

1 **CONTROLLABILITY OF SECOND ORDER DISCRETE-TIME
2 DESCRIPTOR SYSTEMS
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5 **Abstract.** This paper is mainly devoted to controllability of second order discrete-time descrip-
6 tor systems. Characterizations for controllability different concepts are derived and feedback designs
7 are investigated by transforming the system into an appropriate form. Some observability conditions
8 are also studied for these descriptor systems. It shows how the classical conditions for first order
9 discrete-time systems can be generalized to second order discrete-time descriptor systems. We will
10 develop the algebraic approach to establish concise and stably computed condensed forms, which
11 play a key role in our controllability analysis. This work completes the researches about controll-
12 ability/observability of higher order descriptor systems.

13 **Keywords.** Second order systems; Descriptor systems; causal controllability; Complete controlla-
14 bility; Strong controllability; Feedback.

15 **Mathematics Subject Classifications:** 06B99, 34D99, 47A10, 47A99, 65P99. 93B05, 93B07,
16 93B10.

17 **1. Introduction.** In this paper we study the second order descriptor system in
18 discrete-time

$$\begin{aligned} Mx(n+2) + Dx(n+1) + Kx(n) &= Bu(n) \quad \text{for all } n \geq n_0, \\ y(k) &= Cx(k), \\ x(n_0) &= x_0, \quad x(n_0+1) = x_1, \end{aligned} \tag{1.1} \quad \{\text{descriptor 2nd order discrete}\}$$

19 where $M, D, K \in \mathbb{R}^{d,d}$, $B \in \mathbb{R}^{d,p}$, $C \in \mathbb{R}^{q,d}$ are real, constant coefficient matrices.
20 Here $x = \{x(n)\}_{n \geq n_0}$, $u = \{u(n)\}_{n \geq n_0}$ are real-valued vector sequences. System (1.1)
21 is concerned with the singular difference equations (SiDE)

$$Mx(n+2) + Dx(n+1) + Kx(n) = f(n) \quad \text{for all } n \geq n_0. \tag{1.2} \quad \{\text{SiDE 2nd ord}\}$$

22 They arise as mathematical models in various fields such as population dynamics,
23 economics, the discretization of some differential-algebraic equations (DAEs) or par-
24 tial differential equations (PDEs), from sampling in dynamical systems; e.g., see
25 [6, 12, 21, 22, 27]. Recently, solvability and stability of SiDEs of second order has
26 been investigated in [23, 24, 29]. However, controllability for these systems has not
27 been reached although it has been well-studied for both DAEs and SiDEs of first
28 order [5, 11, 19].

29 In classical approach [4, 14, 20, 30, 31], usually new variables are introduced such
30 that a high order system can be reformulated as a first order one. As will be seen
31 later in Examples 2.6 and 2.7, this method, however, is not only non-unique but
32 also has presented some substantial disadvantages from both theoretical and numer-
33 ical viewpoints. These drawbacks include (1) give a wrong prediction on the index
34 and hence, increase the complexity of a numerical solution method, (2) increase the
35 computational effort due to the bigger size of a reformulated system, (3) affect the
36 controllability/observability of the system itself, i.e. a first order resulting system is
37 uncontrollable, even though the original one is.

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38 To overcome these obstacles, the *algebraic approach*, which treats the system
 39 directly without reformulating it, has been studied in [25, 28, 34, 35]. Nevertheless,
 40 the proposed method therein has also presented some additional difficulties as follows.
 41 Firstly, important condensed forms numbered (2.3)-(2.5) are big and complicated,
 42 which is really hard to be generalized for higher order systems. More importantly, the
 43 system transformations are not unitary, and hence, condensed forms and characteristic
 44 values could not be stably computed. Secondly, even though characterizations for the
 45 impulse controllability are given, a feedback strategy to obtain gain matrices is still
 46 missing. Finally, since feedbacks are involved in the system transformations, they
 47 may destroy desired properties, in particular the system observability, see [25, Sec.4].

48 From the observation above, the motivation of this work includes: Firstly, we
 49 want to develop and modify the algebraic method suggested in [25] to make it more
 50 convenient to study different controllability concepts for second order discrete-time
 51 descriptor systems. Secondly, we want to fill in missing gaps in previous researches
 52 that we have mentioned above for causal controllability. In particular, motivated
 53 by recent researches on the control properties of multi-body systems (e.g. [1, 2, 3,
 54 17, 36]), we will study another types of feedback, namely acceleration, beside the
 55 classical displacement/velocity feedbacks. After that, a comparable framework for
 56 controllability of discrete-time systems is set up by using the algebraic approach.
 57 Finally, based on controllability, we derive some characterization for observability of
 58 second order discrete-time descriptor systems.

59 It should be noted, that all results in this paper also carry over to descriptor
 60 systems with time-variable, complex-valued coefficients or higher order descriptor
 61 systems. However, for notational convenience, and because that this is the most
 62 important case in practice, we restrict ourself to time-invariant, real-valued systems
 63 of second order.

64 The outline of this paper is as follows. After recalling some preliminary concepts
 65 and some auxiliary lemmas, in Section 3 we present the the condensed forms (3.4),
 66 (3.11) for (1.1). Based on these, we discuss the causal controllability of (1.1) via differ-
 67 ent types of feedbacks and their characterization. Here we also discuss the advantage
 68 of an acceleration feedback to the causal controllability of the system, while the other
 69 feedbacks fail. In Section 4, making use of (3.4), we analyze other controllability con-
 70 cepts for system (1.1). There, we also highlight a new feature of second order systems
 71 compare to first order ones, as well as the difference between continuous-time and
 72 discrete-time systems. In Section 5, observability for (1.1) is investigated. Finally, we
 73 finish with some conclusion.

74 **2. Preliminaries and auxiliary lemmas.** First let us briefly recall some im-
 75 portant concepts for a first order descriptor system

$$E\xi(n+1) - A\xi(n) = B_1 u(n) \quad \text{for all } n \geq n_0, \quad (2.1) \quad \{\text{SiDE 1st ord}\}$$

76 where $E, A \in \mathbb{R}^{\tilde{d}, \tilde{d}}$, $B_1 \in \mathbb{R}^{\tilde{d}, p}$ for some $\tilde{d} \in \mathbb{N}$. Here we notice that the matrix E
 77 may be rank deficient, and the matrix pair (E, A) is regular, i.e., $\det(\lambda E - A) \neq 0$
 78 in the polynomial sense. It is well-known, that the regularity of the pair (E, A) is
 79 the necessary and sufficient condition for the existence and uniqueness of a solution
 80 to (2.1), see, e.g. [11]. Moreover, the regular pair (E, A) can be transformed to
 81 Kronecker-Weierstraß canonical form (see, e.g. [29]), i.e., there exist nonsingular
 82 matrices U, V such that

$$UEV = \begin{bmatrix} I_{\tilde{d}_1} & 0 \\ 0 & N \end{bmatrix}, \quad UAV = \begin{bmatrix} J & 0 \\ 0 & I_{\tilde{d}_2} \end{bmatrix}, \quad \begin{bmatrix} B_{11} \\ B_{12} \end{bmatrix} = UB_1, \quad (2.2) \quad \{\text{Kronecker}\}$$

83 where N is a nilpotent matrix of nilpotency index ν , i.e., $N^\nu = 0$ and $N^i \neq 0$ for
 84 $i = 1, 2, \dots, \nu - 1$. The index ν is called the index of the pair (E, A) which doesn't
 85 depend on U, V and we write $\text{ind}(E, A) = \nu$. Consequently, the explicit solution of
 86 (2.1) is of the form $\xi(n) = V \begin{bmatrix} \xi_1(n) \\ \xi_2(n) \end{bmatrix}$ with

$$\begin{aligned} \xi_1(n+1) &= J^{n-n_0+1} x(n_0) + \sum_{i=0}^{n-n_0} J^i B_{11} u(n-i), \\ \xi_2(n) &= - \sum_{i=0}^{\nu-1} N^i B_{12} u(n+i) \end{aligned} \quad (2.3) \quad \{\text{solution}\}$$

87 for all $n \geq n_0$.

88 Clearly, the initial condition $\xi(n_0)$ could not be arbitrarily taken. System (2.1) is
 89 called *causal* if the state $\xi(n)$ is determined completely by the initial condition $\xi(n_0)$
 90 and former inputs $u(i)$ with $i = n_0, n_0 + 1, \dots, n$. It is easy to see that if $\text{ind}(E, A) = 1$
 91 then system (2.1) is causal. For a given input sequence $u = \{u(n)\}_{n \geq n_0}$, the set of
 92 consistent initial condition is given by

$$S_0 = \left\{ V \begin{bmatrix} \xi_1(n) \\ \xi_2(n) \end{bmatrix} \mid \xi_1(n_0) \in \mathbb{R}^{\tilde{d}_1}, \xi_2(n_0) = - \sum_{i=0}^{\nu-1} N^i B_{12} u(n+i) \right\}.$$

93 The set \mathcal{R} of *reachable states* or *reachable set* of (2.1) is the set of all vector that
 94 can be reached from some consistent initial vector $\xi(n_0)$ and some input sequence
 95 $\{u(n)\}_{n \geq n_0}$. In fact, for (2.1), it is well-known (e.g. [33]) that

$$\mathcal{R} = \mathbb{R}^{\tilde{d}_1} \oplus \text{Im}\mathcal{K}(N, B_{12}),$$

96 where $\mathcal{K}(N, B_{12}) := [B_{12}, NB_{12}, \dots, N^{\nu-1}B_{12}]$. In particular, if $N = 0$, the fol-
 97 lowing corollary is directly followed.

COROLLARY 2.1. *Assume that the first order, discrete-time descriptor system of the form*

$$\begin{bmatrix} \mathbf{E}_1 \\ 0 \end{bmatrix} \xi(n+1) - \begin{bmatrix} \mathbf{A}_1 \\ \mathbf{A}_2 \end{bmatrix} \xi(n) = \begin{bmatrix} \mathbf{B}_1 \\ \mathbf{B}_2 \end{bmatrix} u(n) \quad \text{for all } n \geq 0,$$

98 where $\begin{bmatrix} \mathbf{E}_1 \\ \mathbf{A}_2 \end{bmatrix}$ is nonsingular, and \mathbf{B}_2 has full row rank. Then the reachable subspace \mathcal{R}
 99 is the whole space \mathbb{R}^d .

100 DEFINITION 2.2. *The first order descriptor system (2.1) is called*

- 101 *i)* completely controllable or C-controllable if for any $x_0 \in \mathbb{R}^n$ and any $x_0^f \in \mathbb{R}^n$
 102 there exist a finite time n_f and an input sequence u such that $x(n_f) = x_0^f$.
- 103 *ii)* controllable on a reachable set or R-controllable if for any $x_0 \in \mathbb{R}^n$ and
 104 any $x_0^f \in \mathbb{R}^n$ there exist a finite time n_f and an input sequence u such that
 105 $x(n_f) = x_0^f$.
- 106 *iii)* causal controllable or Y-controllable if if there exists a feedback $u(k) = Fx(k)$
 107 such that its closed-loop system $Ex(k+1) = (A + B_1F)x(k)$ is causal.
- 108 *iv)* normalizable if there exists a feedback $u(k) = Fx(k+1)$ such that its closed-
 109 loop system $(E + B_1F)x(k+1) = Ax(k)$ is an explicit difference equation,
 110 i.e., $E + B_1F$ is nonsingular.

111 For most classical control design aim, typically, one or more of the following rank
 112 conditions are required

$$\begin{aligned} \mathbf{C0} : & \text{rank} [\alpha E - \beta A, B_1] = \tilde{d} \text{ for all } (\alpha, \beta) \in \mathbb{C}^2 \setminus \{(0, 0)\}, \\ \mathbf{C1} : & \text{rank} [\lambda E - A, B_1] = \tilde{d} \text{ for all } \lambda \in \mathbb{C}, \\ \mathbf{C2} : & \text{rank} [E, AS_\infty(E), B_1] = \tilde{d}, \\ \mathbf{C3} : & \text{rank} [E, B_1] = \tilde{d}, \end{aligned} \quad (2.4) \quad \{\text{rank 1st ord}\}$$

113 where $S_\infty(E)$ is a matrix whose columns span an orthogonal basis of $\ker(E)$. Furthermore,
 114 it should be noted that $\mathbf{C0} = \mathbf{C1} + \mathbf{C3}$. From characterizations of controllability
 115 in [5, 11, 19] and by Kronecker-Weierstraß canonical form we can deduce

116 PROPOSITION 2.3. Consider the first order descriptor system (2.1), whose the
 117 matrix pair (E, A) is regular. Then (2.1) is

- i) C-controllable if and only if $\mathbf{C0}$ holds.
- ii) R-controllable if and only if $\mathbf{C1}$ holds.
- iii) Y-controllable if and only if $\mathbf{C2}$ holds.
- iv) normalizable if and only if $\mathbf{C3}$ holds.

122 For the physical meanings of these controllability concepts and their properties,
 123 we refer the interested readers to classical textbooks [7, 16, 32, 37].

124 DEFINITION 2.4. i) System (1.1) is called regular if there exists an input sequence
 125 $u = \{u(n)\}_{n \geq n_0}$ such that the corresponding IVP (1.1) is uniquely solvable. In this
 126 situation, we also say that the input u and the initial vectors x_0, x_1 are consistent.
 127 ii) In addition, a regular system (1.1) is called causal if for each $n \geq n_0$, $x(n)$ does
 128 not depend on an input u at future time, i.e., $u(n+1), u(n+2), \dots$ but only at present
 129 and past time, i.e., $u(n), u(n-1), \dots, u(n_0)$.

130 DEFINITION 2.5. ([23]) System (1.2) is called strangeness-free if there exists a
 131 nonsingular matrix $P \in \mathbb{R}^{n,n}$ such that by scaling (1.2) with P , we obtain a new
 132 system of the form

$$\begin{aligned} \hat{r}_2 & \left[\begin{array}{c} \hat{M}_1 \\ 0 \\ 0 \\ \hat{v} \end{array} \right] x(n+2) + \left[\begin{array}{c} \hat{D}_1 \\ \hat{D}_2 \\ 0 \\ 0 \end{array} \right] x(n+1) + \left[\begin{array}{c} \hat{K}_1 \\ \hat{K}_2 \\ \hat{K}_3 \\ 0 \end{array} \right] x(n) = \left[\begin{array}{c} \hat{f}_{n,1} \\ \hat{f}_{n,2} \\ \hat{f}_{n,3} \\ 0 \end{array} \right] \text{ for all } n \geq n_0, \end{aligned} \quad (2.5) \quad \{\text{SiDE 2nd order sfree}\}$$

133 where the matrix $[\hat{M}_1^T \quad \hat{D}_2^T \quad \hat{K}_3^T]^T$ has full row rank. Notice that, restricted to the
 134 case that $M = 0$, we obtain exactly the well-known concept strangeness-free for the
 135 first order DAEs in [21].

136 To study control properties of second order descriptor systems, the classical ap-
 137 proach is to reformulate (1.1) in the form of (2.1). In the following example we
 138 demonstrate some critical difficulties that may arise while performing this approach
 139 for SiDEs.

140 EXAMPLE 2.6. Consider (1.1), where the matrix coefficients are

$$M = \begin{bmatrix} 1 & 0 \\ 0 & 0 \end{bmatrix}, \quad D = \begin{bmatrix} 1 & 0 \\ 0 & 0 \end{bmatrix}, \quad K = \begin{bmatrix} 1 & 1 \\ 1 & 0 \end{bmatrix}, \quad B = \begin{bmatrix} 1 \\ 0 \end{bmatrix}. \quad (2.6) \quad \{\text{eq1.4}\}$$

141 In fact, we have at least four ways to reformulate (1.1) as follows

$$\begin{aligned}
 \text{companion form : } & \begin{bmatrix} I & 0 \\ 0 & M \end{bmatrix} \begin{bmatrix} x(n+1) \\ x(n+2) \end{bmatrix} = \begin{bmatrix} 0 & I \\ -K & -D \end{bmatrix} \begin{bmatrix} x(n) \\ x(n+1) \end{bmatrix} + \begin{bmatrix} 0 \\ B \end{bmatrix} u(n), \\
 \text{2nd form: } & \begin{bmatrix} D & M \\ M & 0 \end{bmatrix} \begin{bmatrix} x(n+1) \\ x(n+2) \end{bmatrix} = \begin{bmatrix} -K & 0 \\ 0 & M \end{bmatrix} \begin{bmatrix} x(n) \\ x(n+1) \end{bmatrix} + \begin{bmatrix} B \\ 0 \end{bmatrix} u(n), \\
 \text{3rd form: } & \begin{bmatrix} D & M \\ -M & 0 \end{bmatrix} \begin{bmatrix} x(n+1) \\ x(n+2) \end{bmatrix} = \begin{bmatrix} -K & 0 \\ 0 & -M \end{bmatrix} \begin{bmatrix} x(n) \\ x(n+1) \end{bmatrix} + \begin{bmatrix} B \\ 0 \end{bmatrix} u(n), \\
 \text{4th form : } & \begin{bmatrix} -K & 0 \\ 0 & M \end{bmatrix} \begin{bmatrix} x(n+1) \\ x(n+2) \end{bmatrix} = \begin{bmatrix} 0 & -K \\ -K & -D \end{bmatrix} \begin{bmatrix} x(n) \\ x(n+1) \end{bmatrix} + \begin{bmatrix} 0 \\ B \end{bmatrix} u(n).
 \end{aligned} \tag{2.7} \quad \{\text{first order companion form}\}$$

142 Each form above has its advantage, especially in case that M, K, D has a symmetric or skew-symmetric structure. Now let us check the controllability of these systems by verifying the rank conditions (2.4). Direct computations turns out that only in the fourth form, the index of the matrix pair (E, A) is three, while in the others, the index is four, which suggests a wrong prediction, that $x(n)$ depends also on $u(n+3)$, instead of only $u(n), u(n+1), u(n+2)$.

148 In control theory, classical design approaches usually require that the system is at least Y-controllable. Nevertheless, this is not always fulfilled as shown in Example 149 2.7 below.

151 EXAMPLE 2.7. Consider the artificial descriptor system (1.1) with

$$M = 0, \quad D = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix}, \quad K = \begin{bmatrix} 0 & 0 \\ 1 & 0 \end{bmatrix}, \quad B = \begin{bmatrix} 1 \\ 0 \end{bmatrix}.$$

152 This is in fact a first order system, since $M = 0$. We can directly check that this 153 system is Y-controllable. Nevertheless, all the first order formulations in (2.7) are 154 not. Furthermore, for another input matrix $B = \begin{bmatrix} 0 \\ 1 \end{bmatrix}$ direct computations yield that 155 (1.1) is C-controllable, while all the formulations in (2.7) are not.

156 In view of all these difficulties, it is natural to seek for a suitable first order 157 reformulation that is Y-controllable and be beneficial to study other controllability 158 properties of (1.1). This task will be done in the next section. Two auxiliaries lemmata 159 below will be very useful for our analysis later.

160 LEMMA 2.8. ([24, Lemma 4.1]) Given four matrices $\check{A}, \check{B}, \check{C}$ in $\mathbb{R}^{m,d}$ and \check{D} in 161 $\mathbb{R}^{m,p}$. Then there exists an orthogonal matrix $\check{U} \in \mathbb{R}^{m,m}$ such that

$$\check{U} \begin{bmatrix} \check{A} & \check{B} & \check{C} & | & \check{D} \end{bmatrix} = \left[\begin{array}{ccc|c} \check{A}_1 & \check{B}_1 & \check{C}_1 & \check{D}_1 \\ 0 & \check{B}_2 & \check{C}_2 & 0 \\ 0 & 0 & \check{C}_3 & 0 \\ 0 & 0 & 0 & 0 \\ \hline 0 & \check{B}_4 & \check{C}_4 & \check{D}_4 \\ 0 & 0 & \check{C}_5 & \check{D}_5 \end{array} \right], \tag{2.8} \quad \{\text{eq1.6}\}$$

162 where the matrices $\check{A}_1, \check{B}_2, \check{B}_4, \check{C}_3, \begin{bmatrix} \check{D}_4 \\ \check{D}_5 \end{bmatrix}$ have full row rank.

163 LEMMA 2.9. Let $P = \begin{bmatrix} P_1 \\ P_2 \end{bmatrix} \in \mathbb{R}^{p,d}$, $Q = \begin{bmatrix} 0 \\ Q_2 \end{bmatrix} \in \mathbb{R}^{q,d}$ be two matrices. 164 Furthermore, assume that Q_2 has full row rank. Then there exist a matrix $F \in \mathbb{R}^{d,d}$ such that 165 $P + QF$ has full row rank if and only if P_1 also has full row rank.

166 *Proof.* The necessary part is followed directly from the observation that

$$P + QF = \begin{bmatrix} P_1 \\ P_2 \end{bmatrix} + \begin{bmatrix} 0 \\ Q_2 \end{bmatrix} F = \begin{bmatrix} P_1 \\ P_2 + Q_2 F \end{bmatrix} = \begin{bmatrix} P_1 & 0 \\ P_2 & Q_2 \end{bmatrix} \begin{bmatrix} I \\ F \end{bmatrix}.$$

167 For the sufficient part, see [24, Lemma 2.8]. \square

168 **3. Condensed forms and causal controllability.** In this section, we will
169 modify an *algebraic method* presented in [25] to study the causal controllability (Y-
170 controllability) of system (1.1). The main idea is to transform (1.1) directly, but
171 not reformulate it as a first order one, into so-called *condensed forms*. Moreover,
172 in comparison to [25], the main advantage of our method is two folds. First, the
173 condensed form is much more concise, and can be computed in a stable way. Second,
174 it is helpful to design a suitable feedback that make the closed-loop system to be
175 causal (resp., impulse-free) in the discrete (resp., continuous) time case. Now let us
176 introduce some rank conditions, which generalize the ones in (2.4).

- C21 :** $\text{rank} [\lambda^2 M + \lambda D + K, B] = d$ for all $\lambda \in \mathbb{C}$,
- C22 :** $\text{rank} [M, DS_\infty^1, KS_\infty^2, B] = d$,
- C23 :** $\text{rank} [M, D, B] = d$,
- C24 :** $\text{rank} [M, B] = d$,

177 where columns of S_∞^1 form a basis of kernel M , and columns of S_∞^2 form the basis of

$$\text{kernel} \left[\begin{array}{c} M \\ Z_1^T D \end{array} \right] \setminus \text{kernel} \left[\begin{array}{c} M \\ Z_1^T D \\ Z_3^T K \end{array} \right],$$

178 and columns of Z_1 and of Z_3 span the left null spaces of M and $[M \ D]$, respectively.

179 **DEFINITION 3.1.** Two second order descriptor systems of the form (1.1) with
180 system matrices (M, D, K, B) , and $(\tilde{M}, \tilde{D}, \tilde{K}, \tilde{B})$ are called strongly (left) equivalent
181 if there exist nonsingular matrices $U \in \mathbb{R}^{d,d}$ and $V \in \mathbb{R}^{m,m}$ such that

$$\tilde{M} = UM, \quad \tilde{D} = UD, \quad \tilde{K} = UK, \quad \tilde{B} = UBV,$$

182 We write $(M, D, K, B) \xrightarrow{\ell} (\tilde{M}, \tilde{D}, \tilde{K}, \tilde{B})$.

183 It should be noted that, in contrast to [25, 28, 35], we avoid to perform variable
184 transformations, i.e. $x(n) = W(n)y(n)$ for some nonsingular matrix $W(n)$. This ap-
185 proach will make our analysis more concise and clearer. More importantly, we aim at
186 stably computable condensed forms, which is not available by the approach presented
187 in the references above. Recently, using condensed forms under strongly left equiva-
188 lence transformation, solvability analysis for second order discrete-time systems has
189 been discussed in [24]. Furthermore, we also incorporate another class of equivalent
190 transformations as follows.

191 **DEFINITION 3.2.** Two systems $Mx(n+2) + Dx(n+1) + Kx(n) = Bu(n)$ and
192 $\tilde{M}x(n+2) + \tilde{D}x(n+1) + \tilde{K}x(n) = \tilde{B}u(n)$ are called equivalent under

- 193 i) displacement/position feedback if there exists a matrix $F_d \in \mathbb{R}^{m,d}$ such that

$$(M, D, K, B) \xrightarrow{\ell} (\tilde{M}, \tilde{D}, \tilde{K} + F_d \tilde{B}, \tilde{B}).$$
- 194 ii) velocity feedback if there exists a matrix $F_v \in \mathbb{R}^{m,d}$ such that

$$(M, D, K, B) \xrightarrow{\ell} (\tilde{M}, \tilde{D} + F_v \tilde{B}, \tilde{K}, \tilde{B}).$$

197 *iii) acceleration feedback if there exists a matrix $F_a \in \mathbb{R}^{m,d}$ such that*
 198 $(M, D, K, B) \xrightarrow{\ell} (\tilde{M} + F_a \tilde{B}, \tilde{D}, \tilde{K}, \tilde{B})$.

199 Here F_d, F_v, F_a are called displacement, velocity, acceleration gain matrices.

200 We notice that this concept is equivalent to classical feedback concepts as in
 201 mechanics for continuous-time descriptor systems [26, 27]. Furthermore, in general, a
 202 chosen feedback may contain all acceleration part $F_a x(n+2)$, velocity part $F_v x(n+1)$
 203 and displacement/position part $F_d x(n)$, i.e.,

$$u(n) = -F_a x(n+2) - F_v x(n+1) - F_d x(n). \quad (3.2) \quad \{\text{feedback}\}$$

204 Consequently, the resulting closed-loop system is

$$(M + BF_a)x(n+2) + (D + BF_v)x(n+1) + (K + BF_d)x(n) = 0. \quad (3.3) \quad \{\text{close-loop}\}$$

205 Now let us recall the concept of Y-controllability for system (1.1).

206 DEFINITION 3.3. *The descriptor system (1.1) is called Y-controllable via displace-
 207 ment-velocity-acceleration feedback if there exists a feedback of the form (3.2) such
 208 that the closed-loop system (3.3) is regular and strangeness-free.*

209 LEMMA 3.4. *The Y-controllability is invariant under left equivalent transforma-
 210 tions.*

211 *Proof.* Due to Definition 3.1, by choosing

$$u(n) = -V^{-1}F_a x(n+2) - V^{-1}F_v x(n+1) - V^{-1}F_d x(n)$$

212 the proof is straightforward. \square

213 In the following theorem, we present the first condensed form of system (1.1).

214 THEOREM 3.5. *Consider the descriptor system (1.1). Then there exist two or-
 215 thogonal matrices U, V such that the following identities hold.*

$$U [M, D, K] = \begin{bmatrix} M_1 & D_1 & K_1 \\ 0 & D_2 & K_2 \\ 0 & 0 & K_3 \\ \hline 0 & D_4 & K_4 \\ 0 & 0 & K_5 \\ 0 & 0 & 0 \end{bmatrix}, \quad UBV = \begin{bmatrix} B_{11} & B_{12} & B_{13} \\ 0 & 0 & B_{23} \\ 0 & 0 & 0 \\ \hline 0 & \Sigma_1 & B_{43} \\ 0 & 0 & \Sigma_0 \\ 0 & 0 & 0 \end{bmatrix}, \quad \begin{matrix} r_2 \\ r_1 \\ r_0 \\ \varphi_1 \\ \varphi_0 \\ v \end{matrix} \quad (3.4) \quad \{\text{condensed form 1}\}$$

216 where sizes of the block rows are $r_2, r_1, r_0, \varphi_1, \varphi_0, v$, the matrices $M_1, [D_2]$, K_3 are
 217 of full row rank, and the matrices Σ_1, Σ_0 are nonsingular and diagonal.

218 *Proof.* The proof is followed directly from Lemma 2.8 by consecutively partitioning
 219 two matrices \tilde{D}_5 and \tilde{D}_4 in (2.8) via Singular Value Decompositions. \square

220 Theorem 3.5 has one direct corollary below.

221 COROLLARY 3.6. *In the condensed form (3.4), the condition $r_0 = v_0 = 0$ holds
 222 true if and only if condition C23 holds true, i.e. the matrix $[M, D, B]$ has full row
 223 rank d.*

224 REMARK 3.7. *The orthogonality of U and V guarantees that the condensed form
 225 (3.4) can be numerically stably computed. This is an important advantage, in compar-
 226 ison to the condensed form in Theorem 2.4, [25]. Furthermore, we refer the interested
 227 reader to Remark 2.7 in the same article.*

228 **3.1. Causal controllability via displacement and velocity feedbacks.**

229 Now we are ready to present our first main result about the Y-controllability of (1.1)
230 in Theorem 3.8 below. We emphasize, that due to different roles of feedback types,
231 the characteristic condition for Y-controllability via displacement feedback is more
232 strict than the corresponding one for velocity feedback.

233 THEOREM 3.8. *Consider the second order descriptor system (1.1) and the con-
234 densed form (3.4). Then we have that:*

- 235 i) *System (1.1) is Y-controllable via displacement-velocity feedback if and only if $v = 0$
236 and the matrix $[M_1^T \ D_2^T \ K_3^T]^T$ has full row rank.*
- 237 ii) *System (1.1) is Y-controllable via displacement feedback if and only if $v = 0$ and
238 the matrix $[M_1^T \ D_2^T \ K_3^T \ D_4^T]^T$ has full row rank.*
- 239 iii) *System (1.1) is Y-controllable via velocity feedback if and only if $v = 0$ and the
240 matrix $[M_1^T \ D_2^T \ K_3^T]^T$ has full row rank.*

241 *Proof.* Since the proofs of these three claims are essentially the same, for the sake
242 of brevity we will present only the detailed arguments for part i).

243 **Necessity:** Due to (3.4) we see that

$$[M \ D \ K \mid B] \xrightarrow{\ell} \left[\begin{array}{ccc|ccc} M_1 & D_1 & K_1 & B_{11} & B_{12} & B_{13} \\ 0 & D_2 & K_2 & 0 & 0 & B_{23} \\ 0 & 0 & K_3 & 0 & 0 & 0 \\ \hline 0 & D_4 & K_4 & 0 & \Sigma_1 & B_{43} \\ 0 & 0 & K_5 & 0 & 0 & \Sigma_0 \\ 0 & 0 & 0 & 0 & 0 & 0 \end{array} \right] \begin{matrix} r_2 \\ r_1 \\ r_0 \\ \varphi_1 \\ \varphi_0 \\ v \end{matrix}.$$

244 Thus, by using Gaussian elimination, we obtain

$$[M \ D \ K \mid B] \xrightarrow{\ell} \left[\begin{array}{ccc|ccc} M_1 & D_1^{new} & K_1^{new} & B_{11} & 0 & 0 \\ 0 & D_2 & K_2^{new} & 0 & 0 & 0 \\ 0 & 0 & K_3 & 0 & 0 & 0 \\ \hline 0 & D_4 & K_4^{new} & 0 & \Sigma_1 & 0 \\ 0 & 0 & K_5 & 0 & 0 & \Sigma_0 \\ 0 & 0 & 0 & 0 & 0 & 0 \end{array} \right], \quad (3.5) \quad \text{f eq3.1}$$

245 where by the super script *new* we indicate a (possibly) new matrix at the same block
246 position. This form implies that no matter what feedback has been applied, it will
247 not affect the strangeness property of the upper part of the corresponding system,
248 and hence, system (1.1) is Y-controllable only if the matrix $[M_1^T \ D_2^T \ K_3^T]^T$ has
249 full row rank. Finally, notice that system (1.1) is of square size, so it is regular only
250 if $v = 0$. This completes the necessity part.

251 **Sufficiency:** By applying Lemma 2.9 for the matrices $P = [M_1^T \ D_2^T \ K_3^T]^T$, $Q =$
252 $\begin{bmatrix} 0 & \Sigma_1 & 0 \\ 0 & 0 & \Sigma_0 \end{bmatrix}$ and $G = [D_4^T \ K_5^T]^T$, we see that there exist two matrices F_d , F_v such
253 that the matrix

$$\begin{bmatrix} M_1 \\ D_2 \\ K_3 \\ D_4 + [0 \ \Sigma_1 \ B_{43}] F_v \\ K_5 + [0 \ 0 \ \Sigma_0] F_d \end{bmatrix}$$

254 has full row rank. Consequently, for the displacement-velocity feedback

$$u(n) = -F_v x(n+1)(t) - F_d x(n) \text{ for all } n \geq n_0, \quad (3.6) \quad \{\text{eq5.5}\}$$

255 the closed loop system

$$Mx(n+2) + (D + BF_v)x(n+1) + (K + BF_d)x(n) = 0 \quad (3.7) \quad \{\text{eq5.6}\}$$

256 is strangeness-free. Furthermore, due to the fact that in (3.4) $v = 0$, the closed-loop
257 system (3.7) is regular, and hence, this finishes the proof. \square

258 Making use of (3.4), we can rewrite our system (1.1) as follows

$$\begin{array}{c|cc} \left[\begin{array}{ccc} M_1 & D_1 & K_1 \\ 0 & D_2 & K_2 \\ 0 & 0 & K_3 \end{array} \right] & \left[\begin{array}{c} x(n+2) \\ x(n+1) \\ x(n) \end{array} \right] = & \left[\begin{array}{ccc} B_{11} & B_{12} & B_{13} \\ 0 & 0 & B_{23} \\ 0 & 0 & 0 \end{array} \right] v(n), & \begin{array}{l} r_2 \\ r_1 \\ r_0 \end{array} \\ \hline \left[\begin{array}{ccc} 0 & D_4 & K_4 \\ 0 & 0 & K_5 \\ 0 & 0 & 0 \end{array} \right] & & \left[\begin{array}{ccc} 0 & \Sigma_1 & B_{43} \\ 0 & 0 & \Sigma_0 \\ 0 & 0 & 0 \end{array} \right] & \begin{array}{l} \varphi_1 \\ \varphi_0 \\ v \end{array} \end{array} \quad (3.8) \quad \{\text{system condensed form 1}\}$$

259 where $u(n) = Vv(n)$ for all $n \geq n_0$. Let $z(n) := M_1 x(n+1)$ we can then introduce a
260 new variable $\xi(n) = \begin{bmatrix} z(n) \\ x(n) \end{bmatrix} \in \mathbb{R}^{r_2+d}$ and rewrite system (3.8) in the so-called *minimal
261 extension form*

$$\underbrace{\begin{bmatrix} I_{r_2} & D_1 \\ 0 & M_1 \\ 0 & D_2 \\ 0 & 0 \end{bmatrix}}_{\tilde{E}} \xi(n+1) + \underbrace{\begin{bmatrix} 0 & K_1 \\ -I_{r_2} & 0 \\ 0 & K_2 \\ 0 & K_3 \end{bmatrix}}_{-\tilde{A}} \xi(n) = \underbrace{\begin{bmatrix} B_{11} & B_{12} & B_{13} \\ 0 & 0 & 0 \\ 0 & 0 & B_{23} \\ 0 & 0 & 0 \end{bmatrix}}_{\tilde{B}} v(n). \quad (3.9) \quad \{\text{descriptor minimal extension}\}$$

262
263 THEOREM 3.9. Consider the descriptor system (1.1) and the condensed form
264 (3.4). Furthermore, assume that $v = 0$ and the matrix $[M_1^T \ D_2^T \ K_3^T]^T$ has full
265 row rank. Then the minimal extension form (3.9) is also Y-controllable.

266 Proof. In order to prove the desired claim we will verify the rank condition (2.4).
267 Let $S_\infty(\tilde{E})$ be a full column rank matrix whose columns form an orthogonal basis of
268 the vector space $\ker(\tilde{E})$. Partition $S_\infty(\tilde{E}) = \begin{bmatrix} U_1 \\ V_1 \end{bmatrix} \in \mathbb{R}^{r_2+d, r_2+d}$ correspondingly to
269 (3.9), we see that

$$D_2 V_1 = 0, \quad M_1 V_1 = 0.$$

270 Now we will prove that $K_3 V_1$ has full row rank. To do it first we perform an SVD
271 for the matrix $\begin{bmatrix} M_1 \\ D_2 \end{bmatrix}$, and due to the fact that the matrix $\begin{bmatrix} M_1 \\ D_2 \end{bmatrix}$ has full row rank, it
272 follows that

$$U_2^T \begin{bmatrix} M_1 \\ D_2 \end{bmatrix} V_2 = [\Sigma \ 0],$$

273 where Σ is a nonsingular, diagonal matrix. Hence, $V_1 = V_2 \begin{bmatrix} 0 \\ I \end{bmatrix}$. Partitioning
274 $U_2^T K_3 V_2$ correspondingly, we have $U_2^T K_3 V_2 = [K_{31} \ K_{32}]$. Notice that since the
275 matrix $[M_1^T \ D_2^T \ K_3^T]^T$ has full row rank, K_{32} has full row rank. Thus,

$$K_3 V_1 = U_2 [K_{31} \ K_{32}] V_2^T V_2 \begin{bmatrix} 0 \\ I \end{bmatrix} = U_2 K_{32},$$

which has full row rank. Therefore, we see that

$$\left[\tilde{E} \quad \tilde{A}S_{\infty}(\tilde{E}) \quad \tilde{B} \right] = \left[\begin{array}{cc|c|ccc} I & D_1 & K_1 V_1 & B_{11} & B_{12} & B_{13} \\ 0 & M_1 & U_1 & 0 & 0 & 0 \\ 0 & D_2 & K_2 V_1 & 0 & 0 & B_{23} \\ 0 & 0 & K_3 V_1 & 0 & 0 & 0 \\ \hline 0 & D_4 & K_5 V_1 & 0 & \Sigma_1 & B_{43} \\ 0 & 0 & 0 & 0 & 0 & \Sigma_0 \\ 0 & 0 & 0 & 0 & 0 & 0 \end{array} \right] \quad \begin{matrix} r_2 \\ r_2 \\ r_1 \\ r_0 \\ \varphi_1 \\ \varphi_0 \\ v \end{matrix}$$

has full row rank if and only if $v = 0$. This completes the proof. \square

REMARK 3.10. From Theorems 3.8, 3.9 above, we see that one can interpret the upper part of system (3.8) as a causal uncontrollable part, while the lower part is the causal controllable part. Furthermore, the key point for constructing a suitable first order reformulation to (1.1) (and also for feedback design strategies) is to bring system (1.1) to the form (3.4), where the upper part must be strangeness-free, i.e., $[M_1^T \quad D_2^T \quad K_3^T]^T$ has full row rank. In other words, the index reduction procedure has been performed only for the causal uncontrollable part. Recently, this task has been finished in both theoretical and numerical ways. To keep the brevity of this paper, we will omit the details and refer the interested readers to [24, Section 4]. Below we recall one important result taken from this research.

PROPOSITION 3.11. ([24, Theorem 4.7]) Consider the descriptor system (1.1). Then it has exactly the same solution set as the so-called strangeness-free descriptor system

$$\underbrace{\begin{bmatrix} \hat{M}_1 \\ 0 \\ 0 \\ \hline 0 \\ 0 \\ 0 \end{bmatrix}}_{\hat{M}} x(n+2) + \underbrace{\begin{bmatrix} \hat{D}_1 \\ \hat{D}_2 \\ 0 \\ \hline \hat{D}_4 \\ 0 \\ 0 \end{bmatrix}}_{\hat{D}} x(n+1) + \underbrace{\begin{bmatrix} \hat{K}_1 \\ \hat{K}_2 \\ \hat{K}_3 \\ \hline \hat{K}_4 \\ \hat{K}_5 \\ 0 \end{bmatrix}}_{\hat{K}} x(n) = \underbrace{\begin{bmatrix} \hat{B}_{11} & \hat{B}_{12} & \hat{B}_{13} \\ 0 & 0 & \hat{B}_{23} \\ 0 & 0 & 0 \\ \hline 0 & \hat{\Sigma}_1 & \hat{B}_{43} \\ 0 & 0 & \hat{\Sigma}_0 \\ 0 & 0 & 0 \end{bmatrix}}_{\hat{B}} v(n), \quad \begin{matrix} \hat{r}_2 \\ \hat{r}_1 \\ \hat{r}_0 \\ \hat{\varphi}_1 \\ \hat{\varphi}_0 \\ \hat{v} \end{matrix} \quad \text{(3.10) } \{ \text{descriptor 2nd order sfree} \}$$

for all $t \geq t_0$, where $[\hat{M}_1^T \quad \hat{D}_2^T \quad \hat{K}_3^T]^T$ has full row rank, $\hat{\Sigma}_1$ and $\hat{\Sigma}_0$ are nonsingular and diagonal, and $u(n) = Vv(n)$ for all $n \geq n_0$, where V is nonsingular. Furthermore, if system (1.1) is regular then $\hat{v} = 0$.

Therefore, making use of Theorems 3.8, 3.9 and Proposition 3.11, we can completely analyze the Y-controllability and feedback design of (1.1). We, furthermore, can deduce from these theorems other conditions that help us directly verify the Y-controllability of (1.1) (without any feedback design strategy) as below.

COROLLARY 3.12. Consider the second order descriptor system (1.1) and the condensed form (3.4). Then system (1.1) is Y-controllable via displacement-velocity feedback if and only if condition **C21** is satisfied.

REMARK 3.13. In comparison to the continuous-time case, we see that Corollary 3.12 is similar to Theorem 3.14 i) ([25]). Nevertheless, if one wants to use only one type of feedback (displacement or velocity), then it could lead to extra difficulties, since the condensed form (2.3) ([25]) could not be stably-computed. Therefore, we suggest the reader to use Theorem 3.8.

305 **3.2. Causal controllability via acceleration feedback.** For second order
 306 systems, one can consider different types of feedback (acceleration/velocity/displace-
 307 ment) separately, or mimic them together. In the pioneering work [25], Loose and
 308 Mehrmann considered three feedback types: position, velocity, and position-velocity;
 309 while recently Abdelaziz ([1]) considered displacement-accerleration feedback, and
 310 Zhu and Zhang ([36]) considered the most general form (3.2). In this section, we
 311 will not limit ourself to velocity/displacement feedback as in previous section, but
 312 study also the effectiveness of acceleration feedback. Clearly, to in-cooperate another
 313 feedback type, we need a new condensed form, instead of using (3.4). This is given in
 314 the following theorem.

315 **THEOREM 3.14.** *Consider the descriptor system (1.1). Then, there exist two
 316 orthogonal matrices U, V such that the following identities hold.*

$$U[M, D, K] = \begin{bmatrix} \tilde{M}_1 & \tilde{D}_1 & \tilde{K}_1 \\ 0 & \tilde{D}_2 & \tilde{K}_2 \\ 0 & 0 & \tilde{K}_3 \\ \hline \tilde{M}_4 & \tilde{D}_4 & \tilde{K}_4 \\ 0 & \tilde{D}_5 & \tilde{K}_5 \\ 0 & 0 & \tilde{K}_6 \\ \hline 0 & 0 & 0 \end{bmatrix}, \quad UBV = \begin{bmatrix} 0 & 0 & \tilde{B}_{13} & \tilde{B}_{14} \\ 0 & 0 & 0 & \tilde{B}_{24} \\ 0 & 0 & 0 & 0 \\ \hline 0 & \tilde{\Sigma}_2 & \tilde{B}_{43} & \tilde{B}_{44} \\ 0 & 0 & \tilde{\Sigma}_1 & \tilde{B}_{54} \\ 0 & 0 & 0 & \tilde{\Sigma}_0 \\ \hline 0 & 0 & 0 & 0 \end{bmatrix}, \quad \begin{array}{l} r_2 \\ r_1 \\ r_0 \\ \hline \varphi_2 \\ \varphi_1 \\ \varphi_0 \\ \hline v \end{array} \quad (3.11) \quad \{\text{condensed form 2}\}$$

317 where sizes of the block rows are $r_2, r_1, r_0, \varphi_2, \varphi_1, \varphi_0, v$, the matrices $\begin{bmatrix} \tilde{M}_1 \\ \tilde{M}_4 \end{bmatrix}, \begin{bmatrix} \tilde{D}_2 \\ \tilde{D}_5 \end{bmatrix},$
 318 \tilde{K}_3 are of full row rank, and the matrices $\tilde{\Sigma}_2, \tilde{\Sigma}_1, \tilde{\Sigma}_0$ are nonsingular and diagonal.

319 *Proof.* The proof can be obtained directly by using Theorem 3.5. To keep the
 320 brevity of this paper we will omit the detail. \square

321 The following corollaries are direct consequences of Theorem 3.14 and Lemma
 322 2.8.

323 **COROLLARY 3.15.** *Consider the descriptor system (1.1) and the factorization
 324 (3.11). Then, the following assertions hold true.*

- 325 i) *System (1.1) is Y-controllable via only displacement feedback if and only if in (3.4),
 326 we have $v = 0$ and the matrix $\begin{bmatrix} \tilde{M}_1^T & \tilde{D}_2^T & \tilde{K}_3^T & \tilde{M}_4^T & \tilde{D}_5^T \end{bmatrix}^T$ is of full row rank.*
- 327 ii) *System (1.1) is Y-controllable via displacement-velocity feedback (or velocity feed-
 328 back) if and only if in (3.4), $v = 0$ and the matrix $\begin{bmatrix} \tilde{M}_1^T & \tilde{D}_2^T & \tilde{K}_3^T & \tilde{M}_4^T \end{bmatrix}^T$ is of full
 329 row rank.*

330 **COROLLARY 3.16.** *Consider the descriptor system (1.1) and the factorization
 331 (3.11). Then, for any kind of feedback that involves acceleration ($d\text{-}v\text{-}a, d\text{-}a, v\text{-}a, a$),
 332 system (1.1) is Y-controllable via that feedback type if and only if $v = 0$ and the matrix
 333 $\begin{bmatrix} \tilde{M}_1^T & \tilde{D}_2^T & \tilde{K}_3^T \end{bmatrix}^T$ is of full row rank.*

334 **EXAMPLE 3.17.** *To illustrate the effectiveness of an acceleration feedback, we
 335 consider the discrete-time version of a non-gyroscopic system (e.g. [18])*

$$\begin{bmatrix} 0 & 0 \\ 1 & 0 \end{bmatrix} x(n+2) + \begin{bmatrix} 1 & 0 \\ 0 & 0 \end{bmatrix} x(n) = \begin{bmatrix} 0 \\ 1 \end{bmatrix} u(n). \quad (3.12)$$

336 Here we have that $\tilde{M}_4 = \tilde{K}_3 = [1 \ 0]$, $\tilde{M}_1 = \tilde{D}_2 = \tilde{D}_4 = \tilde{D}_5 = \tilde{K}_6 = []$. Due to
 337 Corollary 3.16i) this system is Y-controllable by acceleration feedback. Furthermore, it

338 is not possible to eliminate the causal behavior by using only displacement and velocity
339 feedbacks, since all the rank conditions in Corollary 3.15 fail.

EXAMPLE 3.18. Similarly, using Corollaries 3.15, 3.16 we see that one could not use only displacement-velocity feedback to eliminate the causal behavior of the system

$$\begin{bmatrix} 0 & 0 & 0 \\ 1 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} x(n+2) + \begin{bmatrix} 0 & 0 & 0 \\ 0 & 0 & 0 \\ 1 & 0 & 0 \end{bmatrix} x(n+1) + \begin{bmatrix} 1 & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} x(n) = \begin{bmatrix} 0 & 0 \\ 1 & 0 \\ 0 & 1 \end{bmatrix} u(n).$$

340 We notice, that we can construct any of four feedback types (d-v-a, d-a, v-a, a) to
341 regularize this system.

342 REMARK 3.19. We also notice, that even though different feedback types can be
343 applied to achieve the causality of the closed-loop systems, two condensed forms (3.4)
344 and (3.11) are still useful to achieve a desired rank for the system, i.e., there is a
345 desired number of zero-, first- and second-order equations. For more details on this
346 issue, we refer the readers to [8, 9, 10].

347 **4. Other controllability concepts and their characterizations.** In this
348 section, using the condensed forms (3.4), (3.9) proposed above, we will discuss other
349 controllability concepts for second order systems. We will also point out the difference
350 between a discrete and continuous time cases and discuss a new feature of second order
351 system as well.

352 DEFINITION 4.1. Consider the descriptor system (1.1).

353 *i)* A set $\mathcal{R} \subseteq \mathbb{R}^n$ is called reachable from the pair (x_0, x_1) if for every $x_0^f \in \mathcal{R}$ there
354 exists an input sequence u that transfers the system in finite time from $x(n_0) = x_0$ to
355 x_f .

356 *ii)* A set $\mathcal{R}_2 \subseteq \mathbb{R}^n \times \mathbb{R}^n$ is called reachable from the pair (x_0, x_1) if for every $(x_0^f, x_1^f) \in \mathcal{R}_2$ there exists an input sequence u that transfers the system in finite time from
357 $x(n_0) = x_0$, $x(n_1) = x_1$ to x_0^f , x_1^f .

358 *iii)* The system is called C-controllable if for any pair $(x_0, x_1) \in \mathbb{R}^n \times \mathbb{R}^n$ and any
359 $x_0^f \in \mathbb{R}^n$ there exist a finite time n_f and an input sequence u such that $x(n_f) = x_0^f$.

360 *iv)* The system is called strongly C2-controllable if for any pair $(x_0, x_1) \in \mathbb{R}^n \times \mathbb{R}^n$ and any pair $(x_0^f, x_1^f) \in \mathbb{R}^n \times \mathbb{R}^n$ there exist a finite time n_f and an input sequence u such that $x(n_f) = x_0^f$, $x(n_f + 1) = x_1^f$.

361 *v)* The system is called R-controllable if any state $x_0^f \in \mathcal{R}$ can be reached from some pair (x_0, x_1) in finite time.

362 *vi)* The system is called R2-controllable if any pair $(x_0^f, x_1^f) \in \mathcal{R}_2$ can be reached from some pair (x_0, x_1) in finite time.

363 It is straightforward to see that all these controllability concepts are invariant under left equivalent transformation. In the following theorem, we give a characterization for the strongly C2- and R2-controllability.

364 THEOREM 4.2. Consider the descriptor system (1.1) and its first order companion form (2.7). Then the following assertions hold true.

365 *i)* System (1.1) is R2-controllable if and only if the system matrix coefficients satisfy condition **C21**.

366 *ii)* Besides that, system (1.1) is strongly C2-controllable if and only if the system matrix coefficients satisfy both conditions **C21** and **C24**.

367 Proof. Following directly from Definition 4.1, we see that system (1.1) is strongly
368 C2-controllable (resp., R2-controllable) if and only if its first order companion form

379 (2.7) is C-controllable (resp., R-controllable). Thus, the proof is directly followed by
 380 checking the rank criteria in Proposition 2.3. \square

381 Now let us come back to the strangeness-free form (3.10). Clearly, we see that
 382 it is reasonable to control $x(n)$ and only the part $M_1x(n+1)$ but not the whole
 383 $x(n+1)$. This fact motivates another concept below, which is more suitable for
 384 singular descriptor systems.

385 DEFINITION 4.3. Consider the descriptor system (1.1) and assume that it is al-
 386 ready in the strangeness-free form (3.10). Then system (1.1) is called C2-controllable
 387 if the minimal extension form (3.9) is C-controllable.

388 LEMMA 4.4. Consider the descriptor system (1.1) and its the strangeness-free
 389 from (3.10) and the minimal extension form (3.9). Then we have that:

- 390 i) System (3.9) is R-controllable if and only if system (3.10) satisfies condition **C21**.
- 391 ii) System (3.9) is C-controllable if and only if system (3.10) satisfies both conditions
C21 and **C23**.
- 393 iii) The constant rank condition **C21** is preserved under the strangeness-free formula-
 394 tion, which transform (1.1) to (3.10).

395 Proof. For notational convenience, within this proof, we will omit the superscript
 396 $\hat{\cdot}$ on all matrices in the strangeness-free form (3.10). Due to Definition 2.3, system
 397 (3.9) is R-controllable (resp. C-controllable) if and only if the matrix coefficients \tilde{E} ,
 398 \tilde{A} , \tilde{B} satisfy the constant rank **C1** (resp., **C0**).

399 i) Condition **C1** applied to system (3.9) reads

$$\text{rank} \begin{bmatrix} \lambda I_{r_2} & \lambda D_1 + K_1 & | & B_{11} & B_{12} & B_{13} \\ -I_{r_2} & \lambda M_1 & | & 0 & 0 & 0 \\ 0 & \lambda D_2 + K_2 & | & 0 & 0 & B_{23} \\ 0 & K_3 & | & 0 & 0 & 0 \\ \hline 0 & \lambda D_4 + K_4 & | & 0 & \Sigma_1 & B_{43} \\ 0 & K_5 & | & 0 & 0 & \Sigma_0 \\ 0 & 0 & | & 0 & 0 & 0 \end{bmatrix} = d + r_2 \text{ for all } \lambda \in \mathbb{C}. \quad (4.1) \quad \{a1\}$$

400 By using matrix row manipulation in order to eliminate λI_{r_2} in the first row, we see
 401 that (4.1) is equivalent to the condition

$$\text{rank} \begin{bmatrix} 0 & \lambda^2 M_1 + \lambda D_1 + K_1 & | & B_{11} & B_{12} & B_{13} \\ -I_{r_2} & \lambda M_1 & | & 0 & 0 & 0 \\ 0 & \lambda D_2 + K_2 & | & 0 & 0 & B_{23} \\ 0 & K_3 & | & 0 & 0 & 0 \\ \hline 0 & \lambda D_4 + K_4 & | & 0 & \Sigma_1 & B_{43} \\ 0 & K_5 & | & 0 & 0 & \Sigma_0 \\ 0 & 0 & | & 0 & 0 & 0 \end{bmatrix} = d + r_2 \text{ for all } \lambda \in \mathbb{C}. \quad (4.2) \quad \{a2\}$$

402 Clearly, this holds true if and only if $\text{rank} [\lambda^2 M + \lambda D + K, B] = d$, which is exactly
 403 the rank condition **C21**.

404 ii) Due to Definition 2.3, we see that **C0** = **C1** + **C3**, and hence we need to prove
 405 that condition **C3** is equivalent to condition **C23**. Now let us look at condition **C3**,

406 which means that the matrix

$$\begin{array}{c|cc|ccc} r_2 & I_{r_2} & D_1 & B_{11} & B_{12} & B_{13} \\ \hline r_2 & 0 & M_1 & 0 & 0 & 0 \\ r_1 & 0 & D_2 & 0 & 0 & B_{23} \\ r_0 & 0 & 0 & 0 & 0 & 0 \\ \hline \varphi_1 & 0 & D_4 & 0 & \Sigma_1 & B_{43} \\ \varphi_0 & 0 & 0 & 0 & 0 & \Sigma_0 \\ v & 0 & 0 & 0 & 0 & 0 \end{array}$$

407 has full row rank ($d + r_2$). Recall that in the strangeness-free form (3.10) the matrix
408 $\begin{bmatrix} M_1 \\ \hat{D}_2 \end{bmatrix}$ has full row rank. Therefore, condition **C3** holds true if and only if $r_0 = v = 0$.
409 Moreover, condition **C23**, which means that the matrix

$$\begin{array}{c|cc|ccc} r_2 & M_1 & D_1 & B_{11} & B_{12} & B_{13} \\ \hline r_1 & 0 & D_2 & 0 & 0 & B_{23} \\ r_0 & 0 & 0 & 0 & 0 & 0 \\ \hline \varphi_1 & 0 & D_4 & 0 & \Sigma_1 & B_{43} \\ \varphi_0 & 0 & 0 & 0 & 0 & \Sigma_0 \\ v & 0 & 0 & 0 & 0 & 0 \end{array}.$$

410 has full row rank, is fulfilled also only when $r_0 = v = 0$. Thus, two conditions **C3** and
411 **C23** are equivalent, and hence, this completes the proof of this part.
412 iii) In order to prove that condition **C21** is preserved under the strangeness-free
413 formulation we only need to prove that it is preserved under one index reduction
414 step. First we notice that for any two strongly equivalent tuples (M, D, K, B) and
415 $(\hat{M}, \hat{D}, \hat{K}, \hat{B})$ we have that

$$[\lambda^2 M + \lambda D + K, B] = U [\lambda^2 M + \lambda D + K, B] \begin{bmatrix} I & 0 \\ 0 & V \end{bmatrix}.$$

416 Thus, rank $[\lambda^2 M + \lambda D + K, B]$ is invariant under strongly equivalent relation. Con-
417 sequently, we may assume that (M, D, K, B) takes the form as in the right hand side
418 of (3.5). For notational convenience, we will omit the super script *new* and rewrite
419 our system as follows.

$$\begin{bmatrix} M_1 \\ 0 \\ 0 \\ 0 \\ 0 \end{bmatrix} x(n+2) + \begin{bmatrix} D_1 \\ D_2 \\ 0 \\ \hat{D}_4 \\ 0 \end{bmatrix} x(n+1) + \begin{bmatrix} K_1 \\ K_2 \\ K_3 \\ K_4 \\ K_5 \\ 0 \end{bmatrix} x(n) = \begin{bmatrix} B_{11} & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & 0 \\ 0 & \Sigma_1 & 0 \\ 0 & 0 & \Sigma_0 \\ 0 & 0 & 0 \end{bmatrix} v(n), \quad \begin{array}{l} r_2 \\ r_1 \\ \frac{r_0}{\varphi_1} (4.3) \\ \varphi_0 \\ v \end{array}$$

420 where M_1, D_2, K_3 have full row rank, and the matrices Σ_0, Σ_1 are digonal and
421 nonsingular. We recall, that due to [24, Lemma 4.4], one step index reduction in
422 the strangeness-free formulation is indeed transforming (4.3) into the new form which

423 reads

$$\underbrace{\begin{bmatrix} S^{(2)}M_1 \\ 0 \\ 0 \\ 0 \\ \hline 0 \\ 0 \\ 0 \end{bmatrix}}_{\tilde{M}} x(n+2) + \underbrace{\begin{bmatrix} S^{(2)}D_1 \\ Z^{(2)}D_1+Z^{(4)}K_2 \\ S^{(1)}D_2 \\ 0 \\ \hline D_4 \\ 0 \\ 0 \end{bmatrix}}_{\tilde{D}} x(n+1) + \underbrace{\begin{bmatrix} S^{(2)}K_1 \\ Z^{(2)}K_1 \\ S^{(1)}K_2 \\ Z^{(1)}K_2 \\ \hline K_3 \\ K_4 \\ K_5 \\ 0 \end{bmatrix}}_{\tilde{K}} x(n) = \underbrace{\begin{bmatrix} S^{(2)}B_{11} & 0 & 0 \\ Z^{(2)}B_{11} & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & 0 \\ \hline 0 & \Sigma_1 & 0 \\ 0 & 0 & \Sigma_0 \\ 0 & 0 & 0 \end{bmatrix}}_{\tilde{B}} v(n) . \quad (4.4) \quad \text{fa4}$$

424 Here, the matrices $S^{(i)}$, $i = 1, 2$, and $Z^{(j)}$, $j = 1, \dots, 5$ satisfy the following conditions.

- 425 i) For $i = 1, 2$, the matrices $\begin{bmatrix} S^{(i)} \\ Z^{(i)} \end{bmatrix} \in \mathbb{R}^{r_i, r_i}$ are orthogonal, and $r_i = d_i + s_i$.
ii) The following identities hold true.

$$\begin{aligned} Z^{(1)}D_2 + Z^{(3)}K_3 &= 0, \\ Z^{(2)}M_1 + Z^{(4)}D_2 + Z^{(5)}K_3 &= 0. \end{aligned}$$

426 Consider the matrix $\left[\lambda^2 \tilde{M} + \lambda \tilde{D} + \tilde{K}, \tilde{B} \right]$, we directly see that

$$\left[\lambda^2 \tilde{M} + \lambda \tilde{D} + \tilde{K}, \tilde{B} \right] = U_\lambda \left[\lambda^2 M + \lambda D + K, B \right],$$

427 where the matrix U_λ is defined as

$$U_\lambda := \begin{bmatrix} \begin{bmatrix} S^{(2)} \\ Z^{(2)} \end{bmatrix} & \begin{bmatrix} 0 \\ \lambda Z^{(4)} \end{bmatrix} & \begin{bmatrix} 0 \\ \lambda^2 Z^{(5)} \end{bmatrix} & 0 & 0 & 0 \\ 0 & \begin{bmatrix} S^{(1)} \\ Z^{(1)} \end{bmatrix} & \begin{bmatrix} 0 \\ \lambda Z^{(3)} \end{bmatrix} & 0 & 0 & 0 \\ 0 & 0 & I_{r_0} & 0 & 0 & 0 \\ 0 & 0 & 0 & I_{\varphi_1} & 0 & 0 \\ 0 & 0 & 0 & 0 & I_{\varphi_0} & 0 \\ 0 & 0 & 0 & 0 & 0 & I_v \end{bmatrix} .$$

428 Since all matrices on the main diagonal are orthogonal, we see that U_λ is nonsingular
429 for all $\lambda \in \mathbb{C}$. Therefore,

$$\text{rank} \left[\lambda^2 \tilde{M} + \lambda \tilde{D} + \tilde{K}, \tilde{B} \right] = \text{rank} \left[\lambda^2 M + \lambda D + K, B \right] \quad \text{for all } \lambda \in \mathbb{C},$$

430 and hence, condition **C21** is preserved under one index reduction step. This finishes
431 our proof. \square

432 In comparison to Theorem 3.9, the advantage of the minimal extension form (3.9)
433 will be proven in the following theorem.

434 THEOREM 4.5. *Consider the descriptor system (1.1), its the strangeness-free from
435 (3.10) and the minimal extension form (3.9). If system (1.1) is R2-controllable then so
436 is system (3.10). Furthermore, if this is the case, then system (3.9) is R-controllable.*

437 *Proof.* Making use of Theorem 4.2 i) and Lemma 4.4 ii) we see that the constant
438 rank condition **C21** holds for the coefficients of system (3.9). As in the proof of
439 Lemma 4.4, due to simple matrix row manipulations, from system (3.9) we see that

$$\text{rank} \left[\lambda \tilde{E} - \tilde{A}, \tilde{B} \right] = \text{rank} \left[\lambda^2 M + \lambda D + K, B \right] + r_2 ,$$

440 and hence, $\text{rank} \left[\lambda \tilde{E} - \tilde{A}, \tilde{B} \right] = d + r_2$. This implies that system (3.9) is R-controllable.
441 \square

442 THEOREM 4.6. Consider the descriptor system (1.1) and its the strangeness-
 443 free from (3.10). Then system (1.1) is C2-controllable if and only if the following
 444 conditions are satisfied.

- 445 i) The matrix coefficients of system (1.1) satisfies condition **C21**.
 446 ii) The matrix coefficients of the strangeness-free system (3.10) satisfies condition
 447 **C23**.

448 Proof. The proof is followed directly from Definition 4.3 and Lemma 4.4. \square

449 The following example shows that condition **C23** is not invariant under the
 450 strangeness-free formulation.

451 EXAMPLE 4.7. Consider the following system

$$\underbrace{\begin{bmatrix} 1 & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix}}_M x(n+2) + \underbrace{\begin{bmatrix} 0 & 1 & 0 \\ 1 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix}}_D x(n+1) + \underbrace{\begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 0 \end{bmatrix}}_K x(n) = \underbrace{\begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}}_B u(n) . \quad \text{(4.6)} \quad \text{eq4.1}$$

452 Due to the strangeness-free formulation in [24], we can shift the second row equation
 453 forward to obtain

$$[1 \ 0 \ 0] x(n+2) + [0 \ 1 \ 0] x(n+1) = 0 .$$

454 By removing this from the first equation, we obtain that $[1 \ 0 \ 0] x(n) = 0$. Therefore,
 455 we obtain the system

$$\begin{bmatrix} 0 & 0 & 0 \\ 1 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} x(n+1) + \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 0 \end{bmatrix} x(n) = \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix} u(n) .$$

456 Analogously, by subtracting the shifted version of the first row equation from the second
 457 equation, we obtain the strangeness-free formulation (2.5) that reads

$$\underbrace{\begin{bmatrix} 0 & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix}}_{\hat{M}} x(n+2) + \underbrace{\begin{bmatrix} 0 & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix}}_{\hat{D}} x(n+1) + \underbrace{\begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 0 \end{bmatrix}}_{\hat{K}} x(n) = \underbrace{\begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}}_{\hat{B}} u(n) . \quad \text{(4.7)} \quad \text{eq4.2}$$

458 Clearly, $\text{rank}[\hat{M}, \hat{D}, \hat{B}] = 3 > 1 = \text{rank}[\hat{M}, \hat{D}, \hat{B}]$. This means that condition
 459 **C23** is not invariant under the strangeness-free formulation.

460 Furthermore, by verifying condition **C21**, we directly see that system (4.6) is R2-
 461 controllable. Indeed, we have that

$$\text{rank} [\lambda^2 M + \lambda D + K \mid B] = \text{rank} \left[\begin{array}{ccc|c} \lambda^2 + 1 & \lambda & 0 & 0 \\ \lambda & 1 & 0 & 0 \\ 0 & 0 & 0 & 1 \end{array} \right] = 3 .$$

462 As obtained above, since $\text{rank}[\hat{M}, \hat{D}, \hat{B}] = 1 < 3$, system (4.6) is not C2-controllable.

463 In fact, from (4.7), it is straightforward that system (4.6) is not C-controllable.

464 REMARK 4.8. As stated in Theorem 4.6, condition **C23** must be required for the
 465 strangeness-free system (3.10) instead of for the original system (1.1). This is the
 466 main difference between discrete and continuous time descriptor systems. In details,
 467 [25, Corollary 3.11 ii) and Theorem 3.18 iv)] imply that the continuous-time version
 468 of system (4.6) is C2-controllable (resp. C-controllable).

469 Naturally, one may ask whether one can verify the $C2$ -controllability of system
 470 (1.1) without performing an index reduction procedure (i.e., without determining the
 471 strangeness-free form (3.10)). In fact, the positive answer is given in the following
 472 theorem.

473 THEOREM 4.9. *Consider the descriptor system (1.1) and its condensed form
 474 (3.4). Then, system (1.1) is $C2$ -controllable if and only if two following conditions
 475 are satisfied.*

- 476 i) *The matrix coefficients of system (1.1) satisfies condition **C21**.*
 477 ii) *In the upper part of system (3.4), $r_0 = v_0 = 0$ and the matrix $\begin{bmatrix} M_1 \\ D_2 \end{bmatrix}$ has full row
 478 rank.*

479 Finally, condition ii) is equivalent to the requirement that $\text{rank}[M, D, B] = d$ and
 480 the matrix $\begin{bmatrix} M_1 \\ D_2 \end{bmatrix}$ has full row rank.

481 Proof. Due to Definition 4.3 system (1.1) is $C2$ -controllable if and only if the
 482 minimal extension form (3.9) is C -controllable. From Definition 2.3 and Lemma 4.4
 483 iii, we see that $\mathbf{C0} = \mathbf{C1} + \mathbf{C3}$ and $\mathbf{C1}$ is equivalent to condition **C21**.

484 Hence, we only need to prove that condition **C3** is equivalent to the claim ii). Now
 485 let us look at condition **C3**, which means that the matrix

$$\begin{array}{c|cc|ccc} r_2 & I_{r_2} & D_1 & B_{11} & B_{12} & B_{13} \\ r_2 & 0 & M_1 & 0 & 0 & 0 \\ r_1 & 0 & D_2 & 0 & 0 & B_{23} \\ \hline r_0 & 0 & 0 & 0 & 0 & 0 \\ \varphi_1 & 0 & D_4 & 0 & \Sigma_1 & B_{43} \\ \varphi_0 & 0 & 0 & 0 & 0 & \Sigma_0 \\ v & 0 & 0 & 0 & 0 & 0 \end{array}$$

486 has full row rank, is fulfilled if and only if $\begin{bmatrix} M_1 \\ D_2 \end{bmatrix}$ has full row rank and $r_0 = v = 0$,
 487 which is nothing else than the claim ii). Finally, the last claim is directly followed
 488 from Corollary 3.6. This completes the proof. \square

489 We summarize the relation between the controllability of the systems discussed
 490 above in Figure 4.1. Now let us discuss the C -controllability of system (1.1). In the
 491 following example we illustrate that for second order systems, C -controllability does
 492 not always imply Y -controllability.

493 EXAMPLE 4.10. *Consider the following system*

$$\underbrace{\begin{bmatrix} 0 & -1 \\ 0 & 0 \end{bmatrix}}_M x(n+2) + \underbrace{\begin{bmatrix} 1 & 1 \\ 0 & 1 \end{bmatrix}}_D x(n+1) + \underbrace{\begin{bmatrix} -1 & 0 \\ -1 & -1 \end{bmatrix}}_K x(n) = \begin{bmatrix} 2 \\ 0 \end{bmatrix} u(n) . \quad (4.8) \quad \text{eq3.6}$$

Clearly, the structure of the pair (M, D) implies that system (4.8) is not Y -controllable.
 By adding the shifted version of the second row equation to the first row, we can
 transform (4.8) to the first order system

$$\begin{bmatrix} 0 & 0 \\ 0 & 1 \end{bmatrix} x(n+1) + \begin{bmatrix} -1 & 0 \\ -1 & -1 \end{bmatrix} x(n) = \begin{bmatrix} 2 \\ 0 \end{bmatrix} u(n) ,$$

494 which can be directly verified that is C -controllable. Thus, C -controllability does not
 495 imply Y -controllability. The same observation can be made for continuous-time sec-
 496 ond order descriptor systems by considering the following system

$$\begin{bmatrix} 0 & -1 \\ 0 & 0 \end{bmatrix} \ddot{x}(t) + \begin{bmatrix} 1 & 1 \\ 0 & 1 \end{bmatrix} \dot{x}(t) + \begin{bmatrix} -1 & 0 \\ -1 & -1 \end{bmatrix} x(t) = \begin{bmatrix} 2 \\ 0 \end{bmatrix} u(t) .$$

497 Example (4.10) suggests, that we should discuss the C-controllability of the strangeness-free formulation (3.10) instead of the original system (1.1). The characterizations
498 of C-controllability for system (1.1) are given in the following theorem.

500 THEOREM 4.11. *Consider the system (1.1) and assume that it is already in the
501 strangeness-free form (3.10). Let \mathcal{R}_{ext} be the reachable set of the minimal extension
502 form (3.9). Let $E_0 = \text{diag}(0_{r_2}, I_d)$. Then the following assertions are equivalent.*

503 i) System (1.1) is C-controllable.

504 ii) System (1.1) is R-controllable and $\text{Im}E_0 \subseteq \mathcal{R}_{ext}$.

505 iii) System (1.1) is R-controllable and $\text{rank}[M, D, B] = d$.

506 Proof. Notice that in system (3.9) $\xi_n = \begin{bmatrix} z_n \\ x_n \end{bmatrix} \in \mathbb{R}^{r_2+d}$, so the equivalence between
507 i) and ii) is straightforward. From the definition of C-controllability and the fact that
508 system (1.1) is square, we have $r_0 = v_0 = 0$. Corollary 3.6, therefore, implies that
509 $\text{rank}[M, D, B] = d$. Hence, we have proved that $i) \Rightarrow iii)$. Now we prove that
510 $iii) \Rightarrow ii)$.

511 Due to Corollary 3.6, we see that $r_0 = v_0 = 0$, and hence the 3rd and 6th rows
512 are not present in the form (3.9). Applying Theorem 3.8 i), in analogous to the
513 sufficiency part, we see that there exist two matrices F_d, F_v such that the matrix

514 $\begin{bmatrix} M_1^T & D_2^T & K_3^T & \tilde{D}_4^T & \tilde{K}_5^T \end{bmatrix}^T$ has full row rank, where

$$\tilde{D}_4 := D_4 + [0 \ \Sigma_1 \ B_{43}] F_v, \quad \tilde{K}_5 := K_5 + [0 \ 0 \ \Sigma_0] F_d.$$

515 Consequently, by introducing a new input function $w = \{w(n)\}$ such that

$$u(n) = -F_v x(n+1)(t) - F_d x(n) + w(n) \quad \text{for all } n \geq n_0,$$

516 we can transform the minimal extension form (3.9) to the closed loop system

$$\begin{array}{c|cc} \begin{bmatrix} I_{r_2} & D_1 \\ 0 & M_1 \\ 0 & D_2 \\ 0 & \tilde{D}_4 \\ 0 & 0 \end{bmatrix} \xi(n+1) + \begin{bmatrix} 0 & K_1 \\ -I_{r_2} & 0 \\ 0 & K_2 \\ 0 & K_4 \\ 0 & \tilde{K}_5 \end{bmatrix} \xi(n) = & \begin{bmatrix} B_{11} & B_{12} & B_{13} \\ 0 & 0 & 0 \\ 0 & 0 & B_{23} \\ 0 & \Sigma_1 & B_{43} \\ 0 & 0 & \Sigma_0 \end{bmatrix} w(n), & \begin{array}{l} r_2 \\ r_2 \\ r_1 \\ \varphi_1 \\ \varphi_0 \end{array} \\ \hline & & \end{array} \quad (4.9) \quad \{\text{eq4.3}\}$$

517 Notice that, since $w(n)$ can be freely chosen like $u(n)$, we neither change the R-
518 controllability or change the reachable set \mathcal{R} of system (1.1). Since the matrix
519 $\begin{bmatrix} M_1^T & D_2^T & \tilde{D}_4^T & \tilde{K}_5^T \end{bmatrix}^T$ has full row rank, the matrix

$$\begin{bmatrix} I_{r_2} & D_1 \\ 0 & M_1 \\ 0 & D_2 \\ 0 & \tilde{D}_4 \\ 0 & \tilde{K}_5 \end{bmatrix}$$

520 also has full row rank, and hence, system (4.9) is regular and strangeness-free. Corol-
521 lary 2.1 applied to system (4.9) implies that the reachable subspace of (4.9) is $\mathcal{R}_{ext} =$
522 \mathbb{R}^{r_2+d} and hence, $\text{Im}E_0 \subseteq \mathcal{R}_{ext}$. This completes the proof. \square

523 By following [11], we can determine the reachable set \mathcal{R} of system (4.9) based on
524 the Kronecker-Weierstraß canonical form of (1.1)

$$\begin{bmatrix} I_{n_1} & 0 \\ 0 & 0 \end{bmatrix} \tilde{\xi}(n+1) = \begin{bmatrix} \bar{A}_1 & 0 \\ 0 & I_{\varphi_0} \end{bmatrix} \tilde{\xi}(n) + \begin{bmatrix} \bar{B}_1 \\ \bar{B}_2 \end{bmatrix} v(n), \quad (4.10) \quad \{\text{eq4.4}\}$$

525 where $n_1 = 2r_2 + r_1 + \varphi_1$. Now we are ready to discuss the R-controllability of the
 526 strangeness-free system (1.1).

527 THEOREM 4.12. Consider the system (1.1) and assume that it is already in the
 528 strangeness-free form (3.10). Let us also consider the system (4.10). Then, system
 529 (1.1) is R-controllable if and only if for the corresponding first order system (4.10)
 530 the matrix product $[0 \ I_{n_1-r_2}] \mathcal{K}(\bar{A}_1, \bar{B}_1)$ has full row rank, where

$$\mathcal{K}(\bar{A}_1, \bar{B}_1) := [\bar{B}_1, \bar{A}_1\bar{B}_1, \dots, \bar{A}_1^{n_1-1}\bar{B}_1], \quad (4.11) \quad \text{[eq4.5]}$$

531 Here the matrix $[0 \ I_{n_1-r_2}] \in \mathbb{R}^{n_1-r_2, n_1}$.

532 Proof. From [11, Chap. 2] we see that the first order system (4.10) has the reachable set $\mathcal{R} = \mathbb{R}^{n_1} \oplus \text{Im}(B_2)$, and (4.10) is R-controllable if and only if $\text{Im}\mathcal{K}(\bar{A}_1, \bar{B}_1) =$
 533 \mathbb{R}^{n_1} . Furthermore, notice that the first r_2 variables of (4.9) come from the trans-
 534 formation of second order system (3.10) to the first order system (4.9) and are not
 535 relevant to consider for R-controllability. Therefore, the proof is straightly followed.
 536 \square

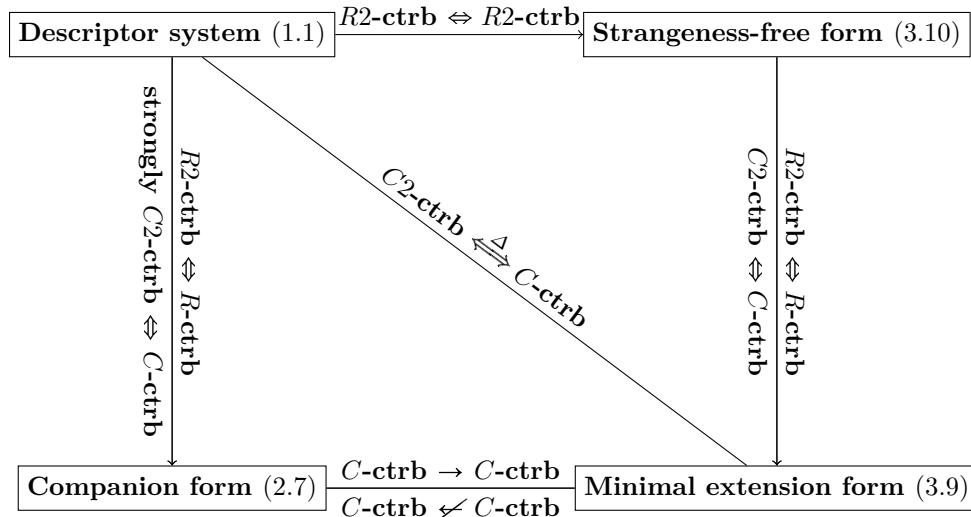


FIG. 4.1. Controllability diagrams of system (1.1) and its reformulations

538 **5. Observability of second order descriptor systems.** In this section we
 539 give a few result about the corresponding observability of system (1.1). For this let
 540 us denote by $\mathcal{P}_{r,2}$ the projection onto the right finite eigenspace corresponding to the
 541 finite eigenvalues of the matrix polynomial $\lambda^2 M + \lambda D + K$, [15]. First we recall three
 542 important concepts.

543 DEFINITION 5.1. *i) System (1.1) is called C-observable if from a response $y = 0$
 544 for the input $u = 0$ it follows that system (1.1) has only one trivial solution $x = 0$.*

545 *ii) It is called R-observable if from a response $y = 0$ for the input $u = 0$ it follows
 546 that $\mathcal{P}_{r,2}x = 0$.*

547 *iii) It is called causal observable (Y-observable) if its state $x(k)$ at any time point k
 548 is uniquely determined by initial condition $(x(0), x(1))$ and the former (k included)
 549 inputs $u(i)$, together with former outputs $y(i)$, $i = 0, \dots, k$.*

550 REMARK 5.2. *Due to linear property of system (1.1), C-observability also means
 551 that for any unknown initial condition $(x(0), x(1))$, there exists a finite integer $k > 0$,
 552 such that the knowledge about former (k included) inputs $u(i)$, together with for-
 553 mer output $y(i)$, $i = 0, \dots, k$ suffices to determine uniquely the initial condition
 554 $(x(0), x(1))$.*

555 It is straightforward to see that all three observability concepts above are invariant
 556 under left equivalent transformation. On the other hand, since the index reduction
 557 procedure, which transforms system (1.1) to the form (3.10), does not alter the so-
 558 lution set of system (1.1), the C- and R-observability are preserved. Furthermore,
 559 due to Remark 3.10, the index reduction procedure has been performed only on the
 560 causal uncontrollable part, which implies that the Y-observability is also preserved.
 561 The following lemma plays the key role in our study about the observability of (1.1).

562 LEMMA 5.3. *Consider system (1.1), the the strangeness-free from (3.10) and
 563 the minimal extension form (3.9). Then, system (3.10) is Y-observable (resp., R-
 564 observable) if and only if system (3.9) is also Y-observable (resp., R-observable).*

565 Proof. Concerning about the Y-observability, the proof is straightforward, since
 566 the transformation from (3.10) to (3.9) keeps both the input and output, while the
 567 second block equation of (3.9) is nothing else than $z(n) = Mx(n+1)$, which does
 568 not have any impact on the causality of the system. About the R-observability, the
 569 proof is essentially the same as the proof of [25, Thm 4.3], so we will omit it to keep
 570 the brevity of this paper. \square

571 Making use of Lemma 5.3, we see that the first order duality of controllability
 572 and observability [11, 13] can be directly extended to the second order case for system
 573 (1.1) and the dual system

$$\begin{aligned} M^T x(n+2) + D^T x(n+1) + K^T x(n) &= C^T u(n) \quad \text{for all } n \geq n_0, \\ y(k) &= Bx(k), \\ x(n_0) &= x_0, \quad x(n_0+1) = x_1. \end{aligned} \tag{5.1} \quad \{\text{dual system}\}$$

574 THEOREM 5.4. *Consider the second order descriptor system (1.1) and the dual
 575 system (5.1). Then the following assertions hold true.*

576 *i) System (1.1) is C-observable if and only if the dual system (5.1) is C2-controllable.*
 577 *ii) System (1.1) is R-observable if and only if the dual system (5.1) is R2-controllable.*
 578 *iii) System (1.1) is Y-observable if and only if the dual system (5.1) is Y-controllable
 579 via displacement-velocity feedback.*

580 Proof. Due to Lemma 5.3, the proof is directly obtained by checking rank condi-
 581 tions for the first order system (3.9), so it will be omitted to keep the brevity of this
 582 paper. \square

584 COROLLARY 5.5. Consider the second order descriptor system (1.1). Then, it is
 585 i) R -observable if and only if

$$\text{rank} \begin{bmatrix} \lambda^2 M + \lambda D + K \\ C \end{bmatrix} = d;$$

586 ii) C -observable if and only if it is R -observable and the matrix coefficients in the
 587 strangeness-free form of the dual system (5.1) satisfy

$$\text{rank} \begin{bmatrix} \hat{M}^T & \hat{D}^T & \hat{C}^T \end{bmatrix} = d; \quad (5.2) \quad \{\text{eq5.1}\}$$

588 iii) Y -observable if and only if

$$\text{rank} \begin{bmatrix} M \\ T_{\infty}^1 D \\ T_{\infty}^2 K \\ C \end{bmatrix} = d,$$

589 where rows of T_{∞}^1 form a basis of cokernel M , and rows of T_{∞}^2 form the basis of

$$\text{cokernel} \begin{bmatrix} M \\ DZ_1 \end{bmatrix} \setminus \text{cokernel} \begin{bmatrix} M \\ DZ_1 \\ KZ_3 \end{bmatrix},$$

590 and rows of Z_1 and of Z_3 form a basis of kernel M and kernel $\begin{bmatrix} M \\ D \end{bmatrix}$, respectively.

591 In analogous to the controllability case, see Example 4.7, here we notice that the
 592 rank condition $\text{rank} [M^T D^T C^T] = d$ implies (5.2), but the converse is not true.
 593 Consequently, system (1.1) may not be Y -observable, even if $\text{rank} [M^T D^T C^T] = d$,
 594 as illustrated in the following example.

EXAMPLE 5.6. Consider the system (1.1) which reads

$$\underbrace{\begin{bmatrix} 1 & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix}}_M x(n+2) + \underbrace{\begin{bmatrix} 0 & 1 & 0 \\ 1 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix}}_D x(n+1) + \underbrace{\begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 0 \end{bmatrix}}_K x(n) = \underbrace{\begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}}_B u(n), \quad (5.3) \quad \{\text{eq5.3}\}$$

$$y(n) = \underbrace{\begin{bmatrix} 0 & 0 & 1 \end{bmatrix}}_C x(n).$$

595 Since the matrices M , D , K are symmetric and $C^T = B$, we see that the dual system
 596 of (5.3) is nothing else than itself. As in Example 4.7, the strangeness-free formulation
 597 of this dual system reads

$$\underbrace{\begin{bmatrix} 0 & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix}}_{\hat{M}^T} x(n+2) + \underbrace{\begin{bmatrix} 0 & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix}}_{\hat{D}^T} x(n+1) + \underbrace{\begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 0 \end{bmatrix}}_{\hat{K}^T} x(n) = \underbrace{\begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}}_{\hat{C}^T} u(n). \quad (5.4) \quad \{\text{eq5.4}\}$$

598 Consequently, the dual system is not C -controllable (and hence not $C2$ -controllable).
 599 Theorem 5.4 i) applied to system (5.3) implies that this system is not C -observable,
 600 despite the fact that $\text{rank} [M^T D^T C^T] = 3$. This agrees with Corollary 5.5 ii), since
 601 $\text{rank} [\hat{M}^T \hat{D}^T \hat{C}^T] = 1 < 3$. Besides that, by direct computation, we see that system
 602 (5.3) is R -observable but not Y -observable.

6. Conclusion and Outlook. In this paper we have presented the theoretical
 analysis for the controllability of linear, second order descriptor systems in discrete-
 time. We have modified an algebraic method proposed in [25, 28] to make it more
 convenient and reliable to apply, in order to study second order descriptor systems. We
 have given several necessary and sufficient conditions, which are numerically verifiable,
 in order to characterize all the fundamental controllability concepts for the considered
 systems. We have pointed out that C-controllable does not imply Y-controllable, and
 have also presented suitable feedback design strategy in order to eliminate the causal
 behavior of the considered systems. Future research includes the generalization of
 this approach to higher order descriptor systems, and also a comparable framework
 for the observability concepts.

REFERENCES

- [1] Taha Abdelaziz, *Eigenstructure assignment by displacement-acceleration feedback for second-order systems*, Journal of Dynamic Systems, Measurement, and Control **138(6)** (2016), 064502–064502–7. 2, 11

[2] Taha Abdelaziz, *Robust solution for second-order systems using displacement-acceleration feedback*, Journal of Control, Automation and Electrical Systems **31** (2019), 2195–3899. 2

[3] Taha H.S. Abdelaziz, *Robust pole assignment using velocity-acceleration feedback for second-order dynamical systems with singular mass matrix*, ISA Transactions **57** (2015), 71 – 84. 2

[4] R.P. Agarwal, *Difference equations and inequalities: Theory, methods, and applications*, Chapman & Hall/CRC Pure and Applied Mathematics, CRC Press, 2000. 1

[5] T. Berger and T. Reis, *Controllability of linear differential-algebraic systems - a survey*, Surveys in Differential-Algebraic Equations I, Differential-Algebraic Equations Forum (A. Ilchmann and T. Reis, eds.), Springer-Verlag, 2013, pp. 1–61. 1, 4

[6] K. E. Brenan, S. L. Campbell, and L. R. Petzold, *Numerical solution of initial-value problems in differential algebraic equations*, 2nd ed., SIAM Publications, Philadelphia, PA, 1996. 1

[7] R. W. Brockett, *Finite dimensional linear systems*, John Wiley and Sons, New York, NY, 1970. 4

[8] A. Bunse-Gerstner, R. Byers, V. Mehrmann, and N. K. Nichols, *Feedback design for regularizing descriptor systems*, Lin. Alg. Appl. **299** (1999), 119–151. 12

[9] D. Chu and V. Mehrmann, *Disturbance decoupling for descriptor systems*, SIAM J. Cont. **38** (2000), 1830–1858. 12

[10] D. Chu, V. Mehrmann, and N. K. Nichols, *Minimum norm regularization of descriptor systems by output feedback*, Lin. Alg. Appl. **296** (1999), 39–77. 12

[11] L. Dai, *Singular control systems*, Springer-Verlag, Berlin, Germany, 1989. 1, 2, 4, 18, 19, 20

[12] Nguyen Huu Du, Vu Hoang Linh, Volker Mehrmann, and Do Duc Thuan, *Stability and robust stability of linear time-invariant delay differential-algebraic equations.*, SIAM J. Matr. Anal. Appl. **34** (2013), no. 4, 1631–1654. 1

[13] G.R. Duan, *Analysis and design of descriptor linear systems*, Advances in Mechanics and Mathematics, Springer New York, 2010. 20

[14] S.N. Elaydi, *An introduction to difference equations*, Undergraduate Texts in Mathematics, Springer New York, 2013. 1

[15] I. Gohberg, P. Lancaster, and L. Rodman, *Matrix polynomials*, Academic Press, New York, NY, 1982. 20

[16] M. Green and D.J.N. Limebeer, *Linear robust control*, Dover Books on Electrical Engineering, Dover Publications, Incorporated, 2012. 4

[17] T. Helmy and S. Abdelaziz, *Robust pole placement for second-order linear systems using velocity-plus-acceleration feedback*, IET Control Theory Applications **7** (2013), no. 14, 1843–1856. 2

[18] P. C. Hughes and R. E. Skelton, *Controllability and Observability of Linear Matrix-Second-Order Systems*, Journal of Applied Mechanics **47** (1980), no. 2, 415–420. 11

[19] Nicholas P. Karampetakis and Anastasia Gregoriadou, *Reachability and controllability of discrete-time descriptor systems*, Internat. J. Control **87** (2014), 235 – 248. 1, 4

[20] W.G. Kelley and A.C. Peterson, *Difference equations: An introduction with applications*, Harcourt/Academic Press, 2001. 1

[21] P. Kunkel and V. Mehrmann, *Differential-algebraic equations – analysis and numerical solu-*

- tion, EMS Publishing House, Zürich, Switzerland, 2006. 1, 4

[22] L. Lang, W. Chen, B. R. Bakshi, P. K. Goel, and S. Ungarala, *Bayesian estimation via sequential monte carlo sampling: constrained dynamic systems*, Automatica **43** (2007), 1615–1622. 1

[23] V. Linh, N. Thanh Nga, and D. Thuuan, *Exponential stability and robust stability for linear time-varying singular systems of second order difference equations*, SIAM J. Matr. Anal. Appl. **39** (2018), no. 1, 204–233. 1, 4

[24] V.H. Linh and H. Phi, *Index reduction for second order singular systems of difference equations*, Lin. Alg. Appl. **608** (2021), 107 – 132. 1, 5, 6, 10, 14, 16

[25] P. Losse and V. Mehrmann, *Controllability and observability of second order descriptor systems*, SIAM J. Cont. Optim. **47**(3) (2008), 1351–1379. 2, 6, 7, 10, 11, 16, 20, 22

[26] P. Losse, V. Mehrmann, L.K. Poppe, and T. Reis, *The modified optimal \mathcal{H}_∞ control problem for descriptor systems*, SIAM J. Cont. **47** (2008), 2795–2811. 7

[27] D. G. Luenberger, *Dynamic equations in descriptor form*, IEEE Trans. Automat. Control **AC-22** (1977), 312–321. 1, 7

[28] V. Mehrmann and C. Shi, *Transformation of high order linear differential-algebraic systems to first order*, Numer. Alg. **42** (2006), 281–307. 2, 6, 22

[29] Volker Mehrmann and Do Duc Thuuan, *Stability analysis of implicit difference equations under restricted perturbations*, SIAM J. Matr. Anal. Appl. **36** (2015), 178 – 202. 1, 2

[30] Lazaros Moysis, Nicholas Karampetakis, and Efstathios Antoniou, *Observability of linear discrete-time systems of algebraic and difference equations*, International Journal of Control **92** (2019), no. 2, 339–355. 1

[31] Lazaros Moysis, Nicholas P. Karampetakis, and Efstathios Antoniou, *Reachability and controllability of discrete-time descriptor systems*, Internat. J. Control **92** (2019), 339 – 355. 1

[32] E.D. Sontag, *Mathematical control theory: Deterministic finite dimensional systems*, Texts in Applied Mathematics, Springer New York, 2013. 4

[33] G. C. Verghese, B. C. Lévy, and T. Kailath, *A generalized state space for singular systems*, IEEE Trans. Automat. Control **AC-26** (1981), 811–831. 3

[34] Lena Wunderlich, *Numerical treatment of second order differential-algebraic systems*, Proc. Appl. Math. and Mech. (GAMM 2006, Berlin, March 27–31, 2006), vol. 6 (1), 2006, pp. 775–776. 2

[35] ———, *Analysis and numerical solution of structured and switched differential-algebraic systems*, Dissertation, Institut für Mathematik, TU Berlin, Berlin, Germany, 2008. 2, 6

[36] Peizhao Yu and Guoshan Zhang, *Eigenstructure assignment and impulse elimination for singular second-order system via feedback control*, IET Control Theory and Applications **10** (2016), 869–876(7). 2, 11

[37] K. Zhou, J.C. Doyle, and K. Glover, *Robust and optimal control*, Feher/Prentice Hall Digital and, Prentice Hall, 1996. 4