

# ECON 616: Lecture Three: The Spectrum

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# Background

- ▶ Overview: Chapters 6 from Hamilton (1994).
- ▶ Technical Details: Chapter 4 from Brockwell and Davis (1987).
- ▶ Other stuff: You might want to look at a digital signals processing textbook, for example: [here](#).

# Cycles as Frequencies

Starting In the 19th Century, economists and others recognized cyclical patterns in economic activity.

Schmupeter distinguished between cycles at different frequencies

- ▶ Kondratieff Cycles – Longwave cycles lasting 50 years (caused by fundamental innovations.)
- ▶ Juglar Cycles – medium cycle (8 years) associated with changes in credit condition.
- ▶ Kitchin Cycles – short run cycles (40 months) associated with information diffusion.

=> model economic activity as a linear combination of periodic function with different frequencies.

## A model of frequencies

Consider the following model for quarterly observations

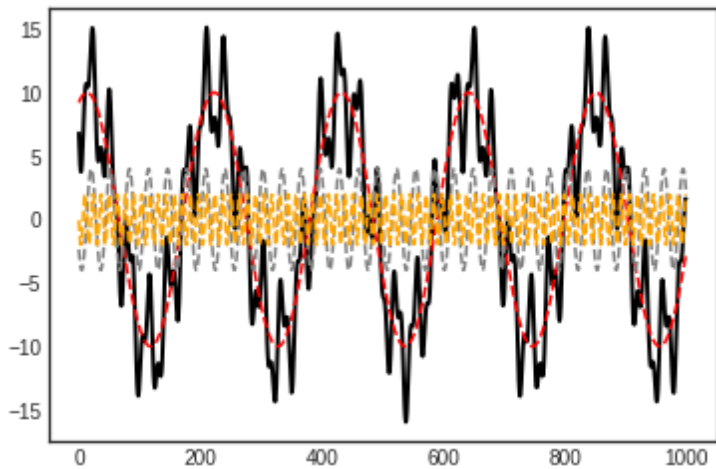
$$X_t = 2 \sum_{j=1}^m a_j \cos(\omega_j t + \theta_j)$$

where  $\theta_j$  is iidU $[-\pi, \pi]$  and  $0 \leq \omega_j < \omega_{j+1} \leq \pi$ . The random variables  $\theta_j$  are determined in the infinite past and simply cause a phase shift. According to Schumpeter's hypothesis  $m$  should be equal to three. The frequencies  $\omega_j$  can be determined as follows.

Cycle	Duration	Frequency
Kondratieff	200 quarters	$\omega_1 = (2\pi)/200 = 0.03$
Juglar	32 quarters	$\omega_2 = (2\pi)/32 = 0.20$
Kitchin	13.3 quarters	$\omega_3 = (2\pi)/13.3 = 0.47$

## A Time Series of this process

$$A = [5, 2, 1], \theta = [0.03, 0.20, 0.47].$$



# The Spectrum

- ▶ The coefficients  $a_1$  to  $a_3$  are the amplitudes of the different cycles
- ▶ If  $a_1$  and  $a_2$  are small then most of the variation in  $X_t$  is due to the Kitchin cycles.
- ▶ The plot of  $a_j^2$  versus  $\omega$  is called the spectrum of  $X_t$ .

## Some math

$$\cos(x + y) = \cos x \cos y - \sin x \sin y \quad (1)$$

$$\sin x \sin y = \frac{1}{2}[\cos(x - y) - \cos(x + y)] \quad (2)$$

$$\cos x \cos y = \frac{1}{2}[\cos(x - y) + \cos(x + y)] \quad (3)$$

$$2 \sin^2 x = 1 - \cos(2x) \quad (4)$$

$$\sin x \cos x = \frac{1}{2} \sin(2x) \quad (5)$$

Moreover,  $\sin^2 x + \cos^2 x = 1$ .

We consider real-valued stochastic processes  $X_t$ , complex numbers will help us summarize sine and cosine expressions using exponential functions.

## More Math

Let  $i = \sqrt{-1}$ .

Euler's formula:

$$e^{i\varphi} = \cos \varphi + i \sin \varphi$$

The formula becomes less mysterious if you rewrite  $e^{i\varphi}$ ,  $\sin \varphi$ , and  $\cos \varphi$  as power series.



# The Plan

- ▶ Rewrite Schumpeter Model
- ▶ Define spectral distribution / density function
- ▶ Examine relationship between autocovariances  $\{\gamma_h\}_{h=-\infty}^{\infty}$  and the spectrum.
- ▶ Discuss very general spectral representation for a stationary stochastic process  $X_t$ .

# Schumpeter Model

$$X_t = 2 \sum_{j=1}^m a_j \cos \theta_j \cos(\omega_j t) - a_j \sin \theta_j \sin(\omega_j t)$$

where  $a_j \cos \theta_j$  and  $a_j \sin \theta_j$  can be regarded as random coefficients.

Eulers formula implies

$$X_t = \sum_{j=-m}^m A(\omega_j) e^{i\omega_j t}$$

where  $\omega_{-j} = -\omega_j$ . Let  $a_{-j} = a_j$  and

This means that

$$A(\omega_j) = \begin{cases} a_j(\cos \theta_{|j|} + i \sin \theta_{|j|}) & \text{if } j > 0 \\ a_j(\cos \theta_{|j|} - i \sin \theta_{|j|}) & \text{if } j < 0 \end{cases}$$

We can verify that:

$$A(\omega_j)e^{i\omega_j t} + A(\omega_{-j})e^{-i\omega_j t} = 2 [a_j \cos \theta_j \cos(\omega_j t) - a_j \sin \theta_j \sin(\omega_j t)]$$

# Moments of Linear Cyclical Models

$$\mathbb{E}[\cos \theta_j] = \frac{1}{2\pi} \int_{-\pi}^{\pi} \cos \theta_j d\theta_j = 0 \quad (6)$$

$$\mathbb{E}[\sin \theta_j] = \frac{1}{2\pi} \int_{-\pi}^{\pi} \sin \theta_j d\theta_j = 0 \quad (7)$$

**Result:** The expectation of  $X_t$  in the linear cyclical model is equal to zero.  $\square$

## Autocovariances

To obtain the autocovariances  $\gamma_h = \mathbb{E}[X_t X_{t-h}]$  we have to calculate the moments  $\mathbb{E}[A(\omega_j)A(\omega_k)]$ .

Let  $j \neq k, j \neq -k$ . Suppose that  $j, k > 0$ .

$$\begin{aligned}\mathbb{E}[A(\omega_j)A(\omega_k)] &= a_j a_k \mathbb{E}[(\cos \theta_j + i \sin \theta_j)(\cos \theta_k + i \sin \theta_k)] \\ &= a_j a_k \mathbb{E}[\cos \theta_j \cos \theta_k + i \cos \theta_j \sin \theta_k i \cos \theta_k \sin \theta_j - \sin \theta_j \sin \theta_k] \\ &= 0\end{aligned}$$

Since  $\phi_j$  and  $\phi_k$  are independent. Similar arguments can be made if  $j$  and  $k$  have different signs.

## Covariance

Let  $j = k$ . Suppose that  $j, k > 0$ .

$$\begin{aligned}\mathbb{E}[A(\omega_j)A(\omega_k)] &= a_j^2 \mathbb{E}[(\cos \theta_j + i \sin \theta_j)^2] \\ &= a_j^2 \mathbb{E}[(\cos^2 \theta_j - \sin^2 \theta_j + i2 \cos \theta_j \sin \theta_j)] \\ &= a_j^2 \mathbb{E}[1 - 2 \sin^2 \theta_j + i2 \cos \theta_j \sin \theta_j] \\ &= a_j^2 \mathbb{E}[\cos(2\theta_j) + i \sin(2\theta_j)] \\ &= 0\end{aligned}\tag{9}$$

In the last step we use the fact that sine and cosine integrate to zero over two cycles. A similar argument can be made for the case  $j, k < 0$

Let  $j = -k$ . Now  $A(\omega_j)$  and  $A(\omega_k)$  are complex conjugates.

$$\mathbb{E}[A(\omega_j)A(\omega_{-j})] = a_j^2 \mathbb{E}[\cos^2 \theta_j + \sin^2 \theta_j] = a_j^2$$

## The upshot

**Result:** The autocovariances of the process  $X_t$  generated by the linear cyclical model are given by

$$\begin{aligned}\gamma_h &= \mathbb{E}[X_t X_{t-h}] \\ &= \sum_{j=-m}^m \sum_{k=-m}^m \mathbb{E}[A(\omega_j) A(\omega_k)] e^{i\omega_j t} e^{i\omega_k (t-h)} \\ &= \sum_{j=-m}^m \mathbb{E}[A(\omega_j) \overline{A(\omega_j)}] e^{i\omega_j h} = \sum_{j=-m}^m a_j^2 e^{i\omega_j h} \quad (10)\end{aligned}$$

Since  $X_t$  is a real valued process the autocovariances can also be written as

$$\gamma_h = 2 \sum_{j=1}^m a_j^2 \cos(\omega_j h) \quad \square$$

# The Spectral Distribution

The spectral distribution function for the process  $X_t$ , defined on the interval  $\omega \in (-\pi, \pi)$ , is

$$S(\omega) = \sum_{j=-m}^m \mathbb{E}[A(\omega_j) \overline{A(\omega_j)}] \{\omega_j \leq \omega\}$$

where  $\{\omega_j \leq \omega\}$  denotes the indicator function that is one if  $\omega_j \leq \omega$ .  $\square$



## Remarks

The spectral distribution is non-negative and continuous from the right.

If the spectral distribution function is evaluated at  $\omega = \pi$  we obtain

$$S(\pi) = \sum_{j=-m}^m \mathbb{E}[A(\omega_j) \overline{A(\omega_j)}] = \sum_{j=-m}^m a_j^2 = \mathbb{E}[X_t^2] \quad (11)$$

The spectral distribution function is symmetric in the sense that for  $\omega > 0$

$$S(-\omega) = S(\pi) - \lim_{n \rightarrow \infty} S((\omega - 1/n)) \quad (12)$$

## Autocovariances, again

The representation of the autocovariances can be expressed as a Riemann-Stieltjes integral. Define a sequence of grids

$$[\omega]^{(n)} = \{\omega_k^{(n)} = 2\pi k/n - \pi\}$$

and  $\Delta_n \omega = \omega_{k+1}^{(n)} - \omega_k^{(n)} = 2\pi/n$ . Moreover, let

$$\Delta_n S(\omega) = S(\omega) - S(\omega - \Delta_n \omega)$$

Roughly,

$$\sum_{k=0}^n e^{i\omega_k^{(n)}h} \Delta_n S(\omega_k^{(n)}) \longrightarrow \sum_{j=-m}^m a_j^2 e^{i\omega_j h}$$

as  $n \rightarrow \infty$ .

## The Upshot

Thus, we can express the autocovariance  $\gamma_h$  as the following integral

$$\gamma_h = \int_{(-\pi, \pi]} e^{i\omega h} dS(\omega)$$

By using a similar argument, we can also obtain an integral representation for the stochastic process  $X_t$ . Define the stochastic process

$$Z(\omega) = \sum_{j=-m}^m A(\omega_j) \{\omega_j \leq \omega\}$$

with orthogonal increments  $\Delta_n Z(\omega) = Z(\omega) - Z(\omega - \Delta_n \omega)$ . Note that the increments are now random variables.

Very roughly,

$$\sum_{k=0}^n e^{i\omega_k^{(n)}t} \Delta_n Z(\omega_k^{(n)}) \longrightarrow \sum_{j=-m}^m A(\omega_j) e^{i\omega_j t}$$

almost surely as  $n \rightarrow \infty$ . Thus, we can express the stochastic process  $X_t$ , generated from the linear cyclical model, as the stochastic integral

$$X_t = \int_{(-\pi, \pi]} e^{i\omega t} dZ(\omega)$$

# Spectral Representation for Stationary Processes

Every zero-mean stationary process has a representation of the form

$$X_t = \int_{(-\pi, \pi]} e^{i\omega h} dZ(\omega)$$

where  $Z(\omega)$  is a orthogonal increment process. Correspondingly, its autocovariance function  $\gamma_h$  can be expressed as

$$\gamma_h = \int_{(-\pi, \pi]} e^{i\omega h} dS(\omega)$$

where  $S(\omega)$  is a non-decreasing right continuous function with  $S(\pi) = \mathbb{E}[X_t^2] = \gamma_0$ .

# Spectral Density Function

Suppose the spectral distribution function is differentiable with respect to  $\omega$  on the interval  $(-\pi, \pi]$ . The spectral density function is defined as

$$s(\omega) = dS(\omega)/d\omega$$

If a process has a spectral density function  $s(\omega)$  then the covariances can be expressed as

$$\gamma_h = \int_{(-\pi, \pi]} e^{ih\omega} s(\omega) d\omega$$

The spectral density uniquely determines the entire sequence of autocovariances. Moreover, the converse is also true.

Consider the sum

$$\begin{aligned} s_n(\omega)^* &= \frac{1}{2\pi} \sum_{h=-n}^n \gamma_h e^{-i\omega h} \\ &= \frac{1}{2\pi} \sum_{h=-n}^n \left[ \int_{(-\pi, \pi]} e^{i\tau h} s(\tau) d\tau \right] e^{-i\omega h} \end{aligned} \quad (13)$$

The sum  $s_n^*(\omega)$  is a Fourier series. If the spectral density  $s(\omega)$  is piecewise smooth then

$$s_n^*(\omega) \longrightarrow s(\omega)$$

Thus, the **spectral density** can be obtained by evaluating the **autocovariance generating function** of  $X_t$  at  $z = e^{-i\omega}$ .

$$s(\omega) = \frac{1}{2\pi} \gamma(e^{-i\omega}) = \frac{1}{2\pi} \sum_{h=-\infty}^{\infty} \gamma_h e^{-i\omega h}$$

where

$$\gamma(z) = \sum_{j=-\infty}^{\infty} \gamma_j z^j$$

# Filter

Suppose  $s_X(\omega)$  is the spectral density function of a process  $X_t$ . Filters are used to dampen or amplify the spectral density at certain frequencies. The spectrum of the filtered series  $Y_t$  is given by

$$s_Y(\omega) = f(\omega)s_X(\omega).$$

where  $f(\omega)$  is the filter function.

Frequency domain trend/cycle analogue

$X_t = \text{low frequency component} + \text{high frequency component}$

Example: For Schumpeter, Kitchin cycle was shortest with  $\omega = 0.47$ . To remove the effects of other cycles from data, we could use the filter

$$f(\omega) = \begin{cases} 0 & \text{if } \omega < 0.4 \\ 1 & \text{otherwise} \end{cases}$$



# Hodrick Prescott filter

A popular filter in the real business cycle literature in macro-economics is the so-called Hodrick Prescott filter.

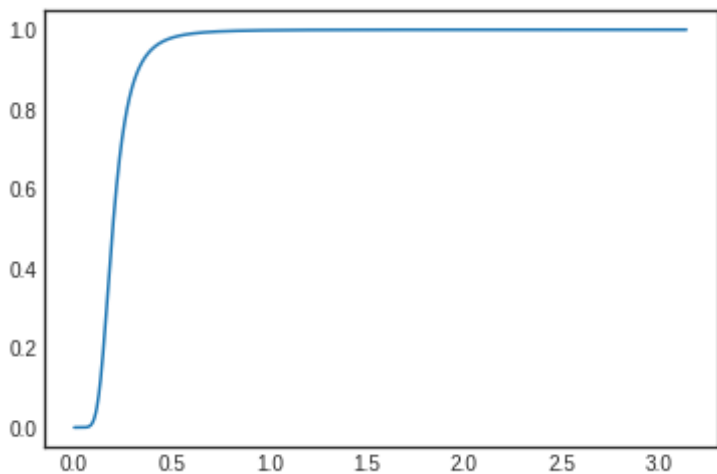
$$f^{\text{HP}}(\omega) = \left[ \frac{16 \sin^4(\omega/2)}{1/1600 + 16 \sin^4(\omega/2)} \right]^2.$$

This filter basically kills long cycles and attenuates medium term ones.

(See Soderlind, 1994.)

## HP Filter

$$f(2\pi/64) = 0.016697846612617945 \quad f(2\pi/32) = 0.4937014515$$



## More on Filters

Subsequently we will consider filters that are linear in the time domain, namely, filters of the form,

$$Y_t = \sum_{h=1}^J c_h X_{t-h} = C(L)X_t$$

where  $C(z)$  is the polynomial function  $\sum_{h=1}^J c_h z^h$ . Recall that

$$X_t = \sum_{j=-m}^m A(\omega_j) e^{i\omega_j t}$$

This means that

Hence,

$$\begin{aligned} X_{t-h} &= \sum_{j=-m}^m A(\omega_j) e^{i\omega_j t} e^{-i\omega_j h} \\ Y_t = C(L)X_t &= \sum_{h=1}^J c_h X_{t-h} \\ &= \sum_{j=-m}^m \left[ A(\omega_j) e^{i\omega_j t} \sum_{h=1}^J c_h e^{-i\omega_j h} \right] \\ &= \sum_{j=-m}^m A(\omega_j) C(e^{-i\omega_j}) e^{i\omega_j t} \\ &= \sum_{j=-m}^m \tilde{A}(\omega_j) e^{i\omega_j t} \end{aligned} \tag{14}$$

## Autocovariance

The autocovariances of  $Y_t$  can therefore be expressed as

$$\mathbb{E}[Y_t Y_{t-h}] = \sum_{j=-m}^m a_j^2 C(e^{-i\omega_j}) C(e^{i\omega_j}) e^{i\omega_j h}$$

Thus, we can define the spectral distribution function of  $Y_t$  as

$$S_Y(\omega) = \sum_{j=-m}^m a_j^2 C(e^{-i\omega_j}) C(e^{i\omega_j})$$

with increments

$$\Delta S_Y(\omega_j) = \Delta S_X C(e^{-i\omega_j}) C(e^{i\omega_j})$$

## Generalization

**Result:** Suppose that  $X_t$  has a spectral density function  $s_X(\omega)$  and  $Y_t = C(L)X_t$ , then the spectral density of the filtered process  $Y_t$  is given by

$$s_Y(\omega) = |C(e^{-i\omega})|^2 s_X(\omega)$$

The function  $C(e^{-i\omega})$  is called transfer function of the filter, and the filter function  $f(\omega) = |C(e^{-i\omega})|^2$  is often called power transfer function.  $\square$

# Examples of Spectrum

## White Noise

$$s(\omega) = \frac{1}{2\pi} \sum_{h=-\infty}^{\infty} \gamma_h e^{-i\omega h} = \frac{\gamma_0}{2\pi}$$

**An AR(1):**  $Y_t = \phi Y_{t-1} + X_t$

Interpret as a linear filter with MA( $\infty$ ) rep:

$Y_t = \sum_{h=0}^{\infty} \phi^h X_{t-h}$ . Thus:

$$\begin{aligned} |C(e^{-i\omega})|^2 &= |[1 - \phi e^{-i\omega}]^{-1}|^2 \\ &= [|1 - \phi \cos \omega + i\phi \sin \omega|^2]^{-1} \\ &= [(1 - \phi \cos \omega)^2 + \phi^2 \sin^2 \omega]^{-1} \\ &= [1 - 2\phi \cos \omega + \phi^2(\cos^2 \omega + \sin^2 \omega)]^{-1}. \quad (15) \end{aligned}$$

which means  $s_Y(\omega) = \frac{\sigma^2/2\pi}{1+\phi^2-2\phi\cos\omega}$ . Note  
 $s_Y(0) \longrightarrow \infty$  as  $\phi \longrightarrow 1$

## More Examples

**Stationary ARMA process:**  $\phi(L)Y_t = \theta(L)X_t$  with  $X_t \sim \text{WN}$ .  
The spectral density is given by

$$s_Y(\omega) = \left| \frac{\theta(e^{-i\omega})}{\phi(e^{-i\omega})} \right|^2 \sigma^2$$

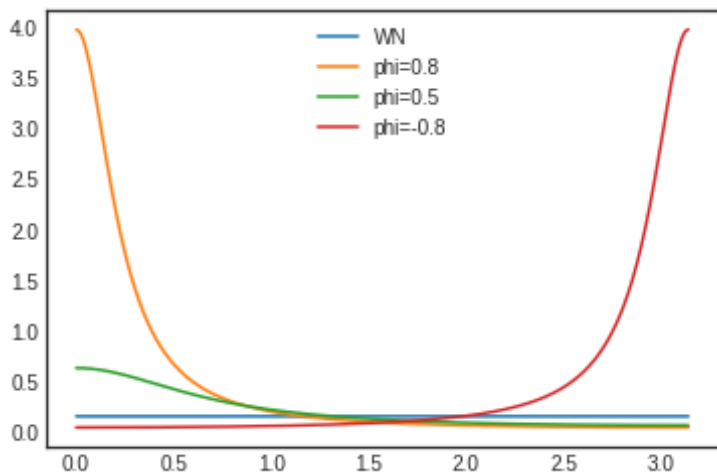
**Sums of processes.** Suppose that  $W_t = Y_t + X_t$ . The spectrum of the process  $W_t$  is simply the sum

$$s_W(\omega) = s_Y(\omega) + s_X(\omega)$$



## Visual

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# Estimation

1. **Parametric** – Pick an ARMA process, estimate in time domain, use filtering results to get spectrum.
2. **Nonparametric** – Estimate autocovariances  $\{\hat{\gamma}_h\}$ , directly write down spectral density. Let's look at this.

Let  $\bar{y} = \frac{1}{T} \sum y_t$  and define the sample covariances

$$\hat{\gamma}_h = \frac{1}{T} \sum_{t=h+1}^T (y_t - \bar{y})(y_{t-h} - \bar{y})$$

An intuitively plausible estimate of the spectrum is the sample periodogram

$$\begin{aligned} I_T(\omega) &= \frac{1}{2\pi} \sum_{j=-T+1}^{T-1} \hat{\gamma}_h e^{-i\omega h} \\ &= \frac{1}{2\pi} \left( \hat{\gamma}_0 + 2 \sum_{h=1}^{T-1} \hat{\gamma}_h \cos(\omega h) \right) \end{aligned} \quad (16)$$

**Result:** The sample periodogram is an asymptotically unbiased estimator of the population spectrum, that is,

$$[I_T(\omega)] \xrightarrow{P} s(\omega) \quad (17)$$

However, it is inconsistent since the variance  $\text{var}[I_T(\omega)]$  does not converge to zero as the sample size tends to infinity.  $\square$

## Smoothed Periodogram

Smoothing: get non-parametric estimators.

To obtain a spectral density estimate at the frequency  $\omega = \omega_*$  we will compute the sample periodogram  $I_T(\omega)$  for some  $\omega_j$ 's in the neighborhood of  $\omega_*$  and simply average them. Define the following band around  $\omega_*$ :

$$B(\omega_*|\lambda) = \left\{ \omega : \omega_* - \frac{\lambda}{2} < \omega \leq \omega_* + \frac{\lambda}{2} \right\} \quad (18)$$

The bandwidth is  $\lambda$ , where  $\lambda$  is a parameter. Moreover, define the “fundamental frequencies” (see Hamilton 1994, Chapter 6.2, for a discussion why these frequencies are “fundamental”)

$$\omega_j = j \frac{2\pi}{T} \quad j = 1, \dots, (T-1)/2 \quad (19)$$

The number of fundamental frequencies in the band  $B(\omega_*)$  is

$$m = \lfloor \lambda T (2\pi)^{-1} \rfloor \quad (20)$$

# Smoothed Periodogram

The smoothed periodogram estimator of  $s(\omega_*)$  is defined as the average

$$\hat{s}(\omega) = \sum_{j=1}^{(T-1)/2} \frac{1}{m} \{\omega_j \in B(\omega_*|\lambda)\} I_T(\omega_j) \quad (21)$$

where  $\{\omega_j \in B(\omega_*|\lambda)\}$  is the indicator function that is equal to one if  $\omega_j \in B(\omega_*|\lambda)$  and zero otherwise.

**Result:** The smoothed periodogram estimator  $\hat{s}(\omega_*)$  of  $s(\omega_*)$  is consistent, provided that the bandwidth shrinks to zero, that is,  $\lambda \rightarrow 0$  as  $T \rightarrow \infty$  and the number of  $\omega_j$ 's in the band  $B(\omega_*|\lambda)$  tends to infinity, that is  $m = \lambda T/(2\pi) \rightarrow \infty$ .  $\square$

## Remarks

- ▶ get smoothed estimates  $\Rightarrow$  need to get  $\lambda$ . Ultimately subjective.
- ▶ Most non-parameterics approaches are based on “Kernel estimates”

The expression  $\{\omega_j \in B(\omega_*)\}$  can be rewritten as follows

$$\begin{aligned}\{\omega_j \in B(\omega_*)\} &= \left\{ \omega_* - \frac{\lambda}{2} < \omega_j \leq \omega_* + \frac{\lambda}{2} \right\} \\ &= \left\{ -\frac{1}{2} < \frac{\omega_j - \omega_*}{\lambda} \leq \frac{1}{2} \right\}\end{aligned}\tag{22}$$

Define

$$K\left(\frac{\omega_j - \omega_*}{\lambda}\right) = \left\{ -\frac{1}{2} < \frac{\omega_j - \omega_*}{\lambda} \leq \frac{1}{2} \right\}\tag{23}$$

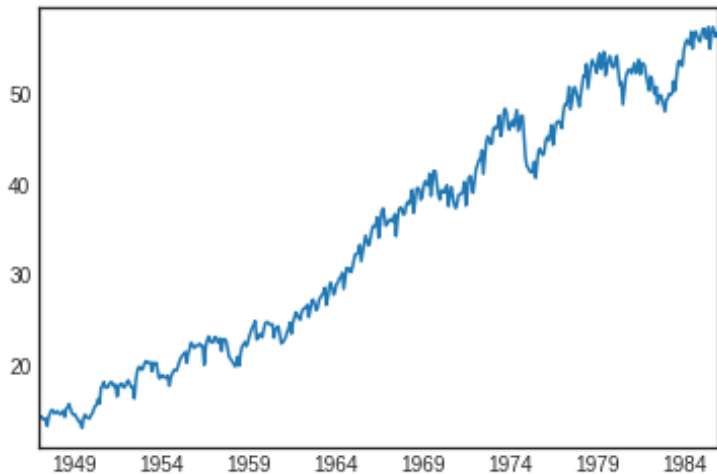
It can be easily verified that

$$\int K\left(\frac{\omega_j - \omega_*}{\lambda}\right) d\omega_* = 1 \quad (24)$$

The function  $K\left(\frac{\omega_j - \omega_*}{\lambda}\right)$  is an example of a Kernel function. In general, a Kernel has the property  $\int K(x)dx = 1$ . Since  $m \approx \lambda(T-1)/2$ , the spectral estimator can be rewritten as

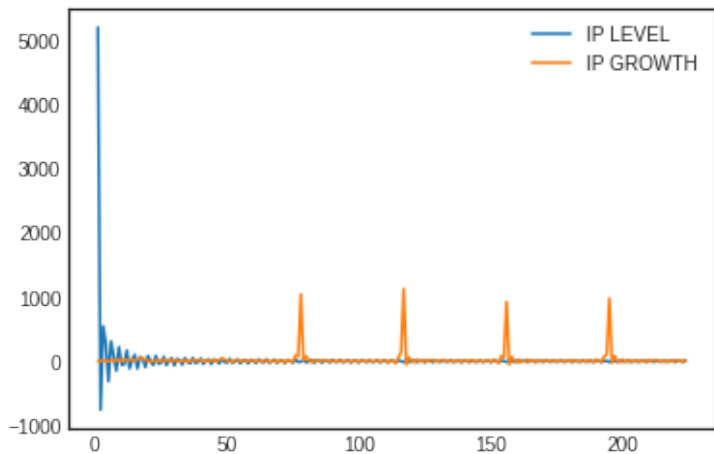
$$\hat{s}(\omega) = \frac{\pi}{\lambda(T-1)/2} \sum_{j=1}^{(T-1)/2} K\left(\frac{\omega_j - \omega_*}{\lambda}\right) I_T(\omega_j) \quad (25)$$

## Application: IP





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## Application: Autocorrelation Consistent Standard Errors

Consider the model

$$y_t = \beta x_t + u_t, \quad u_t = \psi(L)\epsilon_t, \quad \epsilon_t \sim \text{iid}(0, \sigma^2) \quad (26)$$

The OLS estimator is given by

$$\hat{\beta} - \beta = \frac{\sum x_t u_t}{\sum x_t^2} \quad (27)$$

The conventional standard error estimates for  $\hat{\beta}$  are inconsistent if the  $u_t$ 's are serially correlated. However, we can construct a consistent estimate based on non-parametric spectral density estimation. Define  $z_t = x_t u_t$ . We want to obtain an estimate of

$$\text{plim } \Lambda_T = \text{plim } \frac{1}{T} \sum_{t=1}^T \sum_{h=1}^T E[z_t z_h] \quad (28)$$

It can be verified that

$$\sum_{h=-\infty}^{\infty} \gamma_{zz,h} - \frac{1}{T} \sum_{t=1}^T \sum_{h=1}^T E[z_t z_h] \xrightarrow{p} 0 \quad (29)$$

Since

$$s(\omega) = \frac{1}{2\pi} \sum_{h=-\infty}^{\infty} \gamma_{zz,h} e^{-i\omega h} \quad (30)$$

it follows that a consistent estimator of  $\text{plim } \Lambda_T$  is

$$\hat{\Lambda}_T = 2\pi \hat{s}(0) \quad (31)$$

where  $\hat{s}(0)$  is a non-parametric spectral estimate at frequency zero.

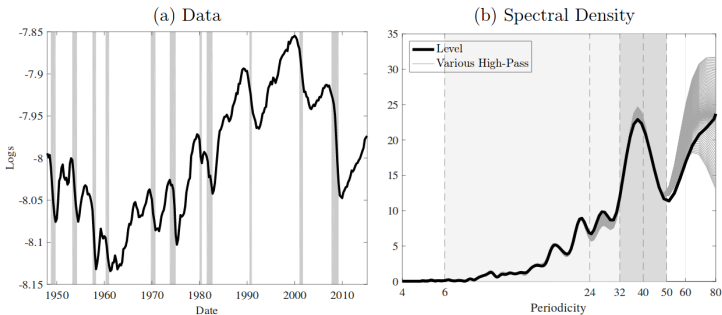
## Application: Beaudry et al. (2020)

Paul Beaudry, Dana Galiza, and Franck Portier (2016):  
“Putting the Cycle Back into Business Cycle Analysis,” NBER  
Working Paper.

- ▶ Re-examines the spectral properties of several cyclically sensitive variables such as hours worked, unemployment and capacity utilization.
- ▶ Document the presence of an important peak in the spectral density at a periodicity of approximately 36-40 quarters.
- ▶ This is cyclical phenomena at the “long end” of the business cycle.
- ▶ Suggests a model (“limit cycles”) to account for this finding.

# The Paper in 1 Picture

Figure 1: Properties of Hours Worked per Capita



# References

- Beaudry, P., Galizia, D., and Portier, F. (2020). Putting the cycle back into business cycle analysis. *American Economic Review*, 110:1–47.
- Brockwell, P. J. and Davis, R. A. (1987). *Time series: Theory and methods*. Springer Series in Statistics.
- Hamilton, J. (1994). *Time Series Analysis*. Princeton University Press, Princeton, New Jersey.