Ex4_2

February 10, 2019

1 Exercise 2

1.1 a)

Analoger Filter:

$$H(s) = \frac{1}{1 + \frac{s}{\omega_c}} \tag{1}$$

Bilineare transformation mit:

$$s = 2f_s \frac{z-1}{z+1} \tag{2}$$

Einsetzen ergibt die generelle Form:

$$H_d(z) = \frac{1}{1 + \frac{2f_s}{\omega_c} \frac{z - 1}{z + 1}} = \frac{1 + z^{-1}}{(1 + \frac{2f_s}{\omega_c}) + (1 - \frac{2f_s}{\omega_c})z^{-1}}$$
(3)

Analoger Tiefpassfilter kann geschrieben werden als:

$$H(s) = \frac{1}{1 + a_1 \frac{s}{\omega_c} + b_1 (\frac{s}{\omega_c})^2} \tag{4}$$

Unser gegebener Tiefpassfilter erster Ordung:

$$H(s) = \frac{1}{1 + \frac{s}{\alpha s}} \tag{5}$$

Normale Notation für die Transferfunktion von Analogen Filtern zweiter Ordnung:

$$H(s) = \frac{B_0 s^2 + B_1 s + B_2}{A_0 s^2 + A_1 s + A_2} \tag{6}$$

Durch den Vergleich der normalen Notation und unserem gegeben Filter ergibt sich:

$$B_0 = 0, B_1 = 0, B_2 = 1, A_0 = \frac{b_1}{\omega_c^2} = 0, A_1 = \frac{a_1}{\omega_c} = \frac{1}{\omega_c}, A_2 = 1$$
 (7)

Nach der Substitution von $s=2f_s\frac{z-1}{z+1}$ wollen wir folgende Form erhalten:

$$H(z) = \frac{b_0 + b_1 z^{-1} + b_2 z^{-2}}{1 + a_1 z^{-1} + a_2 z^{-2}}$$
(8)

Rechenweg:

$$\frac{B_0(2f_s\frac{z-1}{z+1})^2 + B_1(2f_s\frac{z-1}{z+1}) + B_2}{A_0(2f_s\frac{z-1}{z+1})^2 + A_1(2f_s\frac{z-1}{z+1}) + A_2} = \dots = \frac{\frac{B_04f_s^2 + B_12f_s + B_2}{A_04f_s^2 + A_12f_s + A_2} + \frac{-B_08f_s^2 + 2B_2}{A_04f_s^2 + A_12f_s + A_2}z^{-1} + \frac{B_04f_s^2 - B_12f_s + B_2}{A_04f_s^2 + A_12f_s + A_2}z^{-2}}{1 + \frac{-A_08f_s^2 + 2A_2}{A_04f_s^2 + A_12f_s + A_2}z^{-1} \frac{A_04f_s^2 - A_12f_s + A_2}{A_04f_s^2 + A_12f_s + A_2}z^{-2}}$$
(9)

Die Koeffizienten des digitalen Filters sind daher:

$$b_0 = \frac{B2 + 2B_1f_s + 4B_0f_s^2}{A_2 + 2A_1f_s + 4A_0f_s^2} = \frac{1}{1 + \frac{2f_s}{\omega_c}}b_1 = \frac{2B2 - 8B_0f_s^2}{A_2 + 2A_1f_s + 4A_0f_s^2} = \frac{2}{1 + \frac{2f_s}{\omega_c}}b_2 = \frac{B2 + 2B_1f_s - 4B_0f_s^2}{A_2 + 2A_1f_s + 4A_0f_s^2} = \frac{1}{1 + \frac{2f_s}{\omega_c}}a_1 = \frac{1}{1 + \frac{2f_s}{\omega_c}}a_2 = \frac{1}{1 + \frac{2f_s}{\omega_c}a_2}a_2 = \frac{1}{1 + \frac{2f_s}{\omega$$

1.2 b)

$$b_0 = \frac{B2 + 2B_1 f_s + 4B_0 f_s^2}{A_2 + 2A_1 f_s + 4A_0 f_s^2} = \frac{1}{1 + \frac{1}{\tan(\pi \frac{f_c}{f_s})}} b_1 = \frac{2B2 - 8B_0 f_s^2}{A_2 + 2A_1 f_s + 4A_0 f_s^2} = \frac{2}{1 + \frac{1}{\tan(\pi \frac{f_c}{f_s})}} b_2 = \frac{B2 + 2B_1 f_s - 4B_0 f_s^2}{A_2 + 2A_1 f_s + 4A_0 f_s^2} = \frac{1}{1 + \frac{1}{\tan(\pi \frac{f_c}{f_s})}} b_2 = \frac{B2 + 2B_1 f_s - 4B_0 f_s^2}{A_2 + 2A_1 f_s + 4A_0 f_s^2} = \frac{1}{1 + \frac{1}{\tan(\pi \frac{f_c}{f_s})}} b_2 = \frac{B2 + 2B_1 f_s - 4B_0 f_s^2}{A_2 + 2A_1 f_s + 4A_0 f_s^2} = \frac{1}{1 + \frac{1}{\tan(\pi \frac{f_c}{f_s})}} b_2 = \frac{B2 + 2B_1 f_s - 4B_0 f_s^2}{A_2 + 2A_1 f_s + 4A_0 f_s^2} = \frac{1}{1 + \frac{1}{\tan(\pi \frac{f_c}{f_s})}} b_2 = \frac{B2 + 2B_1 f_s - 4B_0 f_s^2}{A_2 + 2A_1 f_s + 4A_0 f_s^2} = \frac{1}{1 + \frac{1}{\tan(\pi \frac{f_c}{f_s})}} b_2 = \frac{B2 + 2B_1 f_s - 4B_0 f_s^2}{A_2 + 2A_1 f_s + 4A_0 f_s^2} = \frac{1}{1 + \frac{1}{\tan(\pi \frac{f_c}{f_s})}} b_2 = \frac{B2 + 2B_1 f_s - 4B_0 f_s^2}{A_2 + 2A_1 f_s + 4A_0 f_s^2} = \frac{1}{1 + \frac{1}{\tan(\pi \frac{f_c}{f_s})}} b_2 = \frac{B2 + 2B_1 f_s - 4B_0 f_s^2}{A_2 + 2A_1 f_s + 4A_0 f_s^2} = \frac{1}{1 + \frac{1}{\tan(\pi \frac{f_c}{f_s})}} b_2 = \frac{B2 + 2B_1 f_s - 4B_0 f_s^2}{A_2 + 2A_1 f_s + 4A_0 f_s^2} = \frac{1}{1 + \frac{1}{\tan(\pi \frac{f_c}{f_s})}} b_2 = \frac{B2 + 2B_1 f_s - 4B_0 f_s^2}{A_2 + 2A_1 f_s + 4A_0 f_s^2} = \frac{1}{1 + \frac{1}{\tan(\pi \frac{f_c}{f_s})}} b_3 = \frac{1}{1 + \frac{1}{\tan(\pi \frac{f_c}{f_s}$$

1.3 c), d), e)

c) calculate the coefficients

```
In [1]: import numpy as np
        import matplotlib.pyplot as plt
        import scipy.signal as sig
        fs = 48000 \# Hz
        fc = 10000 \#Hz
        omc = 2*np.pi*fc
        omc_wrapped = 2*fs *np.tan(np.pi * fc / fs)
        #Coefficients without pre-wrapping:
        b0 = 1 / (1 + 2 *fs / omc)
        b1 = 2 / (1 + 2 *fs / omc)
        b2 = 1 / (1 + 2 *fs / omc)
        b = [b0, b1, b2]
        a1 = 2 / (1 + 2 *fs / omc)
        \#a2 = (1 - 2 * fs / omc) / (1 + 2 * fs / omc)
        a2 = (omc - 2 * fs) / (omc + 2 * fs)
        a = ([1, a1, a2])
        #Coefficients with pre-wrapping
        b0_w = 1 / (1 + 2 *fs / omc_wrapped)
        b1_w = 2 / (1 + 2 *fs / omc_wrapped)
```

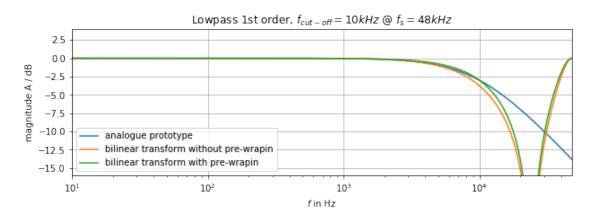
```
b_{wrapped} = [b0_{w}, b1_{w}, b2_{w}]
        a1_w = 2 / (1 + 2 *fs / omc_wrapped)
        a2_w = (1 - 2 * fs / omc_wrapped) / (1 + 2 * fs / omc_wrapped)
        a_{wrapped} = [1, a1_{w}, a2_{w}]
        print("Coefficients of the digital filter in biquad structure:\n")
        print("a = ", a)
        print("b = ",b)
        print("\nCoefficients of the digital filter in biquad structure with pre-wrapping:\n")
        print("a = ",a_wrapped)
        print("b = ",b_wrapped)
Coefficients of the digital filter in biquad structure:
a = [1, 0.7911744635176464, -0.2088255364823536]
b = [0.3955872317588232, 0.7911744635176464, 0.3955872317588232]
Coefficients of the digital filter in biquad structure with pre-wrapping:
a = [1, 0.8683475024126042, -0.13165249758739586]
b = [0.4341737512063021, 0.8683475024126042, 0.4341737512063021]
  d) visualise magnitude transfer function
In [18]: # Analog Filter
         A = [0, 1/omc, 1]
         B = [0, 0, 1]
         #bilinear transform without pre-wraping
         Om, Hd = sig.freqz(b, a, whole=True)
         temp, Ha = sig.freqs(B, A, worN=fs*Om)
         #Compute the frequency response of a digital filter and Analog Filter
         #bilinear transform with pre-wraping
         Om_w, Hd_wrapped = sig.freqz(b_wrapped, a_wrapped, whole=True)
         #freuquency vector
         #f = np.arange(0, fs, 1) #frquency range
         f = Om * fs / (2*np.pi)
         #Plot Magnitude
         plt.figure(figsize=(10,3))
         plt.semilogx(f, 20*np.log10(np.abs(Ha)), label=r'analogue prototype')
         plt.semilogx(f, 20*np.log10(np.abs(Hd)), label=r'bilinear transform without pre-wraping
```

 $b2_w = 1 / (1 + 2 *fs / omc_wrapped)$

```
plt.semilogx(f, 20*np.log10(np.abs(Hd_wrapped)), label=r'bilinear transform with pre-
plt.xlabel(r'$f$ in Hz')
plt.ylabel(r'magnitude A / dB')
plt.grid()
plt.title(r'Lowpass 1st order, $f_{cut-off} = 10 kHz$ @ $f_s = 48 kHz$')
plt.legend()
plt.axis([10, fs, -16, 4])
```

/home/max/anaconda3/lib/python3.7/site-packages/ipykernel_launcher.py:19: RuntimeWarning: dividence/max/anaconda3/lib/python3.7/site-packages/ipykernel_launcher.py:20: RuntimeWarning: dividence/max/anaconda3/lib/python3.7/site-packages/ipykernel_launcher.python3.7/site-packages/ipykernel_launcher.python3.7/site-packages/ipykernel_launcher.python3.7/site-packages/ipykernel_launcher.python3.7/site-packages/ipykernel_launcher.python3.7/site-packages/ipykernel_launcher.python3.7/site-packages/ipykernel_launcher.python3.7/site-packages/ipyke

Out[18]: [10, 48000, -16, 4]



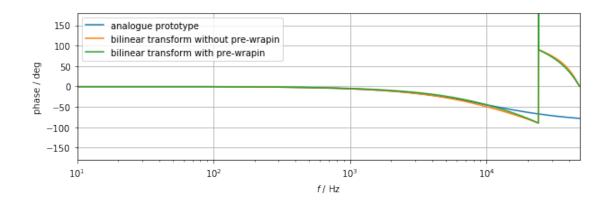
Die billineraren transformation der Frequenzantwort ist verschieden zu der des analogen Filters, durch die first-order approximation während des Mappings von der s-plane zur z-plane. Unter anderem verschiebt sich die corner frequency. Dieser Effekt wird als frequenzy wraping bezeichnet. Durch das pre-wraping, bei der die corner frequency ersetzt, sodass die Frequenz an dieser Stelle korregiert wird.

```
In [14]: #Plot Angle
```

Out[14]: [10, 48000, -180, 180]

e)

```
plt.figure(figsize=(10,3))
  plt.semilogx(f, (180 / np.pi) * np.angle(Ha), label=r'analogue prototype') # transform
  plt.semilogx(f, (180 / np.pi) * np.angle(Hd), label=r'bilinear transform without pre-
  plt.semilogx(f, (180 / np.pi) * np.angle(Hd_wrapped), label=r'bilinear transform with
  plt.xlabel(r'$f$ / Hz')
  plt.ylabel(r'phase / deg')
  plt.grid()
  plt.legend()
  #plt.title(r'Lowpass 1st order, $f_{cut-off} = 10 kHz$ @ $f_s = 48 kHz$')
  plt.axis([10, fs, -180, 180])
```



Die Stabilität und minimale phase des Systems sind durch die billineare Transformation erhalten geblieben.