

6.1

Solution: For $V_{G1} = V_{G2} = 0$ V, $I_{D1} = I_{D2} = 0.4/2 = 0.2$ mA

To obtain

$$V_{D1} = V_{D2} = 0.1 \text{ V}$$

$$V_{DD} - I_{D1,2} R_D = 0.1$$

$$0.9 - 0.2 R_D = 0.1 \rightarrow R_D = 4 \text{ K}\Omega$$

$$\text{For Q1 and Q2: } I_{D1,2} = \frac{1}{2} \mu_a C_{ox} \left(\frac{W}{L}\right)_{1,2} V_{OV}^2 \rightarrow 0.2 = \frac{1}{2} \times 0.4 \left(\frac{W}{L}\right)_{1,2} \times 0.15^2 \rightarrow \left(\frac{W}{L}\right)_{1,2} = 44.4$$

For Q3:

$$0.4 = \frac{1}{2} \times 0.4 \times \left(\frac{W}{L}\right)_3 \times 0.15^2 \rightarrow \left(\frac{W}{L}\right)_3 = 88.8$$

Since Q3 and Q4 form a current mirror with $I_{D3} = 4I_{D4}$:

$$\left(\frac{W}{L}\right)_4 = \frac{1}{4} \left(\frac{W}{L}\right)_3 = 22.2$$

$$V_{GS4} = V_{GS3} = V_{tn} + V_{OV} = 0.4 + 0.15 = 0.55 \text{ V}$$

$$R = \frac{0.9 - (-0.9) - 0.55}{0.1} = 12.5 \text{ K}\Omega.$$

The lower limit on V_{CM} is determined by the need to keep Q₃ operating in saturation. For this to happen, the minimum value of V_{DS} is $V_{OV} = 0.15$ V. Thus

$$V_{ICMmin} = -V_{SS} + V_{OV3} + V_{GS1,2} = -0.9 + 0.15 + 0.4 + 0.15 = -0.2 \text{ V}$$

The upper limit on V_{CM} is determined by the need to keep Q₁ and Q₂ in saturation, thus

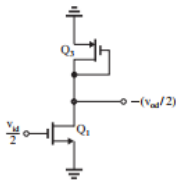
$$V_{ICMmax} = V_t + V_{DD} - \frac{I}{2} R_D = V_{D1,2} + V_{tn} = 0.1 + 0.4 = 0.5 \text{ V}$$

Thus,

$$-0.2 \text{ V} \leq V_{ICM} \leq +0.5 \text{ V}$$

6.2

Solution:



- (a) The figure shows the differential half-circuit. Recalling that the incremental (small-signal) resistance of a diode-connected transistor is given by $(\frac{1}{g_m} || r_o)$, the equivalent load resistance of Q_1 will be: $R_D = \frac{1}{g_{m3}} || r_{o3}$ and the differential gain of the amplifier will be $A_d = \frac{v_{od}}{v_{id}} = g_{m1} [\frac{1}{g_{m3}} || r_{o3} || r_{o1}]$

Since both sides of the amplifier are matched, this expression can be written in a more general way as

$$A_d = g_{m1,2} [\frac{1}{g_{m3,4}} || r_{o3,4} || r_{o1,2}]$$

- (b) Neglecting $r_{o1,2}, r_{o3,4}$ (much larger than $1/g_{m3,4}$),

$$A_d \cong \frac{g_{m1,2}}{g_{m3,4}} = \frac{\sqrt{2\mu_n C_{ox}(W/L)_{1,2}(I/2)}}{\sqrt{2\mu_p C_{ox}(W/L)_{3,4}(I/2)}} = \sqrt{\frac{\mu_n(W/L)_{1,2}}{\mu_p(W/L)_{3,4}}}$$

- (c)

$$\mu_n = 4\mu_p \text{ and all channel lengths are equal, } A_d = 2 \sqrt{\frac{W_{1,2}}{W_{3,4}}} ; A_d = 10 \rightarrow 10 = 2 \sqrt{\frac{W_{1,2}}{W_{3,4}}} \rightarrow \frac{W_{1,2}}{W_{3,4}} = 25$$

6.3

Solution:

The value of R is found as follows:

$$R = \frac{V_{G6} - V_{G7}}{I_{REF}} = \frac{0.8 - (-0.8)}{0.2} = 8 \text{ k}\Omega$$

Since $I = I_{REF}$, Q_3 and Q_6 are matched and are operating at $|V_{OV}| = 1.5 - 0.8 - 0.5 = 0.2 \text{ V}$

Thus,

$$0.2 = \frac{1}{2} \times 0.1 \times \left(\frac{W}{L}\right)_{6,3} \times 0.2^2 \rightarrow \left(\frac{W}{L}\right)_3 = \left(\frac{W}{L}\right)_6 = 100$$

Each of Q_4 and Q_5 is conducting a ds current of $(I/2)$ while Q_7 is conducting a dc current $I_{REF} = I$. Thus, Q_4 and Q_5 are matched and their W/L ratios are equal while Q_7 has twice the (W/L) ration of Q_4 and Q_5 . Thus,

$$\frac{I}{2} = \frac{1}{2} \mu_n C_{ox} \left(\frac{W}{L}\right)_{4,5} V_{OV4,5}^2 ; \text{ where: } V_{OV4,5} = -0.8 - (-1.5) - 0.5 = 0.2 \text{ V}$$

Thus,

$$0.1 = \frac{1}{2} \times 0.25 \times \left(\frac{W}{L}\right)_{4,5} \times 0.04 \rightarrow 0.1 = \frac{1}{2} \times 0.25 \times \left(\frac{W}{L}\right)_{4,5} \times 0.04 \rightarrow \left(\frac{W}{L}\right)_4 = \left(\frac{W}{L}\right)_5 = 20$$

$$\text{And } \left(\frac{W}{L}\right)_7 = 40$$

$$r_{o4} = r_{o5} = \frac{|V_{AP}|}{\frac{I}{2}} = \frac{10}{0.1} = 100 \text{ k}\Omega$$

$$r_{o1} = r_{o2} = \frac{|V_{AP}|}{\frac{I}{2}} = \frac{10}{0.1} = 100 \text{ k}\Omega$$

$$A_d = g_{m1,2}(r_{o1,2} || r_{o4,5}) \rightarrow 50 = g_{m1,2}(100 || 100) \rightarrow g_{m1,2} = 1 \text{ mA/V}$$

But

$$g_{m1,2} = \frac{2(\frac{I}{2})}{|V_{OV1,2}|} \rightarrow 1 = \frac{0.2}{|V_{OV1,2}|} \rightarrow |V_{OV1,2}| = 0.2 \text{ V}$$

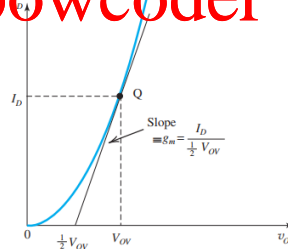


Figure 7.14 The slope of the tangent at the bias point Q intersects the v_{ov} axis at $\frac{1}{2} V_{ov}$. Thus, $g_m = I_D / (\frac{1}{2} V_{ov})$.

The $\frac{W}{L}$ ratio for Q_1 and Q_2 can now be determined from:

$$0.1 = \frac{1}{2} \times 0.1 \times \left(\frac{W}{L}\right)_{1,2} \times 0.2^2 \rightarrow \left(\frac{W}{L}\right)_1 = \left(\frac{W}{L}\right)_2 = 50$$

A summary of the results is provided in the table below:

Transistor	W/L	I_D (mA)	$ V_{GS} $ (V)
Q_1	50	0.1	0.7
Q_2	50	0.1	0.7
Q_3	100	0.2	0.7
Q_4	20	0.1	0.7
Q_5	20	0.1	0.7
Q_6	100	0.2	0.7
Q_7	40	0.2	0.7