

Optimal control for linear port-Hamiltonian descriptor systems

Fabian Common



February 25, 2021

Outline

1 Port-Hamiltonian systems and their properties

2 Optimal control

General descriptor system

$$\begin{aligned} E\dot{x}(t) &= Ax(t) + Bu(t), \quad x(t_0) = x^0 \\ y(t) &= Cx(t) + Du(t) \end{aligned} \tag{1}$$

$$\begin{aligned} E, A \in \mathbb{C}^{n \times n}, B \in \mathbb{C}^{n \times m}, C \in \mathbb{C}^{p \times n}, D \in \mathbb{C}^{p \times m} \\ u : \mathbb{R} \rightarrow \mathbb{C}^m, x : \mathbb{R} \rightarrow \mathbb{C}^n, y : \mathbb{R} \rightarrow \mathbb{C}^m \end{aligned}$$

Port-Hamiltonian descriptor system

$$\begin{aligned}\tilde{E}\dot{x} &= (\tilde{J} - \tilde{R})\tilde{Q}x + (\tilde{B} + \tilde{P})u, \quad x(t_0) = x^0 \\ y &= (\tilde{B} - \tilde{P})^T\tilde{Q}x + (\tilde{S} + \tilde{N})u\end{aligned}\tag{2}$$

$\tilde{E}, \tilde{J}, \tilde{R}, \tilde{Q} \in \mathbb{C}^{n \times n}, \tilde{B}, \tilde{P} \in \mathbb{C}^{n \times m}, \tilde{S}, \tilde{N} \in \mathbb{C}^{p \times m}$

$\tilde{J} = -\tilde{J}^T, \tilde{N} = -\tilde{N}^T, \tilde{R} = \tilde{R}^T \succeq 0, \tilde{S} = \tilde{S}^T \succeq 0, \tilde{Q} = \tilde{Q}^T \succeq 0$

$$\begin{bmatrix} \tilde{R} & \tilde{P} \\ \tilde{P}^T & \tilde{S} \end{bmatrix} \succeq 0$$

Port-Hamiltonian descriptor system

$$\begin{aligned}\tilde{E}\dot{x} &= (\tilde{J} - \tilde{R})\tilde{Q}x + (\tilde{B} + \tilde{P})u, \quad x(t_0) = x^0 \\ y &= (\tilde{B} - \tilde{P})^T \tilde{Q}x + (\tilde{S} + \tilde{N})u\end{aligned}\tag{2}$$

$$\tilde{E}, \tilde{J}, \tilde{R}, \tilde{Q} \in \mathbb{C}^{n \times n}, \tilde{B}, \tilde{P} \in \mathbb{C}^{n \times m}, \tilde{S}, \tilde{N} \in \mathbb{C}^{p \times m}$$

$$\tilde{J} = -\tilde{J}^T, \tilde{N} = -\tilde{N}^T, \tilde{R} = \tilde{R}^T \succeq 0, \tilde{S} = \tilde{S}^T \succeq 0, \tilde{Q} = \tilde{Q}^T \succeq 0$$

$$\begin{bmatrix} \tilde{R} & \tilde{P} \\ \tilde{P}^T & \tilde{S} \end{bmatrix} \succeq 0$$

Port-Hamiltonian descriptor system

$$\begin{aligned}\tilde{E}\dot{x} &= (\tilde{J} - \tilde{R})x + (\tilde{B} + \tilde{P})u, \quad x(t_0) = x^0 \\ y &= (\tilde{B} - \tilde{P})^T x + (\tilde{S} + \tilde{N})u\end{aligned}\tag{2}$$

$$\tilde{E}, \tilde{J}, \tilde{R} \in \mathbb{C}^{n \times n}, \tilde{B}, \tilde{P} \in \mathbb{C}^{n \times m}, \tilde{S}, \tilde{N} \in \mathbb{C}^{p \times m}$$

$$\tilde{J} = -\tilde{J}^T, \tilde{N} = -\tilde{N}^T, \tilde{R} = \tilde{R}^T \succeq 0, \tilde{S} = \tilde{S}^T \succeq 0$$

$$\begin{bmatrix} \tilde{R} & \tilde{P} \\ \tilde{P}^T & \tilde{S} \end{bmatrix} \succeq 0$$

Properties

- passivity
- consistency
- regularity
- stability
- stabilizability
- detectability
- controllability
- observability

Passivity

Definition

We call a control problem *passive*, if there exists a state-dependant storage-function $\mathcal{H} \geq 0$, such that $\frac{d}{dt}\mathcal{H} \leq u^T y$.

[Beattie, Mehrmann, Xu, Zwart 2018]

$$\frac{d}{dt}\mathcal{H} = u^T y - \begin{bmatrix} x^T & u^T \end{bmatrix} \begin{bmatrix} \tilde{R} & \tilde{P} \\ \tilde{P}^T & \tilde{S} \end{bmatrix} \begin{bmatrix} x \\ u \end{bmatrix}$$

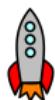
If $\begin{bmatrix} \tilde{R} & \tilde{P} \\ \tilde{P}^T & \tilde{S} \end{bmatrix} \equiv 0$, then the system is called *lossless*.

Consistency and Regularity

Definition



We call a control problem *consistent*, if there exists an input function u for which $\dot{x} = Ax + Bu, x(t_0) = x^0$ has a solution



Definition

We call a control problem *regular*, if it has a unique solution for every initial value that is consistent for the system with input u .

Theorem

If (E, A) is regular, then the control problem is consistent and regular.

Consistency and Regularity

Definition



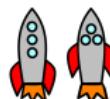
We call a control problem *consistent*, if there exists an input function u for which $\dot{x} = Ax + Bu, x(t_0) = x^0$ has a solution



Definition



We call a control problem *regular*, if it has a unique solution for every initial value that is consistent for the system with input u .



Theorem

If (E, A) is regular, then the control problem is consistent and regular.

Consistency and Regularity

Definition



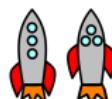
We call a control problem *consistent*, if there exists an input function u for which $\dot{x} = Ax + Bu, x(t_0) = x^0$ has a solution



Definition



We call a control problem *regular*, if it has a unique solution for every initial value that is consistent for the system with input u .



Theorem

If (E, A) is regular, then the control problem is consistent and regular.

Stability

Definition

We call a control problem *stable*, if one (and therefore every) solution to $\dot{x} = Ax$ is bounded on one (and therefore every) interval (t_0, ∞) .

We call it *asymptotically stable*, if additionally $\lim_{x \rightarrow \infty} \|x(t)\| = 0$.

Lemma

asymptotically stable $\Leftrightarrow (E, A)$ regular, $\lambda E - A \prec 0$, all infinite Eigenvalues semisimple

Lemma

stable $\Leftrightarrow (E, A)$ regular, $\lambda E - A \preceq 0$, all purely imaginary Eigenvalues semisimple

Stability

Definition

We call a control problem *stable*, if one (and therefore every) solution to $\dot{x} = Ax$ is bounded on one (and therefore every) interval (t_0, ∞) .

We call it *asymptotically stable*, if additionally $\lim_{x \rightarrow \infty} \|x(t)\| = 0$.

Lemma

asymptotically stable $\Leftrightarrow (E, A)$ regular, $\lambda E - A \prec 0$, all infinite Eigenvalues semisimple

Lemma

stable $\Leftrightarrow (E, A)$ regular, $\lambda E - A \preceq 0$, all purely imaginary Eigenvalues semisimple

Stability

Definition

We call a control problem *stable*, if one (and therefore every) solution to $\dot{x} = Ax$ is bounded on one (and therefore every) interval (t_0, ∞) .

We call it *asymptotically stable*, if additionally $\lim_{x \rightarrow \infty} \|x(t)\| = 0$.

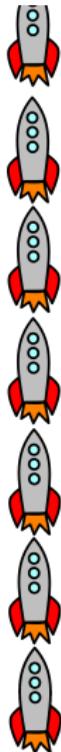
Lemma

asymptotically stable $\Leftrightarrow (E, A)$ regular, $\lambda E - A \prec 0$, all infinite Eigenvalues semisimple

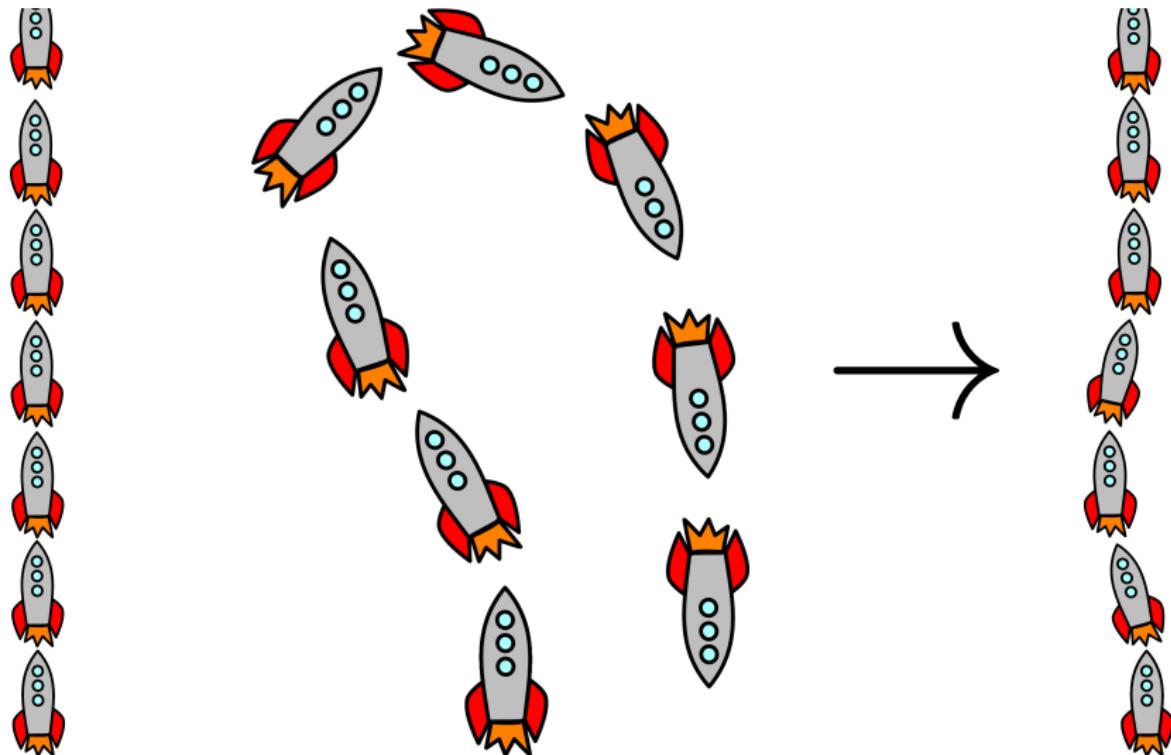
Lemma

stable $\Leftrightarrow (E, A)$ regular, $\lambda E - A \preceq 0$, all purely imaginary Eigenvalues semisimple

Stability and Stabilizability



Stability and Stabilizability



Stabilizability and Detectability

Definition

We call a control problem (asymptotically) *stabilizable*, if for each (t_0, x^0) , there exists a piecewise continuous control function $u(t)$, which is defined for all $t \geq t_0$, s.t. with this u we have $\lim_{x \rightarrow \infty} \|x(t)\| = 0$.

stabilizable

λ with $Re(\lambda) \geq 0$ Eigenvalue of A with corresponding Eigenvektor $x \neq 0$ $\Rightarrow x^T B \neq 0$

detectable

λ with $Re(\lambda) \geq 0$ Eigenvalue of A with corresponding Eigenvektor $x \neq 0$ $\Rightarrow Cx \neq 0$

Stabilizability and Detectability

Definition

We call a control problem (asymptotically) *stabilizable*, if for each (t_0, x^0) , there exists a piecewise continuous control function $u(t)$, which is defined for all $t \geq t_0$, s.t. with this u we have $\lim_{x \rightarrow \infty} \|x(t)\| = 0$.

stabilizable

λ with $Re(\lambda) \geq 0$ Eigenvalue of A with corresponding Eigenvektor $x \neq 0 \Rightarrow x^T B \neq 0$

detectable

λ with $Re(\lambda) \geq 0$ Eigenvalue of A with corresponding Eigenvektor $x \neq 0 \Rightarrow Cx \neq 0$

Stabilizability and Detectability

Definition

We call a control problem (asymptotically) *stabilizable*, if for each (t_0, x^0) , there exists a piecewise continuous control function $u(t)$, which is defined for all $t \geq t_0$, s.t. with this u we have $\lim_{x \rightarrow \infty} \|x(t)\| = 0$.

stabilizable

λ with $\operatorname{Re}(\lambda) \geq 0$ Eigenvalue of A with corresponding Eigenvektor $x \neq 0 \Rightarrow x^T B \neq 0$

detectable

λ with $\operatorname{Re}(\lambda) \geq 0$ Eigenvalue of A with corresponding Eigenvektor $x \neq 0 \Rightarrow Cx \neq 0$

Which transformations are allowed

$$\begin{aligned}\tilde{E}\dot{x} &= (\tilde{J} - \tilde{R})x + (\tilde{B} - \tilde{P})u \\ y &= (\tilde{B} + \tilde{P})^T x + (\tilde{S} + \tilde{N})u\end{aligned}$$

Which transformations are allowed

$$\begin{aligned} Q\tilde{E}\dot{x} &= Q(\tilde{J} - \tilde{R})x + Q(\tilde{B} - \tilde{P})u \\ y &= (\tilde{B} + \tilde{P})^T x + (\tilde{S} + \tilde{N})u \end{aligned}$$

Which transformations are allowed

$$\begin{aligned}\tilde{E}Q^{-1}Q\dot{x} &= (\tilde{J} - \tilde{R})Q^{-1}Qx + (\tilde{B} - \tilde{P})u \\ y &= (\tilde{B} + \tilde{P})^T Q^{-1}Qx + (\tilde{S} + \tilde{N})u\end{aligned}$$

Which transformations are allowed

$$\begin{aligned} Q\tilde{E}Q^{-1}Q\dot{x} &= Q(\tilde{J} - \tilde{R})Q^{-1}Qx + Q(\tilde{B} - \tilde{P})u \\ y &= (\tilde{B} + \tilde{P})^T Q^{-1}Qx + (\tilde{S} + \tilde{N})u \end{aligned}$$

Split the system into its different parts

[Beattie, Gugercin, Mehrmann 2019]: There exists an orthogonal basis transformation, such that the system is of the form

$$\begin{bmatrix} E_{11} & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} \dot{\hat{x}}_1 \\ \dot{\hat{x}}_2 \\ \dot{\hat{x}}_3 \\ \dot{\hat{x}}_4 \\ \dot{\hat{x}}_5 \end{bmatrix} = \begin{bmatrix} J_{11} - R_{11} & J_{12} - R_{12} & J_{13} & J_{14} & 0 \\ J_{21} - R_{21} & J_{22} - R_{22} & J_{23} & J_{24} & 0 \\ J_{31} & J_{32} & J_{33} & 0 & 0 \\ J_{41} & J_{42} & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} \hat{x}_1 \\ \hat{x}_2 \\ \hat{x}_3 \\ \hat{x}_4 \\ \hat{x}_5 \end{bmatrix}$$

$$+ \begin{bmatrix} B_1 - P_1 \\ B_2 - P_2 \\ B_3 \\ B_4 \\ B_5 \end{bmatrix} u \quad y = \begin{bmatrix} B_1 + P_1 \\ B_2 + P_2 \\ B_3 \\ B_4 \\ B_5 \end{bmatrix}^T \begin{bmatrix} \hat{x}_1 \\ \hat{x}_2 \\ \hat{x}_3 \\ \hat{x}_4 \\ \hat{x}_5 \end{bmatrix} + (S + N)u$$

Split the system into its different parts

[Beattie, Gugercin, Mehrmann 2019]: There exists an orthogonal basis transformation, such that the system is of the form

$$\begin{bmatrix} E_{11} & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} \dot{\hat{x}}_1 \\ \dot{\hat{x}}_2 \\ \dot{\hat{x}}_3 \\ \dot{\hat{x}}_4 \\ \dot{\hat{x}}_5 \end{bmatrix} = \begin{bmatrix} J_{11} - R_{11} & J_{12} - R_{12} & J_{13} & J_{14} & 0 \\ J_{21} - R_{21} & J_{22} - R_{22} & J_{23} & J_{24} & 0 \\ J_{31} & J_{32} & J_{33} & 0 & 0 \\ J_{41} & J_{42} & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} \hat{x}_1 \\ \hat{x}_2 \\ \hat{x}_3 \\ \hat{x}_4 \\ \hat{x}_5 \end{bmatrix}$$

$$+ \begin{bmatrix} B_1 - P_1 \\ B_2 - P_2 \\ B_3 \\ B_4 \\ B_5 \end{bmatrix} u \quad y = \begin{bmatrix} B_1 + P_1 \\ B_2 + P_2 \\ B_3 \\ B_4 \\ B_5 \end{bmatrix}^T \begin{bmatrix} \hat{x}_1 \\ \hat{x}_2 \\ \hat{x}_3 \\ \hat{x}_4 \\ \hat{x}_5 \end{bmatrix} + (S + N)u$$

Develop needed lemmata

Lemma

For a quadratic invertible skew-symmetric matrix A its inverse A^{-1} is again skew-symmetric.

Lemma

For a symmetric matrix B the matrix ABA^T is again symmetric.

Lemma

For a skew-symmetric matrix B the matrix ABA^T is again skew-symmetric.

Develop needed lemmata

Lemma

For a quadratic invertible skew-symmetric matrix A its inverse A^{-1} is again skew-symmetric.

Lemma

For a symmetric matrix B the matrix ABA^T is again symmetric.

Lemma

For a skew-symmetric matrix B the matrix ABA^T is again skew-symmetric.

Develop needed lemmata

Lemma

For matrices $R = R^T \succeq 0, J = -J^T \in \mathbb{C}^{n \times n}$ the inverse of the matrix $(J - R)$ can be written as

$$\hat{J} - \hat{R} := (J - R)^{-1}$$

where $\hat{R} = \hat{R}^T \succeq 0, \hat{J} = -\hat{J}^T$.

Lemma

Given two matrices $X, Y \in \mathbb{C}^{n \times m}$.

There exist matrices $A, B \in \mathbb{C}^{n \times m}$ with $X = A - B, Y = A + B$.

Develop needed lemmata

Lemma

For matrices $R = R^T \succeq 0, J = -J^T \in \mathbb{C}^{n \times n}$ the inverse of the matrix $(J - R)$ can be written as

$$\hat{J} - \hat{R} := (J - R)^{-1}$$

where $\hat{R} = \hat{R}^T \succeq 0, \hat{J} = -\hat{J}^T$.

Lemma

Given two matrices $X, Y \in \mathbb{C}^{n \times m}$.

There exist matrices $A, B \in \mathbb{C}^{n \times m}$ with $X = A - B, Y = A + B$.

Develop needed lemmata

Lemma

For matrices $R = R^T \succeq 0, J = -J^T \in \mathbb{C}^{n \times n}$ the inverse of the matrix $(J - R)$ can be written as

$$\hat{J} - \hat{R} := (J - R)^{-1}$$

where $\hat{R} = \hat{R}^T \succeq 0, \hat{J} = -\hat{J}^T$.

Lemma

Given two matrices $X, Y \in \mathbb{C}^{n \times m}$.

There exist matrices $A, B \in \mathbb{C}^{n \times m}$ with $X = A - B, Y = A + B$.

Namely these are

$$A := \frac{1}{2}(X + Y), \quad B := \frac{1}{2}(Y - X)$$

Decompose $[J_{41} \ J_{42}]$

$[J_{41} \ J_{42}] \in \mathbb{C}^{x \times (y+z)}$ has full row rank. We get two cases:

$y \geq x$

$$[J_{41} \ J_{42}] = [R_1 \ 0 \ 0] \hat{Q} \text{ with } R_1 \in \mathbb{C}^{x \times x}$$

$y < x$

$$[J_{41} \ J_{42}] = \begin{bmatrix} * & R_1 & 0 \\ R_2 & 0 & 0 \end{bmatrix} \hat{Q} \text{ with } R_2 \in \mathbb{C}^{y \times y}$$

After the transformation

In the first case we get the system

$$\begin{aligned}\dot{x}_1 &= (J_{11} - R_{11})x_1 + (B_1 - P_1)u \\ y &= (B_1 + P_1)^T x + (S + N)u\end{aligned}\tag{3}$$

In the second case we get the system

$$\begin{aligned}0 &= (J_{11} - R_{11})x_1 + (B_1 - P_1)u \\ y &= (B_1 + P_1)^T x + (S + N)u\end{aligned}\tag{4}$$

After the transformation

We have $R_{11} \succ 0$.

consistency and regularity

$(I, J_{11} - R_{11})$ is regular and $(0, J_{11} - R_{11})$ is regular \Rightarrow The system is consistent and regular.

stability

$J_{11} - R_{11} \prec 0 \Rightarrow$ The system is asymptotically stable.

We only need to look at the symmetric part

Note that $A = \frac{1}{2}(A + A^T) + \frac{1}{2}(A - A^T)$

Lemma

A quadratic matrix A is positive definite if and only if its symmetric part is positive definite.

$$A \succ 0 \Leftrightarrow \frac{1}{2}(A + A^T) \succ 0$$

The same holds for positive semidefinite, negative definite and negative semidefinite.

Stabilizable \Leftrightarrow Detectable

Let λ_0 be a purely imaginary Eigenvalue of $\tilde{J} - \tilde{R}$ with corresponding Eigenvektor x . Then x lies in the kernel of \tilde{R} .

[Mehl, Mehrmann, Wojtylak 2020]

Stabilizable \Leftrightarrow Detectable

Let λ_0 be a purely imaginary Eigenvalue of $\tilde{J} - \tilde{R}$ with corresponding Eigenvektor x . Then x lies in the kernel of \tilde{R} .
[Mehl, Mehrmann, Wojtylak 2020]

$$\begin{aligned} 0 &\leq \begin{bmatrix} x^T & y^T \end{bmatrix} \begin{bmatrix} \tilde{R} & \tilde{P} \\ \tilde{P}^T & \tilde{S} \end{bmatrix} \begin{bmatrix} x \\ y \end{bmatrix} = x^T \tilde{R} x + 2x^T \tilde{P} y + y^T \tilde{S} y \\ &= 2x^T \tilde{P} y + y^T \tilde{S} y \quad \forall \quad y \in \mathbb{C}^m \text{ and multiples of } x \end{aligned}$$

Stabilizable \Leftrightarrow Detectable

Let λ_0 be a purely imaginary Eigenvalue of $\tilde{J} - \tilde{R}$ with corresponding Eigenvektor x . Then x lies in the kernel of \tilde{R} .

[Mehl, Mehrmann, Wojtylak 2020]

$$\begin{aligned}
 0 &\leq \begin{bmatrix} x^T & y^T \end{bmatrix} \begin{bmatrix} \tilde{R} & \tilde{P} \\ \tilde{P}^T & \tilde{S} \end{bmatrix} \begin{bmatrix} x \\ y \end{bmatrix} = x^T \tilde{R} x + 2x^T \tilde{P} y + y^T \tilde{S} y \\
 &= 2x^T \tilde{P} y + y^T \tilde{S} y \quad \forall \quad y \in \mathbb{C}^m \text{ and multiples of } x \\
 \Rightarrow x^T \tilde{P} &= 0 \quad \wedge \quad \tilde{P}^T x = 0
 \end{aligned}$$

Stabilizable \Leftrightarrow Detectable

Let λ_0 be a purely imaginary Eigenvalue of $\tilde{J} - \tilde{R}$ with corresponding Eigenvektor x . Then x lies in the kernel of \tilde{R} .

[Mehl, Mehrmann, Wojtylak 2020]

$$\begin{aligned} 0 &\leq \begin{bmatrix} x^T & y^T \end{bmatrix} \begin{bmatrix} \tilde{R} & \tilde{P} \\ \tilde{P}^T & \tilde{S} \end{bmatrix} \begin{bmatrix} x \\ y \end{bmatrix} = x^T \tilde{R} x + 2x^T \tilde{P} y + y^T \tilde{S} y \\ &= 2x^T \tilde{P} y + y^T \tilde{S} y \quad \forall \quad y \in \mathbb{C}^m \text{ and multiples of } x \\ \Rightarrow x^T \tilde{P} &= 0 \quad \wedge \quad \tilde{P}^T x = 0 \end{aligned}$$

$$x^T (\tilde{B} - \tilde{P}) = x^T \tilde{B} = (\tilde{B}^T x)^T = ((\tilde{B} + \tilde{P})^T x)^T$$

Controllability

C0 $\text{rank}[\alpha E - \beta A, B] = n \quad \forall \quad (\alpha, \beta) \in \mathbb{C}^2$

C1 $\text{rank}[\lambda E - A, B] = n \quad \forall \quad \lambda \in \mathbb{C}$

C2 $\text{rank}[E, AS_\infty(E), B] = n$

Controllability

C0 $\text{rank}[\alpha E - \beta A, B] = n \quad \forall \quad (\alpha, \beta) \in \mathbb{C}^2$

C1 $\text{rank}[\lambda E - A, B] = n \quad \forall \quad \lambda \in \mathbb{C}$

C2 $\text{rank}[E, AS_\infty(E), B] = n$

We call a problem *completely controllable* if C0 holds.

We call a problem *strongly controllable* if C1 and C2 hold.

C0 \Rightarrow C1, C2

C0 \Leftrightarrow C1 and $\text{rank} \begin{bmatrix} E & B \end{bmatrix} = n$

Controllability

$$C0 \ rank[\alpha E - \beta A, B] = n \quad \forall \quad (\alpha, \beta) \in \mathbb{C}^2$$

$$C1 \ rank[\lambda E - A, B] = n \quad \forall \quad \lambda \in \mathbb{C}$$

$$C2 \ rank[E, AS_\infty(E), B] = n$$

We call a problem *completely controllable* if $C0$ holds.

We call a problem *strongly controllable* if $C1$ and $C2$ hold.

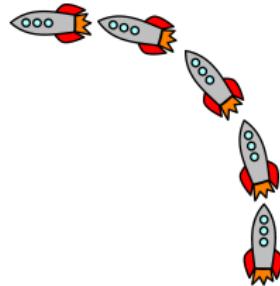
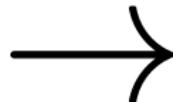
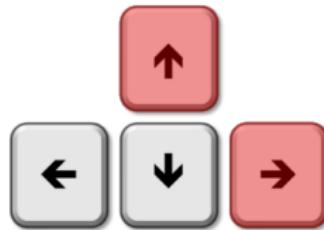
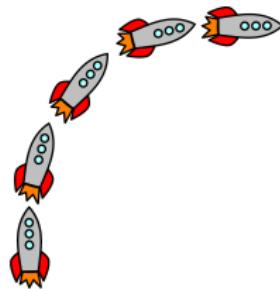
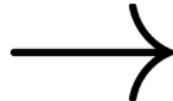
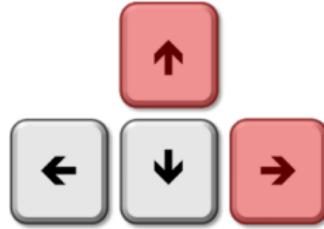
$$C0 \Rightarrow C1, C2$$

$$C0 \Leftrightarrow C1 \text{ and } rank \begin{bmatrix} E & B \end{bmatrix} = n$$

$$C1 \Leftrightarrow \begin{array}{l} \lambda \in \mathbb{C} \text{ Eigenvalue of } A \text{ with corresponding} \\ \text{Eigenvektor } x \neq 0 \end{array} \Rightarrow x^T B \neq 0$$

Observability

Take output $y \hat{=} \text{height above the ground}$



Observability

O0 $\text{rank} \begin{bmatrix} \alpha E - \beta A \\ C \end{bmatrix} = n \quad \forall \quad (\alpha, \beta) \in \mathbb{C}^2$

O1 $\text{rank} \begin{bmatrix} \lambda E - A \\ C \end{bmatrix} = n \quad \forall \quad \lambda \in \mathbb{C}$

O2 $\text{rank} \begin{bmatrix} E \\ T_\infty(E)^T A \\ C \end{bmatrix} = n$

Observability

$$O0 \text{ rank} \begin{bmatrix} \alpha E - \beta A \\ C \end{bmatrix} = n \quad \forall \quad (\alpha, \beta) \in \mathbb{C}^2$$

$$O1 \text{ rank} \begin{bmatrix} \lambda E - A \\ C \end{bmatrix} = n \quad \forall \quad \lambda \in \mathbb{C}$$

$$O2 \text{ rank} \begin{bmatrix} E \\ T_\infty(E)^T A \\ C \end{bmatrix} = n$$

We call a problem *completely observable* if $O0$ holds.

We call a problem *strongly observable* if $O1$ and $O2$ hold.

$O0 \Rightarrow O1, O2$

$$O0 \Leftrightarrow O1 \text{ and rank} \begin{bmatrix} E \\ C \end{bmatrix} = n$$

Observability

$$O0 \text{ rank} \begin{bmatrix} \alpha E - \beta A \\ C \end{bmatrix} = n \quad \forall \quad (\alpha, \beta) \in \mathbb{C}^2$$

$$O1 \text{ rank} \begin{bmatrix} \lambda E - A \\ C \end{bmatrix} = n \quad \forall \quad \lambda \in \mathbb{C}$$

$$O2 \text{ rank} \begin{bmatrix} E \\ T_\infty(E)^T A \\ C \end{bmatrix} = n$$

We call a problem *completely observable* if $O0$ holds.

We call a problem *strongly observable* if $O1$ and $O2$ hold.

$O0 \Rightarrow O1, O2$

$$O0 \Leftrightarrow O1 \text{ and rank} \begin{bmatrix} E \\ C \end{bmatrix} = n$$

$O1 \Leftrightarrow \lambda \in \mathbb{C}$ Eigenvalue of A with corresponding
Eigenvektor $x \neq 0 \Rightarrow Cx \neq 0$

Transform the system to remove S+N

$$\begin{aligned} E\dot{x} &= (\tilde{J} - \tilde{R})x + (\tilde{B} + \tilde{P})u & \Rightarrow & \hat{E}\dot{x} = (\hat{J} - \hat{R})x + (\hat{B} + \hat{P})u \\ y &= (\tilde{B} - \tilde{P})^T x + (\tilde{S} + \tilde{N})u & & y = (\hat{B} - \hat{P})^T x \end{aligned}$$

Transform the system to remove S+N

$$\begin{aligned} E\dot{x} &= (\tilde{J} - \tilde{R})x + (\tilde{B} + \tilde{P})u & \Rightarrow & \hat{E}\dot{x} = (\hat{J} - \hat{R})x + (\hat{B} + \hat{P})u \\ y &= (\tilde{B} - \tilde{P})^T x + (\tilde{S} + \tilde{N})u & & y = (\hat{B} - \hat{P})^T x \end{aligned}$$

Decompose $S + N = XY$ and use the transferfunction

$$\begin{aligned} H(s) &= (\tilde{B} + \tilde{P})^T (s\tilde{E} - (\tilde{J} - \tilde{R}))^{-1} (\tilde{B} - \tilde{P}) + (S + N) \\ &= [(B + P)^T \quad X] \begin{bmatrix} s\tilde{E} - (\tilde{J} - \tilde{R}) & 0 \\ 0 & I \end{bmatrix}^{-1} \begin{bmatrix} (B - P) \\ Y \end{bmatrix} \end{aligned}$$

Transform the system to remove S+N

$$\begin{aligned} E\dot{x} &= (\tilde{J} - \tilde{R})x + (\tilde{B} + \tilde{P})u & \Rightarrow & \hat{E}\dot{x} = (\hat{J} - \hat{R})x + (\hat{B} + \hat{P})u \\ y &= (\tilde{B} - \tilde{P})^T x + (\tilde{S} + \tilde{N})u & y &= (\hat{B} - \hat{P})^T x \end{aligned}$$

Decompose $S + N = XY$ and use the transferfunction

$$\begin{aligned} H(s) &= (\tilde{B} + \tilde{P})^T (s\tilde{E} - (\tilde{J} - \tilde{R}))^{-1} (\tilde{B} - \tilde{P}) + (S + N) \\ &= \begin{bmatrix} (B + P)^T & X \end{bmatrix} \begin{bmatrix} s\tilde{E} - (\tilde{J} - \tilde{R}) & 0 \\ 0 & I \end{bmatrix}^{-1} \begin{bmatrix} (B - P) \\ Y \end{bmatrix} \end{aligned}$$

$$\begin{aligned} \hat{B} &:= \begin{bmatrix} \tilde{B} \\ \frac{1}{2}(X + Y) \end{bmatrix}, \hat{P} := \begin{bmatrix} \tilde{P} \\ \frac{1}{2}(X - Y) \end{bmatrix}, \hat{E} := \begin{bmatrix} \tilde{E} & 0 \\ 0 & 0 \end{bmatrix} \\ \hat{J} &:= \begin{bmatrix} \tilde{J} & 0 \\ 0 & 0 \end{bmatrix}, \hat{R} := \begin{bmatrix} \tilde{R} & 0 \\ 0 & I \end{bmatrix} \end{aligned}$$

Transform the system to remove P

Lemma

The system

$$\begin{aligned}E\dot{x} &= (J - R)x + (B + P)u \\y &= (B - P)^T x\end{aligned}$$

can be reformulated to

$$\begin{aligned}\hat{E}\dot{x} &= (\hat{J} - \hat{R})x + \hat{B}u \\y &= \hat{B}^T x\end{aligned}$$

Controllable \Leftrightarrow Observable

controllable

$\lambda \in \mathbb{C}$ Eigenvalue of $\tilde{J} - \tilde{R}$ with corresponding Eigenvektor $x \neq 0$ $\Rightarrow x^T(\tilde{B} - \tilde{P}) \neq 0$

observable

$\lambda \in \mathbb{C}$ Eigenvalue of $\tilde{J} - \tilde{R}$ with corresponding Eigenvektor $x \neq 0$ $\Rightarrow (\tilde{B} + \tilde{P})^T x \neq 0$

Controllable \Leftrightarrow Observable

controllable

$\lambda \in \mathbb{C}$ Eigenvalue of $\tilde{J} - \tilde{R}$ with corresponding
Eigenvektor $x \neq 0 \Rightarrow x^T \tilde{B} \neq 0$

observable

$\lambda \in \mathbb{C}$ Eigenvalue of $\tilde{J} - \tilde{R}$ with corresponding
Eigenvektor $x \neq 0 \Rightarrow \tilde{B}^T x \neq 0$

What to do if R is not positive semidefinite?

$$\begin{aligned}\dot{x} &= (\tilde{J} - \tilde{R})x + \tilde{B}u \\ y &= \tilde{B}^T x\end{aligned}$$

Schurform

$$\begin{bmatrix} R_{11} & 0 \\ 0 & 0 \end{bmatrix} = URU^T \text{ where } R_{11} \in \mathbb{R}^{I,I}$$
$$\hat{B} := U\tilde{B}$$

What to do if R is not positive semidefinite?

Apply output feedback $u = Ky$, $K = K^T$ blockdiagonal

$$\dot{x} = (\tilde{J} - \tilde{R})x + \tilde{B}u = (\tilde{J} - \tilde{R})x + \tilde{B}K\tilde{B}^T x = (\tilde{J} - \tilde{R} + \tilde{B}K\tilde{B}^T)x$$

What we want

$$\begin{aligned} 0 &\stackrel{!}{\prec} U(\tilde{R} - \tilde{B}K\tilde{B}^T)U^T \\ &= \begin{bmatrix} R_{11} - \hat{B}_{11}K_{11}\hat{B}_{11}^T - \hat{B}_{12}K_{22}\hat{B}_{12}^T & -\hat{B}_{11}K_{11}\hat{B}_{21}^T - \hat{B}_{12}K_{22}\hat{B}_{21}^T \\ -\hat{B}_{21}K_{11}\hat{B}_{11}^T - \hat{B}_{22}K_{22}\hat{B}_{12}^T & -\hat{B}_{21}K_{11}\hat{B}_{21}^T - \hat{B}_{22}K_{22}\hat{B}_{22}^T \end{bmatrix} \\ &:= \begin{bmatrix} X_{11} & X_{12} \\ X_{21} & X_{22} \end{bmatrix} \\ \Leftrightarrow 0 &\stackrel{!}{\prec} X_{11} \quad \wedge \quad 0 \stackrel{!}{\prec} X_{22} - X_{21}X_{11}^{-1}X_{12} \end{aligned}$$

What to do if R is not positive semidefinite?

Apply outputfeedback $u = Ky, K = K^T$ blockdiagonal

$$\dot{x} = (\tilde{J} - \tilde{R})x + \tilde{B}u = (\tilde{J} - \tilde{R})x + \tilde{B}K\tilde{B}^T x = (\tilde{J} - \tilde{R} + \tilde{B}K\tilde{B}^T)x$$

What we want

$$\begin{aligned} 0 &\stackrel{!}{\prec} U(\tilde{R} - \tilde{B}K\tilde{B}^T)U^T \\ &= \begin{bmatrix} R_{11} - \hat{B}_{11}K_{11}\hat{B}_{11}^T - \hat{B}_{12}K_{22}\hat{B}_{12}^T & -\hat{B}_{11}K_{11}\hat{B}_{21}^T - \hat{B}_{12}K_{22}\hat{B}_{22}^T \\ -\hat{B}_{21}K_{11}\hat{B}_{11}^T - \hat{B}_{22}K_{22}\hat{B}_{12}^T & -\hat{B}_{21}K_{11}\hat{B}_{21}^T - \hat{B}_{22}K_{22}\hat{B}_{22}^T \end{bmatrix} \\ &:= \begin{bmatrix} X_{11} & X_{12} \\ X_{21} & X_{22} \end{bmatrix} \\ \Leftrightarrow 0 &\stackrel{!}{\prec} X_{11} \quad \wedge \quad 0 \stackrel{!}{\prec} X_{22} - X_{21}X_{11}^{-1}X_{12} \end{aligned}$$

What to do if R is not positive semidefinite?

Apply outputfeedback $u = Ky, K = K^T$ blockdiagonal

$$\dot{x} = (\tilde{J} - \tilde{R})x + \tilde{B}u = (\tilde{J} - \tilde{R})x + \tilde{B}K\tilde{B}^T x = (\tilde{J} - \tilde{R} + \tilde{B}K\tilde{B}^T)x$$

What we want

$$\begin{aligned}
 0 &\stackrel{!}{\prec} U(\tilde{R} - \tilde{B}K\tilde{B}^T)U^T \\
 &= \begin{bmatrix} R_{11} - \hat{B}_{11}K_{11}\hat{B}_{11}^T - \hat{B}_{12}K_{22}\hat{B}_{12}^T & -\hat{B}_{11}K_{11}\hat{B}_{21}^T - \hat{B}_{12}K_{22}\hat{B}_{22}^T \\ -\hat{B}_{21}K_{11}\hat{B}_{11}^T - \hat{B}_{22}K_{22}\hat{B}_{12}^T & -\hat{B}_{21}K_{11}\hat{B}_{21}^T - \hat{B}_{22}K_{22}\hat{B}_{22}^T \end{bmatrix} \\
 &:= \begin{bmatrix} X_{11} & X_{12} \\ X_{21} & X_{22} \end{bmatrix} \\
 \Leftrightarrow 0 &\stackrel{!}{\prec} X_{11} \quad \wedge \quad 0 \stackrel{!}{\prec} X_{22} - X_{21}X_{11}^{-1}X_{12}
 \end{aligned}$$

Outline

1 Port-Hamiltonian systems and their properties

2 Optimal control

Cost functional - state feedback

$$\mathcal{S}(x(t), u(t))$$

$$\begin{aligned} &= \frac{1}{2} \left(x(t_f)^T M x(t_f) + \int_{t_0}^{t_f} \begin{bmatrix} x(t)^T & u(t)^T \end{bmatrix} \begin{bmatrix} Q & S \\ S^T & R \end{bmatrix} \begin{bmatrix} x(t) \\ u(t) \end{bmatrix} dt \right) \\ &= \frac{1}{2} \left(x(t_f)^T M x(t_f) \right. \\ &\quad \left. + \int_{t_0}^{t_f} x(t)^T Q x(t) + u(t)^T R u(t) + x(t)^T S u(t) + u(t)^T S^T x(t) dt \right) \end{aligned} \tag{5}$$

$$t_0 < t_f \leq \infty, M = M^T, Q = Q^T, R = R^T$$

KKT-System

$$\begin{bmatrix} 0 & A & B \\ A^T & Q & S \\ B^T & S^T & R \end{bmatrix} \begin{bmatrix} \mu \\ x \\ u \end{bmatrix} = \begin{bmatrix} 0 & I & 0 \\ -I & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} \dot{\mu} \\ \dot{x} \\ \dot{u} \end{bmatrix} \quad (6)$$
$$x(t_0) = x^0, \mu(t_f) = Mx(t_f)$$

Hamiltonian Matrix

Solve the third line after u and insert into the other two lines

$$\begin{bmatrix} F & G \\ H & -F^T \end{bmatrix} := \begin{bmatrix} A - BR^{-1}S^T & -BR^{-1}B^T \\ -(Q - SR^{-1}S^T) & -A^T + SR^{-1}B^T \end{bmatrix} \begin{bmatrix} x \\ \mu \end{bmatrix} = \begin{bmatrix} \dot{x} \\ \dot{\mu} \end{bmatrix} \quad (7)$$

$$x(t_0) = x^0, \mu(t_f) = Mx(t_f)$$

Hamiltonian Matrix

Solve the third line after u and insert into the other two lines

$$\begin{bmatrix} F & G \\ H & -F^T \end{bmatrix} := \begin{bmatrix} A - BR^{-1}S^T & -BR^{-1}B^T \\ -(Q - SR^{-1}S^T) & -A^T + SR^{-1}B^T \end{bmatrix} \begin{bmatrix} x \\ \mu \end{bmatrix} = \begin{bmatrix} \dot{x} \\ \dot{\mu} \end{bmatrix} \quad (7)$$

$$x(t_0) = x^0, \mu(t_f) = Mx(t_f)$$

Ansatz: $\mu(t) = X(t)x(t) \Rightarrow X(t_f) = M$

$$\Rightarrow \dot{\mu} = X(t)\dot{x} + \dot{X}(t)x$$

Hamiltonian Matrix

Solve the third line after u and insert into the other two lines

$$\begin{bmatrix} F & G \\ H & -F^T \end{bmatrix} := \begin{bmatrix} A - BR^{-1}S^T & -BR^{-1}B^T \\ -(Q - SR^{-1}S^T) & -A^T + SR^{-1}B^T \end{bmatrix} \begin{bmatrix} x \\ \mu \end{bmatrix} = \begin{bmatrix} \dot{x} \\ \dot{\mu} \end{bmatrix} \quad (7)$$

$$x(t_0) = x^0, \mu(t_f) = Mx(t_f)$$

Ansatz: $\mu(t) = X(t)x(t) \Rightarrow X(t_f) = M$

$$\Rightarrow \dot{\mu} = X(t)\dot{x} + \dot{X}(t)x$$

$$\Rightarrow (\dot{X}(t) - H + F^T X(t) + X(t)F + X(t)G X(t))x = 0$$

Riccati Equations

Differential equation

$$\begin{aligned}\dot{X}(t) &= H - F^T X(t) - X(t)F - X(t)G X(t) \\ X(t_f) &= M\end{aligned}\tag{8}$$

Algebraic equation $t_f = \infty, M = 0, (A, B)$ stabilizable

$$0 = H - F^T X - XF - XGX\tag{9}$$

State feedback

$$\begin{aligned}u(t) &= -R^{-1}(S^T + B^T X(t))x(t) \\ u(t) &= -R^{-1}(S^T + B^T X)x(t)\end{aligned}$$

Cost functional - output feedback

$$\mathcal{S}(y(t), u(t))$$

$$\begin{aligned}
 &= \frac{1}{2} \left(y(t_f)^T M y(t_f) + \int_{t_0}^{t_f} \begin{bmatrix} y(t)^T & u(t)^T \end{bmatrix} \begin{bmatrix} Q & S \\ S^T & R \end{bmatrix} \begin{bmatrix} y(t) \\ u(t) \end{bmatrix} dt \right) \\
 &= \frac{1}{2} \left(y(t_f)^T M y(t_f) \right. \\
 &\quad + \int_{t_0}^{t_f} y(t)^T B(t) Q B(t)^T y(t) + u(t)^T R u(t) + y(t)^T B(t) S u(t) \\
 &\quad \left. + u(t)^T S^T B(t)^T y(t) dt \right)
 \end{aligned}$$

$$t_0 < t_f \leq \infty$$

Insert y

$$\hat{Q} := (\tilde{B} + \tilde{P})Q(\tilde{B} + \tilde{P})^T$$

$$\hat{R} := (\tilde{S} + \tilde{N})^T Q(\tilde{S} + \tilde{N}) + S^T(\tilde{S} + \tilde{N}) + (\tilde{S} + \tilde{N})^T S + R$$

$$\hat{S} := (\tilde{B} + \tilde{P})Q(\tilde{S} + \tilde{N}) + (\tilde{B} + \tilde{P})S$$

We now get the cost functional

$$\mathcal{S}(y(t), u(t))$$

$$= \frac{1}{2} \left(y(t_f)^T M y(t_f) + \int_{t_0}^{t_f} \begin{bmatrix} x(t)^T & u(t)^T \end{bmatrix} \begin{bmatrix} \hat{Q} & \hat{S} \\ \hat{S}^T & \hat{R} \end{bmatrix} \begin{bmatrix} x(t) \\ u(t) \end{bmatrix} dt \right)$$

$$= \frac{1}{2} \left(y(t_f)^T M y(t_f) \right.$$

$$\left. + \int_{t_0}^{t_f} x(t)^T \hat{Q} x(t) + u(t)^T \hat{R} u(t) + x(t)^T \hat{S} u(t) + u(t)^T \hat{S}^T x(t) dt \right)$$

$$t_0 < t_f \leq \infty$$

Ideas to improve finding optimal control via pH structure

Insert pH matrices and see if something cancels out

$$0 = \textcolor{green}{H} - \textcolor{red}{F}^T X - X \textcolor{blue}{F} - X \textcolor{cyan}{G} X$$

Ideas to improve finding optimal control via pH structure

Insert pH matrices and see if something cancels out

$$0 = \textcolor{red}{H} - \textcolor{blue}{F}^T X - X \textcolor{red}{F} - X \textcolor{blue}{G} X$$

$$\begin{aligned} &= [-Q + SR^{-1}S^T] - [A^T - SR^{-1}B^T]X \\ &\quad - X[A - BR^{-1}S^T] - X[-BR^{-1}B^T]X \end{aligned}$$

Ideas to improve finding optimal control via pH structure

Insert pH matrices and see if something cancels out

$$0 = H - F^T X - X F - X G X$$

$$= [-Q + SR^{-1}S^T] - [A^T - SR^{-1}B^T]X \\ - X[A - BR^{-1}S^T] - X[-BR^{-1}B^T]X$$

$$= [-Q + SR^{-1}S^T] - [(\tilde{J} - \tilde{R})^T - SR^{-1}(\tilde{B} - \tilde{P})^T]X \\ - X[(\tilde{J} - \tilde{R}) - (\tilde{B} - \tilde{P})R^{-1}S^T] - X[-(\tilde{B} - \tilde{P})R^{-1}(\tilde{B} - \tilde{P})^T]X$$

Ideas to improve finding optimal control via pH structure

Transform the system with Z

$$Z := \begin{bmatrix} \frac{1}{\sqrt{2}}I & \frac{1}{\sqrt{2}}I & 0 \\ -\frac{1}{\sqrt{2}}I & \frac{1}{\sqrt{2}}I & 0 \\ 0 & 0 & \frac{1}{\sqrt{2}}I \end{bmatrix}$$

$$Z^T \begin{bmatrix} 0 & A & B \\ A^T & Q & S \\ B^T & S^T & R \end{bmatrix} Z Z^T \begin{bmatrix} \mu \\ x \\ u \end{bmatrix} = Z^T \begin{bmatrix} 0 & I & 0 \\ -I & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} Z Z^T \begin{bmatrix} \dot{\mu} \\ \dot{x} \\ \dot{u} \end{bmatrix}$$

$$x(t_0) = x^0, \mu(t_f) = Mx(t_f)$$

Ideas to improve finding optimal control via pH structure

Transform the system with Z

$$0 = \textcolor{red}{H} - \textcolor{blue}{F}^T X - X \textcolor{red}{F} - X \textcolor{blue}{G} X$$

Ideas to improve finding optimal control via pH structure

Transform the system with Z

$$\begin{aligned}0 &= \textcolor{green}{H} - \textcolor{red}{F}^T X - X \textcolor{blue}{F} - X \textcolor{teal}{G} X \\&= [\textcolor{green}{A}^T + \textcolor{green}{A} - \textcolor{green}{Q} + (-\textcolor{red}{B} + \textcolor{red}{S})R^{-1}(-\textcolor{red}{B}^T + \textcolor{red}{S}^T)] \\&\quad - [\textcolor{red}{A}^T - \textcolor{red}{A} + \textcolor{red}{Q} - (-\textcolor{red}{B} + \textcolor{red}{S})R^{-1}(\textcolor{red}{B}^T + \textcolor{red}{S}^T)]X \\&\quad - X[-\textcolor{blue}{A}^T + \textcolor{blue}{A} + \textcolor{blue}{Q} - (\textcolor{blue}{B} - \textcolor{blue}{S})R^{-1}(-\textcolor{blue}{B}^T + \textcolor{blue}{S}^T)] \\&\quad - X[\textcolor{teal}{A}^T + \textcolor{teal}{A} + \textcolor{teal}{Q} - (\textcolor{teal}{B} + \textcolor{teal}{S})R^{-1}(\textcolor{teal}{B}^T + \textcolor{teal}{S}^T)]X\end{aligned}$$

Ideas to improve finding optimal control via pH structure

Transform the system with Z

$$\begin{aligned} 0 &= \textcolor{red}{H} - \textcolor{red}{F}^T X - X \textcolor{blue}{F} - X \textcolor{cyan}{G} X \\ &= [-2\tilde{R} - Q + (-B + S)R^{-1}(-B^T + S^T)] \\ &\quad - [-2\tilde{J} + Q - (-B + S)R^{-1}(B^T + S^T)]X \\ &\quad - X[2\tilde{J} + Q - (B - S)R^{-1}(-B^T + S^T)] \\ &\quad - X[-2\tilde{R} + Q - (B + S)R^{-1}(B^T + S^T)]X \end{aligned}$$

Ideas to improve finding optimal control via pH structure

Choose Q, S, R as $\tilde{R}, \tilde{P}, \tilde{S}$

$$\begin{aligned}
 0 &= \textcolor{green}{H} - \textcolor{red}{F}^T X - X \textcolor{blue}{F} - X \textcolor{teal}{G} X \\
 &= [\tilde{R} + \tilde{P} \tilde{S}^{-1} \tilde{P}^T] \\
 &\quad - [(\tilde{J} - \tilde{R})^T - \tilde{P} \tilde{S}^{-1} (\tilde{B} - \tilde{P})^T] X \\
 &\quad - X [(\tilde{J} - \tilde{R}) - (\tilde{B} - \tilde{P}) \tilde{S}^{-1} \tilde{P}^T] \\
 &\quad + X [(\tilde{B} - \tilde{P}) \tilde{S}^{-1} (\tilde{B} - \tilde{P})^T] X
 \end{aligned}$$

Ideas to improve finding optimal control via pH structure

Choose Q, S, R as $\tilde{R}, \tilde{P}, \tilde{S}$

$$0 = \textcolor{red}{H} - \textcolor{blue}{F}^T X - X \textcolor{blue}{F} - X \textcolor{red}{G} X$$

$$= -\tilde{R} - (\tilde{J} - \tilde{R})^T X - X(\tilde{J} - \tilde{R}) + X \tilde{B} \tilde{S}^{-1} \tilde{B} X$$

Conclusions and Outlook

- pH structure helps us with stabilizing control systems
- We get detectability for free
- If a pH system is controllable, we get observability for free
- Not too much profit from the pH structure regarding optimal control
- Can we improve the algorithms with pH structure?

Thanks

Thank you very much for your
attention

References

- Beattie, Gugercin, Mehrmann (2019) "Structure-preserving Interpolatory Model Reduction for Port-Hamiltonian Differential-Algebraic Systems" In: arXiv:1910.05674
- Mehl, Mehrmann, Wojtylak (2020) "Distance problems for dissipative Hamiltonian systems and related matrix polynomials" In: Linear Algebra and its Applications
- Bunse-Gerstner, Byers, Mehrmann, Nichols (1999) "Feedback Design for Regularizing Descriptor Systems" In: Linear Algebra and its Applications, volume 299, number 1, pages 119-151
- Beattie, Mehrmann, Xu, Zwart (2018) "Linear port-Hamiltonian descriptor systems" In: Mathematics of Control, Signals, and Systems, Volume 30, Article 17