Digital Signal Processing SS 2024 – Exercise 1 Digital Signal Processing Tutorial

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Exercise 1

We have three complex numbers given:

$$c_1 = -5 + 3j$$
 $c_2 = \frac{\sqrt{2}}{2}e^{-\frac{3\pi j}{4}}$ $c_3 = \frac{1}{\sqrt{2}} + \frac{1}{\sqrt{2}}j$

a) We do the calculations by hand. First, let's use Euler's formula to simplify c_2 :

$$c_2 = \frac{\sqrt{2}}{2}e^{-\frac{3\pi j}{4}} = \frac{\sqrt{2}}{2}\left(\cos\left(\frac{3\pi}{4}\right) + j\sin\left(\frac{3\pi}{4}\right)\right) = \frac{\sqrt{2}}{2}\left(-\frac{\sqrt{2}}{2} + \frac{\sqrt{2}}{2}j\right) = -\frac{1}{2} + \frac{1}{2}j$$

With this, let us now compute the following numbers:

$$c_{4} = c_{1} + c_{2} = -5 - \frac{1}{2} + 3j + \frac{1}{2}j = -\frac{11}{2} + \frac{7}{2}j$$

$$c_{5} = c_{1} \cdot c_{2} = (-5 + 3j) \cdot \left(-\frac{1}{2} + \frac{1}{2}j\right) = \frac{5}{2} - \frac{5}{2}j - \frac{3}{2}j - \frac{3}{2} = 1 - 4j$$

$$c_{6} = |c_{3}|^{2} = \left(\sqrt{\left(\frac{1}{\sqrt{2}}\right)^{2} + \left(\frac{1}{\sqrt{2}}\right)^{2}}\right)^{2} = \left(\sqrt{\frac{1}{2} + \frac{1}{2}}\right)^{2} = \left(\sqrt{1}\right)^{2} = 1$$

$$c_{7} = \arg\left(c_{3}\right) = \arctan\left(\frac{1}{\sqrt{2}}, \frac{1}{\sqrt{2}}\right) = \arctan\left(\frac{\frac{1}{\sqrt{2}}}{\frac{1}{\sqrt{2}}}\right) = \arctan\left(1\right) = \frac{\pi}{4}$$

$$c_{8} = \frac{c_{1}}{c_{2}} = \frac{-5 + 3j}{\frac{-1+1j}{2}} = \frac{-10 + 6j}{-1 + 1j} = \frac{(-1 - 1j)(-10 + 6j)}{(-1 - 1j)(-1 + 1j)} = \frac{10 - 6j + 10j + 6}{1 - 1j + 1j + 1} = 8 - 2j$$

b) We check the result using MATLAB.

This is the code we use:

$$c1 = -5 + 3j$$

$$c2 = sqrt(2)/2 * exp((3j*pi)/4)$$

$$c3 = 1/sqrt(2) + 1j/sqrt(2)$$

$$c4 = c1 + c2$$

$$c5 = c1 * c2$$

 $c_9 = c_1 \cdot c_1^*$ = (-5+3j)(-5-3j) = 25+15j-15j+9 = 34

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c6 = abs(c3)^2

c7 = angle(c3)

c8 = c1 / c2

c9 = c1 * c1'
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These are the results:

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\begin{array}{l} c1 &=& -5.0000 \,+\, 3.0000 \,\mathrm{i} \\ c2 &=& -0.5000 \,+\, 0.5000 \,\mathrm{i} \\ c3 &=& 0.7071 \,+\, 0.7071 \,\mathrm{i} \\ c4 &=& -5.5000 \,+\, 3.5000 \,\mathrm{i} \\ c5 &=& 1.0000 \,-\, 4.0000 \,\mathrm{i} \\ c6 &=& 1.0000 \\ c7 &=& 0.7854 \\ c8 &=& 8.0000 \,+\, 2.0000 \,\mathrm{i} \\ c9 &=& 34 \end{array}
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Checks out. (Question: what is the best way to get this output from MATLAB to latex? The way I did it was copying it from the console, but that was rather cumbersome since I needed to remove the blank lines.)

Exercise 2

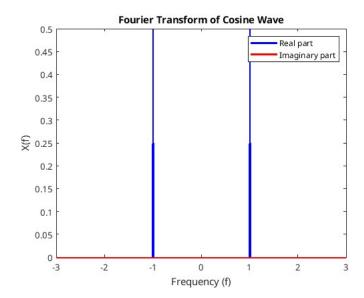
We need to prooof the following relation (Fourier transform pair for the cosine wave)

$$x(t) = \hat{X}\cos(2\pi f_0 t) \leftrightarrow X(f) = \frac{\hat{X}}{2}\delta(f - f_0) + \frac{\hat{X}}{2}\delta(f + f_0)$$

- 1. We use Eulers formula to express the cosine in the time domain as sum of complex exponents.
 - For this we use the formula for the cosine wave from (DSP_2.pdf, page 5) $s(t) = \hat{s}\cos\left(2\pi f_0 t + \varphi_0\right) = \frac{\hat{s}}{2}\left(e^{j(\omega_0 t + \varphi_0)} + e^{-j(\omega_0 t + \varphi_0)}\right)$
 - And we get: $x(t) = \hat{X}\cos(2\pi f_0^t t) = \frac{\hat{X}}{2}e^{j2\pi f_0 t} + \frac{\hat{X}}{2}e^{-j2\pi f_0 t}$
- 2. We express the Fourier Transform using (DSP_2.pdf, page 38).
 - The Fourier Transform of a complex exponential function is a delta function. $x(t) = \hat{X}e^{j2\pi f_0 t} \quad \circ \bullet \quad X(f) = \hat{X}\delta(f f_0)$
- 3. From 1. and 2. we get:

$$\implies X(f) = \frac{\hat{X}}{2}\delta(f - f_0) + \frac{\hat{X}}{2}\delta(f + f_0)$$
 q.e.d

Add a diagram of X(f) in the report (draw the real and imaginary part of X(f) in the same diagram).



Exercise 3

Given are two sines according to the following formula:

$$x_i(t) = \sin(2\pi f_i t)$$
, with $i \in \{1, 2\}$ with $f_1 = 1$ Hz and $f_2 = 3$ Hz.

All two sines are time delayed by $\tau = 0.1$ s to yield

$$y_i(t) = \sin(2\pi f_i(t - 0.1))$$

This corresponds to a phase shift. Thus, the delayed sines may also be written as

$$y_i(t) = \sin\left(2\pi f_i t + \phi_i\right)$$

- a) We have to calculate the phase shifts ϕ_i for each sine and verify that this corresponds to the "Shift Theorem"
- 1. Calculate the phase shifts ϕ_i
 - We know that:

$$\sin(2\pi f_i(t-\tau)) = \sin(2\pi f_i t + \phi_i)$$

• So:

$$2\pi f_i(t - \tau) = 2\pi f_i t + \phi_i$$

$$\implies \phi_i = 2\pi f_i t - 2\pi f_i t \tau - 2\pi f_i t = 2\pi f_i t \tau$$

• So we get:

$$\phi_1 = -2\pi \cdot 1 \cdot 0.1 = -0.2\pi$$

$$\phi_2 = -2\pi \cdot 3 \cdot 0.1 = -0.6\pi$$

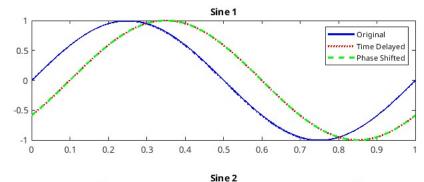
- 2. Verify that this corresponds to the "Shift Theorem" of the Fourier Transform
 - We use the formula for the "Shift Theorem" (DSP_2.pdf, page 41)

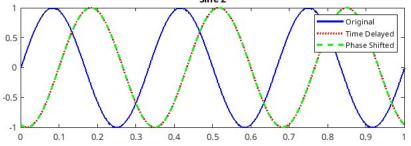
$$x(t-T) \circ - \bullet X(f)e^{-j2\pi fT}$$

• So, since $\phi_i = 2\pi f_i t \tau$, we get

$$x(t-\tau) \circ - \bullet X(f)e^{-j2\pi f\tau} = X(f)e^{-j\phi_i}$$
 q.e.d

b) For both sines in separated plots: Plot the original signal, the time delayed signal and the phase shifted signal. Since the latter two are identical, show this by plotting the first with a solid line and the overlaid one in a different colour with a dashed line. Plot each signal from 0 to 1s. In Matlab use the following time-vector: t = 0:0.001:1.





Exercise 4

We need to find out whether

a)
$$y(t) = (x(t))^2$$
 b) $y(t) = x(t) \cdot \sin(\Omega_0 t)$

are linear and time invariant.

- a) Let $y(t) = (x(t))^2$.
 - 1) Linearity. It is easy to see that the system is not linear:

$$(\alpha x(t))^2 = \alpha^2 (x(t))^2 = \alpha^2 y(t) \neq \alpha y(t).$$

- 2) **Time invariance.** Let $y_1(t) = (x(t-\tau))^2$ be the system in which input is delayed, and let $y_2(t) = y(t-\tau)$ be the system in which the output is delayed. Since $y_1 = (x(t-\tau))^2 = y(t-\tau) = y_2$, we can see that the system is time invariant.
- b) Let $y(t) = x(t) \cdot \sin(\Omega_0 t)$.
 - 1) **Linearity.** Let $y_1(t) = (\alpha x_1(t) + \beta x_2(t)) \sin(\Omega_0 t)$, and let $y_2(t) = \alpha x_1(t) \sin(\Omega_0 t) + \beta x_2(t) \sin(\Omega_0 t)$. Then we can see that

$$y_1(t) = (\alpha x_1(t) + \beta x_2(t))\sin(\Omega_0 t) = \alpha x_1(t)\sin(\Omega_0 t) + \beta x_2(t)\sin(\Omega_0 t) = y_2(t).$$

Therefore, the system is linear.

2) **Time invariance.** Let $y_1(t) = x(t-\tau)\sin(\Omega_0 t)$ be the system in which the input is delayed, and let $y_2(t) = y(t-\tau)$ be the system in which the output is delayed. Since $y_1 = x(t-\tau)\sin(\Omega_0 t) \neq x(t-\tau)\sin(\Omega_0 (t-\tau)) = y(t-\tau) = y_2$, we can see that the system in not time invariant.