

1 On the stability analysis of arbitrarily high-index
2 singular systems with multiple delays

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6 **Abstract**

This paper deals with the class of continuous-time singular linear systems with multiple time-varying delays in a range. The global exponential stability problem of this class of systems is addressed. Delay range-dependent sufficient conditions such that the system is regular, impulse-free and α -stable are developed in the linear matrix inequality (LMI) setting. Moreover, an estimate of the convergence rate of such stable systems is presented. A numerical example is employed to show the usefulness of the proposed results.

7 Keywords: Singular systems, Delay, LMIs, Spectral, Stabilization, Feedback.

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9 **1. Introduction**

Consider the linear singular time-delay system of the form

$$Ex(t) = A_0x(t) + \sum_{i=1}^m A_i x(t - \tau_i) + Bu(t), \quad \text{for all } t \in [t_0, \infty), \quad (1) \quad \{\text{delay-descriptor}\}$$

$$x(t) = \phi(t), \quad \text{for all } t_0 - \tau_m \leq t \leq t_0, \quad (2)$$

10 where $E \in \mathbb{R}^{n,n}$ is allowed to be singular. Here the state is $x : [t_0 - \tau_m, \infty) \rightarrow \mathbb{R}^n$,
11 and the (constant) time-delays satisfy $0 < \tau_1 < \tau_2 < \dots < \tau_m$. The capital letters
12 are real-valued matrices of appropriate dimensions. The system is called *free* (*or*
13 *DDAE*) if we let $u \equiv 0$, i.e., the system reads

$$Ex(t) = A_0x(t) + \sum_{i=1}^m A_i x(t - \tau_i). \quad (3) \quad \{\text{free system}\}$$

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14 **The motivation for the system description 1 in the context of designing
15 controllers lies in its generality in modelling interconnected systems.**

16 The rest of the paper is organized as follows. In Section 2, some definitions
17 concerning about the solution and the system classification are stated. Auxiliary
18 Lemmas about the solution's presentation and the non-advanced test are also
19 recalled. In Section 3, our first main results about the stability of arbitrarily high-
20 index system are given, making use of both approaches above. In Sections 4, we
21 discuss the stabilization problem via the Lyapunov-Krasovskii functional method.
22 Finally, in Section 5, a numerical example and the conclusion are given.

23 **2. Preliminaries**

24 To keep the brevity of this research, we refer the interested readers to Ascher
25 and Petzold [1], Campbell [2], Shampine and Gahinet [3], Ha [4], Ha and
26 Mehrmann [5] for the solvability analysis of the IVP (1).

27 **Definition 1.** *The null solution $x = 0$ of the free system (3) is called exponentially
28 stable if there exist positive constants δ and γ such that for any consistent initial
29 function $\varphi \in C([-\tau, 0], \mathbb{R}^n)$, the solution $x = x(t, \varphi)$ of the corresponding IVP
30 to (3) satisfies*

$$\|x(t)\| \leq \delta e^{-\gamma t} \|\varphi\|_\infty, \text{ for every } t \geq 0.$$

31 **Definition 2.** i) Consider the DDAE (1). The matrix pair (E, A_0) is called regular
32 if the polynomial $\det(\lambda E - A_0)$ is not identically zero.
33 ii) The sets $\sigma(E, A_0, \dots, A_m) := \{\lambda \in \mathbb{C} \mid \det(\lambda E - A_0 - e^{-\lambda \tau_i} A_i) = 0\}$ is called
34 the spectrum of (1).

35 Provided that the pair (E, A_0) is regular, we can transform them to the Kronecker-
36 Weierstraß canonical form as follows.

37 **Lemma 3.** (Dai [6], Kunkel and Mehrmann [7]) *Provided that the matrix pair
38 (E, A_0) is regular, then there exist regular matrices $W, T \in \mathbb{R}^{n,n}$ such that*

$$(WET, WA_0T) = \left(\begin{bmatrix} I & 0 \\ 0 & N \end{bmatrix}, \begin{bmatrix} J & 0 \\ 0 & I \end{bmatrix} \right), \quad (4) \quad \{\text{KW form}\}$$

39 where N is a nilpotent, upper triangular matrix of nilpotency index ν . We also
40 say that the pair (E, A_0) has an index ν , i.e., $\text{ind}(E, A_0) = \nu$. Furthermore, the
41 system (1) is called impulse-free (index 1, or strangeness-free) if $N = 0$.

42 **Remark 1.** In general, the two concepts index and stability are independent. In
 43 fact, Examples 5 in Ha and Nam [8] has illustrated that there exist systems with
 44 arbitrarily high-index (and hence, not impulse-free) which are stable.

45 **Lemma 4.** *For a nilpotent, upper triangular matrix N of nilpotency index ν , the
 46 matrix $I - \lambda N$ is invertible for all $\lambda \in \mathbb{C}$, and $\det(I - \lambda N) = 1$. Furthermore,
 47 the following identity holds true.*

$$(I - \lambda N)^{-1} = I + \sum_{i=1}^{\nu} (\lambda N)^i.$$

48 PROOF. The proof is simple and can be found in classical matrix theory text-
 49 books, for example [9].

50 2.1. System classification

51 It is well-known (see e.g. Bellman and Cooke [10], Hale and Lunel [11])
 52 that in general, time-delayed systems has been classified into three different types
 53 (retarded, neutral, advanced). For example, the time-delayed equation

$$a_0 \dot{x}(t) + a_1 \dot{x}(t - \tau) + b_0 x(t) + b_1 x(t - \tau) = f(t)$$

54 is retarded if $a_0 \neq 0$ and $a_1 = 0$; is neutral if $a_0 \neq 0$, $a_1 \neq 0$; is advanced if
 55 $a_0 = 0$, $a_1 \neq 0$, $b_0 \neq 0$. This classification is based on the smoothness comparison
 56 between $x(t)$ and $x(t - \tau)$. In literature, not only the theoretical but also the
 57 numerical solution has been studied mainly for retarded and neutral systems, due
 58 to their appearance in various applications. For this reason, in Ha [4], Ha and
 59 Mehrmann [5], Unger [12] the authors proposed a concept of *non-advancedness*
 60 for the free system (see Definition 5 below). We also notice, that even though not
 61 clearly proposed, due to the author's knowledge, so far results for delay-descriptor
 62 are only obtained for certain classes of non-advanced systems, e.g. Ascher and
 63 Petzold [1], Shampine and Gahinet [3], Zhu and Petzold [13, 14], Michiels [15],
 64 Phat and Sau [16], Sau et al. [17], Cui et al. [18], Ngoc [19].

65 **Definition 5.** *A regular delay-descriptor system (1) is called non-advanced if for
 66 any consistent and continuous initial function φ , there exists a continuous, piece-
 67 wise differentiable solution $x(t)$.*

68 Making use of Lemma 3, we change the variable $x = Ty$ and scale the whole
 69 system (3) with W to obtain the transformed system

$$\begin{bmatrix} I & 0 \\ 0 & N \end{bmatrix} \dot{y}(t) = \begin{bmatrix} J & 0 \\ 0 & I \end{bmatrix} y(t) + \sum_{i=1}^m \begin{bmatrix} \tilde{A}_{i,1} & \tilde{A}_{i,2} \\ \tilde{A}_{i,3} & \tilde{A}_{i,4} \end{bmatrix} y(t - \tau_i), \quad (5) \quad \{\text{eq9}\}$$

⁷⁰ where $WA_iT = \begin{bmatrix} \tilde{A}_{i,1} & \tilde{A}_{i,2} \\ \tilde{A}_{i,3} & \tilde{A}_{i,4} \end{bmatrix}$ for all $i = 1, \dots, m$. The following lemma gives us
⁷¹ the necessary and sufficient condition for the non-advancedness of system (3).

⁷² **Lemma 6.** *i) System (3) is non-advanced if and only if the matrix coefficients of
⁷³ the transformed system (5) satisfy*

$$N \begin{bmatrix} \tilde{A}_{i,3} & \tilde{A}_{i,4} \end{bmatrix} = \begin{bmatrix} 0 & 0 \end{bmatrix} \text{ for all } i = 1, \dots, m. \quad (6) \quad \{\text{non-advanced cond.}\}$$

⁷⁴ *ii) Consequently, system (5) has exactly the same solution as the so-called index-
⁷⁵ reduced system*

$$\underbrace{\begin{bmatrix} I & 0 \\ 0 & 0 \end{bmatrix}}_{=: \tilde{E}} \dot{y}(t) = \underbrace{\begin{bmatrix} J & 0 \\ 0 & I \end{bmatrix}}_{=: \tilde{A}_0} y(t) + \sum_{i=1}^m \underbrace{\begin{bmatrix} \tilde{A}_{i,1} & \tilde{A}_{i,2} \\ \tilde{A}_{i,3} & \tilde{A}_{i,4} \end{bmatrix}}_{=: \tilde{A}_i} y(t - \tau_i). \quad (7) \quad \{\text{index reduced system}\}$$

PROOF. Partitioning $y := \begin{bmatrix} y_1 \\ y_2 \end{bmatrix}$ conformably, we can rewrite system (5) as follows

$$\begin{aligned} \dot{y}_1 &= Jy_1 + \sum_{i=1}^m \begin{bmatrix} \tilde{A}_{i,1} & \tilde{A}_{i,2} \end{bmatrix} y(t - \tau_i), \\ N\dot{y}_2 &= y_2 + \sum_{i=1}^m \begin{bmatrix} \tilde{A}_{i,3} & \tilde{A}_{i,4} \end{bmatrix} y(t - \tau_i), \end{aligned} \quad (8) \quad \{\text{eq14.2}\}$$

⁷⁶ The second equation has a unique solution

$$y_2(t) = - \begin{bmatrix} \tilde{A}_{i,3} & \tilde{A}_{i,4} \end{bmatrix} y(t - \tau_i) - \sum_{j=1}^{\nu} \sum_{i=1}^m N^i \begin{bmatrix} \tilde{A}_{i,3} & \tilde{A}_{i,4} \end{bmatrix} y^{(j)}(t - \tau_i).$$

⁷⁷ Since the system (3) is non-advanced, then so is system (5). Consequently, $y(t)$
⁷⁸ must not depend on $y^{(j)}(t - \tau_i)$ for all $j = 1, \dots, \nu$ and $i = 1, \dots, m$, which implies
⁷⁹ the identity (12). Then, the second claim is trivially followed.

⁸⁰ **Remark 2.** From Lemma 6 ii), we see that if system (3) is non-advanced, then
⁸¹ there is a linear, bijective mapping $x \mapsto y = T^{-1}x$ (where T is the matrix given
⁸² in the Kronecker-Weierstraß form (4)) between the solution set of the high-index
⁸³ system (3) and the impulse-free system (7). This will play the key role in the
⁸⁴ stability analysis in Section 3.

85 **Remark 3.** Since the numerical computation of the Kronecker-Weierstraß form
 86 (4) is quite complicated and unstable (see [20]), Lemma 6 has more theoretical
 87 than numerical meaning for checking the non-advancedness of (3). Below we
 88 will construct another test, which is more practical.

89 Assume that the pair (E, A_0) is regular with index $\text{ind}(E, A_0) = \nu$. We want
 90 to give a simple check whether the system (3) is non-advanced or not. In analogu-
 91 ous to the case of DAEs, see e.g. Brenan et al. [21], Kunkel and Mehrmann [7],
 92 we aim to extract the so-called *underlying delay equation* of the form

$$\dot{x}(t) = \mathbf{A}_0 x(t) + \sum_{i=1}^m \mathbf{A}_i x(t - \tau_i) + \sum_{i=1}^m \mathbf{F}_i \dot{x}(t - \tau_i), \quad (9) \quad \{\text{underlying DDEs}\}$$

93 from an augmented system consisting of system (3) and its derivatives, which read
 94 in details

$$\frac{d^j}{dt^j} \left(E \dot{x}(t) - A_0 x(t) - \sum_{i=1}^m A_i x(t - \tau_i) \right) = 0, \text{ for all } j = 0, 1, \dots, \nu.$$

We rewrite these equations into the so-called *inflated system*

$$\begin{aligned} & \underbrace{\begin{bmatrix} E & & & \\ -A_0 & E & & \\ & \ddots & \ddots & \\ & & -A_0 & E \end{bmatrix}}_{\mathcal{E}} \underbrace{\begin{bmatrix} \dot{x} \\ \ddot{x} \\ \vdots \\ x^{(\nu+1)} \end{bmatrix}}_{\mathcal{A}_0} = \underbrace{\begin{bmatrix} A_0 & 0 & \dots & 0 \\ 0 & 0 & \dots & 0 \\ \vdots & \vdots & & \vdots \\ 0 & 0 & \dots & 0 \end{bmatrix}}_{\mathcal{A}_0} \underbrace{\begin{bmatrix} x \\ \dot{x} \\ \vdots \\ x^{(\nu)} \end{bmatrix}}_{\mathcal{A}_0} \\ & + \sum_{i=1}^m \underbrace{\begin{bmatrix} A_i & & & \\ & A_i & & \\ & & \ddots & \\ & & & A_i \end{bmatrix}}_{\mathcal{A}_i} \underbrace{\begin{bmatrix} x(t - \tau_i) \\ \dot{x}(t - \tau_i) \\ \vdots \\ x^{(\nu)}(t - \tau_i) \end{bmatrix}}_{\mathcal{A}_i}. \end{aligned} \quad (10) \quad \{\text{inflated}\}$$

Here the matrix coefficients are $\mathcal{E}, \mathcal{A}_0, \mathcal{A}_i \in \mathbb{R}^{(\nu+1)n, (\nu+1)n}$ for all $i = 1, \dots, m$. For the reader's convenience, below we will use MATLAB notations. An underlying delay system (9) can be extracted from (10) if and only if there exists a matrix $P = [P_0 \ P_1 \ \dots \ P_\nu]^T$ in $\mathbb{R}^{(\nu+1)n, n}$ such that

$$\begin{aligned} P^T \mathcal{E} &= [I_n \ 0_{n, \nu n}], \\ P^T \mathcal{A}_i &= [* \ * \ 0_{n, (\nu-1)n}], \text{ for all } i = 1, \dots, m, \end{aligned}$$

95 where $*$ stands for an arbitrary matrix. Consequently, P is the solution to the
 96 following linear systems

$$\begin{bmatrix} \mathcal{E}^T \\ \mathcal{A}_1(:, 2n+1 : end)^T \\ \vdots \\ \mathcal{A}_m(:, 2n+1 : end)^T \end{bmatrix} P = \begin{bmatrix} [I_n \ 0_{n,\nu n}]^T \\ 0_{(\nu-1)n,n} \\ \vdots \\ 0_{(\nu-1)n,n} \end{bmatrix}.$$

97 Therefore, making use of Crammer's rule we directly obtain the simple check for
 98 the non-advancedness of system (3) in the following theorem.

99 **Theorem 7.** Consider the zero-input descriptor system (3) and assume that the
 100 pair (E, A_0) is regular with index $\text{ind}(E, A_0) = \nu$. Then, this system is non-
 101 advanced if and only if the following rank condition is satisfied

$$\text{rank} \begin{bmatrix} \mathcal{E}^T \\ \mathcal{A}_1(:, 2n+1 : end)^T \\ \vdots \\ \mathcal{A}_m(:, 2n+1 : end)^T \end{bmatrix} = \text{rank} \begin{bmatrix} \mathcal{E}^T \\ \mathcal{A}_1(:, 2n+1 : end)^T \\ \vdots \\ \mathcal{A}_m(:, 2n+1 : end)^T \end{bmatrix} \left| \begin{array}{c} [I_n \ 0_{n,\nu n}]^T \\ 0_{(\nu-1)n,n} \\ \vdots \\ 0_{(\nu-1)n,n} \end{array} \right|. \quad (11) \quad \{\text{adv. check eq.}\}$$

102 Theorem 7 applied to the index two case straightly gives us the following
 103 corollary.

104 **Corollary 8.** Consider the zero-input descriptor system (3) and assume that the
 105 pair (E, A_0) is regular with index $\text{ind}(E, A_0) = 2$. Then, system (3) is non-
 106 advanced if and only if the following identity hold true.

$$\text{rank} \begin{array}{|c} \hline E^T & -A_0^T & 0 \\ \hline 0 & E^T & -A_0^T \\ 0 & 0 & E^T \\ \hline 0 & 0 & A_1^T \\ \vdots & \vdots & \vdots \\ 0 & 0 & A_m^T \\ \hline \end{array} = n + \text{rank} \begin{bmatrix} E^T & -A_0^T \\ 0 & E^T \\ \hline 0 & A_1^T \\ \vdots & \vdots \\ 0 & A_m^T \end{bmatrix}. \quad (12) \quad \{\text{check advanced}\}$$

107 3. Stability

108 3.1. Spectral approach

109 The stability analysis of the null solution of (1) in this work is based on a
 110 spectrum determined growth property of the solutions, which allows us to infer

111 stability information from the location of the characteristic roots. For instance,
 112 exponential stability will be related to a strictly negative spectral abscissa (the
 113 supremum of the real parts of the characteristic roots). As we shall see, the spec-
 114 tral abscissa of (1) may not be a continuous function of the delays. Moreover,
 115 this may lead to a situation where infinitesimal delay perturbations destabilise an
 116 exponentially stable system. These properties are very similar to the spectral prop-
 117 erties of neutral equations (see, e.g. [2, Section 2]), which are known to be closely
 118 related to DDAEs [3].

119 **Proposition 9.** ([15, 22]) Consider the linear, homogeneous DDAE (3). Further-
 120 more, assume that it is regular, impulse-free. Then it is stable if and only if the
 121 corresponding spectrum of this system lies entirely on the left half plane and it is
 122 bounded away from the imaginary axis.

123 The following lemma plays the key role in the proof of the main Theorem 11
 124 below.

125 **Lemma 10.** Consider the linear, homogeneous DDAE (3). Furthermore, assume
 126 that it is non-advanced. Then system (3) has the same spectrum (without counting
 127 multiplicity), and also the same stability property as the index-reduced system (7).

128 PROOF. Firstly, we will show that both systems (3) and (7) have the same spec-
 129 trum (without counting multiplicity) as the system (5). Due to the variable trans-
 130 formation $x = Ty$ and the identity

$$W(\lambda E - A_0 - e^{-\lambda \tau_i} A_i) T = \begin{bmatrix} I & 0 \\ 0 & N \end{bmatrix} - \begin{bmatrix} J & 0 \\ 0 & I \end{bmatrix} - e^{-\lambda \tau_i} \begin{bmatrix} \tilde{A}_{i,1} & \tilde{A}_{i,2} \\ \tilde{A}_{i,3} & \tilde{A}_{i,4} \end{bmatrix},$$

131 it is straightforward that

$$\sigma(E, A_0, \dots, A_m) = \sigma \left(\begin{bmatrix} I & 0 \\ 0 & N \end{bmatrix}, \begin{bmatrix} J & 0 \\ 0 & I \end{bmatrix}, \begin{bmatrix} \tilde{A}_{1,1} & \tilde{A}_{1,2} \\ \tilde{A}_{1,3} & \tilde{A}_{1,4} \end{bmatrix}, \dots, \begin{bmatrix} \tilde{A}_{m,1} & \tilde{A}_{m,2} \\ \tilde{A}_{m,3} & \tilde{A}_{m,4} \end{bmatrix} \right). \quad (13) \quad \{\text{eq11}\}$$

Now let us consider the right hand side of (13), due to Lemma 4 we see that for an arbitrary $\lambda \in \mathbb{C}$

$$\begin{aligned} & \det \left(\begin{bmatrix} I & 0 \\ 0 & N \end{bmatrix} - \begin{bmatrix} J & 0 \\ 0 & I \end{bmatrix} - \sum_{i=1}^m e^{-\lambda \tau_i} \begin{bmatrix} \tilde{A}_{i,1} & \tilde{A}_{i,2} \\ \tilde{A}_{i,3} & \tilde{A}_{i,4} \end{bmatrix} \right) \\ &= \det \left(\begin{bmatrix} I & 0 \\ 0 & (I - \lambda N)^{-1} \end{bmatrix} \cdot \begin{bmatrix} I - J - \sum_{i=1}^m e^{-\lambda \tau_i} \tilde{A}_{i,1} & -\sum_{i=1}^m e^{-\lambda \tau_i} \tilde{A}_{i,2} \\ -\sum_{i=1}^m e^{-\lambda \tau_i} \tilde{A}_{i,3} & \lambda N - I - \sum_{i=1}^m e^{-\lambda \tau_i} \tilde{A}_{i,4} \end{bmatrix} \right). \end{aligned}$$

Due to Lemma 4 and (6), we have

$$(I + \sum_{i=1}^{\nu} (\lambda N)^i) \cdot \sum_{i=1}^m e^{-\lambda \tau_i} \tilde{A}_{i,3} = \sum_{i=1}^m e^{-\lambda \tau_i} \tilde{A}_{i,3},$$

$$(I + \sum_{i=1}^{\nu} (\lambda N)^i) \cdot \left(\lambda N - I - \sum_{i=1}^m e^{-\lambda \tau_i} \tilde{A}_{i,4} \right) = -I - \sum_{i=1}^m e^{-\lambda \tau_i} \tilde{A}_{i,4}.$$

Hence, it follows that for any $\lambda \in \mathbb{C}$

$$\det \left(\begin{bmatrix} I & 0 \\ 0 & N \end{bmatrix} - \begin{bmatrix} J & 0 \\ 0 & I \end{bmatrix} - \sum_{i=1}^m e^{-\lambda \tau_i} \begin{bmatrix} \tilde{A}_{i,1} & \tilde{A}_{i,2} \\ \tilde{A}_{i,3} & \tilde{A}_{i,4} \end{bmatrix} \right)$$

$$= \det \left(\begin{bmatrix} I - J - \sum_{i=1}^m e^{-\lambda \tau_i} \tilde{A}_{i,1} & -\sum_{i=1}^m e^{-\lambda \tau_i} \tilde{A}_{i,2} \\ -\sum_{i=1}^m e^{-\lambda \tau_i} \tilde{A}_{i,3} & -I - \sum_{i=1}^m e^{-\lambda \tau_i} \tilde{A}_{i,4} \end{bmatrix} \right),$$

132 which yields that

$$\sigma \left(\begin{bmatrix} I & 0 \\ 0 & N \end{bmatrix}, \begin{bmatrix} J & 0 \\ 0 & I \end{bmatrix}, \begin{bmatrix} \tilde{A}_{1,1} & \tilde{A}_{1,2} \\ \tilde{A}_{1,3} & \tilde{A}_{1,4} \end{bmatrix}, \dots, \begin{bmatrix} \tilde{A}_{m,1} & \tilde{A}_{m,2} \\ \tilde{A}_{m,3} & \tilde{A}_{m,4} \end{bmatrix} \right) = \sigma(\tilde{E}, \tilde{A}_0, \dots, \tilde{A}_m). \quad (14) \quad \{ \text{eq12} \}$$

133 From (13) and (14) we have $\sigma(E, A_0, \dots, A_m) = \sigma(\tilde{E}, \tilde{A}_0, \dots, \tilde{A}_m)$.

134 Now let us consider the stability of (3). Clearly, system (3) is stable if and only if
135 system (5) is also stable, and vice versa. Hence, due to Lemma 6ii) we obtain the
136 desired claim. \square

137 Lemma 10 shows that a non-advanced system (3) has the same spectrum and
138 the same stability property as the index-reduced system (7), which is impulse-free.
139 Therefore, Proposition 9 applied to system (7) leads us to the following theorem.

140 **Theorem 11.** *Consider the free system (3). Furthermore, we assume that the
141 matrix pair (E, A_0) is regular. Then, (3) is exponentially stable if and only if the
142 following assertions hold.*

- 143 i) System (3) is non-advanced.
- 144 ii) The spectrum $\sigma(E, A_0, \dots, A_m)$ lies entirely on the left half plane and it is
145 bounded away from the imaginary axis.

146 **Remark 4.** Again, we notice that due to the complication in computing the Kronecker-
147 Weierstraß form (4), we will not compute the spectrum $\sigma(E, A_0, \dots, A_m)$ based
148 on (4). Instead, we refer the reader to the spectral discretisation approach in
149 [15]. Nevertheless, since this method has only been developed for impulse-free
150 (or index-1) system, we need the pre-processing step as in Lemma 12 below.

151 Let us consider the (reordered) QZ-decomposition ([23]) of the matrix pair
 152 (E, A_0) as follows

$$QEZ^T = \begin{bmatrix} \Sigma_E & \hat{E}_2 \\ 0 & N_E \end{bmatrix}, \quad QA_0Z^T = \begin{bmatrix} J_A & \hat{A}_2 \\ 0 & \Sigma_A \end{bmatrix}, \quad QA_iZ^T = \begin{bmatrix} \hat{A}_{i,1} & \hat{A}_{i,2} \\ \hat{A}_{i,3} & \hat{A}_{i,4} \end{bmatrix}, \quad (15) \quad \{\text{eq15}\}$$

153 where Σ_E, Σ_A are nonsingular, upper triangular matrices, N_E is a nilpotent, upper
 154 triangular matrix.

155 Using the same argument as in Lemma 6, we have the following lemma.

156 **Lemma 12.** *Consider the free system (3) and the QZ-decomposition (15). Then,
 157 the following assertions hold true.*

- 158 i) *System (3) is non-advanced if and only if $N_E \Sigma_A^{-1} [\hat{A}_{i,3} \quad \hat{A}_{i,4}] = 0$ for all $i =$
 159 $1, \dots, m$.*
- 160 ii) *If this is the case, then there is a linear, bijective mapping $x \mapsto y = Zx$ (where
 161 Z is the matrix given in (15)) between the solution set of the high-index system (3)
 162 and the following impulse-free system*

$$\begin{bmatrix} \Sigma_E & \hat{E}_2 \\ 0 & \mathbf{0} \end{bmatrix} \dot{y}(t) = \begin{bmatrix} J_A & \hat{A}_2 \\ 0 & \Sigma_A \end{bmatrix} y(t) + \sum_{i=1}^m \begin{bmatrix} \hat{A}_{i,1} & \hat{A}_{i,2} \\ \hat{A}_{i,3} & \hat{A}_{i,4} \end{bmatrix} y(t - \tau_i). \quad (16) \quad \{\text{impulse free system}\}$$

163 PROOF. The proof is essentially the same as the proof of Lemma 6 and will be
 164 omitted to keep the brevity of this research.

165 3.2. Lyapunov-Krasovskii approach

166 4. Robust stabilization

167 5. Conclusion and Outlook

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