

Chapter #7: Sequential Logic Case Studies

Contemporary Logic Design

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Motivation

- *Flipflops*: most primitive "packaged" sequential circuits

- *More complex sequential building blocks*:

Storage registers, Shift registers, Counters
Available as components in the TTL Catalog

- *How to represent and design simple sequential circuits*: counters

- *Problems and pitfalls when working with counters*:

Start-up States
Asynchronous vs. Synchronous logic

Chapter Overview

Examine Real Sequential Logic Circuits Available as Components

- **Registers for storage and shifting**
- **Random Access Memories**
- **Counters**

Counter Design Procedure

- **Simple but useful finite state machine**
- **State Diagram, State Transition Table, Next State Functions**
- **Excitation Tables for implementation with alternative flipflop types**

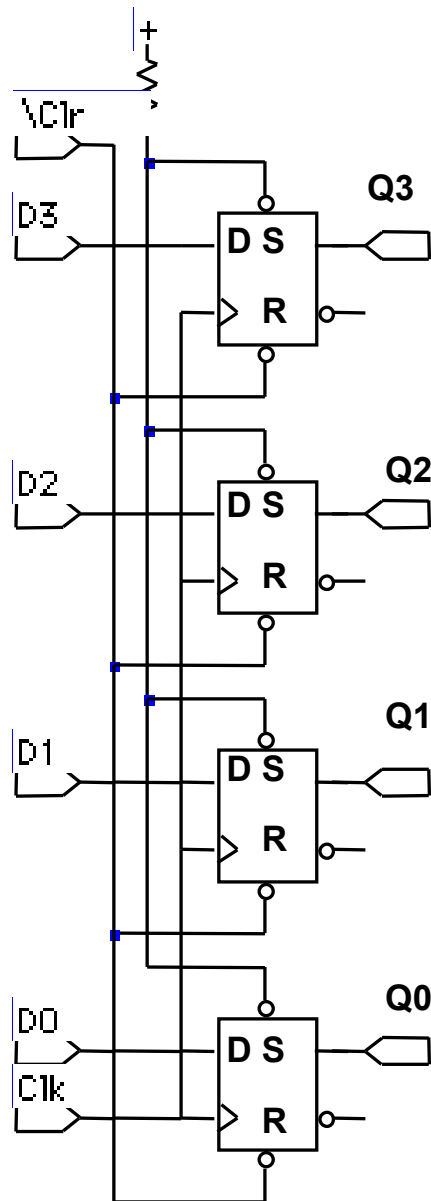
Synchronous vs. Asynchronous Counters

- **Ripple vs. Synchronous Counters**
- **Asynchronous vs. Synchronous Clears and Loads**

Kinds of Registers and Counters

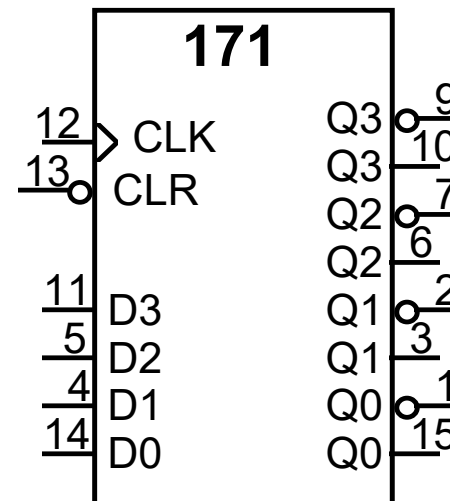
Storage Register

Group of storage elements read/written as a unit



4-bit register constructed from 4 D FFs
Shared clock and clear lines

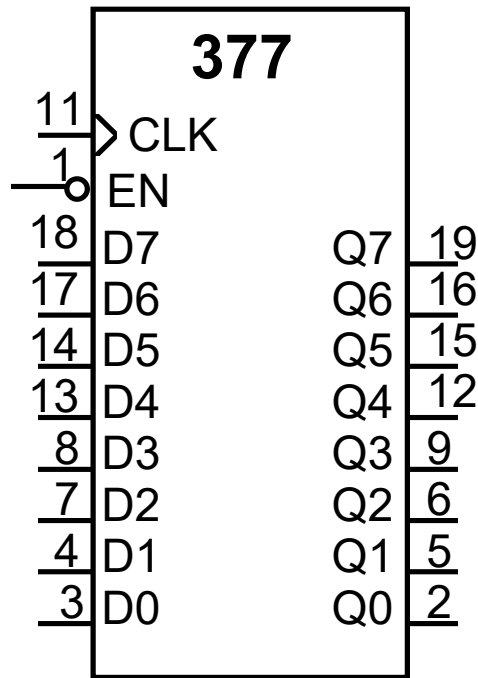
Schematic Shape



TTL 74171 Quad D-type FF with Clear
(Small numbers represent pin #s on package)

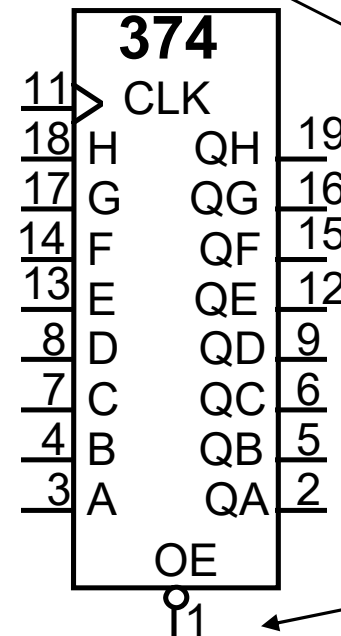
Input/Output Variations

Selective Load Capability
Tri-state or Open Collector Outputs
True and Complementary Outputs



**74377 Octal D-type FFs
with input enable**

*EN enabled low and
lo-to-hi clock transition
to load new data into
register*



**74374 Octal D-type FFs
with output enable**

*OE asserted low
presents FF state to
output pins; otherwise
high impedance*

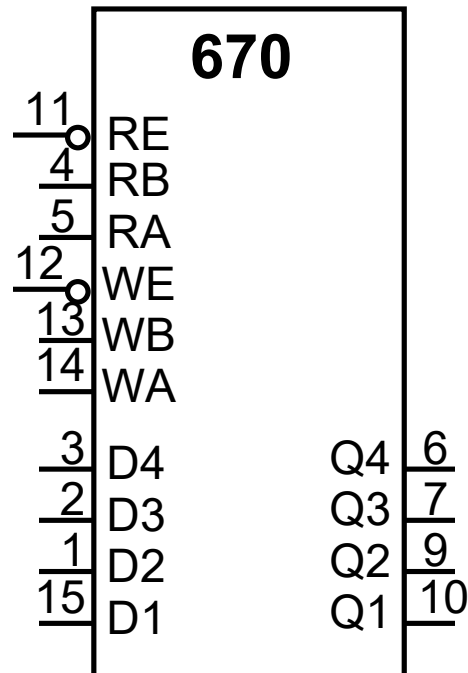
Kinds of Registers and Counters

Register Files

Two dimensional array of flipflops

Address used as index to a particular word

Word contents read or written



Separate Read and Write Enables
Separate Read and Write Address
Data Input, Q Outputs

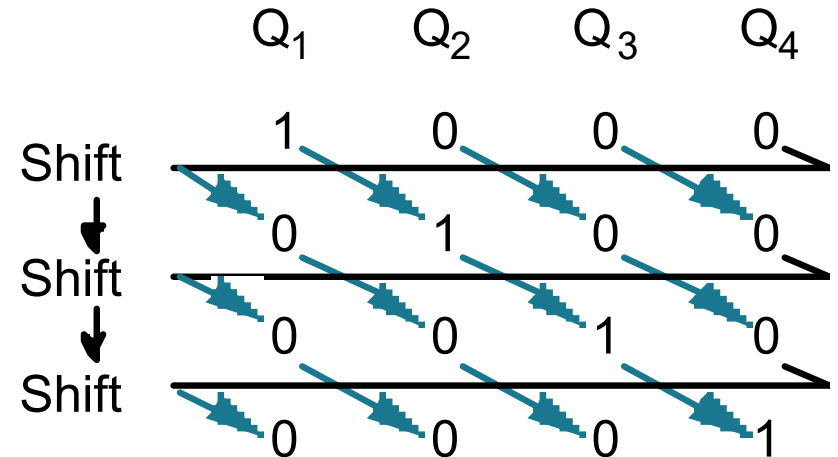
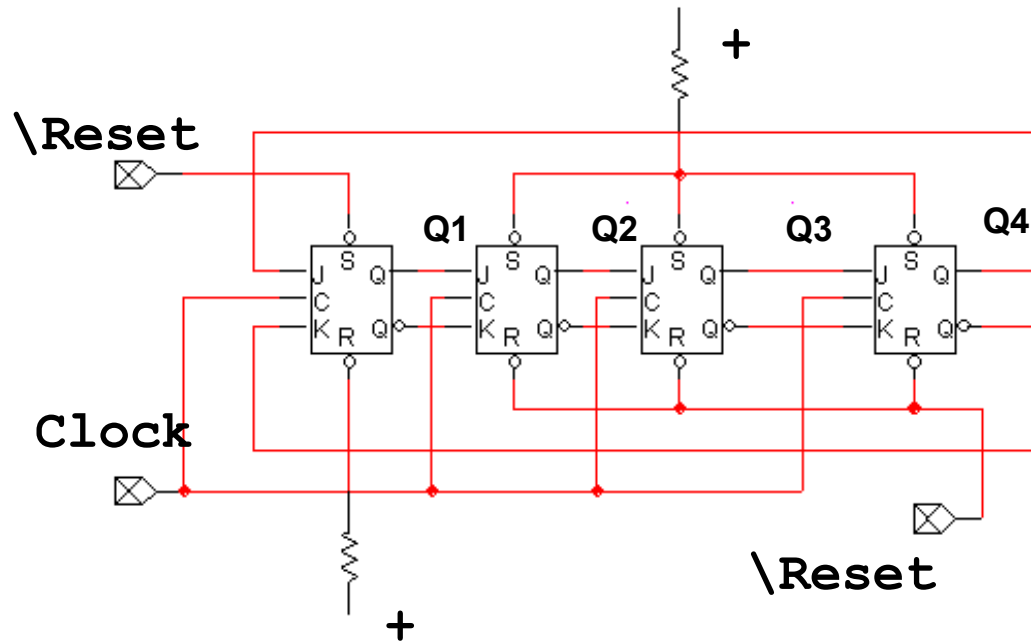
Contains 16 D-ffs, organized as
four rows (words) of four elements (bits)

**74670 4x4 Register File with
Tri-state Outputs**

Kinds of Registers and Counters

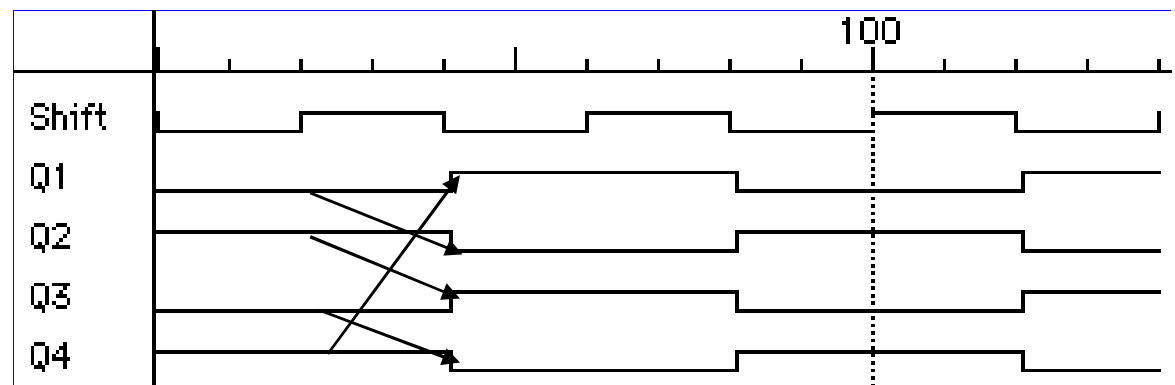
Shift Registers

Storage + ability to circulate data among storage elements



Shift from left storage element to right neighbor on every lo-to-hi transition on shift signal

Wrap around from rightmost element to leftmost element



Master Slave FFs: sample inputs while clock is high; change outputs on falling edge

Contemporary Sequential Ca

J(t)	K(t)	Q(t)	Q(t+Δ)	
0	0	0	0	HOLD
0	0	1	1	
0	1	0	0	RESET
0	1	1	0	
1	0	0	1	SET
1	0	1	1	
1	1	0	1	TOGGLE
1	1	1	0	

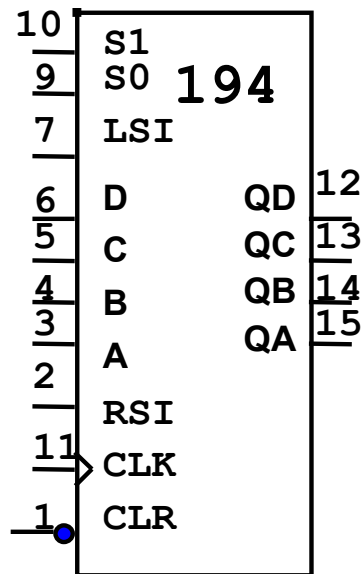
Kinds of Registers and Counters

Shift Register I/O

Serial vs. Parallel Inputs

Serial vs. Parallel Outputs

Shift Direction: Left vs. Right



74194 4-bit Universal
Shift Register

Serial Inputs: LSI, RSI

Parallel Inputs: D, C, B, A

Parallel Outputs: QD, QC, QB, QA

Clear Signal

Positive Edge Triggered Devices

S1, S0 determine the shift function

S1 = 1, S0 = 1: Load on rising clk edge
synchronous load

S1 = 1, S0 = 0: **shift left** on rising clk edge
LSI replaces element **D**

S1 = 0, S0 = 1: **shift right** on rising clk edge
RSI replaces element **A**

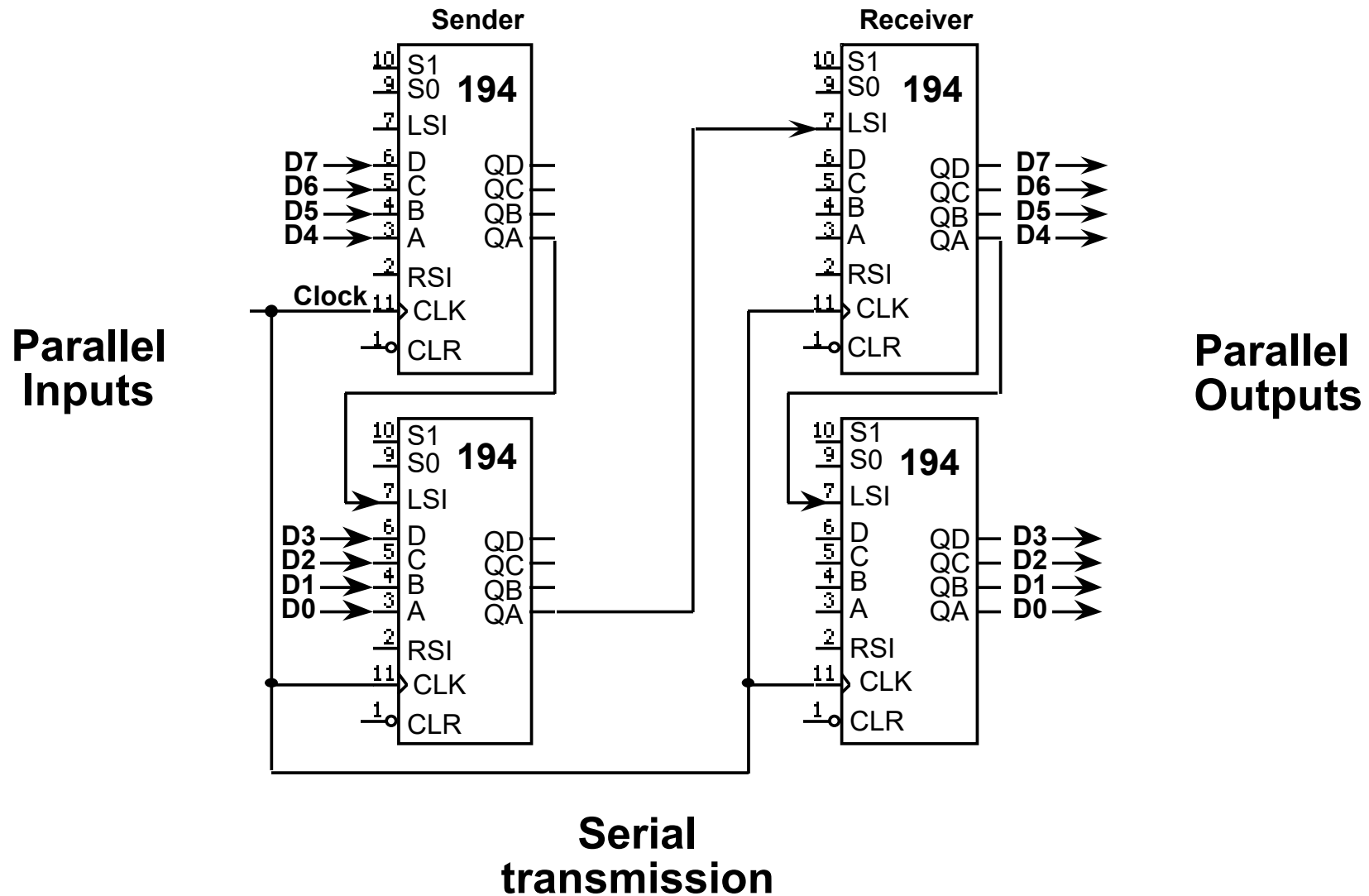
S1 = 0, S0 = 0: hold state

Multiplexing logic on input to each FF!

Shifters well suited for serial-to-parallel conversions,
such as terminal to computer communications

Kinds of Registers and Counters

Shift Register Application: Parallel to Serial Conversion



Kinds of Registers and Counters

Counters

Proceed through a well-defined sequence of states in response to count signal

3 Bit Up-counter: 000, 001, 010, 011, 100, 101, 110, 111, 000, ...

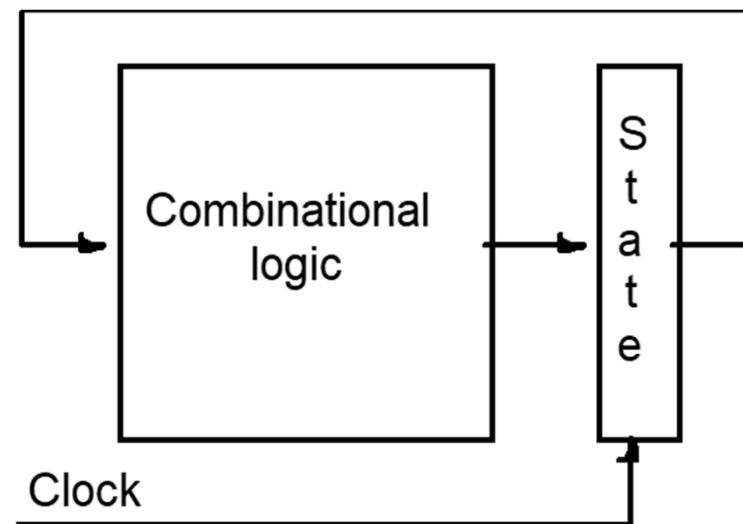
3 Bit Down-counter: 111, 110, 101, 100, 011, 010, 001, 000, 111, ...

Binary vs. BCD vs. Gray Code Counters

A counter is a "degenerate" finite state machine/sequential circuit where the state *is* the only output

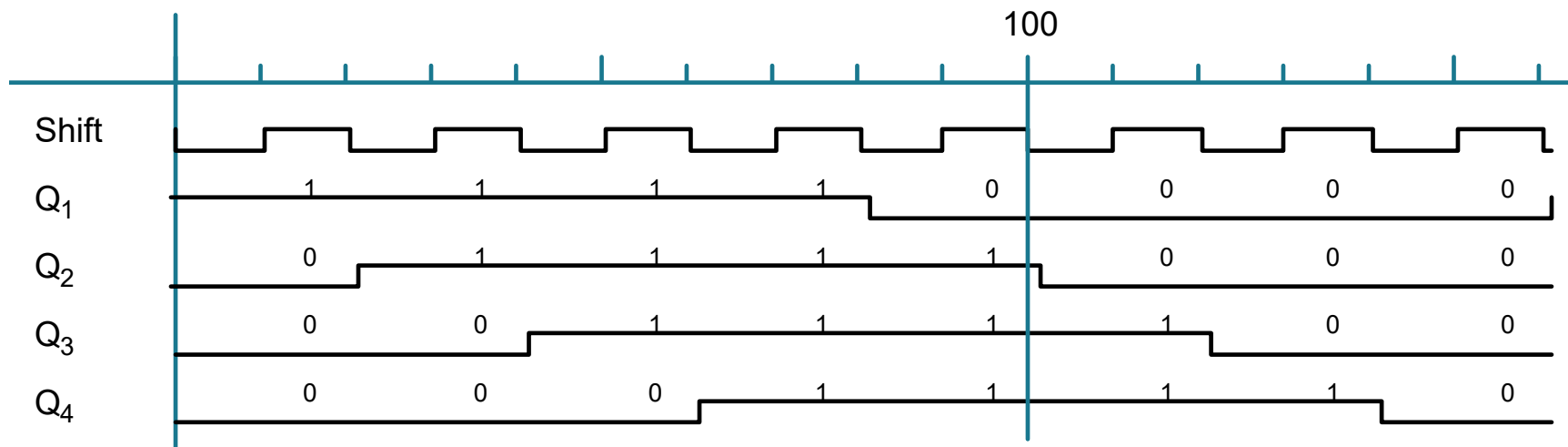
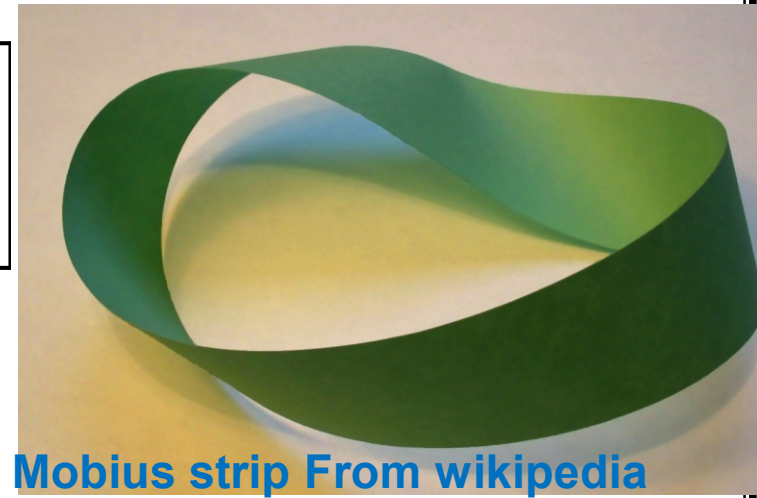
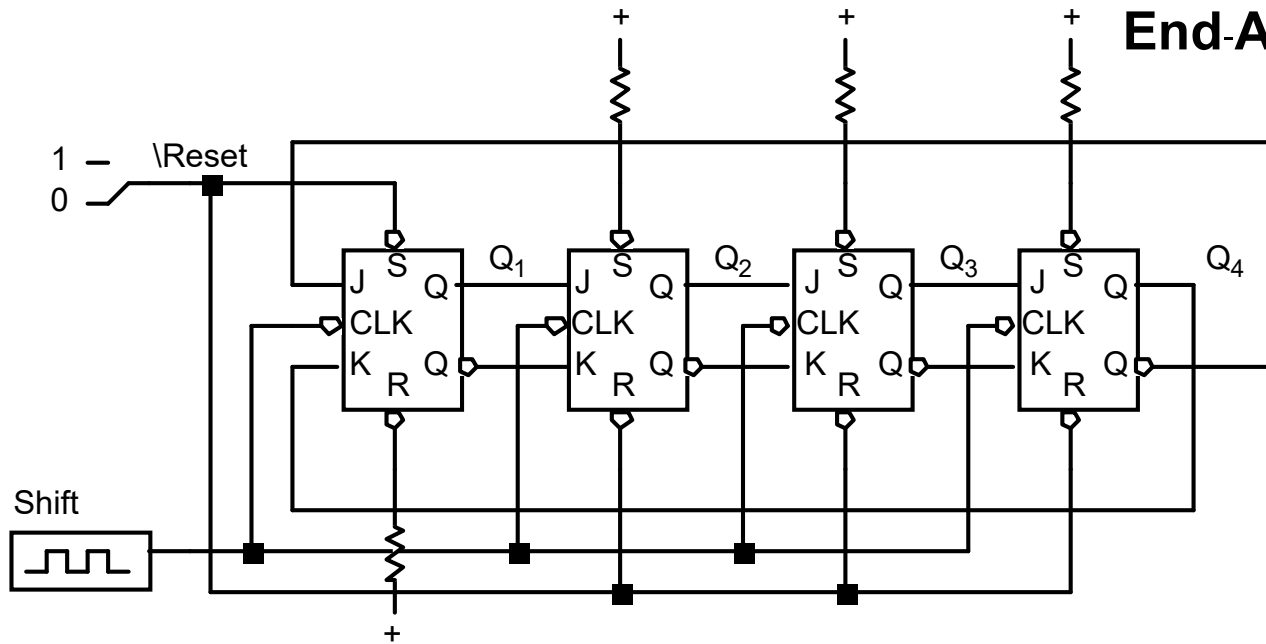
Generic Block Diagram for Clocked Sequential System

state implemented by
latches or edge-triggered FFs



Kinds of Registers and Counters

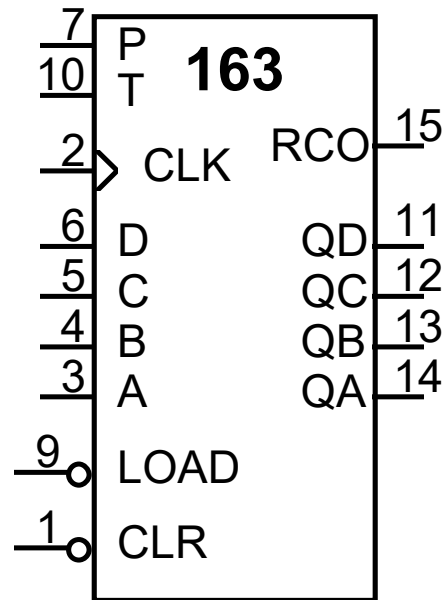
Johnson (Mobius) Counter



8 possible states, single bit change per state, useful for avoiding glitches

Kinds of Registers and Counters

Catalog Counter



**74163 Synchronous
4-Bit Upcounter**

Synchronous Load and Clear Inputs

Positive Edge Triggered FFs

Parallel Load Data from D, C, B, A

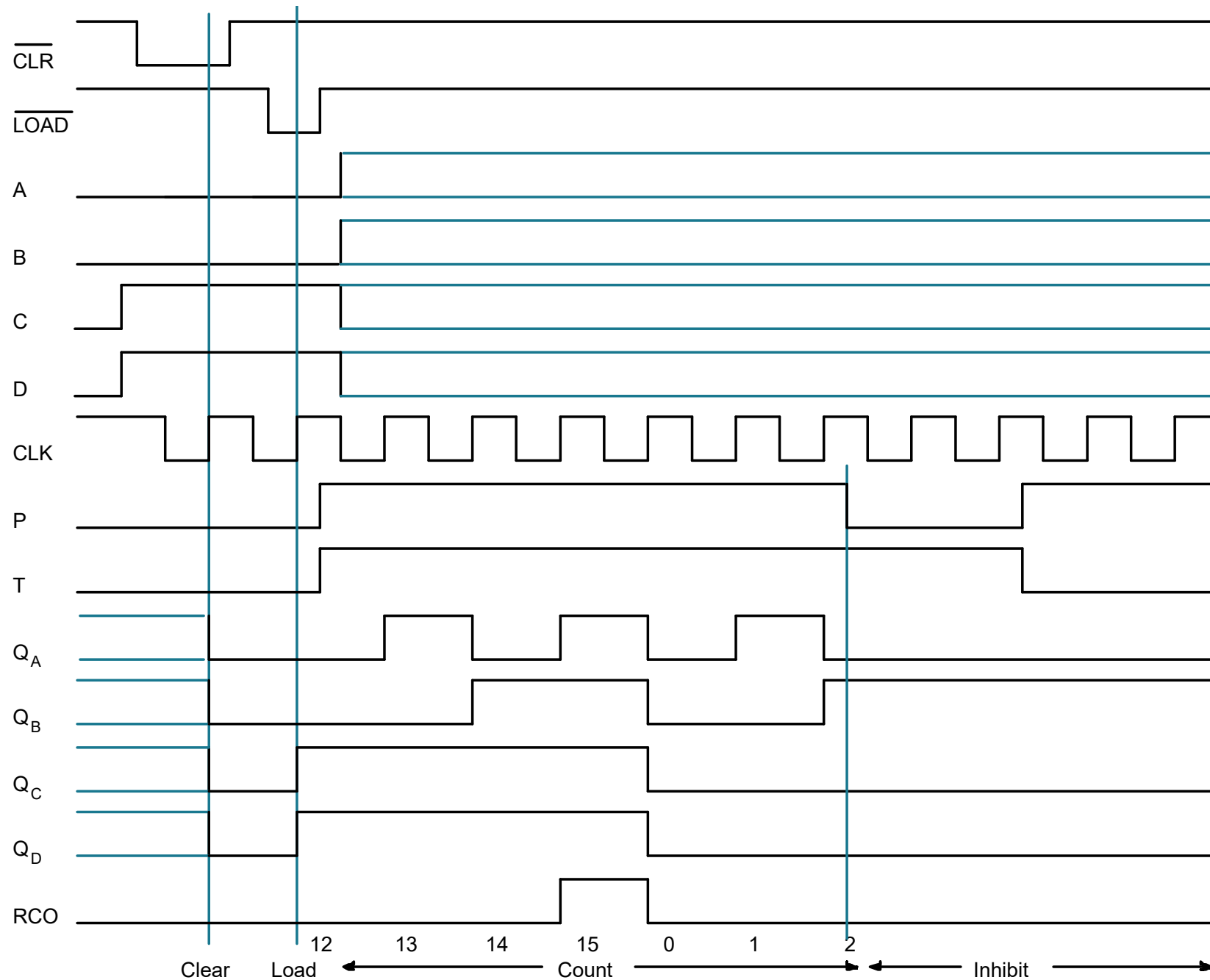
**P, T Enable Inputs: both must be asserted to
enable counting**

**RCO: asserted when counter enters its highest
state 1111, used for cascading counters**
"Ripple Carry Output"

74161: similar in function, asynchronous load and reset

Kinds of Registers and Counters

74163 Detailed Timing Diagram



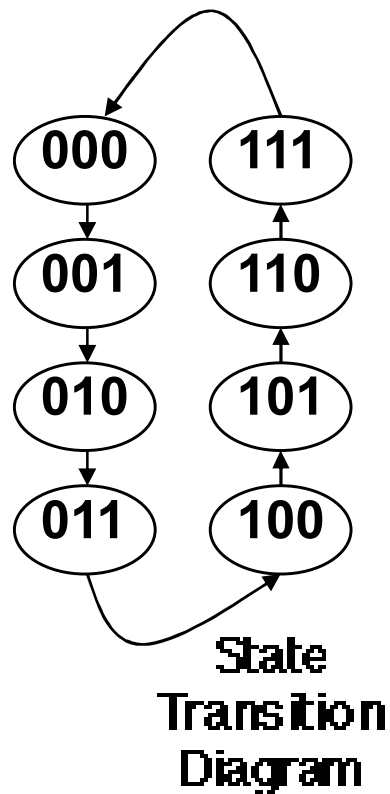
Counter Design Procedure

Introduction

This procedure can be generalized to implement ANY finite state machine

Counters are a very simple way to start:
no decisions on what state to advance to next
current state is the output

Example: 3-bit Binary Upcounter



Present State			Next State			Flip Flops Inputs		
C	B	A	C+	B+	A+	TC	TB	TA
0	0	0	0	0	1			
0	0	1	0	1	0			
0	1	0	0	1	1			
0	1	1	1	0	0			
1	0	0	1	0	1			
1	0	1	1	1	0			
1	1	0	1	1	1			
1	1	1	0	0	0			

State Transition Table

Flipflop Input Table

Q	Q ⁺	R	S	J	K	T
0	0	X	0	0	X	0
0	1	0	1	1	X	1
1	0	1	0	X	1	1
1	1	0	X	X	0	0

Decide to implement with Toggle Flipflops

What inputs must be presented to the T FFs to get them to change to the desired state bit?

This is called
"Remapping the Next State Function"

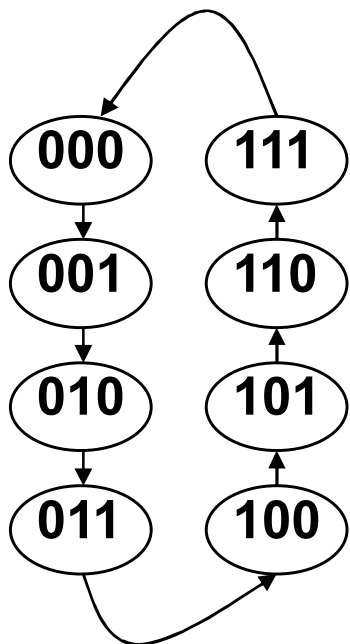
Counter Design Procedure

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current state is the output

Example: 3-bit Binary Upcounter



Present State			Next State			Flip Flops Inputs		
C	B	A	C+	B+	A+	TC	TB	TA
0	0	0	0	0	1	0	0	1
0	0	1	0	1	0	0	1	1
0	1	0	0	1	1	0	0	1
0	1	1	1	0	0	1	1	1
1	0	0	1	0	1	0	0	1
1	0	1	1	1	0	0	1	1
1	1	0	1	1	1	0	0	1
1	1	1	0	0	0	1	1	1

State Transition Table

Flipflop Input Table

Q	Q ⁺	R	S	J	K	T
0	0	X	0	0	X	0
0	1	0	1	1	X	1
1	0	1	0	X	1	1
1	1	0	X	X	0	0

Decide to implement with Toggle Flipflops

What inputs must be presented to the T FFs to get them to change to the desired state bit?

This is called
"Remapping the Next State Function"

Counter Design Procedure

Example Continued

K-maps for Toggle
Inputs:

A \ CB				
	00	01	11	10
0				
1				

TA =

A \ CB				
	00	01	11	10
0				
1				

TB =

A \ CB				
	00	01	11	10
0				
1				

TC =

Resulting Logic Circuit:

Present State			Next State			Flip Flops Inputs		
C	B	A	C+	B+	A+	TC	TB	TA
0	0	0	0	0	1	0	0	1
0	0	1	0	1	0	0	1	1
0	1	0	0	1	1	0	0	1
0	1	1	1	0	0	1	1	1
1	0	0	1	0	1	0	0	1
1	0	1	1	1	0	0	1	1
1	1	0	1	1	1	0	0	1
1	1	1	0	0	0	1	1	1

Example Continued

K-maps for Toggle Inputs:

		C			
		00	01	11	10
A	0	1	1	1	1
	1	1	1	1	1

B
 $TA = 1$

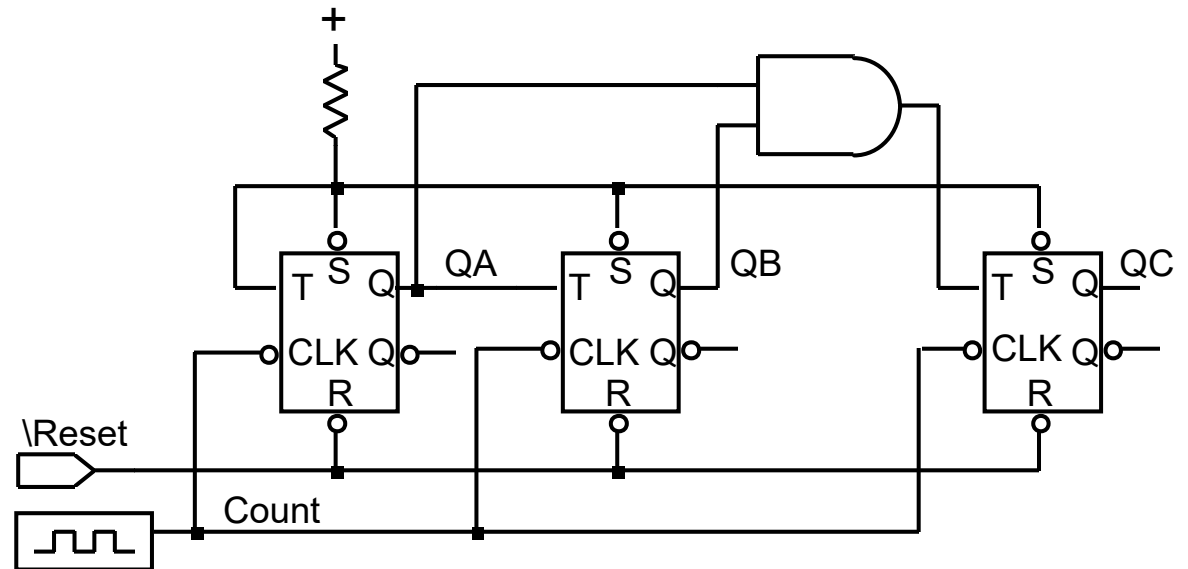
		C			
		00	01	11	10
A	0	0	0	0	0
	1	1	1	1	1

B
 $TB = A$

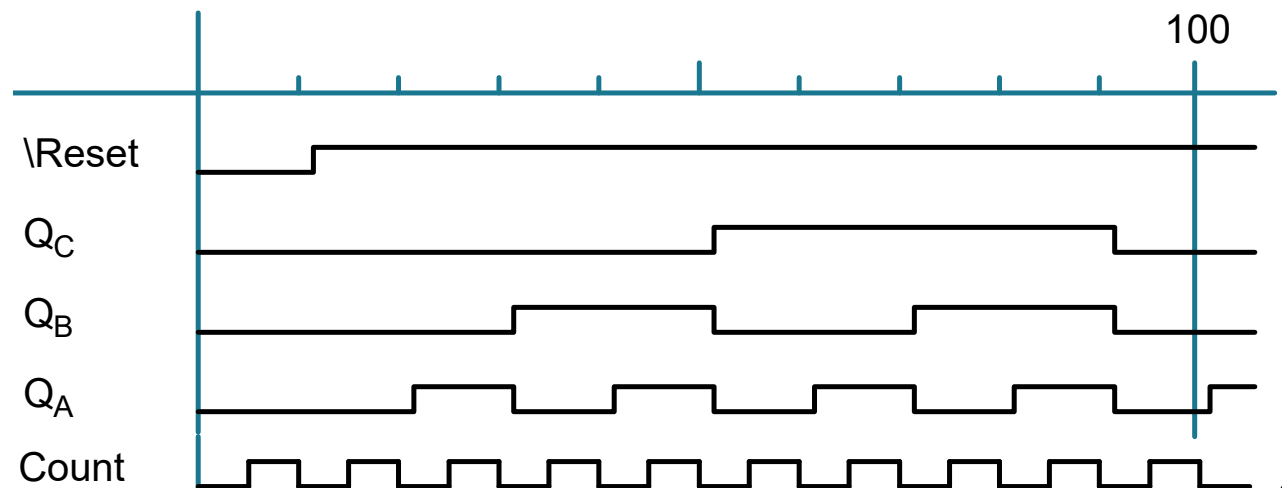
		C			
		00	01	11	10
A	0	0	0	0	0
	1	0	1	1	0

B
 $TC = A \cdot B$

Resulting Logic Circuit:

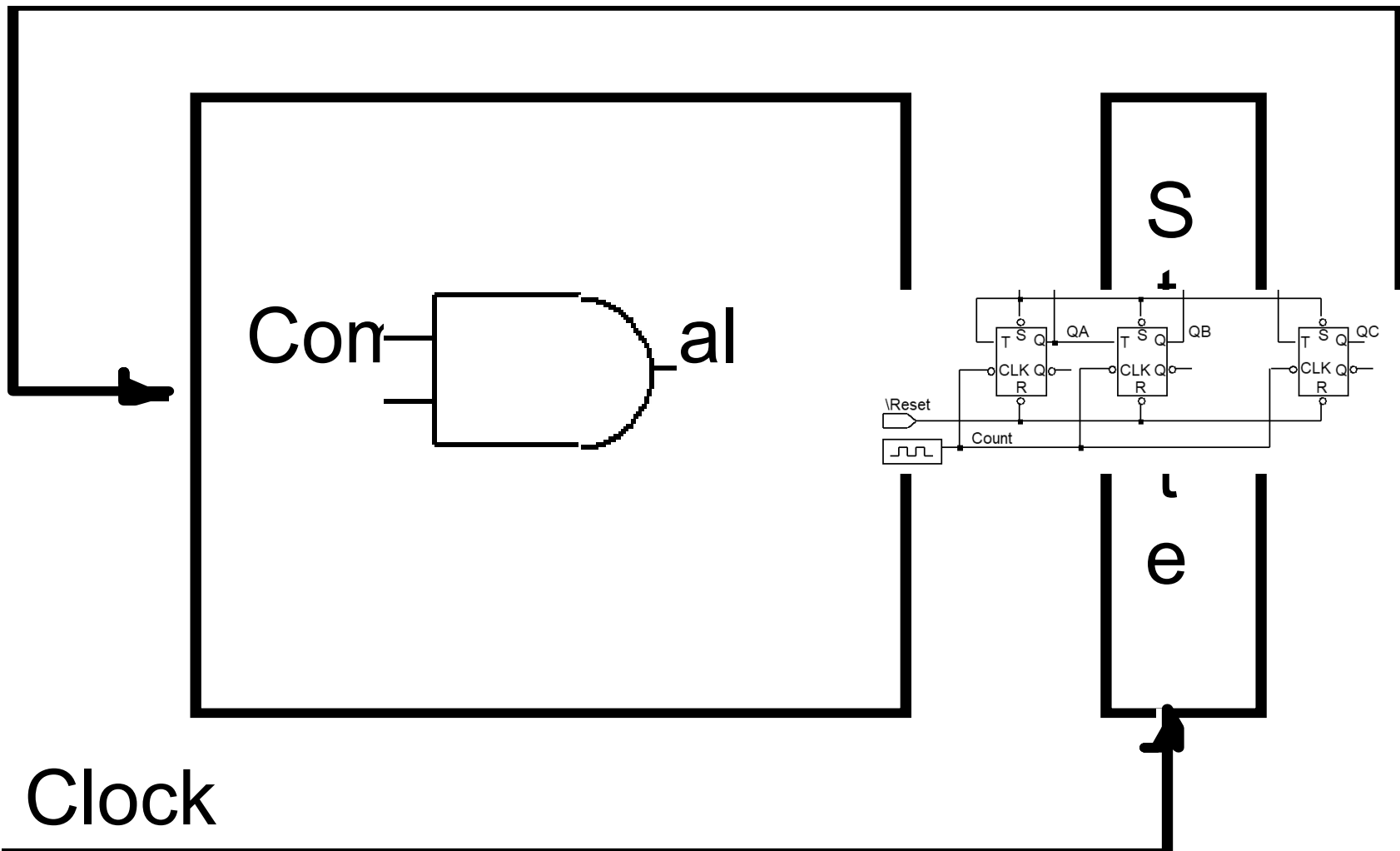
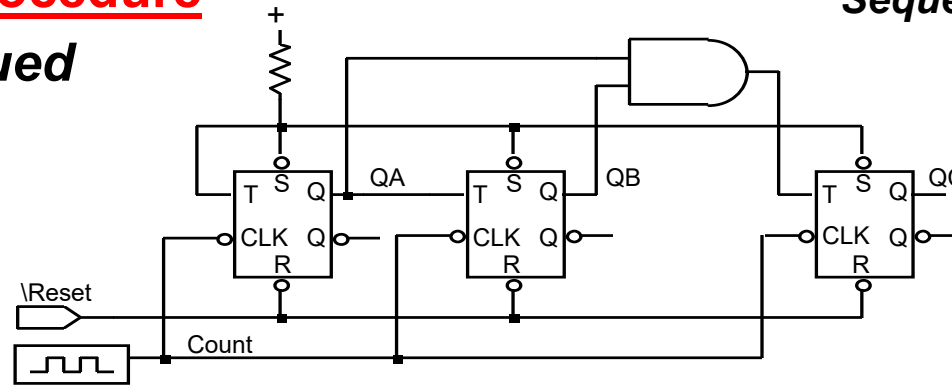


Timing Diagram:



Counter Design Procedure

Example Continued

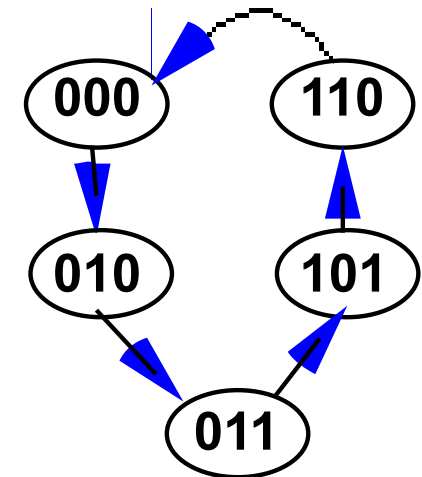


Counter Design Procedure

More Complex Count Sequence

Step 1: Derive the State Transition Diagram

Count sequence: 000, 010, 011, 101, 110



Present States			Next State		
C	B	A	C ⁺	B ⁺	A ⁺
0	0	0	0	1	0
0	0	1	x	x	x
0	1	0	0	1	1
0	1	1	1	0	1
1	0	0	x	x	x
1	0	1	1	1	0
1	1	0	0	0	0
1	1	1	x	x	x

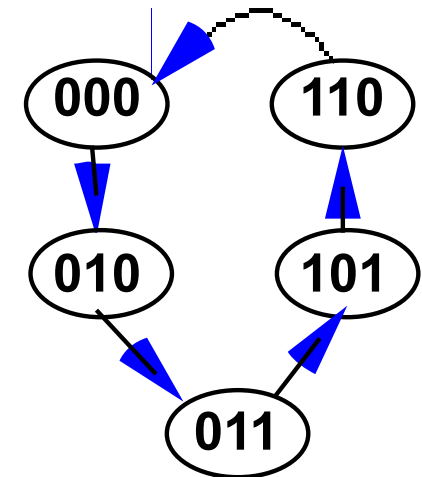
Step 2: State Transition Table

Counter Design Procedure

More Complex Count Sequence

Step 1: Derive the State Transition Diagram

Count sequence: 000, 010, 011, 101, 110



Present States			Next State		
C	B	A	C ⁺	B ⁺	A ⁺
0	0	0	0	1	0
0	0	1	x	x	x
0	1	0	0	1	1
0	1	1	1	0	1
1	0	0	x	x	x
1	0	1	1	1	0
1	1	0	0	0	0
1	1	1	x	x	x

Step 2: State Transition Table

Note the Don't Care conditions

More Complex Count Sequence

Step 3: K-Maps for Next State Functions

A \ CB				
	00	01	11	10
0				
1				

$C_{+} =$

A \ CB				
	00	01	11	10
0				
1				

$B_{+} =$

Present States			Next State		
C	B	A	C_{+}	B_{+}	A_{+}
0	0	0	0	1	0
0	0	1	x	x	x
0	1	0	0	1	1
0	1	1	1	0	1
1	0	0	x	x	x
1	0	1	1	1	0
1	1	0	0	0	0
1	1	1	x	x	x

A \ CB				
	00	01	11	10
0				
1				

$A_{+} =$

More Complex Count Sequence

Step 3: K-Maps for Next State Functions

A \ CB	CB			
	00	01	11	10
0	0	0	0	x
1	x	1	x	1

$C_{+} =$

A \ CB	CB			
	00	01	11	10
0	1	1	0	x
1	x	0	x	1

$B_{+} =$

Present States			Next State		
C	B	A	C_{+}	B_{+}	A_{+}
0	0	0	0	1	0
0	0	1	x	x	x
0	1	0	0	1	1
0	1	1	1	0	1
1	0	0	x	x	x
1	0	1	1	1	0
1	1	0	0	0	0
1	1	1	x	x	x

A \ CB	CB			
	00	01	11	10
0	0	1	0	x
1	x	1	x	0

$A_{+} =$

Present State	Next State	Flip Flop input
0	0	0
0	1	1
1	0	1
1	1	0

Counter Design Procedure

More Complex Counter Sequencing

Step 4: Choose Flipflop Type for Implementation
Use Excitation Table to Remap Next State Functions

Q	Q+	T
0	0	0
0	1	1
1	0	1
1	1	0

Toggle Excitation Table

Present States			Next State		
C	B	A	C ⁺	B ⁺	A ⁺
0	0	0	0	1	0
0	0	1	x	x	x
0	1	0	0	1	1
0	1	1	1	0	1
1	0	0	x	x	x
1	0	1	1	1	0
1	1	0	0	0	0
1	1	1	x	x	x

Toggle Inputs		
TC	TB	TA
0	1	0
x	x	x
0	0	1
1	1	0
x	x	x
0	1	1
1	1	0
x	x	x

Remapped Next State Functions

Counter Design Procedure

More Complex Count Sequence

A \ CB	CB			
	00	01	11	10
0	0	0	1	x
1	x	1	x	0

TC =

A \ CB	CB			
	00	01	11	10
0	1	0	1	x
1	x	1	x	1

TB =

Present States			Next State			Toggle Inputs		
C	B	A	C ⁺	B ⁺	A ⁺	TC	TB	TA
0	0	0	0	1	0	0	1	0
0	0	1	x	x	x	x	x	x
0	1	0	0	1	1	0	0	1
0	1	1	1	0	1	1	1	0
1	0	0	x	x	x	x	x	x
1	0	1	1	1	0	0	1	1
1	1	0	0	0	0	1	1	0
1	1	1	x	x	x	x	x	x

A \ CB	CB			
	00	01	11	10
0	0	1	0	x
1	x	0	x	1

TA =

$$TC = \bar{A} C + A \bar{C} = A \text{ xor } C$$

$$TB = A + \bar{B} + C$$

$$TA = \bar{A} B \bar{C} + \bar{B} C$$

Self-Starting Counters

Start-Up States

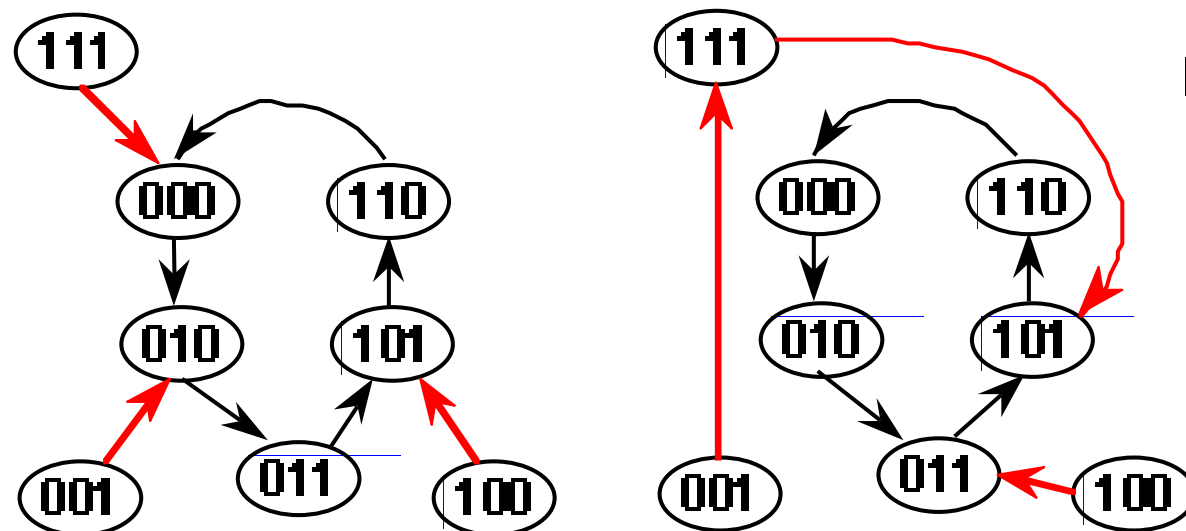
At power-up, counter may be in possible state

Designer must guarantee that it (eventually) enters a valid state

Especially a problem for counters that validly use a subset of states

Self-Starting Solution:

Design counter so that even the invalid states
eventually transition to valid state



Implementation
in Previous
Slide!

*Two Self-Starting State Transition Diagrams
for the Example Counter*

Self-Starting Counters

Deriving State Transition Table from Don't Care Assignment

Inputs to Toggle Flip-flops

State Changes

Inputs to Toggle Flip-flops

		CB		C		
A		00	01	11	10	
0		0	0	1	<u>1</u>	TC
1		<u>1</u>	1	<u>0</u>	0	
		B				

Inputs to Toggle Flip-flops

		CB		C		
A		00	01	11	10	
0		1	0	1	<u>1</u>	TB
1		<u>1</u>	1	<u>1</u>	1	
		B				

Inputs to Toggle Flip-flops

		CB		C		
A		00	01	11	10	
0		0	1	0	<u>1</u>	TA
1		<u>0</u>	0	<u>0</u>	1	
		B				

State Changes

		CB		C		
A		00	01	11	10	
0		0	0	0	<u>0</u>	C+
1		1	1	1	1	
		B				

State Changes

		CB		C		
A		00	01	11	10	
0		1	1	0	<u>1</u>	B+
1		1	0	0	1	
		B				

State Changes

		CB		C		
A		00	01	11	10	
0		0	1	0	<u>1</u>	A+
1		1	1	1	0	
		B				

State Transition Table

Present State			Next State		
C	B	A	C+	B+	A+
0	0	0	0	1	0
0	0	1	1	1	1
0	1	0	0	1	1
0	1	1	1	0	1
1	0	0	<u>0</u>	1	1
1	0	1	1	1	0
1	1	0	0	0	0
1	1	1	1	0	1

Implementation with Different Kinds of FFs

R-S Flipflops

Continuing with the 000, 010, 011, 101, 110, 000, ... counter example

Q	Q+	R	S
0	0	X	0
0	1	0	1
1	0	1	0
1	1	0	X

$$Q_+ = S + R Q$$

RS Excitation Table

Present State			Next State			Remapped Next State					
C	B	A	C+	B+	A+	RC	SC	RB	SB	RA	SA
0	0	0	0	1	0						
0	0	1	X	X	X						
0	1	0	0	1	1						
0	1	1	1	0	1						
1	0	0	X	X	X						
1	0	1	1	1	0						
1	1	0	0	0	0						
1	1	1	X	X	X						

Remapped Next State Functions

Implementation with Different Kinds of FFs

R-S Flipflops

Continuing with the 000, 010, 011, 101, 110, 000, ... counter example

Q	Q+	R	S
0	0	X	0
0	1	0	1
1	0	1	0
1	1	0	X

$$Q_+ = S + R Q^-$$

RS Excitation Table

Present State			Next State			Remapped Next State					
C	B	A	C+	B+	A+	RC	SC	RB	SB	RA	SA
0	0	0	0	1	0	X	0	0	1	X	0
0	0	1	X	X	X	X	X	X	X	X	X
0	1	0	0	1	1	X	0	0	X	0	1
0	1	1	1	0	1	0	1	1	0	0	X
1	0	0	X	X	X	X	X	X	X	X	X
1	0	1	1	1	0	0	X	0	1	1	0
1	1	0	0	0	0	1	0	1	0	X	0
1	1	1	X	X	X	X	X	X	X	X	X

Remapped Next State Functions

Implementation with Different Kinds of FFs

RS FFs Continued

A \ CB				
	00	01	11	10
0				
1				

RC

A \ CB				
	00	01	11	10
0				
1				

SC

A \ CB				
	00	01	11	10
0				
1				

RB

A \ CB				
	00	01	11	10
0				
1				

SB

A \ CB				
	00	01	11	10
0				
1				

RA

A \ CB				
	00	01	11	10
0				
1				

SA

RC =

SC =

RB =

SB =

RA =

SA =

Implementation with Different Kinds of FFs

RS FFs Continued

A \ CB				
	00	01	11	10
0	X	X	1	X
1	X	0	X	0

RC

A \ CB				
	00	01	11	10
0	0	0	1	X
1	X	1	X	0

RB

A \ CB				
	00	01	11	10
0	X	0	X	X
1	X	0	X	1

RA

A \ CB				
	00	01	11	10
0	0	0	0	X
1	X	1	X	X

SC

A \ CB				
	00	01	11	10
0	1	X	0	X
1	X	0	X	1

SB

A \ CB				
	00	01	11	10
0	0	1	0	X
1	X	X	X	0

SA

$$RC = \bar{A}$$

$$SC = A$$

$$RB = A B + B C$$

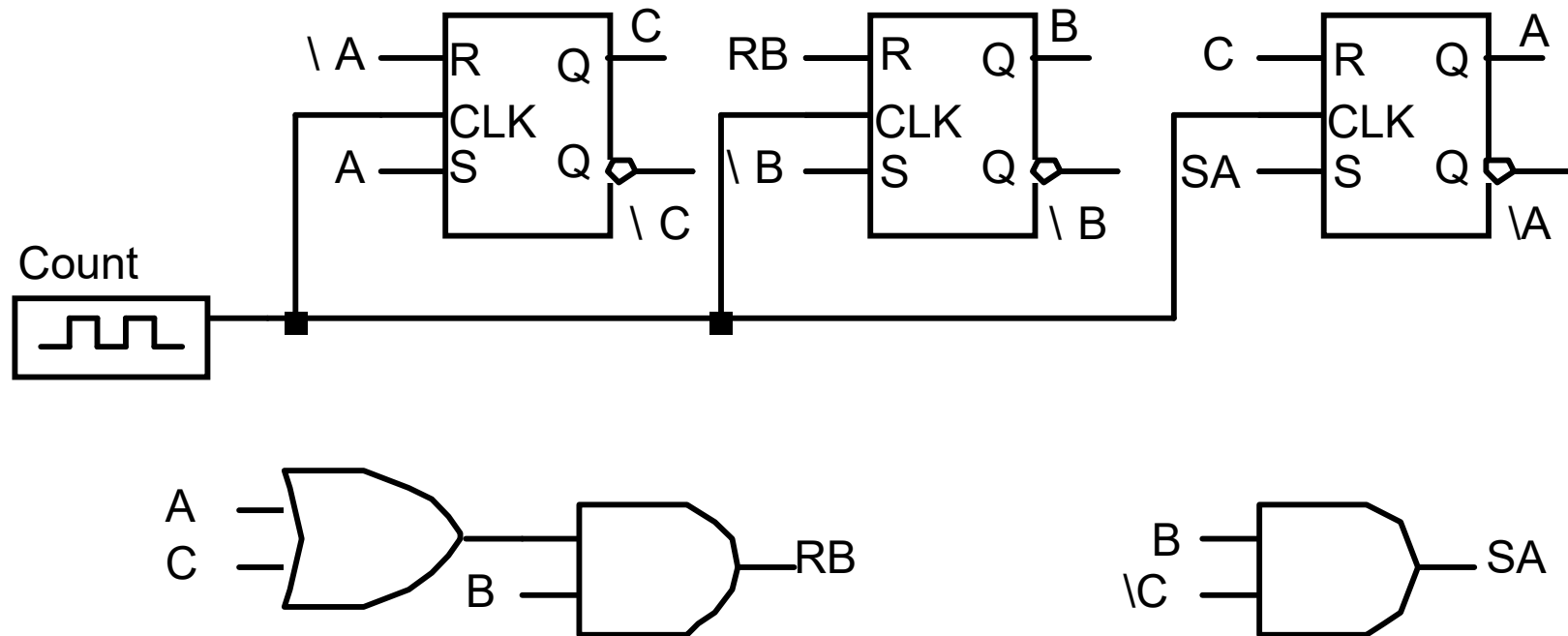
$$SB = \bar{B}$$

$$RA = C$$

$$SA = B \bar{C}$$

Implementation With Different Kinds of FFs

RS FFs Continued



Resulting Logic Level Implementation:
3 Gates, 11 Input Literals + Flipflop connections

Implementation with Different FF Types

J-K FFs

Q	Q+	J	K
0	0	0	X
0	1	1	X
1	0	X	1
1	1	X	0

$$Q_+ = JQ + K\bar{Q}$$

J-K Excitation Table

Present State			Next State			Remapped Next State					
C	B	A	C+	B+	A+	JC	KC	JB	KB	JA	KA
0	0	0	0	1	0						
0	0	1	X	X	X						
0	1	0	0	1	1						
0	1	1	1	0	1						
1	0	0	X	X	X						
1	0	1	1	1	0						
1	1	0	0	0	0						
1	1	1	X	X	X						

Remapped Next State Functions

Implementation with Different FF Types

J-K FFs

Q	Q+	J	K
0	0	0	X
0	1	1	X
1	0	X	1
1	1	X	0

$$Q_+ = JQ + \overline{K}\overline{Q}$$

J-K Excitation Table

Present State			Next State			Remapped Next State					
C	B	A	C+	B+	A+	JC	KC	JB	KB	JA	KA
0	0	0	0	1	0	0	X	1	X	0	X
0	0	1	X	X	X	X	X	X	X	X	X
0	1	0	0	1	1	0	X	X	0	1	X
0	1	1	1	0	1	1	X	X	1	X	0
1	0	0	X	X	X	X	X	X	X	X	X
1	0	1	1	1	0	X	0	1	X	X	1
1	1	0	0	0	0	X	1	X	1	0	X
1	1	1	X	X	X	X	X	X	X	X	X

Remapped Next State Functions

Implementation with Different FF Types

J-K FFs Continued

A \ CB				
	00	01	11	10
0				
1				

JC

A \ CB				
	00	01	11	10
0				
1				

KC

A \ CB				
	00	01	11	10
0				
1				

JB

A \ CB				
	00	01	11	10
0				
1				

KB

A \ CB				
	00	01	11	10
0				
1				

JA

A \ CB				
	00	01	11	10
0				
1				

KA

JC =

KC =

JB =

KB =

JA =

KA =

Implementation with Different FF Types

J-K FFs Continued

A \ CB				
	00	01	11	10
0	0	0	X	X
1	X	1	X	X

JC

A \ CB				
	00	01	11	10
0	X	X	1	X
1	X	X	X	0

KC

A \ CB				
	00	01	11	10
0	1	X	X	X
1	X	X	X	1

JB

A \ CB				
	00	01	11	10
0	X	0	1	X
1	X	1	X	X

KB

A \ CB				
	00	01	11	10
0	0	1	0	X
1	X	X	X	X

JA

A \ CB				
	00	01	11	10
0	X	X	X	X
1	X	0	X	1

KA

$$JC = A$$

$$KC = \bar{A}$$

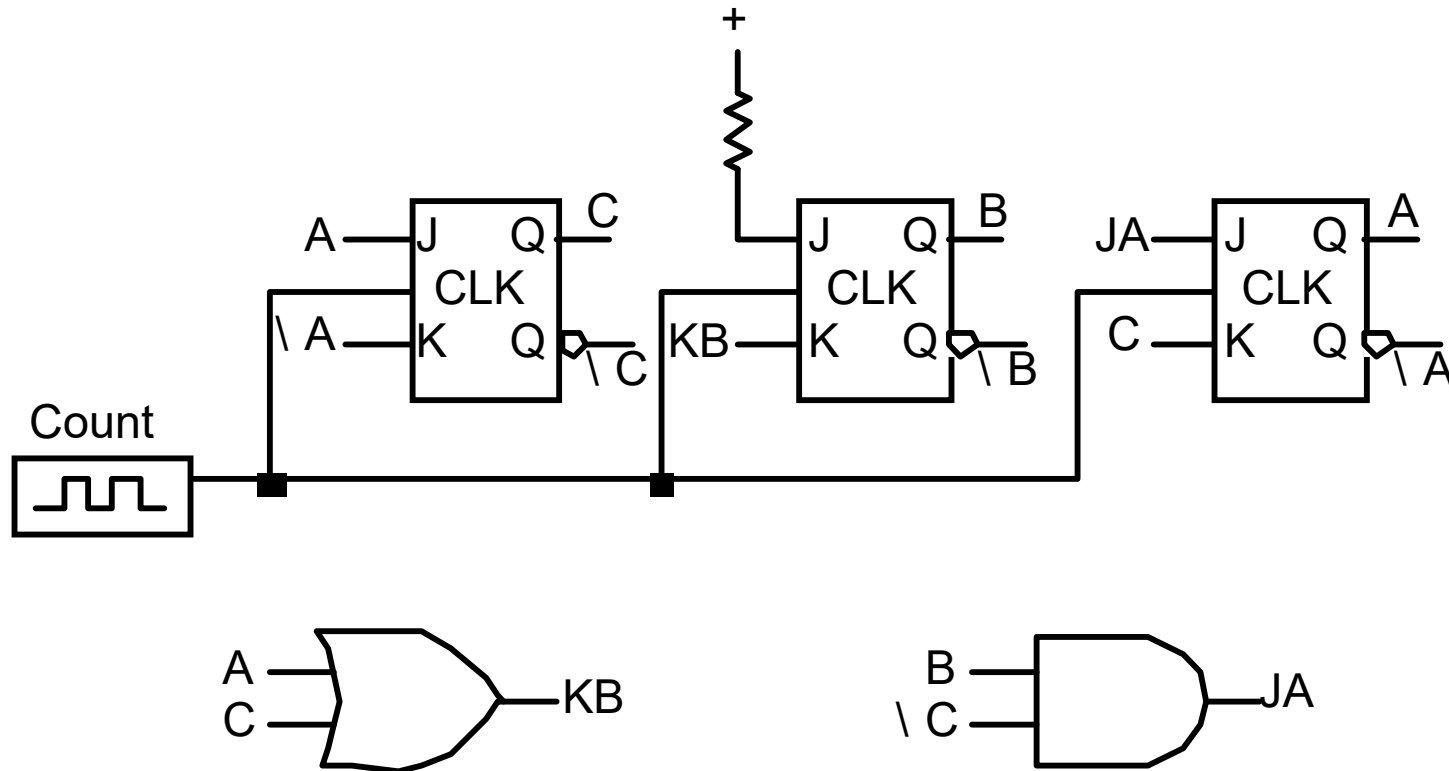
$$JB = 1$$

$$KB = A + C$$

$$JA = B \bar{C}$$

$$KA = C$$

J-K FFs Continued



Resulting Logic Level Implementation:
2 Gates, 10 Input Literals + Flipflop Connections

Implementation with Different FF Types

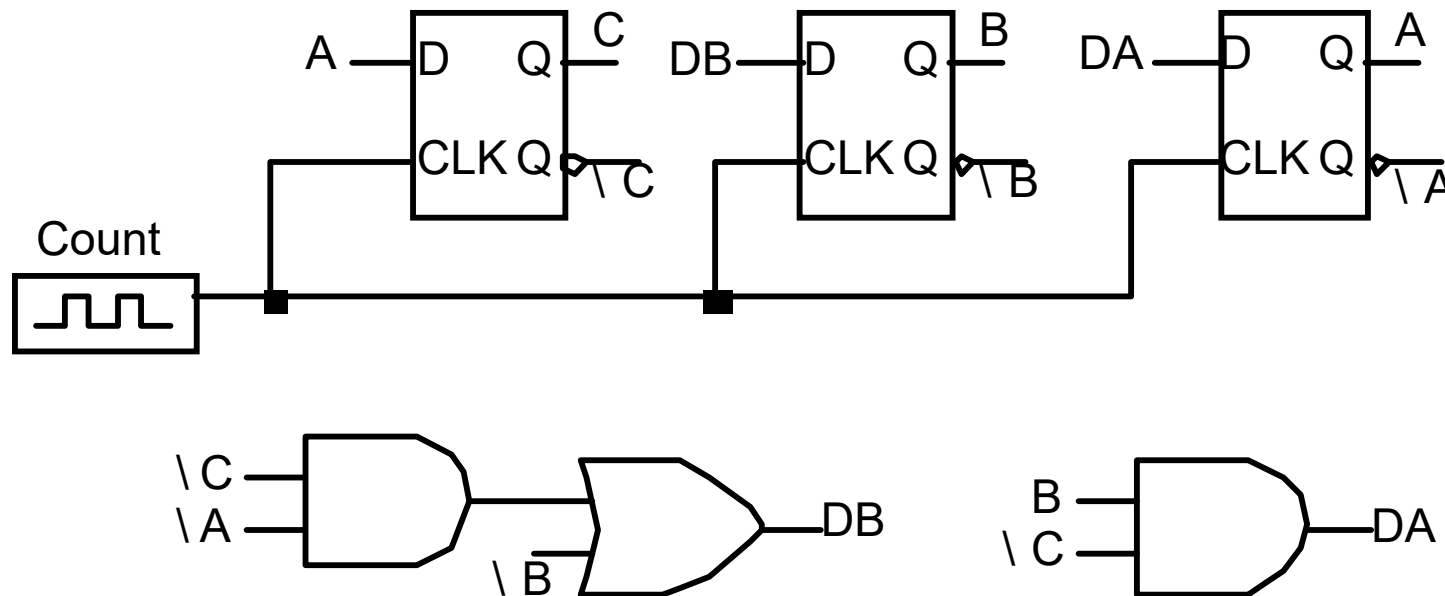
D FFs

Simplest Design Procedure: No remapping needed!

$$DC = A$$

$$DB = \bar{A} \bar{C} + B$$

$$DA = B \bar{C}$$



Resulting Logic Level Implementation:
3 Gates, 8 Input Literals + Flipflop connections

Implementation with Different FF Types

Comparison

- T FFs well suited for straightforward binary counters

But yielded worst gate and literal count for this example!

- No reason to choose R-S over J-K FFs: it is a proper subset of J-K

R-S FFs don't really exist anyway

J-K FFs yielded lowest gate count

Tend to yield best choice for packaged logic where gate count is key

- D FFs yield simplest design procedure

Best literal count

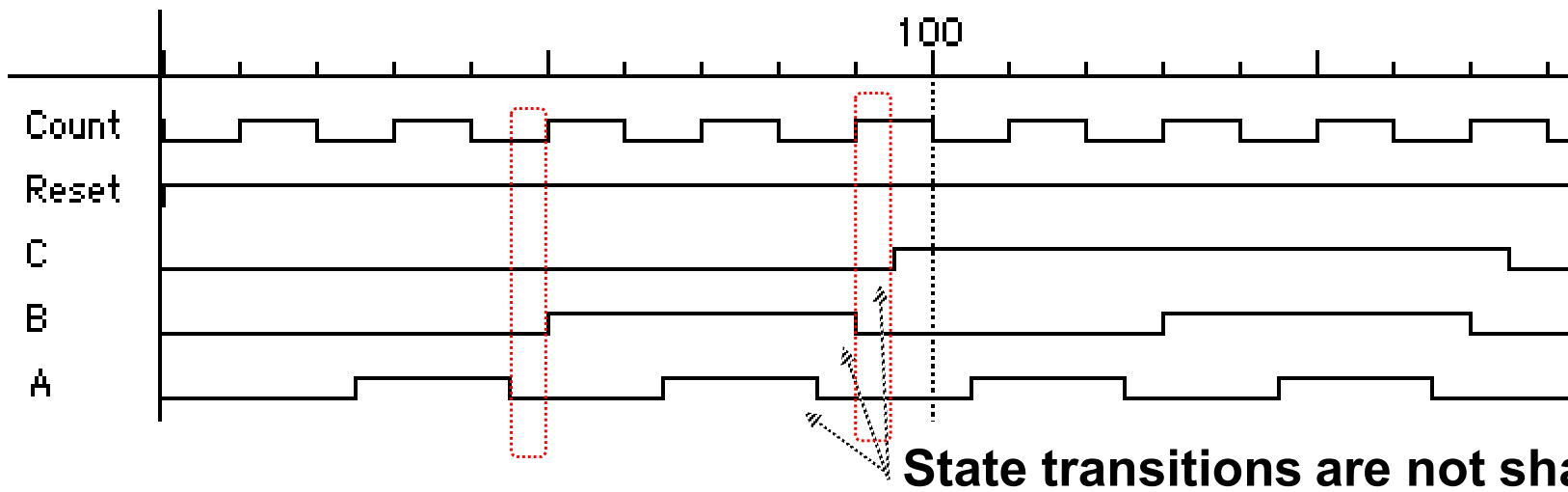
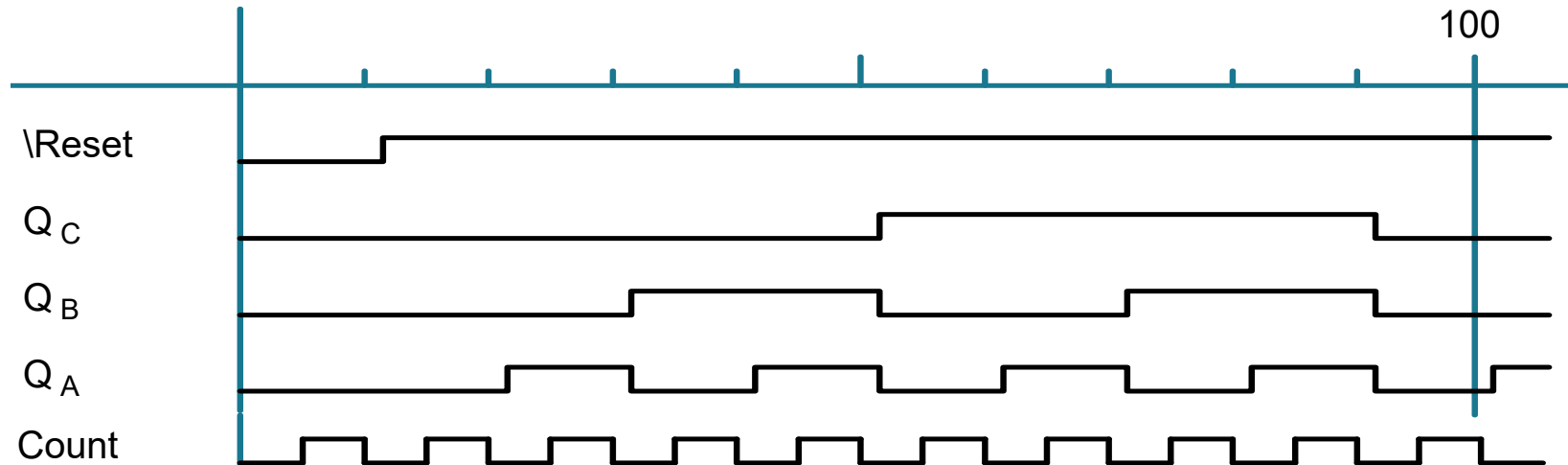
D storage devices very transistor efficient in VLSI

Best choice where area/literal count is the key

Asynchronous vs. Synchronous Counters

Ripple Counters

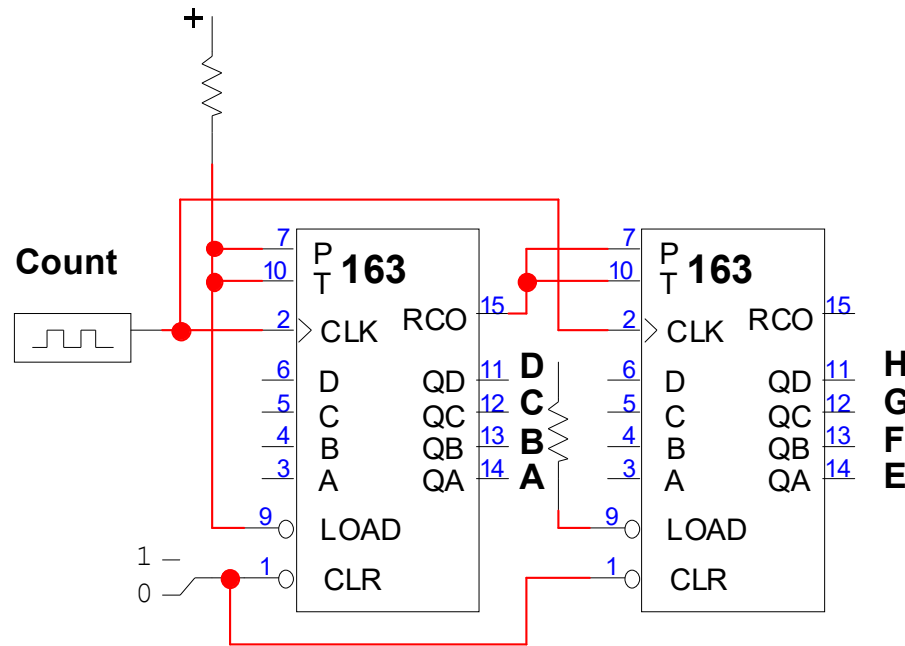
Deceptively attractive alternative to synchronous design style



Can lead to "spiked outputs" from combinational logic
decoding the counter's state

Asynchronous vs. Synchronous Counters

Cascaded Synchronous Counters with Ripple Carry Outputs



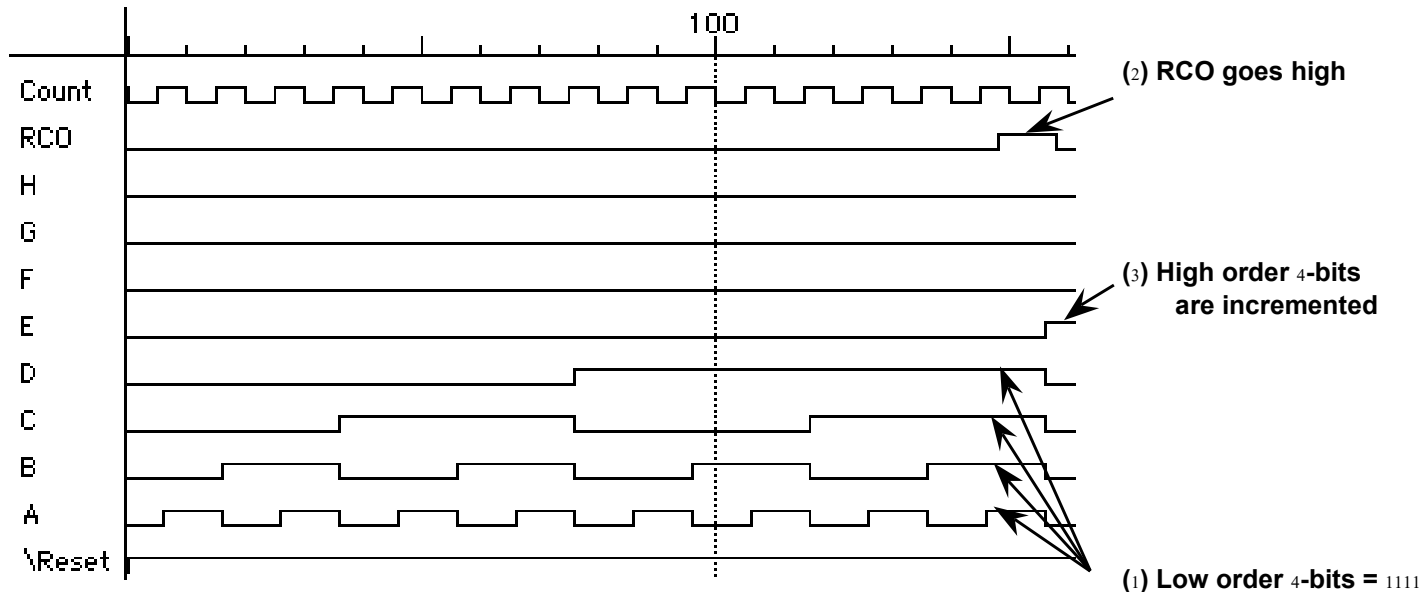
***First stage RCO
enables second stage
for counting***

**RCO asserted
soon after stage
enters state 1111**

**also a function
of the T Enable**

**Downstream stages
lag in their 1111 to
0000 transitions**

**Affects Count period
and decoding logic**

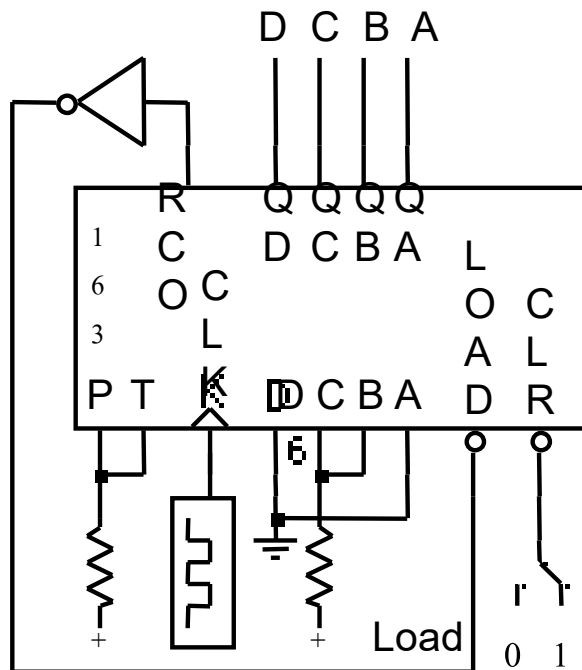


Asynchronous vs. Synchronous Counters

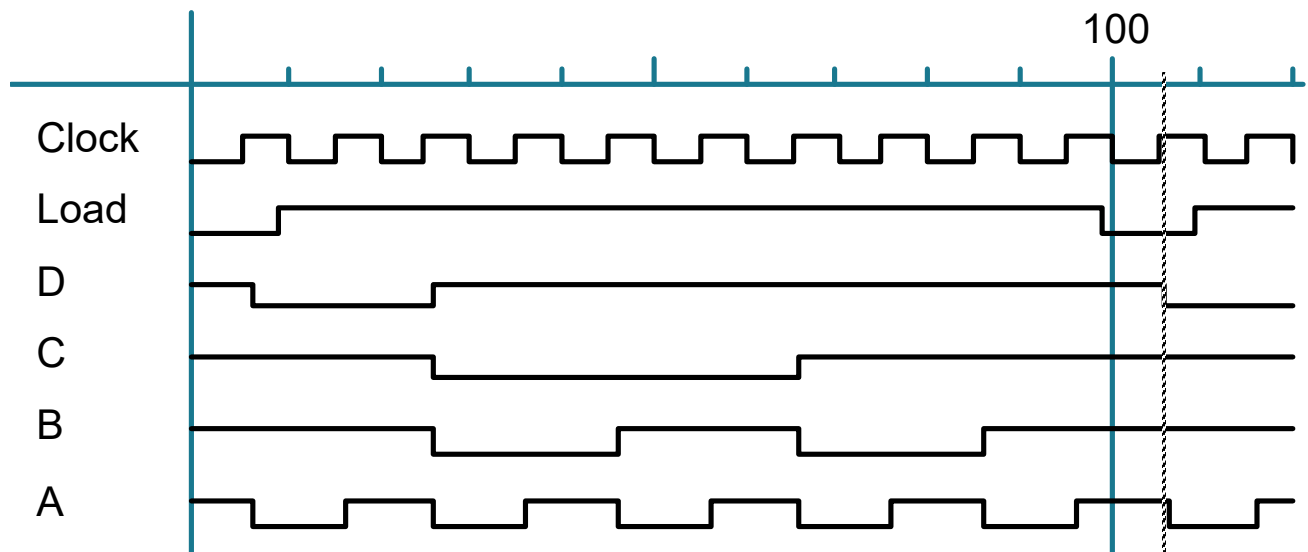
The Power of Synchronous Clear and Load

Starting Offset Counters:

e.g., 0110, 0111, 1000, 1001, 1010, 1011, 1100, 1101, 1111, 0110, ...



0110
*is the state
to be loaded*



Use RCO signal to trigger Load of a new state

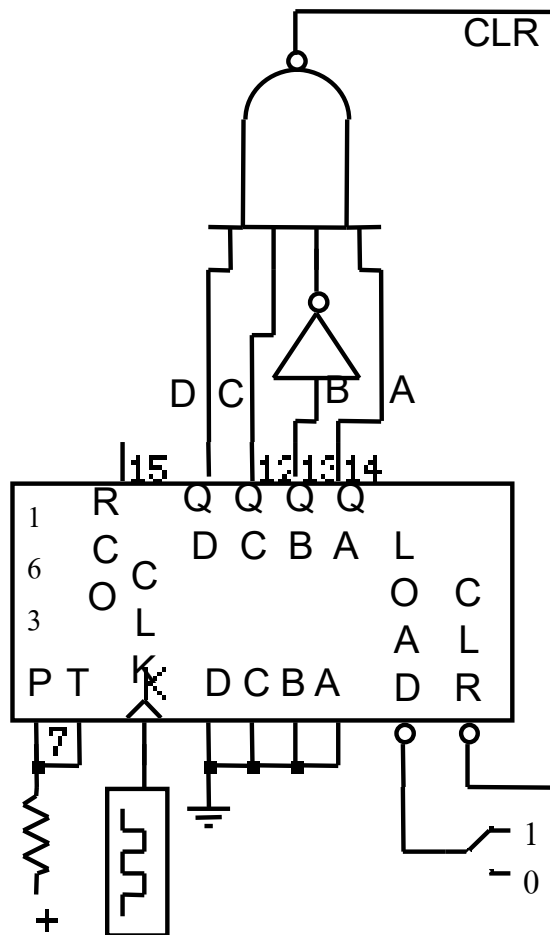
Since 74163 Load is synchronous, state changes only on the next rising clock edge

Asynchronous vs. Synchronous Counters

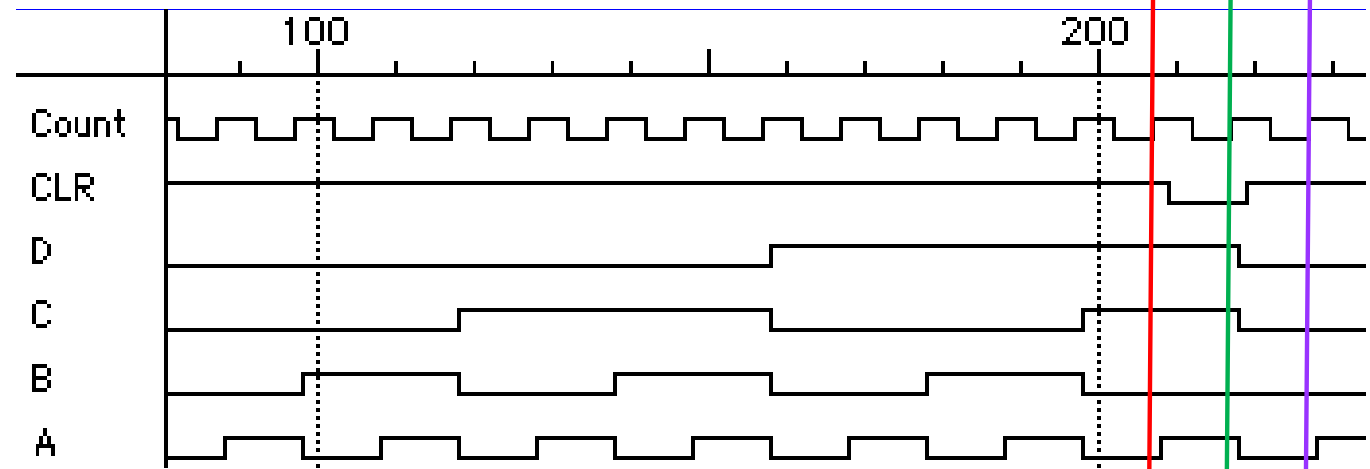
Offset Counters Continued

Ending Offset Counter:

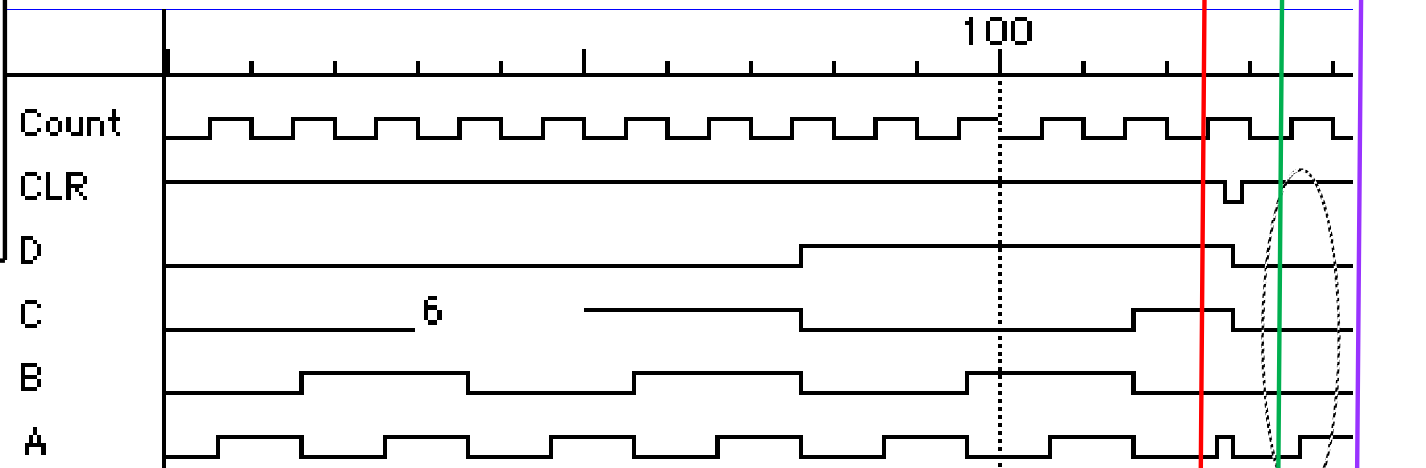
e.g., 0000, 0001, 0010, ..., 1100, 1101, 0000



**Decode state to
determine when to
reset to 0000**



Clear signal takes effect on the rising count edge



**Replace '163 with '161, Counter with Async Clear
Clear takes effect immediately!**

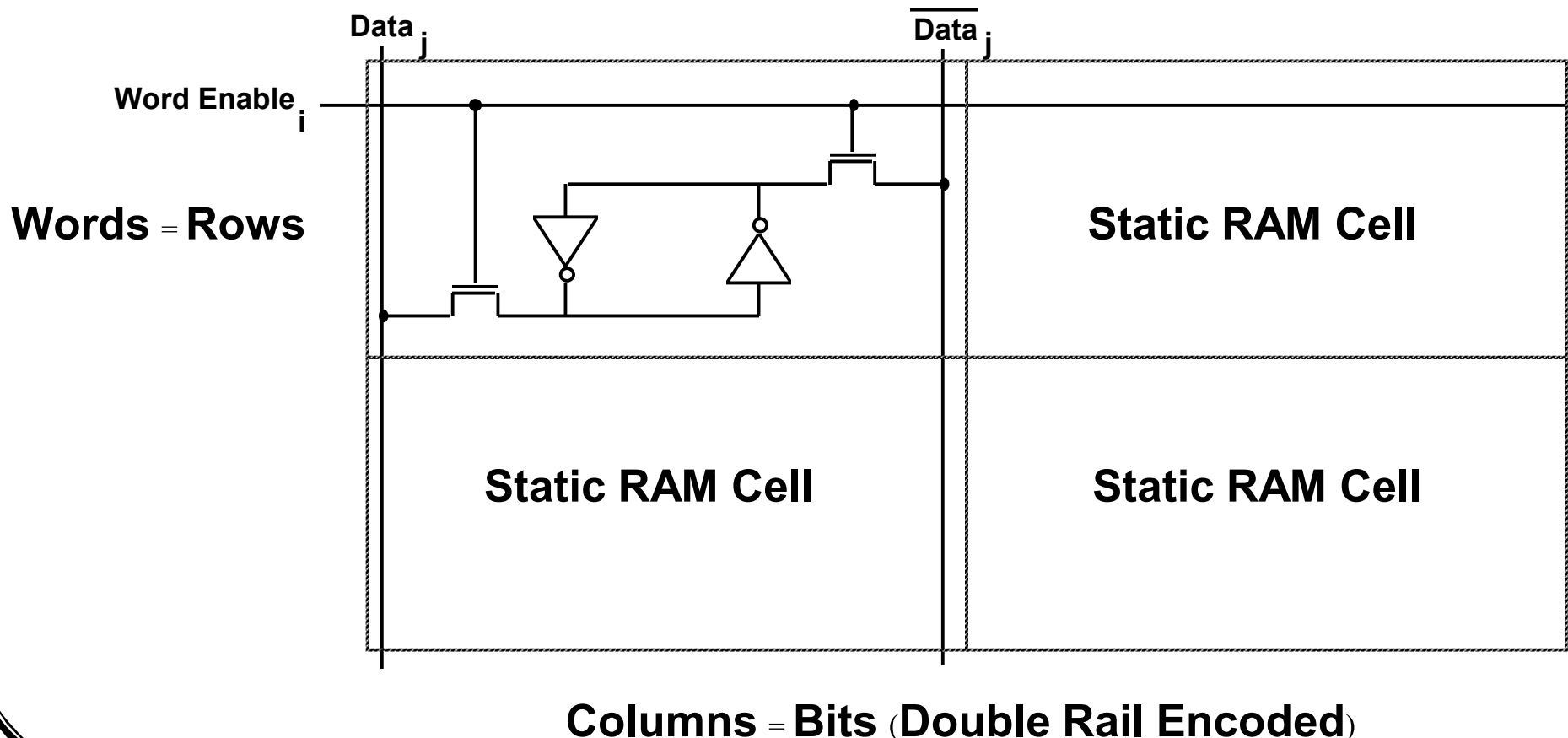
Static RAM

Transistor efficient methods for implementing storage elements

Small RAM: 256 words by 4-bit

Large RAM: 4 million words by 1-bit

We will discuss a 1024 x 4 organization



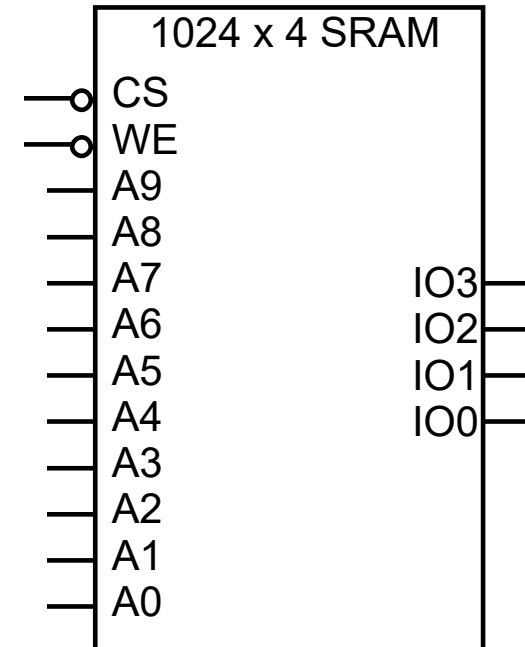
Static RAM Organization

Chip Select Line (active lo)

Write Enable Line (active lo)

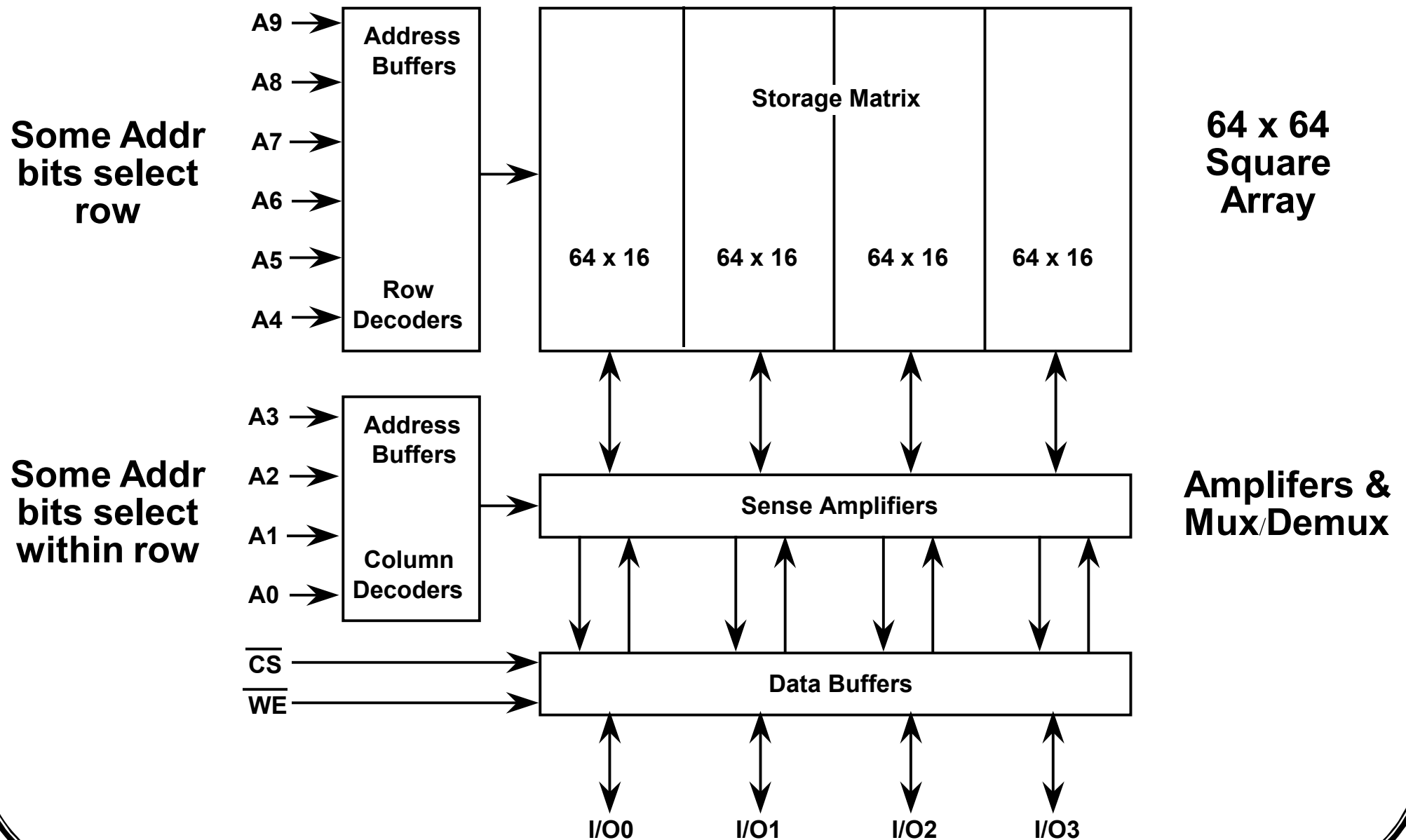
10 Address Lines

4 Bidirectional Data Lines



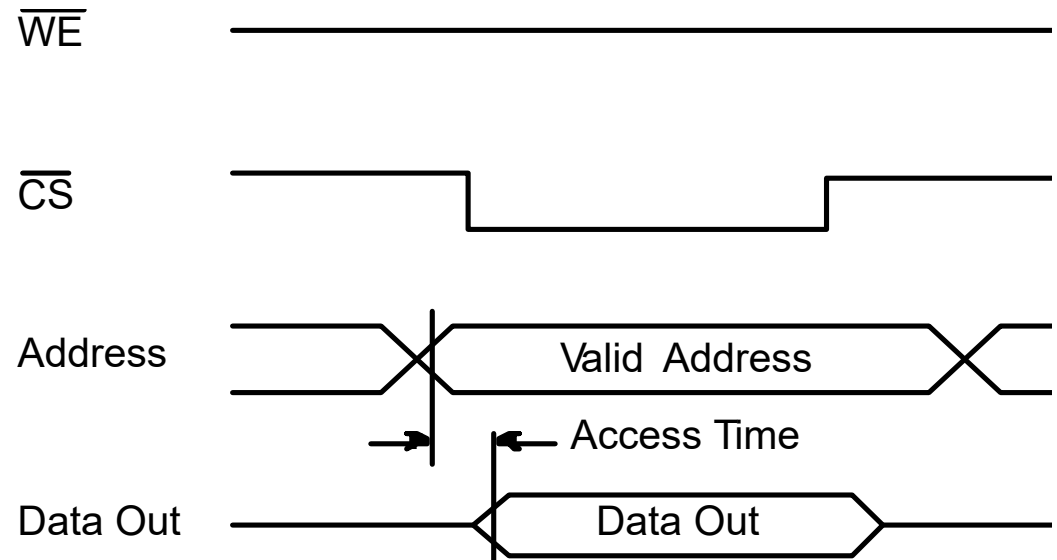
RAM Organization

Long thin layouts are not the best organization for a RAM

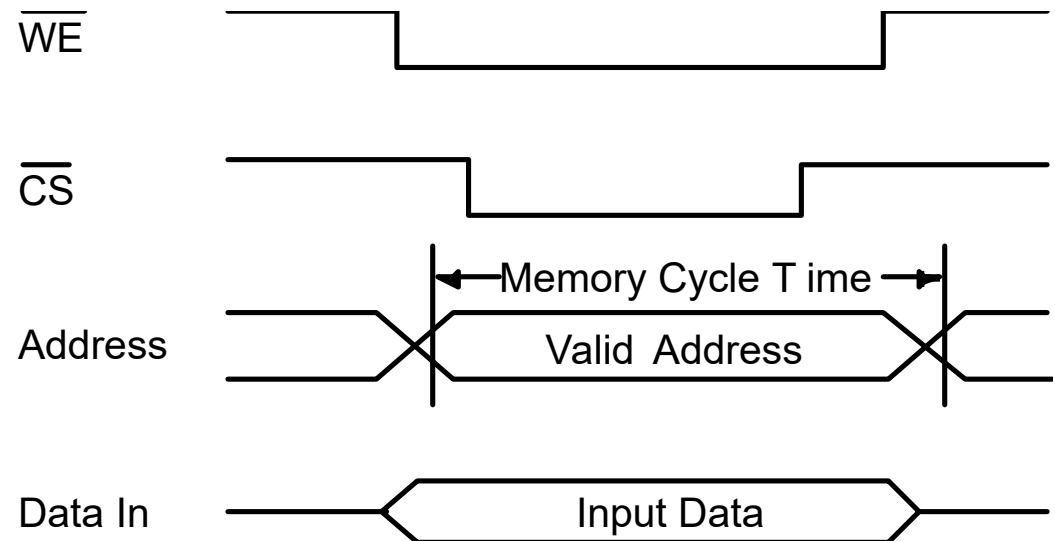


RAM Timing

Simplified Read Timing

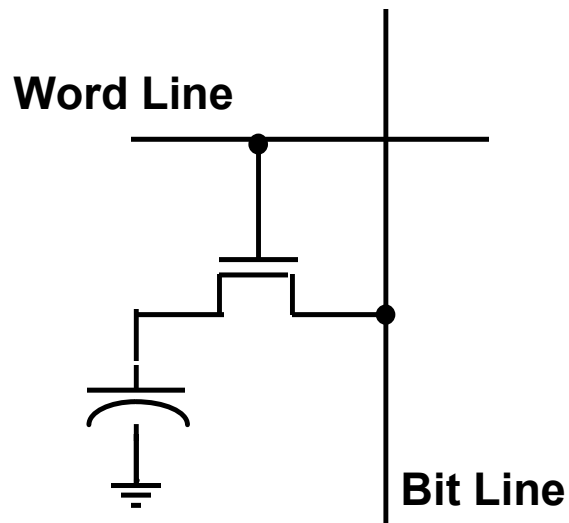


Simplified Write Timing



Random Access Memories

Dynamic RAMs



1 Transistor (+ capacitor) memory element

Read: Assert Word Line, Sense Bit Line

Write: Drive Bit Line, Assert Word Line

Destructive Read-Out

Need for Refresh Cycles: storage decay in ms

Internal circuits read word and write back

Random Access Memories

DRAM Organization

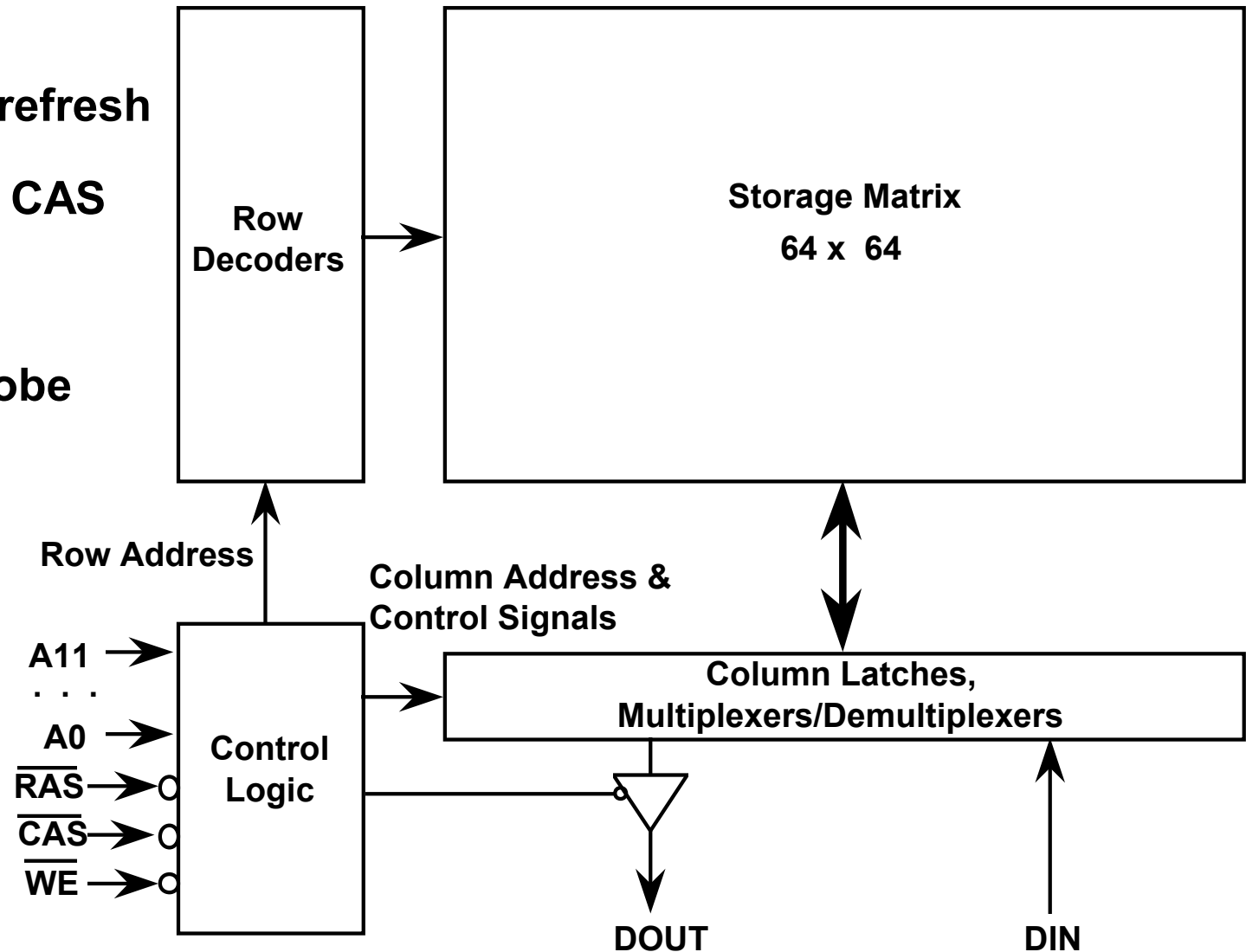
Long rows to simplify refresh

Two new signals: RAS, CAS

Row Address Strobe

Column Address Strobe

replace Chip Select



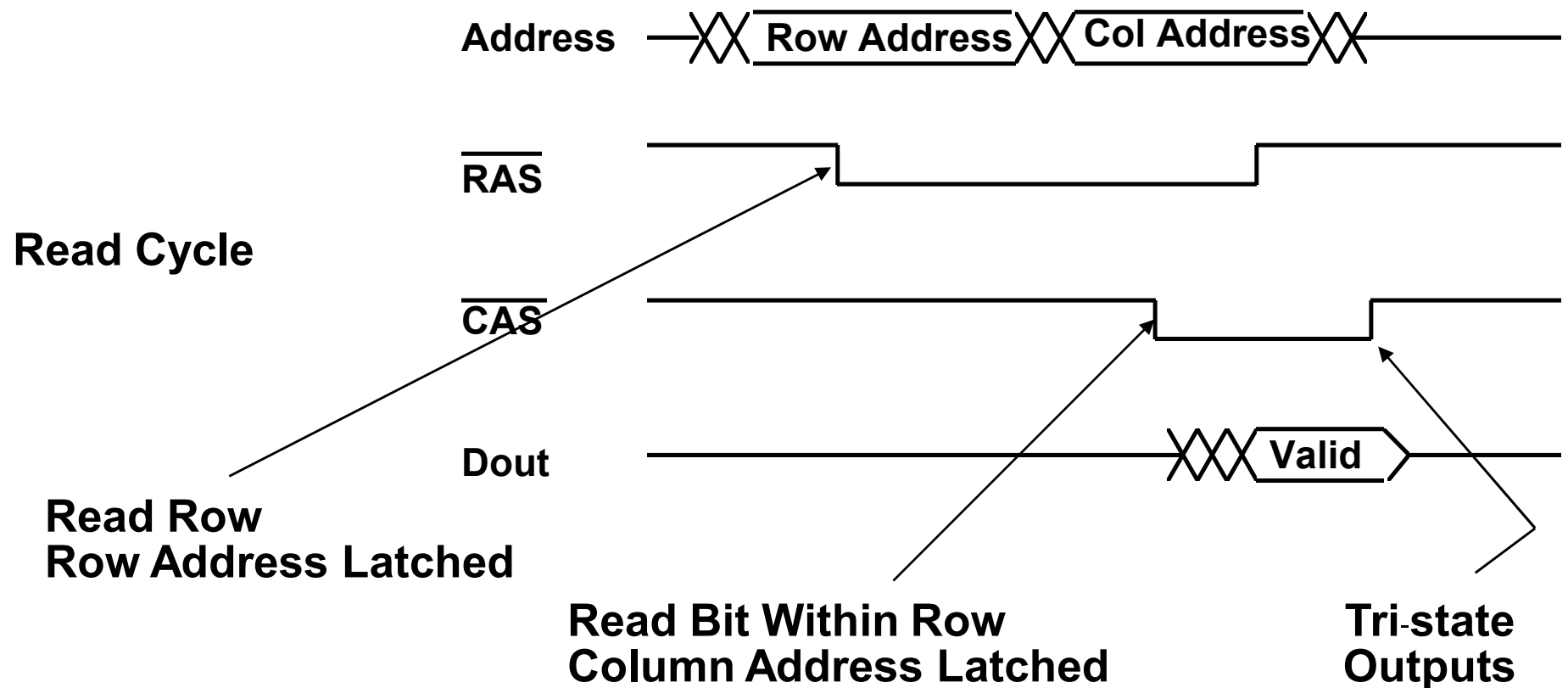
4096 x 1 bit DRAM

Random Access Memory

RAS, CAS Addressing

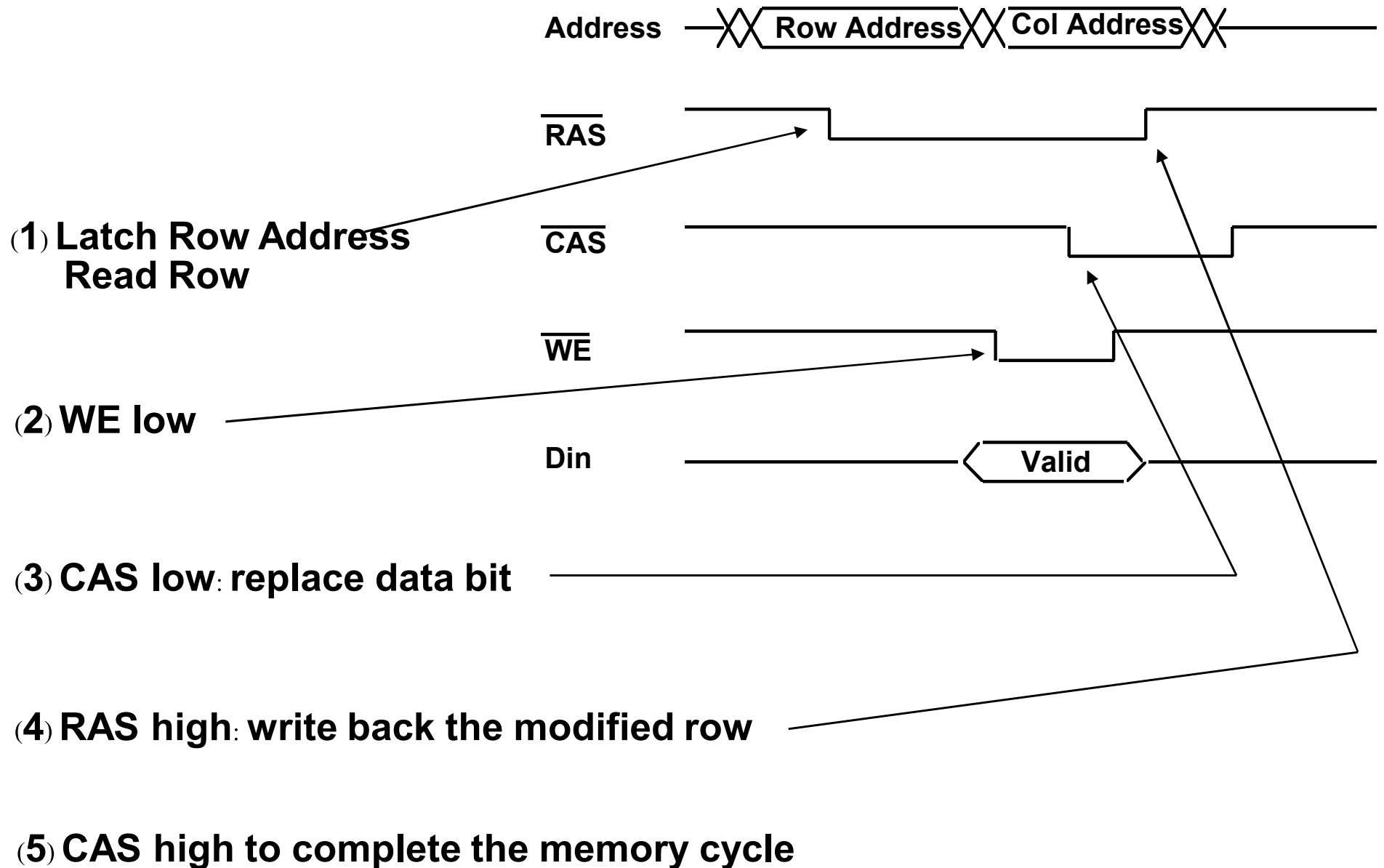
Even to read 1 bit, an entire 64-bit row is read!

Separate addressing into two cycles: Row Address, Column Address
Saves on package pins, speeds RAM access for sequential bits!



Random Access Memory

Write Cycle Timing



Random Access Memory

RAM Refresh

Refresh Frequency:

4096 word RAM -- refresh each word once every 4 ms

Assume 120ns memory access cycle

This is one refresh cycle every 976 ns (1 in 8 DRAM accesses)!

But RAM is really organized into 64 rows

This is one refresh cycle every 62.5 μ s (1 in 500 DRAM accesses)

Large capacity DRAMs have 256 rows, refresh once every 16 μ s

RAS-only Refresh (RAS cycling, no CAS cycling)

External controller remembers last refreshed row

Some memory chips maintain refresh row pointer

CAS before RAS refresh: if CAS goes low before RAS, then refresh

Random Access Memory

DRAM Variations

Page Mode DRAM:

read/write bit within last accessed row without RAS cycle

RAS, CAS, CAS, ..., CAS, RAS, CAS, ...

New column address for each CAS cycle

Static Column DRAM:

like page mode, except address bit changes signal new cycles rather than CAS cycling

on writes, deselect chip or CAS while address lines are changing

Nibble Mode DRAM:

like page mode, except that CAS cycling implies next column address in sequence -- no need to specify column address after first CAS

Works for 4 bits at a time (hence "nibble")

RAS, CAS, CAS, CAS, CAS, RAS, CAS, CAS, CAS, ...

Chapter Summary

- **The Variety of Sequential Circuit Packages**
Registers, Shifters, Counters, RAMs
- **Counters as Simple Finite State Machines**
- **Counter Design Procedure**
 1. **Derive State Diagram**
 2. **Derive State Transition Table**
 3. **Determine Next State Functions**
 4. **Remap Next State Functions for Target FF Types**
Using Excitation Tables; Implement Logic
- **Different FF Types in Counters**
J-K best for reducing gate count in packaged logic
D is easiest design plus best for reducing wiring and area in VLSI
- **Asynchronous vs. Synchronous Counters**
Avoid Ripple Counters! State transitions are not sharp
Beware of potential problems when cascading synchronous counters

Offset counters: easy to design with synchronous load and clear
Never use counters with asynchronous clear for this kind of application