### **CS221: Digital Design**

# RTL Design: Serial Multiplier

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### **Outline**

- RTL Design
- Modulo 14 Counter Example
- Serial Multiplier Example
- Can we automate this RTL design process?
  - Given C code/Parallel Code

## Reference Material for Lec 33, 34, 35

- Chapter 8 of Mano Book
  - Design at Register Transfer Level
  - Classic Example: Serial Binary Multiplication
- Chapter 15 of Kumar Book
  - Section 15.5.1: Data path Subsystem for Binary Multiplier

# ASM Charts: An Complete Example Ref: Mano Book

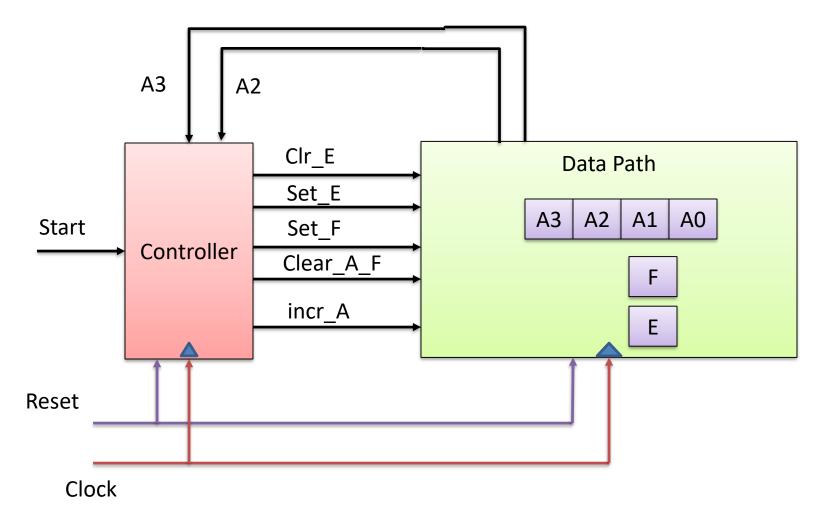
# **ASM Charts: An Example**

### **Mod 14 counter:**

- There is a 4 bit counter (A)
- E specify: less than 12 or less than 4
- EF=11 specify : value =12,
  - It time to reset after next counting
- E depends of A<sub>2</sub>, F depends on A<sub>2</sub>, A<sub>3</sub>

If 
$$A_2=1 \rightarrow E=1$$
, else E=0  
 $A_3A_2=1$ ,  $\rightarrow F=1$ , counter reset

# **AMS DP+CP: to be High Level**

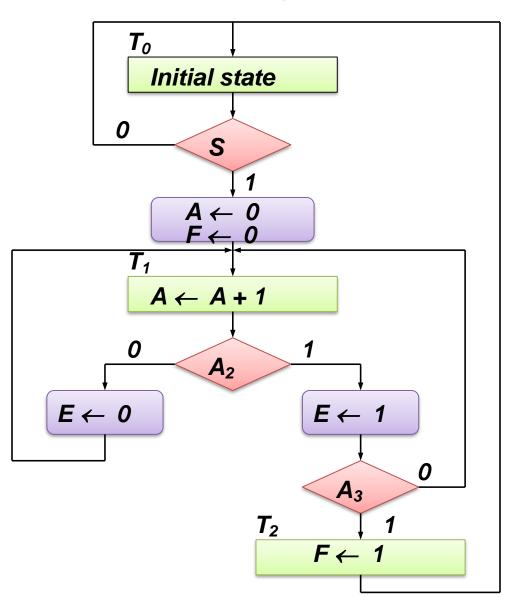


# **ASM Charts: An Example**

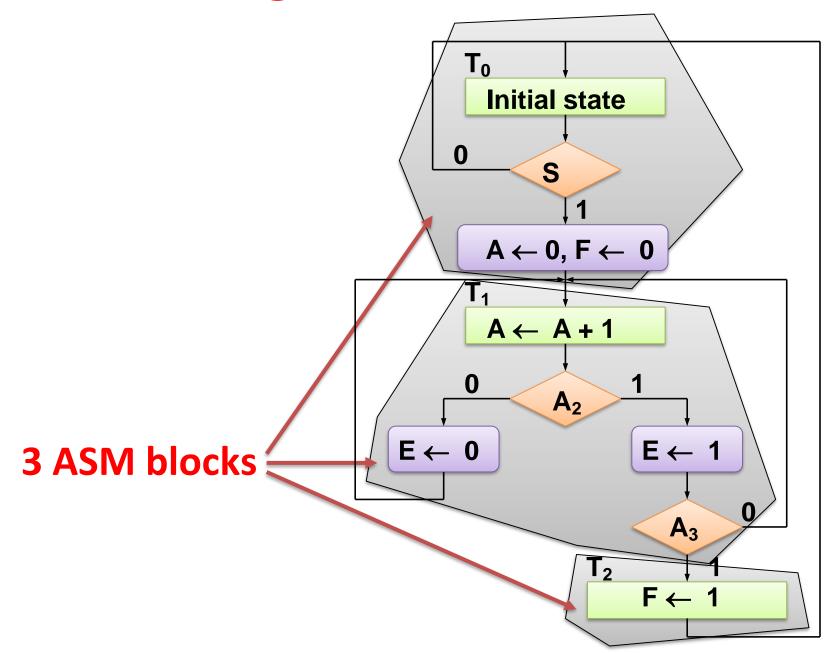
- A is a register;
- A<sub>i</sub> stands for i<sup>th</sup> bit of the A register.

$$A = A_3 A_2 A_1 A_0$$

 E and F are single-bit flipflops.

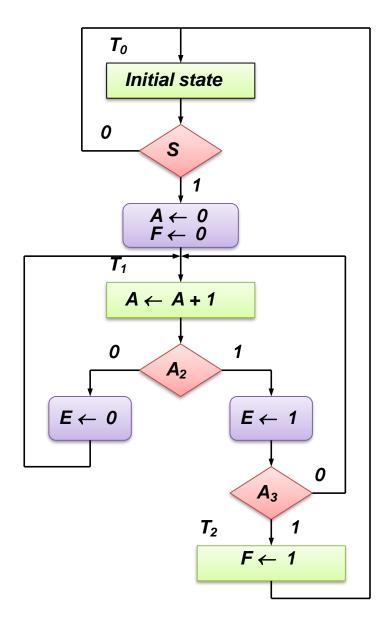


- Operations of ASM can be illustrated through a timing diagram.
- Two factors which must be considered are
  - Operations in an ASM block occur at the same time in one clock cycle
  - Decision boxes are dependent on the status of the *previous clock cycle* (that is, they do not depend on operations of current block)



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CTR	E, F	Conditions	State
0000	1,0	A <sub>2</sub> =0,A <sub>3</sub> =0	T1
0001	0,0		
0010	0,0		
0011	0,0		
0100	0,0	A <sub>2</sub> =1,A <sub>3</sub> =0	
0101	1,0		
0110	1,0		
0111	1,0		
1000	1,0	A <sub>2</sub> =0,A <sub>3</sub> =1	
1001	0,0		
1010	0,0		
1011	0,0		
1100	0,0	A <sub>2</sub> =1,A <sub>3</sub> =1	
1101	1,0		T2
1101	1,1		то



# **ASM Chart => Digital System**

- ASM chart describes a digital system. From ASM chart, we may obtain:
  - Controller logic (via State Table/Diagram)
  - Architecture/Data Processor
- Design of controller is determined from the decision boxes and the required state transitions.
- Design requirements of data processor can be obtained from the operations specified with the state and conditional boxes.

### **ASM Chart => Controller**

### Procedure:

- Step 1: Identify all states and assign suitable codes.
- Step 2: Formulate state table using

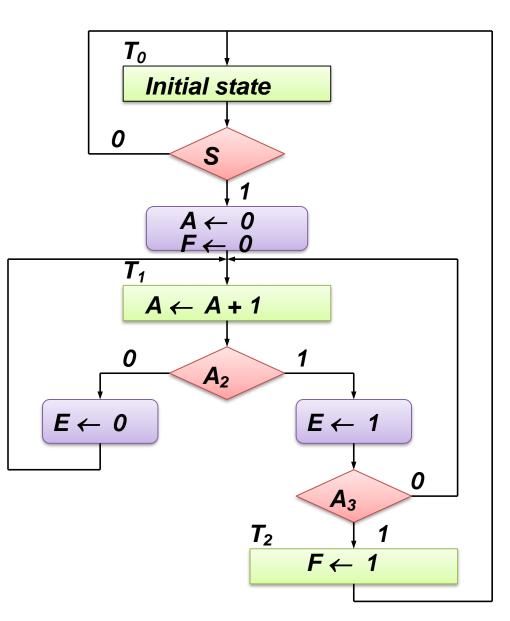
**State** from state boxes

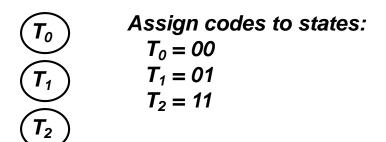
**Inputs** from decision boxes

**Outputs** from operations of state/conditional boxes.

Step 3: Obtain state/output equations and draw circuit.

### **ASM Chart => Controller**





Present state		inputs			Next state		outputs		
G <sub>1</sub>	$G_0$	S	A <sub>2</sub>	<b>A</b> <sub>3</sub>	G <sub>1</sub> <sup>+</sup>	$G_0^+$	T <sub>0</sub>	T <sub>1</sub>	T <sub>2</sub>
0	0	0	X	X	0	0	1	0	0
0	0	1	X	X	0	1	1	0	0
0	1	X	0	X	0	1	0	1	0
0	1	X	1	0	0	1	0	1	0
0	1	X	1	1	1	1	0	1	0
1	1	X	X	X	0	0	0	0	1

Inputs from conditions in decision boxes.

Outputs = present state of controller.

### **ASM Chart => Architecture/Data Processor**

- Architecture is more difficult to design than controller.
- Nevertheless, it can be deduced from the ASM chart. In particular, the operations from the ASM chart determine:
  - What registers to use
  - How they can be connected
  - What operations to support
  - How these operations are activated.
- Guidelines:
  - always use high-level units
  - simplest architecture possible.

# ASM Chart => Architecture/Data Processor

### Various operations are:

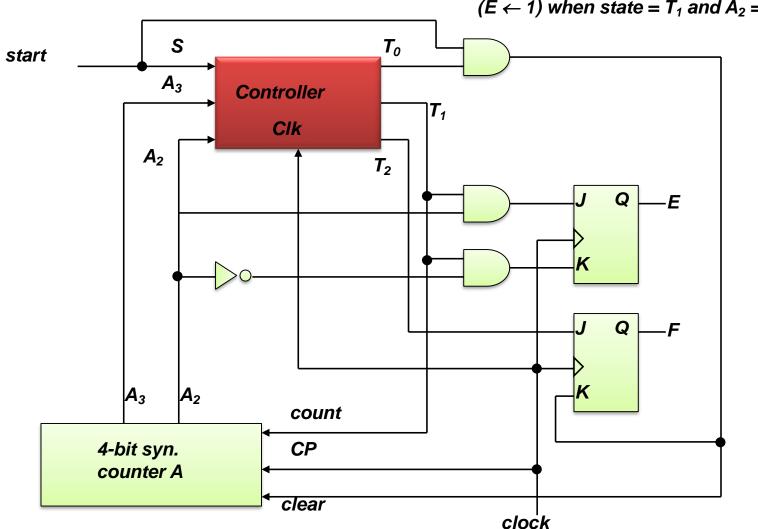
- Counter incremented (A  $\leftarrow$  A + 1) when state =  $T_1$ .
- Counter cleared (A  $\leftarrow$  0) when state = T<sub>0</sub> and S = 1.
- E is set (E ← 1) when state =  $T_1$  and  $A_2$  = 1.
- E is cleared (E ← 0) when state =  $T_1$  and  $A_2$  = 0.
- F is set (F ← 0) when state =  $T_2$ .
- F is cleared (F  $\leftarrow$  0) when state = T<sub>0</sub> and S = 1.

### • Deduce:

- One 4-bit register A (e.g.: 4-bit synchronous counter with clear/increment).
- Two flip-flops needed for E and F (e.g.: JK/D flip-flops).

# 

 $(A \leftarrow A + 1)$  when state =  $T_1$ .  $(A \leftarrow 0)$  when state =  $T_0$  and S = 1.  $(E \leftarrow 1)$  when state =  $T_1$  and  $A_2 = 1$ .



# RTL/ASM example for 8 bit Sequential Multiplier

Ref: Mano Book

# **RTL: Multiplier Example**

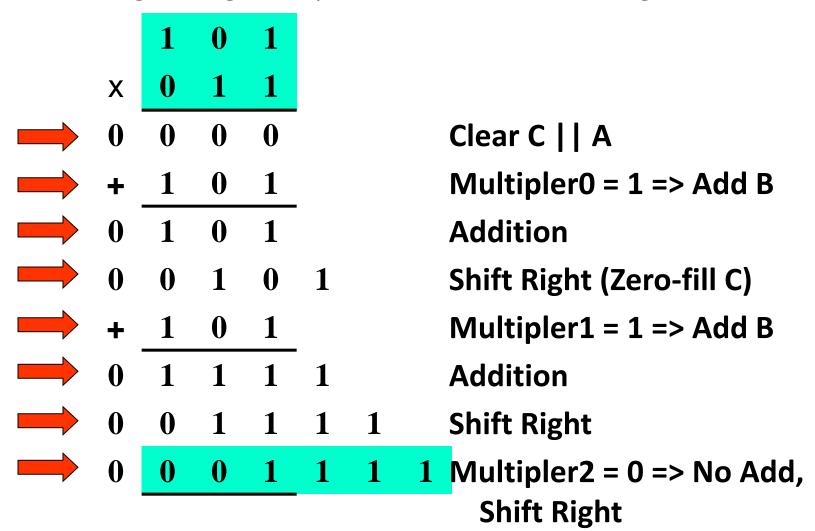
- Example: (101 x 011) Base 2
- Note that the partial product summation for n digits, base
   2 numbers requires adding up to n digits (with carries) in a column.
- Note also n x m digit multiply generates up to an m + n digit result (same as decimal).

Partial products are: 101 x 0, 101 x 1, and 101 x 1

			1	0	1
		X	0	1	1
			1	0	1
		1	0	1	
	0	0	0		
0	0	1	1	1	1

# Example (1 0 1) x (0 1 1) Again

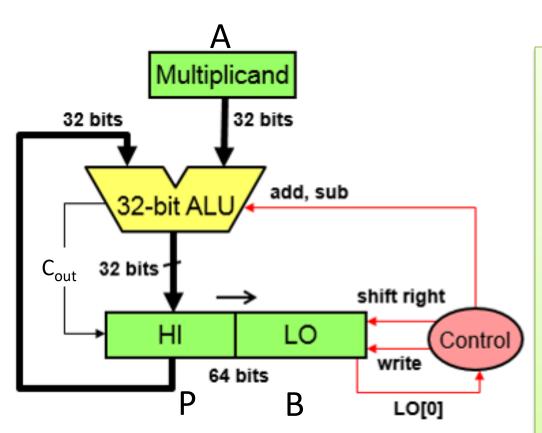
Reorganizing example to follow hardware algorithm:



# **Binary Multiplication**

- Polynomial Multiplication
  - You can think Binary number A, B as polynomials
  - $-A(X)=A_{n-1}.2^{n-1}+....+A_2.2^2+A_1.2^1+A_0.2^0$
  - Multiply polynomial to get another one
  - Time complexity: O(n²) Basic serial Algorithm,
     O(n¹.5) for divide conquer approach, O(n lgn) using
     FFT
- Booth Algorithm reduce number Partial addition
  - 99 Represent using 100-1, 95 is 100-5
  - 111 represent as 1000-1 in binary: possibly reduce

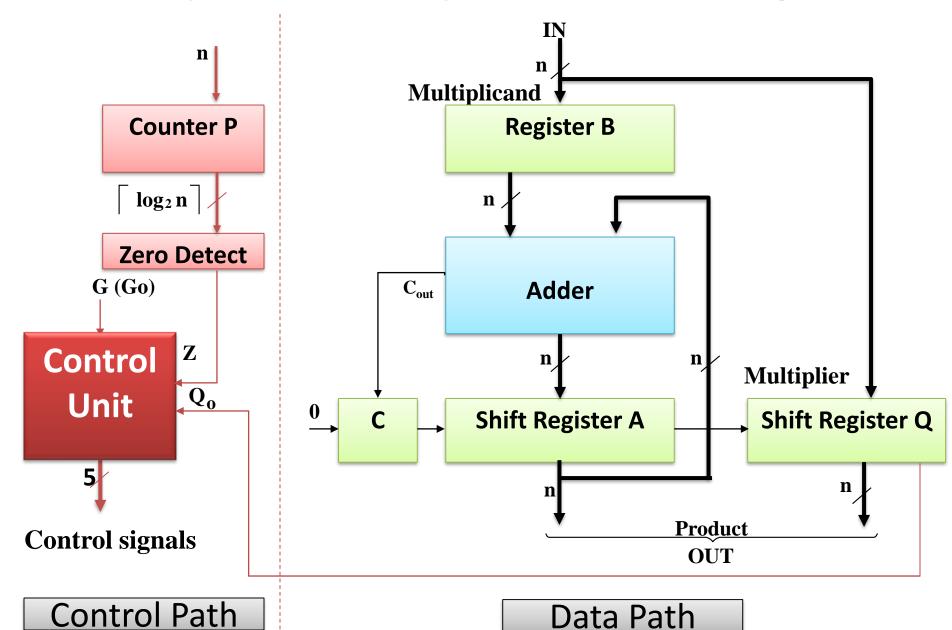
# Sequential Multiplier



### **Control Algorithm:**

- P ← 0, A ← multiplicand, B ← multiplier //Initialization
- 2. If LSB of B==1 then add A to P else add 0
- 3. Shift [P][B] right 1
- 4. Repeat steps 2 and 3 n-1 times.
- 5. [P][B] has product.

# Multiplier Example: Block Diagram



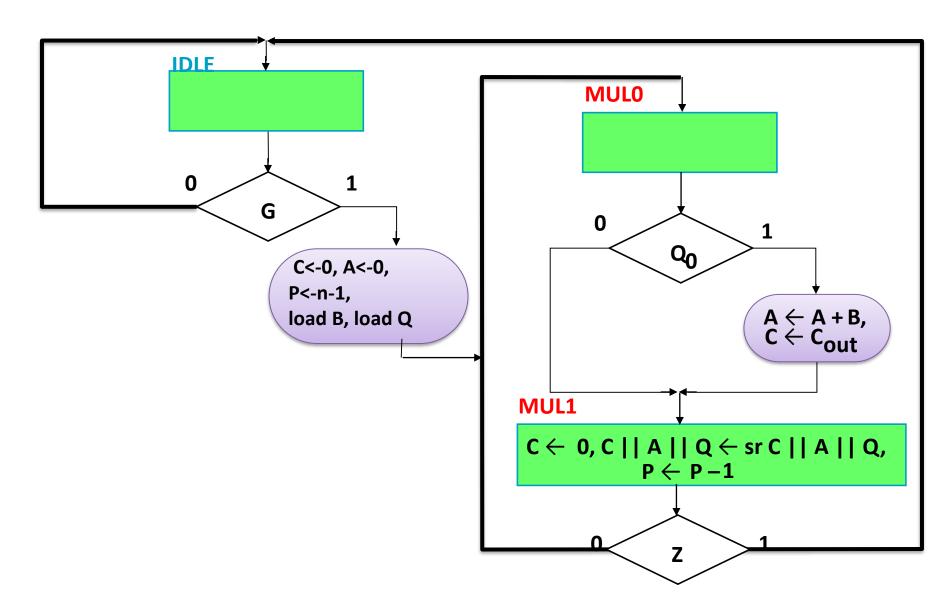
### **Multiplier Example: Operation**

- Step1: The multiplicand (top operand) is loaded into register B.
- Step2: The multiplier (bottom operand) is loaded into register Q.
- Step3: Register C||A is initialized to 0 when G becomes 1.
- Step4: The partial products are summed iteratively in register C||A||Q.

### **Multiplier Example: Operation**

- Step5: Each multiplier bit, beginning with the LSB, is processed (if bit is 1, use adder to add B to partial product; if bit is 0, do nothing)
- Step6: C||A||Q is shifted right using the shift register
  - Partial product bits fill vacant locations in Q as multiplier is shifted out
  - If overflow during addition, the outgoing carry is recovered from C during the right shift
- Step 7: Steps 5 and 6 are repeated until Counter P = 0 as detected by Zero detect.
  - Counter P is initialized in step 4 to n 1, n = number of bits in multiplier

# Multiplier Example: ASM Chart



### Multiplier Example: ASM Chart (continued)

Combined Mealy - Moore output model

### IDLE - state

- Input G is used as the condition for starting the multiplication, and
- C, A, and P are initialized and Load B, Load Q

### MULO - state

– Conditional addition is performed based on the value of  $Q_0$ .

### MUL1 - state

- Right shift is performed to capture the partial product and position the next bit of the multiplier in  $Q_0$
- the terminal count of 0 for down counter P is used to sense completion or continuation of the multiply.

# **Multiplier Example: Control Signal Table**

Module	Micro Operations	Control Signal Name	Control Expression
Reg A	A <- 0 A <- A +B C   A  Q <- sr (C   A  Q)	Load_regs Add_regs Shift CAQ	IDLE.G MULO.Q <sub>0</sub> MUL1
Reg B	B <- IN	Load_regs	LOAD_B
FF C	C <- 0 C <- Cout	Clear C Load	IDLE.G+MUL1 
Reg Q	Q <- IN C   A  Q <- sr (C   A  Q )	Load_regs ShiftCAQ	LOAD_Q
Ctr P	P <- N-1 P <- P-1	Load_regs Decr_P	MUL1

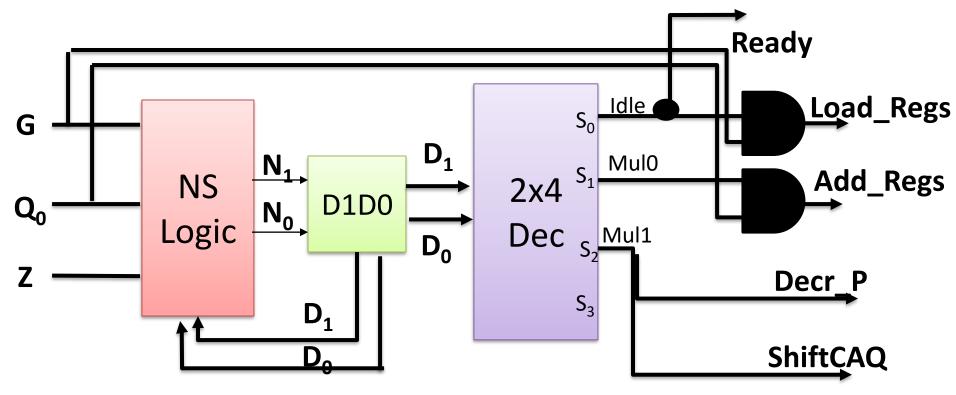
### **Multiplier Example: Control Circuit**

PS	PS	Inputs			NS	Ready		Decr_ P	Add_r egs	Shift_ CAQ
		G	Q0	Z						
Idle	00	0	x	X	00	1				
Idle	00	1	X	x	01	1	1			
Mul0	01	x	0	X	10			1		
Mul0	01	x	1	X	10				1	
Mul1	10	x	x	0	01			1		1
Mul1	10	X	x	1	00			1		1

Ready=Idle Load\_Regs=Idle.G Add\_regs=Mul0.Q<sub>0</sub>

Decr\_P=Mul0
ShiftCAQ=Mul1

### **Multiplier Example: Control Circuit**



$$N1=D_1'D_0$$
  
 $N0=D_1'D_0'G+D_1D_0'Z'$ 

Ready=Idle Load\_Regs=Idle.G Decr\_P=Mul1
Add\_regs=Mul0.Q<sub>0</sub> ShiftCAQ=Mul1