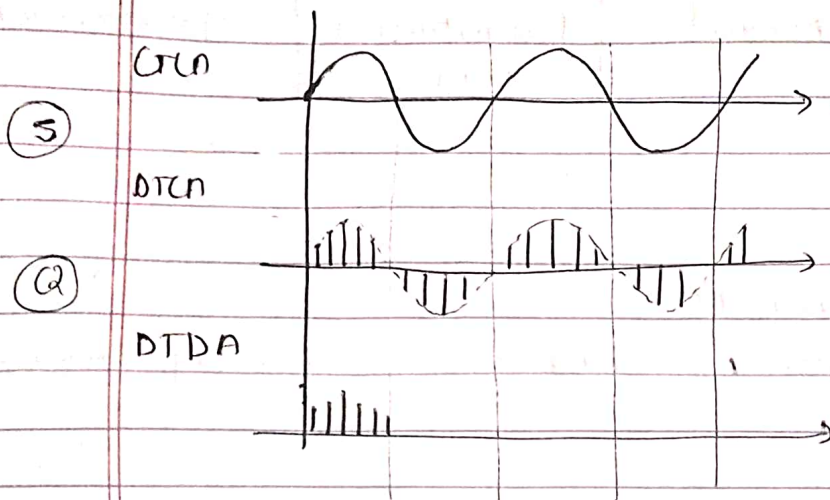
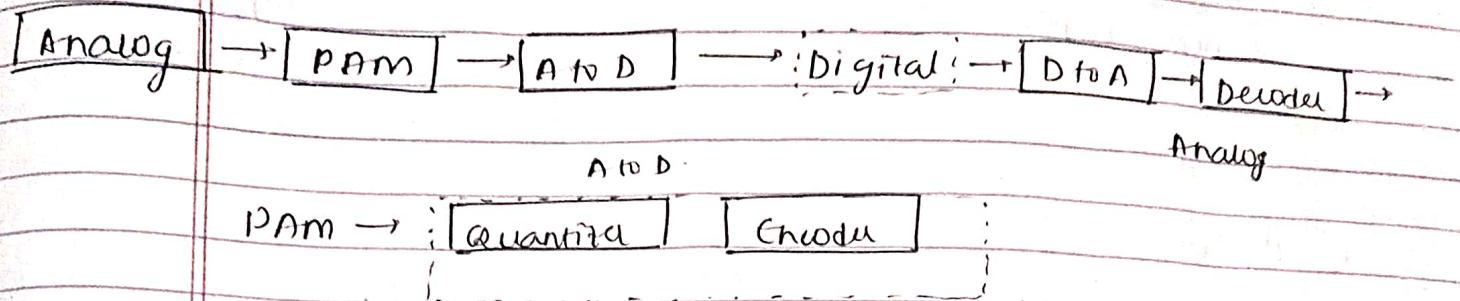
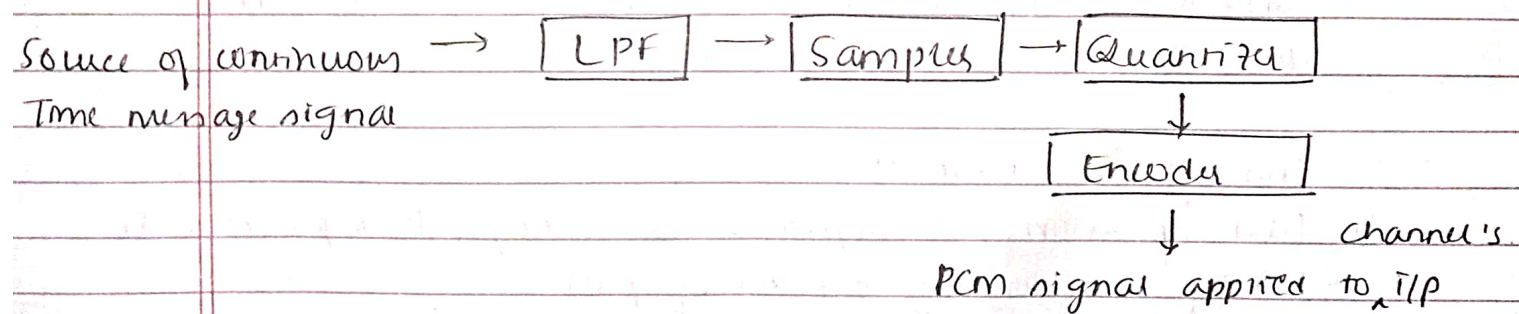


* PCM:

Discrete Time, discrete amplitude waveform using process by means of which analog signals can be directly represented as sequence of coded pulses.



* Transmitter:



1) Basic operations performed :

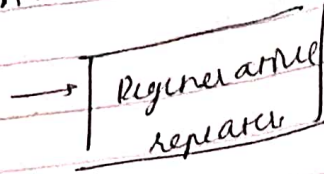
- Sampling
- Quantizing.
- Encoding.

2) LPF : to prevent message signals aliasing.

3) A to D converter : Quantize & Encoder.

Transmission path:

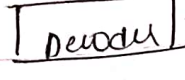
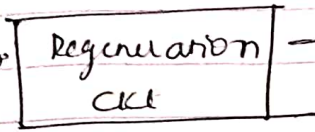
distorted PCM
signal produced at
channel output



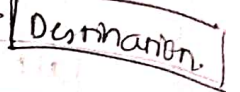
Regenerated
PCM signal
applied to the
receiver.

Receiver:

Final
channel
O/p



PAM



Sampling in Transmitter:

- 1) Incoming message signal is sampled with train of rectangular pulses

Quantization in Transmitter:

- 1) The PAM representation of m(t) is then quantized in A/D converter, providing new representation of signal that is discrete in both time & amplitude.

Encoding in Transmitter:

- 1) Last operation on signal is the coding, to represent each binary codeword by sequence of pulses.

Ex: symbol 1 : used to represent presence of pulse.

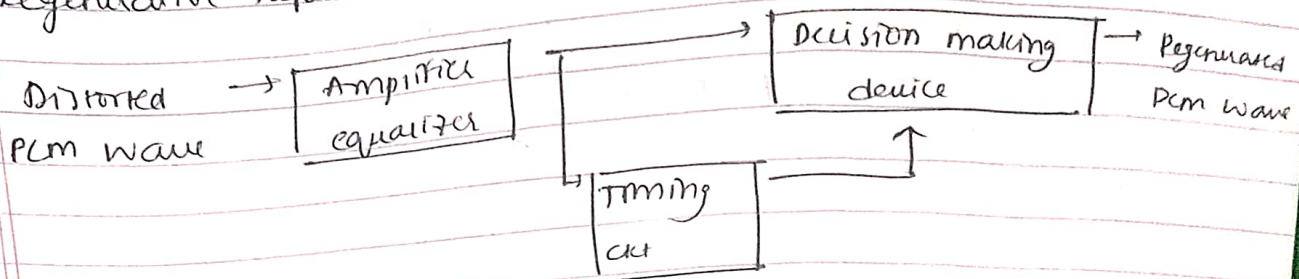
symbol 0 : " " " absence " "

Receiver's opⁿ:

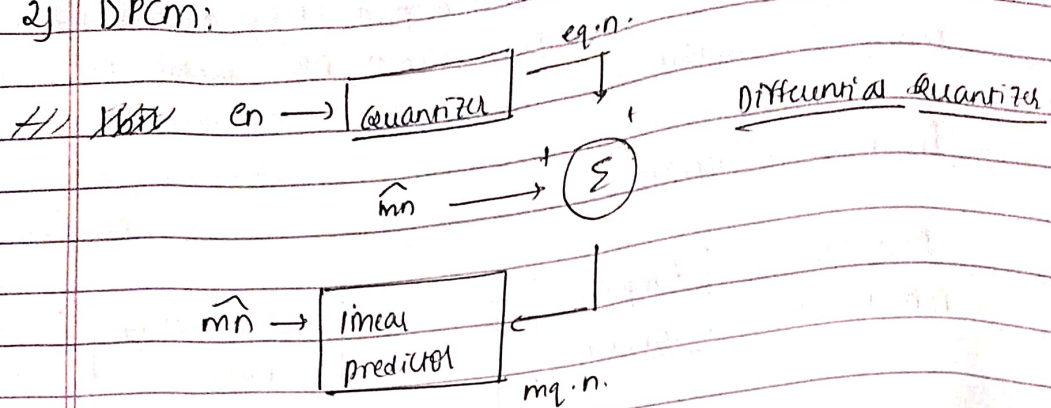
- 1) 1st operation is to regenerate the received pulses.
- 2) These clean pulses are then regrouped into codewords & decoded into quantized PAM signals.
- 3) signal reconstruction →

An estimate of original signal is produced by passing distorted output through a low pass filter whose $f_c =$ message bandwidth.

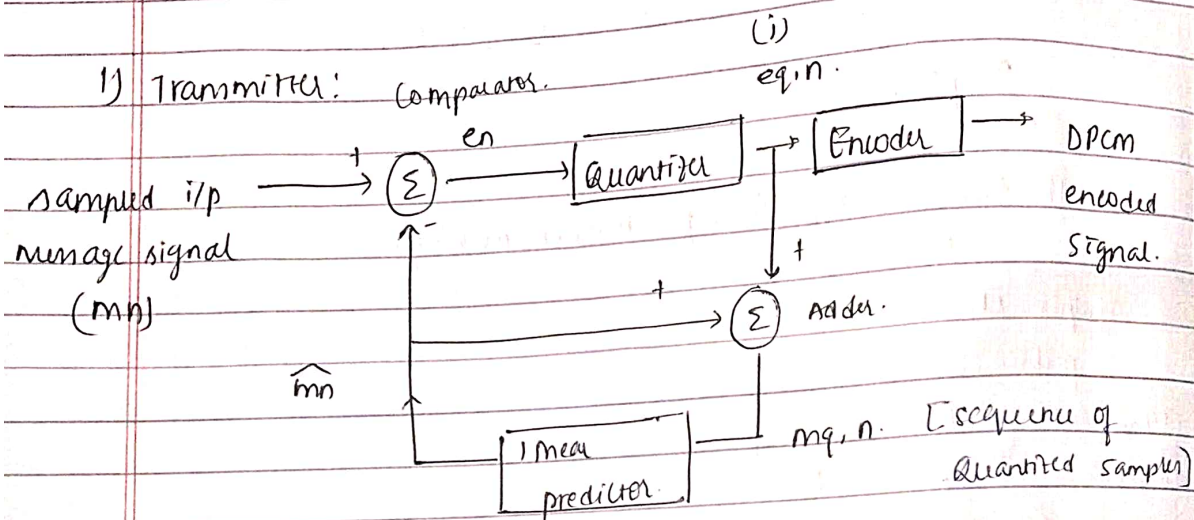
Regenerative repeater:



2) DPCM:



1) Transmitter:



$$eq.n = en + qn \rightarrow \text{quantization noise}$$

$$mq.n = eq.n + \widehat{mn} \rightarrow \text{predicted value of original message signal}$$

$$\therefore mq.n = eq.n + en + qn$$

$$en = mn - \widehat{mn}$$

$$mq.n = eq.n + mn - \widehat{mn} + qn$$

$$en = mn - \widehat{mn}$$

$$\Rightarrow mq.n = eq.n + \widehat{mn}$$

$$= en + qn + \widehat{mn}$$

$$= mn - \widehat{mn} + qn + \widehat{mn}$$

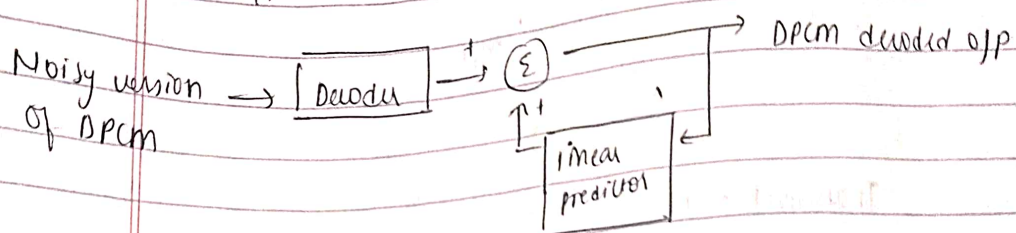
$$\boxed{mq.n = mn + qn}$$

Quantized version of original signal.

Ops at Transmitter:

- 1) Given predicted message signal \hat{m}_n , comparator computes e_n (prediction error).
- 2) e_n is quantized to produce e_{qn} (quantized version of e_n).
- 3) Adder produces quantized version of $m_n - \hat{m}_n$, m_{qn} .
- 4) \hat{m}_n is produced by applying sequence of quantized samples to a linear FIR predictor of order P .

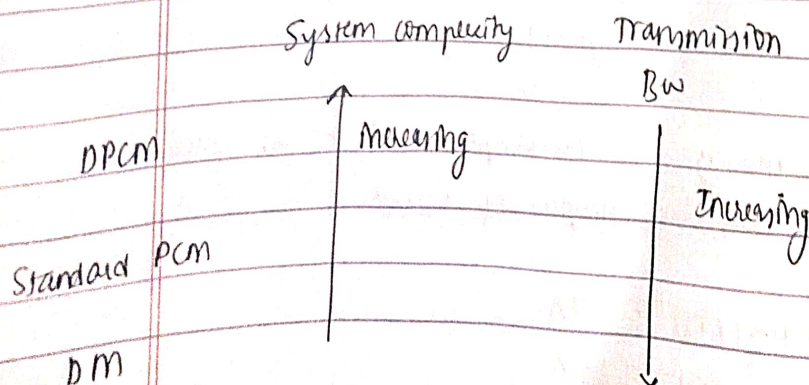
* Receiver:



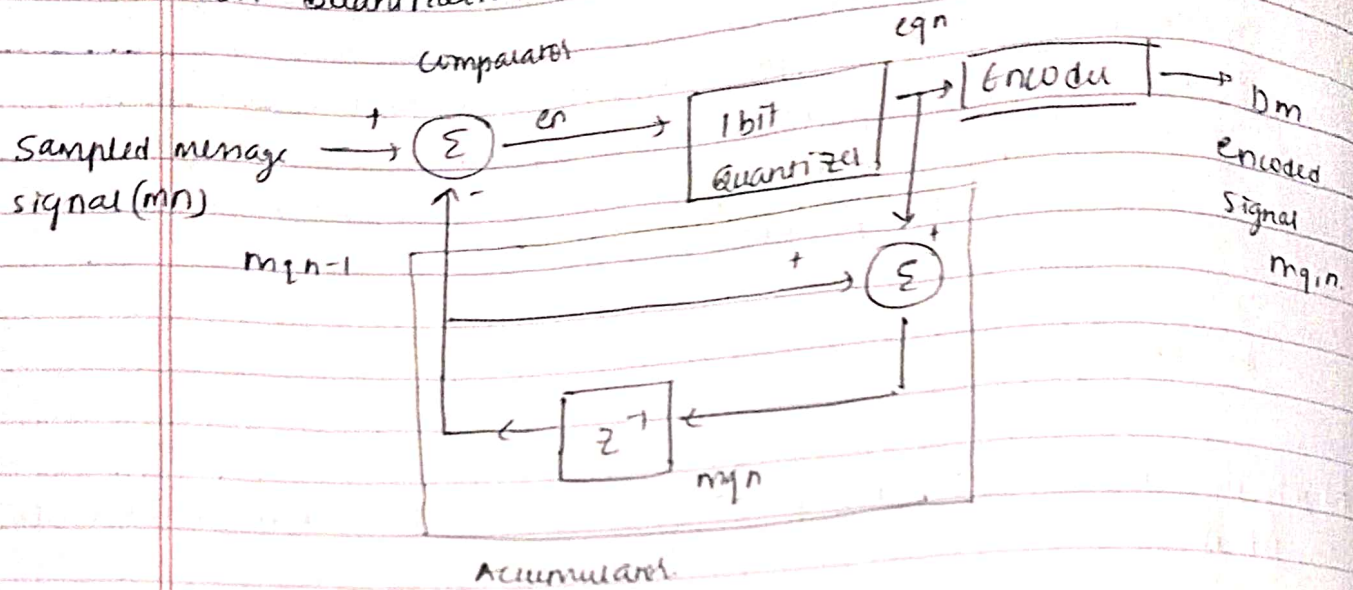
$$\begin{aligned}
 m_{qn} &= \hat{m}_n + e_{qn} \\
 &= m_n - e_n + e_{qn} \\
 &= e_n + q_n + m_n - e_n \\
 m_{qn} &= q_n + m_n
 \end{aligned}$$

→ Ops at Receiver:

- 1) sample structure
- 2) Decoder reconstructs quantized version of prediction error (e_{qn}).
- 3) Estimate of original o/p message sample m_n is then computed by applying decoder o/p to the same predictor as used in the Transmitter.



- 3) Delta modulation: Provides staircase approxⁿ to oversampled $m(n)$
 - 1 bit quantization.



Transmitter opⁿ:

- 1) Try to reduce system complexity by:
- (i) single bit quantizer: only 2 decision levels $\pm \Delta$
 - (ii) single unit delay element: primitive form of predictor.

$$e_n = m_n - \hat{m}_n$$

$$= m_n - m_{q,n-1}$$

$$e_n = \begin{cases} +\Delta & \text{if } e_n > 0 \\ -\Delta & \text{if } e_n < 0 \end{cases}$$

$$m_{q,n} = m_{q,n-1} + e_n$$

or accumulator:

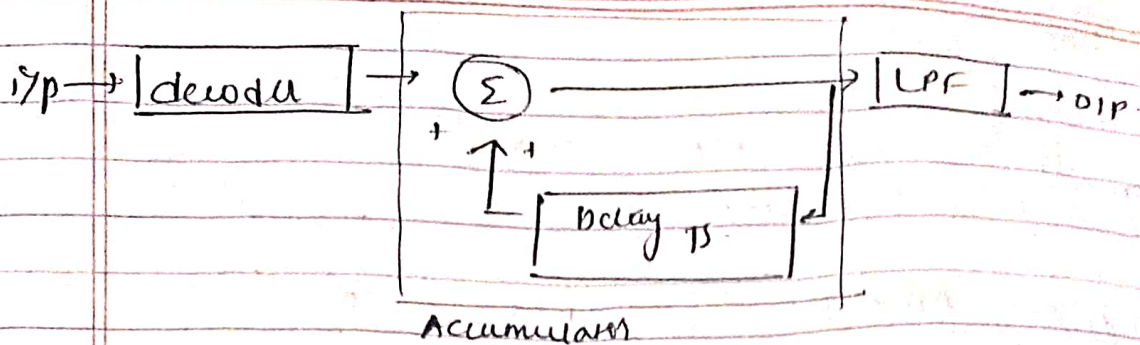
If $e_n = +ve$: $\hat{m}_n = m_{q,n-1}$ is increased by Δ , hence encoder sends out symbol 1.

If $e_n = -ve$: $\hat{m}_n = m_{q,n-1}$ is reduced by Δ , hence encoder sends out symbol 0.

- 2) Incoming message are oversampled - to increase correlation between adjacent samples of signals.

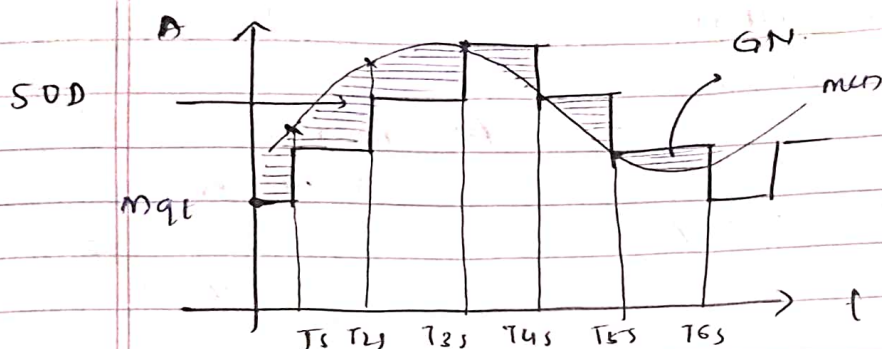
$$3) m(nT_s) > m_q(nT_s) = +\Delta$$

$$m(nT_s) < m_q(nT_s) = -\Delta$$



→ m_{qch} is reconstructed by passing sequence of '1's & -1's pulses, produced at decoder o/p.

→ The quantization noise in high frequency m_{qch} is rejected by passing through "LPF".



Noise:-

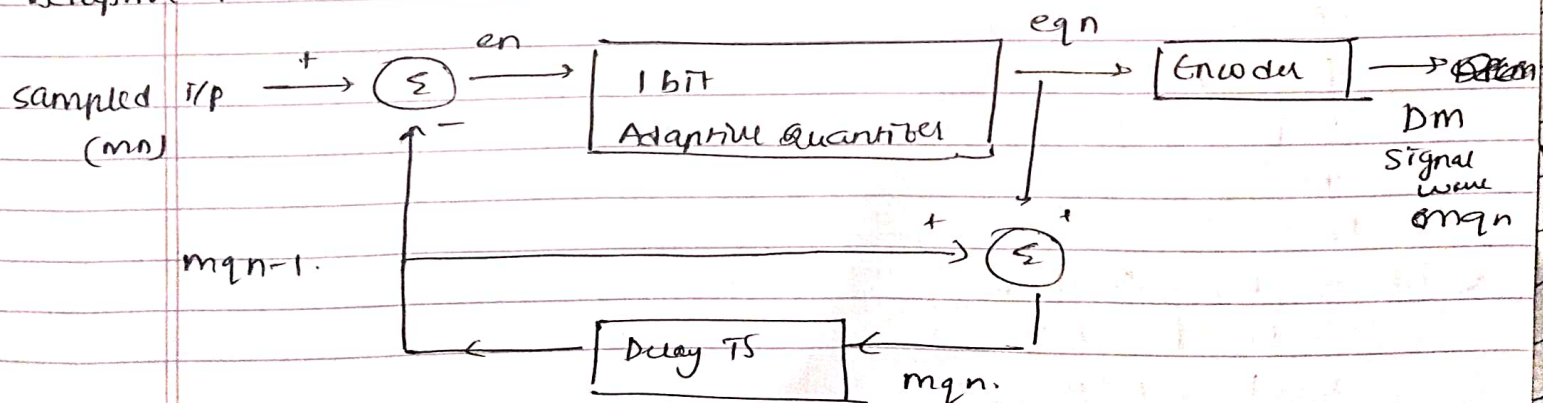
Slope overload distortion

Granular noise

If $x(t)$ is much higher than $x_{q(t)}$, over a long duration, then $x_{q(t)}$ will not be able to follow $x(t)$.

— When i/p signal $x(t)$ is relatively constant in amplitude $x_{q(t)}$ is bouncing up & down which will cause GN distortion.

Adaptive Noise modulation:



2)

SNR:

$$SNR = \frac{PS}{PN} = \frac{\text{Power signal}}{\text{Power noise}}$$

Consider input 'm' with range $[-m_{\max}, m_{\max}]$
 continuous amplitude

$$\Delta = \frac{m_{\max} - (-m_{\max})}{L}$$

$$\Delta = \frac{2m_{\max}}{L}$$

Let R = no. of bits / sample

$$L = 2^R$$

$$\Rightarrow \Delta = \frac{2m_{\max}}{2^R} \quad \text{--- (1)}$$

∴ The Quantization error 'Q' for uniform mid-rise quantizer will have samples.

$$-\frac{\Delta}{2} \leq q \leq \frac{\Delta}{2}$$

$$f_Q(q) = \begin{cases} \frac{1}{\Delta} & : -\frac{\Delta}{2} \leq q \leq \frac{\Delta}{2} \\ 0 & : \text{otherwise} \end{cases}$$

$$\sigma^2_Q = E(Q^2)$$

$$\Rightarrow = \int_{-\Delta/2}^{+\Delta/2} f_Q(q) \cdot q^2 \cdot dq$$

$$= \frac{1}{\Delta} \int_{-\Delta/2}^{+\Delta/2} q^2 \cdot dq$$

$$= \frac{1}{3\Delta} \left[\frac{\Delta^3}{8} + \frac{\Delta^3}{8} \right]$$

$$= \frac{1}{3\Delta} \left[\frac{2\Delta^3}{8} \right]$$

$$= \frac{\Delta^2}{12}$$

$$\sigma^2_Q = \frac{1}{12} \times \left[\frac{4m^2_{\max}}{2^{2R}} \right]$$

$$= \frac{1}{3} \times \left[\frac{m^2_{\max}}{2^{2R}} \right]$$

$$(SNR)_0 = \frac{P}{\sigma_Q^2}$$

$$= \frac{P}{1 \times m^2_{\max}} \times 3 \times 2^{2R}$$

$$\text{Let } P = \frac{Am^2}{2}$$

$$\text{but } \frac{Am^2}{2} = \frac{m^2_{\max}}{2}$$

$$= \frac{m^2_{\max}}{2 \times m^2_{\max}} \times 3 \times 2^{2R}$$

$$= \frac{3 \times 2^{2R}}{2}$$

$$(SNR)_0 = \frac{3 \times 2^{2R}}{2}$$

Take log on both sides \Rightarrow

$$10 \log (SNR)_0 = 10 \log \left(\frac{3 \times 2^{2R}}{2} \right)$$

$$10 \log (SNR)_0 = 10 \log(1.5) + 20 \log 2 \\ = 1.8 + 6R \quad \text{//}$$