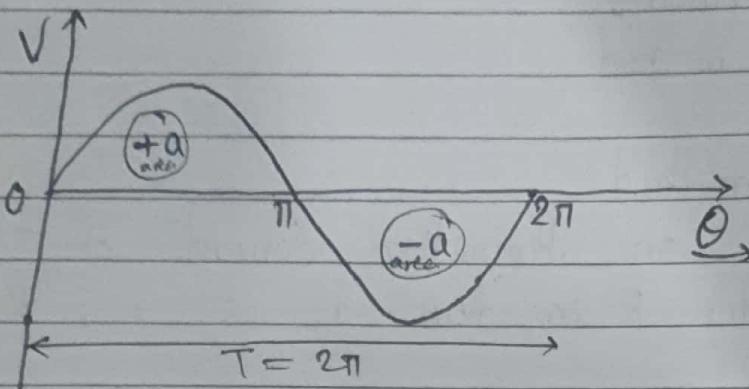
$$\therefore \text{or, } I^2 = \frac{1}{T} \int_0^T i^2 dt$$

$$\therefore I = \sqrt{\frac{1}{T} \int_0^T i^2 dt}$$

$$V = \sqrt{\frac{1}{T} \int_0^T v^2 dt}$$

in terms of voltage.

Average Value of a Sinusoidal wave form:



$$\text{Avg. value} = \frac{+a + (-a)}{2\pi} = 0$$

By Integration Method:

$$\begin{aligned}
 \text{Avg. value} &= \frac{1}{T} \int_0^T V d\theta \\
 &= \frac{1}{2\pi} \int_0^{2\pi} V_{\max} \sin \theta d\theta \quad [\because V = V_{\max} \sin \theta] \\
 &= \frac{1}{2\pi} \int_0^{2\pi} V_{\max} \cdot \sin \theta d\theta \\
 &= \frac{V_{\max}}{2\pi} \left[\sin \theta - \cos \theta \right]_0^{2\pi} \\
 &= \frac{V_{\max}}{2\pi} \times 0 \quad [\because \cos 2\pi = 0]
 \end{aligned}$$

$$\therefore \text{Avg. value} = 0$$

R.M.S. value of Sinusoidal Wave form:-

$$\text{V}^2 = \frac{1}{T} \int_0^T v^2 d\theta$$

$$\Rightarrow V = \sqrt{\frac{1}{T} \int_0^T v^2 d\theta}$$

$$V_{rms} = \sqrt{\frac{1}{2\pi} \int_0^{2\pi} V_{max}^2 \cdot \sin^2 \theta d\theta}$$

$$V_{rms} = \sqrt{\frac{V_{max}^2}{4\pi} \int_0^{2\pi} (1 - \cos 2\theta) d\theta}$$

$$V_{rms} = \sqrt{\frac{V_{max}^2}{4\pi} \left[\theta - \frac{\sin 2\theta}{2} \right]_0^{2\pi}}$$

$$V_{rms} = \frac{V_{max}}{2} \sqrt{\frac{1}{\pi} \left[2\pi - \frac{\sin 4\pi}{2} \right]}$$

$$V_{rms} = \frac{V_{max}}{2} \sqrt{\frac{1}{\pi} \times 2\pi}$$

$$V_{rms} = \frac{V_{max}}{\sqrt{2}}$$

∴ $V_{rms} = 0.707 V_{max}$ (in terms of voltage)
Also,

$$I_{rms} = 0.707 I_{max}$$
 (in terms of current)

Form factor (K_f)

The ratio of R.M.S. value to average value.

$$\begin{aligned} K_f &= \frac{\text{R.M.S. value}}{\text{Avg. value}} \\ &= \frac{0.707 V_{\max}}{0.637 V_{\max}} \\ \therefore K_f &= 1.11 \end{aligned}$$

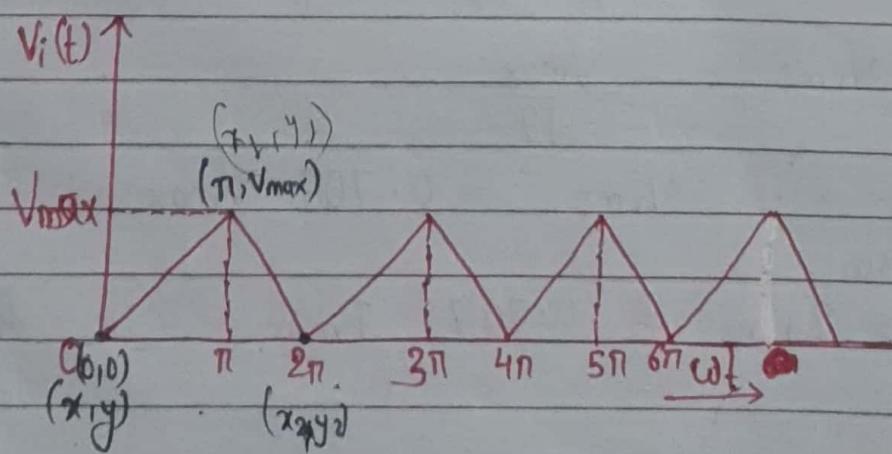
Peak factor (K_p) / Amplitude factor / Crest factor:

It is defined as the ratio of maximum value to R.M.S. value.

$$\begin{aligned} K_p &= \frac{\text{max}^m \text{ value (Peak value)}}{\text{R.M.S. value}} \\ &= \frac{\text{max}^m \text{ value}}{0.707 \cdot \text{max}^m \text{ value}} \\ \therefore K_p &= 1.414 \end{aligned}$$

problems

1) Q. Find the avg. value or R.M.S. value of given waveform.



$$V_{avg} = \frac{1}{T} \int_0^T v \cdot d(\omega t)$$

$$V_{rms} = \sqrt{\frac{1}{T} \int_0^T v^2 d(\omega t)}$$

solution :-

$$V_{avg} = \frac{1}{T} \int_0^T v \cdot d(\omega t)$$

$$= \frac{1}{2\pi} \left[\int_0^{\pi} \frac{V_{max}(\omega t)}{\pi} \cdot d(\omega t) + \int_{\pi}^{2\pi} \left(-\frac{V_{max} \cdot \omega t}{\pi} + 2V_{max} \right) d(\omega t) \right]$$

$$= \frac{1}{2\pi} \left[\frac{V_{max} \cdot \omega^2}{\pi} \int_0^{\pi} t dt \right] -$$

$$\frac{V_{max} \cdot \omega^2}{\pi} \int_{\pi}^{2\pi} t dt$$

$$+ 2V_{max} \cdot \omega \int_{\pi}^{2\pi} dt \right]$$

$$= \frac{1}{2\pi} \cdot V_{max}$$

Rough

$$x_1 = 0, y_1 = 0$$

$$x_2 = \pi, y_2 = V_{max}$$

$$y - y_1 = \frac{y_2 - y_1}{x_2 - x_1} (x - x_1)$$

$$y = \frac{V_{max}}{\pi} x$$

$$\Rightarrow V_i(t) = \frac{V_{max}}{\pi} (\omega t)$$

Again,

$$x_1 = \pi, y_1 = V_{max}$$

$$x_2 = 2\pi, y_2 = 0$$

So,

$$y - V_{max} = -\frac{V_{max}}{\pi} (x - \pi)$$

$$\Rightarrow V_i(t) - V_{max} = -\frac{V_{max}}{\pi} (\omega t - \pi)$$

$$\Rightarrow V_i(t) = -\frac{V_{max} \cdot \omega t}{\pi} + \frac{V_{max}}{\pi} \pi + V_{max}$$

$$\Rightarrow V_i(t) = -\frac{V_{max} \cdot \omega t}{\pi} + 2V_{max}$$

$$= V_{max} \left[-\frac{\omega t}{\pi} + 2 \right]$$

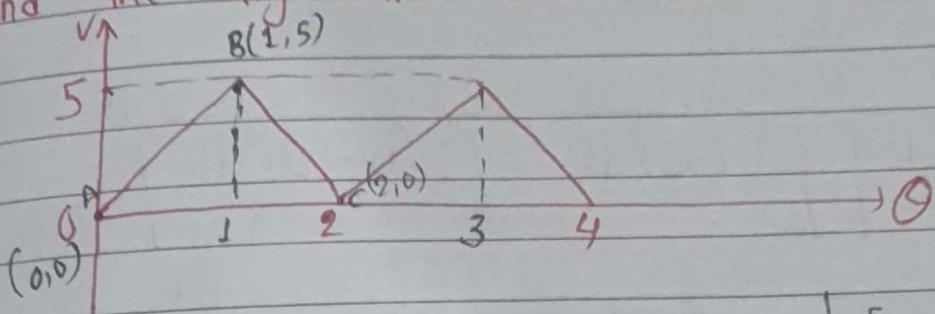
$$\begin{aligned}
 V_{avg} &= \frac{1}{2\pi} \left[\int_0^{\pi} V_{max}(\omega t) \cdot d(\omega t) + \int_{\pi}^{2\pi} -\frac{V_{max}}{\pi} (\omega t - 2\pi) d(\omega t) \right] \\
 &= \frac{V_{max}}{2\pi^2} \left[\int_0^{\pi} \omega t \cdot d(\omega t) - \int_{\pi}^{2\pi} (\omega t - 2\pi) d(\omega t) \right] \\
 &= \frac{V_{max}}{2\pi^2} \left[\left[\frac{(\omega t)^2}{2} \right]_0^{\pi} - \left[\frac{(\omega t)^2}{2} \right]_{\pi}^{2\pi} + 2\pi \left[\omega t \right]_{\pi}^{2\pi} \right] \\
 &= \frac{V_{max}}{2\pi^2} \left[\left[\frac{\pi^2}{2} - \frac{0}{2} \right] - \left[\frac{4\pi^2}{2} + \frac{\pi^2}{2} + 2\pi(2\pi - \pi) \right] \right] \\
 &= \frac{V_{max}}{2\pi^2} \left[-\frac{2\pi^2}{2} + 2\pi^2 \right] \\
 &= \frac{V_{max}}{2}
 \end{aligned}$$

Avg. value of the a.c. voltage is given = $\frac{V_{max}}{2}$
Now,

RMS Value for the given a.c. waveform:

$$\begin{aligned}
 V_{rms} &= \sqrt{\frac{1}{2\pi} \left[\int_0^{\pi} V_i(t)^2 \cdot d(\omega t) + \int_{\pi}^{2\pi} V_i(t)^2 \cdot d(\omega t) \right]} \\
 &= \sqrt{\frac{1}{2\pi} \left[\frac{V_{max}^2}{\pi^2} \left\{ \left[\frac{\omega t}{3} \right]^3 \right\}_{0}^{\pi} + \frac{1}{2\pi} \cdot -\frac{V_{max}^2}{\pi^2} \int_{\pi}^{2\pi} (\omega t)^2 - 2\omega t \cdot 2\pi + 4\pi^2 \right]} \\
 &= \frac{V_{max}}{\pi} \sqrt{\frac{1}{2\pi} \left\{ \frac{\pi^3}{3} - 0 - \frac{1}{3} \left[(\omega t)^3 - 4\pi(\omega t)^2 + 4\pi^2 \omega t \right]_{\pi}^{2\pi} \right\}} \\
 &= \frac{V_{max}}{\pi} \sqrt{\frac{\pi^2}{3}} = \frac{V_{max}}{\sqrt{3}}
 \end{aligned}$$

2. Q. Find the Avg. value and RMS. Value :



So,

$$\text{Avg. } V_{\text{avg.}} = \frac{1}{T} \int_0^T V d\theta$$

$$= \frac{1}{2} \left[\int_0^1 5\theta d\theta + \int_1^2 (-50+10)\theta d\theta \right]$$

$$= \frac{1}{2} \left[5 \left[\frac{\theta^2}{2} \right]_0^1 - 5 \left(\frac{\theta^2}{2} \right)_1^2 + 10[\theta]_0^2 \right]$$

$$= \frac{1}{2} \left[\frac{5}{2} - \frac{15}{2} + 10 \right]$$

$$= \frac{1}{2} \times \left[\frac{5-15+20}{2} \right]$$

$$= \frac{1}{2} \times \frac{10}{2}$$

$$= \frac{1}{2} \times 5$$

$$= 2.5V$$

For AB,

$$x_1 = 0, x_2 = 1$$

$$y_1 = 0, y_2 = 5$$

So,

$$y - 0 = 5(x - 0)$$

$$\Rightarrow y = 5x$$

$$\Rightarrow V = 5\theta$$

From BC,

$$x_1 = 1, y_1 = 5$$

$$x_2 = 2, y_2 = 0$$

$$y - y_1 = \frac{y_2 - y_1}{x_2 - x_1} (x - x_1)$$

$$\text{or, } y - 5 = \frac{-5}{2-1} (x - 1)$$

$$\Rightarrow y - 5 = -5x + 5$$

$$\Rightarrow y = -5x + 10$$

$$\Rightarrow V = -5\theta + 10$$

$$V_{rms} = \sqrt{\frac{1}{T} \int_0^T v^2 d\theta}$$

$$= \sqrt{\frac{1}{2} \left[\int_0^1 (5\theta)^2 d\theta + \int_1^2 (10 - 5\theta)^2 d\theta \right]}$$

$$= \sqrt{\frac{1}{2} \left[25 \left[\frac{\theta^3}{3} \right]_0^1 + \int_1^2 (100 - 100\theta + 25) d\theta \right]}$$

$$= \sqrt{\frac{1}{2} \left[25 \times \frac{1}{3} + 100 \left[\theta^2 \right]_1^2 - 100 \cdot \frac{\left[\theta^2 \right]_1^2}{2} + 25 \left[\frac{\theta^3}{3} \right]_1^2 \right]}$$

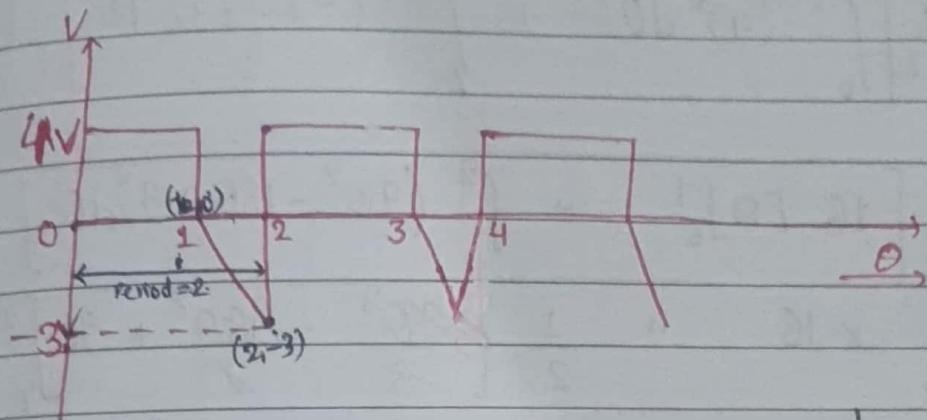
$$= \sqrt{\frac{1}{2} \left[\frac{25}{3} + 100 - 150 + \frac{175}{3} \right]}$$

$$= \sqrt{\frac{1}{2} \times \frac{50}{3}}$$

$$= \sqrt{\frac{25}{3}}$$

$$= \frac{5}{\sqrt{3}}$$

3) find the avg. value, r.m.s. value and form factor of given wave form shown below:



solution :-

Here, 0 to 1 , voltage = 4V
1 to 2 , voltage = -3V

$$x_1 = 1, y_1 = 0$$

$$x_2 = 2, y_2 = -3$$

Now,

$$\begin{aligned} \text{Avg. value} &= \frac{1}{T} \int_0^T V \cdot d\theta \\ &= \frac{1}{2} \left[\int_0^1 4 d\theta + \int_1^2 (-3\theta + 3) d\theta \right] \\ &= \frac{1}{2} \left[4 \cdot [0]_0^1 + -3 \left[\frac{\theta^2}{2} \right]_1^2 + 3 [0]_1^2 \right] \\ &= \frac{1}{2} \left[4 \times 1 - 3 \times \frac{2^2}{2} + 3 \times \frac{1^2}{2} + 3 \times 2 - 3 \times 1 \right] \\ &= \frac{1}{2} \left[4 - 6 + \frac{3}{2} + 6 - 3 \right] \\ &= \frac{1}{2} \left[\frac{8 + 3 - 6}{2} \right] \\ &= \frac{5}{4} = 1.25 \text{ V} \end{aligned}$$

Rough

$$y - y_1 = \frac{y_2 - y_1}{x_2 - x_1} (x - x_1)$$

$$y - 0 = \frac{-3 - 0}{2 - 1} (x - 1)$$

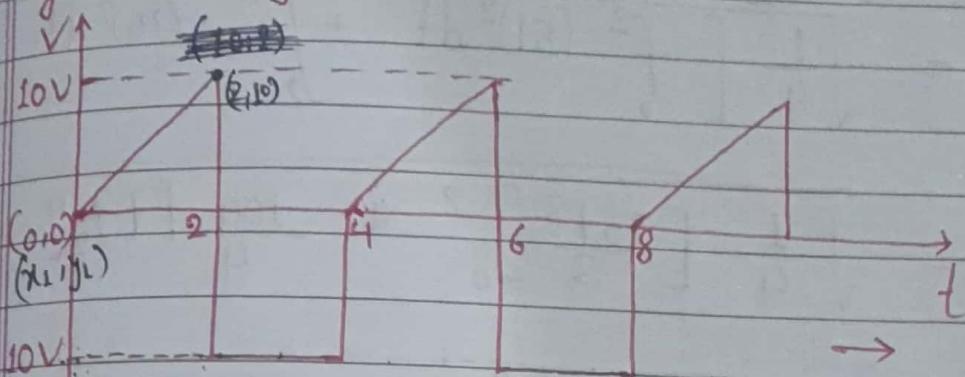
$$\Rightarrow y = -3x + 3$$

$$\Rightarrow V = -3\theta + 3$$

$$\begin{aligned}
 \text{RMS. value} &= \sqrt{\frac{1}{T} \int_0^T v^2 d\theta} \\
 &= \sqrt{\frac{1}{2} \left[\int_0^1 (4)^2 d\theta + \int_1^2 (-3\theta + 3)^2 d\theta \right]} \\
 &= \sqrt{\frac{1}{2} \left[16 [\theta]_0^1 + \int_1^2 (9\theta^2 - 18\theta + 9) d\theta \right]} \\
 &= \sqrt{\frac{1}{2} \times 16 + \frac{1}{2} \left[\frac{3\theta^3}{3} - \frac{18\theta^2}{2} + 9\theta \right]_1^2} \\
 &= \sqrt{8 + \frac{1}{2} [3 \times 2^3 - 3 \times 1^3 - 9 \times 2^2 + 9 \times 1^2 + 9 \times 2 - 9 \times 1]} \\
 &= \sqrt{8 + \frac{1}{2} [24 - 3 - 36 + 9 + 18 - 9]} \\
 &= \sqrt{8 + \frac{1}{2} \times 9} \\
 &= \sqrt{\frac{19}{2}} = 3.08
 \end{aligned}$$

$$\begin{aligned}
 \text{Form factor (K_f)} &= \frac{\text{rms value}}{\text{avg. value}} \\
 &= \frac{3.08}{1.25} \\
 &= 2.464
 \end{aligned}$$

4) find the avg. value, rms value & form factor of the given triangular waveform.

PV
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$$0 \leq t \leq 2$$

$$2 \leq t \leq 4$$

Soln :-

$$x_1 = 0, y_1 = 0$$

$$x_2 = 2, y_2 = 10$$

Now,

$$\text{Avg. value} = \frac{1}{T} \int_0^T V dt$$

$$= \frac{1}{4} \left[\int_0^2 5t \cdot dt + \int_{02}^4 10 dt \right]$$

$$= \frac{1}{4} \left[\left[\frac{5t^2}{2} \right]_0^2 + 10[t]^4_2 \right]$$

$$= \frac{1}{4} \left[\frac{5 \times 2^2}{2} + 10 \times 4 - 10 \times 2 \right]$$

$$= \frac{1}{4} [10 + 40 - 20]$$

$$= \frac{30}{4}$$

$$= 7.5 V$$

Rough

~~$y = x$~~

~~$y = \frac{10}{2}(x-0)$~~

$$y = 5x$$

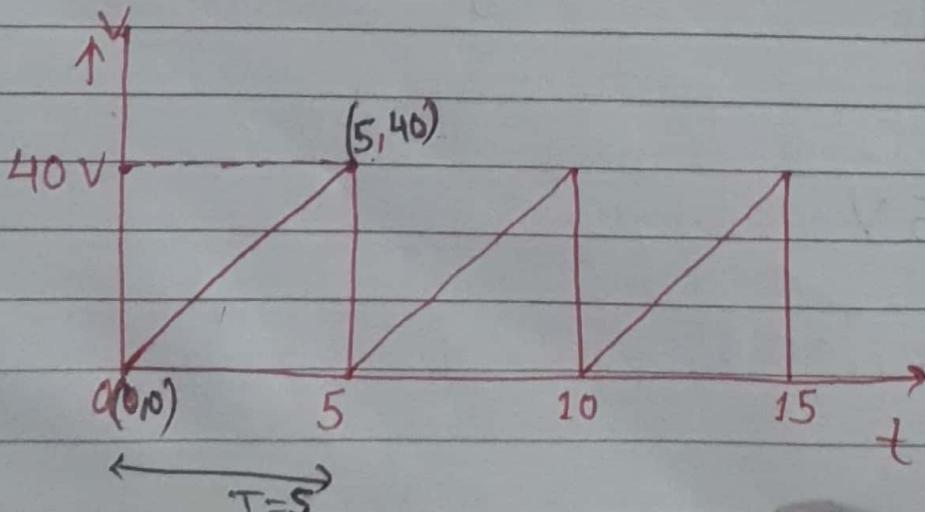
$$V = 5t - ①$$

$$V = 10 - ②$$

$$\begin{aligned}
 \text{R.M.S. Value} &= \sqrt{\frac{1}{T} \int_0^T v^2 dt} \\
 &= \sqrt{\frac{1}{4} \left[\int_0^2 (5t)^2 dt + \int_2^4 (10)^2 dt \right]} \\
 &= \sqrt{\frac{1}{4} \left[\frac{25t^3}{3} \Big|_0^2 + \frac{100t}{4} \Big|_2^4 \right]} \\
 &= \sqrt{\frac{1}{4} \times \frac{25 \times 2^3}{3} + 25 \times 4 - 25 \times 2} \\
 &\quad \cancel{+ \frac{50}{3}} \\
 &= \cancel{7.67} \quad 8.16
 \end{aligned}$$

$$\begin{aligned}
 \text{Form factor } (K_f) &= \frac{\text{rms value}}{\text{avg. value}} \\
 &= \frac{8.16}{7.5} \\
 &= 1.08
 \end{aligned}$$

- 5) Compute the form factor of a given sawtooth wave if
 PU 2019
 vimp
- (triangular wave).



Sol :-

First method:

$$x_1 = 0, y_1 = 0$$

$$x_2 = 5, y_2 = 40$$

$$y - 0 = \frac{840}{5}x - 0$$

Now,

$$\begin{aligned} \text{Avg. value} &= \frac{1}{T} \int_0^T v dt \\ &= \frac{1}{5} \int_0^5 8t dt \\ &= \frac{1}{5} \left[\frac{8t^2}{2} \right]_0^5 \\ &= \frac{1}{5} \times 4 \times 5^2 \\ &= 20 V \end{aligned}$$

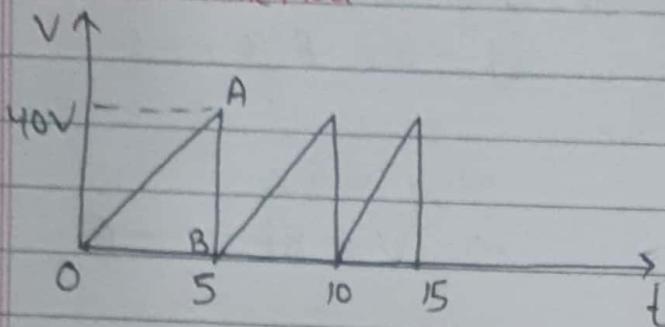
Now,

$$\begin{aligned} \text{-RMS value} &= \sqrt{\frac{1}{T} \int_0^T v^2 dt} \\ &= \sqrt{\frac{1}{5} \int_0^5 64t^2 dt} \\ &= \sqrt{\frac{1}{5} \left[64 \times \frac{t^3}{3} \right]_0^5} = \sqrt{\frac{1}{5} \left[64 \times \frac{5^3}{3} \right]} \end{aligned}$$

$$\begin{aligned} &= \sqrt{\frac{1}{5} \times 250} \\ &= 23.09 V \end{aligned}$$

$$\begin{aligned} \text{Form factor (kf)} &= \frac{\text{rms value}}{\text{avg. value}} \\ &= \frac{23.09}{20} \\ &= 1.173 \end{aligned}$$

Second method:



$$y = mx + c$$

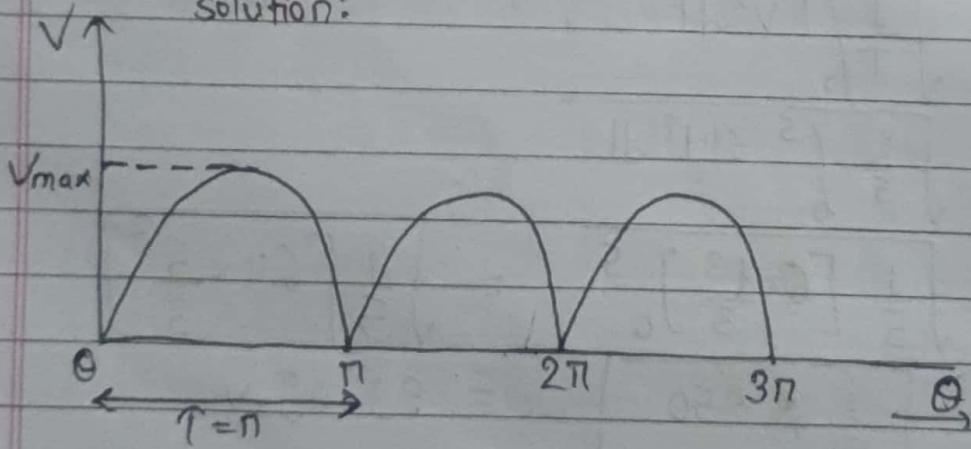
$$= \frac{40}{5}x + c$$

$$\left[m = \frac{\text{height}}{\text{base length}} \right]$$

$$\therefore y = 8x$$

$$\Rightarrow V = 8t$$

- 6) The output voltage (full wave rectifier) shown below. find the average value and rms value (effective & virtual value). solution:



solution:

$$\text{Avg. value} = \frac{1}{T} \int_0^T V d\theta$$

$$= \frac{1}{\pi} \int_0^{\pi} V_{\max} \sin \theta d\theta$$

\because sinusoidal wave form
 $V = V_{\max} \sin \theta$

$$= \frac{V_{\max}}{\pi} \left[-\cos \theta \right]_0^{\pi} = \frac{V_{\max}}{\pi} [-\cos \pi + \cos 0]$$

$$\text{So, Avg. value} = \frac{V_{\max}}{\pi} \times 2$$

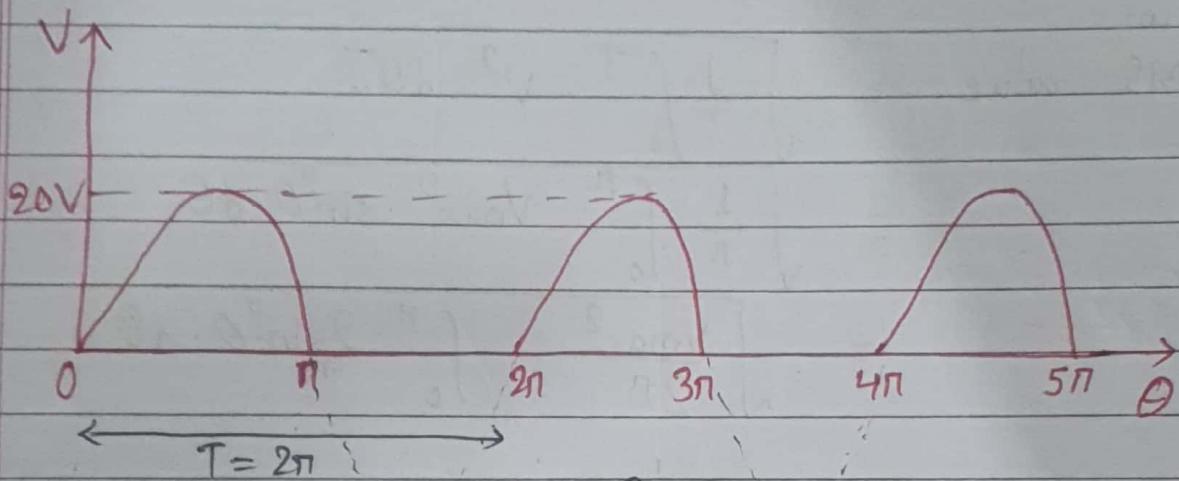
$$= 0.636 V_{\max}$$

Now,

$$\begin{aligned}
 \text{RMS value} &= \sqrt{\frac{1}{T} \int_0^T V^2 d\theta} \\
 &= \sqrt{\frac{1}{\pi} \int_0^{\pi} V_{\max}^2 \cdot \sin^2 \theta d\theta} \\
 &= \sqrt{\frac{V_{\max}^2}{2\pi} \int_0^{\pi} 2\sin^2 \theta \cdot d\theta} \\
 &= \sqrt{\frac{V_{\max}^2}{2\pi} \int_0^{\pi} (1 + \cos 2\theta) d\theta} \\
 &= \sqrt{\frac{V_{\max}^2}{2\pi} \left[\theta + \frac{\sin 2\theta}{2} \right]_0^{\pi}} \\
 &= \sqrt{\frac{V_{\max}^2}{2\pi} \left[\pi + \frac{\sin 2\pi}{2} - \frac{\sin 0}{2} \right]} \\
 &= \sqrt{\frac{V_{\max}^2}{2\pi} \times \pi} \\
 &= \frac{V_{\max}}{\sqrt{2}}
 \end{aligned}$$

$$= 0.707 V_{\max}$$

7) The output voltage (half wave rectifier) whose maximum value is 20V. find the avg. value, rms value (virtual or effective value), peak factor and form factor.



$$\begin{aligned}
 \text{Avg. value} &= \frac{1}{T} \int_0^T v \cdot d\theta \\
 &= \frac{1}{2\pi} \left[\int_0^\pi 20 \sin \theta \, d\theta + \int_{\pi}^{2\pi} 0 \cdot \sin \theta \, d\theta \right] \\
 &= \frac{1}{2\pi} \left[20 [-\cos \theta]_0^\pi + 0 \right] \\
 &= \frac{20}{2\pi} [-\cos \pi + \cos 0] \\
 &= \frac{20}{2\pi} \times 2 \\
 &= \frac{20}{\pi} = 6.37 \text{ V}
 \end{aligned}$$

$$\begin{aligned}
 \text{rms value} &= \sqrt{\frac{1}{T} \int_0^T v^2 \, d\theta} \\
 &= \sqrt{\frac{1}{2\pi} \int_0^\pi 400 \sin^2 \theta \, d\theta}
 \end{aligned}$$

$$\begin{aligned}
 &= \frac{1}{2\pi f} \sqrt{\frac{200}{2\pi} \left[0 + \frac{\sin 2\theta}{2} \right]_0^\pi} \\
 &= \cancel{\frac{1}{2\pi f}} \sqrt{\frac{200}{2\pi} [\pi + 0 - 0 - 0]} \\
 &= \cancel{\frac{1}{2\pi f}} \sqrt{\frac{200\pi}{2\pi}} = \sqrt{100} \\
 &= \sqrt{\frac{200\pi}{4\pi^2}} = 10V \\
 &= 3.99
 \end{aligned}$$

Peak factor (k_p) = $\frac{\text{max}^m \text{ value}}{\text{rms}}$

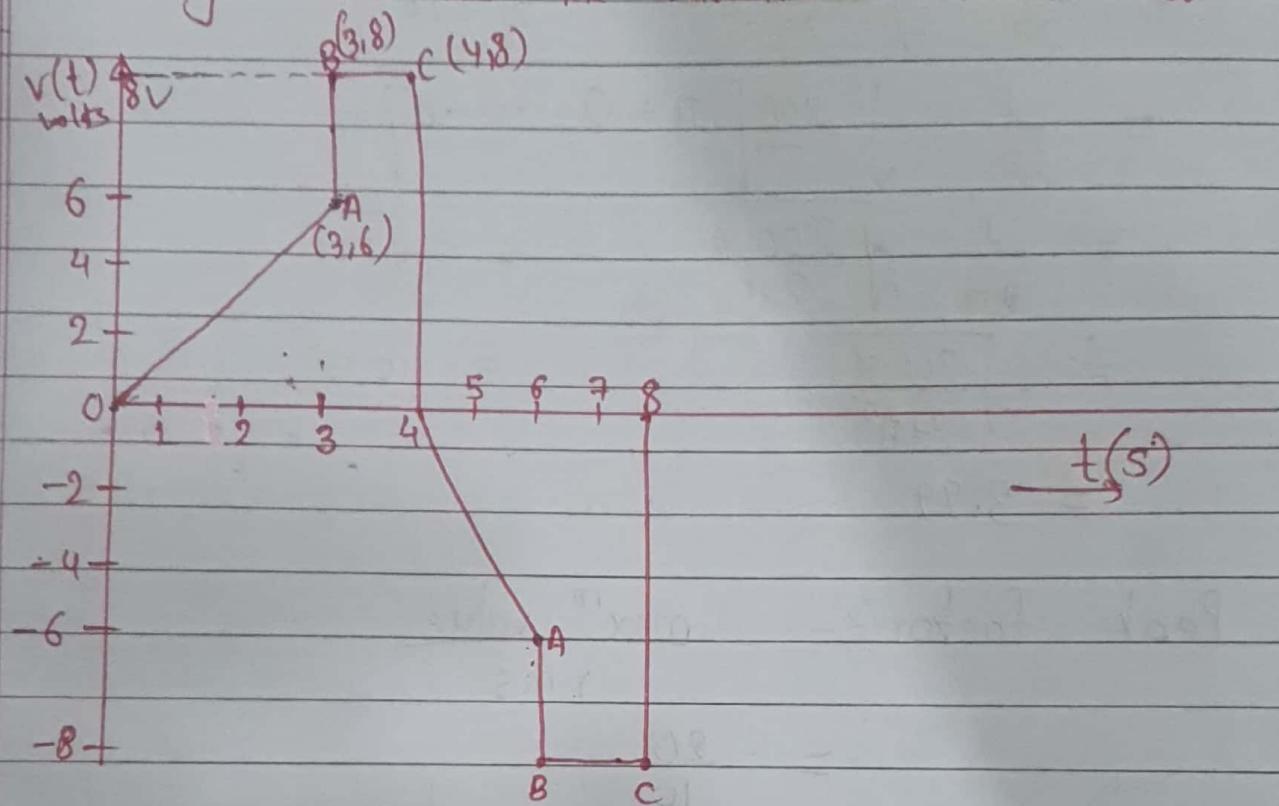
$$\begin{aligned}
 &= \frac{20}{10} \\
 &= 2
 \end{aligned}$$

form factor (k_f) = $\frac{\text{rms value}}{\text{avg. value}}$

$$\begin{aligned}
 &= \frac{10}{6.36} \\
 &= 1.572
 \end{aligned}$$

8) A voltage wave has the variation as shown below:

2016 fall
PU



solution: Case 1 - ~~(OA)~~ (OA)

$$x_1 = 0, y_1 = 0 \quad y = \frac{6}{3}x$$

$$x_2 = 3, y_2 = 6 \quad \Rightarrow y = 2x$$

$$\Rightarrow V = 2t \quad \text{--- (1)}$$

Case -2 (AB)

$$x_1 = 3, y_1 = 6$$

$$x_2 = 3, y_2 = 8$$

$$\Rightarrow y - 6 = \frac{2}{0}(x - 3)$$

$$\Rightarrow 0 = 2x - 6$$

Case -3 (BC)

$$x_1 = 3, y_1 = 8$$

$$x_2 = 4, y_2 = 8$$

So,

$$y - 8 = \frac{0}{1}(x - 3)$$

$$\Rightarrow y = 8$$

$$\Rightarrow V = 8$$

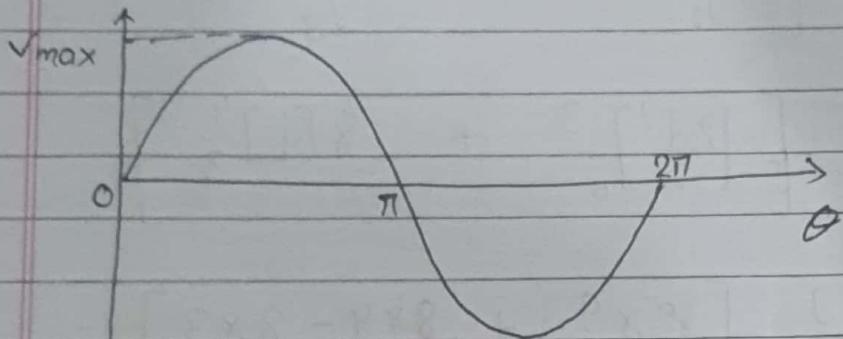
$$\begin{aligned}
 \text{Avg. value} &= \frac{1}{T} \int_0^T v \, dt \\
 &= \frac{1}{4} \left[\int_0^3 2t \cdot dt + \int_3^4 8 \, dt \right] \\
 &= \frac{1}{4} \left[\left[\frac{2t^2}{2} \right]_0^3 + 8[t]_3^4 \right] \\
 &= \frac{1}{4} \left[\frac{2 \times 9}{2} + 8 \times 4 - 8 \times 3 \right] \\
 &= \frac{1}{4} \times 17 \\
 &= 4.25V
 \end{aligned}$$

$$\begin{aligned}
 \text{RMS value} &= \sqrt{\frac{1}{T} \int_0^T v^2 \, dt} \\
 &= \sqrt{\frac{1}{4} \left[\int_0^3 4t^2 \cdot dt + \int_3^4 64 \, dt \right]} \\
 &= \sqrt{\frac{1}{4} \cdot \frac{4}{3} \left[\frac{t^3}{3} \right]_0^3 + \frac{1}{4} \times 64 \left[t \right]_3^4} \\
 &= \sqrt{\frac{3^3}{3} + 16 \times 4 - 16 \times 3} \\
 &= \sqrt{9 + 64 - 48} = \sqrt{25} \\
 &= 4.365V
 \end{aligned}$$

$$\begin{aligned}
 P_{avg} &= \frac{V_{rms}^2}{R} = \frac{5^2}{50} \quad [\text{given in ques}] \\
 &= 0.5 \text{ watt}
 \end{aligned}$$

Q9) Show that the peak factor of sine value is equal to 1.414.

solution:

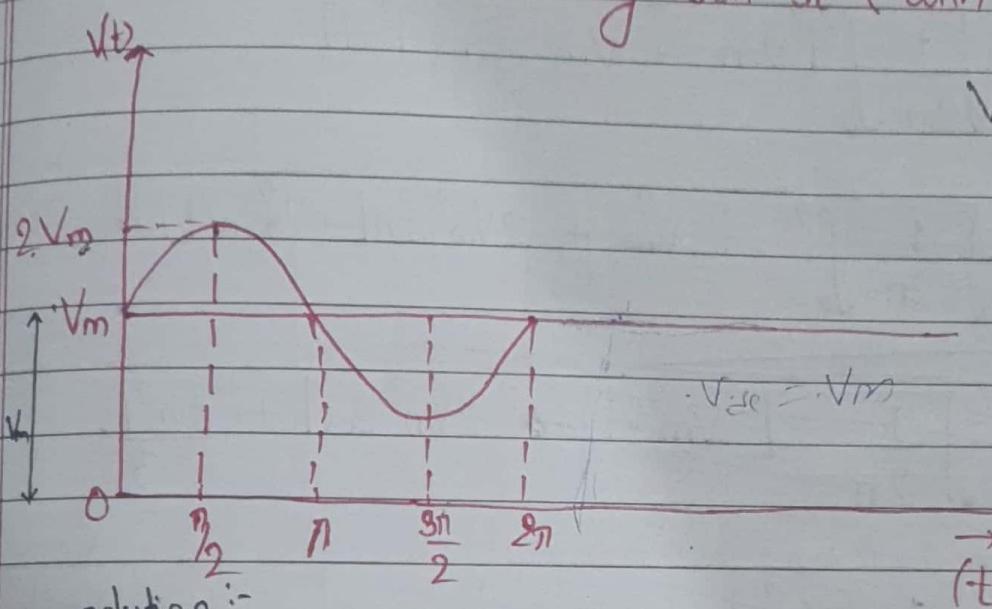


$$\begin{aligned}
 V_{rms} &= \sqrt{\frac{1}{T} \int_0^T V^2 d\theta} \\
 &= \sqrt{\frac{1}{2\pi} \int_0^{2\pi} V_{max}^2 \sin^2 \theta d\theta} \\
 &= 0.707 V_{max}
 \end{aligned}$$

$$\text{Peak value} = \frac{V_{max}}{V_{rms}}$$

$$\begin{aligned}
 &= \frac{V_{max}}{0.707 V_{max}} \\
 &= 1.414
 \end{aligned}$$

10) Determine the average value and rms value in which the ac waveform is ~~receiving~~ residing on dc (with a magnitude V_m).



$$V_{rms} = \sqrt{\frac{3}{2}} V_m$$

solution :-

Here the waveform eqⁿ for a given waveform is

for dc component:

$$V_{d.c.} = V_m$$

~~$V = V_{max} \cdot \sin t$~~

for ac component:

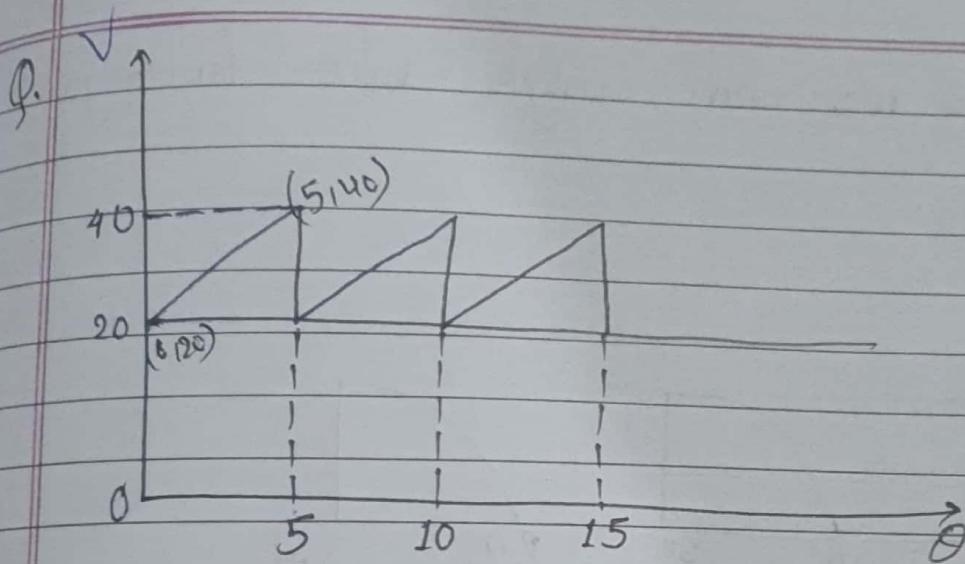
$$V_{ac} = V_m \sin t$$

So, that, 'total wave form eqⁿ' are

$$\begin{aligned} V &= V_{dc} + V_{ac} \\ &= V_m + V_m \sin t \quad 0 \leq t \leq 2\pi \end{aligned}$$

Now, avg. value for one complete cycle is

$$\begin{aligned} V_{avg} &= \frac{1}{2\pi} \int_0^{2\pi} (V_m + V_m \sin t) dt \\ &= \frac{1}{2\pi} \left[V_m t + V_m [-\cos t] \right]_0^{2\pi} + \frac{1}{2\pi} [V_m - \cancel{V_m}]_0^{2\pi} \\ &= \frac{1}{2\pi} \times 2\pi V_m - \frac{1}{2\pi} \times V_m [(-1 + 1)] = V_m - 0 = V_m \end{aligned}$$



d.c. component = 20 V

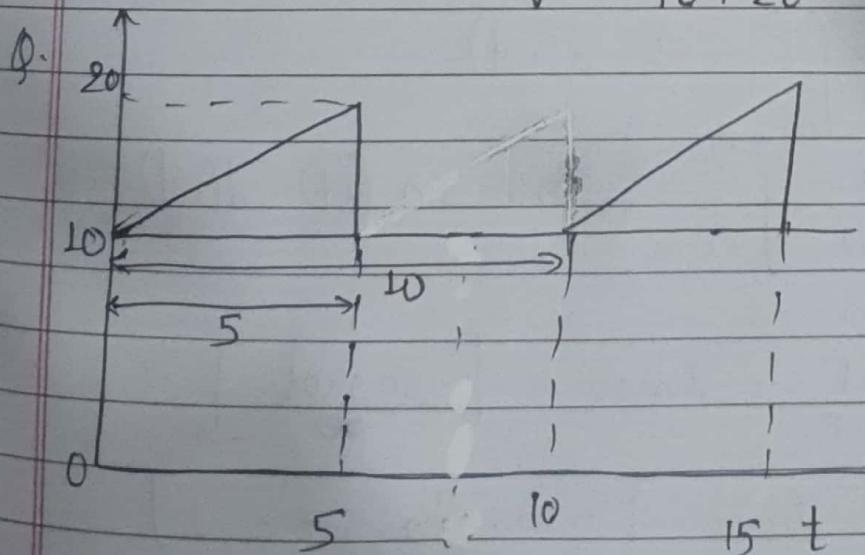
ac component = ~~8.40~~ [0 ≤ t ≤ 5]

$$\rightarrow y - 20 = \frac{40 - 20}{5}(x - 0)$$

$$y = 4x + 20$$

$$V = 40 + 20$$

$$\begin{cases} y = mx + c \\ V = \frac{\text{total ht.}}{\text{base l.}} x + c \\ V = \frac{20}{5}x + 20 \\ 5 = 4x + 20 \\ 5 = 4t \end{cases}$$



~~$dc = 10$~~

~~$ac = 2t$~~

$$V_{\max} = \frac{1}{T} \int_0^T V dt$$

$$= \frac{1}{10} \int_0^{10} (10 + 2t) dt$$

$$+ \int_5^{10} 10 dt$$

~~$ac = 2t + 10$~~

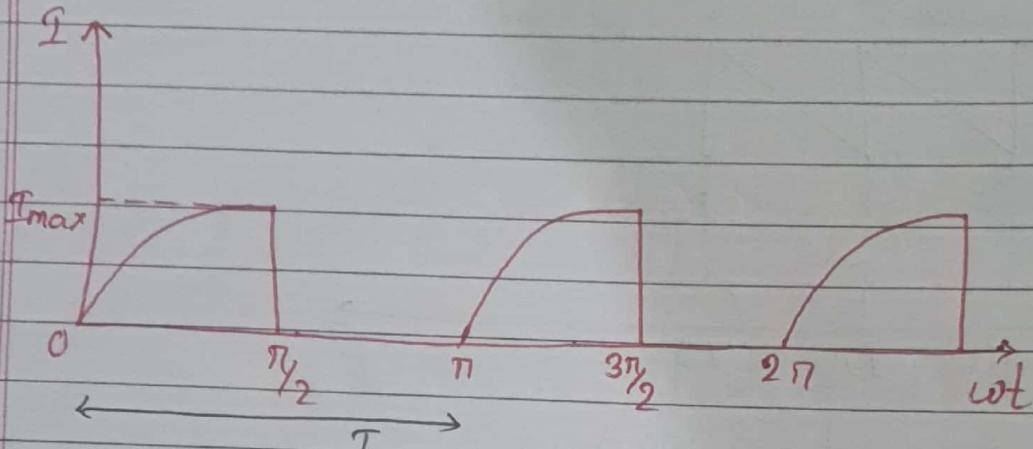
~~$dc = 10V \quad (0 \leq t \leq 10)$~~

~~$y = mx + c$~~

~~$= \frac{10}{5}x = \frac{\text{height}}{\text{base length}}$~~

~~$= 2x$~~

Q. Determine the rms and average value for a given waveform:



$$I = I_{\max} \cdot \sin \omega t \quad \text{--- (i) } (0 \leq \omega t \leq \frac{\pi}{2})$$

$$I = I_{\max} \cdot \sin \omega t \quad \text{--- (ii) } (\frac{\pi}{2} \leq \omega t \leq \pi)$$

So,

$$\begin{aligned} I_{\max} I_{\text{avg}} &= \frac{1}{T} \int_0^T I \, d(\omega t) \\ &= \frac{1}{\pi} \int_0^{\pi/2} I_{\max} \sin(\omega t) \, d(\omega t) \end{aligned}$$

$$= \frac{1}{\pi} \cdot I_{\max} \left[-\cos \omega t \right]_0^{\pi/2}$$

$$= \frac{I_{\max}}{\pi} \left[-\cos \frac{\pi}{2} + \cos 0 \right]$$

$$= \frac{I_{\max}}{\pi} [0 + 1]$$

$$I_{\text{avg}} = \frac{I_{\max}}{\pi}$$

$$I_{rms} = \sqrt{\frac{1}{\pi} \int_0^{\pi/2} (I_{max} \cdot \sin \omega t)^2 d(\omega t)}$$

$$= \sqrt{\frac{1}{\pi} \cdot I_{max}^2 \cdot \frac{1}{2} \int_0^{\pi/2} (1 - \cos 2(\omega t)) d(\omega t)}$$

$$= \sqrt{\frac{I_{max}^2}{2\pi}} \left[\omega t - \frac{\sin 2(\omega t)}{2} \right]_0^{\pi/2}$$

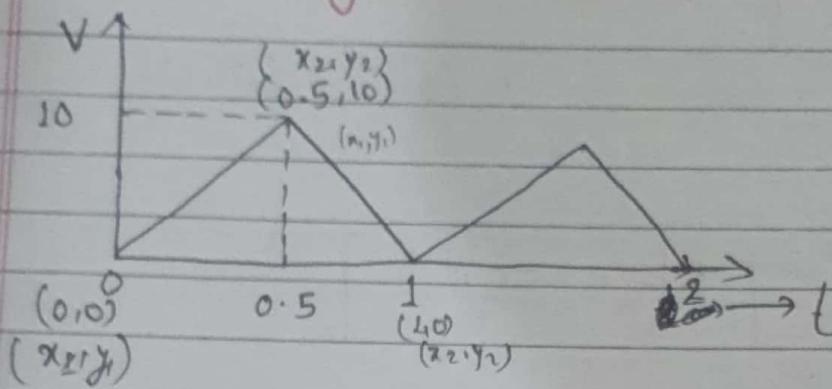
$$= \sqrt{\frac{I_{max}^2}{2\pi}} \left(\frac{\pi}{2} - \frac{0}{2} \right)$$

$$= \sqrt{\frac{I_{max}^2}{2\pi} \times \frac{\pi}{2}}$$

$$I_{rms} = \frac{I_{max}}{2}$$

2014 Fall

Q. Find the avg. value & rms value:



$$\text{avg.} = 50.5$$

$$\text{rms} = 5.77$$

$$y - y_1 = \frac{y_2 - y_1}{x_2 - x_1} (x - x_1)$$

$$\Rightarrow y = \frac{10}{0.5} x \\ = V = 20t$$

Now Also,

$$y - 10 = \frac{-10}{0.5} (x - 0.5)$$

$$\text{or } y - 10 = -20(x - 0.5)$$

$$\text{or } y = -20x + 10 + 10$$

$$\Rightarrow y = -20x + 20$$

$$\Rightarrow V = -20t + 20$$

Now,

$$\text{Avg. value} = \frac{1}{T} \int_0^T V dt$$

$$= \frac{1}{1} \left[\int_0^{0.5} 20t \cdot dt + \int_{0.5}^1 (-20t + 20) dt \right]$$

$$= \left[\frac{20t^2}{2} \right]_0^{0.5} + \left[\frac{-20t^2 + 20t}{2} \right]_0^{0.5}$$

$$= [10t^2]_0^{0.5} + [-10t^2 + 20t]_0^1$$

$$= (10 \times 0.5^2) + (-10 \times 1^2 + 20 \times 1 + 10 \times 0.5^2 - 20 \times 0.5)$$

$$= 2.5 - 10 + 20 + 2.5 - 10$$

$$= 5$$

Now,

$$\text{RMS value} = \sqrt{\frac{1}{T} \int_0^T v^2 dt}$$

$$= \sqrt{\frac{1}{1} \left[\int_0^{0.5} 400t^2 dt + \int_{0.5}^1 (20-20t)^2 dt \right]}$$

$$= \sqrt{\int_0^{0.5} 400t^2 dt + \int_{0.5}^1 (400 - 800t + 400t^2) dt}$$

$$= \sqrt{400 \left[\frac{t^3}{3} \right]_0^{0.5} + 400 \left[t - \frac{9t^2}{2} + \frac{t^3}{3} \right]_{0.5}^1}$$

$$= \sqrt{400 \left[\frac{0.5^3 - 0^3}{3} \right] + 400 \left[1 - 0.5 - \frac{9 \times 1^2}{2} + \frac{9 \times 0.5^2}{2} + \frac{1^3 - 0.5^3}{3} \right]}$$

$$= 20 \sqrt{\frac{4}{24} + 0.5, -1 + 0.25 + \frac{1}{3} - \frac{1}{24}}$$

$$= 20 \times \frac{1}{\sqrt{2}} \quad 5.77$$

$$= 1.67$$

A.C.

Civil low

In C, I leads V.

~~∴~~ V leads I in L.

Steady state response of pure ckt:

- 1) phase relation in a pure resistor
- 2) " " " " " inductor
- 3) " " " " " capacitor

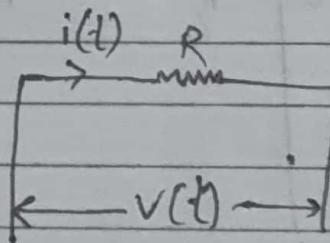
② Phase relation in a pure resistor

When a sinusoidal voltage of certain magnitude is applying to a resistor certain amount of sine-wave current passing through it. So, we know the relation of $V(t)$ & $i(t)$.

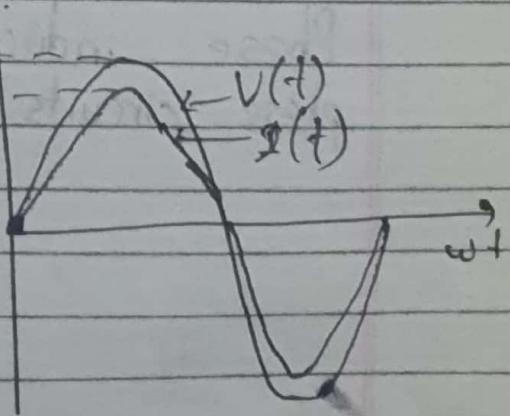
$$V(t) = i(t) \cdot R$$

Consider the function,

$$i(t) = I_{\max} \cdot \sin \omega t$$



Both voltage & current in the phase but amplitude is different according to the value of resistor.



Voltage & currents said to phase since phase is difference is zero.

$$Z = R$$

2) Phase relation in a pure ~~resistor~~ inductor:

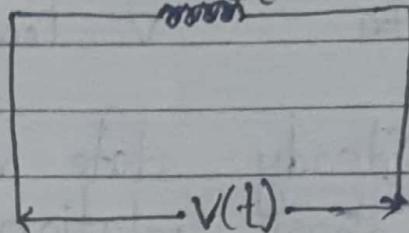
$$v(t) = L \cdot \frac{di}{dt}$$

$$L \frac{di}{dt}$$

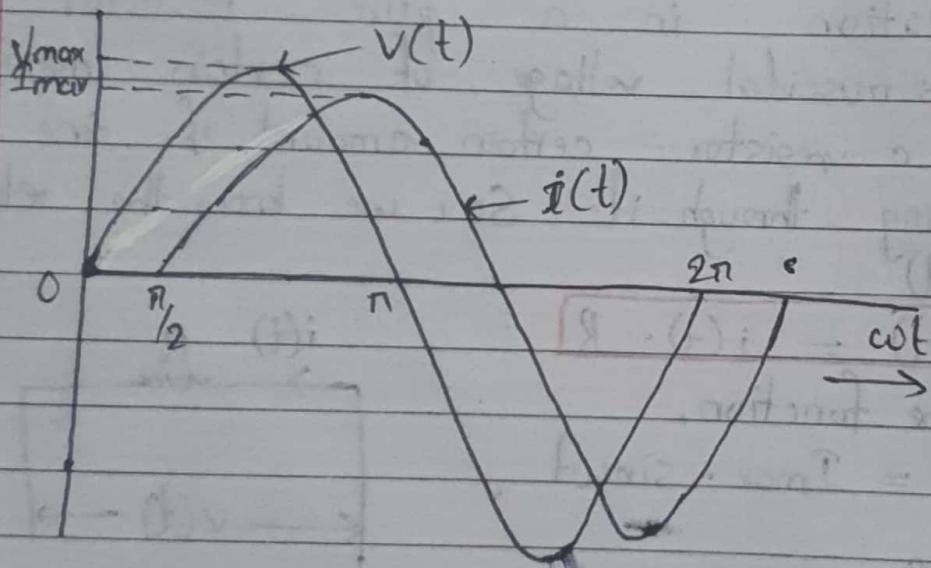
$$i(t) = I_{\max} \cdot \sin(\omega t)$$

$$v(t) = L \cdot \frac{d}{dt}(I_{\max} \sin \omega t)$$

$$v(t) = \omega L \cdot I_{\max} \cos \omega t$$



$$\therefore V(t) = V_{\max} \cdot \cos \omega t$$



Phase inductor voltage and current are out of phase circuits lags behind the voltage by 90°

3) Phase resistor in a pure capacitor:

$$V(t) = \frac{1}{C} \int i dt \quad \textcircled{1} \quad [\text{The relation bet' voltage & current}]$$

$$i(t) = I_{\max} \cdot \sin \omega t$$

$$V(t) = \frac{1}{C} \int I_{\max} \cdot \sin \omega t d(\omega t)$$

$$= \frac{1}{\omega C} I_{\max} (-\cos \omega t)$$

$$= \frac{I_{\max} \sin(\omega t - \pi/2)}{\omega C}$$

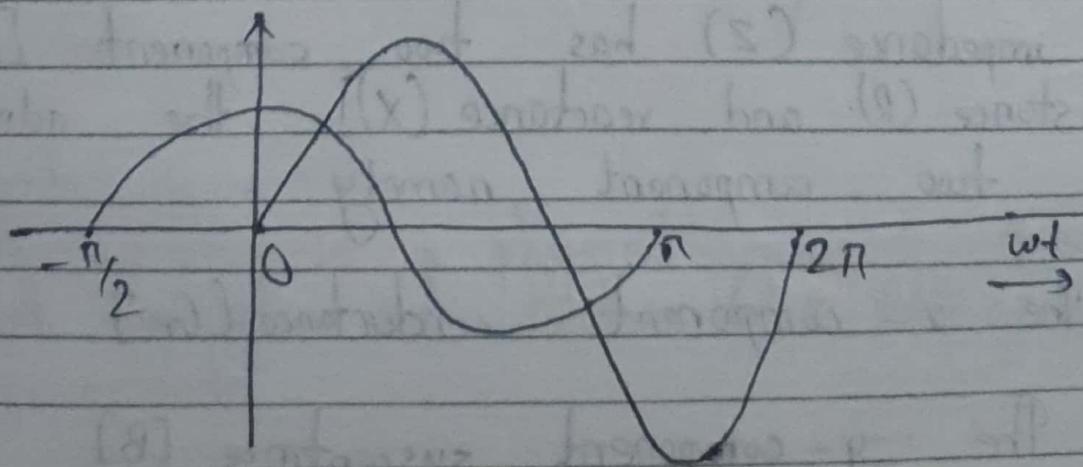
$$\therefore V(t) = V_{\max} \cdot \sin(\omega t - \pi/2)$$

where, $V_{\max} = \frac{I_{\max}}{\omega C}$

$$\Rightarrow \frac{V_{\max}}{I_{\max}} = Z$$

$$\therefore Z = \frac{1}{\omega C}$$

$$\text{Impedance (Z)} = X_C = \frac{1}{\omega C}$$



1) Polar form $\rightarrow e^{j\theta}$

2) Rectangular form $\rightarrow x + jy$
 $4 + j5 \quad ?$ rectangular form
 $r = 6.40 \quad ?$ polar form
 $\theta = 51.34 \quad ?$

Polar to rectangular form :

- Rec $(6.40, 51.34) \Rightarrow x = 4, y = 5$
- $\Rightarrow 4 + j5$

$$\frac{2.13 \angle 6.4 \angle 51.34}{5 \angle 5^\circ} = 2.13 \angle 46.34^\circ$$

$$\frac{5 + j6}{8 + j10} = \frac{7.81 \angle 50.19^\circ}{12.81 \angle 51.34^\circ}$$

X Admittance (Y)

It is the reciprocal of impedance (Z) of an a.c. circuit. So,

$$Y = \frac{1}{Z}$$

Thus, As impedance (Z) has two components [via resistance (R) and reactance (X)], the admittance has also two component namely

i) The x - component conductance [G]

ii) The y - component susceptance [B]

$$\text{Conductance } (G) = Y \cos \phi = \frac{1}{Z} \therefore \frac{R}{Z} = \frac{R}{Z^2}$$

$$\& \text{ Susceptance } (B) = Y \sin \phi = \frac{1}{Z} \cdot \frac{X}{Z} = \frac{X}{Z^2}$$

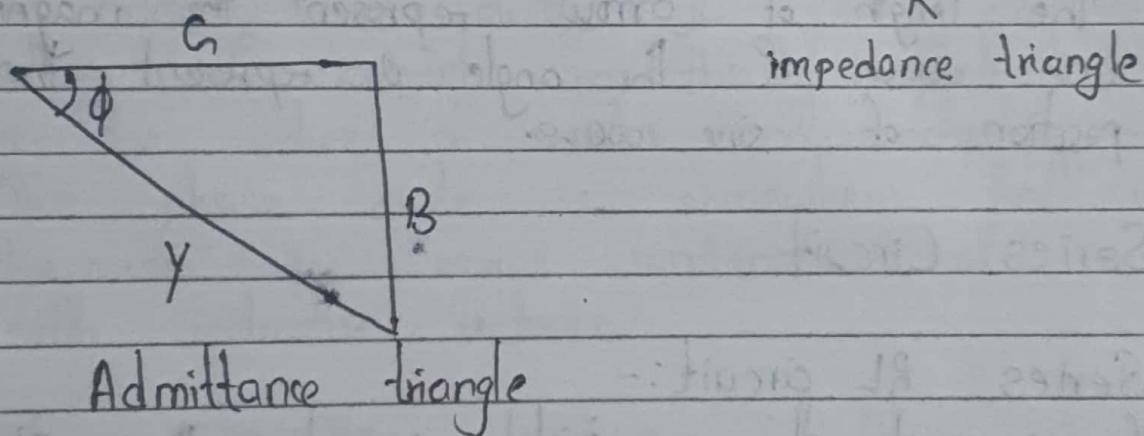
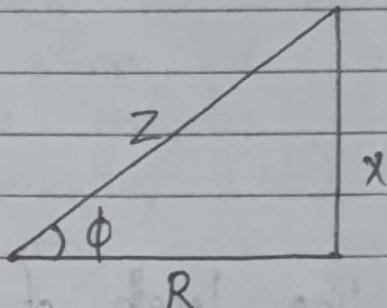
$$\therefore \text{ Admittance } (Y) = \sqrt{G^2 + B^2}$$

The unit's of

$$G = \text{mho}$$

$$B = \text{ohm}^{-1}$$

$$Y = \text{simon's}$$



Phasor diagram

Imp A phasor diagram can be used to represent a sine-wave in terms of magnitude and its angular position.

Vector diagram

A vector is a physical quantity that have the both magnitude and direction (X, Y, Z) or polar co-ordinate

Phasor diagram:
E.g.

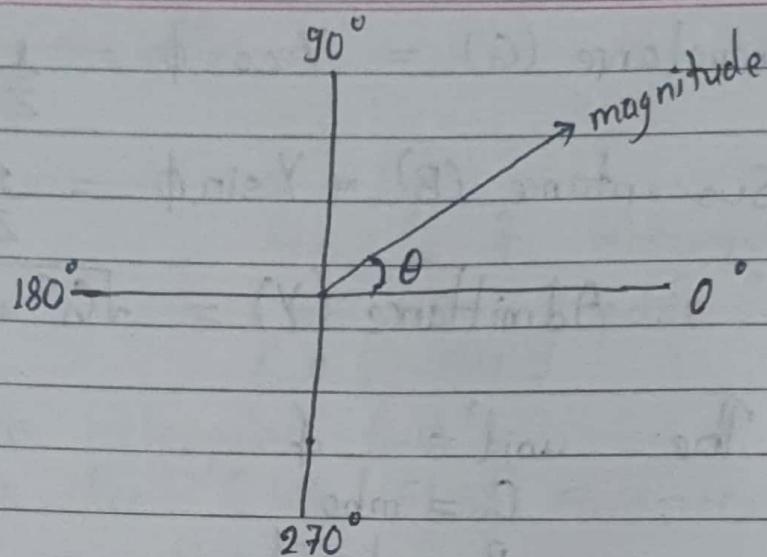


fig (a)

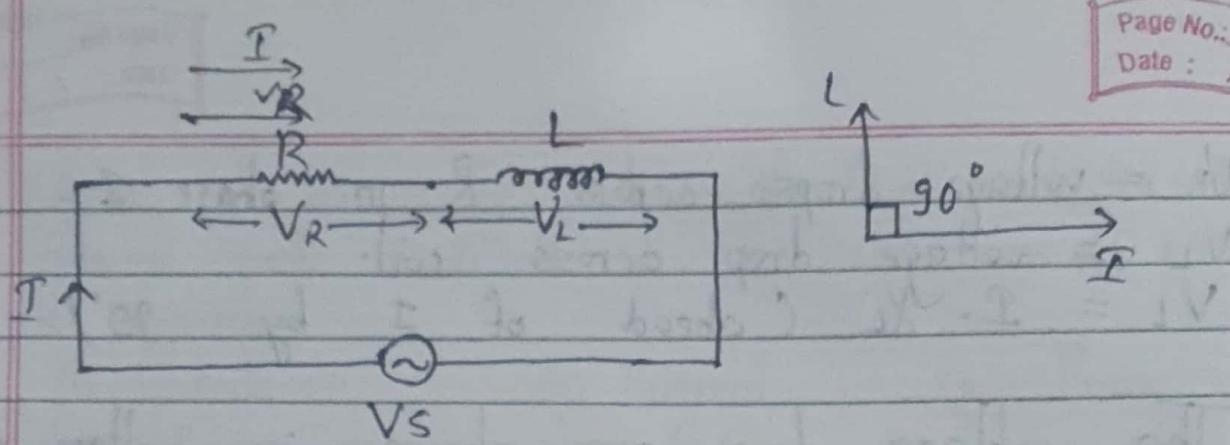
The length of arrow represent the magnitude of sine wave & the angle θ represent the angular position of sine wave.

Series Circuit

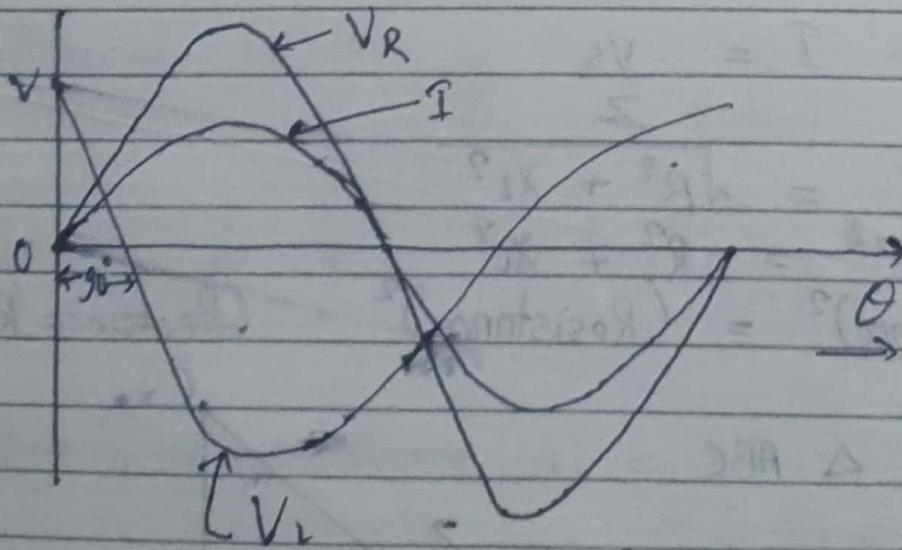
1) Series RL circuit:-

If we apply the sinusoidal input in the RL ckt, all the current in the elements & all the voltage across the elements are sinusoidal. In the analysis of RL ckt we find

- i) impedance
- ii) current
- iii) phase angle
- iv) Voltage drops



The resistor voltage (V_R) and the current (I) are in same phase with each other but lag behind the source voltage (V_s). The inductor voltage V_L leads the source voltage (V_s). The phase angle between current and voltage in the inductor is always 90° . The amplitude of voltage & current in the complete ckt depend on the value of resistor and inductive reactance. The phase angle is somewhere bet' 0 to 90° (because the series combination of resistance with inductive reactance).



V_s = RMS value of the applied voltage
 I = " " " " " resultent current.

V_R = voltage drops across R in phase I.

V_L = voltage drop across coil.

$V_L = I \cdot X_L$ (ahead of I by 90°).

The voltage drop are shown in voltage triangle:

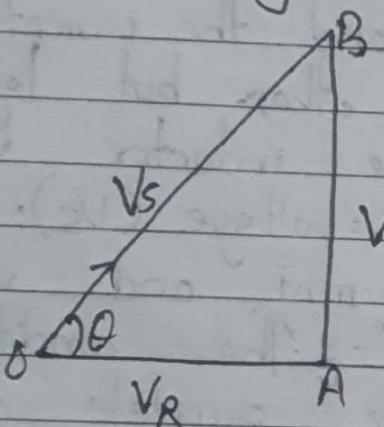


fig (a) - voltage triangle

$OA = \text{ohmic drop}$

$AB = \text{inductive drop}$

$OB = \text{vector sum of two } V_s$

$$V_s = \sqrt{V_R^2 + V_L^2}$$

$$V_s = \sqrt{(IR)^2 + (IX_L)^2}$$

$$V_s = I \sqrt{R^2 + X_L^2}$$

$$I = \frac{V_s}{\sqrt{R^2 + X_L^2}}$$

$$[\because Z = \sqrt{R^2 + X_L^2}]$$

$$\therefore I = \frac{V_s}{Z}$$

$$Z = \sqrt{R^2 + X_L^2}$$

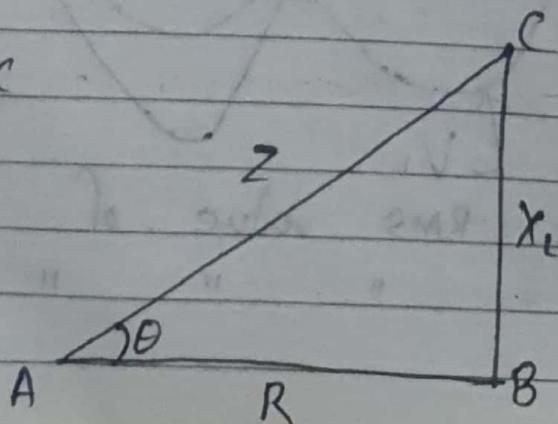
$$Z^2 = R^2 + X_L^2$$

$$(\text{Impedance})^2 = (\text{Resistance})^2 + (\text{Reactance})^2$$

Impedance $\triangle ABC$

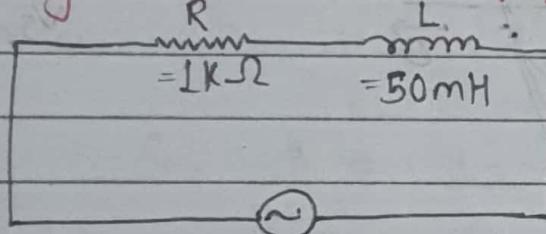
$$\tan \theta = \frac{X_L}{R}$$

$$\theta = \tan^{-1} \left(\frac{X_L}{R} \right)$$



Q. 1) Circuit shown in fig. below of $1\text{k}\Omega$ connected to + in series with 50 mH coil, a $10\text{ V rms}, 10\text{ kHz}$ signal is applied

- (i) Impedance
- (ii) Current
- (iii) Phase angle
- (iv) Voltage across resistance (V_R)
- (v) Voltage across inductor (V_L)



$$(i) \text{ Inductor } X_L = \omega L = 2\pi f \cdot L = 2\pi \times 10^4 \times 10^3 \times 50 \times 10^{-3} = 3141.5 \Omega$$

So,

$$\begin{aligned} \text{Impedance } (Z) &= \sqrt{R^2 + X_L^2} \\ &= \sqrt{(10^3)^2 + (3141.5)^2} \\ &= 3296.9 \Omega \end{aligned}$$

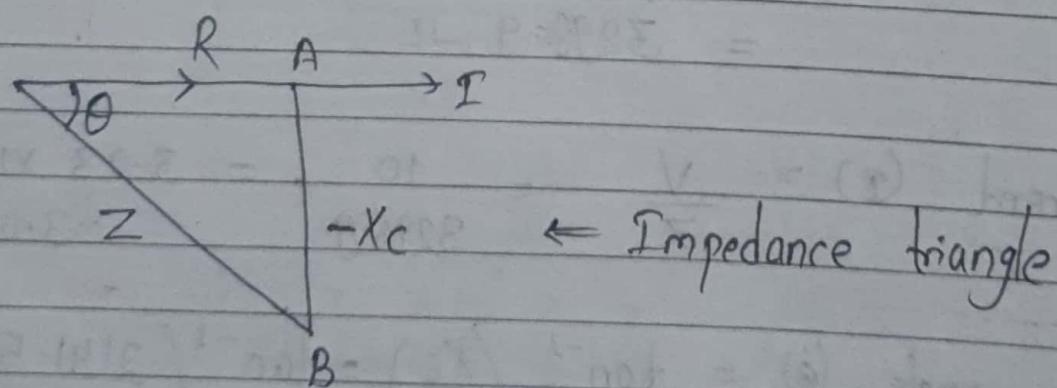
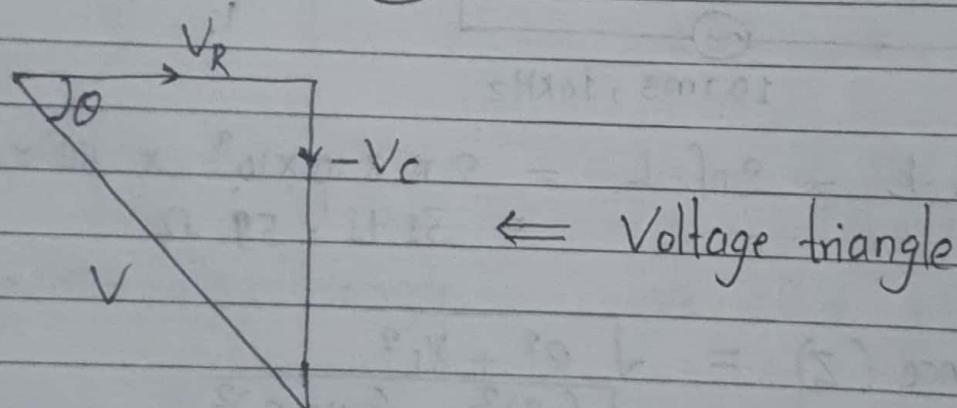
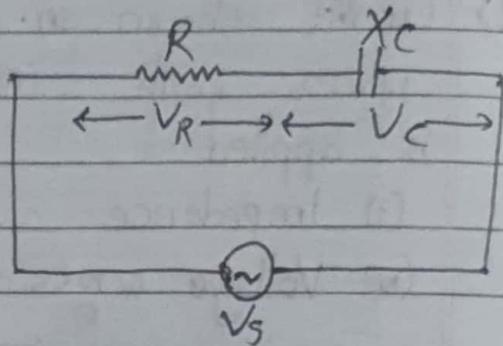
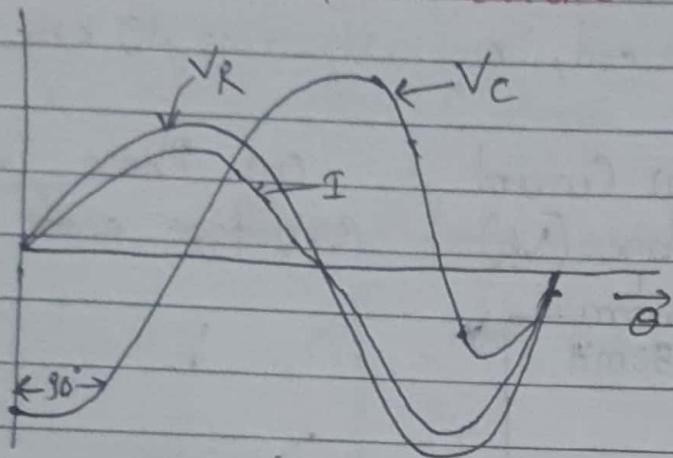
$$(ii) \text{ Current } (I) = \frac{V}{Z} = \frac{10}{3296.9} = 3.03 \times 10^{-3} \text{ amp} = 3.03 \text{ mA}$$

$$(iii) \text{ Phase angle } (\theta) = \tan^{-1} \left(\frac{X_L}{R} \right) = \tan^{-1} \left(\frac{3141.5}{10^3} \right) = 72.34^\circ$$

$$(iv) \text{ Voltage across resistance } (V_R) = IR = 3.03 \times 10^{-3} \times 10^3 = 3.03 \text{ V}$$

$$(v) \text{ Voltage across inductor } (V_L) = I \cdot X_L = 3.03 \times 10^{-3} \times 3141.5 = 9.51 \text{ V}$$

2) Series R-C Circuit



$$V_R = IR \quad \text{drop across the resistor (R)}$$

$$V_C = I \cdot X_C \quad \text{drop across the capacitor}$$

Capacitive reactance taken as -ve.

$$V_s^2 = V_R^2 + (-V_C)^2$$

$$\text{or } V_s = \sqrt{(IR)^2 + (-IX_C)^2}$$

$$\text{or } V_s = I \sqrt{R^2 + X_C^2}$$

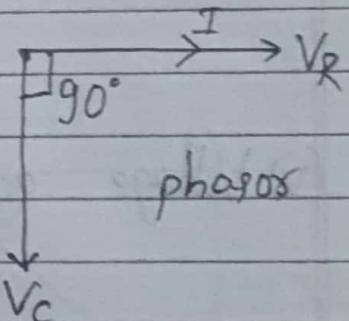
$$\therefore I = \frac{V_s}{\sqrt{R^2 + X_C^2}}$$

$$\text{Impedance } (Z) = \sqrt{R^2 + X_C^2}$$

$$\text{So, } I = \frac{V_s}{Z}$$

$$\tan \theta = -\frac{X_C}{R}$$

$$\theta = \tan^{-1} \left(-\frac{X_C}{R} \right)$$



- Q. The pure resistance of 50Ω in series with pure capacitor of $100\mu F$. This series combination is connected across $100V$, $50Hz$ supply. find (i) impedance (ii) current (iii) power factor (iv) phase angle (v) voltage across resistor (vi) voltage across capacitor (vii) draw vector diagram.

Solution:

$$\text{Given Capacitance, } X_C = \frac{1}{\omega C} = \frac{1}{2\pi f \cdot C} = \frac{1}{2\pi \times 50 \times 100 \times 10^{-6}} = 31.83 \Omega$$

$$\begin{aligned} \text{(ii)} \quad Z &= \sqrt{R^2 + X_C^2} \\ &= \sqrt{50^2 + (31.83)^2} \\ &= 59.36 \Omega \end{aligned}$$

(II) Current, $I = \frac{V_s}{Z} = \frac{100}{59.36} = 1.68$ amp

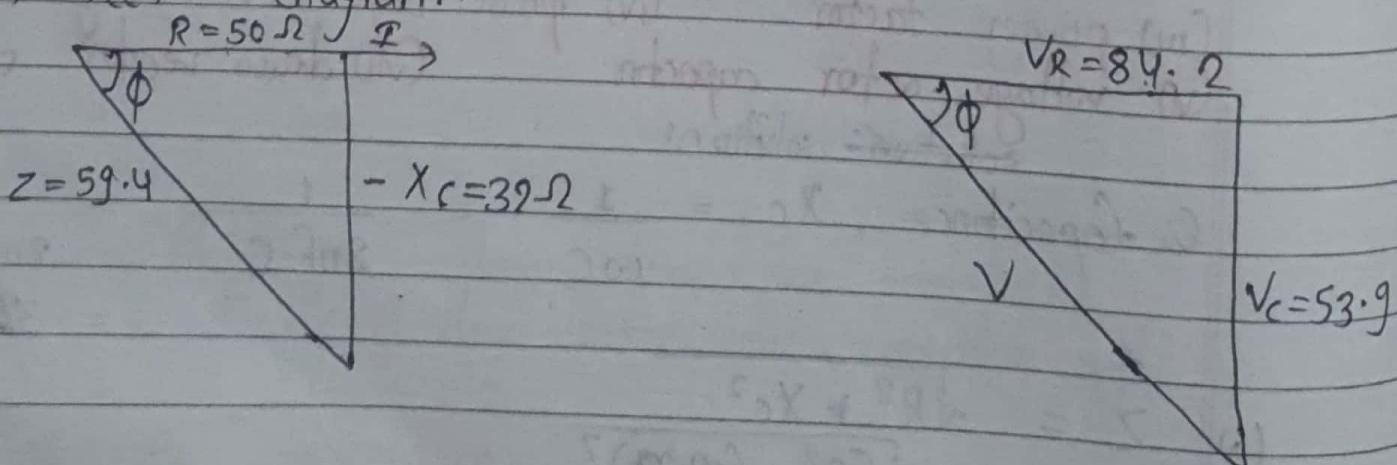
(III) power factor = $\frac{R}{Z} = \frac{50}{59.36} = 0.84$ ~~with~~

(IV) phase angle = $\tan^{-1} \left(\frac{-X_c}{R} \right)$
 $= \tan^{-1} \left(\frac{-32}{50} \right)$
 $= -32.62^\circ$

(V) Voltage across resistor (V_R) = IR
 $= 1.68 \times 50 = 84$ V

(VI) Voltage across capacitor (V_c) = $I \cdot X_c$
 $= 1.68 \times 32 = 53.76$ V

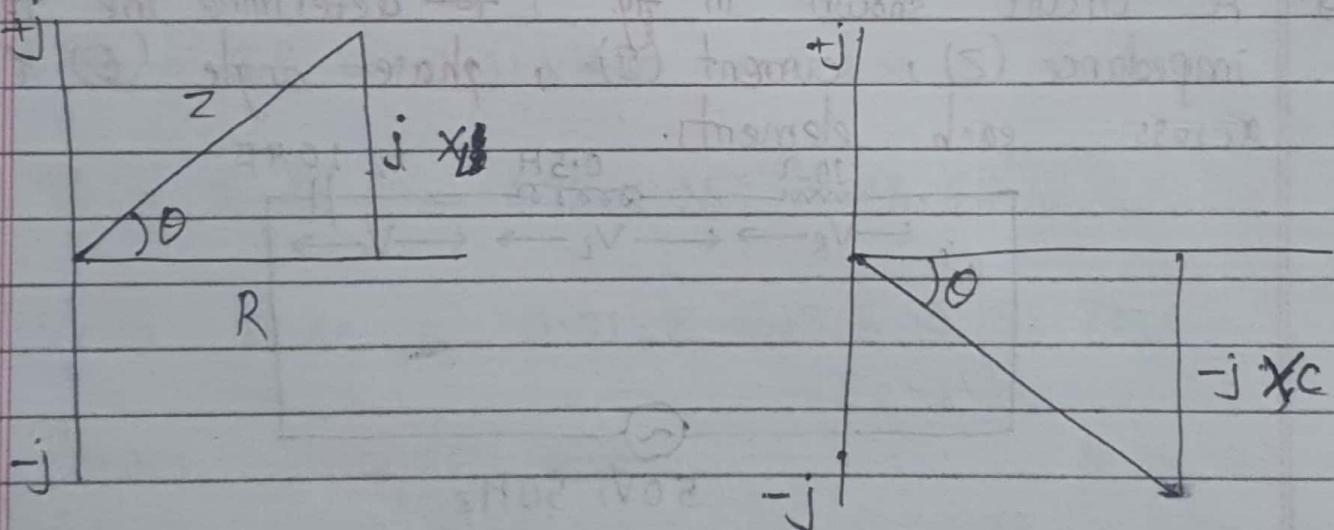
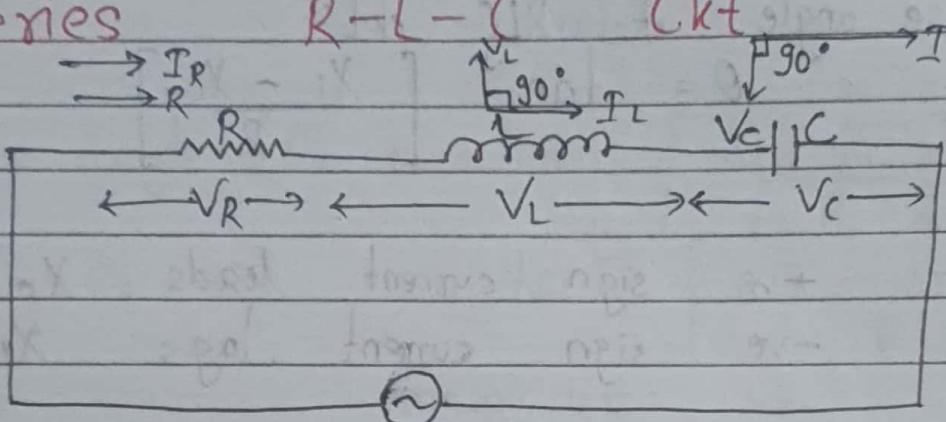
(VII) Vector diagram:



Series

3) Series

R-L-C Ckt



$$V_s^2 = (IR)^2 + (IX_L - IX_C)^2$$

$$\therefore V_s = I \sqrt{R^2 + (X_L - X_C)^2}$$

$$I = \frac{V_s}{\sqrt{R^2 + (X_L - X_C)^2}}$$

$$\therefore I = \frac{V_s}{Z} ; Z = \sqrt{R^2 + (X_L - X_C)^2}$$

power consumed =

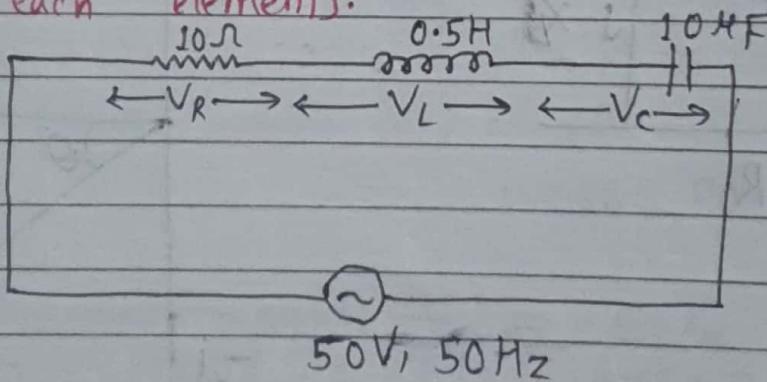
Similarly,

phase angle of RLC ckt:

$$\theta = \tan^{-1} \left[\frac{X_L - X_C}{R} \right]$$

Note:- +ve sign current leads $X_C > X_L$
 -ve sign current lags $X_L > X_C$

- Q. A circuit shown in fig., to determine the total impedance (Z), current (I), phase angle (θ) & voltage across each elements.



$$X_L = \omega L = 2\pi f \cdot L = 314 \times 0.5 = 157 \Omega$$

$$X_C = \frac{1}{\omega C} = \frac{1}{2\pi f C} = \frac{1}{2\pi \times 50 \times 10 \times 10^{-6}} \\ = 318.5 \Omega$$

$$Z = \sqrt{R^2 + (X_L - X_C)^2} \\ = \sqrt{10^2 + (157 - 318.5)^2} \\ = 161.8 \Omega$$

$$\text{Current } (I) = \frac{V}{Z} = \frac{50}{161.8} = 0.31 \text{ amp}$$

$$\begin{aligned}\text{phase angle } (\theta) &= \tan^{-1} \left[\frac{X_L - X_C}{R} \right] \\ &= \tan^{-1} \left[\frac{157 - 318.5}{10} \right] \\ &= -86.45^\circ\end{aligned}$$

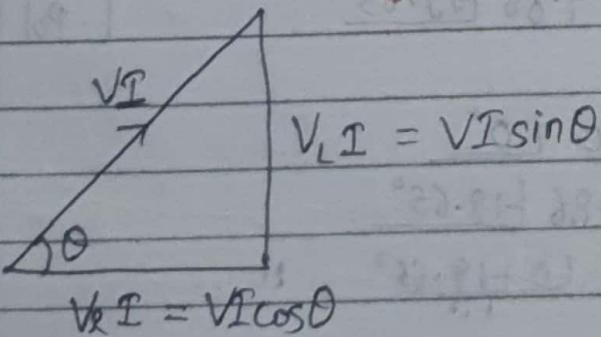
$$V_R = IR = 0.31 \times 10 = 3.1 V$$

$$V_L = I \cdot X_L = 0.31 \times 157 = 48.67 V$$

$$V_C = I \cdot X_C = 0.31 \times 318.5 = 98.73 V$$

Power Triangle

VVI
formula



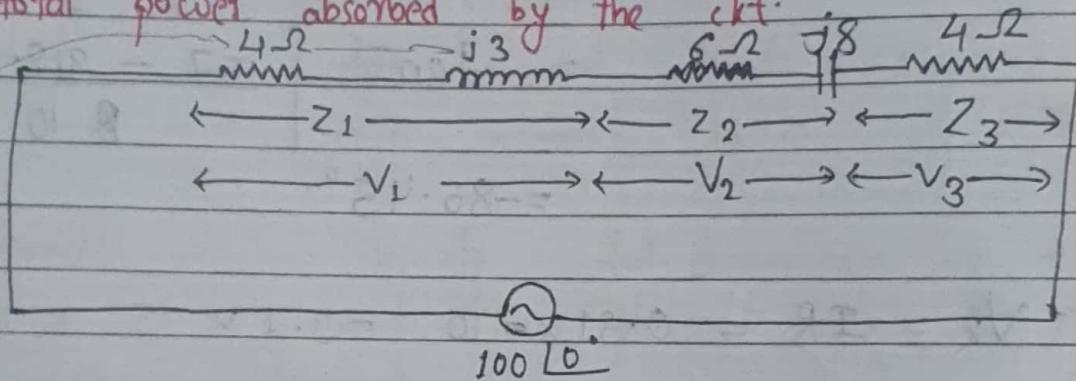
- ① Active power (True or Real power) (P) = $V_I \cos \phi$ W
- ② Reactive power (Q) = $V_I \sin \phi$ VAR
- ③ Apparent power (S) = V_I VA
- ④ Active component (wattful component) = $I \cos \phi$
- ⑤ Reactive comp. (wattless " ") = $I \sin \phi$

$$f = 50 \text{ Hz}$$

④ Reactive component (wattless comp.) = $I \sin \phi$

- Q. In a figure below, calculate (i) current, (ii) voltage drops V_1 , V_2 & V_3 (iii) power absorbed by each impedance & the total power absorbed by the ckt.

if L is given
then $Z = R + jX_L$



$$\text{Voltage } (V) = 100 \angle 0^\circ =$$

$$\begin{aligned} Z &= Z_1 + Z_2 + Z_3 \\ &= (4+j3) + (6-j8) + (4+j0) \\ (\text{rect form}) &= (14 - 5j) \angle 2 \\ (\text{polar form}) &\Rightarrow 14.86 \angle 19.65^\circ \end{aligned}$$

in calculator
[Pol (14, -5)]

Now,

$$\begin{aligned} I &= \frac{V}{Z} = \frac{100 \angle 0^\circ}{14.86 \angle 19.65^\circ} \\ &= \frac{100}{14.86} \angle 0 + 19.65^\circ \\ &\Rightarrow 6.72 \angle 19.65^\circ \end{aligned}$$

Again,

$$\begin{aligned} V_1 &= IR_1 \quad V_1 = IZ_1 = 6.72 \angle 19.65^\circ \times (4+j3) \\ &= 6.72 \angle 19.65^\circ \times 5\sqrt{36.86}^\circ \\ &= 33.6 \angle 56.51^\circ \\ &= (18.54 + j 28.02) \text{ V} \end{aligned}$$

$$V_2 = IZ_2$$

$$= 6.72(19.65^\circ) \times (6-j8)$$

$$= 6.72(19.65^\circ) \times 10(-53.13)$$

$$= \cancel{6.72}^{67.2} (-33.48)$$

$$= (56.05 - 37.07j)V$$

Hence capacitive.

Now,

$$V_3 = IZ_3$$

$$= 6.72(19.65^\circ) \times 4$$

$$= 26.88(19.65^\circ)V$$

Numericals

* Power absorbed by each elements:

$$\begin{aligned}P_1 &= I^2 R_1 \\&= (6.72)^2 \times 4 \\&= 180.63 \text{ watt}\end{aligned}$$

$$\begin{aligned}P_2 &= I^2 R_2 \\&= (6.72)^2 \times 6 \\&= 270.95 \text{ watt}\end{aligned}$$

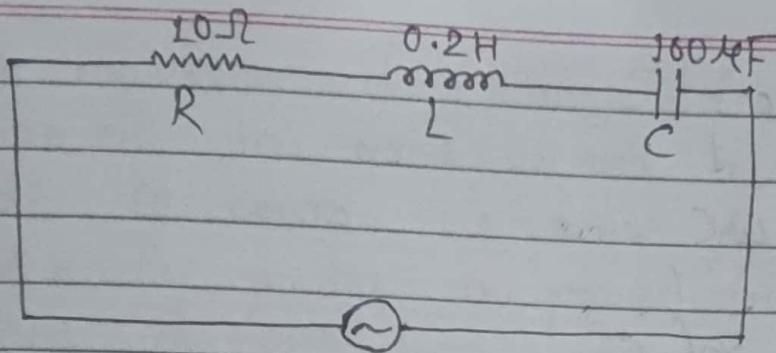
$$\begin{aligned}P_3 &= I^2 R_3 \\&= (6.72)^2 \times 4 \\&= 180.63 \text{ watt}\end{aligned}$$

Now,

Power absorbed by the circuit,

$$\begin{aligned}P_{\text{total}} &= P_1 + P_2 + P_3 \\&= (180.63 + 270.95 + 180.63) \text{ watt} \\&= 632.21 \text{ watt}\end{aligned}$$

- Q. A resistance of 10Ω , inductance of 0.2 H and capacitor of $100 \mu\text{F}$ are connected in series across 200V , 50 Hz supply mains. Determine
- (i) impedance
 - (ii) current
 - (iii) voltage across R , L & C
 - (iv) power factor angle of lag (PF)
 - (v) power consumed in watt
 - (vi) VA



$$X_L = \omega L = 2\pi f L = 314 \times 0.2 = 62.8 \Omega$$

$$X_C = \frac{1}{\omega C} = \frac{1}{2\pi f \cdot C} = \frac{1}{314 \times 100 \times 10^{-6}} = 31.85 \Omega$$

$$\begin{aligned} \text{(i) Impedance } Z &= \sqrt{R^2 + (X_L - X_C)^2} \\ &= \sqrt{10^2 + (62.8 - 31.85)^2} \\ &= 32.52 \Omega \end{aligned}$$

$$\begin{aligned} \text{(ii) current, } I &= \frac{V}{Z} = \frac{200}{32.52} \\ &= 6.15 \text{ amp} \end{aligned}$$

(iii) Voltage across Resistor: R, L, C :

$$\begin{aligned} V_R &= IR \\ &= 6.15 \times 10 = 61.5 V \end{aligned}$$

$$\begin{aligned} V_L &= I X_L = I \times \omega L \\ &= I \times 2\pi f \cdot L \\ &= 6.15 \times 2\pi \times 50 \times 0.2 \\ &= 386.41 V \end{aligned}$$

$$\begin{aligned}
 V_C &= I \cdot X_C \\
 &= I \cdot \frac{1}{\omega C} \\
 &= I \cdot \frac{1}{2\pi f \cdot C} \\
 &= \frac{6.51}{314 \times 100 \times 10^{-6}} \\
 &= 207.32 \text{ V}
 \end{aligned}$$

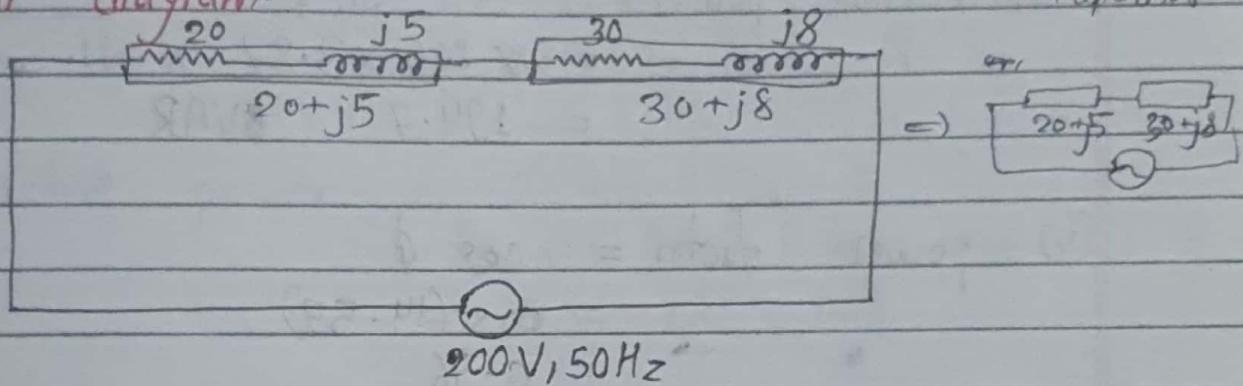
(iv) power factor angle of lag = $\frac{R}{Z} = \cos\phi$

$$\begin{aligned}
 &= \frac{10}{32.52} \\
 &= 0.307
 \end{aligned}$$

(v) Power consumed = $VI \cos\phi$

$$\begin{aligned}
 &= 200 \times 6.15 \times 0.307 \\
 &= 377.61
 \end{aligned}$$

Q. Two impedances of $(20+j5)\Omega$ and $(30+j8)\Omega$ are connected in series across a 200 V, 50Hz supply. Find: (i) current (ii) active power (iii) apparent power (iv) reactive power (v) power factor & also draw a phasor diagram.



$$\begin{aligned}Z_T &= Z_1 + Z_2 \\&= (20 + j5) + (30 + j8) \\&= (50 + 13j) \Omega \\&= 51.66 \angle 14.57^\circ\end{aligned}$$

$$V = 200 + j0 = 200 \angle 0^\circ$$

$$(i) I = \frac{V}{Z} = \frac{200 \angle 0^\circ}{51.66 \angle 14.57^\circ} = 3.87 \angle -14.57^\circ$$

~~Now,~~

$$(ii) \cos \phi = \frac{R}{Z}$$

~~$\cos \phi = \cos(14.57^\circ)$~~

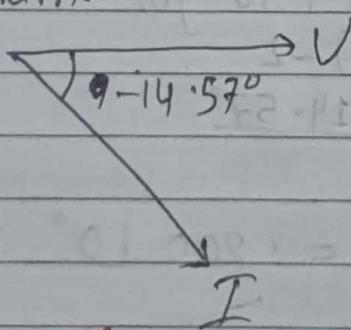
$$\begin{aligned}&\text{active power} = VI \cos \phi \\&= 200 \angle 0^\circ \times 3.87 \angle -14.57^\circ \\&= 200 \times 3.87 \times \cos(14.57^\circ) \\&= 11,297.18 \text{ W} \quad \cancel{W} \quad 749.10 \text{ kW}\end{aligned}$$

$$\begin{aligned}
 \text{(iii) apparent power} &= VI \\
 &= 200 \times 3.87 \\
 &= 774 \text{ VA}
 \end{aligned}$$

$$\begin{aligned}
 \text{(iv) reactive power} &= VI \sin \phi \\
 &= 200 \times 3.87 \times \sqrt{1 - \cos^2(14.57)} \\
 &= 194.7 \text{ VAR}
 \end{aligned}$$

$$\begin{aligned}
 \text{(v) power factor} &= \cos \phi \\
 &= \cos(14.57) \\
 &= 0.96
 \end{aligned}$$

Phasor diagram:



- Q Two impedances of $(5 + j10) \Omega$ & $(10 + j5) \Omega$ are connected in series across 200V, 50Hz. Find (i) current (ii) active power (iii) reactive power (iv) apparent power (v) power factor & also draw phasor diagram.

solution:

$$\begin{aligned}
 Z_1 &= (5 + j10) \Omega = 11.18 \angle 63.43^\circ \\
 Z_2 &= (10 + j5) \Omega = 11.18 \angle 26.56^\circ
 \end{aligned}$$

$$\begin{aligned}
 Z_T &= Z_1 + Z_2 = (5 + j10) + (10 + j5) \\
 &= (15 + j15) \Omega = 21.21 \angle 45^\circ
 \end{aligned}$$

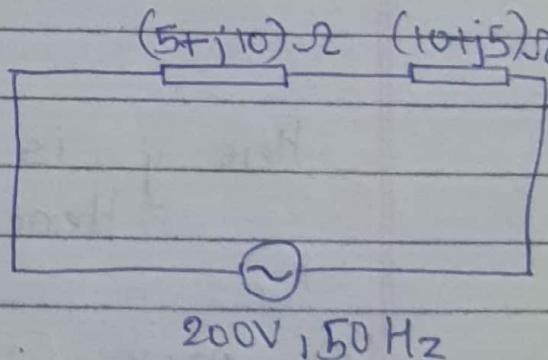
$$(i) I = \frac{V}{Z} = \frac{200}{21.21 \angle 45^\circ} = 9.42 \angle -45^\circ$$

$$\begin{aligned}(ii) \text{ active power} &= VI \cos \phi \\ &= 200 \times 9.42 \times \cos(45^\circ) \\ &= 1332.18 \text{ watt}\end{aligned}$$

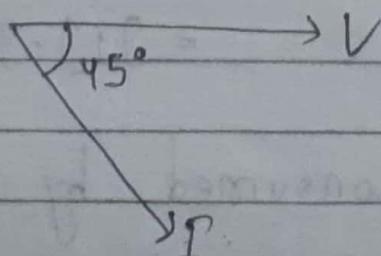
$$\begin{aligned}(iii) \text{ reactive power} &= VI \sin \phi \\ &= 200 \times 9.42 \times \sin(45^\circ) \\ &= 1332.18 \text{ VAR}\end{aligned}$$

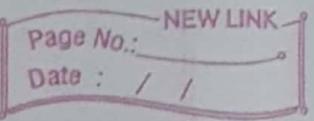
$$\begin{aligned}(iv) \text{ apparent power} &= VI \\ &= 200 \times 9.42 \\ &= 1884 \text{ VA}\end{aligned}$$

$$\begin{aligned}(v) \text{ power factor} &= \cos \phi \\ &= \cos(45^\circ) \\ &= \cancel{\frac{1}{\sqrt{2}}} \\ &= 0.70\end{aligned}$$



Phasor diagram:





- Q. The ac voltage $V = (80 + j 60)$ volt applied to a ckt. resulting current of $(4 + j 10)$ ampere. Find
 (i) impedance of ckt whether it is inductive or capacitive
 (ii) power consumed
 (iii) phase angle
 (iv) reactive power

solution:

$$V = 80 + j 60 = 100 \angle 36.86^\circ$$

$$I = 4 + j 10 = 10.77 \angle 68.19^\circ$$

Now,

$$\begin{aligned} \text{(i) Impedance } (Z) &= \frac{V}{I} \\ &= \frac{100 \angle 36.86^\circ}{10.77 \angle 68.19^\circ} \\ &= 9.28 \angle -31.33^\circ \\ &= 7.92 - j 4.82 \end{aligned}$$

Here j is -ve.

Hence capacitive.

$$\begin{aligned} \text{(ii) power consumed} &= VI \cos \phi \\ &= 100 \times 10.77 \times \cos(31.33^\circ) \\ &= 919.95 \text{ watt} \end{aligned}$$

\therefore Power consumed by given ckt = 919.95 watt

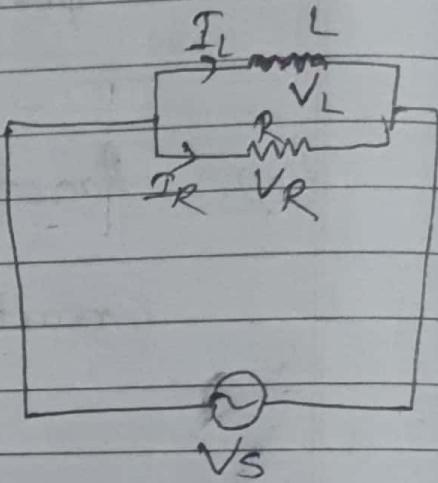
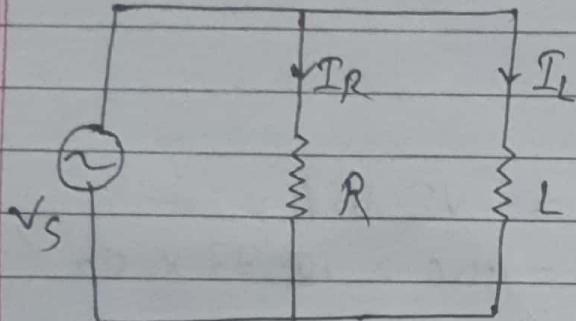
(iv) phase angle (ϕ) = 31.33°

(v) Reactive power:

$$\begin{aligned}\text{reactive power} &= VI \sin \phi \\ &= 100 \times 10.77 \times \sin (31.33) \\ &= 560 \text{ VAR}\end{aligned}$$

Parallel Circuit

1) Parallel R-L Ckt



Applied voltage = V_s

voltage drop across (R) = V_R

" " " " (L) = V_L

Currents I_L lags behind the applied voltage.

$$I_R = \frac{V_s}{R}$$

$$I_L = \frac{V_s}{X_L}$$

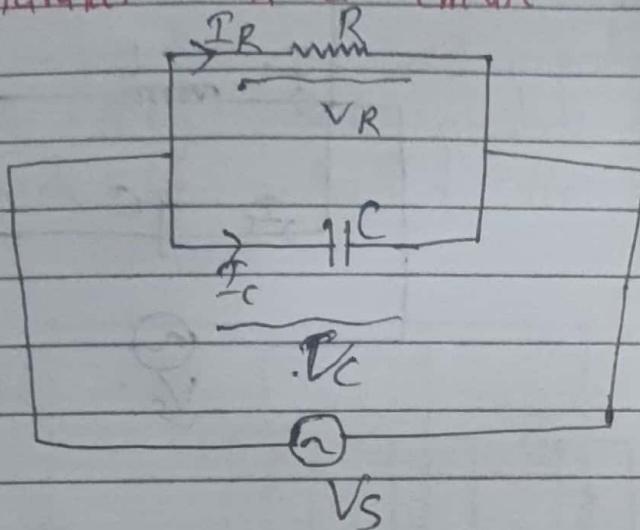
$$I = \sqrt{I_R^2 + I_L^2}$$

$$\tan \phi = \frac{I_L}{I_R} = \frac{\frac{V_s}{X_L}}{\frac{V_s}{R}} = \frac{R}{X_L}$$

$$\phi = \tan^{-1} \left(\frac{R}{X_L} \right)$$

$$\cos \phi = \frac{I_R}{\sqrt{I_R^2 + I_L^2}} = \frac{I_R}{I} = \text{power factor}$$

2) Parallel R-C circuit



$$I_R = \frac{V_s}{R} \quad \text{or} \quad V_s = I_R \cdot R$$

$$I_C = \frac{V_s}{X_C} \quad \text{or} \quad V_s = I_C \cdot X_C$$

$$\text{Resultant currents (I)} = \sqrt{I_R^2 + I_C^2}$$

$$\text{Power factor (PF)} = \frac{I_R}{\sqrt{I_R^2 + I_C^2}}$$

$$\tan \phi = \frac{I_C}{I_R} = \frac{V_s/X_C}{V_s/R} = \frac{R}{X_C}$$

$$\text{Phase angle } (\phi) = \tan^{-1} \left(\frac{R}{X_C} \right)$$

$$\cos \phi = \frac{I_R}{\sqrt{I_R^2 + I_C^2}}$$

3) Parallel R-L-C ckt

$$V_s = I_R \cdot R$$

$$V_s = I_L \cdot X_L$$

$$V_s = I_C \cdot X_C$$

$$I_R = \frac{V_s}{R}$$

$$I_L = \frac{V_s}{X_L}$$

$$I_C = \frac{V_s}{X_C}$$

The resultant or total current ;

$$I = \sqrt{I_R^2 + (I_L - I_C)^2}$$

$$\text{Power factor} = \cos \phi = \frac{I_R}{I}$$

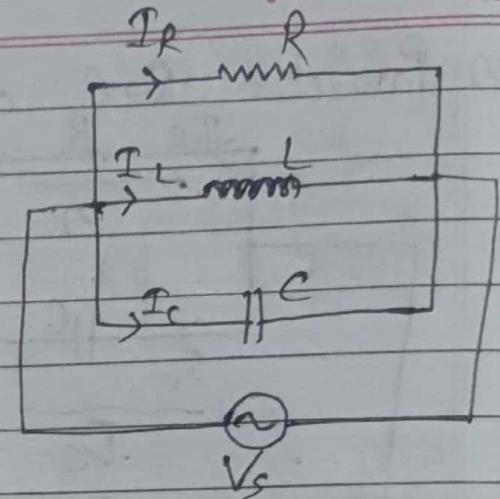
$$= \frac{I_R}{\sqrt{I_R^2 + (I_L - I_C)^2}}$$

$$\tan \phi = \frac{I_L - I_C}{I_R}$$

$$\phi = \tan^{-1} \left(\frac{I_L - I_C}{I_R} \right)$$

$$= \tan^{-1} \left(\frac{\frac{V_s}{X_L} - \frac{V_s}{X_C}}{\frac{V_s}{R}} \right)$$

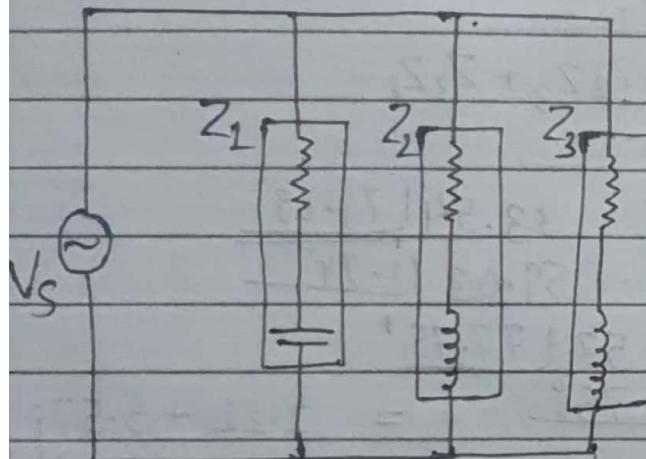
$$= \tan^{-1} \left[R \left(\frac{1}{X_L} - \frac{1}{X_C} \right) \right]$$



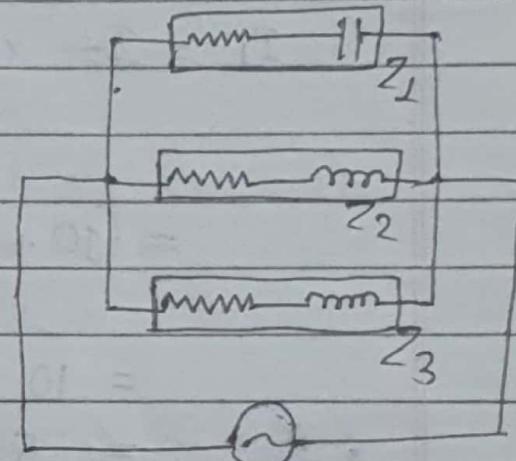
Problems of R-LC Ckt:

Total current of 10 amp current flows through parallel combination of three impedances as $(2-j5)\Omega$, $(6+j3)\Omega$ & $(3+j4)\Omega$. Calculate the current flowing through each branch. Also find the power factor of the combination.

Solution:



or,



$$Z_1 = (2 - j5) \Omega$$

$$Z_2 = (6 + j3) \Omega$$

$$Z_3 = (3 + j4) \Omega$$

$$Z_1 \cdot Z_2 = (2 - j5) \times (6 + j3) =$$

$$Z_2 \cdot Z_3 =$$

$$Z_3 \cdot Z_1 =$$

$$Z_1 \cdot Z_2 \cdot Z_3 =$$

$$(a + bj)(c + dj)$$

$$= (ac - bd) + (ad + bc)j$$

$$\frac{1}{Z_T} = \frac{1}{Z_1} + \frac{1}{Z_2} + \frac{1}{Z_3}$$

$$Z_1 \cdot Z_2 = (2 - j5) \times (6 + j3) = (27 - j24) \Omega = 36.12 \angle -41.63^\circ$$

$$Z_2 \cdot Z_3 = (6 + j3) \times (3 + j4) = (6 + j33) \Omega = 33.54 \angle 79.69^\circ$$

$$Z_3 \cdot Z_1 = (3 + j4) \times (2 - j5) = (26 - j7) \Omega = 26.92 \angle -15.06^\circ$$

$$Z_1 \cdot Z_2 \cdot Z_3 = (2 - j5)(6 + j3)(3 + j4) = (177 + j36) \Omega$$

$$= 180.62 \angle 11.49^\circ$$

$$\begin{aligned}
 Z_1 \cdot Z_2 + Z_2 \cdot Z_3 + Z_3 \cdot Z_1 &= (27 - j 24) + (6 + j 33) + \\
 &\quad (26 - j 7) \\
 &= (59 + j 2) \angle 2 \\
 &= 59.03 \angle 1.94^\circ
 \end{aligned}$$

Now,

$$I_1 = I_2 \times \frac{Z_2 \cdot Z_3}{Z_1 Z_2 + Z_2 Z_3 + Z_3 Z_1}$$

$$Z_1 Z_2 + Z_2 Z_3 + Z_3 Z_1$$

$$= (10 + j 0) \times \frac{33.54 \angle 79.69^\circ}{59.03 \angle 1.94^\circ}$$

$$= 10 \times 0.57 \angle 77.75^\circ$$

$$= 5.7 \angle 77.75^\circ = 1.21 + 5.57j$$

$$\begin{aligned}
 I_2 &= I_2 \times \frac{Z_3 \cdot Z_1}{Z_1 Z_2 + Z_2 Z_3 + Z_3 Z_1} \\
 &= 10 \times \frac{26.92 \angle -15.06^\circ}{59.03 \angle 1.94^\circ} \\
 &= 4.56 \angle -17^\circ
 \end{aligned}$$

①

$$\begin{aligned}
 I_3 &= I_2 \times \frac{Z_1 \cdot Z_2}{Z_1 Z_2 + Z_2 Z_3 + Z_3 Z_1} \\
 &= 10 \times \frac{36.12 \angle -41.63^\circ}{59.03 \angle 1.94^\circ} \\
 &= 6.1 \angle -43.57^\circ
 \end{aligned}$$

$$\text{Total } Z, \quad Z_T = \underline{Z_1 Z_2 Z_3}$$

$$= \underline{Z_1 Z_2 + Z_2 Z_3 + Z_3 Z_1}$$

~~$$= \frac{180.62 \angle 11.49^\circ}{360.12 \angle -41.63^\circ + 38.54 \angle 79.69^\circ + 180.62 \angle 11.49^\circ}$$~~

$$= \frac{180.62 \angle 11.49^\circ}{59.03 \angle 1.94^\circ}$$

$$= 3.05 \angle 9.55^\circ$$

$$= (3.01 + j 0.51) \Omega$$

$$\begin{aligned} \text{power factor} &= \cos \phi \\ &= \cos(9.55^\circ) \\ &= 0.98 \end{aligned}$$

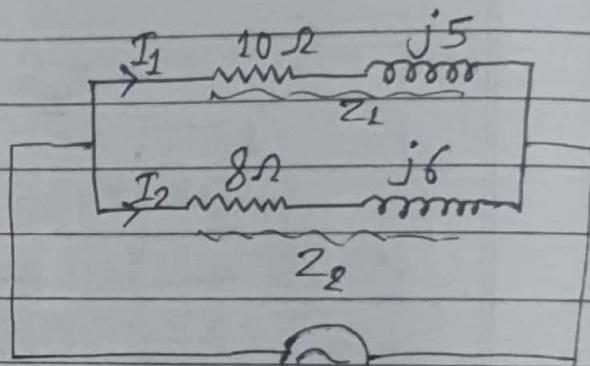
Q. Two impedances given by $Z_1 = (10 + j5) \Omega$ & $Z_2 = (8 + j6) \Omega$ are joined in parallel and connected across the voltage of $V = (200 + j0) V$. Calculate the circuit current, its phase & the branch current & draw the vector diagram.

Solution:-

$$Z_1 = (10 + j5) \Omega = 11.18 \angle 26.56^\circ$$

$$Z_2 = (8 + j6) \Omega = 10 \angle 36.86^\circ$$

$$V = 200 V$$



$$\begin{aligned}
 Z_T &= \frac{Z_1 \cdot Z_2}{Z_1 + Z_2} = \frac{(10 + j5)(8 + j6)}{(10 + j5) + (8 + j6)} \\
 &= \frac{50 + j100}{18 + j11} \\
 &= \frac{50 \angle 0^\circ}{14.86 \angle 47.72^\circ} = \frac{111.8 \angle 63.43^\circ}{21.09 \angle 31.42^\circ} \\
 &= 5.30 \angle 32.01^\circ
 \end{aligned}$$

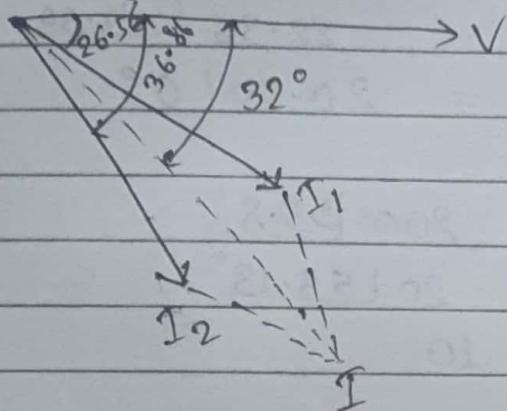
$$\text{Power factor} = \cos \phi = \cos(32.01) = 0.847$$

Now,

$$\left(\frac{Z_2 \times Z_2}{Z_1 + Z_2} \right) I_1 = \frac{V_s}{Z_1} = \frac{200 + j0}{11.18 \angle 26.56^\circ} = 17.88 \angle -26.56^\circ$$

$$I_2 = \frac{V_s}{Z_2} = \frac{200}{10 \angle 36.86^\circ} = 20 \angle -36.86^\circ$$

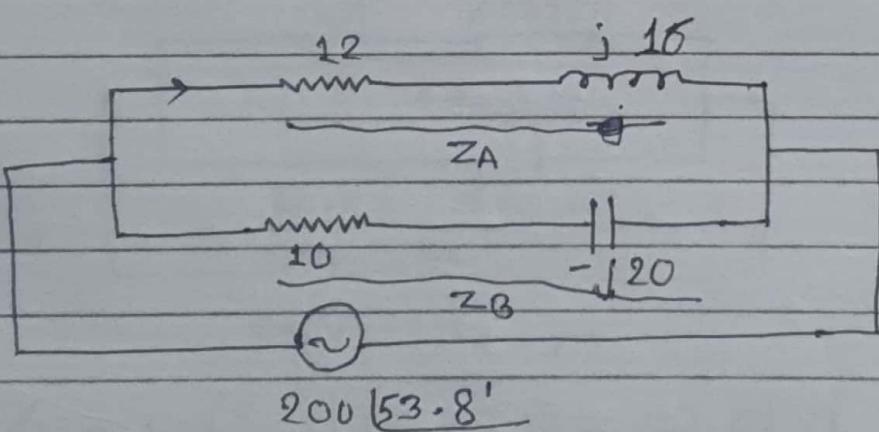
Phasor diagram:



Q. A voltage of $200 \angle 53.8^\circ$ is applied across impedance in parallel. The value of impedance are $(12 + j16) \Omega$ & $(10 - j20) \Omega$. Determine the KVA, KWT KVAR & KW in each branch & power factor of whole ckt.

solution:

Ans



$$I_A = 10 + j0$$

$$I_B = -4.0 + j8$$

$$K_W = 1.2$$

$$KVAR = 1.6$$

$$KVA = 2$$

$$pf = \cos \phi = \frac{R}{Z}$$

Here,

$$V = 200 \angle 53.8^\circ$$

Also,

$$\ell \quad Z_1 = (12 + j16) \Omega = 20 \angle 53.13^\circ$$

$$Z_2 = (10 - j20) \Omega = 22.36 \angle -63.43^\circ$$

$$Z_1 \cdot Z_2 = (12 + j16)(10 - j20) = 440 - 80j = 447.21 \angle -10.30^\circ$$

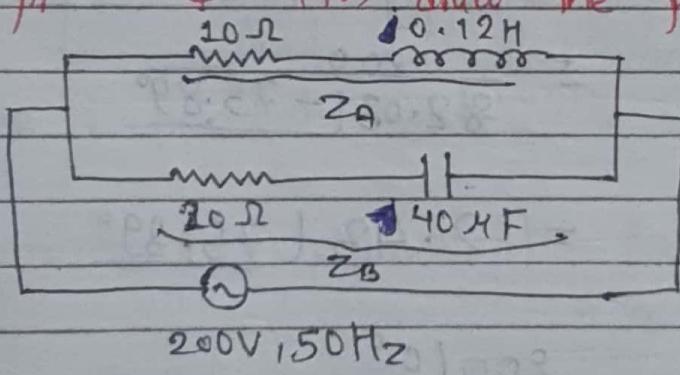
$$Z_1 + Z_2 = 12 + j16 + 10 - j20 = (22 - 4j) \Omega = 22.36 \angle -10.30^\circ$$

$$Z_T = \frac{Z_1 Z_2}{Z_1 + Z_2} = \frac{447.91 \angle -10.30^\circ}{22.36 \angle -10.30^\circ} \\ = 20.00 \angle 0^\circ$$

$$I_A = \frac{V}{Z_1} = \frac{200 \angle 53.8^\circ}{20 \angle 53.13^\circ} \\ = 10$$

$$I_B = \frac{V}{Z_2} = \frac{200 \angle 53.8^\circ}{22.36 \angle -63.46^\circ} \\ =$$

Q. Two circuits A & B are connected in parallel across 200V, 50Hz main supply. Ckt A consists of resistance of $10\ \Omega$ & inductance of $0.12\ H$ connected in series, ckt B consists of a resistance $20\ \Omega$ in series with capacitor of $40\ \mu F$. Calculate
 (i) current in each branch, (ii) source current
 (iii) pf & (iv) draw the phasor diagram.



$$X_L = 2\pi f L = 314 \times 0.12 = 37.67\ \Omega$$

$$X_C = \frac{1}{2\pi f C} = \frac{1}{314 \times 40 \times 10^{-6}} = 79.61\ \Omega$$

$$Z_A = 10 + j 37.7\ \Omega$$

$$Z_B = 20 - j 79.61\ \Omega$$

$$\begin{aligned}
 Z_{eq} &= \frac{Z_A \cdot Z_B}{Z_A + Z_B} \\
 &= 62.11 \angle 53.65^\circ \\
 \Rightarrow Z_{eq} &= (36.81 + j 50.02) - \Omega
 \end{aligned}$$

Branch currents:

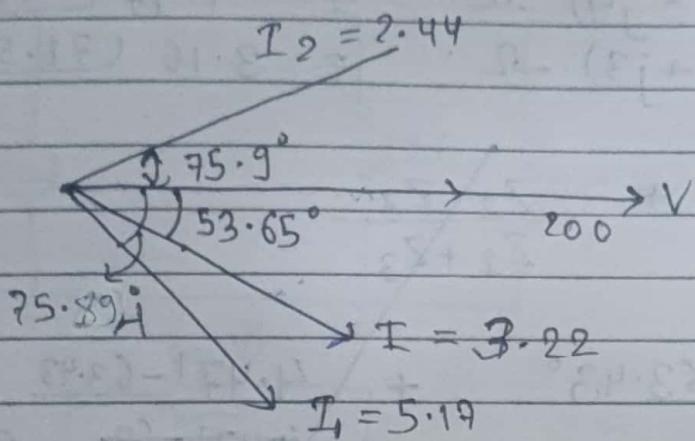
$$\begin{aligned}
 (i) \quad I_A &= \frac{V}{Z_A} = \frac{200 \angle 0^\circ}{10 + j 37.7} \\
 &= \frac{200}{39 \angle 75.14^\circ} \\
 &= 5.12 \angle -75.14^\circ
 \end{aligned}$$

$$\begin{aligned}
 I_B &= \frac{V}{Z_B} = \frac{200 \angle 0^\circ}{20 - j 79.61} \\
 &= \frac{200}{82.081 \angle -75.89^\circ} \\
 &= 2.43 \angle 75.89^\circ
 \end{aligned}$$

$$\begin{aligned}
 (ii) \quad I &= \frac{V}{Z_{eq}} = \frac{200 \angle 0^\circ}{62.11 \angle 53.65^\circ} \\
 &= 3.22 \angle -53.65^\circ
 \end{aligned}$$

$$\begin{aligned}
 (iii) \text{ power factor (pf)} &= \cos \phi = \cos(53.65) \\
 &= 0.59
 \end{aligned}$$

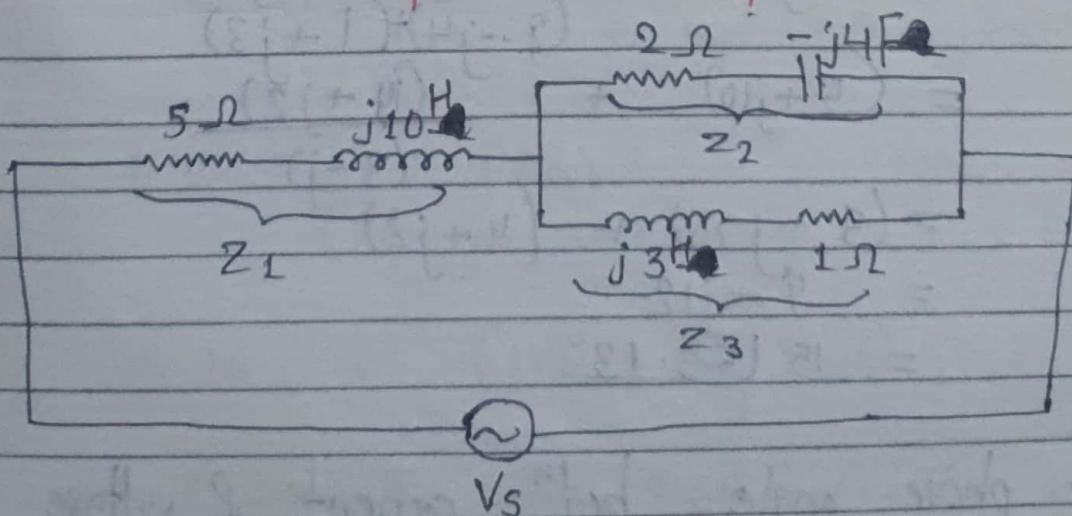
(iv) phasor diagram



Compound ckt

In many cases, the ac ckt to be analysed consist of combination of series & parallel impedance ckt. This types are known as series-parallel or compound ckt. Compound ckt can be simplified in the same manner as series - parallel ckt.

Q.1 Determine the equivalent Impedance of the given ckt.



$$Z_1 = (5+j10) \Omega = 11.18 \angle 63.43^\circ$$

$$Z_2 = (2-j4) \Omega = 4.47 \angle -63.43^\circ$$

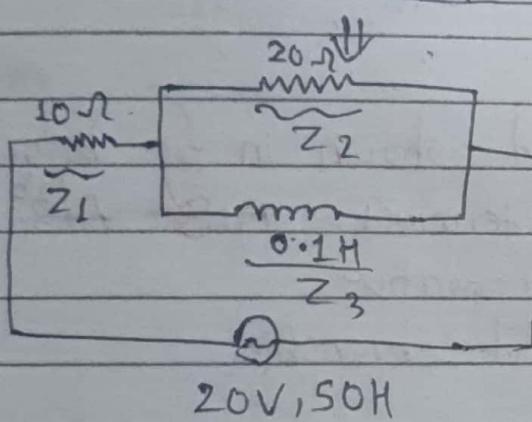
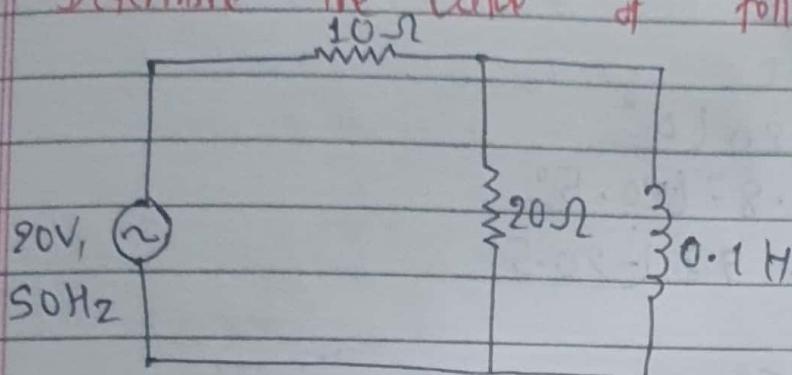
$$Z_3 = (1+j3) \Omega = 3.16 \angle 71.56^\circ$$

$$\begin{aligned} Z_T &= Z_1 + \frac{Z_2 \cdot Z_3}{Z_2 + Z_3} \\ &= 11.18 \angle 63.43^\circ + \frac{4.47 \angle -63.43^\circ \times 3.16 \angle 71.56^\circ}{4.47 \angle -63^\circ 3 - j 1} \\ &= 11.18 \angle 63.43^\circ + 14.12 \angle 8.18^\circ \\ &= 11.18 \angle 63.43^\circ + 4.46 \angle -10.25^\circ \end{aligned}$$

$$\begin{aligned} Z_T &= Z_1 + \frac{Z_2 \cdot Z_3}{Z_2 + Z_3} \\ &= (5+j10) + \frac{(2-j4)(1+j3)}{(2-j4)(1+j3)} \\ &= (5+j10) + \frac{(4+j2)}{(3-j)} \\ &= (5+j10) + (4+j2) \\ &= 9+j12 \\ &= 15 \angle 53.13^\circ \end{aligned}$$

∴ phase angle b/w current & voltage is
 $\cos \phi = \cos (53.13)$
 $= 0.60$

Q.2) Determine the value of following: (i) Z_T , (ii) I_T , (iii) ϕ .



$$\begin{aligned}
 Z_1 &= (10 + j0) \Omega \\
 Z_2 &= (20 + j0) \Omega \\
 Z_3 &= 0 + j \cdot X_L \\
 &= 0 + j \cdot 2\pi f L \\
 &= 0 + j \times 314 \times 0.1 \\
 &= (0 + j \cdot 31.4) \Omega
 \end{aligned}$$

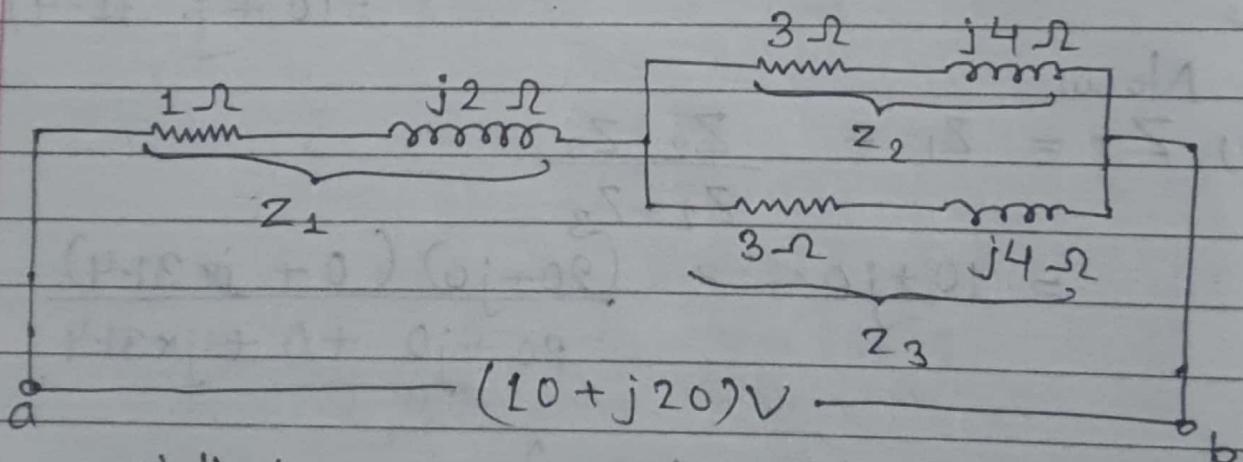
Now,

$$\begin{aligned}
 (i) Z_T &= Z_1 + \frac{Z_2 \cdot Z_3}{Z_2 + Z_3} \\
 &= 10 + j0 + \frac{(20 + j0)(0 + j \times 31.4)}{20 + j0 + 0 + j \times 31.4} \\
 &= 10 + \frac{(0 + 628j)}{(20 + j \cdot 31.4)} \\
 &= 10 + (14.22 + j 9.06) \\
 &= (24.22 + j 9.06) \Omega \\
 &= 25.86 \angle 20.5^\circ
 \end{aligned}$$

$$\begin{aligned}
 \text{(ii)} \quad I_T &= \frac{V_S}{Z_T} \\
 &= \frac{90 \angle 0^\circ}{25.87 \angle 20.5^\circ} \\
 &= 0.77 \angle -20.5^\circ
 \end{aligned}$$

$$\text{(iii)} \quad \phi = 20.5^\circ$$

- Q For the series-parallel circuit shown in fig. below, determine
- total impedance bet' terminal A₁B₁ A₂B₂ state if it is inductive or capacitive
 - voltage across the parallel branch.
 - phase angle.



Solution:

$$Z_1 = (1 + j2) \Omega$$

$$Z_2 = (3 + j4) \Omega$$

$$Z_3 = (3 + j4) \Omega$$

We have,

$$\begin{aligned}
 \text{(i)} \quad Z_T &= Z_1 + \frac{Z_2 \cdot Z_3}{Z_2 + Z_3} \\
 &= (1 + j2) + \frac{(3 + j4)(3 + j4)}{(3 + j4) + (3 + j4)}
 \end{aligned}$$

$$= (1 + j2) + \frac{-7 + 24j}{6 + j8}$$

$$= 1 + j2 + \frac{3}{2} + 2j$$

$$\therefore Z_T = (2.5 + j4) \Omega$$

Hence, the given ckt is inductive.

~~Diagram~~

$$I_T = \frac{V_s}{Z_T} = \frac{10 + j20}{2.5 + j4}$$

$$= 4.72 + j0.45$$

$$= 4.73 \angle 5.44^\circ$$

i.e. The current lags behind the voltage by 5.44° .

(ii) Voltage across parallel branch,

$$v = I_T \cdot Z_{eq}$$

$$= 4.73 \angle 5.44^\circ \times 4.5 \angle 61.18^\circ$$

$$Z_{eq} = \frac{Z_2 \cdot Z_3}{Z_2 + Z_3} = 1.5 + j2$$

Now,

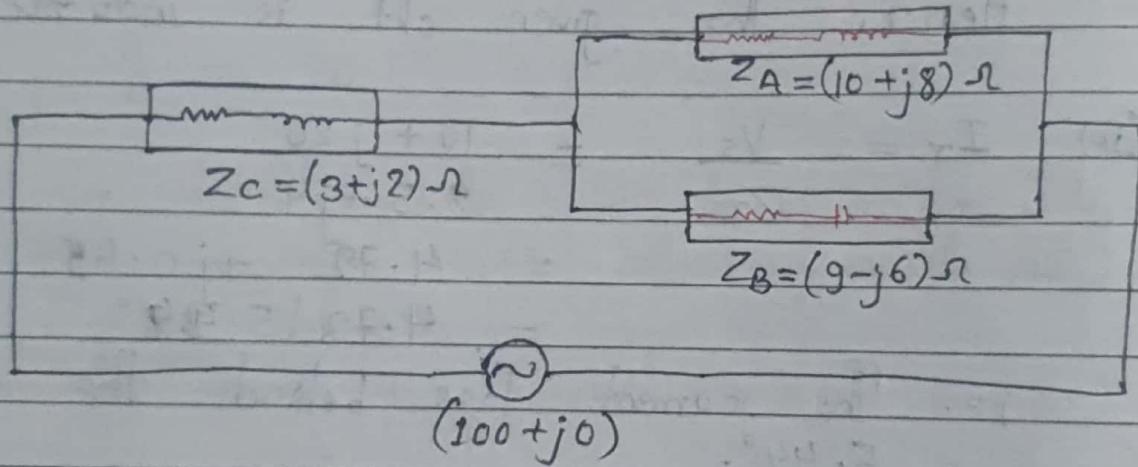
$$v = I_T \cdot Z_{eq}$$

$$= 4.73 \angle 5.44^\circ \times (1.5 + j2)$$

$$= 4.73 \angle 5.44^\circ \times 2.5 \angle 53.13^\circ$$

$$= 11.85 \angle 58.57^\circ$$

Q. In series - parallel ckt, shown in fig. below, the two parallel branches A & B are in series with C. The impedances are $Z_A = (10 + j8) \Omega$, $Z_B = (9 - j6) \Omega$ & $Z_C = (3 + j2) \Omega$ & the voltage across the ckt is $(100 + j0)$ V. Find the current I_A , I_B & phase angle bet' them.



$$\begin{aligned}
 Z_T &= Z_C + \frac{Z_A \cdot Z_B}{Z_A + Z_B} \\
 &= (3 + j2) + \frac{(10 + j8)(9 - j6)}{(10 + j8) + (9 - j6)} \\
 &= (3 + j2) + \frac{(138 + 12j)}{(19 + j2)} \\
 &= (3 + j2) + (7.24 - 0.13j) \\
 &= 10.24 + j1.87
 \end{aligned}$$

Total current, $I_T = \frac{V_s}{Z_T}$

$$= \frac{(100 + j0)}{10.25 + j1.87}$$

$$= \frac{9.51 - 10.34}{9.59} j.44 - 1.72j$$

$$= 9.59 \angle -10.34^\circ$$

The current lags behind the voltage by 10.34° .

Now,

Current in branch A is

$$I_A = I_T \times \frac{Z_B}{Z_A + Z_B}$$

$$Z_A + Z_B$$

$$= 9.59 \angle -10.34^\circ \times \frac{9 - j6}{19 + j2}$$

$$= 9.59 \angle -10.34^\circ \times \frac{10.82 \angle -33.69^\circ}{19 \cdot 10 \angle 6.00^\circ}$$

$$= 5.43 \angle -50.03^\circ$$

$$I_B = I_T \times \frac{Z_A}{Z_A + Z_B}$$

$$Z_A + Z_B$$

$$= 9.59 \angle -10.34^\circ \times \frac{(10 + j8)}{19 + j2}$$

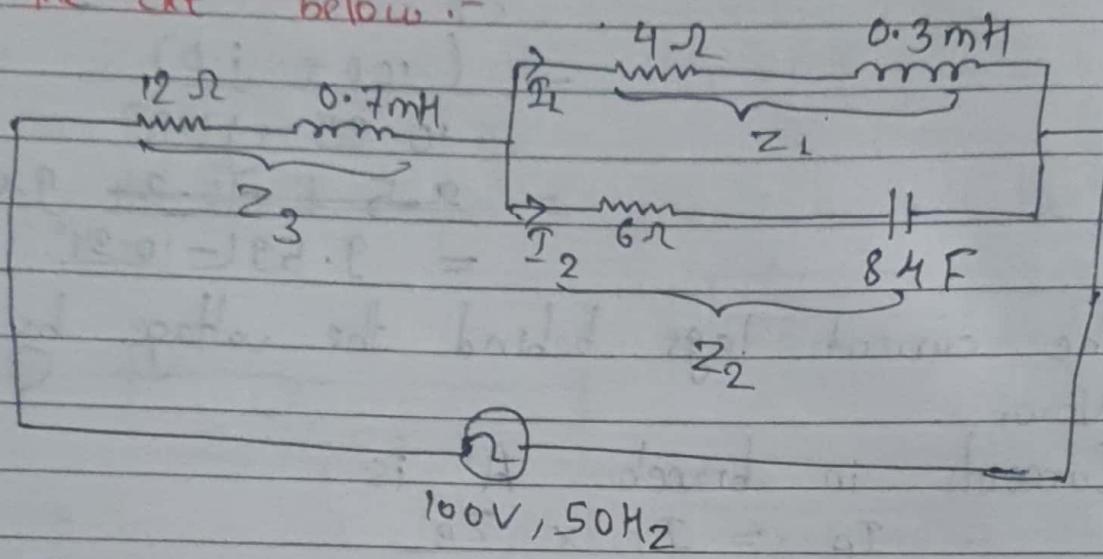
$$= 6.42 \angle 22.32^\circ$$

\therefore The phase between I_A & I_B is

$$\phi = 50.04^\circ + 22.32^\circ$$

$$= 72.36^\circ$$

Q. Determine the total impedance & current in each branch of the ckt below:-



$$\begin{aligned}
 Z_3 &= 12 + j \cdot X_L \\
 &= 12 + j \cdot 2\pi f \cdot L = 12 + j \cdot 314 \times 0.7 \times 10^{-3} \\
 &= (12 + j 0.22) \Omega
 \end{aligned}$$

$$\begin{aligned}
 Z_1 &= 4 + j \cdot X_L \\
 &= 4 + j \cdot 2\pi f \cdot L = 4 + j \cdot 314 \times 0.3 \times 10^{-3} \\
 &= (4 + j 0.09) \Omega
 \end{aligned}$$

$$\begin{aligned}
 Z_2 &= 6 - j \cdot \frac{1}{314 \times 8 \times 10^{-6}} \\
 &= 6 - j \cdot 398.09
 \end{aligned}$$

Now,

$$\begin{aligned}
 Z_T &= Z_3 + \frac{Z_1 Z_2}{Z_1 + Z_2} \\
 &= (12 + j 0.22) + \frac{(4 + j 0.09)(6 - j 398.09)}{(4 + j 0.09) + (6 - j 398.09)} \\
 &= (12 + j 0.22) + \frac{(59.83 - 1591.8j)}{10 - 398j}
 \end{aligned}$$

$$= 12 + j0.22 + 4 + 0.05j$$

$$= 16 + j0.27$$

$$= 16(0.97)$$

Now:

$$\text{Current } I_T = \frac{V}{Z_T} = \frac{100}{16(0.97)} = 6.25 - 0.97^\circ$$

(Ans)

$\theta = 42.0^\circ$

S

Power factor and its significance:

Power factor is cosine of angle between applied voltage & the resultant current flowing in circuit, the power factor is normally called lagging when the current lags the applied voltage, & leading when current leads the voltage. It is also the ratio of active power (P) to apparent power (S).

Importance of pf

$$pf = \frac{\text{active power}}{\text{apparent power}}$$

$$It\ is\ also\ ratio\ of\ active\ power\ to\ apparent\ power$$

$$pf = \frac{\text{watt}}{\text{VA}} = \frac{VI \cos\phi}{VI} = \cos\phi$$

In other words, pf. is also the ratio of resistance to impedance of the ckt.

$$pf = \frac{\text{Resistance}(R)}{\text{Impedance}(Z)} = 0 \text{ to } 1$$

$$\left[\cos\phi = \frac{R}{Z} \right] \quad (\text{0 to 1})$$

Low pf possesses the following drawbacks:

- The current ~~needed~~ obtained to obtain a given power is high, this in-turn results in the increased resistive losses, thereby resulting in decreased efficiency.
- It causes poor voltage regulation.
- It limits the output of generator & transformer, since higher current is drawn out from these results in

the increase in temp.

- d) It causes fall in terminal voltage.
- e) It causes more power loss in line & thereby reducing line efficiency.
- f) It reduces kWh, since the true power is loss.
- g) It reduces the torque of the consumer motor.

Use of Capacitor for improving pf.

In order to improve pf, capacitor are connected in parallel with the ckt (say an electric motor), since the capacitor draw current leading in phase by $\pi/2$ radian (90°). So, leading current compensates (reduces) the current req'd by inductive load, thereby the pf improves.

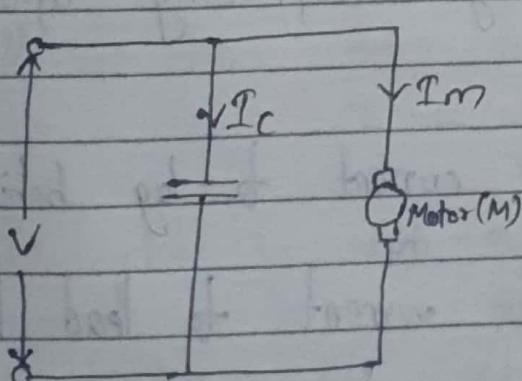


fig 'a' - use of capacitor for improving pf of a motor

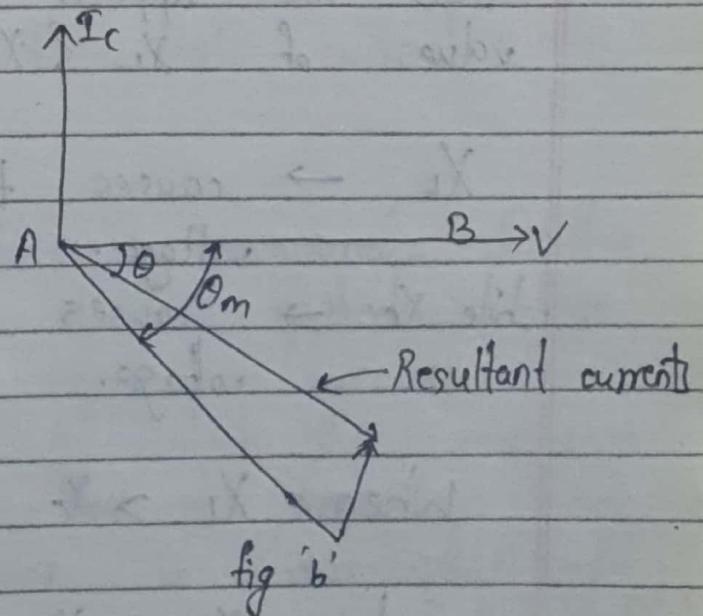


fig 'b'

Resonance in Series and Parallel RLC ckt:

- i) Series Resonance
- ii) Parallel Resonance

Series Resonance

In many of electrical ckt, resonance is very important phenomenon. The study of resonance is very useful particularly in the area of communication. For example, the ability of radio receiver to select the certain frequency, transmitted by a station & eliminate the frequencies from other stations is based on the principle of resonance.

In R-L-C series ckt:

In R-L-C series ckt, the current lags behind or leads the applied voltage depending upon the value of X_L & X_C .

$X_L \rightarrow$ causes total current to lag behind the applied voltage.

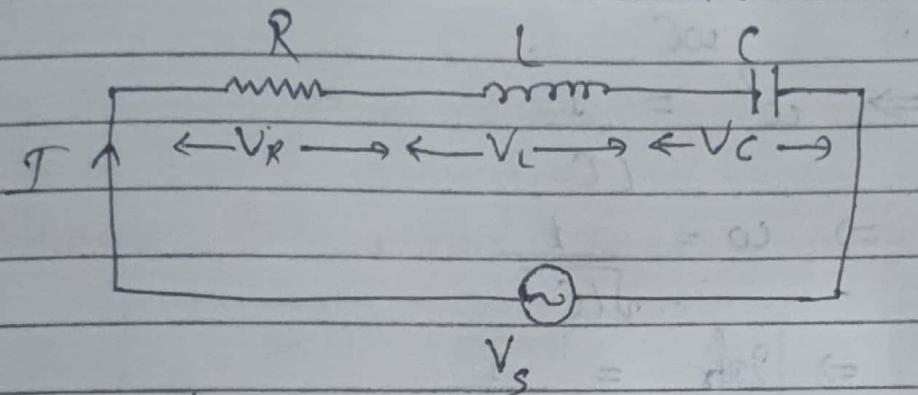
while $X_C \rightarrow$ causes total current to lead behind the applied voltage.

When $X_L > X_C$ [ckt is predominantly inductive]

when $X_C > X_L$ [ckt is predominantly capacitive]

However, if one of the parameters of RLC ckt varies in such a way that current in the ckt is inphase with applied voltage. Then ckt is said to be in resonance.

Consider a series R-L-C ckt shown in fig-(a) below:



The total impedance for the series RLC ckt:

$$Z = R + j(X_L - X_C)$$

$$= R + j\left(\omega L - \frac{1}{\omega C}\right)$$

It is clear from the ckt that current

$$I = \frac{V_s}{Z}$$

The ckt is said to be in resonance if the current in is in phase with applied voltage.

In series RLC ckt, resonance occurs when $X_L = X_C$, the frequency at which the resonance occurs is called the resonant frequency. Since, $X_L = X_C$; then impedance in series RLC ckt is purely resistive. At the resonant frequency (f_r), the voltage across the capacitance & inductance are equal in magnitude since they are 180° out of phase.

with each other. Hence the zero voltage appear across the LC combination.

At resonance,

$$X_L = X_C$$

$$\Rightarrow \omega L = \frac{1}{\omega C}$$

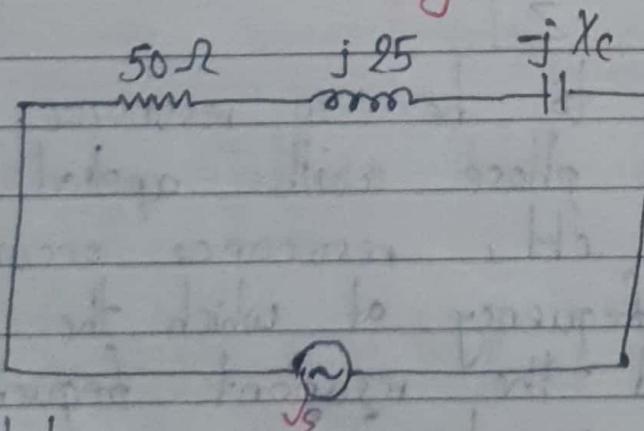
$$\Rightarrow \omega^2 = \frac{1}{LC}$$

$$\Rightarrow \omega = \frac{1}{\sqrt{LC}}$$

$$\Rightarrow 2\pi f_r = \frac{1}{\sqrt{LC}}$$

$$\Rightarrow f_r = \frac{1}{2\pi\sqrt{LC}}$$

Q. Determine the value of a capacitive reactance & impedance at resonance in fig. below!



solution:

$$X_L = X_C$$

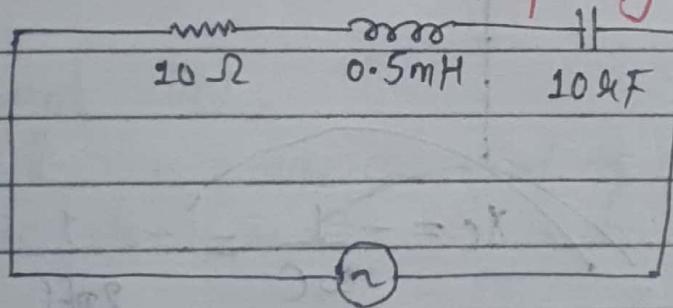
$$\Rightarrow X_L = 25$$

$$X_C = 25$$

$$\therefore \frac{1}{\omega C} = 25$$

The value of impedance at resonance
 $Z = R$
 $\Rightarrow Z = 50\Omega$

Q. Determine the resonant frequency for the ckt shown below:



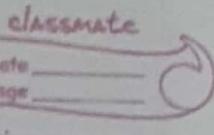
Here,

$$f_r = \frac{1}{2\pi\sqrt{LC}}$$

$$= \frac{1}{2\pi\sqrt{0.5 \times 10^{-3} \times 10 \times 10^{-6}}}$$

$$= 9950.79 \text{ Hz}$$

Difference betⁿ Series resonance & Parallel re.



Series Resonance

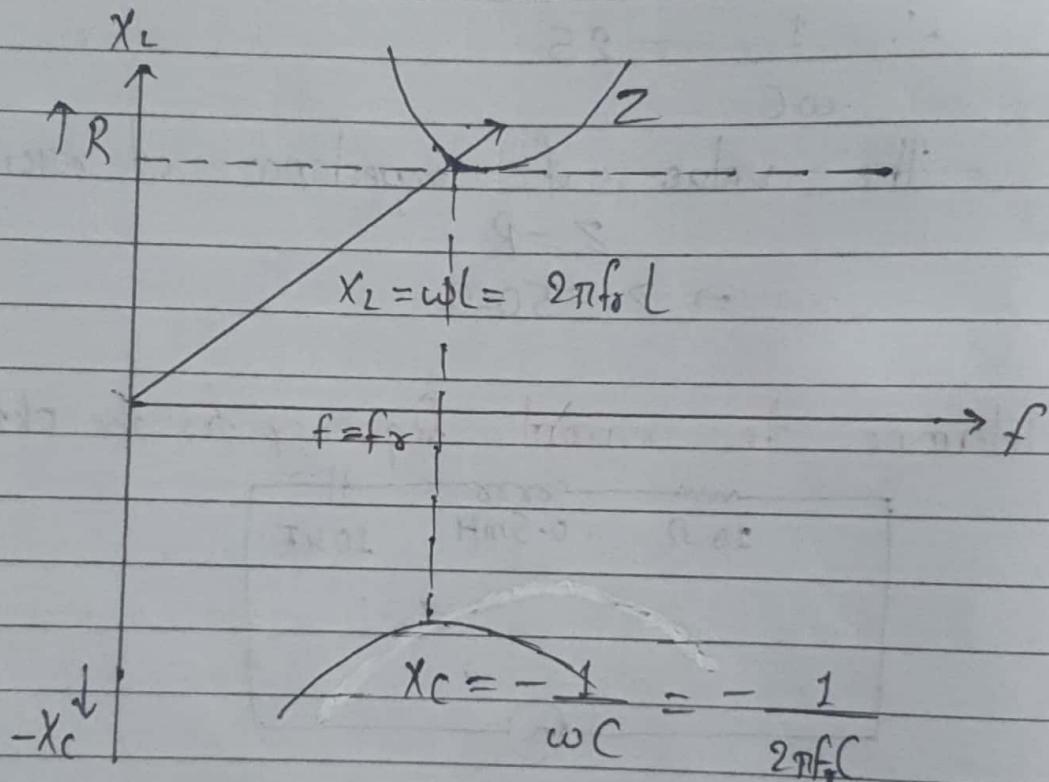
- ① Series resonance occurs when the arrangement of the components creates the minimum impedance.
- ② It refers to the resonance that occurs in circuits where capacitors and inductors are connected in series.
- ③ In series resonance, a series RLC ckt contains minimum impedance at resonant frequency.
- ④ In series resonance, maximum possible current flows through the ckt.
- ⑤ acceptor ckt

Parallel Resonance

- ① Parallel resonance occurs when the arrangement of components creates the largest impedance.
- ② It refers to the resonance that occurs in circuit where the capacitors and inductors are connected in parallel.
- ③ In parallel resonance, a parallel RLC ckt contains maximum impedance at resonant frequency.
- ④ In parallel resonance, minimum possible current flows through the ckt.
- ⑤ rejector ckt

Imp
PQ

Variation of resistance, inductive reactance, capacitive reactance, impedance & current with frequency in RLC series circuit.



(a) Resistance:

It is independent of frequency & st. line represents st. line.

(b) Inductive reactance (X_L):

$$X_L = \omega L = 2\pi f L$$

X_L is directly proportional to fr. i.e. X_L increases linearly with frequency and passes through origin.

(c) Capacitive reactance (X_C):

$$X_C = \frac{1}{\omega C} = \frac{1}{2\pi f_0 C}$$

X_C is inversely prop. to f . Its graph is asymptote to horizontal axis.

(d) Impedance (Z):

$$Z = \sqrt{R^2 + (X_L - X_C)^2}$$

$$\Rightarrow Z = \sqrt{R^2 + \left(\omega L - \frac{1}{\omega C}\right)^2}$$

At low frequency, Z is large because X_C is large. Also, at high frequency Z is again large because X_L is large.

(e) Current (I):

It is reciprocal of diff impedance so when Z is low, I is high & vice versa.

Imp
solute
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Resonance Curve

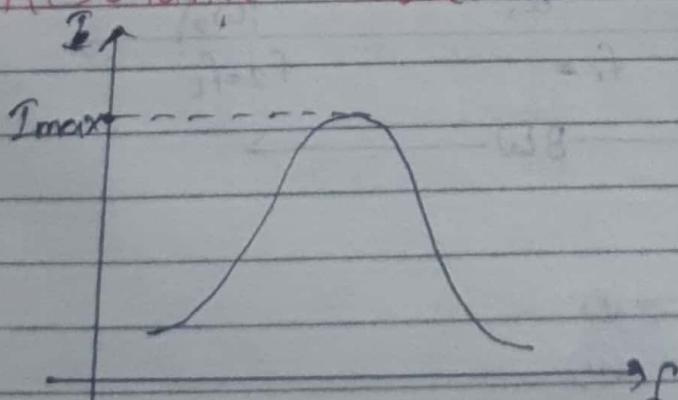


fig: Resonance curve

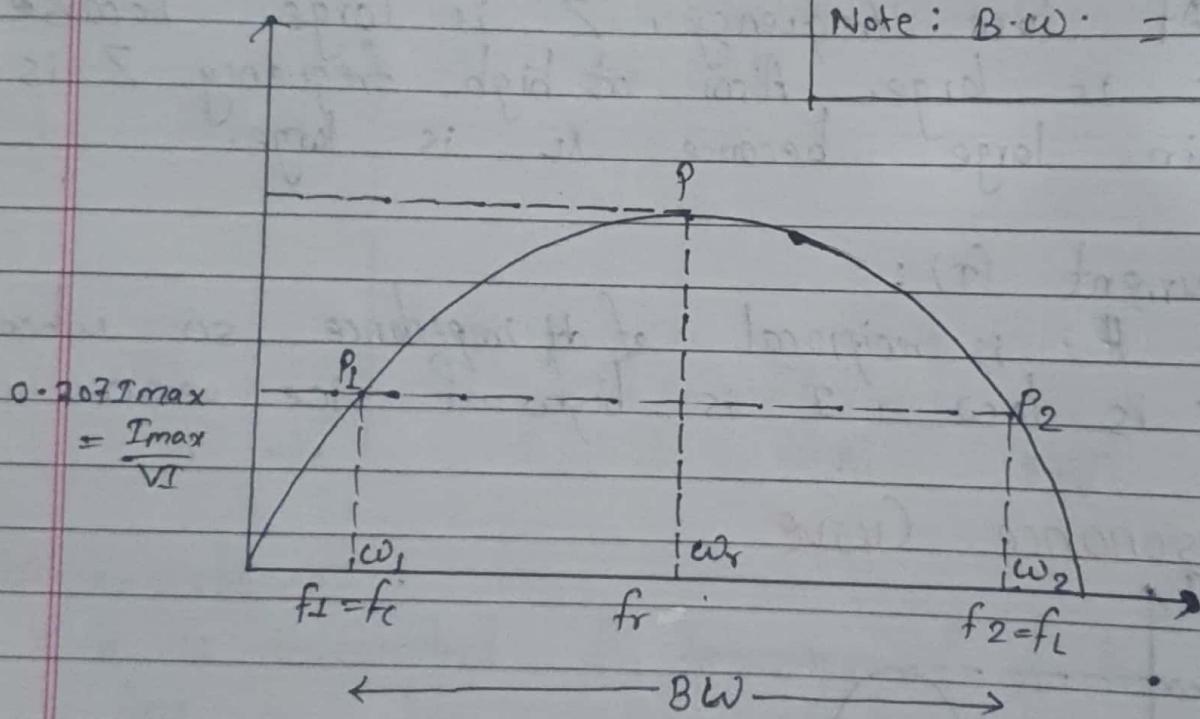
The graphical relation bet' magnitude of current & the frequency of applied emf is called the resonance curve. It is clear that the current increases as the frequency increases. It becomes

maximum at particular frequency, this frequency is called resonant frequency.

most imp Band width of a RLC ^{series} ckt

The band width of any system is the range of frequencies. It is denoted by $BW = f_2 - f_1$ shown in fig. below: The unit of band width is Hz.

Note: $B.W. = \frac{f_r}{Q}$ (quality factor)



$$0.707 I_{max} = \frac{I_{max}}{\sqrt{2}}$$

$$X_C - X_L = R$$

$$X_L - X_C = R$$

$$\frac{1}{\omega_1 C} - \omega_1 L = R \quad \text{--- (1)}$$

$$\omega_2 L - \frac{1}{\omega_2 C} = R \quad \text{--- (2)}$$

Equating ① & ⑩, we get

$$\frac{1}{\omega_1 C} - \omega_1 L = \omega_2 L - \frac{1}{\omega_2 C}$$

$$\Rightarrow \frac{1}{C} \left(\frac{1}{\omega_1} + \frac{1}{\omega_2} \right) = L(\omega_1 + \omega_2)$$

$$\Rightarrow \frac{1}{C} \left(\frac{\omega_1 + \omega_2}{\omega_1 \omega_2} \right) = L (\omega_1 + \omega_2)$$

$$\Rightarrow \frac{1}{C} \left[\frac{1}{\omega_1 \omega_2} \right] = L \quad \text{--- (III)}$$

$$\omega_1 \omega_2 = \frac{1}{LC}$$

$$[\because \underline{\omega_1 \omega_2} = \underline{\omega^2}]$$

~~$$\omega^2 = \frac{1}{LC}$$~~
$$\text{--- (IV)}$$

If we add eqn ① & ⑩, we get

$$\frac{1}{\omega_1 C} - \omega_1 L + \omega_2 L - \frac{1}{\omega_2 C} = 2R \quad \text{--- (V)}$$

Since, $\omega r^2 = \frac{1}{LC}$

$$\Rightarrow C = \frac{1}{\omega r^2} \cdot L$$

$$\Rightarrow \omega_1 \omega_2 = \omega r^2$$

$$(\omega_2 - \omega_1)L + \frac{(\omega_2 - \omega_1)\omega r^2 L}{\omega r^2} = 2R \quad \text{--- (VI)}$$

$$(\omega_2 - \omega_1)L + (\omega_2 - \omega_1)L = 2R$$

$$\Rightarrow 2L(\omega_2 - \omega_1) = 2R$$

$$\Rightarrow \omega_2 - \omega_1 = \frac{R}{L} \quad \text{--- (VI)}$$

$$f_2 - f_1 = \frac{R}{2\pi L} \quad \text{--- (VIII)}$$

$$\text{Band width} = \frac{R}{2\pi L}$$

$$f_r - f_1 = \frac{R}{4\pi L} \quad \text{--- (IX)}$$

$$f_1 = f_r - \frac{R}{4\pi L} \Rightarrow \text{The lower freq. limit}$$

$$f_2 = f_r + \frac{R}{4\pi L} \Rightarrow \text{The upper freq. limit.}$$

(imp for short note)
 Quality factor (Q) of R-L-C series ckt (Q) :

Quality factor (Q) is defined as the ratio of resonant freq. to the band width of the ckt.
 i.e.

$$Q = \frac{\text{resonant freq.}}{\text{band width}}$$

$$= \frac{1/2\pi f_r C}{R/2\pi L}$$

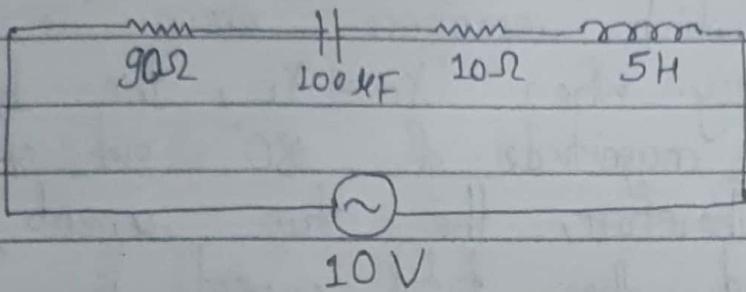
$$\therefore Q = \frac{1}{R} \sqrt{\frac{L}{C}}$$

It is also the ratio of inductive reactance of the coil to its resistance.

So,

$$Q = \frac{X_L}{R}$$

Q. In given ckt below; determine the value of Q at resonant frequency & band width of the ckt.



$$f_r = \frac{1}{2\pi\sqrt{LC}}$$

$$= \frac{1}{2\pi\sqrt{5 \times 100 \times 10^{-6}}}$$

$$= 7.11 \text{ Hz}$$

$$\text{Quality factor (Q)} = \frac{X_L}{R}$$

$$= \frac{\omega L}{R}$$

$$= \frac{2\pi f_r \cdot L}{R}$$

$$= \frac{2\pi \times 7.11 \times 5}{100}$$

$$= 2.23$$

$$\text{Band - width} = \frac{f_r}{Q}$$

$$= \frac{7.11}{2.23}$$

$$= 3.18 \text{ Hz}$$

Parallel Resonance ckt:

Basically, parallel resonance occurs when $X_C = X_L$. The frequency at which resonance occurs is called the resonant frequency. When $X_C = X_L$, the branch currents are equal in magnitude & 180° out of phase with each other. Therefore, the two currents cancel each other out, and the total current is 'zero'.

Consider the ckt shown in fig. below :

$$X_L = X_C$$

for total admittance,

$$Y = \frac{1}{R_L + j \cdot X_L} + \frac{1}{R_C - j X_C}$$

$$Y = \frac{R_L - j \cdot X_L}{R_L^2 + j \cdot X_L^2} + \frac{R_C + j X_C}{R_C^2 + j X_C^2}$$

$$Y = \frac{R_L - j \cdot \omega L}{R_L^2 + \omega^2 L^2} + \frac{R_C + j X_C}{R_C^2 + \frac{1}{\omega^2 C^2}}$$

$$= \underbrace{-\frac{R_L}{R_L^2 + \omega^2 L^2} + \frac{R_C}{R_C^2 + \frac{1}{\omega^2 C^2}}}_{\text{At resonance, } \frac{1}{\omega^2 C^2} \rightarrow 0} + j \left[\frac{\frac{1}{\omega^2 C^2}}{R_C^2 + \frac{1}{\omega^2 C^2}} \right] - \left(\frac{\omega L}{R_L^2 + \omega^2 L^2} \right) \quad \text{--- (1)}$$

At resonance, the susceptance part becomes zero.

$$\frac{\omega^2 C}{R_C^2 + \frac{1}{\omega^2 C^2}} = \frac{\omega \cdot L}{R_L^2 + \omega^2 L^2} \quad \text{--- (1)}$$

$$\omega_r \cdot L \left[R_c^2 + \frac{1}{\omega_r^2 C^2} \right] = \frac{1}{\omega_r \cdot C} \left[R_L^2 + \omega_r^2 L^2 \right]$$

$$\omega_r^2 \left[R_c^2 + \frac{1}{\omega_r^2 C^2} \right] = \frac{1}{LC} \left[R_L^2 + \omega_r^2 L^2 \right]$$

$$\omega_r^2 \cdot R_c^2 - \frac{\omega_r^2 \cdot L}{C} = \frac{1}{LC} \left[R_L^2 - \frac{L}{C^2} \right]$$

$$\omega_r^2 \left[R_c^2 - \frac{L}{C} \right] = \frac{1}{LC} \left[R_L^2 - \frac{L}{C} \right]$$

$$\Rightarrow \boxed{\omega_r = \frac{1}{\sqrt{LC}} \sqrt{\frac{R_L^2 - \frac{L}{C}}{R_c^2 - \frac{L}{C}}}} \quad \text{--- (iii)}$$

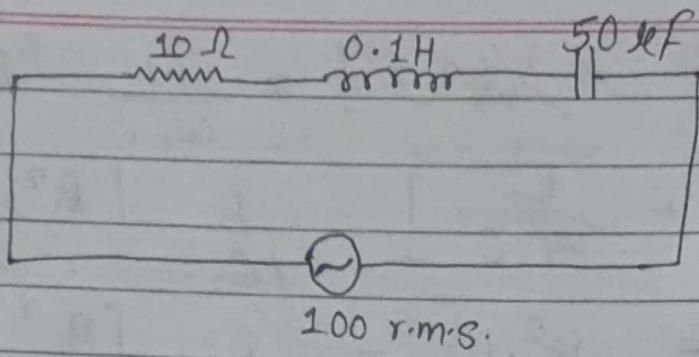
The condition for resonant frequency is given by eq? (iii) . In special case,

$$R_L = R_c$$

$$\text{So, } \omega_r = \frac{1}{\sqrt{LC}}$$

$$\Rightarrow \cancel{\boxed{f_r = \frac{1}{2\pi\sqrt{LC}}}}$$

- Q. For the ckt shown in fig. below, determine the frequency at which the ckt resonates. Also, find the voltage across the inductor at resonance & the Q factor of the ckt.



$$X_L = \omega L$$

$$X_C = \frac{1}{\omega C}$$

$$2\pi f_r \cdot L = \frac{1}{2\pi f_r \cdot C}$$

$$\Rightarrow f_r = \frac{1}{2\pi\sqrt{LC}} = \frac{1}{2\pi\sqrt{0.1 \times 50 \times 10^{-6}}} = 71.18 \text{ Hz}$$

The current passing through the ckt at resonance

$$I = \frac{V}{R} = \frac{100}{10} = 10 \text{ amp}$$

The voltage drop across the inductor

$$V_L = IX_L = I \times \omega L = I \times 2\pi f_r L \\ = 10 \times 2\pi \times 71.18 \times 0.1 \\ = 447.2 \text{ V}$$

The quality factor ~~(Q)~~,

$$Q = \frac{X_L}{R} = \frac{\omega L}{R}$$

$$= \frac{2\pi f_r L}{R}$$

$$= \frac{2\pi \times 71.18 \times 0.1}{10} \\ = 4.47$$