

# First-Order Transients in Linear Electrical Networks (2)

Lecture 13

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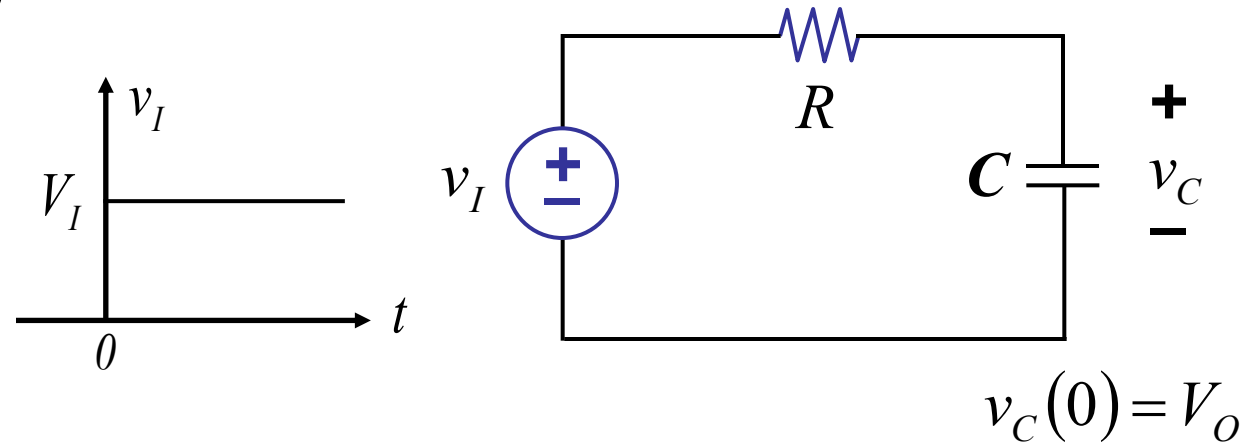
Computer Science and Engineering

Seoul National University

***Slide credits: Prof. Anant Agarwal at MIT***

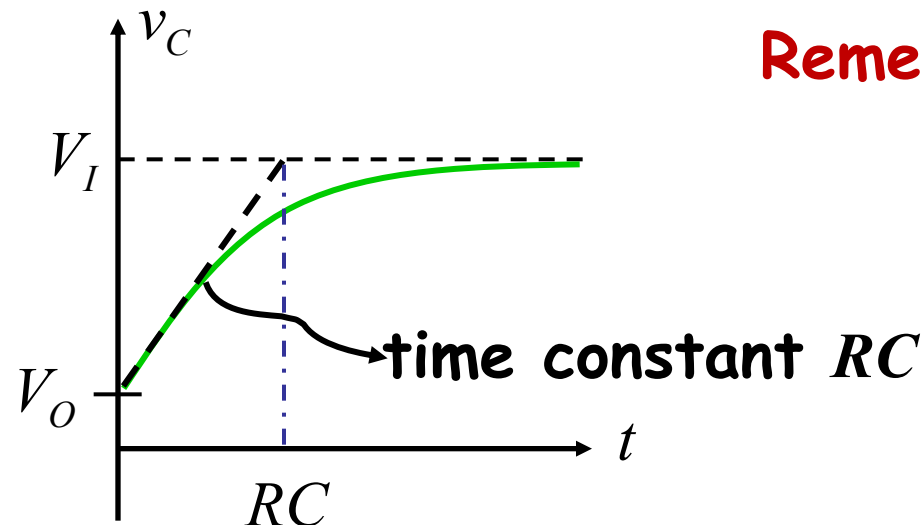
V-code: ???

# Review: Analysis of RC Circuits



$$v_C = V_I + (V_O - V_I) e^{\frac{-t}{RC}} \quad \text{—————} \quad \textcircled{1}$$

**Remember this!**



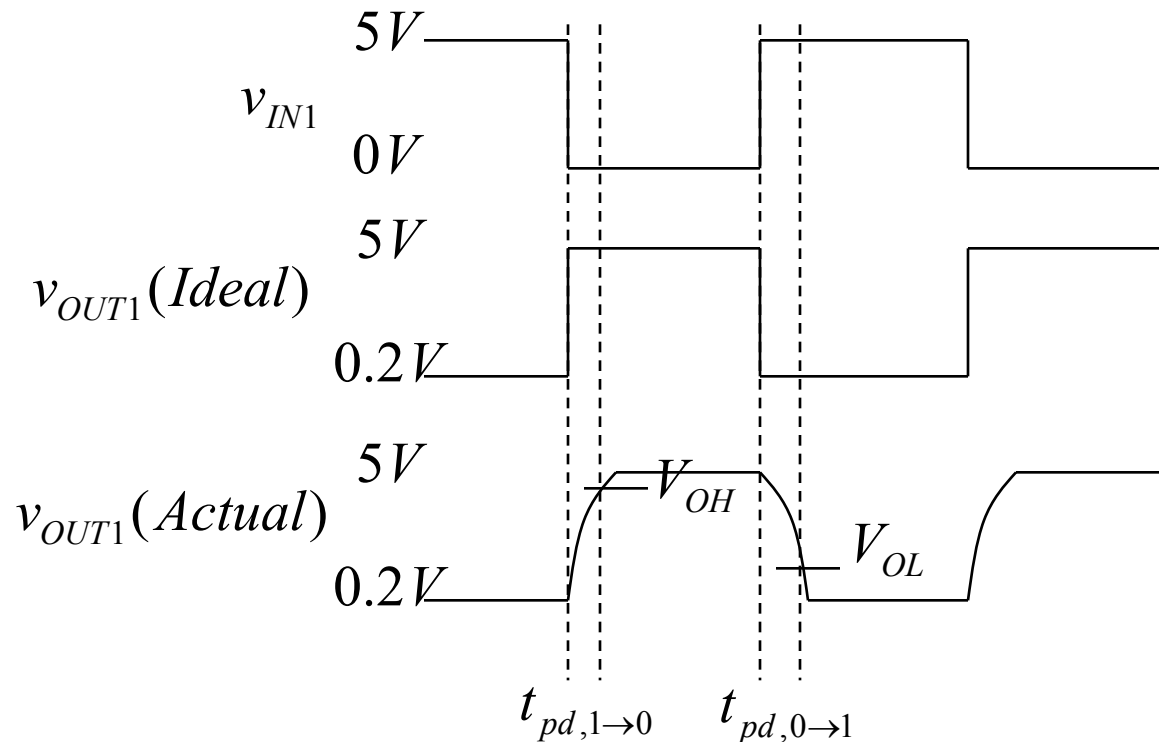
# Outline

**Textbook: 10.4, 10.5, 10.6.2, 10.7**

- **Propagation Delay and the Digital Abstraction**
- State and State Variables
- Digital Memory

# Propagation Delay

## ■ Definitions



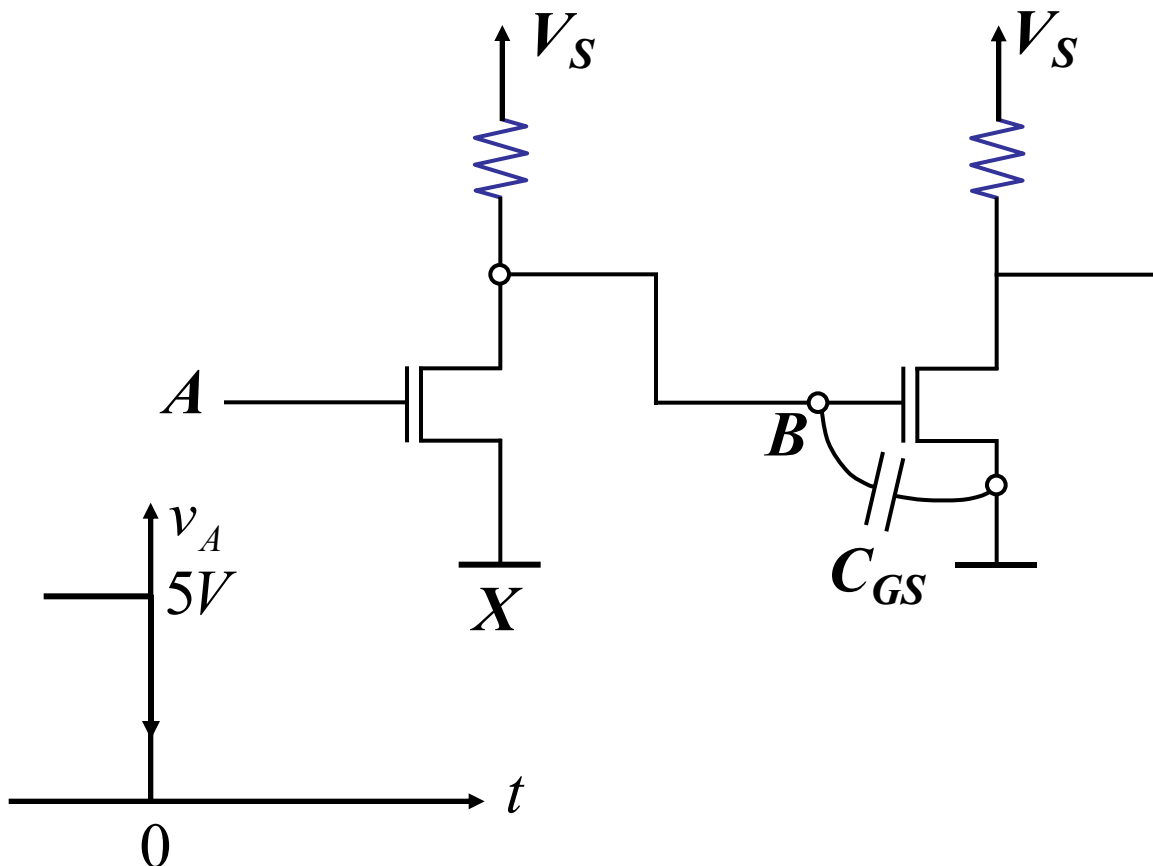
**Propagation delay**  $t_{pd} = \max(t_{pd,1 \rightarrow 0}, t_{pd,0 \rightarrow 1})$

V-code: ???

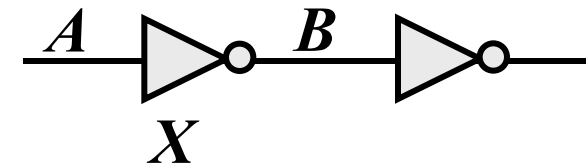
# Propagation Delay

## Let's apply the result to an inverter.

First, rising delay  $t_r$  at  $B$

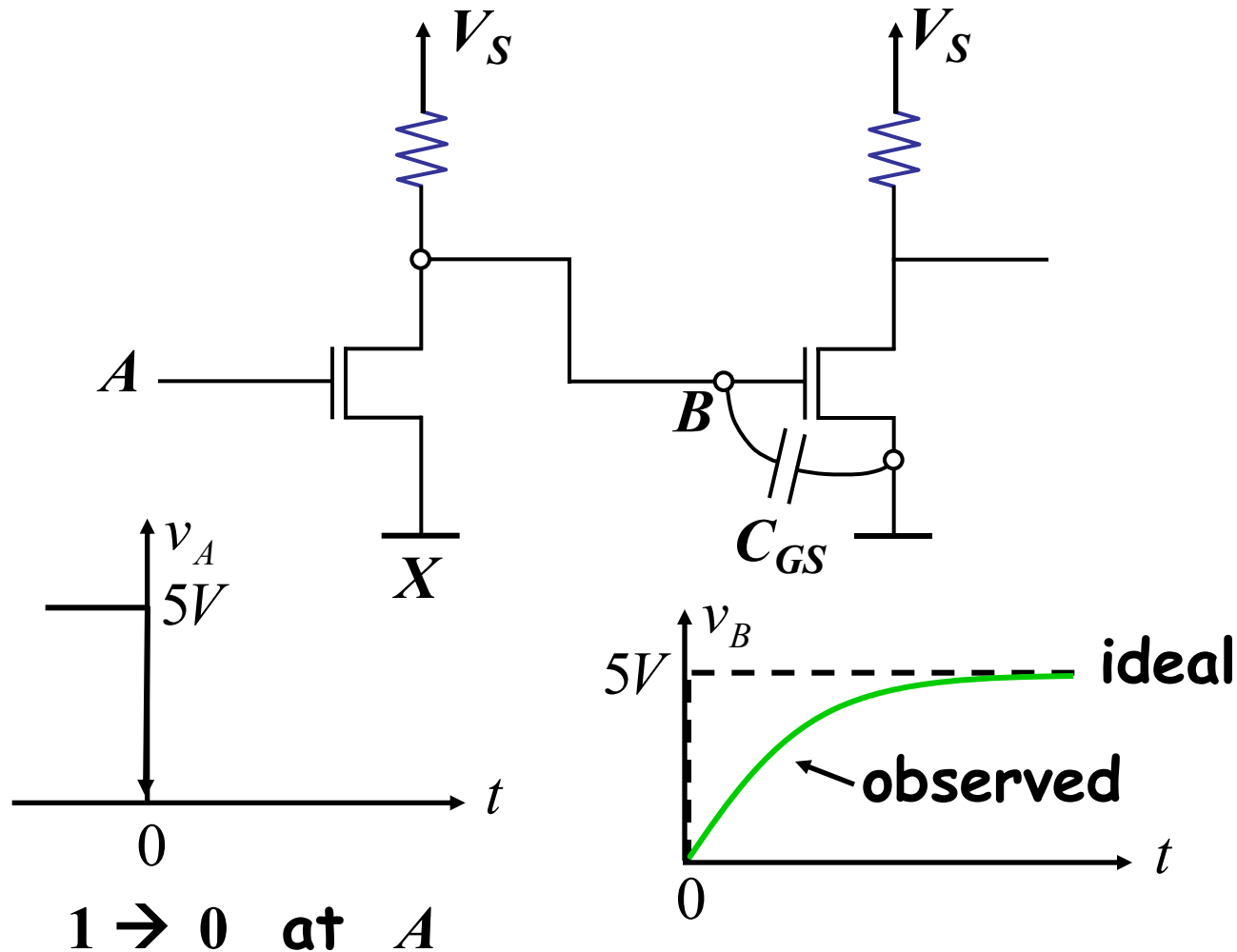


**1  $\rightarrow$  0 at  $A$**



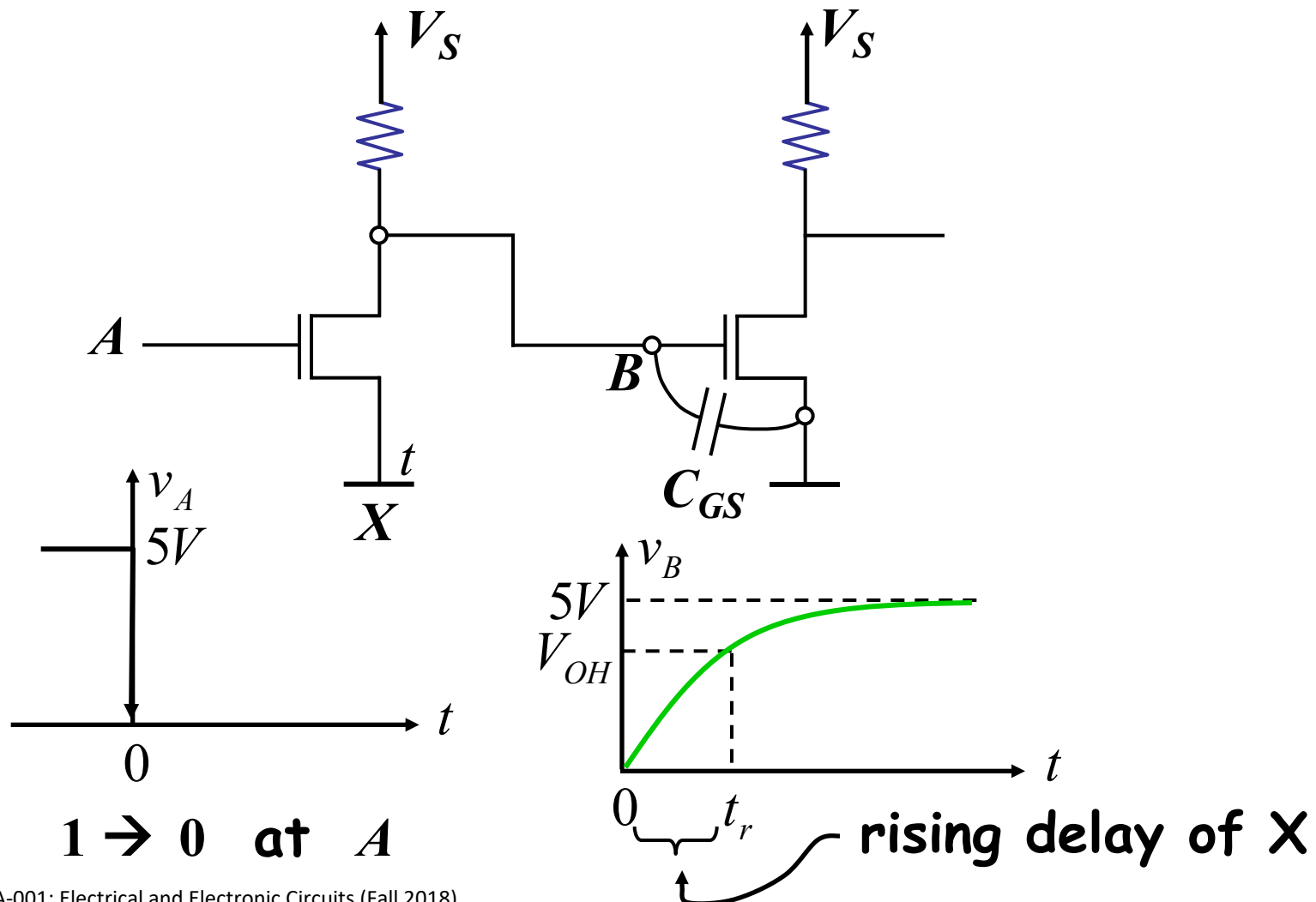
# Propagation Delay

First, rising delay  $t_r$  at  $B$



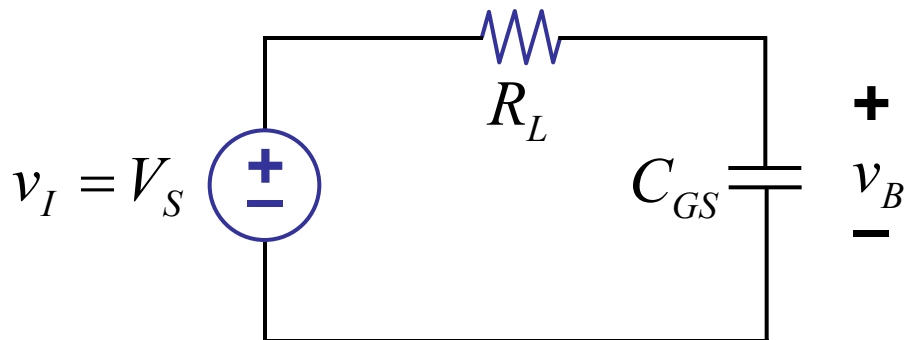
# Propagation Delay

First, rising delay  $t_r$  at  $B$



# Propagation Delay

Equivalent circuit for  $0 \rightarrow 1$  at  $B$



$$\begin{aligned} v_I &= V_S \\ v_B(0) &= 0 \quad \text{for } t \geq 0 \end{aligned}$$

From ①

$$v_B = V_S + (0 - V_S) e^{\frac{-t}{R_L C_{GS}}}$$

Now, we need to find  $t$  for which  $v_B = V_{OH}$ .



# Propagation Delay

First, rising delay  $t_r$  at  $B$

Or

$$v_{OH} = V_S - V_S e^{\frac{-t}{R_L C_{GS}}}$$

Find  $t_r$  :

$$V_S e^{\frac{-t_r}{R_L C_{GS}}} = V_S - V_{OH}$$

$$\frac{-t_r}{R_L C_{GS}} = \ln \frac{V_S - V_{OH}}{V_S}$$

$$t_r = -R_L C_{GS} \ln \frac{V_S - V_{OH}}{V_S}$$

# Propagation Delay

First, rising delay  $t_r$  at  $B$

Or

$$v_{OH} = V_S - V_S e^{\frac{-t}{R_L C_{GS}}}$$

Find  $t_r$  :

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$$t_r = -R_L C_{GS} \ln \frac{V_S - V_{OH}}{V_S}$$

e.g.

$$R_L = 1K$$

$$V_S = 5V$$

$$C_{GS} = 0.1 pF$$

$$V_{OH} = 4V$$

$$t_r = -1 \times 10^3 \times 0.1 \times 10^{-12} \ln \frac{5-4}{5}$$

$$= 0.16 ns$$

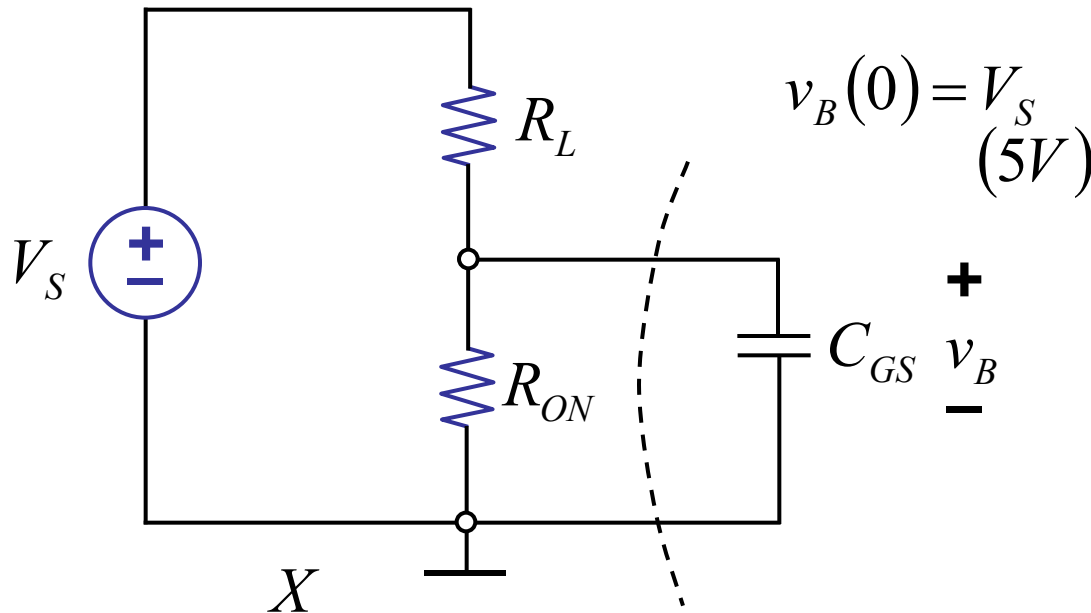
$$RC = 0.1 ns !$$

# Propagation Delay

## Falling Delay $t_f$

Falling delay  $t_f$  is  
the  $t$  for which  $v_B$  falls to  $V_{OL}$

Equivalent circuit for  $1 \rightarrow 0$  at  $B$

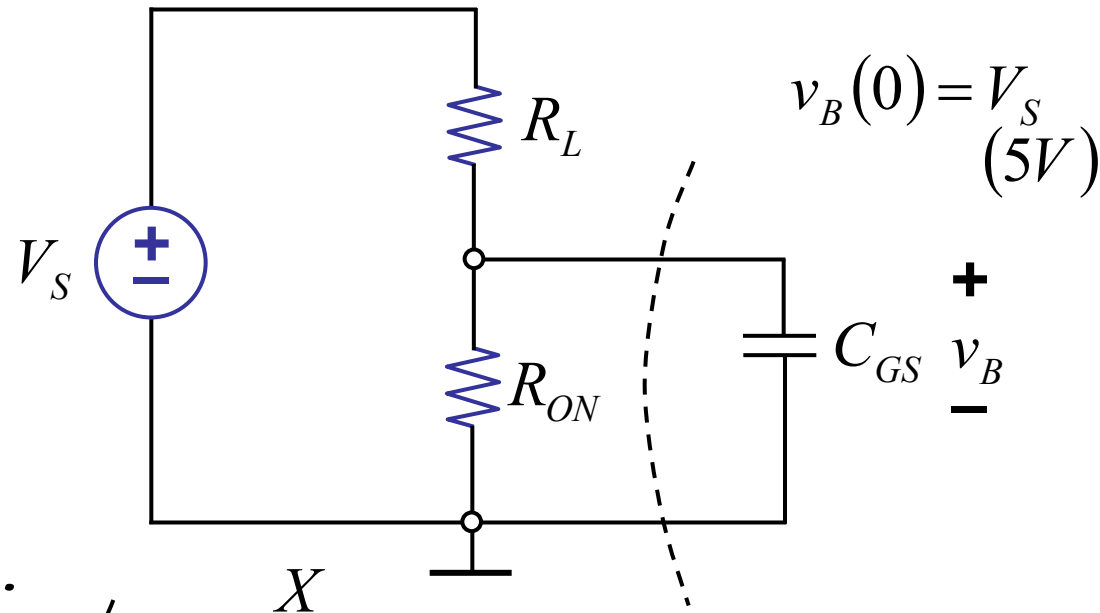


V-code: ???

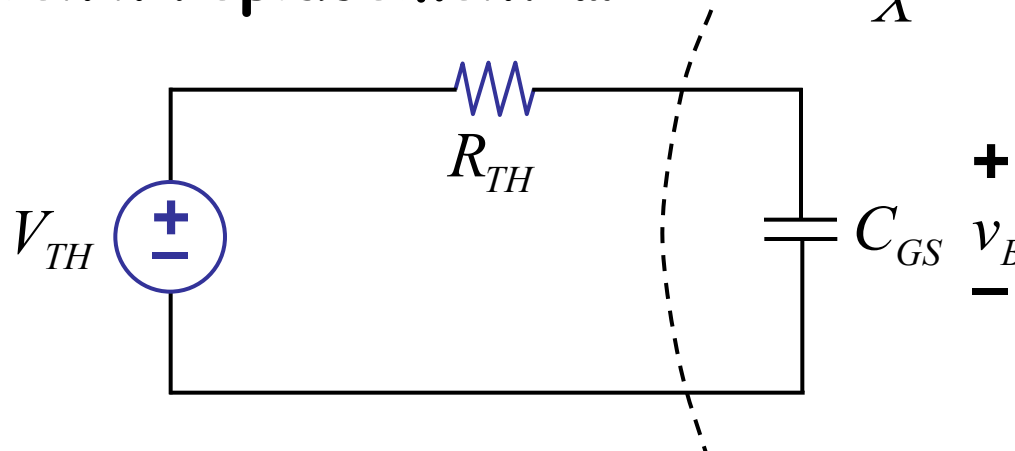
# Propagation Delay

## Falling Delay $t_f$

### Equivalent circuit for $1 \rightarrow 0$ at $B$



Thévenin replacement ...



$$R_{TH} = R_L \parallel R_{ON}$$

$$V_{TH} = V_S \frac{R_{ON}}{R_{ON} + R_L}$$

# Propagation Delay

## Falling Delay $t_f$

From ①

$$v_B = V_{TH} + (V_S - V_{TH}) e^{\frac{-t}{R_{TH}C_{GS}}}$$

Falling decay  $t_f$  is  
the  $t$  for which  $v_B$  falls to  $V_{OL}$

$$V_{OL} = V_{TH} + (V_S - V_{TH}) e^{\frac{-t_f}{R_{TH}C_{GS}}}$$

or

$$t_f = -R_{TH}C_{GS} \ln \frac{V_{OL} - V_{TH}}{V_S - V_{TH}}$$

# Propagation Delay

## Falling Delay $t_f$

$$t_f = -R_{TH} C_{GS} \ln \frac{V_{OL} - V_{TH}}{V_S - V_{TH}}$$

**e.g.**      $R_L = 1K$               $V_S = 5V$       $R_{ON} = 10\Omega$

$$C_{GS} = 0.1 pF \quad V_{OL} = 1V$$

$$R_{TH} \approx 10\Omega, \quad V_{TH} \approx 0V$$

$$t_f = -10 \cdot 0.1 \cdot 10^{-12} \ln \frac{1}{5}$$

$$= 1.6 ps$$

$$RC = 1 ps !$$

# Outline

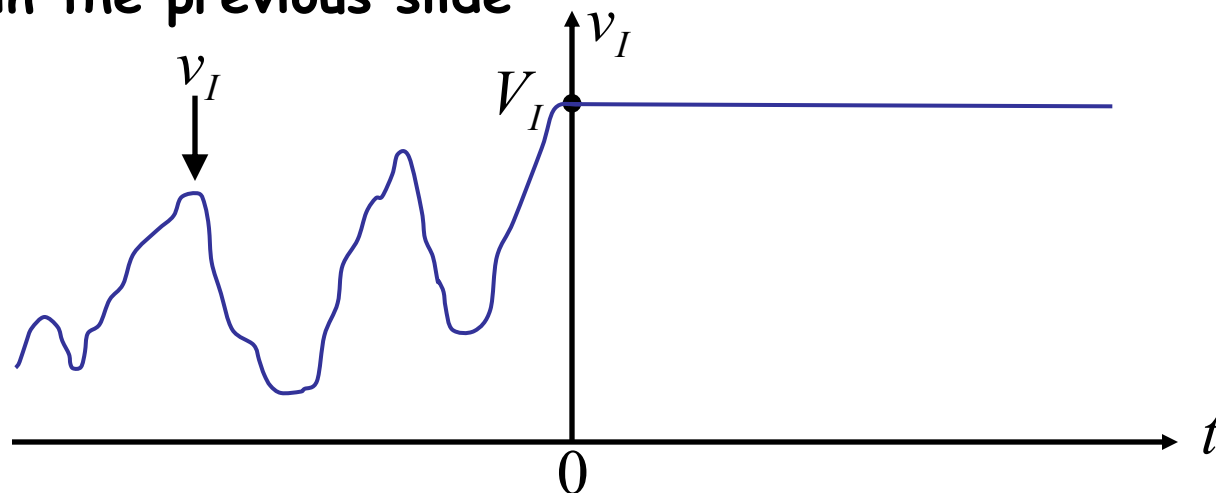
**Textbook: 10.4, 10.5, 10.6.2, 10.7**

- Propagation Delay and the Digital Abstraction
- **State and State Variables**
- Digital Memory

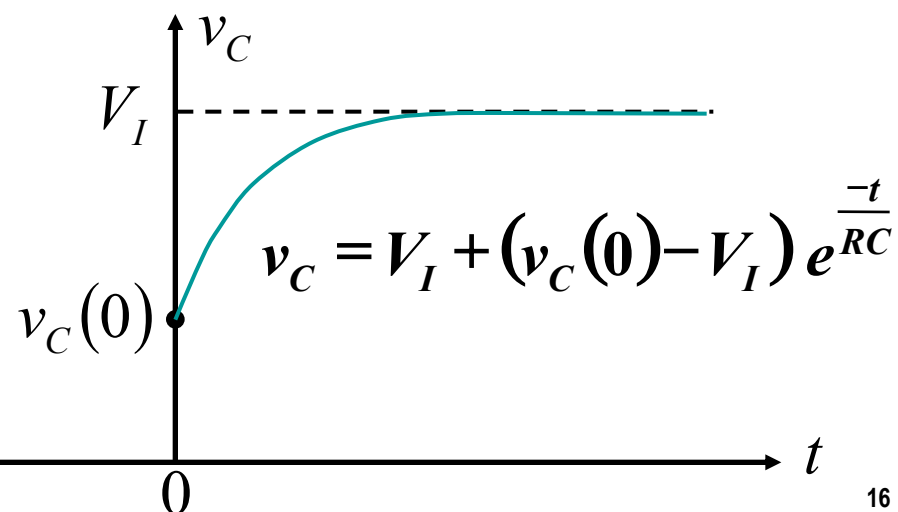
# State and State Variables

This lecture will dwell on the memory property of capacitors.

For the RC circuit in the previous slide



Notice that the capacitor voltage for  $t \geq 0$  is independent of the form of the input voltage before  $t = 0$ . Instead, it depends only on the capacitor voltage at  $t = 0$ , and the input voltage for  $t = 0$ .

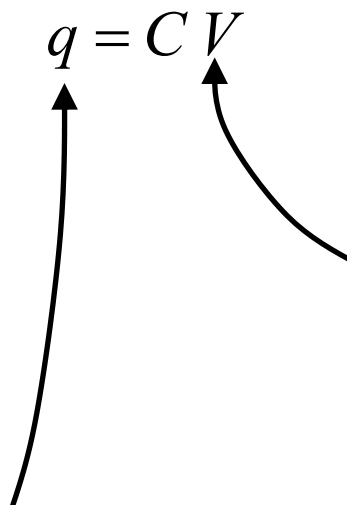




# State and State Variables

**State** : summary of past inputs relevant to predicting the future

$q = C V$



for linear capacitors,  
capacitor voltage  $V$   
is also state variable

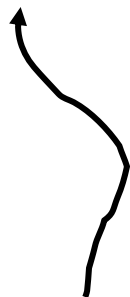
state variable, actually

# State and State Variables

Back to our simple RC circuit ①

$$v_C = f(v_C(0), v_I(t))$$

$$v_C = V_I + (v_C(0) - V_I) e^{\frac{-t}{RC}}$$



Summarizes the past input  
relevant to predicting future  
behavior

# State and State Variables

**An alternative method to solve the problem:**  
**Solve the transient problem by superposition!**

$$\frac{dv_C}{dt} = -\frac{v_C}{RC} + \frac{v_I(t)}{RC}$$

$$\frac{d}{dt} (\text{state variable}) = K_1 (\text{State variable present value}) + K_2 (\text{input variable})$$

**Total Solution = zero-input response + zero-state response**

# State and State Variables

**An alternative method to solve the problem:**  
**Solve the transient problem by superposition!**

$$\frac{dv_C}{dt} = -\frac{v_C}{RC} + \frac{v_I(t)}{RC}$$

**Zero-input response:**  $\frac{dv_C}{dt} = -\frac{v_C}{RC}$

**Zero-state response:**  $\frac{dv_C}{dt} = -\frac{v_C}{RC} + \frac{v_I(t)}{RC}$

# State and State Variables

We are often interested in circuit response for

- zero state  $v_C(\mathbf{0}) = 0$
- zero input  $v_I(\mathbf{t}) = 0$

Correspondingly,

- zero state response or *ZSR*

$$v_C = V_I - V_I e^{\frac{-t}{RC}} \quad \text{—————} \quad \textcircled{2}$$

**Total solution**  
=  $\textcircled{2} + \textcircled{3}$

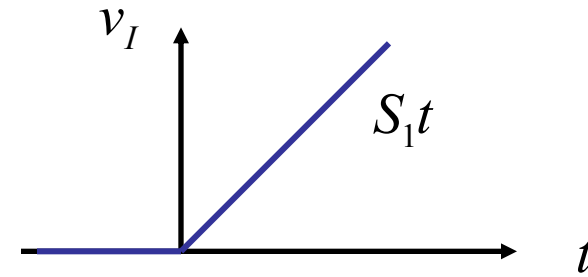
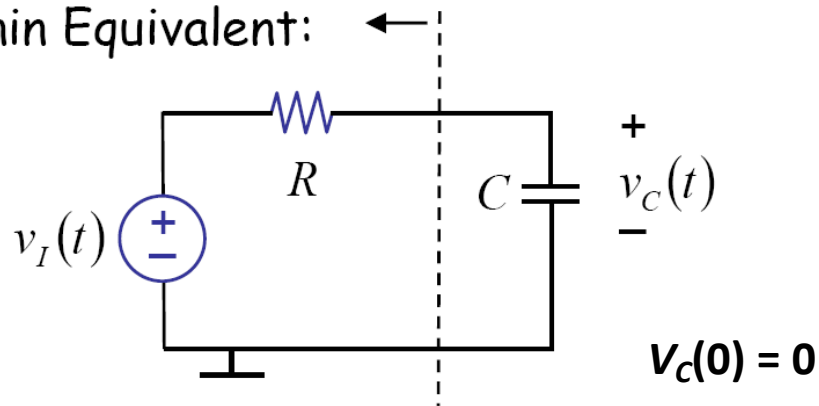
- zero input response or *ZIR*

$$v_C = v_C(0) e^{\frac{-t}{RC}} \quad \text{—————} \quad \textcircled{3}$$

# State and State Variables

- Ramp input as an example: Total solution = ZSR (+ ZIR)

Thévenin Equivalent:



Ramp input

# Outline

**Textbook: 10.4, 10.5, 10.6.2, 10.7**

- Propagation Delay and the Digital Abstraction
- State and State Variables
- **Digital Memory**

V-code: ???

# Digital Memory

One application of STATE



Why memory?

**DIGITAL MEMORY**

Or, why is combinational logic insufficient?

## Examples

- Consider adding 6 numbers on your calculator

$$2 + 9 + 6 + 5 + 3 + 8$$

**M+**

("Add the displayed value to the memory")

- "Remembering" transient inputs

(	)	mc	m+	m-	mr	AC	+/-	%	÷
2 <sup>nd</sup>	x <sup>2</sup>	x <sup>3</sup>	x <sup>y</sup>	e <sup>x</sup>	10 <sup>x</sup>	7	8	9	×
1/x	√x	∛x	∜x	ln	log <sub>10</sub>	4	5	6	-
x!	sin	cos	tan	e	EE	1	2	3	+
Rad	sinh	cosh	tanh	π	Rand	0	.		=



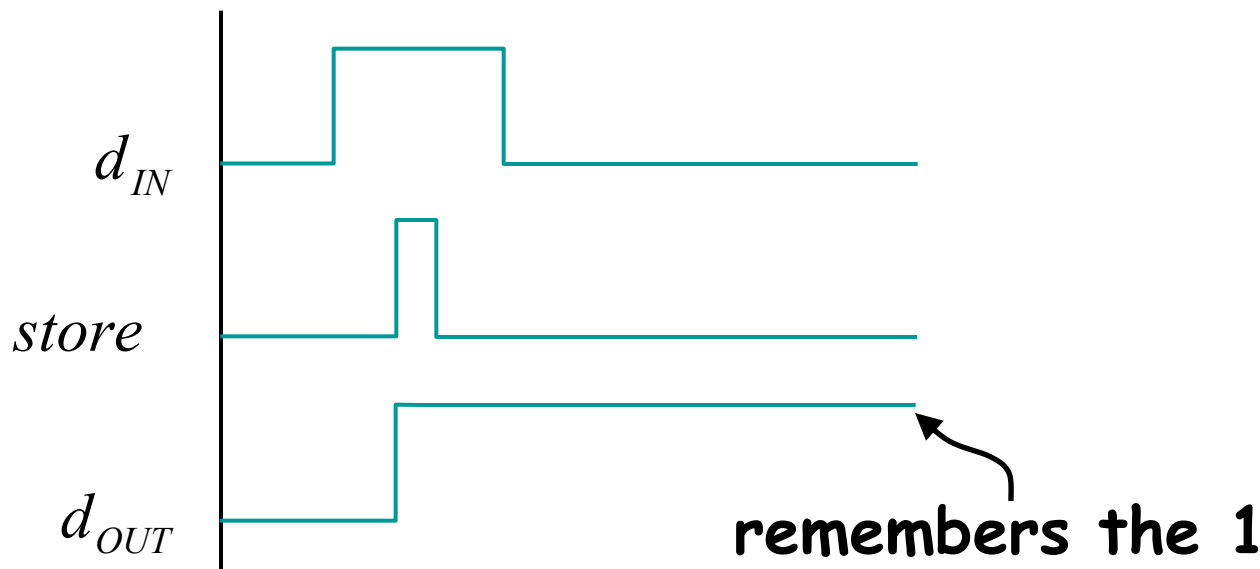
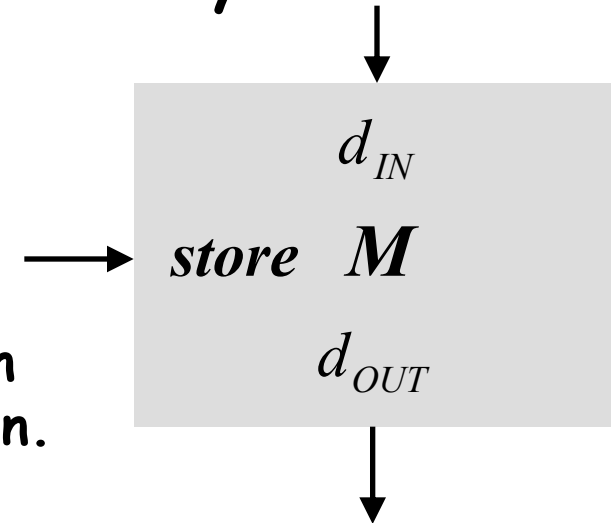
V-code: ???

# Digital Memory

## Memory Abstraction

A 1-bit memory element

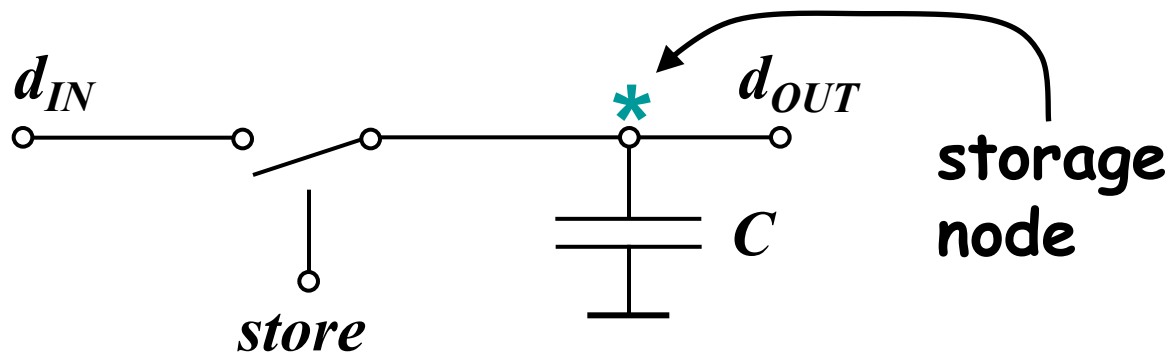
Remembers input when *store* goes high.  
Like a camera that records input ( $d_{IN}$ ) when the user presses the shutter release button.  
The recorded value is visible at  $d_{OUT}$ .



# Digital Memory

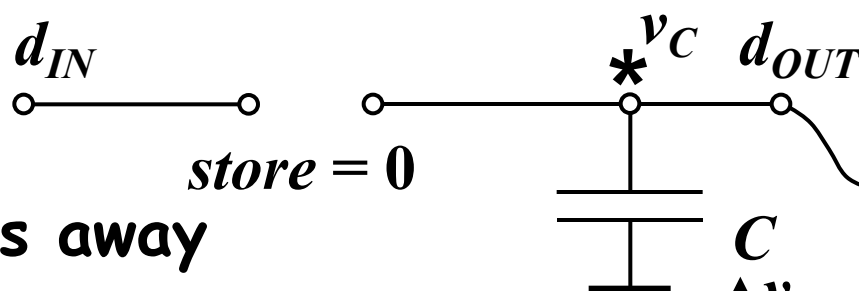
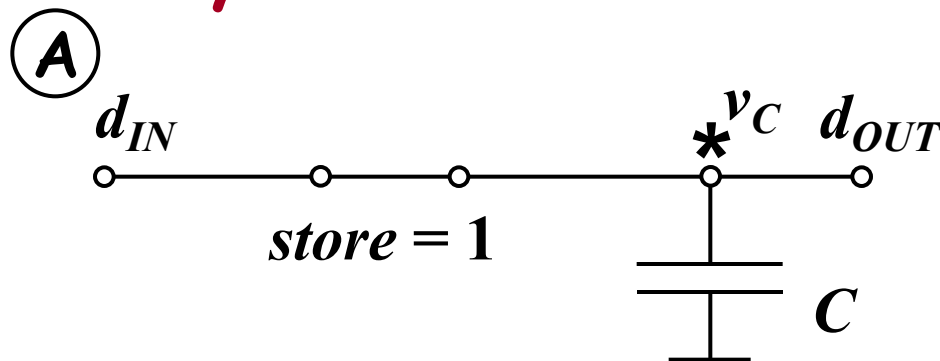
## Building a memory element ...

### Ⓐ First attempt



# Digital Memory

## Building a memory element ...



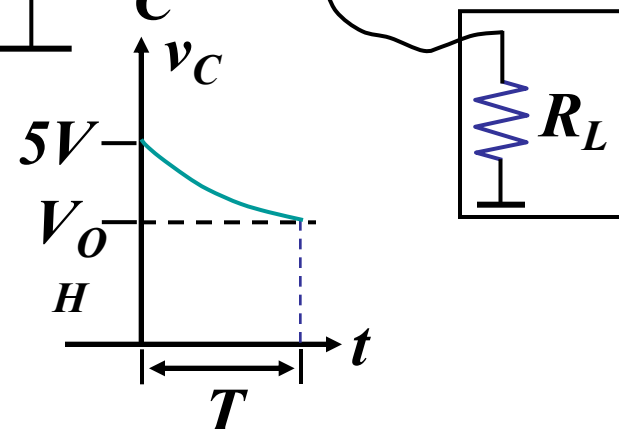
Stored value leaks away

$$v_C = 5 \cdot e^{\frac{-t}{R_L C}}$$

from ②

$$T = -R_L C \ln \frac{V_{OH}}{5}$$

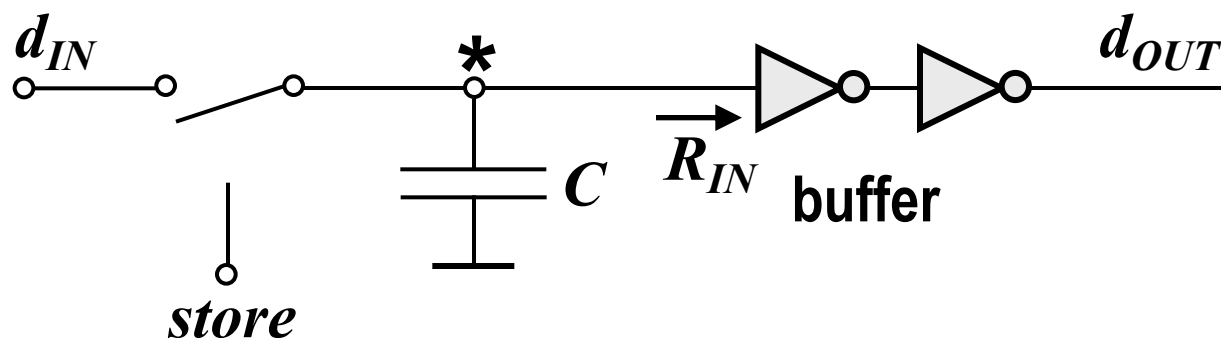
$store \text{ pulse width} \gg R_{ON} C$



# Digital Memory

## Building a memory element ...

Ⓑ Second attempt → buffer



Input resistance  $R_{IN}$

$$T = -R_{IN}C \ln \frac{V_{OH}}{5}$$

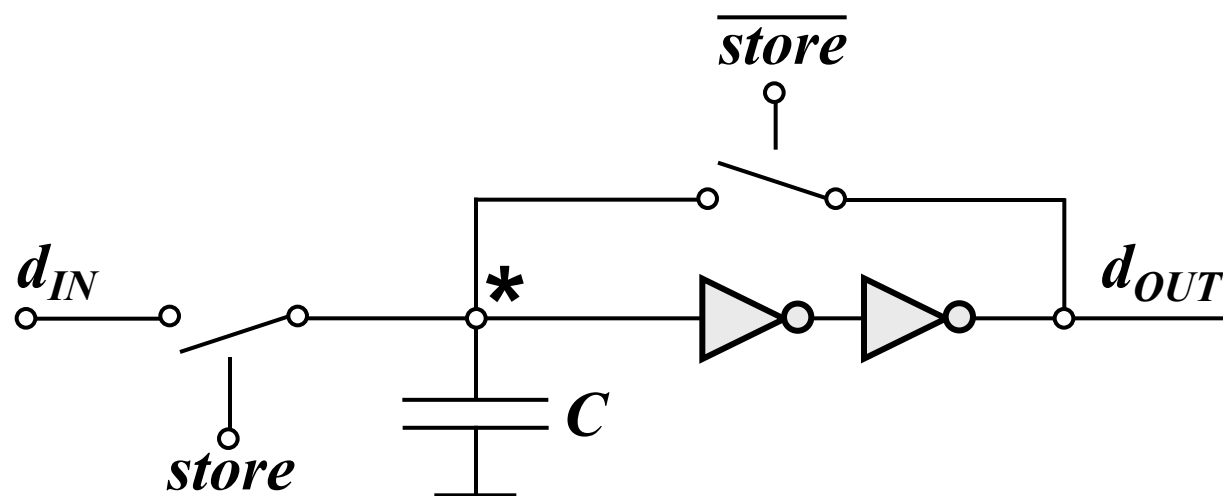
$$R_{IN} \gg R_L$$

Better, but still not perfect.

# Digital Memory

## Building a memory element ...

③ Third attempt → buffer + refresh



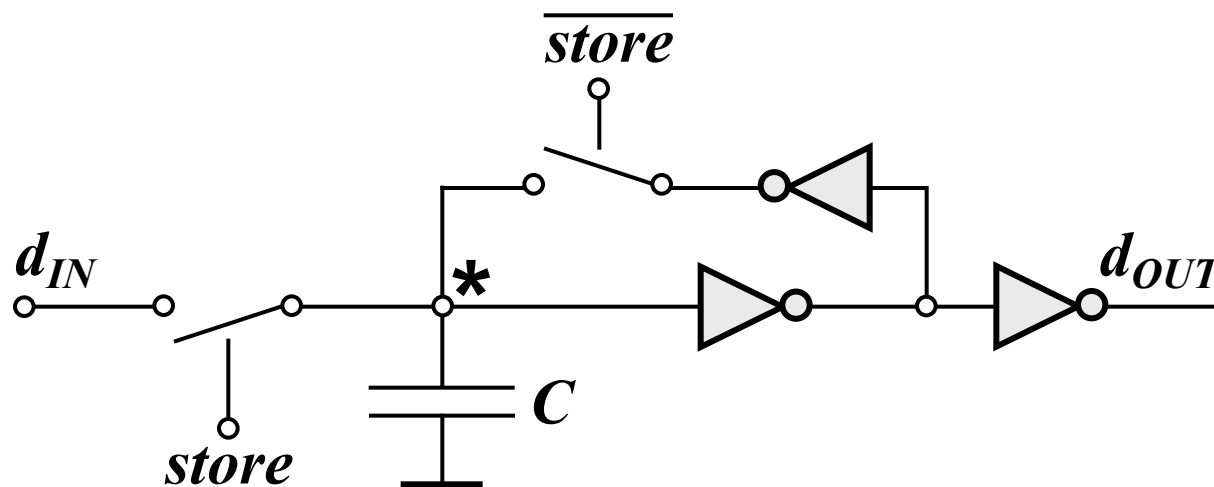
Does this work?

No. External value can influence storage node.

# Digital Memory

## Building a memory element ...

④ Fourth attempt → buffer + decoupled refresh

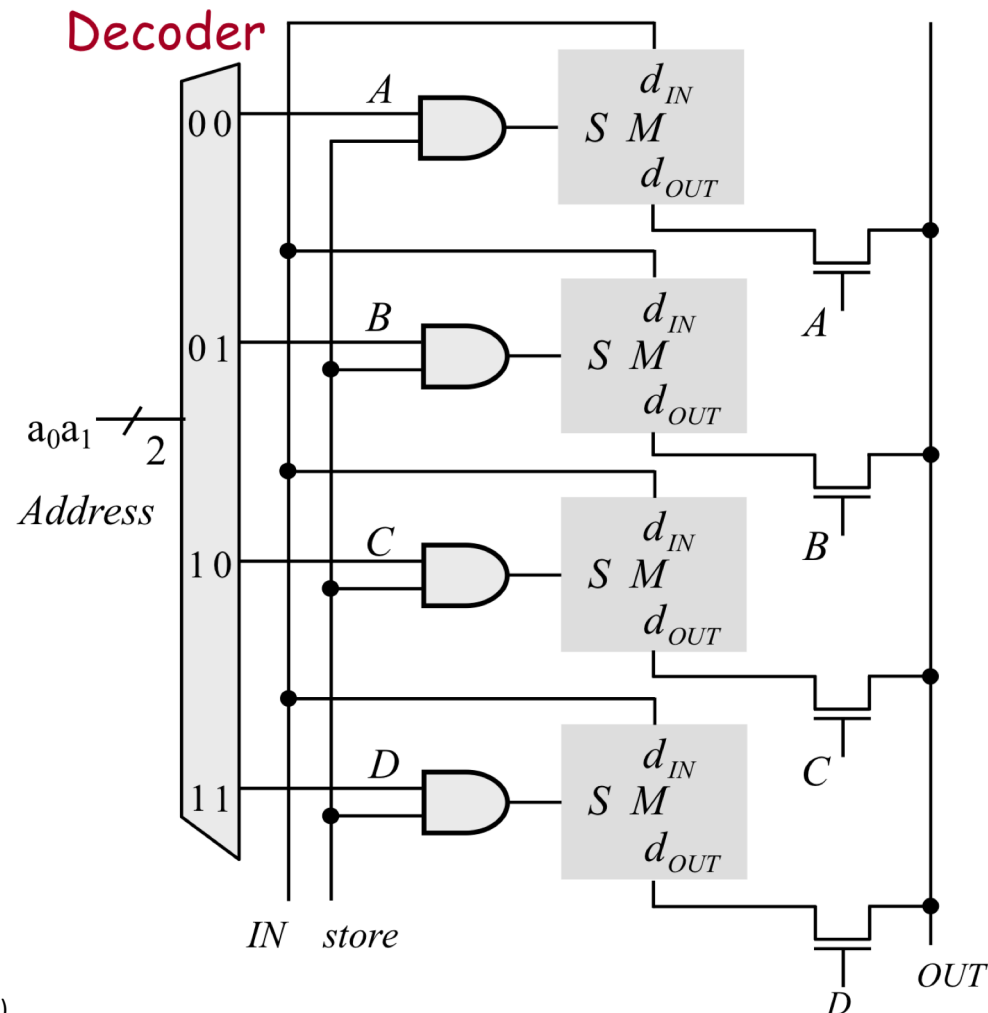
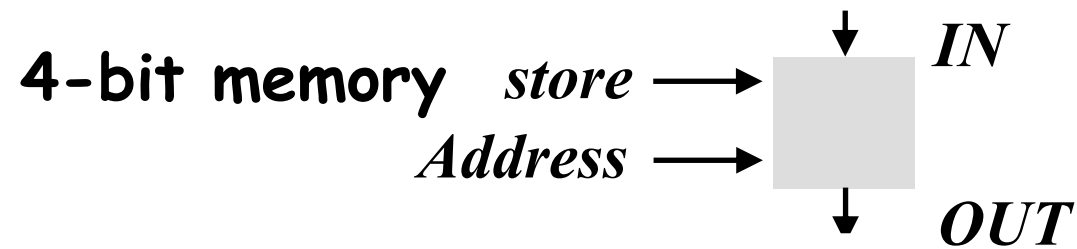


Works!

V-code: ???

# Digital Memory

## A Memory Array



# Digital Memory

## Truth table for decoder

$a_0$	$a_1$	$A$	$B$	$C$	$D$
0	0	1	0	0	0
0	1	0	1	0	0
1	0	0	0	1	0
1	1	0	0	0	1