

# LECTURE 4

#### **❖** Difference between circuit and Network

- An electrical network is an interconnection of electrical elements such as resistors, inductors, capacitors, transmission lines, voltage and current sources, switches etc.
- An electrical circuit is a network that has a closed loop, giving a return path for the current.
- All the circuits must be network but all the networks may or may not be circuit.

## **Difference between Hypothesis, Theory and Law**

- A hypothesis is a limited explanation of a phenomenon. It is a reasonable guess based on something that you observe in the natural world.
- A scientific theory is an in-depth empirical explanation of the observed phenomenon. A scientific theory consists of one or more hypotheses that have been supported by repeated testing.
- A theorem is a mathematical statement for the phenomena that has been proved, or can be proved.
- A law is a statement about an observed phenomenon and generally it rely on a concise mathematical equation used to describe an action under certain circumstances.

# > Superposition Theorem

- ✓ The basic principle of superposition states that if the effect produced in a system is directly proportional to the cause, then the overall effect produced in the system, due to a number of causes acting jointly, can be determined by superposing (adding) its effects of each source acting separately.
- ✓ The superposition principle is only applicable to linear networks and systems. A linear network comprises of independent sources, linear dependent sources and linear passive elements like resistor, inductor, capacitor and transformer.

A device is called linear if it is characterized by an equation of the form y = mx, where m is a constant and not a function of x.

Thus,  $y = x^3$  is a nonlinear equation because in this case,  $m = x^2$  is a function of the independent variable x.

Let the sources  $v_1$  and  $v_2$  are applied to a linear network with zero initial conditions. If  $v_1$  gives  $x_1$  and  $v_2$  gives  $x_2$ , then  $(v_1 + v_2)$  gives  $(x_1 + x_2)$ 

✓ In a linear network with several sources (which include the equivalent sources due to initial conditions) the overall response at any point in the network is equal to the sum of individual response of each source, considered separately, the other sources being made inoperative.

# > Superposition Theorem

## **✓ Procedure of Superposition Theorem**

The following steps are required to find the response in a particular branch using superposition theorem.

- **Step 1** Find the response in a particular branch of a network by considering one independent source and making inoperative of other sources by
  - (a) short-circuiting the voltage sources and replacing them by their series impedance and
  - (b) open-circuiting the current sources and substituting them by their shunt impedances.
- **Step 2** Repeat Step 1 for all independent sources present in the network.
- Step 3 Add all the responses in order to get the overall response in a particular branch when all independent sources are present in the network.

## > Superposition Theorem

# **✓** Explanation

- Let us consider any linear network containing bilateral linear impedances and energy sources as shown in **Fig. 2.1**.
- Show that the current flowing in any element is the algebraic sum of the currents that are separately caused to flow in that element by each energy source.

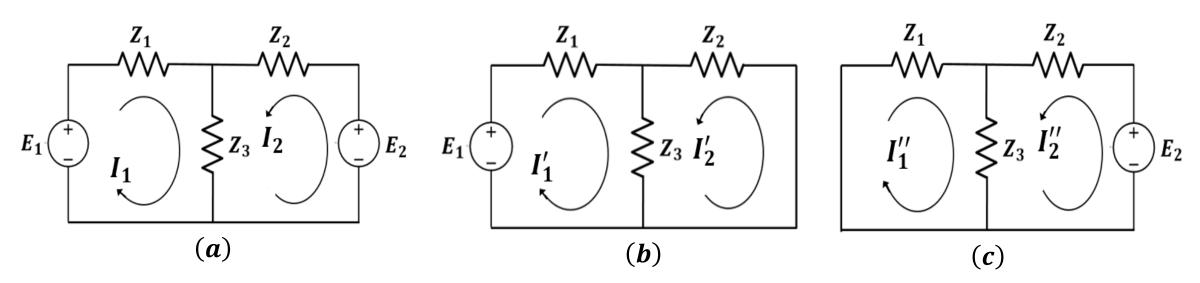


Fig. 2.1. Circuits for Superposition Theorem

# > Superposition Theorem

# **✓** Explanation

Consider Fig. 1(a)

$$E_1 = I_1(Z_1 + Z_3) + I_2Z_3$$

$$E_2 = I_1 Z_3 + I_2 (Z_2 + Z_3)$$

Solving

$$I_1 = \frac{Z_2 + Z_3}{Z_1 Z_2 + Z_2 Z_3 + Z_3 Z_1} E_1 - \frac{Z_3}{Z_1 Z_2 + Z_2 Z_3 + Z_3 Z_1} E_2$$

$$I_2 = \frac{-Z_3}{Z_1 Z_2 + Z_2 Z_3 + Z_3 Z_1} E_1 + \frac{Z_1 + Z_3}{Z_1 Z_2 + Z_2 Z_3 + Z_3 Z_1} E_2$$

# > Superposition Theorem

# **✓** Explanation

Making  $E_2$  inoperative as in **Fig. 1(b)** 

$$E_1 = I_1' (Z_1 + Z_3) + I_2' Z_3$$

$$0 = I_1' Z_3 + I_2' (Z_2 + Z_3)$$

Solving

$$I_1' = \left[ \frac{Z_2 + Z_3}{Z_1 Z_2 + Z_2 Z_3 + Z_3 Z_1} \right] E_1$$

$$I_2' = \left[ \frac{-Z_3}{Z_1 Z_2 + Z_2 Z_3 + Z_3 Z_1} \right] E_1$$

# > Superposition Theorem

# **✓** Explanation

Now, making  $E_1$  inoperative as in **Fig. 1(c)** 

$$0 = I_1^{"}(Z_1 + Z_3) + I_2^{"}Z_3$$

$$E_2 = I_1^{"} Z_3 + I_2^{"} (Z_2 + Z_3)$$

Solving

$$I_1^{"} = \left[ \frac{-Z_3}{Z_1 Z_2 + Z_2 Z_3 + Z_3 Z_1} \right] E_2$$

$$I_2^{"} = \left[ \frac{Z_1 + Z_3}{Z_1 Z_2 + Z_2 Z_3 + Z_3 Z_1} \right] E_2$$

# > Superposition Theorem

## **✓** Explanation

Now, 
$$I_1 = I_1' + I_1''$$
 and  $I_2 = I_2' + I_2''$ 

Hence, the superposition theorem is proved.

#### **✓** Definition

The superposition theorem for electrical circuit states that for a linear system the response (voltage or current) in any branch of a linear bilateral circuit having more than one independent source equals the algebraic sum of the responses caused by each independent source acting alone, where all the other independent sources are replaced by their internal impedances

# > Superposition Theorem

# **✓** Explanation

Consider a linear network N having l independent loops. The loop equations are

$$[Z][I] = [E] \longrightarrow (1)$$

where the order of Z is  $(l \times l)$ , I is  $(l \times 1)$  and E is  $(l \times 1)$ .

The solution of equation can be written as

$$[I] = [Z]^{-1} [E] = [Y] [E] \longrightarrow (2)$$

$$: I_i = \sum_{j=1}^l Y_{ij} E_j \quad i = 1, 2, \dots \dots l \quad ----- (3)$$

# > Superposition Theorem

## **✓** Explanation

Applying the superposition theorem making  $E_j = 0$  for all j except j = k and for k = 1, 2, ..., l and adding the individual responses

$$I_i = \sum_{k=1}^{l} Y_{ik} E_k; \quad i = 1, 2, \dots l$$
 which is same as Eq. (3)

First assume all  $E_j$  except  $E_1$  to be zero. This will give the current in loop 1 as  $Y_{11}E_1$ , in loop 2 as  $Y_{21}E_1$  etc. Similarly, when all the sources except  $E_2$  are made inoperative, the current in loop 1 is  $Y_{12}E_2$ , loop 2 is  $Y_{22}E_2$  and so on. The procedure is repeated in turn for the remaining sources  $E_3$ ,  $E_4$ , ... ...  $E_l$  and the corresponding loop responses determined. On applying the superposition principle, the individual responses are added together to give the resulting response.

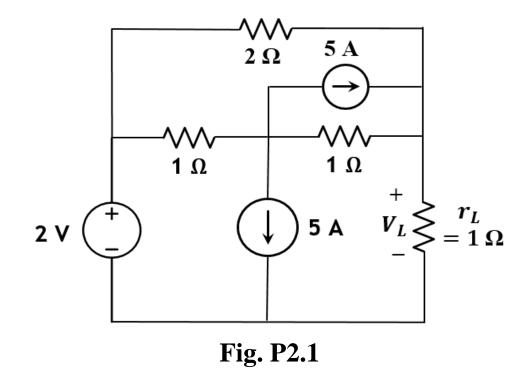
For the first loop,  $I_1 = Y_{11}E_1 + Y_{12}E_2 + \cdots + Y_{1l}E_l$ 

Which is the same as that obtained by solving the simultaneous equations.

## > Superposition Theorem

## Example – P2.1

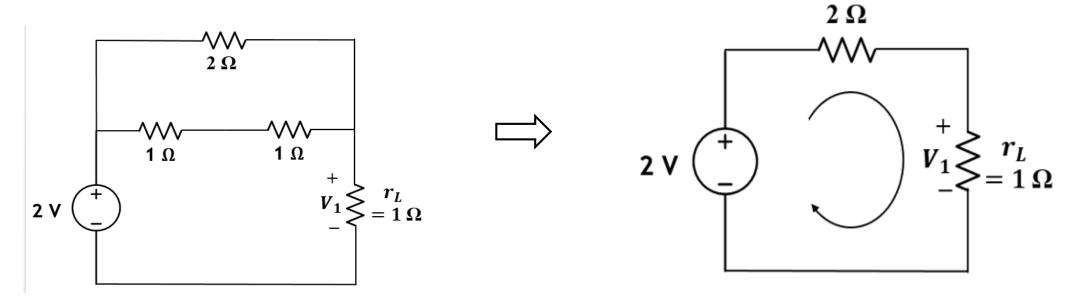
Find  $V_L$  in the circuit as shown in **Fig. P2.1** using Superposition Theorem.



## > Superposition Theorem

## **Solution of Example – P2.1**

Let us first take the 2 V source deactivating the current sources.



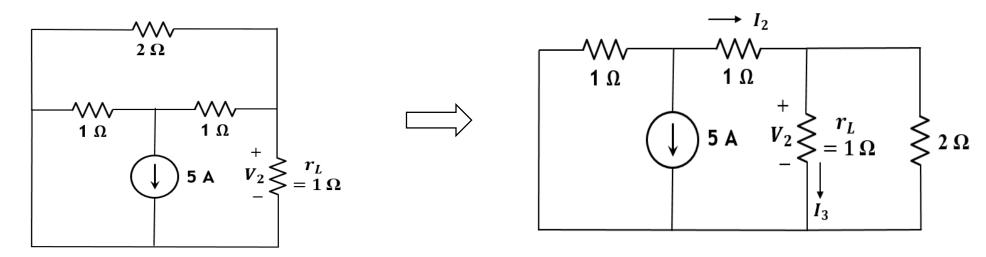
$$i_1 = \frac{2}{\frac{2 \times 2}{2 + 2}} + 1 = 1 A$$

 $V_1(drop\ across\ r_L\ due\ to\ 2\ V\ source) = 1 \times 1\ V = 1\ V$ 

## > Superposition Theorem

#### **Solution of Example – P2.1**

Next taking the lower current source only



$$i_{2} = (-5) \times \frac{1}{1+1+\frac{2}{3}} = (-5) \times \frac{3}{8} = -\frac{15}{8} A$$

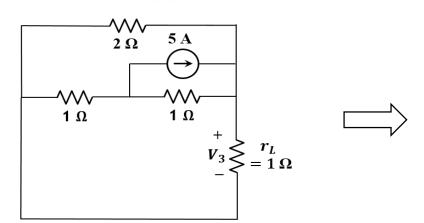
$$i_{3} = (-5) \times \frac{1}{1+1+\frac{2}{3}} \times \frac{2}{2+1} = (-5) \times \frac{3}{8} \times \frac{2}{3} = -\frac{5}{4} A$$

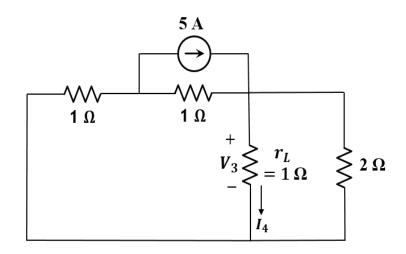
$$\therefore V_{2} = -\frac{5}{4} \times 1 = -\frac{5}{4} V$$

## > Superposition Theorem

## **Solution of Example – P2.1**

Next taking the upper current source only





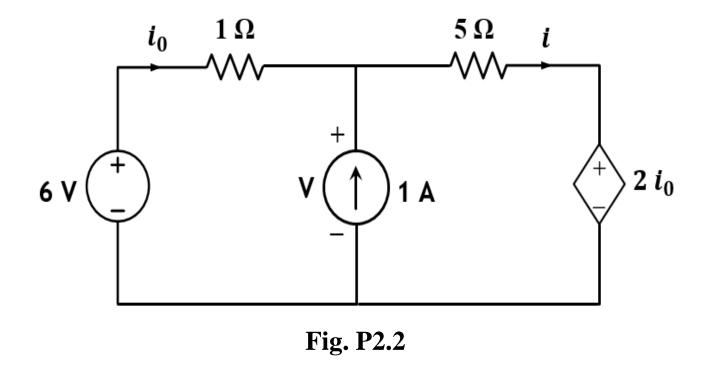
$$i_4 = 5 \times \frac{1}{1 + \frac{2}{3}} \times \frac{2}{2 + 1} = 5 \times \frac{3}{5} \times \frac{2}{3} = 2 A$$
  

$$\therefore V_3 = 2 \times 1 = 2 V \qquad \therefore V = V_1 + V_2 + V_3 = 1 + \left(-\frac{5}{4}\right) + 2 = \frac{7}{4} V$$

## > Superposition Theorem

## Example – P2.2

Find  $i_0$  and i from the circuit as shown in **Fig. P2.2** using Superposition Theorem.



## > Superposition Theorem

## **Solution of Example – P2.2**

Assuming only 6 V source to be active, with reference to Fig. P2.2 (a)

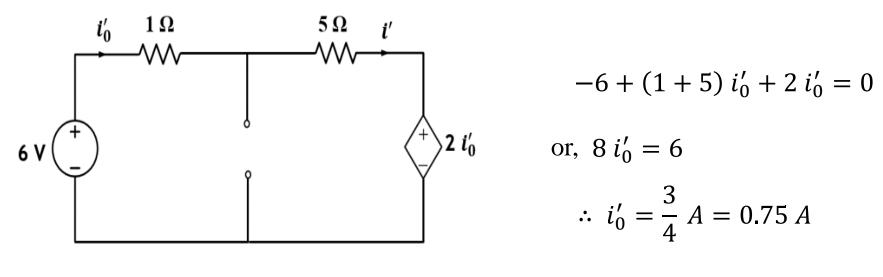


Fig. P2.2 (a)

## > Superposition Theorem

## **Solution of Example – P2.2**

Next, assuming 1 A source active only, with reference to Fig. P2.2 (b)

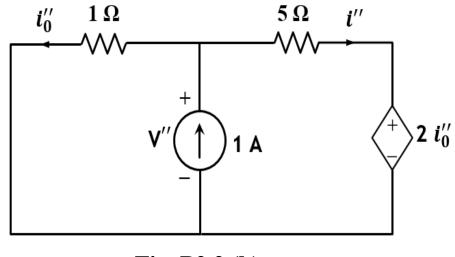


Fig. P2.2 (b)

Using the principle of Superposition,

$$1 = i_0'' + i'' = \frac{V''}{1} + \frac{V'' - 2i_0''}{5} = 1.2 \, V'' - 0.4 \, i_0''$$

But, 
$$i_0^{\prime\prime} = \frac{v^{\prime\prime}}{1}$$

We finally get  $1 = 1.2 i_0'' - 0.4 i_0'' = 0.8 i_0''$ 

$$i_0'' = \frac{1}{0.8} = 1.25 A$$

$$i'' = \frac{V'' - 2i_0''}{5} = \frac{-i_0''}{5} = -0.25 A$$

$$i_0 = i'_0 + i''_0 = (0.75 - 1.25) A = -0.5 A$$

and 
$$i = i'_0 + i''_0 = (0.75 - 0.25) A = 0.5 A$$

## > Reciprocity Theorem

- ✓ A linear network is said to be reciprocal or bi-lateral, if it remains invariant due to the interchange of position of cause and effect in the network.
  - Consider two loops A and B of a network N.
  - An ideal voltage source E in loop A produces a current I in loop B.
  - An identical source in loop B produces the same current I in loop A.
  - The network is said to be reciprocal. The dual is also true.
- ✓ A reciprocal network comprises of linear time invariant bi-lateral passive elements.
  - It is applicable to resistors, capacitors, inductors (with and without coupling) and transformers.
  - Both dependent and independent sources are not permissible. The sources with zero state response are considering only.

# > Reciprocity Theorem

#### **✓** Proof

Let us consider a network N having only one driving voltage source  $E = E_k$  in loop K. So, the current response in loop m due to the voltage source is

$$I_m = Y_{mk} E_k$$

Now, interchange the positions of cause and effect. The same voltage source  $E = E_m$  is placed in loop m and hence the current response in loop k is

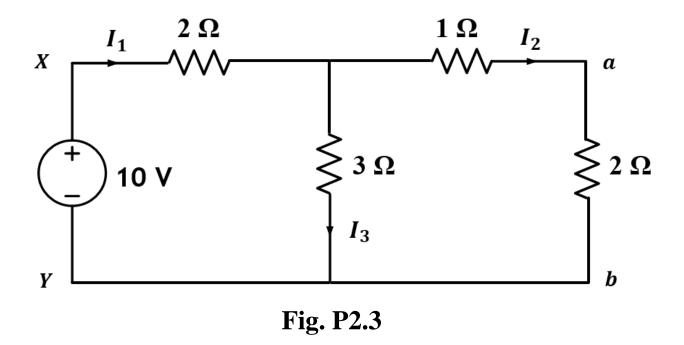
$$I_k = Y_{km}E_m$$

- $I_k$  will be equal to  $I_m$  provided  $Y_{km} = Y_{mk}$ .
- This is the condition for reciprocity.
- The admittance matrix Y is symmetric as  $Y_{km} = Y_{mk}$  for all m and k.

# > Reciprocity Theorem

## Example – P2.3

Show the application of reciprocity theorem in the network as shown in Fig. P2.3



# > Reciprocity Theorem

## **Solution of Example – P2.3**

The equivalent resistance across X-Y of the circuit as shown in **Fig. P2.3**, is given by

$$R_{eq} = [(2+1) \parallel 3+2] = 3.5 \,\Omega$$

$$I_1 = \frac{10}{3.5} = 2.86 A$$

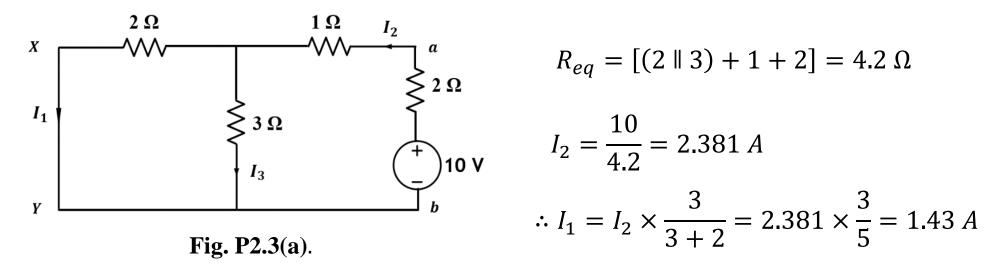
$$I_2 = 2.86 \times \frac{3}{3+3} = 1.43 A$$

$$I_3 = 2.86 - 1.43 = 1.43 A$$

# > Reciprocity Theorem

## **Solution of Example – P2.3**

Now, consider the voltage source is changed in position and placed in series with 2  $\Omega$  resistance as shown in **Fig. P2.3(a)**.



Hence, we observe that when the source is in branch X - Y of **Fig. P2.3**, the a - b branch current is 1.43 A. Again, when the source is in branch a - b of **Fig. P2.3(a)**, the X - Y branch current becomes 1.43 A.

Hence, this proves the reciprocity theorem.

# LECTURE 5

#### > Thevenin's Theorem

- ✓ **Definition:** Any two terminal linear network containing impedances and energy sources may be replaced by an independent voltage source of generated voltage  $V_g$  and internal impedance  $Z_g$ .
  - $V_g$  is the open circuit voltage at the terminals and  $Z_g$  is the impedance viewed at the terminals when all the independent energy sources are replaced by their internal impedances.
- ✓ The theorem is commonly known as Thevenin's theorem, in honor of Leon Charles Thevenin, a French telegraph engineer.

#### ✓ Proof:

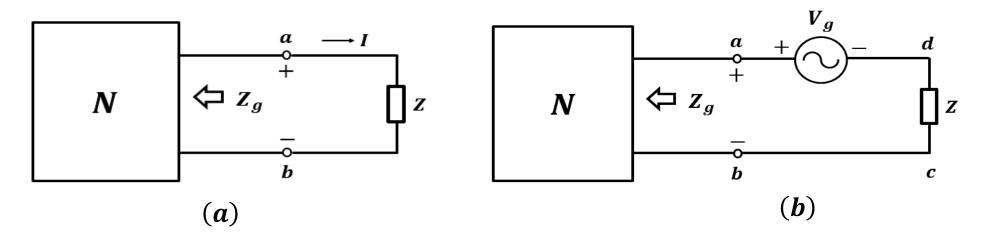


Fig. 2.2. Circuits for Thevenin Theorem

#### > Thevenin's Theorem

- $\checkmark$  A linear network N containing impedances and energy sources. It is connected to a load impedance Z at the terminals a, b.
- $\checkmark$  The open circuit voltage across a, b is  $V_g$  when Z is removed.
- $\checkmark$  Let the voltage source  $V_q$  be connected in series with Z with polarities as shown in **Fig. 2.2.**
- $\checkmark$  The net voltage in the loop *abcd* will now be zero for all values of Z. The current through Z is thus zero under these conditions.
- $\checkmark$  If I is the current supplied by the network and I' is the current supplied by the added source to Z, then

$$I + I' = 0 \longrightarrow (1)$$

 $\checkmark$  By superposition theorem, the current supplied by the added source  $V_g$  is given by

$$I' = -\frac{V_g}{Z + Z_g} \longrightarrow (2)$$

where  $Z_g$  is the impedance viewed at the terminals a, b when all the independent energy sources within the network are replaced by their internal impedances.

24

#### > Thevenin's Theorem

 $\checkmark$  We get from Eq. (2)

$$I = \frac{V_g}{Z + Z_g}$$

$$\therefore V = ZI = V_g - Z_gI \longrightarrow (3)$$

Here, V is the voltage across Z.

 $\checkmark$  Eq. (3) shows that with respect to the terminals a, b the circuit of Fig. 2.2(a) may be replaced by the equivalent circuit of Fig. 2.3.

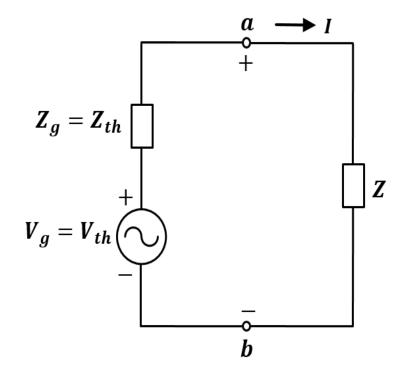


Fig. 2.3. Thevenin Equivalent Circuit

✓ It is seen that the network to the left of the terminals a, b is replaced by an independent voltage source of generated voltage  $V_g$  and internal impedance  $Z_g$ .

 $V_g$  is the open circuit voltage at a, b and  $Z_g$  is the impedance measured back at a, b with all the internal independent energy sources deactivated and replaced by their impedances.

The voltage source in **Fig. 2.3** is called the equivalent Thevenin Source and impedance in **Fig. 2.3** is called the Thevenin Impedance.

25

#### > Thevenin's Theorem

#### **✓** Illustration

Find the current  $I_3$ , flowing through the load having impedance of  $Z_3$  as shown in **Fig I1(a)** by Thevenin Theorem.  $Z_1$   $Z_2$ 

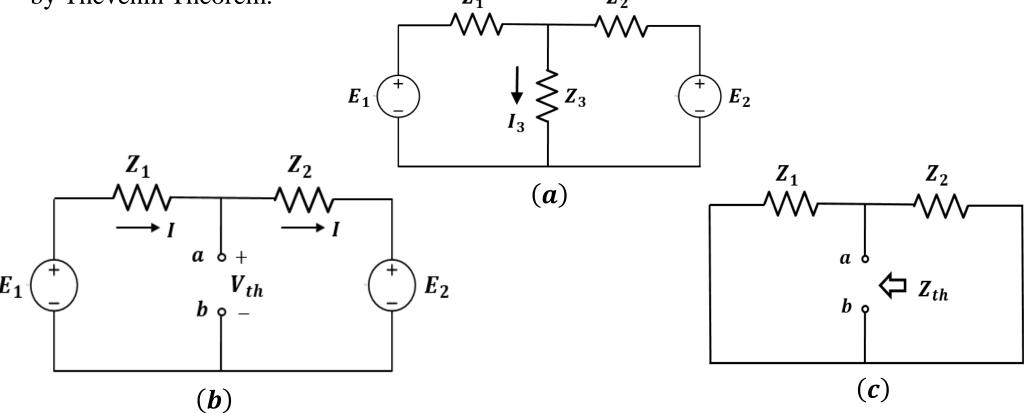


Fig. I1. Circuit for illustration of Thevenin Theorem

#### > Thevenin's Theorem

#### **✓** Illustration

From **Fig. I1(b)**, we get 
$$I = \frac{E_1 - E_2}{Z_1 + Z_2}$$

From **Fig. I1(c)**, we get

$$Z_{th} = \frac{Z_1 Z_2}{Z_1 + Z_2}$$

$$\therefore I_3 = \frac{V_{th}}{Z_{th} + Z_3}$$

$$= \frac{Z_2}{Z_1 Z_2 + Z_2 Z_2 + Z_2 Z_4} E_1 + \frac{Z_1}{Z_1 Z_2 + Z_2 Z_2 + Z_2 Z_4} E_2$$

From **Fig. I1(b)**, we get

$$V_{ab} = E_1 - IZ_1$$

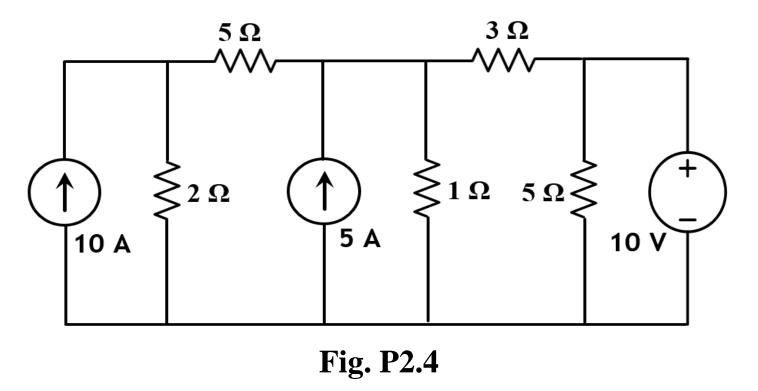
or, 
$$V_{ab} = E_1 - \frac{E_1 - E_2}{Z_1 + Z_2} \cdot Z_1$$

$$V_{th} = V_{ab} = \frac{Z_2}{Z_1 + Z_2} E_1 + \frac{Z_1}{Z_1 + Z_2} E_2$$

#### > Thevenin's Theorem

# Example – P2.4

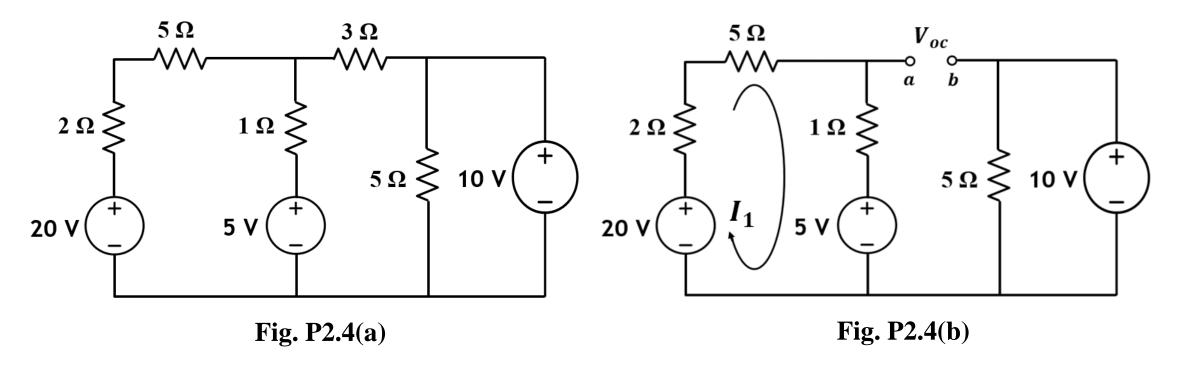
Find the current in the 3  $\Omega$  resistor of the circuit as shown in **Fig. P2.4** 



#### > Thevenin's Theorem

## **Solution of Example – P2.4**

The circuit as shown in **Fig. P2.4(a)** is obtained after transforming current sources into voltage sources. The circuit as shown in **Fig. P2.4(b)** is obtained after removing 3  $\Omega$  resistor.



#### > Thevenin's Theorem

#### **Solution of Example – P2.4**

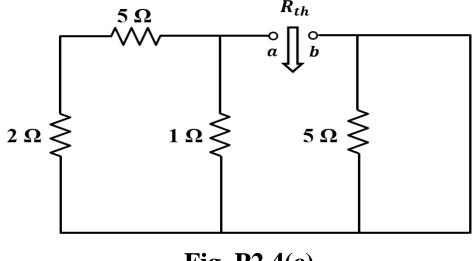
The circulating current  $(I_1)$  in the left most loop is then given by

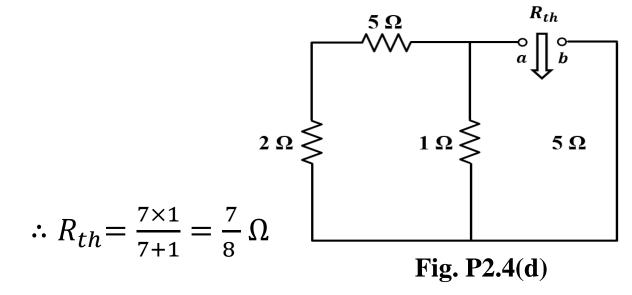
$$I_1 = \frac{20-5}{2+5+1} = \frac{15}{8} A$$

$$I_1 = \frac{20 - 5}{2 + 5 + 1} = \frac{15}{8} A \qquad \therefore V_{oc} = \therefore V_{th} = 20 - \frac{15}{8} (2 + 5) - 10 V = -3.125 V$$

To find the Thevenin's resistance of the given circuit, the sources are deactivated as shown in **Fig.** 

**P2.4(a)** and **Fig. P2.4(d)**.

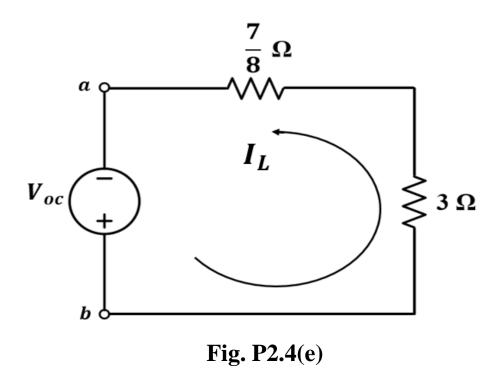




#### > Thevenin's Theorem

#### **Solution of Example – P2.4**

The Thevenin's Equivalent Circuit is given in **Fig. P2.4(e)** 



The current through 3  $\Omega$  resistor is given by

$$I_L = \frac{V_{oc}}{\frac{7}{8} + 3} = \frac{3.125}{\frac{7}{8} + 3} = 0.806 \, A$$

#### > Norton's Theorem

- ✓ **Definition:** Any two terminal linear network containing impedances and energy sources may be replaced by an independent current source  $I_g$  in parallel with an admittance  $Y_g$ .
  - $I_g$  is the short-circuit current between the terminals a, b and  $Y_g$  is the admittance viewed at the terminals a, b when all the independent energy sources are replaced by their internal admittances.
- ✓ Norton's theorem was independently derived in 1926 by Siemens & Halske researcher Hans Ferdinand Mayer (1895–1980) and Bell Labs engineer Edward Lawry Norton.

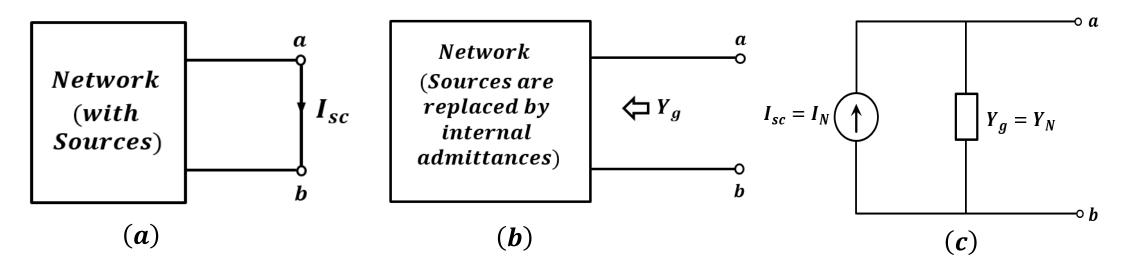


Fig. 2.3. Circuits for Norton Theorem

#### > Norton's Theorem

✓ **Definition:** Any two terminal linear network containing enery sources and resistances is equivalent to a constant current source and a parallel resistance when viewed from its output terminals.

The constant current is equal to the current which flows in a short-circuit placed across the terminals a, b and parallel resistances is the resistance of the network when viewed from these open circuited terminals after all voltage and current sources have been removed and replaced by their internal resistances.

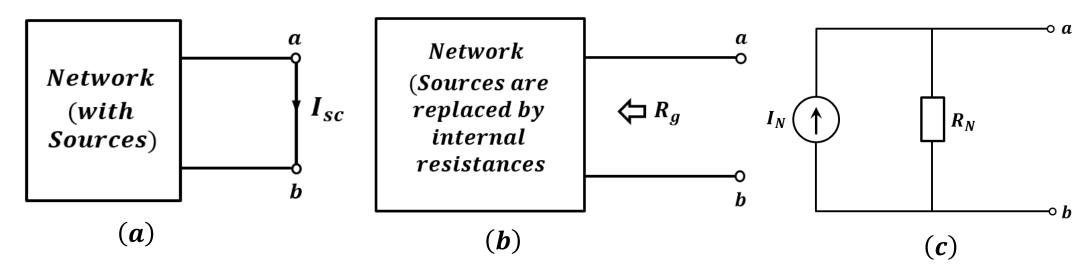


Fig. 2.4. Circuits for Norton Theorem

#### > Norton's Theorem

#### **✓** How to Nortonize a given circuit

The step-wise procedure of Norton theorem is given below:

- 1. Remove the resistance (if any) across the two given terminals and put a short-circuit across them.
- 2. Compute the short-circuit current  $I_{sc}$
- 3. Remove all voltage sources but retain their internal resistances(if any). Similarly, remove all current sources and replace them by open circuit i.e. by infinite resistance
- 4. Next, find the resistance  $R_i$  (also called  $R_N$ ) of the network as looked into from the given terminals. It is exactly the same as  $R_{th}$ .
- 5. The current source  $(I_{sc})$  joined in parallel across  $R_i$  between the two terminals gives Norton's Equivalent Circuit.

#### > Norton's Theorem

#### **✓** Illustration

Find the current  $I_3$ , flowing through the load having impedance of  $Z_3$  as shown in **Fig I1(a)** by Norton Theorem.  $Z_1$   $Z_2$ 

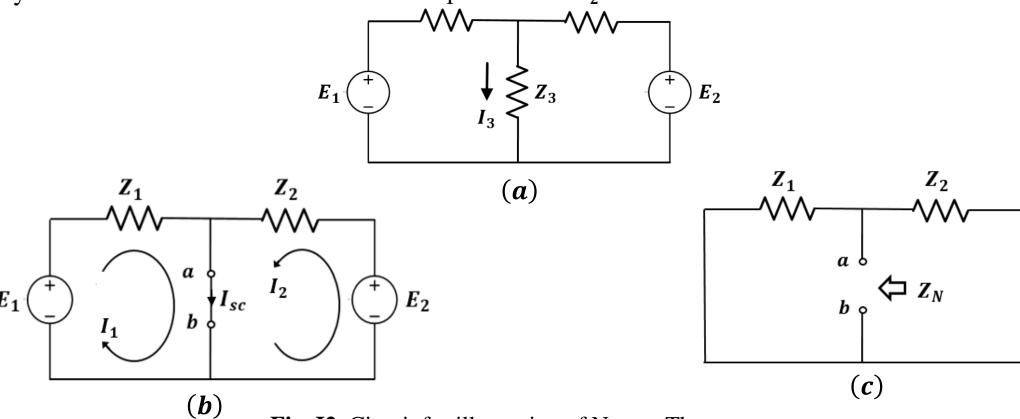


Fig. I2. Circuit for illustration of Norton Theorem

#### > Norton's Theorem

#### **✓** Illustration

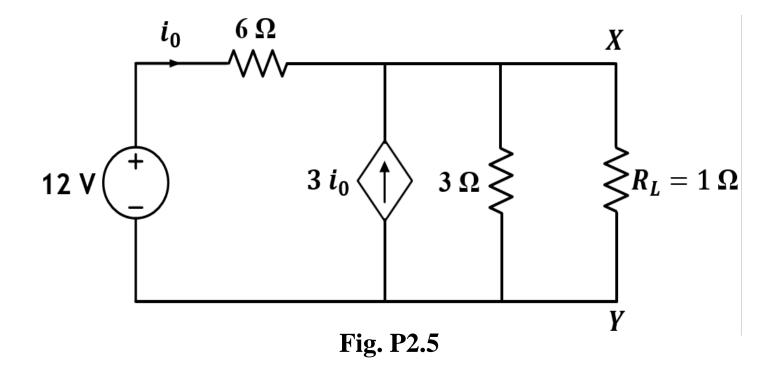
From **Fig. I2(b)**, we get 
$$I_{sc} = \frac{E_1}{Z_1} + \frac{E_2}{Z_2}$$

From **Fig. I2(c)**, we get 
$$Z_N = \frac{Z_1 Z_2}{Z_1 + Z_2}$$

#### > Norton's Theorem

# Example – P2.5

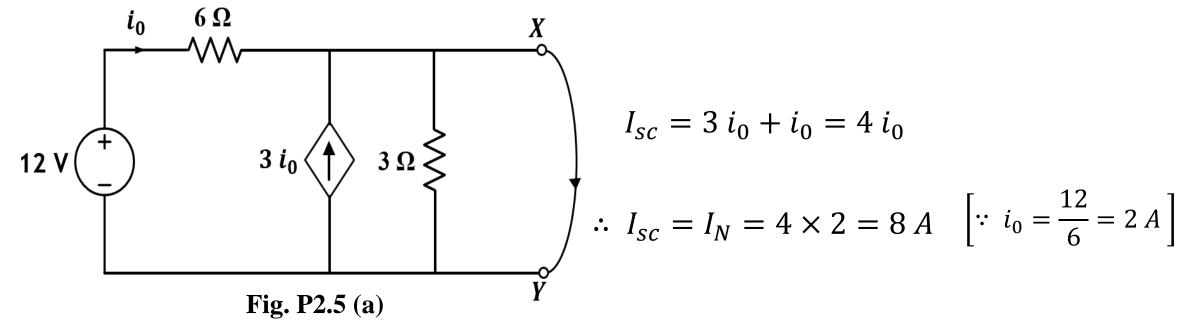
Find the current through  $R_L$  in the circuit of **Fig. P2.5** using Norton's Theorem.



#### > Norton's Theorem

# **Solution of Example – P2.5**

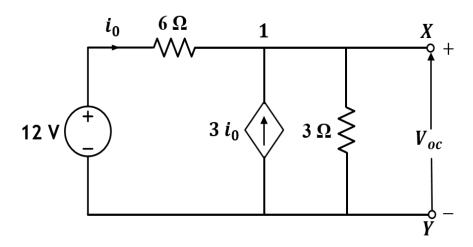
Let us first remove  $R_L$  from X-Y terminals and short X-Y as shown in Fig. P2.5 (a)



#### > Norton's Theorem

# **Solution of Example – P2.5**

Let us now remove the short circuit and the circuit is open circuited at X - Y as shown in Fig. P2.5 (b)



**Fig. P2.5** (b)

$$\therefore R_{int} = \frac{V_{oc}}{I_{sc}} = \frac{8}{8} = 1 \Omega$$

Nodal analysis at node 1 gives

$$i_0 + 3 i_0 - \frac{V_{oc}}{3} = 0$$

or, 
$$4i_0 - \frac{V_{oc}}{3} = 0$$

or, 
$$4 \times \left(\frac{12 - V_{oc}}{6}\right) - \frac{V_{oc}}{3} = 0$$

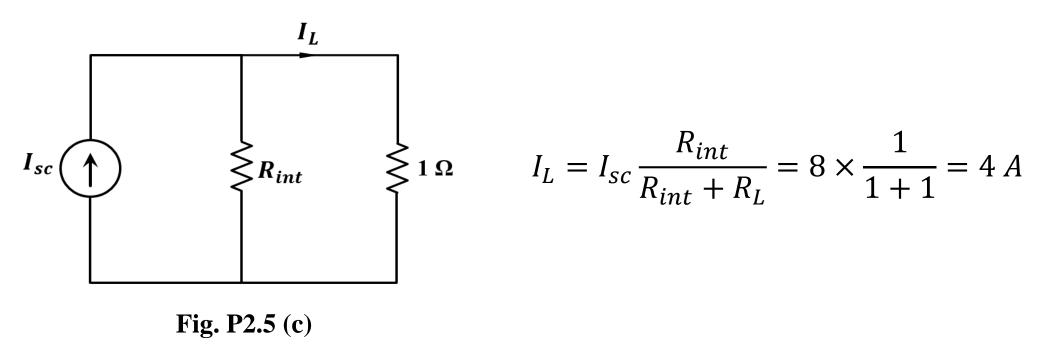
or, 
$$8 - \frac{2V_{oc}}{3} - \frac{V_{oc}}{3} = 0$$

$$V_{oc} = 8 V$$

#### > Norton's Theorem

# **Solution of Example – P2.5**

Norton's equivalent circuit is shown in Fig. P2.5 (c).



# LECTURE 6

#### > Maximum Power Transfer Theorem

The goal of 'Maximum Power Transfer Theorem' is either to receive or transmit maximum power but the overall efficiency of a network supplying maximum power is only 50%. The power output  $(P_{max})$  and efficiency  $(\eta)$  for maximum power

condition is shown in **Fig. 2.5**.

■ The application of the theorem for power transmission and distribution networks is limited.

• Although the efficiency is reduced, the theorem is particularly useful for analysing electronic and communication networks when power involved is only a few milliwatts or microwatts.

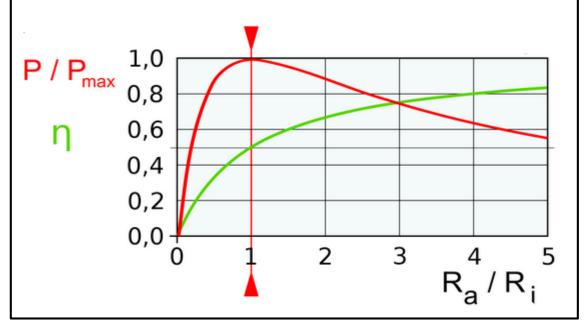


Fig. 2.5. Maximum output and efficiency

- > Maximum Power Transfer Theorem
- ✓ Condition of Maximum Power Transfer for different networks
  - 1. Purely resistive circuit, the load resistance is varied: Maximum power will be delivered from a voltage source to a load when the load resistance is equal to the internal resistance of the source.
  - 2. Reactance present, load resistance and reactance are independently varied:

    Maximum power will be delivered from a voltage source to a load when the load impedance is the complex conjugate of the source impedance.
  - 3. Reactance present, the magnitude but not the angle of the load impedance is varied: Maximum power is delivered from a voltage source to the load impedance when the magnitude of the load impedance is equal to the magnitude of source impedance.

#### > Maximum Power Transfer Theorem

- **✓** Condition of Maximum Power Transfer for different networks
  - i) Purely resistive circuit and the load resistance is variableProof:

A voltage source of generated voltage  $V_g$  and internal resistance  $R_g$  is connected to a load resistance R as shown in **Fig. 2.6.** 

The load current is

$$I = \frac{V_g}{R + R_g}$$

The power delivered to the load is

$$P = I^2 R = \left(\frac{V_g}{R + R_g}\right)^2 R$$

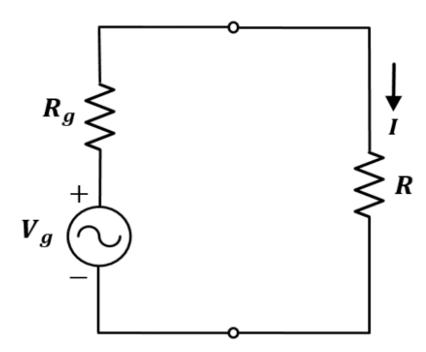


Fig. 2.6. Circuit having resistive load

- > Maximum Power Transfer Theorem
  - **✓** Condition of Maximum Power Transfer for different networks
    - i) Purely resistive circuit and the load resistance is variable

#### **Proof:**

The power is maximized by changing R. So,

$$\frac{dP}{dR} = 0$$
or, 
$$\frac{V_g^2 \left[ \left( R + R_g \right)^2 - 2R(R + R_g) \right]}{\left( R + R_g \right)^4} = 0$$
or, 
$$\left( R + R_g \right)^2 = 2R \left( R + R_g \right)$$

$$\therefore R = R_g$$

This is the condition for transmitting maximum power from source to load.

#### > Maximum Power Transfer Theorem

- **✓ Condition of Maximum Power Transfer for different networks** 
  - i) Purely resistive circuit and the load resistance is variable

The output power for the condition of  $R = R_g$  is

$$P_{out} = \left(\frac{V_g}{R + R_g}\right)^2 R = \left(\frac{V_g}{R + R}\right)^2 R = \frac{V_g^2}{4R}$$

and the input power for the same condition is

$$P_{in} = V_g \times I = V_g \times \frac{V_g}{R + R_g} = \frac{V_g^2}{2R}$$

So, the efficiency under this condition is

$$\eta = \frac{P_{out}}{P_{in}} \times 100 \% = 50 \%$$

The efficiency, 50% means one-half of the total generated power is dissipated within the source and the remaining other-half is utilized by the load.

#### > Maximum Power Transfer Theorem

- **✓** Condition of Maximum Power Transfer for different networks
  - 2. Reactance present, load resistance and reactance are independently varied:

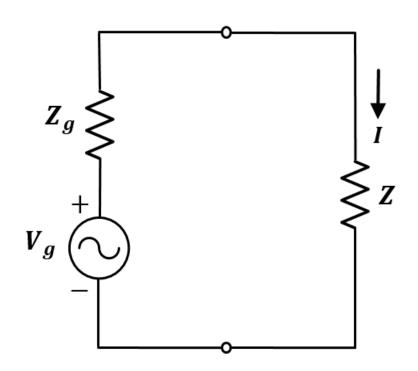
#### **Proof:**

A voltage source of generated voltage  $V_g$  and internal impedance  $Z_g$  is connected to a load impedance Z as shown in **Fig. 2.7.** 

where  $Z_g = R_g + jX_g$ , internal impedance of the source and Z = R + jX, load impedance.

The load current is

$$I = \frac{V_g}{(R+R_g)+j(X+X_g)}$$



**Fig. 2.7.** Circuit having load with resistance and reactance.

- > Maximum Power Transfer Theorem
  - **✓** Condition of Maximum Power Transfer for different networks
    - 2. Reactance present, load resistance and reactance are independently varied:

#### **Proof:**

The power delivered to the load is

$$P = I^{2}R = \frac{V_{g}^{2} R}{(R + R_{g})^{2} + (X + X_{g})^{2}}$$

The power is maximized by changing X, we must have

$$\frac{dP}{dX} = 0$$

$$\therefore X = -X_g$$

- > Maximum Power Transfer Theorem
  - **✓** Condition of Maximum Power Transfer for different networks
    - 2. Reactance present, load resistance and reactance are independently varied:

#### **Proof:**

Clearly, P is maximum with respect to reactance variation when  $X = -X_g$ . Under this condition

$$P = \frac{\left(V_g\right)^2 R}{\left(R + R_g\right)^2}$$

Maximum power transfer, P to the load will be maximum when  $Z = R_g - jXg = Z_g^*$ .

So, the load impedance is the complex conjugate of the internal impedance of the source.

#### > Maximum Power Transfer Theorem

- ✓ Condition of Maximum Power Transfer for different networks
  - 3. Reactance present, the magnitude but not the angle of the load impedance is varied:

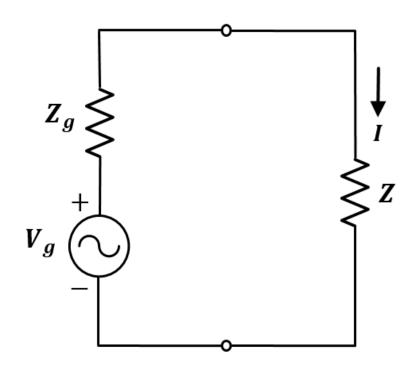
#### **Proof:**

A voltage source of generated voltage  $V_g$  and internal impedance  $Z_g$  is connected to a load impedance Z as shown in **Fig. 2.8.** 

where  $Z_g = R_g + jX_g$ , internal impedance of the source and  $Z = |Z| \cos \theta + j|Z| \sin \theta$ , load impedance.

The load current is

$$I = \frac{V_g}{(|Z|\cos\theta + R_g) + j(|Z|\sin\theta + X_g)}$$



**Fig. 2.8.** Circuit having load with resistance and reactance.

- > Maximum Power Transfer Theorem
  - **✓** Condition of Maximum Power Transfer for different networks
    - 3. Reactance present, the magnitude but not the angle of the load impedance is varied: Proof:

The power delivered to the load is

$$P = \frac{\left(V_g\right)^2 |Z| \cos \theta}{\left(R_g + |Z| \cos \theta\right)^2 + \left(X_g + |Z| \sin \theta\right)^2}$$

For maximum power transfer we must have

$$\frac{dP}{d|Z|} = 0$$

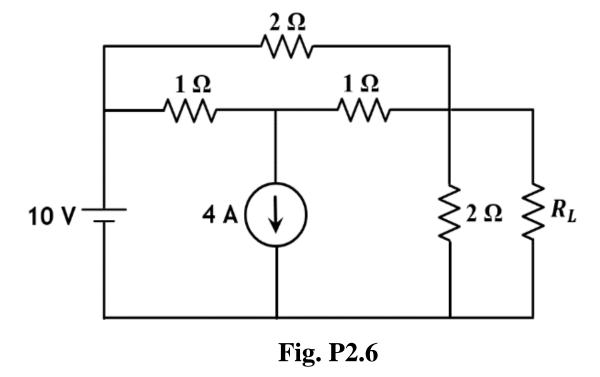
Under this condition and simplifying, we get  $(R_g)^2 + (X_g)^2 = |Z|^2$  $\therefore |Z_g| = |Z|$ 

This is the condition of maximum power transfer and it holds for a transformer

#### > Maximum Power Transfer Theorem

# Example – P2.6

Find the value of  $R_L$  for maximizing power transfer and also find the value of maximum power for the circuit as shown in **Fig. P2.6.** 



## > Maximum Power Transfer Theorem

# **Solution of Example – P2.6**

The circuit for finding Thevenin Voltage is as shown in Fig. P2.6 (a).

Apply KCL at node A, we get

$$\frac{V_A - 10}{1} + \frac{V_A - V_B}{1} + 4 = 0$$

or, 
$$V_A - 10 + V_A - V_B + 4 = 0$$
 10 V  $\top$ 

$$\therefore 2 V_A - V_B = 6 \longrightarrow (1)$$

Apply KCL at node B, we get

$$\frac{V_B}{1} \sim \frac{V_B - V_A}{1} + \frac{V_B - 10}{2} + \frac{V_B}{2} = 0$$

or, 
$$2V_B - 2V_A + V_B - 10 + V_B = 0$$

or, 
$$4 V_B - 2V_A = 10$$

$$\therefore 2 V_B - V_A = 5$$

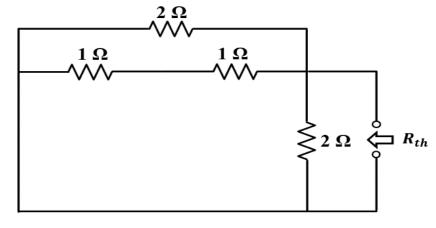
Multiply Eq. (1) by 1 and Eq. (2) by 2, then add and we get

$$\therefore V_B = V_{Th} = \frac{16}{3} V$$

#### > Maximum Power Transfer Theorem

# **Solution of Example – P2.6**

The circuit for finding Thevenin Resistance is as shown in Fig. P2.6 (b).



**Fig. P2.6** (b)

$$R_{th} = [2 \| (1+1)] \| 2 = (1 \| 2) = \frac{2}{3} \Omega$$

The Thevenin equivalent circuit along with load resistance is shown in **Fig. P2.6** (c)

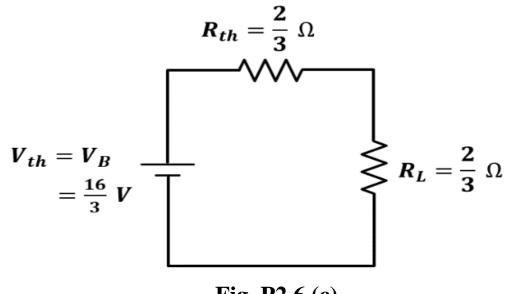


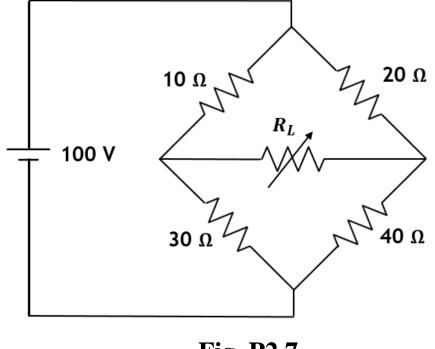
Fig. P2.6 (c)

: Maximum power transfer, 
$$P_L = \frac{(V_{th})^2}{4 R_L} = \left(\frac{16}{3}\right)^2 \times \frac{1}{4 \times \frac{2}{3}} = \frac{16 \times 16}{3 \times 3} \times \frac{3}{8} = \frac{32}{3} = 10.67 \text{ watt}$$

#### > Maximum Power Transfer Theorem

# Example – P2.7

Find the value of load resistance  $R_L$  to receive maximum power from the source and also find the maximum power delivered to the load in the circuit as shown in **Fig. P2.7.** 



**Fig. P2.7** 

