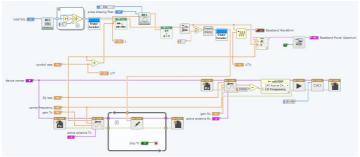
Lab 4 - Binary Phase Shift Keying (BPSK) Via USRP

Exercise 1 - BPSK Transmitter

Aim

In this exercise, we constructed a BPSK Transmitter

Diagram:



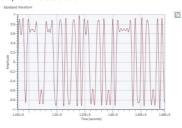
Observation

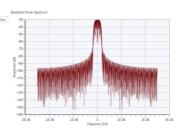
Values:

| Carrier frequency | 400 MHz |
|----------------------|---|
| IQ Rate | 200 kHz (Note: This sets the value of 1/Tx) |
| Gain | 0 dB |
| Active Antenna | TX1 |
| Symbol rate | 10,000 symbols/s |
| Message Length | 1000 bits |
| Pulse shaping filter | Root Raised |

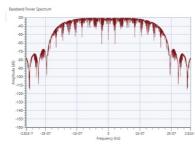
For a IQ Rate of 200k and a symbol rate of 10k the number of samples per symbol is 20.

Graphs for values above:



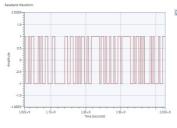


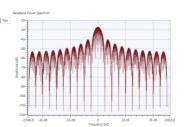
Main lobe:



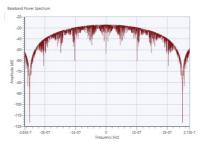
Main lobe bandwidth: 4e-07 Hz

No pulse shaping filter graphs:





No pulse shaping filter main lobe:



Main lobe bandwidth: 5.2e-07 Hz

- We observed that without a pulse shaping filter, the sideband lobes continued for a larger frequency range. This is because without a pulse shaping filter, higher frequency elements are required due to the rapid change in the message signal.
- Spectral rolloff is much faster using the pulse shaping filter

Exercise 2 - BPSK Receiver

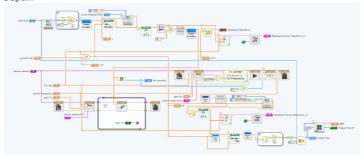
Aim

The steps needed to obtain the transmitted signals are:

- Ine steps needed to obtain the transmitted signals are:

 1. Channel Estimation: This is required to remove phase ambiguity caused by the channel and the USRP oscillators. Because the Tx and Rx channels are both using free running oscillators, there will be some ambiguous phase offset that is highly dependent on the drift and skew of the oscillators themselves. The channel estimator attempts to correct this problem by reading the received pilots which you inserted in the Tx signal then performing the LSE channel estimation to find the channel transfer function. The channel transfer function is then inverted on the Rx signal to remove phase offset.
- 2. Matched Filtering: We will use a root-raised-cosine receiver filter. This filter's impulse response $g_{\pi X}[n]$ is matched to the pulse shape $g_{\tau X}[n]$ of the transmitted pulses. The matched filter gives optimum performance in the presence of additive white Gaussian noise.
- 3. <u>Pulse Synchronization</u>: The matched filter output is an analog baseband signal that must be sampled once per symbol time, i.e. once every T seconds. Because of filtering, propagation delays, and distortion caused by the communication channel, it is necessary to determine the optimum time to take these samples. A sub-VI called PulseAlign has been provided to align the baseband signal.
- 4. Sampling: The Decimate function will sample the aligned baseband waveform at index 0 and every T seconds thereafter
- 5. <u>Detection</u>: Once the baseband waveform has been sampled, each sample must be examined to determine whether it represents a symbol of value 1 or a 0.
- Symbol Mapping: The detected symbol values must be converted to bits. For binary PSK, this step is easily included in the detection step.

In this exercise, we constructed the BPSK Receiver



Observation

| Tx Gain (dB) | Rx Gain (dB) | BER1 | BER2 | BER3 | BER4 | BER5 | Average BER |
|--------------|--------------|-------|-------|-------|-------|-------|-------------|
| 0 | 0 | 0.499 | 0.508 | 0.002 | 0.002 | 0.003 | 0.203 |
| -35 | -15 | 0.505 | 0.470 | 0.002 | 0.496 | 0.536 | 0.389 |
| -37 | -15 | 0.190 | 0.506 | 0.463 | 1 | 0.480 | 0.528 |
| -40 | -15 | 0.487 | 0.466 | 1 | 0.470 | 0.469 | 0.587 |

Exercise 3 - Error Correction Coding

Δim

Observation

- BER Module does not allow the program to compile.
- After doing all the steps, the program still did not work until the following steps were done:
- Needed to turn the output from the final pulse align module to real from complex.
- Needed to put a large constant into the number of samples port of the niUSRP Fetch Rx Data port so that it doesn't stop receiving data before the transmitter finishes transmitting.

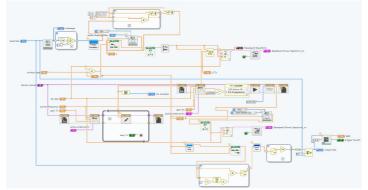
Finally, when it was working, the BER was much lower:

| Tx Gain (dB) | Rx Gain (dB) | BER1 | BER2 | BER3 | BER4 | BER5 | Average BER |
|--------------|--------------|-------|-------|-------|-------|-------|-------------|
| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| -35 | -15 | 0.001 | 0 | 0.438 | 0.369 | 0 | 0.162 |
| -37 | -15 | 0.452 | 0.256 | 0.353 | 0.490 | 0.457 | 0.402 |
| -40 | -15 | 1 | 0.457 | 0.480 | 0.506 | 0.501 | 0.589 |

Exercise 4 - Differential Phase Shift Keying (DPSK)

Aim

In this exercise, we constructed a DPSK Encoder and Decoder and added it to the transmitter and receivers



Observation

| Tx Gain (dB) | Rx Gain (dB) | BER1 | BER2 | BER3 | BER4 | BER5 | Average BER |
|--------------|--------------|-------|-------|-------|-------|-------|-------------|
| 0 | 0 | 0 | 0 | 0 | 0.507 | 0.449 | 0.191 |
| -35 | -15 | 0.491 | 0 | 0 | 0.489 | 0 | 0.196 |
| -37 | -15 | 0.060 | 0.481 | 0.022 | 0.456 | 0.001 | 0.204 |
| -40 | -15 | 1 | 0.171 | 0.501 | 1 | 0.495 | 0.633 |

Comparison between DPSK and BPSK

With the data we collected, DPSK performed better until the Tx Gain was reduced to -40. In theory, DPSK is just sending the differences, so has protection from any phase flipping effect that may occur between the receiver and transmitter.

However, since DPSK only transmits differences, it is possible that errors get propogated forward which is why it performed worse at very low gain.