

1. (a)

Syndrome

2028-10-31

Let E_i denote symbol energy of s_i (Signal is real)

$$\begin{aligned}
 E_1 &= \int_{-\infty}^{\infty} (-3A p(t))^2 dt \\
 &= 9A^2 \underbrace{\int_{-\infty}^{\infty} (p(t))^2 dt}_{\text{unit-energy}}
 \end{aligned}$$

$$E_2 = \int_{-\infty}^{\infty} (-A_p(t))^2 dt$$

$$= A^2 \int_{-\infty}^{\infty} p(t)^2 dt$$

$$= A^2 \text{ initial energy}$$

$$\begin{aligned} E_J &= \int_{-\infty}^{\infty} (Ap(t))^2 dt \\ &= A^2 \int_{-\infty}^{\infty} (p(t))^2 dt \\ &= A^2 \underbrace{\int_0^T (p(t))^2 dt}_{\text{Varil. energy}} \end{aligned}$$

$$\begin{aligned} E_p &= \int_{-\infty}^{\infty} (1/\rho(t))^2 dt \\ &= g\Delta^2 \int_{-\infty}^{\infty} (\rho(t))^2 dt \\ &= g\Delta^2 \text{ initial energy} \end{aligned}$$

Tunipro bubble \Rightarrow ~~average~~ sy

$$\Rightarrow E_s = \frac{E_1 + E_2 + E_3 + E_4}{4} = 5A^2$$

(b)

$$\text{Correlator: } V(t) = p(t) * y(t) = \int_0^T$$

Dirk Sy

Define signal $s(t)$ ~~$s(t')$~~ $s(t, x) = x A_p(t)$

~~s(t; A)~~

Correlator usage

$$p(t) := \underline{V}(t) = p(t) * s(t, A)$$

~~(EZ, X, t)~~
~~(twice has zero mean)~~

$$V(t) = p(t+T) \star s(t, A)$$

$$= \int_{-\infty}^{\infty} p(\tau + T - t) s(\tau; A) d\tau$$

$$= \int_{-\infty}^{\infty} p(\tau + T - \tau) s(\tau; \alpha) p(\tau) d\tau$$

~~$\check{V}(\tau) = p(\tau)$~~

$$= \int_{-\infty}^{\tau} p(\tau) s(\tau; \alpha) d\tau$$

$$= \lambda A \int_0^T (p(\tau))^2 d\tau$$

$\exists x A$ ~~and every~~ even
 $(\forall t) \exists s$ ~~symmetric~~
~~such that~~ $t = \frac{1}{s}$

$$\text{mean} = \frac{1}{T} \int_0^T x(t) A(p(t))^2 dt$$

$$= x A \int_0^T (p(1+x))^2 dt$$

λ (lambdah) \rightarrow $T \rightarrow T$

$$\check{V}(\vec{x}) = \int_{\vec{r}=\infty}^{\vec{r}=\vec{x}} P(\vec{r}) d\vec{r}$$

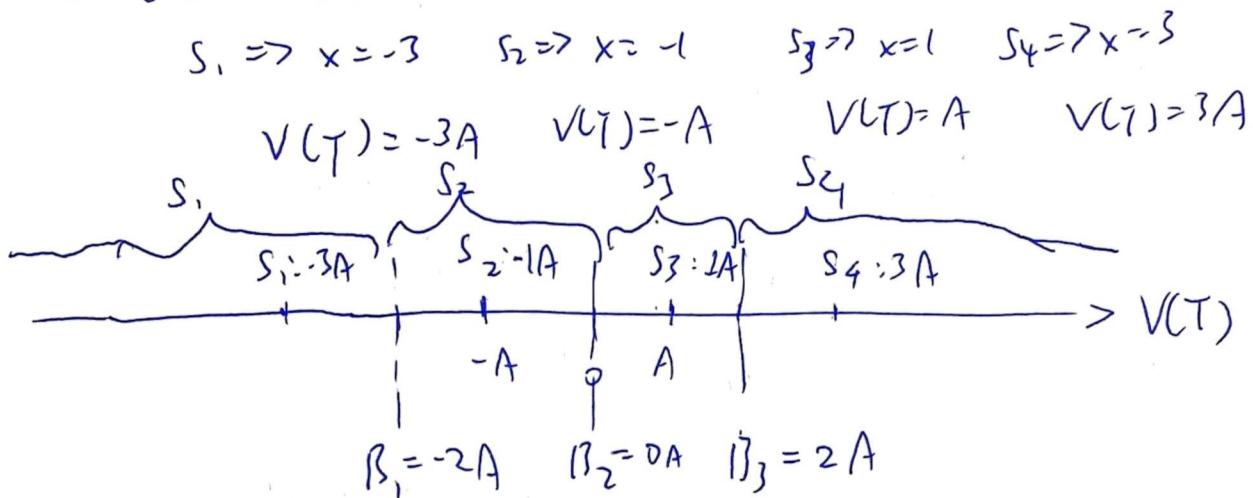
$$\check{Y}(T) = \int_0^\infty p(T-\tau) S(\tau; A) d\tau$$

$$= \int_0^T \int_0^T \rho(t-T) s$$

$$= \int_{-\infty}^{\infty} p(\tau - T) * A p(\tau) d\tau$$

$$-x A \int_0^{\infty} p(\tau) \tau dt = x A \text{ (total energy)}$$

(b) (continue)



boundary should be in the middle of two signals adjacent signals as they are equiprobable.

Decision rule :

$$\hat{s}(t) = \begin{cases} s_1(t) & \text{if } V(t) \leq -2A \\ s_2(t) & \text{if } -2A < V(t) \leq 0A \\ s_3(t) & \text{if } 0A < V(t) \leq 2A \\ s_4(t) & \text{if } 2A < V(t) \end{cases}$$

(c) ~~take over time T~~
~~noise variance!~~ $\sigma^2 = \sqrt{\frac{N_0}{2} T}$

$P_{els_1} = \alpha \left(\frac{A}{\sigma} \right)^2$

$P_{els_2} = 2 \alpha \left(\frac{A}{\sigma} \right)$

$P_{els_3} = 2 \alpha \left(\frac{A}{\sigma} \right)$

$P_{els_4} = 2 \alpha \left(\frac{A}{\sigma} \right)$

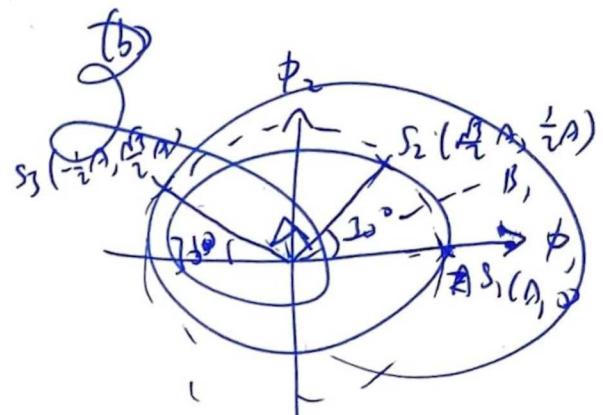
$P_e = \frac{P_{els_1} + P_{els_2} + P_{els_3} + P_{els_4}}{4} \quad (\text{equiprobable})$

$$= \frac{1}{2} \alpha \left(\frac{A}{\sigma} \right)^2 = \frac{1}{2} \alpha \left(\sqrt{\frac{2A^2}{N_0 T}} \right)^2 = \frac{1}{2} \alpha \left(\sqrt{\frac{2E_s}{SN_0 T}} \right)$$

2. (a)

$$d_{12} = \sqrt{(A - A \cos \alpha)^2 + (0 - A \sin \alpha)^2}$$

$$\begin{aligned} &= \sqrt{A^2(1 - \cos^2 \alpha) + A^2 \sin^2 \alpha} \\ &= A \sqrt{1 - 2 \cos^2 \alpha} \\ &= A \sqrt{2(1 - \cos 2\alpha)} \\ &= A \sqrt{2} \end{aligned}$$



Signal coordinates

$$S_1 : (A, 0)$$

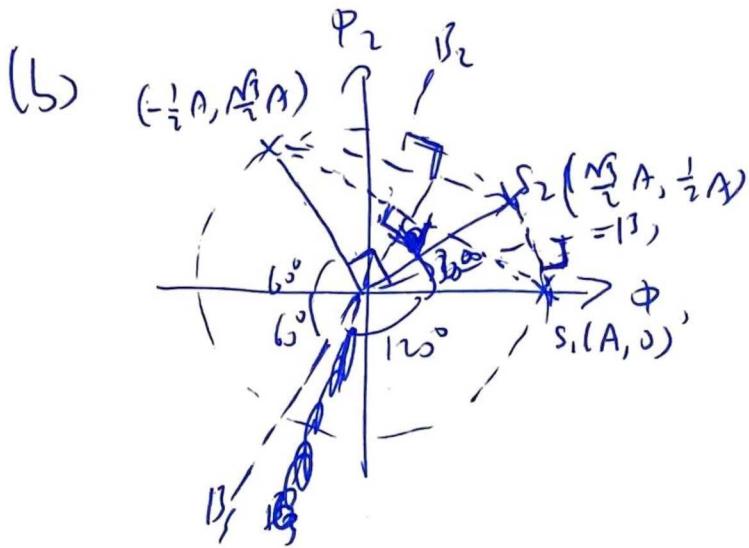
$$S_2 : (A \cos \alpha = A \frac{\sqrt{3}}{2} A, A \sin \alpha = \frac{1}{2} A)$$

$$S_3 : (-A \sin \alpha = -\frac{\sqrt{3}}{2} A, A \cos \alpha = \frac{1}{2} A)$$

$$\begin{aligned} d_{12} &= \sqrt{A^2(1 - \cos^2 \alpha) + \left(\frac{1}{2}A\right)^2} = \sqrt{\left(A - \frac{A \cos \alpha}{2}\right)^2 + \left(0 - \frac{1}{2}A\right)^2} \\ &= A \sqrt{1 - \frac{3}{4} + \frac{1}{4} + \frac{1}{4}} = A \sqrt{2 - \sqrt{3}} \end{aligned}$$

$$\begin{aligned} d_{13} &= \sqrt{A^2(1 - \cos(1 + \frac{1}{2}))^2 + \left(0 - \frac{\sqrt{3}}{2}A\right)^2} \\ &= \sqrt{A^2(1 - \cos(1 + \frac{1}{2}))^2 + \left(0 - \frac{\sqrt{3}}{2}A\right)^2} \\ &\approx \sqrt{A^2} = A \end{aligned}$$

$$\begin{aligned} d_{23} &= \sqrt{(A \cos \alpha - A \cos 1)^2 + (A \sin \alpha - A \sin 1)^2} \\ &= \sqrt{\left(\frac{\sqrt{3}}{2}A + \frac{1}{2}A\right)^2 + \left(\frac{1}{2}A - \frac{\sqrt{3}}{2}A\right)^2} \\ &\approx A \sqrt{\frac{3}{4} + \frac{1}{4} + \frac{1}{4} + \frac{1}{4} - \frac{\sqrt{3}}{2} + \frac{3}{4}} = A \sqrt{2} \end{aligned}$$



The optimal decision boundaries should be bisector of the ~~segment~~ two adjacent signals, when all symbols are equiprobable.

$$B_1: y = \tan\left(\frac{\pi}{12}x\right) \quad (x \geq 0)$$

$$B_2: y = \tan\left(\frac{5\pi}{12}x\right) \quad (x \geq 0)$$

$$B_3: y = \tan$$

$$B_1: y = x \tan\left(\frac{\pi}{12}\right) = x \frac{1 - \cos\frac{\pi}{6}}{\sin\frac{\pi}{6}} = x \frac{1 - \sqrt{3}/2}{1/2} = \cancel{2} \cdot (2 - \sqrt{3})x \quad (x \geq 0)$$

$$B_2: y = x \tan\left(\frac{5\pi}{12}\right) = x \frac{1}{\tan\left(\frac{\pi}{12}\right)} = \frac{1}{2 - \sqrt{3}}x = \cancel{2} \cdot (2 + \sqrt{3})x \quad (\cancel{x \geq 0})$$

$$B_3: y = x \tan\left(\frac{4\pi}{3}\right) = x \tan\left(\frac{\pi}{3}\right) = \cancel{\frac{\sqrt{3}}{2}x} \quad (x \leq 0)$$

(c) Equi-probable ~~noise~~ noise per unit time: $\sigma = \sqrt{\frac{N_0}{2}}$

$$P_e \leq \frac{1}{3} [P_{e|S_1} + P_{e|S_2} + P_{e|S_3}] \quad P_{e|S_1} \leq Q\left(\frac{d_{12}}{2\sigma}\right) + Q\left(\frac{d_{13}}{2\sigma}\right)$$

$$P_{e|S_2} \leq Q\left(\frac{d_{12}}{2\sigma}\right) + Q\left(\frac{d_{23}}{2\sigma}\right)$$

$$= \frac{2}{3} \left[Q\left(\frac{d_{12}}{2\sigma}\right) + Q\left(\frac{d_{13}}{2\sigma}\right) + Q\left(\frac{d_{23}}{2\sigma}\right) \right] \quad P_{e|S_3} \leq Q\left(\frac{d_{13}}{2\sigma}\right) + Q\left(\frac{d_{23}}{2\sigma}\right)$$

$$= \frac{2}{3} \left(Q\left(\frac{A\sqrt{2\sqrt{3}}}{\sqrt{2N_0}}\right) + Q\left(\frac{A\sqrt{3}}{\sqrt{2N_0}}\right) + Q\left(\frac{A\sqrt{5}}{\sqrt{2N_0}}\right) \right)$$

Assume symbol time is T .

3. (a) The optimum receiver consists of M correlators
~~consists for~~ using $s_m(t)$ as the template
more specifically, each $s_m(t)$ correlation does
the following: \star -correlation:
$$s_m(t+T) \star y(t)$$

~~advanced so that~~

When $t=T$ (assume $t=0$ is when symbol is received),
the result of the ~~above~~ \star -correl.
$$(s_m(t+T) \star y(t))(T)$$
 is ~~very~~ saved.

This result is also the projection of the signal
during $[0, T]$ to each of $s_m(t)$ (up to scaling),
which is the same for all signals as $\|s_m\|^2 = E_s$.

To fix the scaling, just divide by $\sqrt{E_s}$. Let this

Then finally, finally, collecting this number

into ~~multiple~~ a tuple to result be x_m .

Finally, collect the ~~to~~ M results into a coordinate
in the signal space

$$\therefore (x_1, x_2, \dots, x_M).$$

Each orthogonal signal $s_m(t)$ has the
coordinate

$$(0, \dots, 0, \underbrace{\sqrt{E_s}}_{\text{m-th entry}}, 0, \dots, 0)$$

Compute distance of (x_1, \dots, x_M) to all orthogonal
signals, and choose $s_m(t)$ with the smallest distance.

(b) Assume sy

Each orthogonal signal $s_m(t)$ has the coordinate in signal space
 $\underbrace{(0, \dots, 0, \sqrt{E_s}, 0, \dots, 0)}_{\text{M numbers}}$
 \downarrow
 sum by threshold

\therefore distance between any two $s_i(t)$ and $s_j(t)$ ($i \neq j$)

$$= \sqrt{E_s + E_s} = \sqrt{2} E_s$$

$$\begin{aligned} P_{\text{err}} &= P_{\text{elsm}} \leq (M-1) \cdot 2 \left(\frac{\sqrt{2} E_s}{2 \sigma} \right) && \text{noise per unit time: } \sigma = \sqrt{\frac{N_0}{2}} \\ &= (M-1) \cdot 2 \left(\frac{E_s}{\sigma} \right) \left(\sqrt{\frac{E_s}{N_0}} \right) \end{aligned}$$

Unprobable

In orthogonal modulation,

$$\begin{aligned} P_e &\leq \frac{1}{2(M-1)} \cdot \frac{M}{2} \cdot \frac{P_{\text{err}}}{M-1} \\ &= \frac{1}{2} \cdot 2 \left(\frac{E_s}{\sigma} \right) \left(\sqrt{\frac{E_s}{N_0}} \right) \end{aligned}$$

$$4.(a) \langle S_a, S_b \rangle = \int_0^T p(t) dt$$

Assume $n \neq m, n \neq -m, n \neq 0, n \neq 0$

$p(n)$ has unit energy

$$\begin{aligned} \langle S_a, S_b \rangle &= \int_0^T 2E_s \cos(2\pi nt/T) p(t) \cos(2\pi mt/T) (p(t))^2 dt \\ &= \int_0^T 2E_s \underbrace{\left(\cos(2\pi(n+m)t/T) + \cos(2\pi(n-m)t/T) \right)}_{\substack{\text{constant} \\ \text{integrals to zero over } [0, T]}} (p(t))^2 dt \\ &= 0 \end{aligned}$$

as $n \neq m$

some integrals to zero over $[0, T]$

as $n \neq m$

constant within $[0, T]$

as $n \neq m$

For $\langle S_b, S_c \rangle$

$$\begin{aligned} \langle S_b, S_c \rangle &= \int_0^T 2E_s \cos(2\pi nt/T) \cos(2\pi mt/T + \theta) (p(t))^2 dt \\ &= \int_0^T 2E_s \underbrace{\left(\cos(2\pi(m+n)t/T + \theta) + \cos(2\pi(m-n)t/T + \theta) \right)}_{\substack{\text{const} \\ \text{int to zero over } [0, T] \\ \text{as } n \neq m}} (p(t))^2 dt \\ &\quad \underbrace{\text{int to zero over } [0, T]}_{\substack{\text{as } n \neq m}} \end{aligned}$$

last over $[0, T]$

$$\begin{aligned} \langle S_a, S_c \rangle &= \int_0^T 2E_s \underbrace{\cos(2\pi nt/T) \cos(2\pi mt/T + \theta)}_{\text{const}} (p(t))^2 dt \\ &= \int_0^T 2E_s \underbrace{\left(\cos(4\pi nt/T + \theta) + \cos(-\theta) \right)}_{\substack{\text{int to zero over } [0, T] \\ \text{as } n \neq 0}} (p(t))^2 dt \\ &\quad \underbrace{\text{const over } [0, T]}_{\substack{\text{as } n \neq 0}} \\ &= \int_0^T E_s \cos \theta (p(t))^2 dt \\ &= E_s \cos \theta \underbrace{\int_0^T (p(t))^2 dt}_{\text{unit energy}} \\ &= E_s \cos \theta \end{aligned}$$

(b)

$$\begin{aligned}
 \text{(b)} \quad \|S_a(t)\|^2 &= \int_0^T 2E_s \cos^2(2\pi n t/T) (p(t))^2 dt \\
 &= \int_0^T E_s \left(\underbrace{\cos(4\pi n t/T) + 1}_{\text{int } \frac{t}{T} \text{ over } [0, T]} \right) \underbrace{(p(t))^2 dt}_{\text{int over } [0, T]} \\
 &= E_s \int_0^T (p(t))^2 dt \\
 &\approx E_s \quad \text{most energy}
 \end{aligned}$$

Similarly for S_b, S_c (and same result)

$$\text{so } \|S_a(t)\| = \|S_b(t)\| = \|S_c(t)\| = \sqrt{E_s}$$

$$\begin{aligned}
 d_{ab} &= \sqrt{\|S_a - S_b\|^2} \\
 &= \sqrt{\|S_a\|^2 - 2\langle S_a, S_b \rangle + \|S_b\|^2} \\
 &= \sqrt{E_s - 2\cancel{E_s} + E_s} \\
 &= \sqrt{2E_s}
 \end{aligned}$$

$$\begin{aligned}
 d_{bc} &= \sqrt{\|S_b - S_c\|^2} \\
 &= \sqrt{\|S_b\|^2 - 2\langle S_b, S_c \rangle + \|S_c\|^2} \\
 &= \sqrt{E_s - 2\cancel{E_s} + E_s} \\
 &= \sqrt{2E_s}
 \end{aligned}$$

$$\begin{aligned}
 d_{ac} &= \sqrt{\|S_a - S_c\|^2} \\
 &= \sqrt{\|S_a\|^2 - 2\langle S_a, S_c \rangle + \|S_c\|^2} \\
 &= \sqrt{E_s - 2\cancel{E_s} + E_s} \\
 &= \sqrt{2E_s - 2E_s \cos(\theta)}
 \end{aligned}$$

(5) We noise per unit time, $\sigma = \sqrt{\frac{N_0}{2}}$ $2\sigma = \sqrt{2N_0}$

$$P_{e/s_a} \leq Q\left(\frac{d_{ab}}{2\sigma}\right) + 2\left(\frac{d_{ac}}{2\sigma}\right)$$

$$P_{e/s_b} \leq 2\left(\frac{d_{ab}}{2\sigma}\right) + 2\left(\frac{d_{bc}}{2\sigma}\right)$$

$$P_{e/s_c} \leq 2\left(\frac{d_{ac}}{2\sigma}\right) + 2\left(\frac{d_{bc}}{2\sigma}\right)$$

P A Equiprobable

$$\Rightarrow P_e \leq \frac{2}{3} \cancel{\sum} (P_{e/s_a} + P_{e/s_b} + P_{e/s_c})$$

$$= \frac{2}{3} \left(Q\left(\frac{d_{ab}}{2\sigma}\right) + Q\left(\frac{d_{ac}}{2\sigma}\right) + Q\left(\frac{d_{bc}}{2\sigma}\right) \right)$$

$$= \frac{2}{3} \left(2Q\left(\sqrt{\frac{T_b}{N_0}}\right) + 2\left(\sqrt{\frac{T_b - T_b \cos \theta}{N_0}}\right) \cancel{+ 2\left(\sqrt{\frac{T_b}{N_0}}\right)} \right)$$

(c)

$$P_e \leq \frac{2}{3} \left(2Q\left(\sqrt{10}\right) + 2\left(\sqrt{10(1 - \cos(\frac{\pi}{3}))}\right) \right)$$

$$= \frac{2}{3} \left(2Q\left(\sqrt{6}\right) + 2\left(\sqrt{\frac{10}{2}}\right) 2\left(\sqrt{5}\right) \right)$$

$$\approx \frac{2}{3} \left(2(2.0007888) + 0.0176757 \right)$$

$$\approx 0.004492777$$