Basics of Systems and Control Theory for pyMOR

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pyMOR Online Course 2020



$$y(t) = \begin{array}{c} \hat{x}(t) + B \\ \hat{y}(t) = \begin{array}{c} C \\ \hat{x}(t) + D \\ \hat{y}(t) = \end{array} \begin{array}{c} \hat{x}(t) + D \\ \hat{x}(t) + D \\ \hat{y}(t) = \begin{array}{c} \hat{x} \\ \hat{x}(t) + D \\ \hat{x}(t) + D \\ \hat{x}(t) + D \\ \hat{y}(t) = \begin{array}{c} \hat{x} \\ \hat{x}(t) + D \\ \hat{x}(t) + D \\ \hat{x}(t) + D \\ \hat{x}(t) + D \\ \hat{y}(t) = \begin{array}{c} \hat{x} \\ \hat{x}(t) + D \\ \hat{x}(t) + D$$

PYMOR Outline

- 1 Linear Time-Invariant (LTI) Systems
- 2 Transfer Function and Realizations
- 3 System Analysis
- 4 A Selection of MOR Methods

Only continuous-time systems

Discrete-time is treated in [1]

No differential-algebraic systems

For DAE aspects see [6, 3, 4, 5]

No non-linearities

No parameter dependencies

- 1 Linear Time-Invariant (LTI) Systems
 - Setting for this course
 - Examples
- 2 Transfer Function and Realizations
- 3 System Analysis
- 4 A Selection of MOR Methods

First-order State-space Systems

(PMOR: LTIModel)

$$\dot{\mathbf{x}}(t) = \mathbf{A}\mathbf{x}(t) + \mathbf{B}\mathbf{u}(t),$$

 $\mathbf{y}(t) = \mathbf{C}\mathbf{x}(t) + \mathbf{D}\mathbf{u}(t).$ (\(\Sigma\))

Here

- $\mathbf{x}(t) \in \mathbb{R}^n$ is called the state,
- $\mathbf{u}(t) \in \mathbb{R}^m$ is called the input,
- $\mathbf{y}(t) \in \mathbb{R}^p$ is called the output

of the LTI system. Correspondingly, we have

$$\mathbf{A} \in \mathbb{R}^{n \times n}$$
, $\mathbf{B} \in \mathbb{R}^{n \times m}$, $\mathbf{C} \in \mathbb{R}^{p \times n}$ and $\mathbf{D} \in \mathbb{R}^{p \times m}$.

We assume $t \in [0, \infty)$, $\mathbf{x}(0) = 0$.

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$$\mathbf{A} \in \mathbb{R}^{n \times n}, \quad \mathbf{B} \in \mathbb{R}^{n \times m}, \quad \mathbf{C} \in \mathbb{R}^{p \times n}.$$

We assume $t \in [0, \infty)$, $\mathbf{x}(0) = 0$.

First-order State-space Systems

(PMOR: LTIModel)

$$\mathbf{E}\dot{\mathbf{x}}(t) = \mathbf{A}\mathbf{x}(t) + \mathbf{B}\mathbf{u}(t),$$

 $\mathbf{y}(t) = \mathbf{C}\mathbf{x}(t).$ (\(\Sigma\)

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- $\mathbf{x}(t) \in \mathbb{R}^n$ is called the state,
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of the LTI system. Correspondingly, we have

$$\mathbf{E}, \mathbf{A} \in \mathbb{R}^{n \times n}, \quad \mathbf{B} \in \mathbb{R}^{n \times m}, \quad \mathbf{C} \in \mathbb{R}^{p \times n}.$$

We assume $t \in [0, \infty)$, $\mathbf{x}(0) = 0$ and \mathbf{E} invertible.

Second-order State-space Systems

(₱**MOR**: SecondOrderModel)

$$\begin{split} \mathbf{M}\ddot{\mathbf{x}}(t) + \mathbf{E}\dot{\mathbf{x}}(t) + \mathbf{K}\mathbf{x}(t) &= \mathbf{B}\mathbf{u}(t), \\ \mathbf{y}(t) &= \mathbf{C}_{\mathsf{v}}\dot{\mathbf{x}}(t) + \mathbf{C}_{\mathsf{p}}\mathbf{x}(t). \end{split}$$

Here

- $\mathbf{x}(t) \in \mathbb{R}^n$ is called the position,
- $\dot{\mathbf{x}}(t) \in \mathbb{R}^n$ is called the velocity,
- $\mathbf{u}(t) \in \mathbb{R}^m$ is called the input,
- $\mathbf{y}(t) \in \mathbb{R}^p$ is called the output

of the LTI system. Correspondingly, we have

$$\mathbf{M}, \mathbf{E}, \mathbf{K} \in \mathbb{R}^{n \times n}, \quad \mathbf{B} \in \mathbb{R}^{n \times m}, \quad \mathbf{C}_{v}, \mathbf{C}_{p} \in \mathbb{R}^{p \times n}.$$

Heat Equation [MORWiki thermal block] I

For $t \in (0,T)$, $\xi \in \Omega$ and initial values

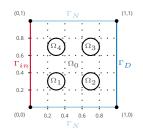
$$\theta(0,\xi) = 0$$
, for $\xi \in \Omega$,

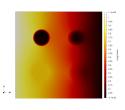
consider

$$\partial_t \theta(t,\xi) + \nabla \cdot (-\sigma(\xi)\nabla \theta(t,\xi)) = 0,$$

with boundary conditions

$$\begin{split} \sigma(\xi)\nabla\theta(t,\xi)\cdot n(\xi) &= u(t) &\quad t\in(0,T), \xi\in\Gamma_{in},\\ \sigma(\xi)\nabla\theta(t,\xi)\cdot n(\xi) &= 0 &\quad t\in(0,T), \xi\in\Gamma_{N},\\ \theta(t,\xi) &= 0 &\quad t\in(0,T), \xi\in\Gamma_{D}. \end{split}$$





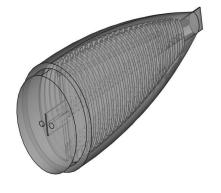
Heat Equation [MORWiki thermal block] II

Finite element semi-discretization in space

- ullet pairwise inner products of ansatz functions $\leadsto E$
- ullet discretized spatial operator + Dirichlet boundary condition $\leadsto {f A}$
- ullet discretized non-zero Neumann boundary condition $\leadsto B$
- $\bullet\;$ average temperatures on the inclusions $\leadsto C$
- n = 7488
- m = 1
- p = 4

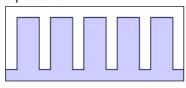
An Artificial Fishtail [MORWiki Artificial Fishtail] I

Construction:



Fluid Elastomer Actuation:

no pressure



under pressure



An Artificial Fishtail [MORWiki Artificial Fishtail] II

Variables:

- displacement $\vec{s}(t, \vec{z})$
- strain $\underline{\vec{\varepsilon}}(\vec{s}(t,\vec{z}))$
- stress $\underline{\vec{\sigma}}(\vec{s}(t, \vec{z}))$

Material parameters:

- density ρ
- Lamé parameters λ , μ

Basic principle:

$$\begin{split} \underline{\vec{\varepsilon}}(\vec{s}(t,\vec{z})) &= \frac{1}{2} \left(\nabla \vec{s}(t,\vec{z}) + \nabla^\mathsf{T} \vec{s}(t,\vec{z}) \right) & \text{(kinematic equation)} \\ \underline{\vec{\sigma}}(\vec{s}(t,\vec{z})) &= \lambda \operatorname{tr}((\underline{\vec{\varepsilon}}(\vec{s}(t,\vec{z}))) \, \underline{I} + 2\mu \underline{\vec{\varepsilon}}(\vec{s}(t,\vec{z}))) & \text{(material equation)} \\ \rho \frac{\partial^2 \vec{s}(t,\vec{z})}{\partial t^2} &= \nabla \cdot \underline{\vec{\sigma}}(\vec{s}(t,\vec{z})) + \vec{f}(t,\vec{z}) & \text{(equation of motion)} \end{split}$$

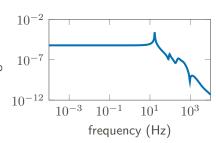
+ initial and boundary conditions

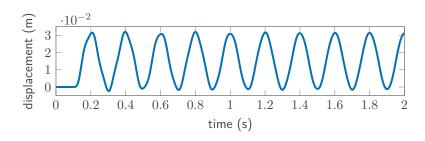
An Artificial Fishtail [MORWiki Artificial Fishtail] III

FEM semi-discretization:

$$\begin{aligned} \mathbf{M}\ddot{\mathbf{x}}(t) + \mathbf{E}\dot{\mathbf{x}}(t) + \mathbf{K}\mathbf{x}(t) &= \mathbf{B}\mathbf{u}(t), & \text{product}\\ \mathbf{y}(t) &= \mathbf{C}_{\mathrm{p}}\mathbf{x}(t), & \text{product}\\ \mathbf{y}(t) &= \mathbf{K}_{\mathrm{p}}\mathbf{x}(t), & \text{product}\\ \mathbf{y}(t) &= \mathbf{K}_{\mathrm{p}}\mathbf{x}(t),$$

- $M, E, K > 0, C_v = 0$
- n = 779232, m = 1, p = 3.





- 1 Linear Time-Invariant (LTI) Systems
- 2 Transfer Function and Realizations
 - Laplace Transform
 - Transfer Function
 - Realizations
 - Projection-based MOR
- 3 System Analysis
- 4 A Selection of MOR Methods

Definition

Let $f:[0,\infty)\to\mathbb{R}^n$ be exponentially bounded with bounding exponent α . Then

$$\mathcal{L}\left\{f\right\}(s) := \int_{0}^{\infty} f(\tau) e^{-s\tau} d\tau$$

for $\mathrm{Re}(s)>\alpha$ is called the **Laplace transform** of f. The process of forming the Laplace transform is called **Laplace transformation**.

It can be shown that the integral converges uniformly in a domain with $\mathrm{Re}(s) \geq \beta$ for all $\beta > \alpha$.

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Allows us to map time signals to frequency signals.

Theorem

Let $f, g, h: [0, \infty) \to \mathbb{R}^n$ be given. Then the following two statements hold true:

a) The Laplace transformation is linear, i. e., if f and g are exponentially bounded, then $h:=\gamma f+\delta g$ is also exponentially bounded and

$$\mathcal{L}\left\{h\right\} = \gamma \mathcal{L}\left\{f\right\} + \delta \mathcal{L}\left\{g\right\}$$

holds for all $\gamma, \delta \in \mathbb{C}$.

b) If $f\in \mathcal{PC}^1([0,\infty),\mathbb{R}^n)$ and \dot{f} is exponentially bounded, then f is exponentially bounded and

$$\mathcal{L}\{\dot{f}\}(s) = s\mathcal{L}\{f\}(s) - f(0).$$

PYMOR Transfer Function and Realizations | Laplace Transform

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$$\mathcal{L}\{\dot{f}\}(s) = s\mathcal{L}\{f\}(s) - f(0).$$

- $X(s) := \mathcal{L}\{\mathbf{x}\}(s), U(s) := \mathcal{L}\{\mathbf{u}\}(s), \text{ and } Y(s) := \mathcal{L}\{\mathbf{y}\}(s)$
- $\mathbf{A}\mathbf{x}(t) + \mathbf{B}\mathbf{u}(t) \rightsquigarrow \mathbf{A}X(s) + \mathbf{B}U(s)$
- $\mathbf{y}(t) = \mathbf{C}\mathbf{x}(t) \rightsquigarrow Y(s) = \mathbf{C}X(s)$
- $sX(s) := \mathcal{L}\{\dot{x}\}(s)$ since $\mathbf{x}(0) = 0$

Rational Matrix Function Representation

In summary we have:

- $s\mathbf{E}X(s) = \mathbf{A}X(s) + \mathbf{B}U(s)$
- $Y(s) = \mathbf{C}X(s)$

Thus the mapping from inputs to outputs in frequency domain can be expressed as

$$\mathbf{H}(s) = \mathbf{C}(s\mathbf{E} - \mathbf{A})^{-1}\mathbf{B}.$$

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$$\mathbf{H}(s) = \mathbf{C}(s\mathbf{E} - \mathbf{A})^{-1}\mathbf{B}.$$

Analogously, for second-order systems we get

$$\mathbf{H}(s) = (s\mathbf{C}_{\mathsf{v}} + \mathbf{C}_{\mathsf{p}}) \left(s^2 \mathbf{M} + s\mathbf{E} + \mathbf{K} \right)^{-1} \mathbf{B}.$$

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 \mathbf{H} is analytic in $\mathbb{C} \setminus \Lambda(\mathbf{E}, \mathbf{A})$, or $\mathbb{C} \setminus \Lambda(\mathbf{M}, \mathbf{E}, \mathbf{K})$, respectively

Important Representations of $\hat{H}(s)$

(Laurent) series expansion

$$\mathbf{H}(s) = \sum_{k=0}^{\infty} (s - s_0)^k M_k(s_0) \quad \mathbf{H}(s) = \sum_{k=0}^{\infty} s^{-k} M_k(\infty)$$

The matrices $M_k(s_0)$ are called **moments** of ${\bf H}$. At infinity they are also referred to as **Markov parameters**.

Pole Residue Form

Let (λ_i, w_i, v_i) be the eigentriplets of the pair (\mathbf{A}, \mathbf{E}) with no degenerate eigenspaces. Then we have

$$\mathbf{H}(s) = \sum_{i=1}^{n} \frac{R_i}{s - \lambda_i},$$

where $R_i = (\mathbf{C}v_i)(w_i^{\mathsf{H}}\mathcal{B})$, assuming $w_i^{\mathsf{H}}v_i = 1$.

The representation of H using (E,A,B,C) is not unique.

In fact for any invertible matrix $\mathbf{T} \in \mathbb{R}^{n \times n}$, we have

$$\mathbf{H}(s) = \mathbf{C}(s\mathbf{E} - \mathbf{A})^{-1}\mathbf{B}$$
$$= \mathbf{C}\mathbf{T}^{-1}\mathbf{T}(s\mathbf{E} - \mathbf{A})^{-1}\mathbf{T}^{-1}\mathbf{T}\mathbf{B}$$
$$= \mathbf{C}\mathbf{T}^{-1}(s\mathbf{T}\mathbf{E}\mathbf{T}^{-1} - \mathbf{T}\mathbf{A}\mathbf{T}^{-1})^{-1}\mathbf{T}\mathbf{B}$$

and thus a system given, by $(TET^{-1}, TAT^{-1}, TB, CT^{-1})$ realizes the exact same input/output behavior.

Definition

- ullet All sets of matrices leading to the same function ${f H}$ are called its **realizations**.
- ullet The matrix ${f T}$ above is called **state-space transformation**.

Important Realizations

Minimal Realizations

Can we realize ${f H}$ with less equations?

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Can we realize \mathbf{H} with less equations?

Truncated Realizations

Can we introduce a small error to get even less equations?

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Balanced Realizations



Can we find state coordinates that allow us to decide what is important?

McMillan Degree and Minimal Realization

Example

Realizations can even be of different dimensions. Take for example:

$$\mathbf{E}=\mathbf{I}$$
 the identity, $\mathbf{A}=egin{bmatrix} -11 & 0 \ 0 & -5 \end{bmatrix}$, $\mathbf{B}=egin{bmatrix} 1 \ 1 \end{bmatrix}$ and $\mathbf{C}=egin{bmatrix} 1 & 0 \end{bmatrix}$.

Truncating the second state component does not change H.

Definition

There exists a minimum number of equations necessary to describe \mathbf{H} . The state dimension n of this minimal set of equations is called **McMillan degree** of the system. A realization of \mathbf{H} with this dimension is called **minimal realization**.

$$\mathbf{E}\dot{\mathbf{x}}(t) - \mathbf{A}\mathbf{x}(t) - \mathbf{B}\mathbf{u}(t) = 0,$$

$$\mathbf{y}(t) - \mathbf{C}\mathbf{x}(t) - \mathbf{D}\mathbf{u}(t) = 0.$$

$$\begin{split} \mathbf{E}\mathbf{V}\dot{\hat{\mathbf{x}}}(t) - \mathbf{A}\mathbf{V}\hat{\mathbf{x}}(t) - \mathbf{B}\mathbf{u}(t) &= e_{\mathsf{res}}(t), \\ \mathbf{y}(t) - \mathbf{C}\mathbf{V}\hat{\mathbf{x}}(t) - \mathbf{D}\mathbf{u}(t) &= e_{\mathsf{output}}(t). \end{split}$$

Step I: Use truncated state transformation

Replace

$$\mathbf{x}(t) \approx \mathbf{V}\hat{\mathbf{x}}(t)$$

with $\mathbf{V} \in \mathbb{R}^{n \times r}$ and $\hat{\mathbf{x}}(t) \in \mathbb{R}^r$.

$$\begin{split} \mathbf{V}^\mathsf{T} \mathbf{E} \mathbf{V} \dot{\hat{\mathbf{x}}}(t) - \mathbf{V}^\mathsf{T} \mathbf{A} \mathbf{V} \hat{\mathbf{x}}(t) - \mathbf{V}^\mathsf{T} \mathbf{B} \mathbf{u}(t) &= 0, \\ \mathbf{y}(t) - \mathbf{C} \mathbf{V} \hat{\mathbf{x}}(t) - \mathbf{D} \mathbf{u}(t) &= e_\mathsf{output}(t). \end{split}$$

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Step II: Mitigate transformation error

Suppress truncation residual through left projection.

ullet one-sided method: use ${f V}$ again.

$$\begin{split} \mathbf{W}^\mathsf{T} \mathbf{E} \mathbf{V} \dot{\hat{\mathbf{x}}}(t) - \mathbf{W}^\mathsf{T} \mathbf{A} \mathbf{V} \hat{\mathbf{x}}(t) - \mathbf{W}^\mathsf{T} \mathbf{B} \mathbf{u}(t) &= 0, \\ \mathbf{y}(t) - \mathbf{C} \mathbf{V} \hat{\mathbf{x}}(t) - \mathbf{D} \mathbf{u}(t) &= e_\mathsf{output}(t). \end{split}$$

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Step II: Mitigate transformation error

Suppress truncation residual through left projection.

- ullet one-sided method: use ${f V}$ again.
- two-sided method: find $\mathbf{W} \in \mathbb{R}^{n \times r}$.

 $\hat{\mathbf{A}} pprox \mathbf{A}$

Reduced order model (ROM)

(PV**♦O**R: LTIPGReductor)

Define
$$\hat{\mathbf{E}} = \mathbf{W}^\mathsf{T} \mathbf{E} \mathbf{V}$$
, $\hat{\mathbf{A}} = \mathbf{W}^\mathsf{T} \mathbf{A} \mathbf{V} \in \mathbb{R}^{r \times r}$, $\hat{\mathbf{B}} = \mathbf{W}^\mathsf{T} \mathbf{B} \in \mathbb{R}^{r \times m}$ and $\hat{\mathbf{C}} = \mathbf{C} \mathbf{V} \in \mathbb{R}^{p \times r}$. Then

$$\hat{\mathbf{E}}\dot{\hat{\mathbf{x}}}(t) = \hat{\mathbf{A}}\hat{\mathbf{x}}(t) + \hat{\mathbf{B}}\mathbf{u}(t),$$

$$\hat{\mathbf{y}}(t) = \hat{\mathbf{C}}\hat{\mathbf{x}}(t) + \mathbf{D}\mathbf{u}(t)$$
(ROM)

approximates the dynamics of the full-order model (Σ) with output error

$$\mathbf{y}(t) - \hat{\mathbf{y}}(t) = e_{\text{output}}(t).$$

- \bullet We call the corresponding transfer function $\hat{H}.$
- Model order reduction (MOR) \leadsto Find $\mathbf{W}, \mathbf{V} \in \mathbb{R}^{n \times r}$ such that $e_{\mathsf{output}}(t)$ is small in a suitable sense.
- We will see energy-based and interpolation-based methods today and tomorrow.

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 - System Norms and Hardy Spaces
 - Frequency-Domain Analysis
- 4 A Selection of MOR Methods

We have

$$Y(s) = \mathbf{H}(s)U(s)$$

and

$$\hat{Y}(s) = \hat{\mathbf{H}}(s)U(s).$$

Question

What are suitable norms such that

$$||y - \hat{y}|| \le ||\mathbf{H} - \hat{\mathbf{H}}|| ||u||?$$

The Banach Space $\mathcal{H}_{\infty}^{p imes m}$

$$\mathcal{H}_{\infty}^{p\times m}:=\left\{G:\mathbb{C}^{+}\rightarrow\mathbb{C}^{p\times m}\ :\ G\ \text{is analytic in}\ \mathbb{C}^{+}\ \text{and}\ \sup_{s\in\mathbb{C}^{+}}\left\Vert G(s)\right\Vert _{2}<\infty\right\}.$$

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 $\mathcal{H}_{\infty}^{p imes m}$ is a Banach space equipped with the \mathcal{H}_{∞} -norm

$$||G||_{\mathcal{H}_{\infty}} := \sup_{\omega \in \mathbb{R}} ||G(i\omega)||_2.$$

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Can show:
$$\|\mathbf{y} - \hat{\mathbf{y}}\|_{\mathcal{L}_2} \leq \left\|\mathbf{H} - \hat{\mathbf{H}}\right\|_{\mathcal{H}_{\infty}} \|\mathbf{u}\|_{\mathcal{L}_2}.$$

This bound can even be shown to be sharp.

The Hilbert Space $\mathcal{H}_2^{p imes m}$

$$\mathcal{H}_2^{p imes m}:=\left\{G:\mathbb{C}^+ o\mathbb{C}^{p imes m}\ :\ G ext{ is analytic in }\mathbb{C}^+ ext{ and }
ight.$$

$$\sup_{\xi>0} \int_{-\infty}^{\infty} \|G(\xi + i\omega)\|_{F}^{2} d\omega < \infty \right\}.$$

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 $\mathcal{H}_2^{p \times m}$ is a Hilbert space with the inner product

$$\langle F, G \rangle_{\mathcal{H}_2} := \frac{1}{2\pi} \int_{-\infty}^{\infty} \operatorname{tr} \left(F(i\omega)^{\mathsf{H}} G(i\omega) \right) d\omega$$

and induced norm

$$||G||_{\mathcal{H}_2} := \langle G, G \rangle_{\mathcal{H}_2}^{1/2} = \left(\frac{1}{2\pi} \int_{-\infty}^{\infty} ||G(i\omega)||_F^2 d\omega\right)^{1/2}.$$

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Can show:
$$\|\mathbf{y} - \hat{\mathbf{y}}\|_{\mathcal{L}_{\infty}} \leq \left\|\mathbf{H} - \hat{\mathbf{H}}\right\|_{\mathcal{H}_{2}} \|\mathbf{u}\|_{\mathcal{L}_{2}}.$$

System Gramians and \mathcal{H}_2 -trace-formula

A system (Σ) with $\Lambda(\mathbf{E}, \mathbf{A}) \subset \mathbb{C}^-$ is called **asymptotically stable**. Then, all state trajectories decay exponentially as $t \to \infty$ and

a) the infinite controllability and observability **Gramians** exist:

$$\mathbf{P} = \int_0^\infty e^{\mathbf{E}^{-1}\mathbf{A}t} \mathbf{E}^{-1} \mathbf{B} \mathbf{B}^\mathsf{T} \mathbf{E}^{-\mathsf{T}} e^{\mathbf{A}^\mathsf{T} \mathbf{E}^{-\mathsf{T}} t} dt$$
$$\mathbf{E}^\mathsf{T} \mathbf{Q} \mathbf{E} = \int_0^\infty e^{\mathbf{A}^\mathsf{T} \mathbf{E}^{-\mathsf{T}} t} \mathbf{C}^\mathsf{T} \mathbf{C} e^{\mathbf{E}^{-1} \mathbf{A} t} dt.$$

b) P, Q solve the two Lyapunov equations

$$APE^{\mathsf{T}} + EPA^{\mathsf{T}} = -BB^{\mathsf{T}}, \quad A^{\mathsf{T}}QE + E^{\mathsf{T}}QA = -C^{\mathsf{T}}C$$

c) the \mathcal{H}_2 -norm can be expressed as

$$\left\|\mathbf{H}\right\|_{\mathcal{H}_{2}}^{2}=\mathrm{tr}\big(\mathbf{C}\mathbf{P}\mathbf{C}^{\mathsf{T}}\big)=\mathrm{tr}\big(\mathbf{B}^{\mathsf{T}}\mathbf{Q}\mathbf{B}\big)\,.$$

Bode Plots

The Bode plot for **H** consists of a **magnitude plot** and a **phase plot**.

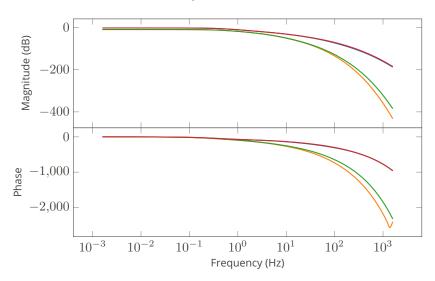
Bode magnitude plot

- component-wise graph of the function $|\mathbf{H}(\mathrm{i}\omega)|$ for frequencies $\omega \in [\omega_{\min}, \omega_{\max}] \subset \mathbb{R}.$
- ω -axis is logarithmic.
- magnitude is given in decibels, i.e., $|\mathbf{H}(i.)|$ is plotted as $20 \log_{10}(|\mathbf{H}(i.)|)$.

Bode phase plot

- component-wise graph of the function $\arg \mathbf{H}(\mathrm{i}\omega)$ for frequencies $\omega \in [\omega_{\min}, \omega_{\max}] \subset \mathbb{R}.$
- ω -axis is logarithmic.
- phase is given in degrees on a linear scale.

Bode Plot for the Thermal Block Example



(Sigma) Magnitude Plots

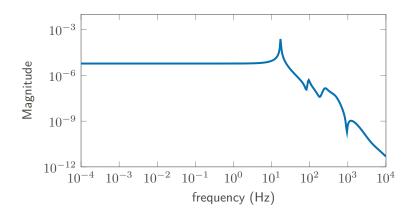
Sigma magnitude plot

- 2-norm-wise graph of the function $\mathbf{H}(\mathrm{i}\omega)$ for frequencies $\omega \in [\omega_{\min}, \omega_{\max}] \subset \mathbb{R}$.
- ω -axis is logarithmic.

The name is due to the fact that for a given matrix M the norm $\|M\|_2$ is given by its largest singular value.

The real sigma magnitude plot depicts all singular values as functions of ω .

Sigma Magnitude Plot for the Artificial Fishtail



- 1 Linear Time-Invariant (LTI) Systems
- 2 Transfer Function and Realizations
- System Analysis
- 4 A Selection of MOR Methods
 - Balancing Based MOR
 - Moments and Interpolation

Idea:

• The system (Σ), in realization ($\mathbf{E} = \mathbf{I}, \mathbf{A}, \mathbf{B}, \mathbf{C}$), is called balanced, if the solutions \mathbf{P}, \mathbf{Q} of the Lyapunov equations

$$\mathbf{AP} + \mathbf{PA}^\mathsf{T} + \mathbf{BB}^\mathsf{T} = 0, \quad \mathbf{A}^\mathsf{T}\mathbf{Q} + \mathbf{QA} + \mathbf{C}^\mathsf{T}\mathbf{C} = 0,$$

satisfy:
$$\mathbf{P} = \mathbf{Q} = \operatorname{diag}(\sigma_1, \dots, \sigma_n)$$
 where $\sigma_1 \geq \sigma_2 \geq \dots \geq \sigma_n > 0$.

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$$\begin{split} \boldsymbol{\mathcal{T}} : (\mathbf{I}, \mathbf{A}, \mathbf{B}, \mathbf{C}) &\mapsto (\mathbf{I}, \mathbf{T}\mathbf{A}\mathbf{T}^{-1}, \mathbf{T}\mathbf{B}, \mathbf{C}\mathbf{T}^{-1}) \\ &= \left(\left[\begin{array}{cc} \mathbf{A}_{11} & \mathbf{A}_{12} \\ \mathbf{A}_{21} & \mathbf{A}_{22} \end{array} \right], \left[\begin{array}{cc} \mathbf{B}_{1} \\ \mathbf{B}_{2} \end{array} \right], \left[\begin{array}{cc} \mathbf{C}_{1} & \mathbf{C}_{2} \end{array} \right] \right). \end{split}$$

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• Truncation \leadsto reduced order model: $(\mathbf{I}, \hat{\mathbf{A}}, \hat{\mathbf{B}}, \hat{\mathbf{C}}) = (\mathbf{I}, \mathbf{A}_{11}, \mathbf{B}_1, \mathbf{C}_1).$

Implementation: The Square Root Method

The SR Method

(₱**%O**₹: BTReductor)

1. Compute (Cholesky) factors of the solutions to the Lyapunov equation,

$$\mathbf{P} = \mathbf{S}^\mathsf{T} \mathbf{S}, \quad \mathbf{Q} = \mathbf{R}^\mathsf{T} \mathbf{R}.$$

Implementation: The Square Root Method

The SR Method

(PMOR: BTReductor)

1. Compute (Cholesky) factors of the solutions to the Lyapunov equation,

$$P = S^T S$$
, $Q = R^T R$.

2. Compute singular value decomposition

$$\mathbf{S}\mathbf{R}^\mathsf{T} = \left[egin{array}{ccc} \mathbf{U}_1, \ \mathbf{U}_2 \end{array}
ight] \left[egin{array}{ccc} oldsymbol{\Sigma}_1 \ & oldsymbol{\Sigma}_2 \end{array}
ight] \left[egin{array}{ccc} oldsymbol{V}_1^\mathsf{T} \ oldsymbol{V}_2^\mathsf{T} \end{array}
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3. Define

$$\mathbf{W} := \mathbf{R}^\mathsf{T} \mathbf{V}_1 \Sigma_1^{-1/2}, \qquad \mathbf{V} := \mathbf{S}^\mathsf{T} \mathbf{U}_1 \Sigma_1^{-1/2}.$$

4. Then the reduced order model is $(\mathbf{W}^\mathsf{T} \mathbf{A} \mathbf{V}, \mathbf{W}^\mathsf{T} \mathbf{B}, \mathbf{C} \mathbf{V})$.

Properties

- Lyapunov balancing preserves asymptotic stability.
- ullet We have the **a priori error bound**: $\left\|\mathbf{H} \hat{\mathbf{H}} \right\|_{\mathcal{H}_{\infty}} \leq 2\sum_{k=r+1}^n \sigma_k$

Variants

(₱**MO**R: BRBTReductor, LQGBTReductor)

Other versions for special classes of systems or applications exist, such as

positive-real balancing,

(passivity-preserving)

bounded-real balancing,

(contractivity-preserving)

• linear-quadratic Gaussian balancing.

(stability preserving)

(aims at low-order output feedback controllers)

The given ones all compute ${f P},\ {f Q}$ as solutions of **algebraic Riccati equations** of the form:

$$0 = \tilde{A}\mathbf{P}\tilde{E}^{\mathsf{T}} + \tilde{E}\mathbf{P}\tilde{A}^{\mathsf{T}} + \tilde{B}\tilde{B}^{\mathsf{T}} \pm \tilde{E}\mathbf{P}\tilde{C}^{\mathsf{T}}\tilde{C}\mathbf{P}\tilde{E}^{\mathsf{T}}$$
$$0 = \tilde{A}^{\mathsf{T}}\mathbf{Q}\tilde{E} + \tilde{E}^{\mathsf{T}}\mathbf{Q}\tilde{A} + \tilde{C}^{\mathsf{T}}\tilde{C} \pm \tilde{E}^{\mathsf{T}}\mathbf{Q}\tilde{B}\tilde{B}^{\mathsf{T}}\mathbf{Q}\tilde{E}.$$

Tools I

Lemma (Neumann series)

Let $\mathbf{A}\in\mathbb{C}^{n\times n}$ with spectral radius $\rho(\mathbf{A})<1$ be given. Then $\mathbf{I}-\mathbf{A}$ is invertible and it holds that

$$(\mathbf{I} - \mathbf{A})^{-1} = \sum_{k=0}^{\infty} \mathbf{A}^k.$$

Will be important to identify the actual shape of Markov parameters and system moments.

Tools II

Definition ((polynomial) Krylov subpace)

Given an invertible matrix $\mathbf{A} \in \mathbb{R}^{n \times n}$ and a vector $\mathbf{b} \in \mathbb{R}^n$ the k-dimensional (polynomial) Krylov subspace is defined as

$$\mathcal{K}_k(\mathbf{A}, \mathbf{b}) := \operatorname{span}\{\mathbf{b}, \mathbf{Ab}, \mathbf{A}^2\mathbf{b}, \dots, \mathbf{A}^{k-1}\mathbf{b}\}.$$

Definition (rational Krylov subpace)

Given an invertible matrix $\mathbf{A} \in \mathbb{R}^{n \times n}$ a vector $\mathbf{b} \in \mathbb{R}^n$ and a vector of shifts $s \in \mathbb{R}^k$ the k-dimensional rational Krylov subspace is defined as

$$\mathcal{K}_k(\mathbf{A}, \mathbf{b}, s) := \operatorname{span}\left\{ (s_1 \mathbf{I} - \mathbf{A})^{-1} \mathbf{b}, (s_2 \mathbf{I} - \mathbf{A})^{-1} \mathbf{b}, \dots, (s_k \mathbf{I} - \mathbf{A})^{-1} \mathbf{b} \right\}.$$

Orthonormal bases of these spaces should be computed via the **Arnoldi iteration**.

Padé-type approximations

Goal

Match the coefficients $\mathbf{M}_k(s_0)$ or $\mathbf{M}_k(\infty)$ in

$$\mathbf{H}(s) = \sum_{k=0}^{\infty} (s - s_0)^k M_k(s_0) \quad \mathbf{H}(s) = \sum_{k=0}^{\infty} s^{-k} M_k(\infty)$$

Motivation

(assume: m=p=1, s large enough)

$$\begin{split} \mathbf{H}(s) &= \mathbf{C}(s\mathbf{E} - \mathbf{A})^{-1}\mathbf{B} = \frac{1}{s}\mathbf{C}\underbrace{\left(\mathbf{I} - \frac{1}{s}\mathbf{E}^{-1}\mathbf{A}\right)^{-1}}_{=\sum_{k=0}^{\infty} \frac{1}{s^k}(\mathbf{E}^{-1}\mathbf{A})^k} \mathbf{E}^{-1}\mathbf{B} \\ &= \sum_{k=1}^{\infty} \mathbf{C}(\mathbf{E}^{-1}\mathbf{A})^{k-1}\mathbf{E}^{-1}\mathbf{B}\frac{1}{s^k}. \end{split}$$

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(assume: m=p=1, s large enough)

$$\begin{split} \mathbf{H}(s) &= \mathbf{C}(s\mathbf{E} - \mathbf{A})^{-1}\mathbf{B} = \frac{1}{s}\mathbf{C}\underbrace{\left(\mathbf{I} - \frac{1}{s}\mathbf{E}^{-1}\mathbf{A}\right)^{-1}}_{=\sum_{k=0}^{\infty} \frac{1}{s^k}(\mathbf{E}^{-1}\mathbf{A})^k} \mathbf{E}^{-1}\mathbf{B} \\ &= \sum_{k=1}^{\infty} \mathbf{C}(\mathbf{E}^{-1}\mathbf{A})^{k-1}\mathbf{E}^{-1}\mathbf{B}\frac{1}{s^k}. \end{split}$$

Therefore, we have

$$M_k(\infty) = \begin{cases} 0, & \text{if } k = 0, \\ \mathbf{C}(\mathbf{E}^{-1}\mathbf{A})^{k-1}\mathbf{E}^{-1}\mathbf{B}, & \text{if } k \geq 1. \end{cases} \text{ where } \mathbf{V} = \mathcal{K}_r(\mathbf{E}^{-1}\mathbf{A}, \mathbf{E}^{-1}\mathbf{B})$$

Padé-type approximations

Approximation at ∞

$$\mathbf{V} = \mathcal{K}_r(\mathbf{E}^{-1}\mathbf{A}, \mathbf{E}^{-1}\mathbf{B}), \quad \mathbf{W} = \mathbf{V} \text{ or } \mathbf{W} = \mathcal{K}_r(\mathbf{A}^\mathsf{T}\mathbf{E}^{-\mathsf{T}}, \mathbf{C}^\mathsf{T})$$

Approximation at $s_0 = 0$

$$\mathbf{V} = \mathcal{K}_r(\mathbf{A}^{-1}\mathbf{E}, \mathbf{A}^{-1}\mathbf{B}), \quad \mathbf{W} = \mathbf{V} \text{ or } \mathbf{W} = \mathcal{K}_r(\mathbf{E}^\mathsf{T}\mathbf{A}^{-\mathsf{T}}, \mathbf{C}^\mathsf{T})$$

Approximation at $s_0 \in (0, \infty)$

$$\mathbf{V} = \mathcal{K}_r((s_0 \mathbf{E} - \mathbf{A})^{-1} \mathbf{E}, (s_0 \mathbf{E} - \mathbf{A})^{-1} \mathbf{B}), \quad \mathbf{W} = \mathbf{V}$$

or

$$\mathbf{W} = \mathcal{K}_r (\mathbf{E}^\mathsf{T} (s_0 \mathbf{E}^\mathsf{T} - \mathbf{A}^\mathsf{T})^{-1}, \mathbf{C}^\mathsf{T})$$

Multi-point Moment Matching, Interpolation and IRKA/TSIA

Approximation at s_1, \ldots, s_r

$$\mathbf{V} = \mathcal{K}_r(s, \mathbf{E}^{-1}\mathbf{A}, \mathbf{E}^{-1}\mathbf{B}), \qquad \mathbf{W} = \mathbf{V} \text{ or } \mathbf{W} = \mathcal{K}_r(s, \mathbf{A}^\mathsf{T}\mathbf{E}^{-\mathsf{T}}, \mathbf{C}^\mathsf{T}).$$

- $\mathbf{W} = \mathbf{V}$ as above matches first r moments of (Σ).
- $\mathbf{W} \neq \mathbf{V}$ as above matches first 2r moments of (Σ).
- ullet $\mathbf{W}
 eq \mathbf{V}$ as above actually achieves Hermite interpolation of \mathbf{H} , see, e.g., [2].

How do we choose s_1, \ldots, s_r ?

\mathcal{H}_2 -optimal MOR

Find
$$\mathbf{s} = [s_1, \dots, s_r]^\mathsf{T}$$
, such that $\left\| \mathbf{H} - \hat{\mathbf{H}} \right\|_{\mathcal{H}_2}$ is minimized.

IRKA iterative improvement of $\mathbf s$ using $\Lambda(\hat{\mathbf E}_j,\hat{\mathbf A}_j)$.

(PMOR: IRKAReductor)

TSIA run a fixed point iteration on the first order necessary conditions.

(PMOR: TSIAReductor)

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