

1) 23.3

Verify that the LQG/LQR controller makes the closed loop system asymptotically stable

$$\dot{\hat{x}} = (A - LC)\hat{x} + Bu + Ly \quad u = Kx \quad \dot{x} = Ax + Bu \quad e = x - \hat{x}$$

$$\dot{\hat{x}} = (A - LC - BK)\hat{x} + LCx \quad \dot{x} = Ax - BK\hat{x}$$

$$\begin{aligned} \dot{e} = \dot{x} - \dot{\hat{x}} &= Ax - BK\hat{x} - A\hat{x} + LC\hat{x} - LCx + BK\hat{x} \\ &= (A - LC)e \end{aligned}$$

This is stable so the estimation will go to zero

$$\dot{x} = Ax - BK\hat{x} = Ax - BKx = (A - BK)x \quad \text{which is stable, so the system is stable}$$

2) 23.4

Verify that the solution to (23.17) is given by (23.18)

$$\begin{cases} Ax_{eq} + Bu_{eq} = 0 \\ r = Gx_{eq} + Hu_{eq} \end{cases} \Leftrightarrow \begin{bmatrix} -A & B \\ G & H \end{bmatrix} \begin{bmatrix} -x_{eq} \\ u_{eq} \end{bmatrix} = \begin{bmatrix} 0 \\ r \end{bmatrix}$$

Let $P(s) = \begin{bmatrix} sI - A & B \\ -G & H \end{bmatrix}$ be the rosenbrock matrix then

$P(0) = \begin{bmatrix} -A & B \\ -G & H \end{bmatrix}$ if $P(0)$ is full rank - i.e. number of inputs is larger than the number of outputs then

$$\begin{bmatrix} -x_{eq} \\ u_{eq} \end{bmatrix} = P(0)^{-1} (P(0)P(0)^{-1}) \begin{bmatrix} 0 \\ r \end{bmatrix} \quad \text{proof: sub into equation 23.17}$$

$$P(0)^{-1} P(0) \begin{bmatrix} 0 \\ r \end{bmatrix} = \begin{bmatrix} 0 \\ r \end{bmatrix} \Rightarrow \begin{bmatrix} 0 \\ r \end{bmatrix} = \begin{bmatrix} 0 \\ r \end{bmatrix}$$

3) 23.5

Verify that the LQG/LQR setpoint controller (23.21) makes the closed loop system asymptotically stable.

$$\begin{aligned} \dot{\bar{x}} &= (A - LC - BK) \bar{x} - L(y - Lx_{eq}) & u &= K\bar{x} + u_{eq} & u_{eq} &= Nr & y_{eq} &= rCF \\ \bar{x} &= x_{eq} - \hat{x} & \tilde{x} &= x - x_{eq} & e &= x - \hat{x} \end{aligned}$$

$$\begin{aligned} \dot{e} &= \dot{x} - \dot{\hat{x}} = Ax + Bu + (A - LC - BK) \bar{x} - L(y - Lx_{eq}) \\ &= Ax + BK\bar{x} + Bu_{eq} + A\bar{x} - LC\bar{x} - BK\bar{x} - L(y - Lx_{eq}) \\ &= Ax + A\bar{x} + Bu_{eq} - LC\bar{x} - LCx + LCx_{eq} \\ &= Ax + A(x_{eq} - \hat{x}) + Bu_{eq} - LC(x_{eq} - \hat{x}) - LCx + LCx_{eq} \\ &= (A - LC)(x - \hat{x}) + \underbrace{Ax_{eq} + Bu_{eq} - LCx_{eq}}_{\text{at equilibrium this is zero}} \end{aligned}$$

$$= (A - LC) e$$

This shows that $\hat{x} \rightarrow x$ as $t \rightarrow \infty$ (23.21)
shows that $\bar{x} = x_{eq} - \hat{x} \rightarrow 0$ as $t \rightarrow \infty$ thus the system is stable ✓

4) 23.6

Show that for a single controlled output ($l=1$), we can take $u_{eq}=0$ in (23.17) when the matrix A has an eigenvalue at the origin and this mode is observable through z .

If this mode is observable through z then $\text{rank} \begin{bmatrix} A - \lambda I \\ G \end{bmatrix} = n$

$$\begin{bmatrix} -A & B \\ -G & H \end{bmatrix} \begin{bmatrix} -x_{eq} \\ u_{eq} \end{bmatrix} = \begin{bmatrix} 0 \\ r \end{bmatrix} \quad \text{Let } u_{eq}=0 \text{ then } \begin{bmatrix} -Ax_{eq} \\ -Gx_{eq} \end{bmatrix} = \begin{bmatrix} 0 \\ r \end{bmatrix}$$

This means $Ax_{eq} = 0$ & $-Gx_{eq} = r$.
Since A has an eigenvalue at the origin, then A is singular & x_{eq} must be in the null space of A . Since this mode is observable through G s.t. $Gx_{eq} \neq 0$, then x_{eq} can be chosen such that $-Gx_{eq} = r$.

Remember $x_{eq} \in N(A) = \alpha v_0$ where α is a scalar and v_0 is the eigen vector corresponding to the 0 eigen value.

$-Gx_{eq} = \alpha(-Gv_0) = r$ where α can be a tuning value to get any value r desired. ✓

7)

show that $\det \begin{bmatrix} z_i I - A & B \\ -C & D \end{bmatrix} = \det(z_i I - A) \det(G(z_i))$

$$\begin{bmatrix} z_i I - A & B \\ -C & D \end{bmatrix} = \begin{bmatrix} z_i I - A & 0 \\ -C & I \end{bmatrix} \begin{bmatrix} I & (z_i I - A)^{-1} B \\ 0 & G(z_i) \end{bmatrix}$$

$$\det \left(\begin{bmatrix} z_i I - A & 0 \\ -C & I \end{bmatrix} \begin{bmatrix} I & (z_i I - A)^{-1} B \\ 0 & G(z_i) \end{bmatrix} \right) = \det \begin{bmatrix} z_i I - A & 0 \\ -C & I \end{bmatrix} \det \left(\begin{bmatrix} I & (z_i I - A)^{-1} B \\ 0 & G(z_i) \end{bmatrix} \right)$$

$$= \det(z_i I - A) \det(G(z_i)) \quad \checkmark$$

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5)

```
A = [2 0 0 ; 0 -1 0; 0 0 -1];  
B = [1 0 ; 1 0; 0 1];  
C = [1 0 2; 0 -1 0];  
D = [1 0; 1 0];
```

```
sys = ss(A,B,C,D);
```

a)

Where are the poles of the system? What is the multiplicity of each pole?

```
syms s
```

```
G_s = C*inv(s*eye(3)-A)*B +D;  
% Poles of the system are 2,-1,-1
```

b)

What are the invariant zeros of the system(finite and infinite)?

```
P = [s*eye(3)-A, B; -C D]
```

```
% An invariant zero will cause P to lose rank.  
rank_2 = rank(subs(P,2))  
rank_1 = rank(subs(P,-1))  
rank_0 = rank(subs(P,0))
```

```
% Invariant zeros are 0, inf and 2
```

c) Transmission zeros are the invariant zeros that are not eigenvalues of A.

```
% Since 2 is an eigen value of A, 0 is the only transmission zero.
```

```

P =

[ s - 2,      0,      0, 1, 0]
[      0, s + 1,      0, 1, 0]
[      0,      0, s + 1, 0, 1]
[     -1,      0,     -2, 1, 0]
[      0,      1,      0, 1, 0]

```

```

rank_2 =

4

```

```

rank_1 =

5

```

```

rank_0 =

4

```

6)

```

clear all;
A = [-1 0 0; 0 -2 0; 0 0 -2];
B = [2 -2; -2 4; -4 2];
C = [1 1 0; 1 0 1];
D = [0 0; 0 0];
sys = ss(A,B,C,D);

```

a) Use the transmission zero to find an input $u(t)$ and the initial

condition $x(0)$ that will result in $y(t) = 0$ for all time.

```

lambda_A = eig(A);      % Eigen values are -2, -2, -1
inv_zeros = tzero(sys); % Invariant zeros are 1
                        % The transmission zero must be 1 since 1 is
                        not an
                        % eigen value of A.

% Now that we have a transmission zero, we can find an x(s) and u(s)
  that
% are in the null space of P(s=transmission_zero).

```

```

syms s
P = [s*eye(3)-A, B; -C D];
Pz = [1*eye(3) - A,B; -C D];
nu = null(double(subs(P,1)));
nu = null(Pz)

% Thus the solution is x(0) = -[2 -2 -2]' and u(t) = [-1 1]'exp(t)'

```

b) Verify your solution by calculating the output for the given input

```

xo = nu(1:3);
uo = -nu(4:5);

% opts = odeset('RelTol',1e-2,'AbsTol',1e-8);

[t,x] = ode45(@(t,x) prob6(t,x,A,B,C,D,uo), [0 2],xo);

y = C*x';

sum(y,2)

syms t
u = uo.*exp(0:0.01:2);

[y,x] = lsim(ss(A,B,C,0),uo.*exp(0:0.01:5),0:0.01:5,xo);
sum(y)

function dx = prob6(t,x,A,B,C,D,uo)

u = uo*exp(t);

dx = A*x+B*u;

end

nu =

    0.5345
   -0.5345
   -0.5345
   -0.2673
    0.2673

ans =

    1.0e-04 *

```

0.2060
0.2060

ans =

1.0e-03 *
-0.2190 -0.2190

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Question 8

```
Xo = [250,0,0,0]';

th_max = 0.5;           % Throttle max
el_max = 25*pi/180; % Elevator max

%
% x = [Airspeed, Angle of attack, Pitch Rate, Pitch Angel]'
% u = [Throttle, elevator]'

% State Transition Model
A = [-0.038   18.984   0       -32.174;...
      -0.001  -0.632   1       0;...
      0       -0.759  -0.518   0;...
      0       0       1       0];

% Control-Input Model
B = [10.1      0;...
      0       -0.0086;...
      0.025    -0.011;...
      0       0];

C = zeros(2,4);
C(1:2,1:2) = eye(2);
D = 0;

SYS = ss(A,B,C,D);

% Input Weights
R1 = diag([1/th_max^2; 1/el_max^2]); % Bryson's method
r = 1;                               % Scaling Factor
R = r*R1;

% State Weights
max_angle_dev = 0.5*pi/180;
Q1 = diag([0.1, 1/max_angle_dev^2, 1/max_angle_dev^2, 1/
max_angle_dev^2 ]); % Bryson's method
q = 1;
% Scaling Factor
Q = q*Q1;
```

a

Set $q = 1$ and let $r = [1000, 100, 10, 7]$. Plot the damping ratio vs oscillation frequency for the closed-loop short period mode obtained using LQR optimal. Include the open loop short period modes too.

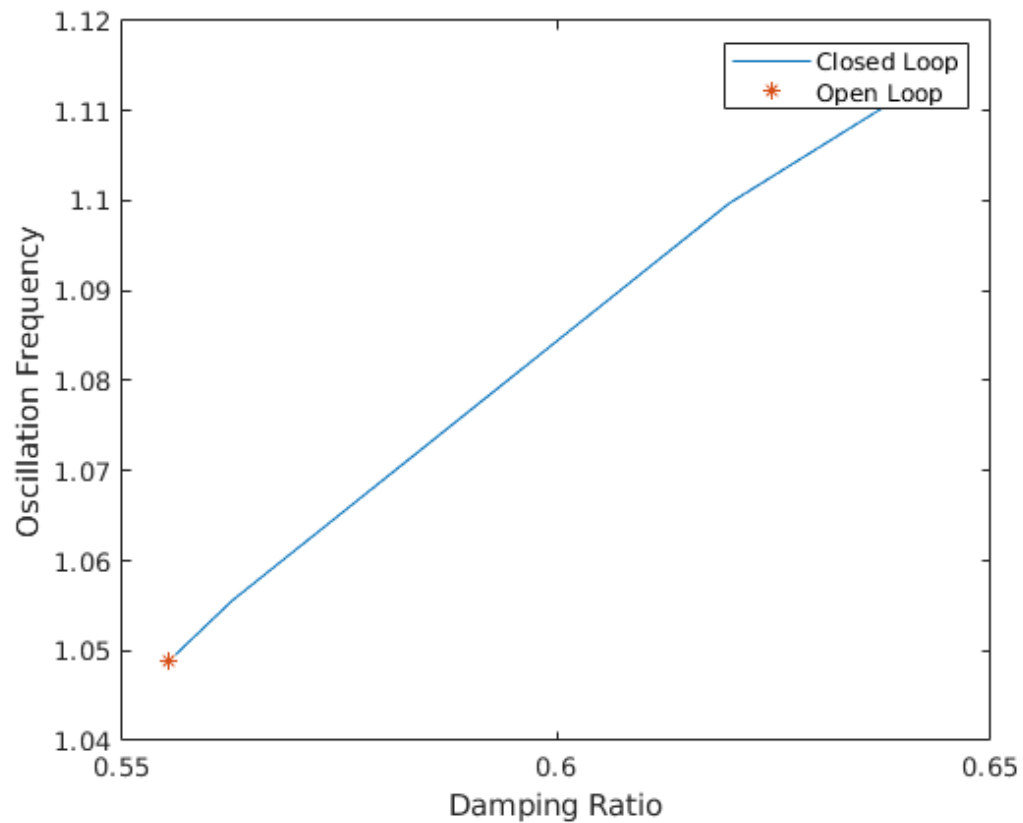
```
% Openloop
[Wn,Z,P] = damp(SYS);

% Close loop
Wnc = zeros(4,4);
Zc = zeros(4,4);
Pc = zeros(4,4);
ii = 1;
for r = [7,10,100,1000]

    R = r*R1;
    [K,S,E] = lqr(A,B,Q,R,0);           % Compute K gain
    Ac = A-B*K;
    Bc = zeros(4,2);
    SYSs = ss(Ac,Bc,C,0);
    [Wnc(:,ii),Zc(:,ii),Pc(:,ii)] = damp(SYSs);
    ii = ii +1;
end

figure(1), clf;
plot(Zc(4,:),Wnc(4,:));
hold on
plot(Z(4),Wn(4),'*');
xlabel("Damping Ratio");
ylabel("Oscillation Frequency");
legend("Closed Loop","Open Loop")

% It seems that as r becomes smaller, the closed loop system short
period
% modes converge to the open loop system short period modes.
```



b

Choose an R matrix that maximizes use of the elevator and throttle deflection without exceeding their bounds given an initial perturbation of $x_0 = [20, 0.01, -0.01, 0.02]'$. Make plots of the uncontrolled and controlled responses to verify your design.

```

r = 54;          % Scale
R = r*R1;
xo = [20, 0.01, -0.01, 0.02]'; % Initial Conditions
xo = [xo;xo];
[K,S,E] = lqr(A,B,Q,R,0);

[t,x] = ode45(@(t,x) aircraftDynamics(t,x,A,B,C,K,[ ]),[0 10],xo);
% closed loop
[t_ol,x_ol] = ode45(@(t,x) aircraftDynamics(t,x,A,B,C,[ ],[ ]),[0
10],xo); % Open loop
u = -K*x(:,1:4)';

figure(2),clf;

subplot(4,1,1);

plot(t,x(:,1));
hold on
plot(t_ol,x_ol(:,1));

```

```

title("Airspeed")
xlabel("time (s)");
ylabel("m/s");
legend("Openloop", "Closed Loop")

subplot(4,1,2);
plot(t,x(:,2));
hold on
plot(t_ol,x_ol(:,2));
title("Angle of Attack")
xlabel("time (s)");
ylabel("rads");
legend("Openloop", "Closed Loop")

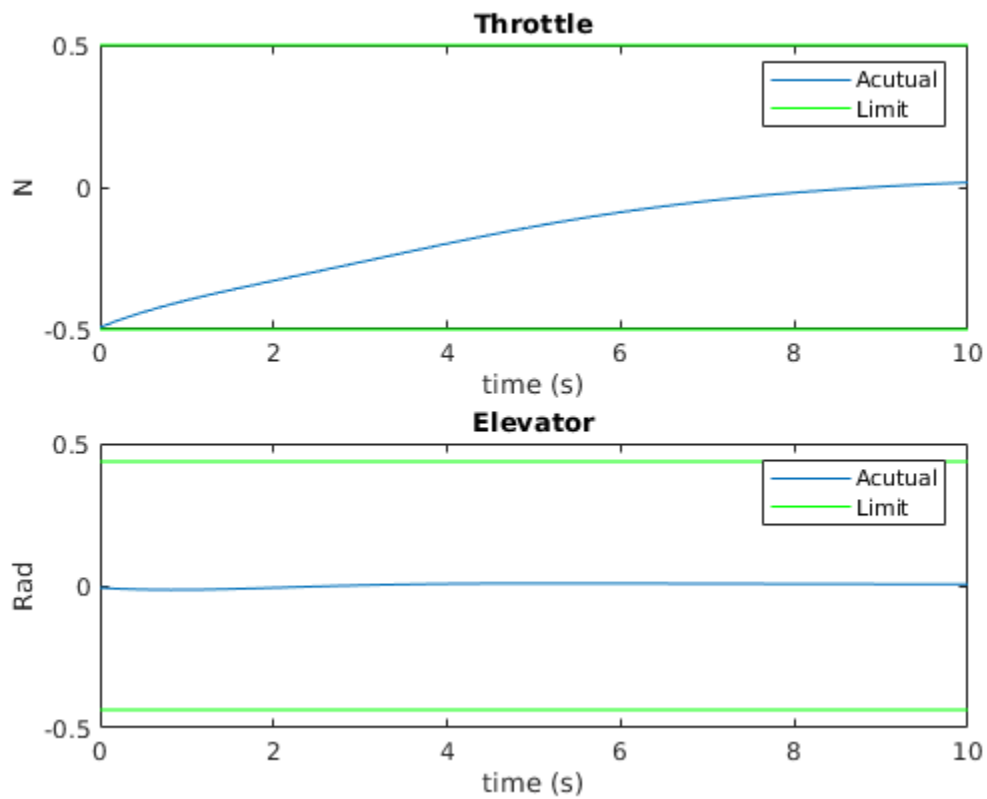
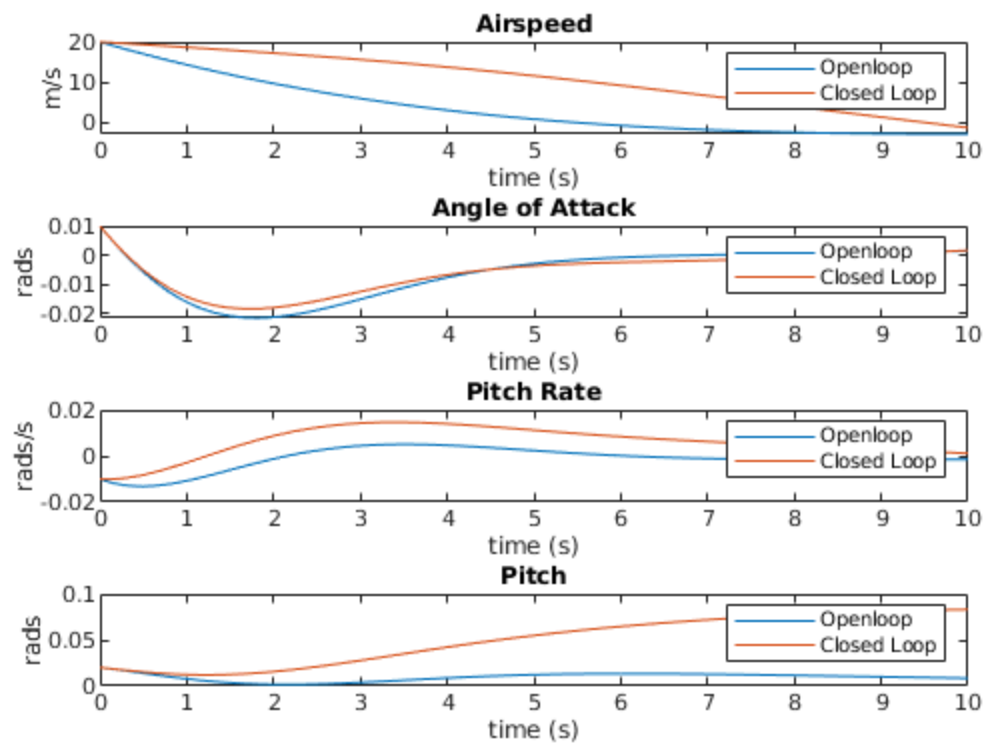
subplot(4,1,3);
plot(t,x(:,3));
hold on
plot(t_ol,x_ol(:,3));
title("Pitch Rate")
xlabel("time (s)");
ylabel("rads/s");
legend("Openloop", "Closed Loop")

subplot(4,1,4);
plot(t,x(:,4));
hold on
plot(t_ol,x_ol(:,4));
title("Pitch")
xlabel("time (s)");
ylabel("rads");
legend("Openloop", "Closed Loop")

figure(3),clf;
subplot(2,1,1);
plot(t,u(1,:));
hold on
plot(t,ones(1,length(t))*th_max,'g');
plot(t,-ones(1,length(t))*th_max,'g');
title("Throttle")
xlabel("time (s)");
ylabel("N");
legend("Acutual","Limit")

subplot(2,1,2)
plot(t,u(2,:));
hold on
plot(t,ones(1,length(t))*el_max,'g');
plot(t,-ones(1,length(t))*el_max,'g');
title("Elevator")
xlabel("time (s)");
ylabel("Rad");
legend("Acutual","Limit")

```



c)

Implement a kalman filter observer. And we wish to control the airspeed and the flight path angle which are the only variables measured.

```
xeq = [10,0,0,0]';
ueq = -B \ A*xeq;
A*xeq + B*ueq;

SYSk = ss(A, [B B*B'],C,0);

R = diag([1,10^-5]);
Q = B*10^-4*B';
[est,L,P] = kalman(SYSk,Q,R);

xo = [20, 0.01, -0.01, 0.02]'; % Initial Conditions
xo = [xo;xo*1.5];
[tk,xk] = ode45(@(t,x) aircraftDynamics(t,x,A,B,C,K,L),[0 10],xo); %
Open loop

figure(4),clf;

subplot(4,1,1);

plot(tk,xk(:,1));
hold on
plot(tk,xk(:,5));
title("Airspeed")
xlabel("time (s)");
ylabel("m/s");
legend("True", "Estimate")

subplot(4,1,2);
plot(tk,xk(:,2));
hold on
plot(tk,xk(:,6));
title("Angle of Attack")
xlabel("time (s)");
ylabel("rads");
legend("True", "Estimate")

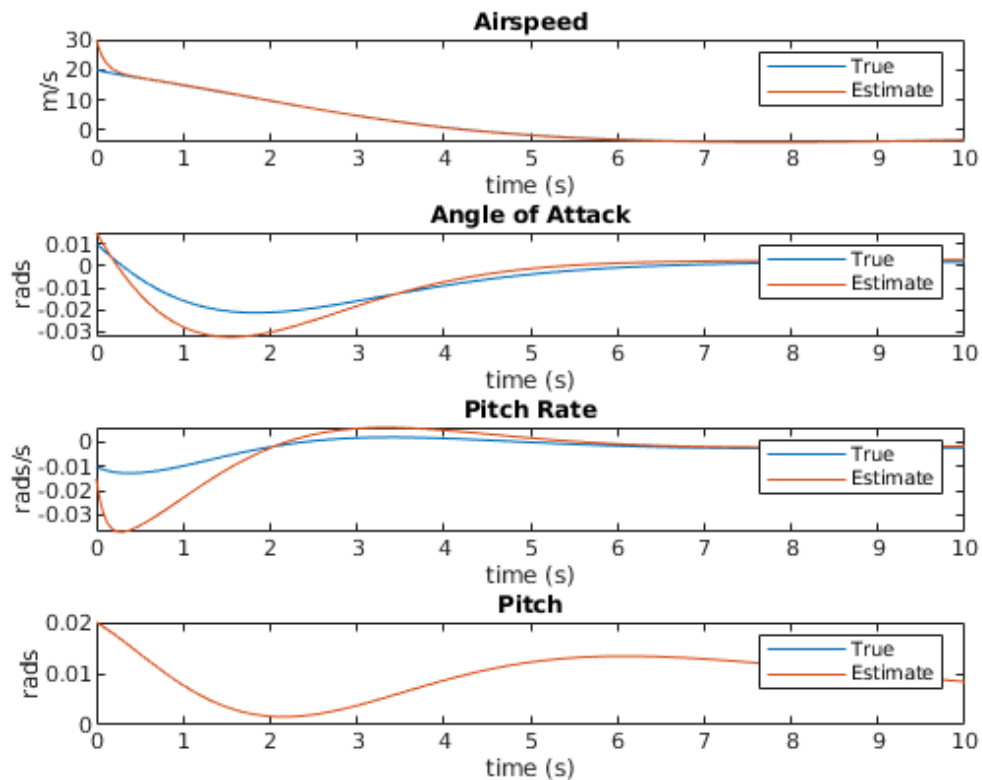
subplot(4,1,3);
plot(tk,xk(:,3));
hold on
plot(tk,xk(:,7));
title("Pitch Rate")
xlabel("time (s)");
ylabel("rads/s");
legend("True", "Estimate")

subplot(4,1,4);
plot(t,x(:,4));
hold on
```

```

plot(t,x(:,8));
title("Pitch")
xlabel("time (s)");
ylabel("rads");
legend("True", "Estimate")

```



d

Use the observer gains and controller gains derived in parts c) and d) from hw #5 to regulate your system to their trim conditions.

```

p = [-5+j,-5-j,-3+0.14j,-3-0.14j];
K_old = place(A,B,p);
L_old = place(A',C',p*10)';

[t_old,x_old] = ode45(@(t,x) aircraftDynamics(t,x,A,B,C,K_old,L_old),
[0 10],xo); % Using old gains.

% By inspecting the plots, the LQR/LQG design seems to be much more
% efficient.

figure(5),clf;

subplot(4,1,1);

```

```

plot(tk,xk(:,5));
hold on
plot(t_old,x_old(:,5));
title("Airspeed")
xlabel("time (s)");
ylabel("m/s");
legend("LQR/LQG", "Old Design")

subplot(4,1,2);
plot(tk,xk(:,6));
hold on
plot(t_old,x_old(:,6));
title("Angle of Attack")
xlabel("time (s)");
ylabel("rads");
legend("LQR/LQG", "Old Design")

subplot(4,1,3);
plot(tk,xk(:,7));
hold on
plot(t_old,x_old(:,7));
title("Pitch Rate")
xlabel("time (s)");
ylabel("rads/s");
legend("LQR/LQG", "Old Design")

subplot(4,1,4);
plot(tk,xk(:,8));
hold on
plot(t_old,x_old(:,8));
title("Pitch")
xlabel("time (s)");
ylabel("rads");
legend("LQR/LQG", "Old Design")

function dxdt = aircraftDynamics(t,x,A,B,C,K,L)

    z = x(1:4);
    zh = x(5:8);

    if isempty(K)
        u = zeros(2,1);
    else
        u = -K*zh;
    end

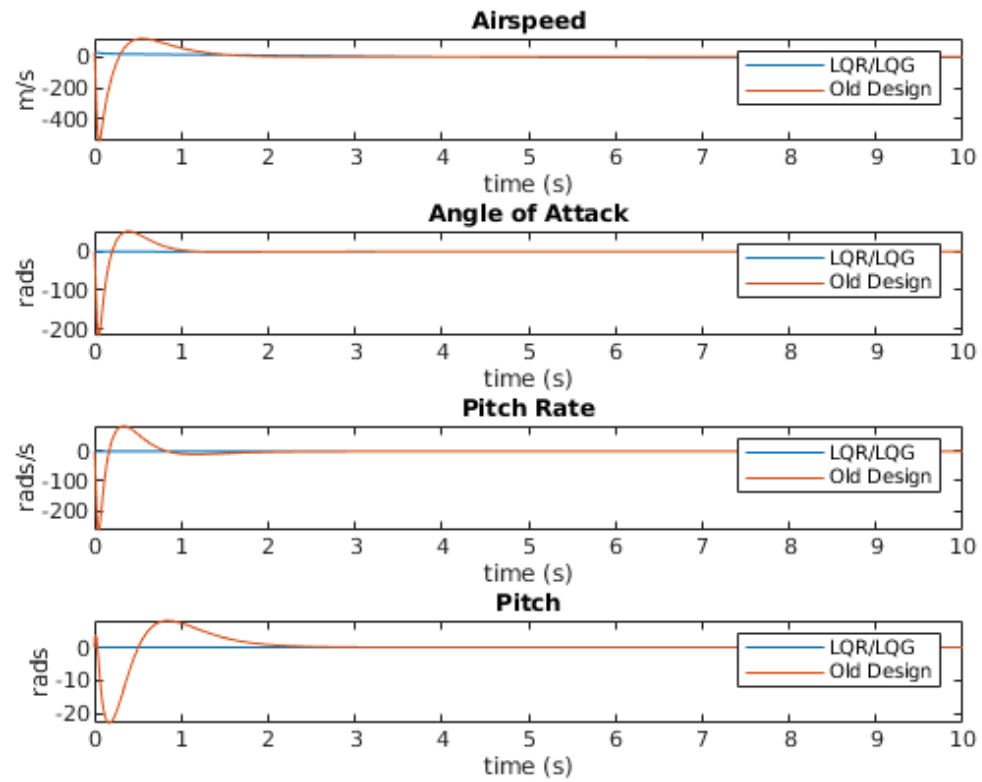
    z_dot = A*z + B*u;

    if isempty(L)
        zh_dot = z_dot;
    else
        zh_dot = A*zh + B*u - L*C*(zh - z);
    end

```

```
dxdt = [z_dot;zh_dot];
```

```
end
```



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