

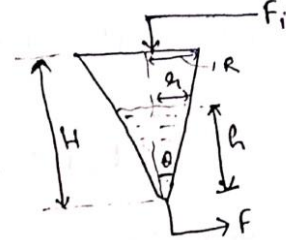
## Assignment 2 Solution

1.

**Solution:**

$$\tan \theta = \frac{r}{h} = \frac{R}{H}$$

$$r = \frac{Rh}{H}$$



Apply material balance over a canonical tank

Accumulation = in-out

$$\frac{dV}{dt} = F_{in} - F \quad \left\{ \text{at any time, area } A = \pi r^2 = \pi \left( \frac{Rh}{H} \right)^2 \right\}$$

$$\frac{dh}{dt} = \frac{F_{in}}{A} - \frac{F}{A}$$

$$\frac{dh}{dt} = \frac{F_{in} H^2}{\pi R^2 h^2} - \frac{F H^2}{\pi R^2 h^2} = \frac{\alpha F_{in}}{h^2} - \frac{\alpha F}{h^2}$$

$$\frac{dh}{dt} = \frac{\alpha F_{in}}{h^2} - \frac{A_0 \alpha k h}{h^2}$$

$$\frac{dh}{dt} = \frac{\alpha F_{in}}{h^2} - \beta h^{-1}$$

Now linearize the above equation,

$$\frac{dh}{dt} = \frac{\alpha F_i}{h^2} - \frac{\beta}{h} = F(h)$$

$$\left[ \frac{\partial F}{\partial h} \right]_{h=h_s} = -\frac{2\alpha F_i}{h_s^3} + \frac{\beta}{h_s^2} = A_1$$

$$\left[ \frac{\partial F}{\partial F_i} \right]_{h=h_s} = \frac{\alpha}{h_s^2} = A_2$$

$$\frac{dh}{dt} = \left( \frac{\partial F}{\partial h} \right)_{h=h_s} (h - h_s) + \left( \frac{\partial F}{\partial F_i} \right)_{F=F_{is}} (F_i - F_{is})$$

$$h' = h - h_s$$

$$F_i' = F_i - F_{is}$$

$$\frac{dh'}{dt} = A_1 h' + A_2 F_i'$$

$$sH(s) = A_1 H(s) + A_2 F_i(s)$$

$$\frac{H(s)}{F_i(s)} = \frac{A_2}{s - A_1}$$

$$\text{Time constant} = \frac{1}{A_1}$$

$$\text{Gain} = \frac{A_2}{A_1}$$

2. For given differential eqn.

$$\frac{d^2 y}{dt^2} + 5 \frac{dy}{dt} + 6y = f(t)$$

a) For unit step

$$\frac{d^2 y}{dt^2} + 5 \frac{dy}{dt} + 6y = 1$$

Solution of this equation will be  $y = y_c + y_p$

CF is

$$m^2 + 5m + 6 = 0$$

$$m = -2, -3$$

$$y_c = c_1 e^{-2t} + c_2 e^{-3t}$$

Now applying the BC

$$y(0) = 0, \left( \frac{dy}{dt} \right)_{(0)} = 0$$

We get

$$c_1 = \frac{-1}{2}$$

$$c_2 = \frac{1}{3}$$

PI is

$$0 + 5(0) + 6y_p = 1 \Rightarrow y_p = \frac{1}{6}$$

The final solution is

$$y = \frac{-1}{2} e^{-2t} + \frac{1}{3} e^{-3t} + \frac{1}{6}$$

b) For Ramp input

$$\frac{d^2 y}{dt^2} + 5 \frac{dy}{dt} + 6y = t$$

Solution of this equation will be  $y = y_c + y_p$

CF is

$$m^2 + 5m + 6 = 0$$

$$m = -2, -3$$

$$y_c = c_1 e^{-2t} + c_2 e^{-3t}$$

Now applying the BC

$$y(0) = 0, \left( \frac{dy}{dt} \right)_{(0)} = 0$$

We get

$$c_1 = \frac{-1}{2}$$

$$c_2 = \frac{1}{3}$$

PI is

$$y_p = (D+2)^{-1} (D+3)^{-1} (t)$$

$$y_p = \frac{1}{6} \left( 1 - \frac{D}{2} \right) \left( 1 - \frac{D}{3} \right) (t)$$

$$y_p = \frac{1}{6} \left[ 1 - \frac{5D}{6} + \frac{D^2}{6} \right] (t)$$

$$y_p = \frac{1}{6} \left[ t - \frac{5}{6} \right]$$

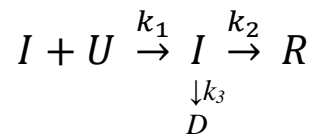
The final solution is

$$y = \frac{-1}{2} e^{-2t} + \frac{1}{3} e^{-3t} + \frac{1}{6} \left[ t - \frac{5}{6} \right]$$

3. Solve according to question 2(a).

4.

**Solution:**



$$\frac{dC_I}{dt} = -k_1 C_I C_U - k_2 C_I - k_3 C_I$$

$$\frac{dC_I}{dt} = -k_1 C_I C_U - (k_2 + k_3) C_I$$

Linearising  $C_I C_U$  about its steady state,  $\bar{C}_I$  and  $\bar{C}_U$

$$C_I C_U = \bar{C}_I \bar{C}_U + \bar{C}_I (C_U - \bar{C}_U) + \bar{C}_U (C_I - \bar{C}_I)$$

$X - \bar{X} = \hat{X}$ , deviation variable

$$C_I C_U = \bar{C}_I \bar{C}_U + \bar{C}_I \widehat{C}_U + \bar{C}_U \widehat{C}_I$$

$$\frac{dC_I}{dt} + (k_2 + k_3) C_I = k_1 \bar{C}_I \bar{C}_U + k_1 \bar{C}_I \widehat{C}_U + k_1 \bar{C}_U \widehat{C}_I$$

Subtract  $(k_2 + k_3) \bar{C}_I$  from both sides,

$$\text{LHS} = \frac{dC_I}{dt} + (k_2 + k_3) (C_I - \bar{C}_I)$$

$$\text{RHS} = k_1 \bar{C}_I \bar{C}_U - (k_2 + k_3) \bar{C}_I + k_1 \bar{C}_I \widehat{C}_U + k_1 \bar{C}_U \widehat{C}_I$$

We have,  $\frac{d\widehat{C_I}}{dt} + (k_2 + k_3)\widehat{C_I} = k_1\overline{C_I}\widehat{C_U} + k_1\overline{C_U}\widehat{C_I}$

$$\frac{d\widehat{C_I}}{dt} + (k_2 + k_3 - k_1\overline{C_U})\widehat{C_I} = k_1\overline{C_I}\widehat{C_U}$$

$$\left(\frac{1}{k_2 + k_3 - k_1\overline{C_U}}\right)\frac{d\widehat{C_I}}{dt} + \widehat{C_I} = \left(\frac{k_1\overline{C_I}}{k_2 + k_3 - k_1\overline{C_U}}\right)\widehat{C_U}$$

$$\text{Solution, } \widehat{C_I} = \widehat{C_U} \left(\frac{k_1\overline{C_I}}{k_2 + k_3 - k_1\overline{C_U}}\right)(1 - e^{(k_2 + k_3 - k_1\overline{C_U})(-t)})$$

5.

**Solution:**

$$A \rightarrow R \rightarrow S$$

Now rate equation

$$r_A = \frac{dC_A}{dt} = -k_1 C_A$$

$$\text{This implies } C_A = C_{A0} e^{-k_1 t} \quad (1)$$

$$r_R = \frac{dC_R}{dt} = k_1 C_A - k_2 C_R \quad (2)$$

From equation 2

$$\frac{dC_R}{dt} = k_1 C_A - k_2 C_R$$

$$\frac{dC_R}{dt} + k_2 C_R = k_1 C_A$$

This is of the form of linear equation

$$\frac{dy}{dx} + py = Q$$

So we can solve it by using an integration factor method

$$IF = e^{\int k_2 dt}$$

$$\text{This implies } C_R e^{k_2 t} = \int k_1 C_A e^{k_2 t} + c$$

On solving the above equation with BC that at  $t=0$   $C_{R0} = 0$

$$C_R = \frac{k_1 C_{A0}}{(k_2 - k_1)} (e^{-k_1 t} - e^{-k_2 t}) \quad (3)$$

For maximum value of R we need to differentiate equation 3 from where we get the for maximum value of  $C_R$

$$\frac{dC_R}{dt} = \frac{k_1 C_{A0}}{(k_2 - k_1)} (-k_1 e^{-k_1 t} + k_2 e^{-k_2 t}) = 0$$

$$\text{On solving we get } t = \frac{\ln \frac{k_2}{k_1}}{(k_2 - k_1)}$$

This implies  $t = \frac{1}{k_{logmean}}$

6.

**Solution:**

For P only controller

$$u = K_c e(t) \quad \text{Where, } e(t) = (y_{sp} - y)$$

So, equation becomes,

$$\tau \frac{dy}{dt} - y = KK_c (y_{sp} - y)$$

$$\tau \frac{dy}{dt} + y(KK_c - 1) = KK_c y_{sp} \quad ; \text{ Closed loop governing equation ODE}$$

Closed loop characteristic equation:

$$\tau\lambda + (KK_c - 1) = 0$$

$$\lambda = -\frac{(KK_c - 1)}{\tau} \quad \text{for stable system } \lambda \text{ should be negative}$$

$$\text{So, } KK_c - 1 \geq 0 \Rightarrow K_c \geq \frac{1}{K}$$

Hence if  $K_c > \frac{1}{K}$  the system with P controller is always stable.

7.

**Solution:**

The general ODE for second order system is:  $\tau^2 \frac{d^2 y}{dt^2} + 2\xi\tau \frac{dy}{dt} + y = Ku(t)$

The solution of above differential equation is  $y = y_p + y_c$

The complementary part  $y_c$  is the solution of the equation  $\tau^2 \frac{d^2 y}{dt^2} + 2\xi\tau \frac{dy}{dt} + y = 0$

Hence the characteristic eqn. of the above ODE is

$$\tau^2 m^2 + 2\xi\tau m + 1 = 0$$

$$m = \frac{-2\xi\tau \pm \sqrt{4\xi^2\tau^2 - 4\tau^2}}{2\tau^2}$$

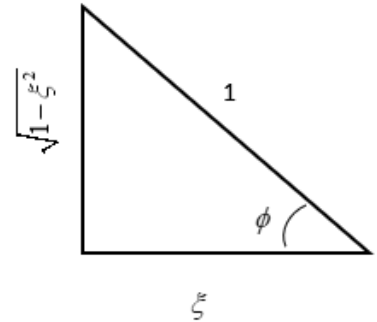
$$m = \frac{-\xi}{\tau} \pm \frac{\sqrt{\xi^2 - 1}}{\tau}$$

For  $\xi < 1$  ; oscillatory response

$$m = \frac{-\xi}{\tau} \pm \frac{\sqrt{\xi^2 - 1}}{\tau} = \frac{-\xi}{\tau} \pm j \frac{\sqrt{1 - \xi^2}}{\tau}$$

$$\text{hence, } \lambda_1 = \frac{-\xi}{\tau} + j \frac{\sqrt{1-\xi^2}}{\tau} \quad \lambda_2 = \frac{-\xi}{\tau} - j \frac{\sqrt{1-\xi^2}}{\tau}$$

$$\text{defining, } \tan \phi = \frac{\sqrt{1-\xi^2}}{\xi}$$



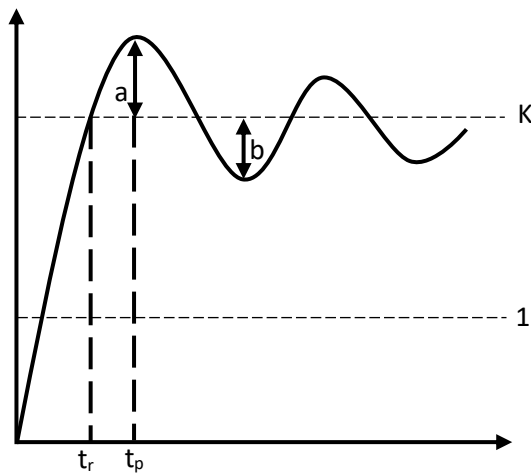
we get on solving,

$$\text{a) } y(t) = K \left[ 1 - \frac{\exp\left(\frac{-\xi}{\tau} t\right)}{\sqrt{1-\xi^2}} \sin\left(\frac{\sqrt{1-\xi^2}}{\tau} t + \phi\right) \right] \quad \{\text{sin term led to oscillation}\}$$

$$\text{As } \xi \rightarrow 0 \quad y(t) = K \left[ 1 - \sin\left(\frac{t}{\tau} + \frac{\pi}{2}\right) \right], \text{ sustained oscillations.}$$

**Note:**

- Real part of the root decides how quick will the oscillation will die out with time.
- Imaginary part decides the amplitude and time period.



- Decay ratio = b/a
- POR = a/k
- Rise time =  $t_r$
- Peak time =  $t_p$

$$\text{b) } \frac{t_r}{\tau} = \frac{\pi - \phi}{\sin \phi}$$

$$\text{c) } \frac{t_p}{\tau} = \frac{\phi}{\sin \phi}$$

$$\text{d) } \frac{a}{k} = \exp(-\pi \cot \phi)$$

$$\text{e) } \frac{b}{a} = \exp(-2\pi \cot \phi)$$

$$\text{f) } \frac{T_s}{\tau} = \frac{-\ln(f \cdot \sin \phi)}{\cos \phi}$$

$$y(t) = K \left[ 1 - \frac{\exp\left(\frac{-\xi}{\tau} t\right)}{\sqrt{1-\xi^2}} \sin\left(\frac{\sqrt{1-\xi^2}}{\tau} t + \phi\right) \right]$$

$$y(t) = K \left[ 1 - \frac{\exp(-\xi \omega_n t)}{\sqrt{1-\xi^2}} \sin(\omega t + \phi) \right]$$

$$\Rightarrow y(t) = 1 - \frac{\exp(-\xi \omega_n t)}{\sqrt{1-\xi^2}} \sin(\omega t + \phi)$$

In order to find the settling time, we assume that it will settle in  $\pm f$  % of the final value. This definition gives the following value

$$f = \frac{\exp(-\xi \omega_n t)}{\sqrt{1-\xi^2}} \sin(\omega t + \phi)$$

This above equation gives us the conservative estimate of  $\cos(\omega_n \sqrt{1-\xi^2} t - \phi) = 1$  at the settling time so the equation comes as follows

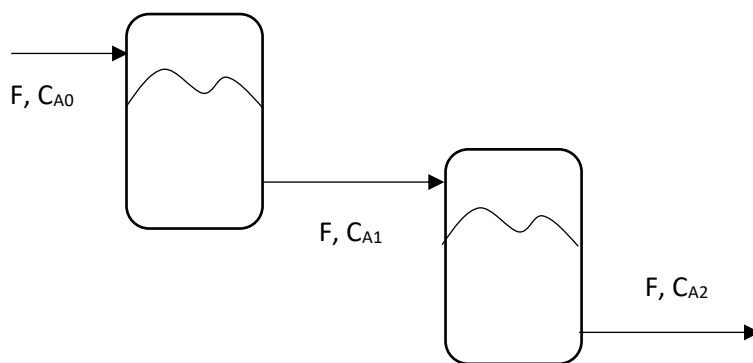
$$T_s = \frac{-\ln(f \sqrt{1-\xi^2})}{\xi \omega_n}$$

$$T_s = \frac{-\ln(f \cdot \sin \phi)}{\omega_n \cdot \cos \phi}$$

8.

**Solution:**

### Dynamics of CSTR in Series, with step change in the Inlet Concentration, Servo Response



$$\frac{V dC_{A1}}{dt} = F C_{A0} - F C_{A1} \quad \frac{V dC_{A2}}{dt} = F C_{A1} - F C_{A2} \quad \frac{V}{F} = \tau$$

Equations leads to,

$$\frac{\tau dC_{A1}}{dt} = C_{A0} - C_{A1}, \text{ Eq (a)}$$

$$\frac{\tau dC_{A2}}{dt} = C_{A1} - C_{A2}, \text{ Eq (b)}$$

Solving Equation (a) by use of integrating factor

I.F. is  $e^{\frac{t}{\tau}}$

Hence solution to the equation (a) is  $e^{\frac{t}{\tau}} C_{A1} = e^{\frac{t}{\tau}} C_{A0} + I$

$$C_{A0} = \overline{C_{A0}} + \Delta C_{A0}$$

From initial condition i.e. at  $t=0$ ,  $C_{A1}=C_{A0}$  we get

$$C_{A1} = \overline{C_{A0}} + \Delta C_{A0} \left(1 - e^{-\frac{t}{\tau}}\right)$$

Similary solve equation (b) after use I.F we get

$$e^{\frac{t}{\tau}} C_{A2} = \int \frac{1}{\tau} e^{\frac{t}{\tau}} C_{A1} dt$$

Integrating above equation by parts as

$$e^{\frac{t}{\tau}} C_{A2} = e^{\frac{t}{\tau}} C_{A1} - \int C'_{A1} e^{\frac{t}{\tau}} dt + I \quad \text{where } C'_{A1} = \frac{\Delta C_{A0}}{\tau} e^{-\frac{t}{\tau}}$$

$$e^{\frac{t}{\tau}} C_{A2} = e^{\frac{t}{\tau}} C_{A1} - \int \frac{\Delta C_{A0}}{\tau} e^{-\frac{t}{\tau}} e^{\frac{t}{\tau}} dt + I$$

$$C_{A2} = C_{A1} - \Delta C_{A0} \frac{t}{\tau} e^{-\frac{t}{\tau}} + I e^{-\frac{t}{\tau}}$$

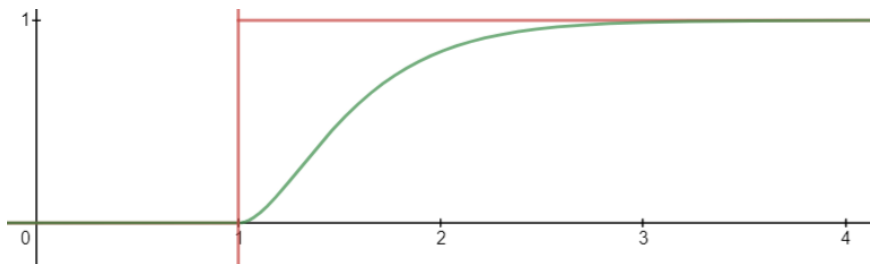
$$C_{A2} = \overline{C_{A0}} + \Delta C_{A0} \left(1 - e^{-\frac{t}{\tau}}\right) - \Delta C_{A0} \frac{t}{\tau} e^{-\frac{t}{\tau}} + I e^{-\frac{t}{\tau}}$$

From initail condition we get  $I=0$

$$C_{A2} = \overline{C_{A0}} + \Delta C_{A0} \left(1 - e^{-\frac{t}{\tau}} - \frac{t}{\tau} e^{-\frac{t}{\tau}}\right)$$

So, the response of outlet concentration,  $C_{A2}$  a unit step in inlet concentration,  $C_{A0}$  is given by,

$$C_{A2} - \overline{C_{A0}} = \Delta C_{A0} \left(1 - \exp\left(-\frac{t}{\tau}\right) \left(1 + \frac{t}{\tau}\right)\right)$$





**9. Solution:**

Let  $V_r$  = volume of each tank

$$A_1 = \rho_1 C_{p1} V_r$$

$$A_2 = \rho_2 C_{p2} V_r$$

$$B_1 = w_1 C_{p1}$$

$$B_2 = w_2 C_{p2}$$

The energy balance over 4 tanks gives ;

$$A_2 \frac{dT_6}{dt} = B_2(T_4 - T_6) + K(T_3 - T_6)$$

$$A_2 \frac{dT_4}{dt} = B_2(T_2 - T_4) + K(T_5 - T_4)$$

$$A_1 \frac{dT_5}{dt} = B_1(T_3 - T_5) + K(T_4 - T_5)$$

$$A_1 \frac{dT_3}{dt} = B_1(T_1 - T_3) + K(T_6 - T_3)$$

**10. For PI controller**

$$u = k_c e \left\{ 1 + \frac{1}{\tau_i} \int e dt \right\}$$

We know that heated tank follows 1<sup>st</sup> order kinetics whose ODE is given as

$$\tau \frac{dy}{dt} + y = ku$$

Now,

$$\tau \frac{dy}{dt} + y = k k_c (y_{sp} - y) \left\{ 1 + \frac{1}{\tau_i} \int (y_{sp} - y) dt \right\}$$

$$\tau \frac{d^2 y}{dt^2} + \frac{dy}{dt} (1 + k k_c) + \frac{y (k k_c)}{\tau_i} = \{ y_{sp} + \tau_i \frac{dy_{sp}}{dt} \} \frac{k k_c}{\tau_i}$$

Characteristic Equation

$$\tau \lambda^2 + \lambda (1 + k k_c) + \frac{(k k_c)}{\tau_i} = 0$$

$$\lambda^2 + \lambda \frac{(1 + k k_c)}{\tau} + \frac{(k k_c)}{\tau \tau_i} = 0$$

On solving this equation by quadratic formula

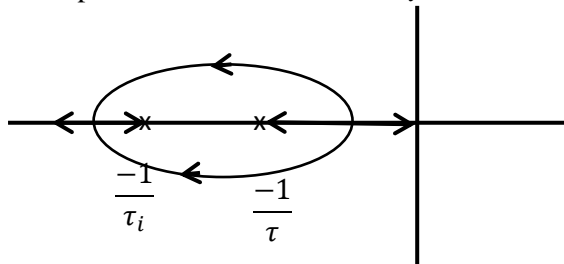
$$\lambda = -\frac{(1 + k k_c)}{2\tau} \pm \frac{(1 + k k_c)}{\tau} \sqrt{1 - \frac{4k\tau}{\tau_i} \frac{(k_c)}{(1 + k k_c)^2}}$$

Now for judging the nature of roots

$$D = (1 - kk_c)^2 + 4kk_c(1 - r)$$

$$r = \frac{\tau}{\tau_i}$$

For complex root  $r > 1$   $\tau > \tau_i$



Hence a PI controller is stable.

## 11. Solution:

a) For 1<sup>st</sup> order system

$$u = K_c e(t) \quad \text{Where, } e(t) = (y_{sp} - y)$$

So, equation becomes,

$$\tau \frac{dy}{dt} + y = K K_c (y_{sp} - y)$$

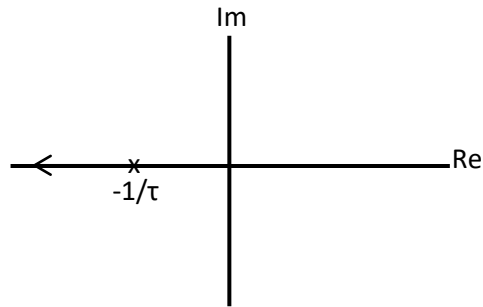
$$\tau \frac{dy}{dt} + y (K K_c + 1) = K K_c y_{sp} \quad ; \text{ Closed loop governing equation ODE}$$

Closed loop characteristic equation:

$$\tau \lambda + (K K_c + 1) = 0$$

$$\lambda = -\frac{(K K_c + 1)}{\tau} \quad \text{for stable system } \lambda \text{ should be negative}$$

Now, vary  $K_c$  from 0 to infinity  $\longrightarrow$  giving rise to root locus



Hence always stable.

b) For 2<sup>nd</sup> order system

$$u = k_c e\{t\}$$

We know

$$\tau_1 \tau_2 \frac{d^2 y}{dt^2} + (\tau_1 + \tau_2) \frac{dy}{dt} + y = k k_c (y_{sp} - y)$$

$$\tau_1 \tau_2 \frac{d^2 y}{dt^2} + \frac{dy}{dt} (\tau_1 + \tau_2) + y (1 + k k_c) = y_{sp} k k_c$$

Characteristic Equation

$$\tau_1 \tau_2 \lambda^2 + \lambda (\tau_1 + \tau_2) + (1 + k k_c) = 0$$

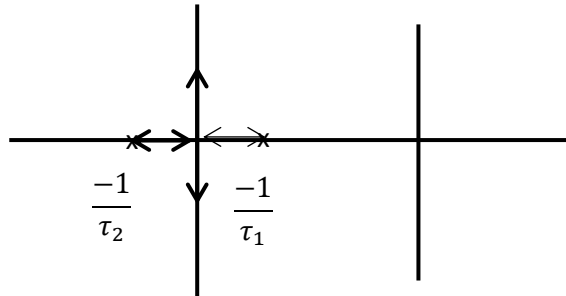
$$\lambda^2 + \lambda \frac{(\tau_1 + \tau_2)}{\tau_1 \tau_2} + \frac{(1 + k k_c)}{\tau_1 \tau_2} = 0$$

On solving this equation by quadratic formula

$$\lambda = -\frac{(\tau_1 + \tau_2)}{2 \tau_1 \tau_2} \pm \frac{1}{2 \tau_1 \tau_2} \sqrt{(\tau_1 + \tau_2)^2 - 4 \tau_1 \tau_2 (1 + k k_c)}$$

Now for judging the nature of roots

$$D = (\tau_1 - \tau_2)^2 - 4\tau_1\tau_2kk_C$$



Hence a P controller is stable.