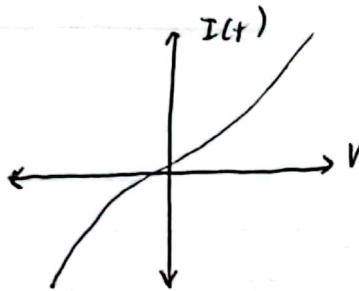
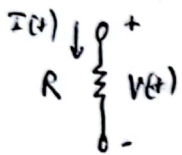


EE 536a 8/26/24 Notes

- Hashemi OH MW 9-10 am PHE 631
- Midterm is Oct. 28th
- Final is Dec 13th

Resistors



~~$i(t) = f(v(t))$~~
 ~~$i(t) = f(v(t))$~~
 $i(t) = f(v(t))$ Any function can be a resistor!

- Linear Resistor $\Rightarrow f = (\frac{1}{R})v(t)$

Linear System

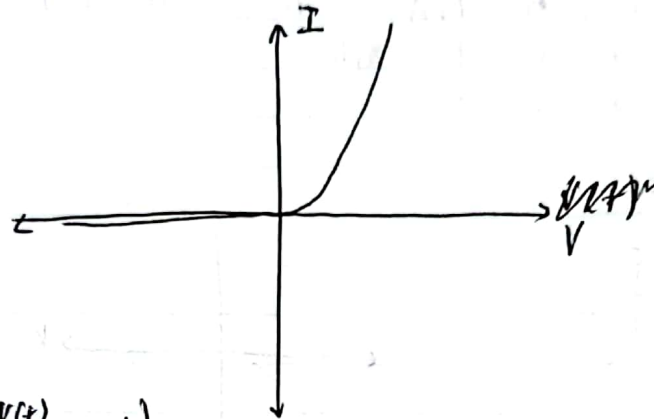
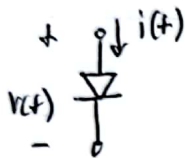


1. $ax(t) \rightarrow ay(t)$
 2. $\begin{cases} x_1(t) \rightarrow y_1(t) \\ x_2(t) \rightarrow y_2(t) \end{cases}$
- Scaling Property
 Superposition Property
- $x_1 + x_2 \rightarrow y_1 + y_2$

Example

$$\begin{cases} x_1 \rightarrow 2x_1 + 1 \\ x_2 \rightarrow 2x_2 + 1 \end{cases} \quad x_1 + x_2 \rightarrow 2(x_1 + x_2) + 1 \quad \text{Not linear!}$$

Diode - nonlinear!



$$i(t) = I_s \left(e^{\frac{v(t)}{nV_T}} - 1 \right)$$

$$V_T = \frac{kT}{q} \approx 25.8 \text{ mV @ room temp.}$$

k = Boltzmann const.

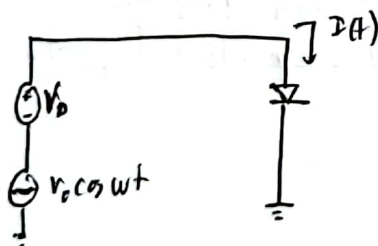
T = absolute temp.

q = electron charge

n = non-ideality const. Ideally, $n=1$

I_s = saturation current (proportional to diode area)

- This class only deals with linear components.



$$I(t) = I_s \left[e^{\frac{V_0 + V_0 \cos \omega t}{nV_T}} - 1 \right]$$

$$= I_s e^{\frac{V_0}{nV_T}} e^{\frac{V_0}{nV_T} \cos \omega t} - I_s$$

Assumption

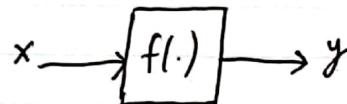
$$\frac{V_0}{nV_T} \ll 1 \quad (\text{factor of } 10)$$

$$e^x = 1 + x + \frac{1}{2!} x^2 + \frac{1}{3!} x^3 + \dots$$

$$\text{if } x \ll 1 \Rightarrow e^x \approx 1 + x$$

$$\begin{aligned}
 \overset{\substack{V_0 \\ nV_T} \ll 1}{I(t)} &\approx I_s \left[e^{\frac{V_0}{nV_T}} \left(1 + \frac{V_0}{nV_T} \cos \omega_0 t \right) - 1 \right] \\
 &= \underbrace{I_s \left[e^{\frac{V_0}{nV_T}} - 1 \right]}_{\substack{I_{DC} \\ = I(V_0 \cos \omega_0 t) = 0}} + \underbrace{\left(\frac{I_s e^{\frac{V_0}{nV_T}}}{nV_T} \right)}_{\substack{g_m \\ \text{transconductance} \\ = \text{linear!}}} (V_0 \cos \omega_0 t) \quad \text{AC component}
 \end{aligned}$$

This process of linearizing a non-linear function is referred to as small signal analysis assuming the input is very small.

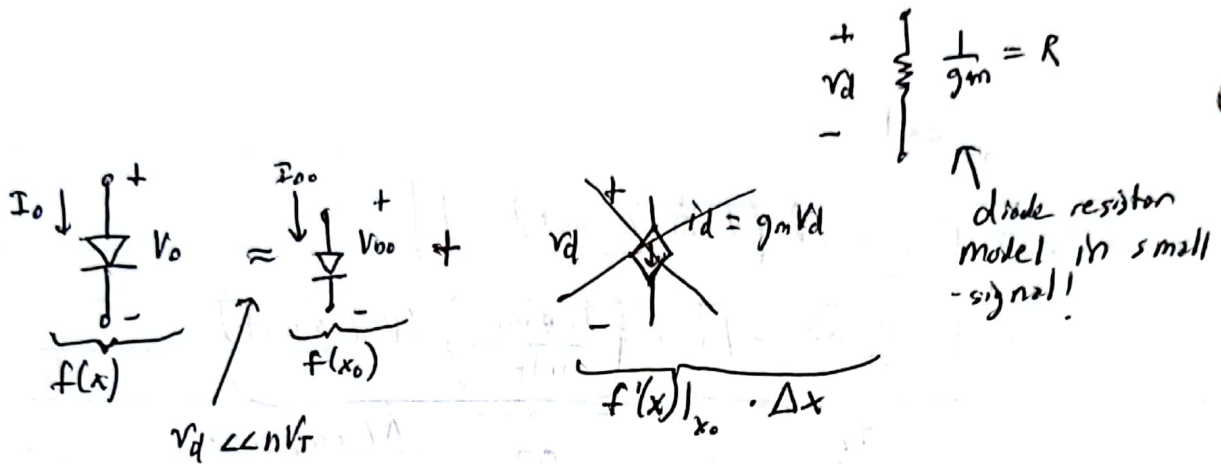


$$\begin{aligned}
 x &= x_0 + \Delta x, \quad \Delta x \ll x_0 \\
 f(x) &= f(x_0 + \Delta x) \\
 &= f(x_0) + f'(x) \big|_{x=x_0} \Delta x + \frac{1}{2!} f''(x) \big|_{x=x_0} (\Delta x)^2 \approx f(x_0) + f'(x) \big|_{x=x_0} \Delta x \\
 &\quad \text{Taylor Series Expansion} \quad \quad \quad \text{if } \Delta x = \text{small.}
 \end{aligned}$$

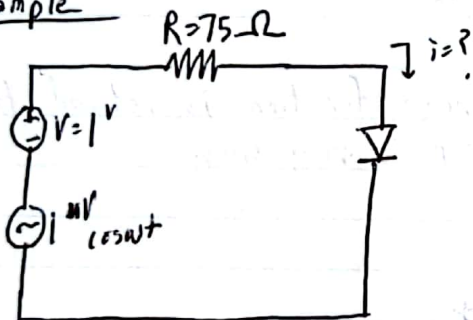
$$\begin{array}{c} \text{+} \\ \downarrow \\ \text{diode symbol} \end{array} \quad \underbrace{V_0}_{DC} + \underbrace{V_d}_{\substack{AC \\ \text{small signal}}} = V_0$$

$$I_D \approx \underbrace{I_s \left(e^{\frac{V_{D0}}{nV_T}} - 1 \right)}_{f(x_0)} + \left. \frac{dI_D}{dV_D} \right|_{V_{D0}} V_d$$

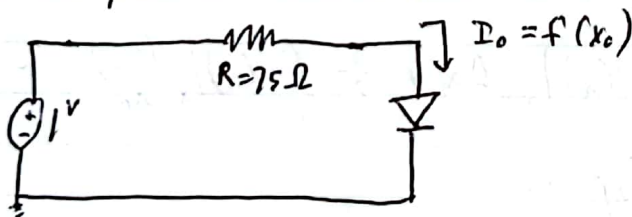
$$\begin{aligned}
 \left. \frac{dI_D}{dV_D} \right|_{V_{D0}} &= \frac{I_s}{nV_T} e^{\frac{V_{D0}}{nV_T}} \bigg|_{V_{D0}} = \frac{I_s}{nV_T} e^{\frac{V_{D0}}{nV_T}} \quad \text{simplified linearized diode eq.} \\
 &\approx \frac{I_s \left(e^{\frac{V_{D0}}{nV_T}} - 1 \right)}{nV_T} = \frac{I_{D0}}{nV_T} \triangleq g_m
 \end{aligned}$$



Example



1. Solve DC problem - assume $AC = 0$



for DC, have to solve nonlinear eq.!

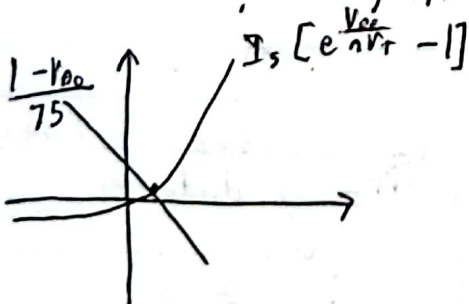
$$I_{D0} = I_s [e^{\frac{V_{D0}}{nV_T}} - 1] \text{ A}$$

$$I_{D0} = \frac{1V - V_{D0}}{75}$$

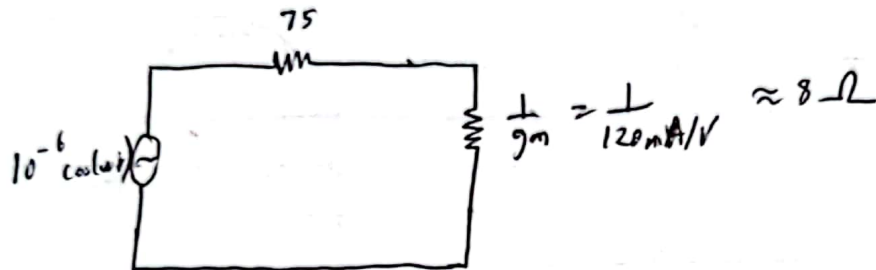
$$\rightarrow \boxed{I_{D0} = 3mA} \quad (Approx)$$

$$V_{D0} = 0.7V$$

How to solve? Use graphical methods!



2. AC analysis



$$g_m = \frac{I_{D0}}{n V_T} = \frac{3 \text{ mA}}{25 \text{ mV}} = 120 \text{ mA/V}$$

$V_T = 25 \text{ mV}$
 $n = 1$

$$i_d = \frac{10^{-6} \cos(\omega t)}{75 + 8}$$

$$I_D = I_{D0} + i_d$$

Transistor



$$I_{DS} = f[V_D, V_S, V_G, V_B]$$

NFET

n-type Field Effect Transistor channel

