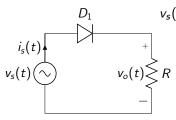
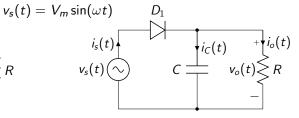
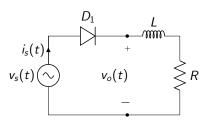
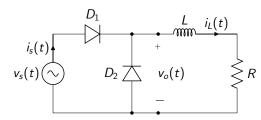
## Single-phase half-wave (1-pulse) rectifiers





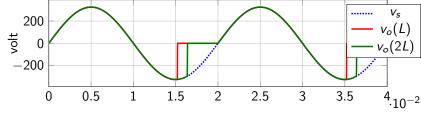


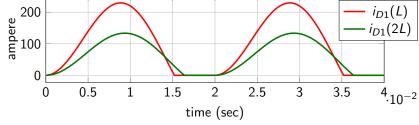




#### 1-ph half-wave rectifier [(R-L load) / (L-filter and R-load)]

Waveforms in periodic steady state





 $v_s = 230\sqrt{2}\sin(2\pi \cdot 50 \cdot t)$ V, R=0.5 $\Omega$ , L=6.5mH,  $i_{D1} = i_s$ 

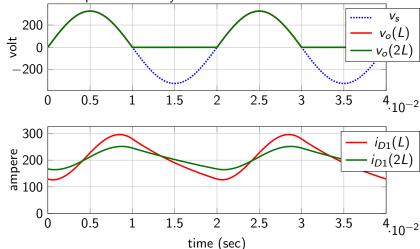


#### 1-ph half-wave rectifier (Free-wheeling diode, L-filter and R-load)

Waveforms: Start-up transient and periodic steady state volt -200·10<sup>-3</sup> ampere ·10<sup>-3</sup> time (sec)

#### 1-ph half-wave rectifier (Free-wheeling diode, L-filter and R-load)

Waveforms in periodic steady state

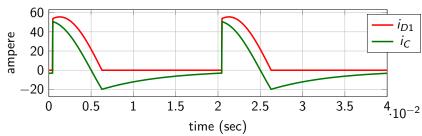


 $v_s = 230\sqrt{2}\sin(2\pi \cdot 50 \cdot t)$ V, R=0.5 $\Omega$ , L=6.5mH,  $i_{D1} = i_s$  Diode currents?

## 1-ph half-wave rectifier (Capacitive filter and R load)

Waveforms in periodic steady state



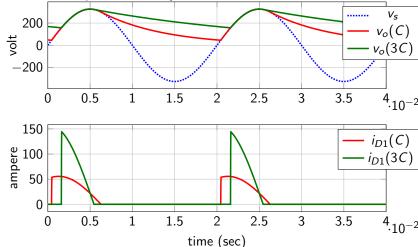


 $v_s = 230\sqrt{2}\sin(2\pi \cdot 50 \cdot t)V$ , R=15 $\Omega$ , C=0.5mF,  $i_{D1} = i_s$ 



# 1-ph half-wave rectifier (Capacitive filter and R-load)

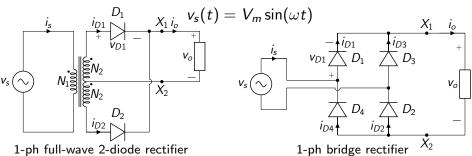
Waveforms in periodic steady state

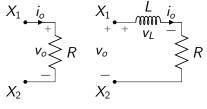


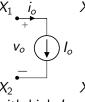
 $v_s = 230\sqrt{2}\sin(2\pi \cdot 50 \cdot t)$ V, R=15 $\Omega$ , C=0.5mF,  $i_{D1} = i_s$  vs  $i_o$ !

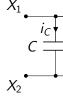


## Single-phase full-wave (2-pulse) rectifiers







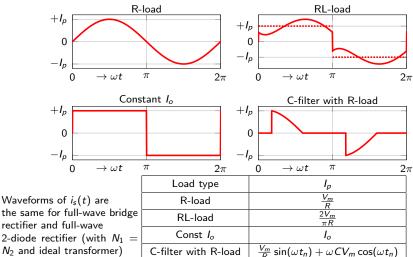




R with high L

EN 313 - Power Electronics

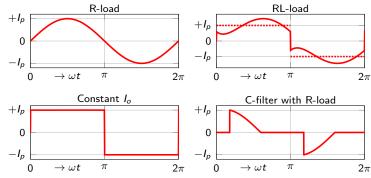
# 1-ph full-wave rectifier: Waveforms of input current $i_s(t)$





 $\omega t_n$  is the instant at which the diodes turn-ON

# 1-ph full-wave rectifier: Waveforms of input current $i_s(t)$

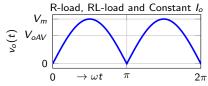


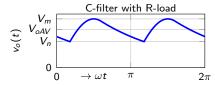
In the case of constant current  $(I_o)$  load,

- $I_{sRMS} = I_o; \quad \text{Fundamental } I_{s1RMS} = \frac{4I_o}{\pi} \frac{1}{\sqrt{2}} = 0.9I_o; \quad \text{n}^{th} \text{ harmonic } I_{snRMS} = \frac{I_{s1RMS}}{n}$
- ▶ DPF = 1 and  $DF_1 = 0.9$  ⇒ power factor PF = 0.9
- $ightharpoonup i_s(t)$  doesn't contain even harmonics
- ► Total harmonic distortion  $I_{sTHD} = 48.43\%$



# 1-ph full-wave rectifier: Waveforms of output voltage $v_o(t)$





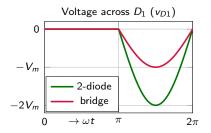
 $V_n = V_m \sin(\omega t_n)$  and  $\omega t_n$  is the instant at which the diodes turn-ON

 $K = \omega RC$ 

	$\mathbf{v}_n = \mathbf{v}_m \sin(\omega t_n)$ and $\omega t_n$ is the instant at which the diodes turn-ON				
	Load type	$V_{oAV}$			
	R-Load	21/			
	RL-Load	$\frac{2V_m}{}$			
	Const $I_o$	$\pi$			
	C-filter with R-load	$\frac{V_m}{\pi} \left[ \cos(\omega t_n) + \sqrt{1 + K^2} - \frac{K^2}{\sqrt{1 + K^2}} e^{\frac{-\tan^{-1}K - \omega t_n}{K}} \right]$			

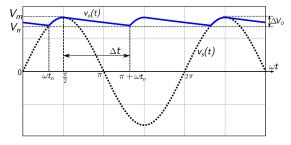


#### 1-ph full-wave rectifier: Voltage across diode



- Peak inverse voltage (PIV) across each diode in a 2-diode rectifier (with  $N_1 = N_2$ ) is twice that in a bridge rectifier
- In practical diodes,  $v_{D1} \neq 0$  when  $D_1$  is ON. Forward voltage drop across two diodes appears between source and load in bridge rectifier, while the center-tap (2-diode) rectifier has only one diode-voltage-drop between source and load

## 1-ph full-wave rectifier: Design of C-filter



Specifications: (i) Input voltage and frequency ( $V_m$  and  $\omega$ ) (ii) average output power ( $P_o$ ) (iii) peak-to-peak ripple in output voltage ( $\Delta V_o$ )

Assumptions: (i) Diodes stop conducting at  $\omega t=\pi/2$  and (ii) ripple in output power is negligible

Peak-to-peak ripple 
$$\Delta V_o = V_m - V_n = V_m - [V_m \sin(\omega t_n)] \Rightarrow \omega t_n = \sin^{-1}[1 - (\Delta V_o/V_m)]$$

Output power  $P_o$  is supplied only by the capacitor for the duration  $\Delta t = \frac{(\pi/2) + \omega t_n}{\omega}$ 



Change in energy of the capacitor  $\frac{1}{2}C\left(V_m^2-V_n^2\right)=P_o\Delta t$ 

#### Connection of multiple diodes

#### Common cathode connection

Diode with its anode at the most positive voltage conducts

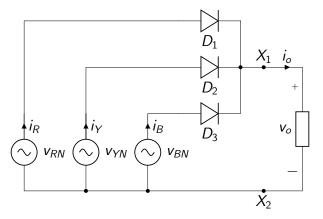
#### Common anode connection

Diode with its cathode at the most negative voltage conducts





## 3-ph half-wave (3-pulse) rectifier



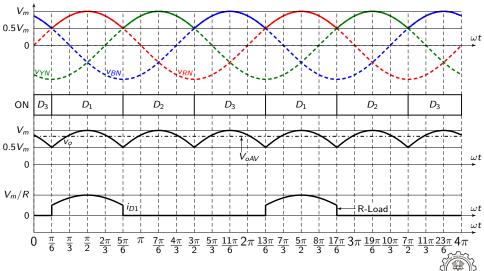
$$v_{RN} = V_m \sin(\omega t)$$
  $v_{YN} = V_m \sin\left(\omega t - \frac{2\pi}{3}\right)$   $v_{BN} = V_m \sin\left(\omega t + \frac{2\pi}{3}\right)$ 

 $i_R = i_{D1}$  and similar definitions for  $i_Y \& i_B$ 



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#### 3-ph half-wave (3-pulse) rectifier



## 3-ph half-wave (3-pulse) rectifier

Average output voltage

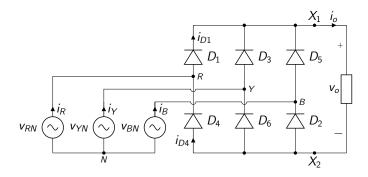
$$V_{oAV} = rac{1}{(2\pi/3)} \int_{rac{\pi}{6}}^{rac{5\pi}{6}} V_m \sin(\omega t) \ d(\omega t) = rac{3\sqrt{3}V_m}{2\pi}$$

- Peak-to-peak ripple in output voltage  $V_{OPP} = 0.5 V_m$
- ► Each diode conducts for 120° (one-third of a line cycle)
- ▶ Input current  $i_R(t)$  contains DC component (finite average value)





## 3-ph full-wave/bridge (6-pulse) rectifier



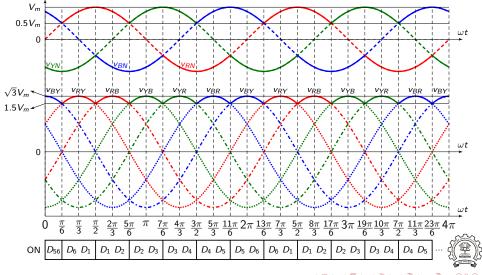
$$v_{RN} = V_m \sin(\omega t)$$
  $v_{YN} = V_m \sin\left(\omega t - \frac{2\pi}{3}\right)$   $v_{BN} = V_m \sin\left(\omega t + \frac{2\pi}{3}\right)$ 

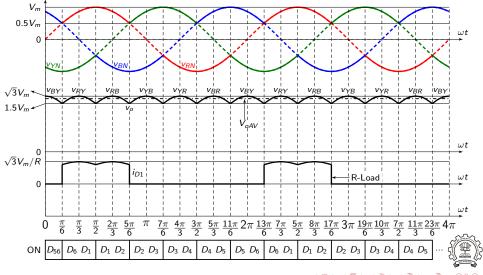
$$v_{BN} = V_m \sin\left(\omega t + \frac{2\pi}{3}\right)$$

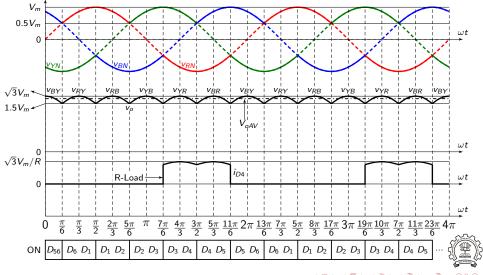
 $i_R = i_{D1} - i_{D4}$  and similar definitions for  $i_Y \& i_B$ 

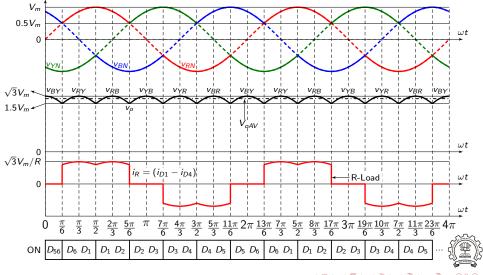
PIV for each diode is  $\sqrt{3}V_m$ 











Average output voltage

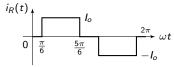
$$V_{oAV} = \frac{1}{(\pi/3)} \int_{\frac{\pi}{6}}^{\frac{\pi}{2}} v_{RY}(t) \ d(\omega t)$$
$$= \frac{3}{\pi} \int_{\frac{\pi}{6}}^{\frac{\pi}{2}} \sqrt{3} V_m \sin\left(\omega t + \frac{\pi}{6}\right) \ d(\omega t) = \frac{3\sqrt{3} V_m}{\pi}$$

- Maximum value of  $v_o(t)$  is peak line-to-line voltage  $\sqrt{3}V_m$
- Peak-to-peak ripple in output voltage  $V_{oPP} = \sqrt{3}V_m 1.5V_m = 0.232V_m$
- ► Frequency of output voltage ripple is 300Hz (6 times the line frequency)
- $\blacktriangleright$  Each diode conducts for  $120^\circ$  (one-third of a line cycle), two diodes conduct together
- Input current  $i_R(t)$  doesn't contain even harmonics; doesn't contain triplen harmonics





In the case of constant current  $(I_o)$  load,  $i_R(t)$  has the shape of a quasi-square wave



- $I_{RRMS} = I_o \sqrt{\frac{2}{3}}; \quad \text{Fundamental } I_{R1RMS} = \frac{\sqrt{6}}{\pi} I_o; \quad \text{n}^{th} \text{ harmonic } I_{RnRMS} = \frac{I_{R1RMS}}{n}$
- ▶ DPF = 1 and  $DF_1 = 0.955 \Rightarrow$  power factor PF = 0.955
- ▶ Order of harmonics in  $i_R(t)$  is 5,7,11,13,17,19,... or  $6m \pm 1$ , m = 1, 2, 3, ...
- ▶ Total harmonic distortion  $I_{RTHD} = 31.08\%$





#### Effect of source inductance: Commutation (Constant load current Io)

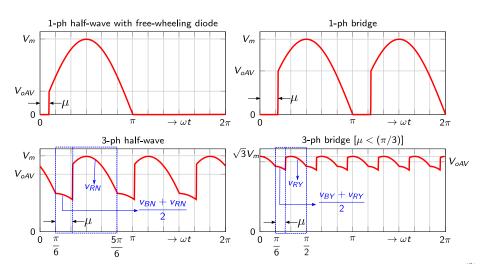
- Commutation: The process of current being transferred from one diode to the other
- Commutating inductance  $L_s$  is in series with  $v_s$  for 1-phase rectifiers
- For 3-phase rectifiers,  $L_s$  is in series with each of the sources,  $v_{RN}$ ,  $v_{YN}$  and  $v_{BN}$
- Commutation interval (in terms of angle) is denoted by  $\mu$
- Effects of commutation
  - Load regulation (Variation of output voltage with load)
  - Change in the shape of source (input) current waveform

Type of rectifier	$\cos \mu$	$V_{oAV}$	
1-ph half-wave (free-wheeling diode)	$1 - \frac{\omega L_s I_o}{V_m}$	$\frac{V_m}{\pi} \left( 1 - \frac{\omega L_s I_o}{2V_m} \right)$	
1-ph bridge	$1 - \frac{2\omega L_s I_o}{V_m}$	$\frac{2V_m}{\pi}\left(1-\frac{\omega L_s I_o}{V_m}\right)$	
3-ph half-wave	$1 - \frac{2\omega L_s I_o}{\sqrt{3} V_m}$	$\frac{3\sqrt{3}V_m}{2\pi}\left(1-\frac{\omega L_s I_o}{\sqrt{3}V_m}\right)$	
3-ph bridge	$1 - \frac{2\omega L_s I_o}{\sqrt{3} V_m}$	$\frac{3\sqrt{3}V_m}{\pi}\left(1-\frac{\omega L_s I_o}{\sqrt{3}V_m}\right)$	





# Effect of source inductance: Waveforms of output voltage $v_o(t)$

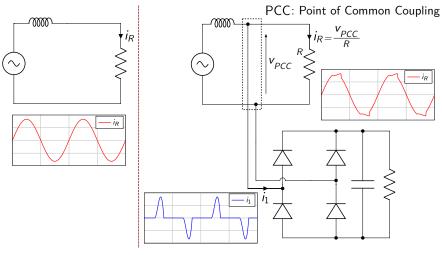






Pavan Kumar Hari (DESE, IIT Bombay)

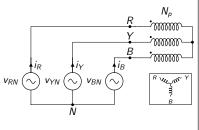
#### Input current of a rectifier: Effect on other loads



#### 3-phase 12-pulse series-type diode rectifier

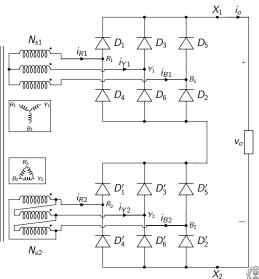
$$\frac{N_{s2}}{N_p} = \sqrt{3} \frac{N_{s1}}{N_p}$$

 $V_{R2Y2}$  and  $V_{R1Y1}$  have equal magnitudes  $V_{R2Y2}$  leads  $V_{R1Y1}$  by 30°



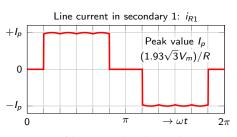
$$i_{R} = \left(\frac{N_{s1}}{N_{p}}\right) i_{R1} + \left(\frac{N_{s2}}{N_{p}}\right) \left(\frac{i_{R2} - i_{B2}}{3}\right)$$

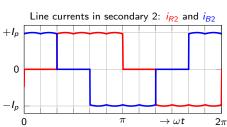
Ideal transformer

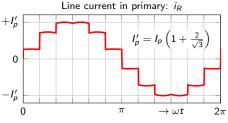


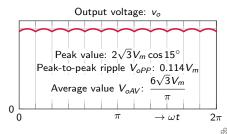
45 / 120

#### 3-phase 12-pulse diode rectifier: R-load





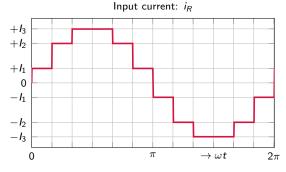




 $N_{s1} = N_p$ 

## 3-phase 12-pulse diode rectifier (constant current Io load)

$$N_{s1} = N_p$$



$$I_1 = \frac{I_o}{\sqrt{3}}$$

$$I_2 = I_o \left( 1 + \frac{1}{\sqrt{3}} \right)$$

$$I_3 = I_o \left( 1 + \frac{2}{\sqrt{3}} \right)$$

- $I_{RRMS} = I_o \sqrt{\frac{4}{3} + \frac{2}{\sqrt{3}}}; \quad \text{Fundamental } I_{R1RMS} = \frac{2\sqrt{6}}{\pi}I_o$
- ▶ DPF = 1 and  $DF_1 = 0.9886 \Rightarrow$  power factor PF = 0.9886
- ▶ Order of harmonics in  $i_R(t)$  is 11,13,23,25,... or  $12m \pm 1$ , m = 1,2,3,...
- ► Total harmonic distortion  $I_{RTHD} = 15.24\%$



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#### Module 1: Summary

- Diode characteristics and loss calculations
- ▶ 1-pulse, 2-pulse, 3-pulse, 6-pulse and 12-pulse rectifiers
- ▶ R-load, constant current load (R with high L) and C-filter with R-load
- ► Effect of source inductance: commutation, load regulation
- Periodic steady-state and measures of periodic waveforms
- Fourier analysis and waveform symmetries
- Harmonic distortion at point of common coupling



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