DMUSIC Algorithm

Pedro Pons-Villalonga, 2021

1 Singular Value Decomposition

A matrix $M \in \mathbb{C}^{m \times n}$ can be factorized as:

$$M = USV^* \tag{1}$$

Where:

- $U \in \mathbb{C}^{m \times m}$ is unitary.
- $S \in \mathbb{R}^{m \times n}$ is a rectangular diagonal matrix of positive real numbers.
- $V \in \mathbb{C}^{n \times n}$ is unitary.
- V^* denotes the complex conjugate of V.

The elements of the diagonal of S are called $singular\ values$, and are usually sorted in descending order (important). The columns of U and V are orthonormal bases, and we can say that the matrix M maps the vectors \mathbf{v}_i to the stretched unit vector $\sigma_i \mathbf{u}_i$, where σ_i is the i-th singular value (i-th element of the diagonal of S), as:

$$MV = USV^*V = US \tag{2}$$

2 DMUSIC

Consider a signal y(t), that we assume can be written as a sum of K (damped) harmonics and white noise W(t):

$$y(t) = \sum_{k=1}^{K} A_k e^{-a_k t + i\omega_k t} + W(t) \equiv \sum_{k=1}^{K} A_k e^{s_k t} + W(t)$$
 (3)

Or, for a digitized signal:

$$y(n) = \sum_{k=1}^{K} A_k e^{s_k \tau n} + W(n)$$
 (4)

Where $\tau=1/F_{\rm s}$, the inverse of the sampling frequency, is the time interval between digitizations.

We want to recover from the signal the frequencies ω_k and damping rates a_k . For that, if the signal is digitized, we make the assumption that we can predict the signal in an interval of length N using the preceding J (> K) values:

$$y(n) = \sum_{i}^{J} c_{(J-i)} y(n-i)$$
 (5)

We can build a *prediction matrix* $A \in \mathbb{R}^{(N-J)\times J}$, so that:

$$\begin{pmatrix} y(1) & y(2) & \cdots & y(J) \\ y(2) & y(3) & \cdots & \\ \vdots & & \ddots & \\ y(N-J) & \cdots & & y(N) \end{pmatrix} \begin{pmatrix} c_1 \\ c_2 \\ \vdots \\ c_J \end{pmatrix} = \begin{pmatrix} y(J+1) \\ y(J+2) \\ \vdots \\ y(N) \end{pmatrix}$$
(6)

The matrix A is a Hankel matrix. If the noise was zero, the matrix A can be expressed as a Vandermonde decomposition:

$$A = S_L C S_R^* \tag{7}$$

Where $C \in \mathbb{C}^{K \times K}$ is a diagonal matrix, and the k-th columns of the matrices $S_L \in \mathbb{C}^{(N-J) \times K}$ and $S_R \in \mathbb{C}^{J \times K}$ are given by the so-called *left* and *right signal vectors*:

$$\vec{r}_L(s_k) = \begin{pmatrix} 1 & e^{s_k \tau} & e^{2s_k \tau} & \cdots & e^{(N-J-1)s_k \tau} \end{pmatrix}^T$$
 (8)

$$\vec{r}_R(s_k) = \begin{pmatrix} 1 & e^{s_k \tau} & e^{2s_k \tau} & \cdots & e^{(J-1)s_k \tau} \end{pmatrix}^T$$
 (9)

From the right signal vectors we can define the signal space:

$$\Omega^{S} \equiv \operatorname{span}\left(\left\{\vec{r}_{R}(s_{k})\right\}\right) \tag{10}$$

and the null space:

$$\Omega^N \equiv \ker\left(A\right) \tag{11}$$

We can perform a singular value decomposition on matrix A, obtaining:

$$A = USV^* \tag{12}$$

In this case, $U \in \mathbb{C}^{(N-J)\times (N-J)}$, $V \in \mathbb{C}^{J\times J}$ are square matrices and $S \in \mathbb{R}^{(N-J)\times J}$ is a diagonal rectangular matrix of real, positive numbers.

In the absence of noise, the rank of S is K, and only the first K singular values (elements of the diagonal of S) are non-zero. The first K columns of U and V, then, contain all the information from the signal, and the first K columns of V span the signal space Ω^S . Consequently, as the columns $\{K+1, \dots, J\}$ of V correspond with null singular values, they belong to the null space.

If there is noise, the separation between signal space and null space (or, more appropriately, $noise\ space$) becomes less clear. However, if the signal-to-noise ratio is sufficiently high, we can safely assume that the information of the signal remains in the first K columns of U and V.

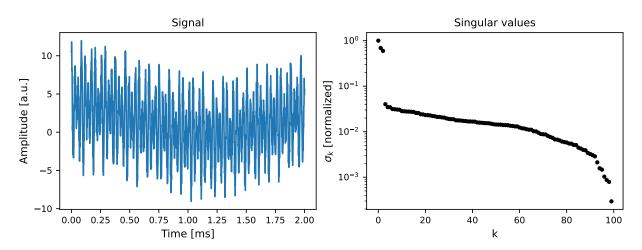


Figure 1: Synthetic (noisy) signal and singular values of the SVD decomposition of its A matrix. N=200, K=3.

We want, given a real signal (of which we do not know the number of harmonics K) to estimate its dominant frequencies and corresponding damping rates. For that, we need a good estimate of K, which can be obtained by looking at the SVD decomposition of the signal. Lower values of K can misidentify signal components as noise, and viceversa for higher values of K.

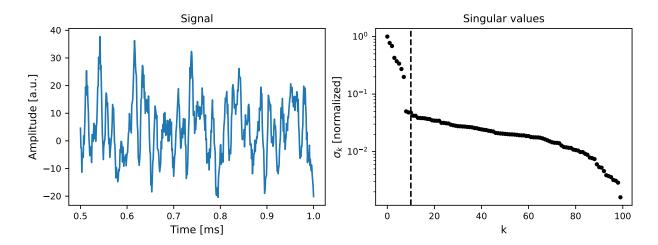


Figure 2: Synthetic (noisy) signal and singular values of its SVD decomposition. N=200, K=10. The separation between signal and noise space is not that clear, maybe in part because the amplitudes of some harmonics are small.

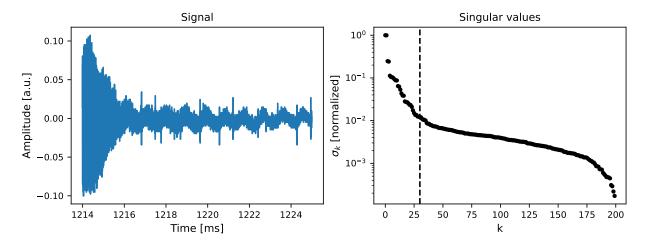


Figure 3: Real signal and singular values. Vertical line at K=30 shows the justification for a commonly used value of K. N=400.

We can define a new matrix, V_N , given by the columns $(K+1,\cdots,J)$ of V. Then, all vectors $\vec{r}(s) \in \mathbb{C}^J$:

$$\vec{r}(s) = (1, e^{s\tau}, e^{2s\tau}, \dots, e^{(J-1)s\tau})$$
(13)

that verify:

$$\left\| V_N \vec{r}(s) \right\| = 0 \tag{14}$$

Will be orthogonal to the noise space and thus belong to the signal space. In the presence of noise, the true equality is not necessarily satisfied. However, the function:

$$P(s) \equiv \frac{1}{\left\|V_N \vec{r}(s)\right\|} \tag{15}$$

will still present maxima when $s \simeq s_k$, so we can use it to estimate the spectrum of the signal. To do that, we must perform a scan over a suitable grid. As we are interested only in the frequencies of the maxima, we integrate over the real part of the grid to get the frequency estimator:

$$p(\omega) = \int_{a_0}^{a_1} P(-a + i\omega) \, da \tag{16}$$

3 Implementation

A simple, reasonably fast implementation has been made with Python and MATLAB. We must realize that the bulk of the computational effort will be spent in the matrix multiplication $V_N \vec{r}(s)$. Python is relatively slow with loops, but the matrix multiplication routines of the package numpy are optimized. Because of that, the optimal solution consists in the creation of a big matrix R of all the $\vec{r}(s)$, that will be multiplied by V_N only once. This multiplication is done in parallel automatically.

Algorithm 1 DMUSIC spectrogram

```
procedure DMUSIC_SPEC(t, y, N, K, overlap)

    b t: time array, y: raw signal

     J \leftarrow N//2
                                                                                                                   > For performance
    f_{\text{arr}} \leftarrow \text{linspace}(0, F_{\text{nyq}}, n_f)
                                                                    \triangleright nf: number of frequencies to scan (typ. \sim 500)
     a_{arr} \leftarrow linspace(0, -1, n_a)
                                                              > na: number of damping factors to scan (typ. ~ 50)
    \mathbf{R} \leftarrow \mathbf{Empty} \ \mathbf{matrix} \in \mathbb{C}^{J \times (n_f \cdot n_a)}

    This way it is only created once

    for f_i \leftarrow w_arr do
          for a_i \leftarrow a_a rr do
               s \leftarrow a_i + i f_i
               Make \vec{r}(s) (normalized)
               Append \vec{r}(s) to R
          end for
     end for
     Find nsteps
    P_f \leftarrow \text{Empty matrix} \in \mathbb{R}^{\text{nsteps} \times n_f}
    i ← 0
     while i < len(y) do
         ym \leftarrow Hankel(y[i:i+N])
          U, S, V_T \leftarrow \text{svd}(ym)
          V_N \leftarrow V_T[\texttt{K}:(\texttt{N-J})]
                                                                                              \triangleright Last rows of V_T, starting at K
                                                                                                                             \triangleright P_s \in \mathbb{R}^{n_f \cdot n_a}
         P_s \leftarrow \text{norm}(V_N R)
         i \leftarrow i + (1 - overlap) \cdot N
          Append to P_f \leftarrow \text{Integral of } P_s \text{ over } a_{\text{arr}} \text{ in intervals of length } n_a
     end while
    P_f \leftarrow 10 \cdot \log_{10}(P_f/\text{max}(P_f))
end procedure
```

References

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