

# Basics of Systems and Control Theory for pyMOR

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August 23, 2022

pyMOR School 2022



$$y(t) = \begin{array}{c} \hat{x}(t) + B \\ \hat{y}(t) = \begin{array}{c} C \\ \hat{x}(t) + D \\ \hat{y}(t) = \end{array} \begin{array}{c} \hat{x}(t) + D \\ \hat{x}(t) + D \\ \hat{y}(t) = \begin{array}{c} \hat{x} \\ \hat{x}(t) + D \\ \hat{x}(t) + D \\ \hat{x}(t) + D \\ \hat{y}(t) = \begin{array}{c} \hat{x} \\ \hat{x}(t) + D \\ \hat{x}(t) + D \\ \hat{x}(t) + D \\ \hat{x}(t) + D \\ \hat{y}(t) = \begin{array}{c} \hat{x} \\ \hat{x}(t) + D \\ \hat{x}(t) + D$$

# MOR | Outline

- 1 Linear Time-Invariant (LTI) Systems
- 2 Transfer Function and Realizations
- 3 System Analysis
- 4 A Selection of MOR Methods

Only continuous-time systems
 Discrete-time is treated in [Ant05]

No differential-algebraic systems
 For DAE aspects see [Voi19, GSW13, MS05, Sty04]

No non-linearities

No parameter dependencies

- 1 Linear Time-Invariant (LTI) Systems
  - Setting for this course
  - Examples
- 2 Transfer Function and Realizations
- 3 System Analysis
- 4 A Selection of MOR Methods

# First-order State-space Systems

(PMOR: LTIModel)

$$\dot{\mathbf{x}}(t) = \mathbf{A}\mathbf{x}(t) + \mathbf{B}\mathbf{u}(t),$$
 
$$\mathbf{y}(t) = \mathbf{C}\mathbf{x}(t) + \mathbf{D}\mathbf{u}(t).$$
  $(\Sigma)$ 

### Here

- $\mathbf{x}(t) \in \mathbb{R}^n$  is called the state,
- $\mathbf{u}(t) \in \mathbb{R}^m$  is called the input,
- $\mathbf{y}(t) \in \mathbb{R}^p$  is called the output

of the LTI system. Correspondingly, we have

$$\mathbf{A} \in \mathbb{R}^{n \times n}, \qquad \mathbf{B} \in \mathbb{R}^{n \times m}, \qquad \mathbf{C} \in \mathbb{R}^{p \times n} \quad \text{and} \quad \mathbf{D} \in \mathbb{R}^{p \times m}.$$

We assume  $t \in [0, \infty)$ ,  $\mathbf{x}(0) = 0$ .

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# First-order State-space Systems

(PMOR: LTIModel)

$$\begin{aligned} \mathbf{E}\dot{\mathbf{x}}(t) &= \mathbf{A}\mathbf{x}(t) + \mathbf{B}\mathbf{u}(t), \\ \mathbf{y}(t) &= \mathbf{C}\mathbf{x}(t). \end{aligned} \tag{$\Sigma$}$$

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$$\mathbf{E}, \mathbf{A} \in \mathbb{R}^{n \times n}, \qquad \mathbf{B} \in \mathbb{R}^{n \times m}, \qquad \mathbf{C} \in \mathbb{R}^{p \times n}.$$

We assume  $t \in [0, \infty)$ ,  $\mathbf{x}(0) = 0$  and  $\mathbf{E}$  invertible.

# **Second-order State-space Systems**

(PMOR: SecondOrderModel)

$$\begin{split} \mathbf{M}\ddot{\mathbf{x}}(t) + \mathbf{E}\dot{\mathbf{x}}(t) + \mathbf{K}\mathbf{x}(t) &= \mathbf{B}\mathbf{u}(t), \\ \mathbf{y}(t) &= \mathbf{C}_{\mathrm{V}}\dot{\mathbf{x}}(t) + \mathbf{C}_{\mathrm{p}}\mathbf{x}(t). \end{split}$$

### Here

- $\mathbf{x}(t) \in \mathbb{R}^n$  is called the position,
- $\dot{\mathbf{x}}(t) \in \mathbb{R}^n$  is called the velocity,
- $\mathbf{u}(t) \in \mathbb{R}^m$  is called the input,
- $\mathbf{y}(t) \in \mathbb{R}^p$  is called the output

of the LTI system. Correspondingly, we have

M. E. 
$$K \in \mathbb{R}^{n \times n}$$
.

$$\mathbf{B} \in \mathbb{R}^{n \times m}$$

$$\mathbf{M}, \mathbf{E}, \mathbf{K} \in \mathbb{R}^{n \times n}, \qquad \mathbf{B} \in \mathbb{R}^{n \times m}, \qquad \mathbf{C}_{\mathbf{V}}, \mathbf{C}_{\mathbf{p}} \in \mathbb{R}^{p \times n}.$$

# Heat Equation [MORWiki thermal block] I

For  $t \in (0, T)$ ,  $\xi \in \Omega$  and initial values

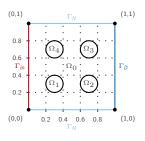
$$\theta(0,\xi) = 0$$
, for  $\xi \in \Omega$ ,

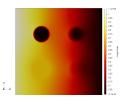
consider

$$\partial_t \theta(t,\xi) + \nabla \cdot (-\sigma(\xi)\nabla \theta(t,\xi)) = 0,$$

with boundary conditions

$$\begin{split} \sigma(\xi)\nabla\theta(t,\xi)\cdot n(\xi) &= u(t) &\quad t\in(0,\mathit{T}),\,\xi\in\Gamma_{\mathit{in}},\\ \sigma(\xi)\nabla\theta(t,\xi)\cdot n(\xi) &=0 &\quad t\in(0,\mathit{T}),\,\xi\in\Gamma_{\mathit{N}},\\ \theta(t,\xi) &=0 &\quad t\in(0,\mathit{T}),\,\xi\in\Gamma_{\mathit{D}}. \end{split}$$





# Heat Equation [MORWiki thermal block] II

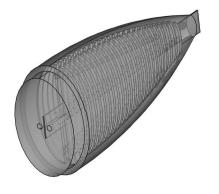
### Finite element semi-discretization in space

- pairwise inner products of ansatz functions → E
- discretized spatial operator + Dirichlet boundary condition → A
- discretized non-zero Neumann boundary condition → B
- ullet average temperatures on the inclusions  $\leadsto$   ${f C}$

- n = 7488
- m = 1
- p = 4

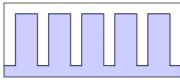
# An Artificial Fishtail [MORWiki Artificial Fishtail] I

### Construction:



### Fluid Elastomer Actuation:

# no pressure



# under pressure



# An Artificial Fishtail [MORWiki Artificial Fishtail] II

### Variables:

- displacement  $\vec{s}(t, \vec{z})$
- strain  $\underline{\vec{\varepsilon}}(\vec{s}(t, \vec{z}))$
- stress  $\vec{\underline{\sigma}}(\vec{s}(t,\vec{z}))$

# Material parameters:

- density  $\rho$
- Lamé parameters  $\lambda$ ,  $\mu$

# **Basic principle:**

$$\begin{split} \underline{\vec{\varepsilon}}(\vec{s}(t,\vec{z})) &= \frac{1}{2} \left( \nabla \vec{s}(t,\vec{z}) + \nabla^{\mathsf{T}} \vec{s}(t,\vec{z}) \right) & \text{(kinematic equation)} \\ \underline{\vec{\sigma}}(\vec{s}(t,\vec{z})) &= \lambda \operatorname{tr}((\underline{\vec{\varepsilon}}(\vec{s}(t,\vec{z}))) \, \underline{l} + 2\mu \underline{\vec{\varepsilon}}(\vec{s}(t,\vec{z}))) & \text{(material equation)} \\ \rho \frac{\partial^2 \vec{s}(t,\vec{z})}{\partial t^2} &= \nabla \cdot \underline{\vec{\sigma}}(\vec{s}(t,\vec{z})) + \vec{f}(t,\vec{z}) & \text{(equation of motion)} \end{split}$$

+ initial and boundary conditions

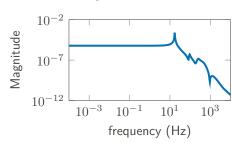
# An Artificial Fishtail [MORWiki Artificial Fishtail] III

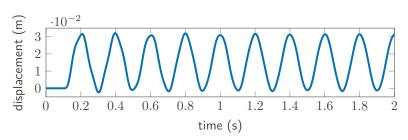
FEM semi-discretization:

$$\begin{split} \mathbf{M}\ddot{\mathbf{x}}(t) + \mathbf{E}\dot{\mathbf{x}}(t) + \mathbf{K}\mathbf{x}(t) &= \mathbf{B}\mathbf{u}(t), \\ \mathbf{y}(t) &= \mathbf{C}_{\mathbf{p}}\mathbf{x}(t), \end{split}$$

with

- $\mathbf{M}, \mathbf{E}, \mathbf{K} > 0$ ,  $\mathbf{C}_{\mathbf{v}} = 0$ ,
- n = 779232, m = 1, p = 3.





- 1 Linear Time-Invariant (LTI) Systems
- 2 Transfer Function and Realizations
  - Laplace Transform
  - Transfer Function
  - Realizations
  - Projection-based MOR
- 3 System Analysis
- 4 A Selection of MOR Methods

Then

$$\mathcal{L}\left\{f\right\}\left(\mathsf{s}\right) := \int_{0}^{\infty} f(\tau) \mathsf{e}^{-\mathsf{s}\tau} \mathsf{d}\tau$$

for  $Re(s) > \alpha$  is called the **Laplace transform** of f. The process of forming the Laplace transform is called **Laplace transformation**.

It can be shown that the integral converges uniformly in a domain with  $Re(s) \ge \beta$  for all  $\beta > \alpha$ .

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It can be shown that the integral converges uniformly in a domain with  $Re(s) \ge \beta$  for all  $\beta > \alpha$ .

Allows us to map time signals to frequency signals.

# **MOR** | Transfer Function and Realizations | Laplace Transform

### Theorem

Let  $f, g, h : [0, \infty) \to \mathbb{R}^n$  be given. Then the following two statements hold true:

a) The Laplace transformation is linear, i. e., if f and g are exponentially bounded, then  $h:=\gamma f+\delta g$  is also exponentially bounded and

$$\mathcal{L}\left\{h\right\} = \gamma \mathcal{L}\left\{f\right\} + \delta \mathcal{L}\left\{g\right\}$$

holds for all  $\gamma, \delta \in \mathbb{C}$ .

b) If  $f \in \mathcal{PC}^1([0,\infty),\mathbb{R}^n)$  and  $\dot{f}$  is exponentially bounded, then f is exponentially bounded and

$$\mathcal{L}\{\dot{f}\}(\mathbf{s}) = \mathbf{s}\mathcal{L}\{f\}(\mathbf{s}) - f(0).$$



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b) If  $f \in \mathcal{PC}^1([0,\infty),\mathbb{R}^n)$  and  $\dot{f}$  is exponentially bounded, then f is exponentially bounded and

$$\mathcal{L}\{\dot{f}\}(s) = s\mathcal{L}\{f\}(s) - f(0).$$

- $X(s) := \mathcal{L}\{\mathbf{x}\}(s), \ U(s) := \mathcal{L}\{\mathbf{u}\}(s), \ \text{and} \ Y(s) := \mathcal{L}\{\mathbf{y}\}(s)$
- $\mathbf{A}\mathbf{x}(t) + \mathbf{B}\mathbf{u}(t) \rightsquigarrow \mathbf{A}X(s) + \mathbf{B}U(s)$
- $\mathbf{v}(t) = \mathbf{C}\mathbf{x}(t) \rightsquigarrow Y(s) = \mathbf{C}X(s)$

# **Rational Matrix Function Representation**

### In summary we have:

- sEX(s) = AX(s) + BU(s)
- $Y(s) = \mathbf{C}X(s)$

Thus the mapping from inputs to outputs in frequency domain can be expressed as

$$\mathbf{H}(s) = \mathbf{C}(s\mathbf{E} - \mathbf{A})^{-1}\mathbf{B}.$$

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Analogously, for second-order systems we get

$$\mathbf{H}(s) = (s\mathbf{C}_{v} + \mathbf{C}_{p}) (s^{2}\mathbf{M} + s\mathbf{E} + \mathbf{K})^{-1}\mathbf{B}.$$

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**H** is analytic in  $\mathbb{C} \setminus \Lambda(\mathbf{E}, \mathbf{A})$ , or  $\mathbb{C} \setminus \Lambda(\mathbf{M}, \mathbf{E}, \mathbf{K})$ , respectively

# Important Representations of $\hat{H}(s)$

### (Laurent) series expansion

$$\mathbf{H}(s) = \sum_{k=0}^{\infty} (s - s_0)^k M_k(s_0)$$
  $\mathbf{H}(s) = \sum_{k=0}^{\infty} s^{-k} M_k(\infty)$ 

The matrices  $M_k(s_0)$  are called **moments** of **H**. At infinity they are also referred to as **Markov parameters**.

### Pole Residue Form

Let  $(\lambda_i, w_i, v_i)$  be the eigentriplets of the pair  $(\mathbf{A}, \mathbf{E})$  with no degenerate eigenspaces. Then we have

$$\mathbf{H}(s) = \sum_{i=1}^{n} \frac{R_i}{s - \lambda_i},$$

where  $R_i = (\mathbf{C}v_i)(w_i^{\mathsf{H}}\mathbf{B})$ , assuming  $w_i^{\mathsf{H}}v_i = 1$ .

# **MOR** | Transfer Function and Realizations | Realizations

The representation of H using (E, A, B, C) is not unique.

In fact for any invertible matrix  $\mathbf{T} \in \mathbb{R}^{n \times n}$ , we have

$$\begin{aligned} \mathbf{H}(s) &= \mathbf{C}(s\mathbf{E} - \mathbf{A})^{-1}\mathbf{B} \\ &= \mathbf{C}\mathbf{T}^{-1}\mathbf{T}(s\mathbf{E} - \mathbf{A})^{-1}\mathbf{T}^{-1}\mathbf{T}\mathbf{B} \\ &= \mathbf{C}\mathbf{T}^{-1}\left(s\mathbf{T}\mathbf{E}\mathbf{T}^{-1} - \mathbf{T}\mathbf{A}\mathbf{T}^{-1}\right)^{-1}\mathbf{T}\mathbf{B} \end{aligned}$$

and thus a system given, by  $(\mathbf{TET}^{-1},\mathbf{TAT}^{-1},\mathbf{TB},\mathbf{CT}^{-1})$  realizes the exact same input/output behavior.

### Definition

- All sets of matrices leading to the same function H are called its realizations.
- The matrix T above is called state-space transformation.

# Important Realizations

Minimal Realizations

Can we realize  ${\bf H}$  with less equations?

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Can we introduce a small error to get even less equations?

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Balanced Realizations



Can we find state coordinates that allow us to decide what is important?

# McMillan Degree and Minimal Realization

### Example

Realizations can even be of different dimensions. Take for example:

$$\mathbf{E} = \mathbf{I} \text{ the identity, } \mathbf{A} = \begin{bmatrix} -11 & 0 \\ 0 & -5 \end{bmatrix} \text{, } \mathbf{B} = \begin{bmatrix} 1 \\ 1 \end{bmatrix} \text{ and } \mathbf{C} = \begin{bmatrix} 1 & 0 \end{bmatrix}.$$

Truncating the second state component does not change H.

### Definition

There exists a minimum number of equations necessary to describe  $\mathbf{H}$ . The state dimension n of this minimal set of equations is called  $\mathbf{McMillan}$  degree of the system. A realization of  $\mathbf{H}$  with this dimension is called minimal realization.

$$\begin{aligned} \mathbf{E}\dot{\mathbf{x}}(t) - \mathbf{A}\mathbf{x}(t) - \mathbf{B}\mathbf{u}(t) &= 0, \\ \mathbf{y}(t) - \mathbf{C}\mathbf{x}(t) - \mathbf{D}\mathbf{u}(t) &= 0. \end{aligned}$$

$$\begin{split} \mathbf{E}\mathbf{V}\dot{\hat{\mathbf{x}}}(t) - \mathbf{A}\mathbf{V}\hat{\mathbf{x}}(t) - \mathbf{B}\mathbf{u}(t) &= e_{\mathrm{res}}(t), \\ \mathbf{y}(t) - \mathbf{C}\mathbf{V}\hat{\mathbf{x}}(t) - \mathbf{D}\mathbf{u}(t) &= e_{\mathrm{output}}(t). \end{split}$$

### Step I: Use truncated state transformation

Replace

$$\mathbf{x}(t) \approx \mathbf{V}\hat{\mathbf{x}}(t)$$

with  $\mathbf{V} \in \mathbb{R}^{n \times r}$  and  $\hat{\mathbf{x}}(t) \in \mathbb{R}^{r}$ .

$$\begin{split} \mathbf{V}^{\mathsf{T}}\mathbf{E}\mathbf{V}\dot{\hat{\mathbf{x}}}(t) - \mathbf{V}^{\mathsf{T}}\mathbf{A}\mathbf{V}\dot{\hat{\mathbf{x}}}(t) - \mathbf{V}^{\mathsf{T}}\mathbf{B}\mathbf{u}(t) &= 0,\\ \mathbf{y}(t) - \mathbf{C}\mathbf{V}\dot{\hat{\mathbf{x}}}(t) - \mathbf{D}\mathbf{u}(t) &= e_{\mathrm{output}}(t). \end{split}$$

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### Step II: Mitigate transformation error

Suppress truncation residual through left projection.

• one-sided method: use **V** again.

$$\begin{split} \mathbf{W}^{\mathrm{T}}\mathbf{E}\mathbf{V}\dot{\hat{\mathbf{x}}}(t) - \mathbf{W}^{\mathrm{T}}\mathbf{A}\mathbf{V}\hat{\mathbf{x}}(t) - \mathbf{W}^{\mathrm{T}}\mathbf{B}\mathbf{u}(t) &= 0,\\ \mathbf{y}(t) - \mathbf{C}\mathbf{V}\hat{\mathbf{x}}(t) - \mathbf{D}\mathbf{u}(t) &= e_{\mathrm{output}}(t). \end{split}$$

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Replace

$$\mathbf{x}(t) \approx \mathbf{V}\hat{\mathbf{x}}(t)$$

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### Step II: Mitigate transformation error

Suppress truncation residual through left projection.

- one-sided method: use V again.
- two-sided method: find  $\mathbf{W} \in \mathbb{R}^{n \times r}$ .

 $\hat{\mathbf{A}} \approx \mathbf{A}$ 

### Reduced order model (ROM)

(PMOR: LTIPGReductor)

Define  $\hat{\mathbf{E}} = \mathbf{W}^T \mathbf{E} \mathbf{V}$ ,  $\hat{\mathbf{A}} = \mathbf{W}^T \mathbf{A} \mathbf{V} \in \mathbb{R}^{r \times r}$ ,  $\hat{\mathbf{B}} = \mathbf{W}^T \mathbf{B} \in \mathbb{R}^{r \times m}$  and  $\hat{\mathbf{C}} = \mathbf{C} \mathbf{V} \in \mathbb{R}^{p \times r}$ . Then

$$\hat{\mathbf{E}}\dot{\hat{\mathbf{x}}}(t) = \hat{\mathbf{A}}\hat{\mathbf{x}}(t) + \hat{\mathbf{B}}\mathbf{u}(t),$$

$$\hat{\mathbf{y}}(t) = \hat{\mathbf{C}}\hat{\mathbf{x}}(t) + \mathbf{D}\mathbf{u}(t)$$
(ROM)

approximates the dynamics of the full-order model  $(\Sigma)$  with output error

$$\mathbf{y}(t) - \hat{\mathbf{y}}(t) = e_{\text{output}}(t).$$

- We call the corresponding transfer function  $\hat{\mathbf{H}}$ .
- Model order reduction (MOR)  $\leadsto$  Find  $\mathbf{W}, \mathbf{V} \in \mathbb{R}^{n \times r}$  such that  $e_{\text{output}}(t)$  is small in a suitable sense.
- We will focus on energy-based and interpolation-based methods today.

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  - System Norms and Hardy Spaces
  - Frequency-Domain Analysis
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We have

$$Y(s) = \mathbf{H}(s)U(s)$$

and

$$\hat{Y}(s) = \hat{\mathbf{H}}(s)U(s).$$

#### Question

What are suitable norms such that

$$||y - \hat{y}|| \le ||\mathbf{H} - \hat{\mathbf{H}}|| \, ||u||?$$

The Banach Space  $\mathcal{H}_{\infty}^{p \times m}$ 

$$\mathcal{H}_{\infty}^{\rho\times m}:=\left\{G:\mathbb{C}^{+}\rightarrow\mathbb{C}^{\rho\times m}\ :\ G\text{ is analytic in }\mathbb{C}^{+}\text{ and }\sup_{s\in\mathbb{C}^{+}}\left\Vert G(s)\right\Vert _{2}<\infty\right\}.$$

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 $\mathcal{H}_{\infty}^{p imes m}$  is a Banach space equipped with the  $\mathcal{H}_{\infty}$ -norm

$$\left\| \mathbf{G} \right\|_{\mathcal{H}_{\infty}} := \sup_{\omega \in \mathbb{R}} \left\| \mathbf{G}(\mathrm{i}\omega) \right\|_{2}.$$

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Can show: 
$$\|\mathbf{y} - \hat{\mathbf{y}}\|_{\mathcal{L}_2} \leq \|\mathbf{H} - \hat{\mathbf{H}}\|_{\mathcal{H}_{\infty}} \|\mathbf{u}\|_{\mathcal{L}_2}.$$

This bound can even be shown to be sharp.

## The Hilbert Space $\mathcal{H}_2^{\rho \times m}$

$$\mathcal{H}_2^{
ho imes m}:=\left\{\mathit{G}:\mathbb{C}^+ o\mathbb{C}^{
ho imes m}\ :\ \mathit{G}\ ext{is analytic in }\mathbb{C}^+\ ext{and}
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### The Hilbert Space $\mathcal{H}_2^{p \times m}$

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 $\mathcal{H}_2^{p \times m}$  is a Hilbert space with the inner product

$$\langle F, G \rangle_{\mathcal{H}_2} := \frac{1}{2\pi} \int_{-\infty}^{\infty} \operatorname{tr}\left(F(i\omega)^{\mathsf{H}} G(i\omega)\right) d\omega$$

and induced norm

$$\left\|\mathbf{G}\right\|_{\mathcal{H}_{2}}:=\left\langle\mathbf{G},\mathbf{G}\right\rangle_{\mathcal{H}_{2}}^{1/2}=\left(\frac{1}{2\pi}\int_{-\infty}^{\infty}\left\|\mathbf{G}(\mathrm{i}\omega)\right\|_{\mathrm{F}}^{2}\mathrm{d}\omega\right)^{1/2}.$$

# The Hilbert Space $\mathcal{H}_2^{p \times m}$

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$$\sup_{\xi>0}\int_{-\infty}^{\infty}\left\| \mathit{G}(\xi+\mathrm{i}\omega)\right\|_{\mathrm{F}}^{2}\mathrm{d}\omega<\infty\right\}.$$

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and induced norm

$$\|\mathbf{G}\|_{\mathcal{H}_2} := \langle \mathbf{G}, \mathbf{G} \rangle_{\mathcal{H}_2}^{1/2} = \left(\frac{1}{2\pi} \int_{-\infty}^{\infty} \|\mathbf{G}(\mathrm{i}\omega)\|_{\mathrm{F}}^2 \, \mathrm{d}\omega\right)^{1/2}.$$

$$\|\mathbf{y} - \hat{\mathbf{y}}\|_{\mathcal{L}_{\infty}} \leq \|\mathbf{H} - \hat{\mathbf{H}}\|_{\mathcal{H}_{2}} \|\mathbf{u}\|_{\mathcal{L}_{2}}.$$

### System Gramians and $\mathcal{H}_2$ -trace-formula

A system  $(\Sigma)$  with  $\Lambda(\mathbf{E}, \mathbf{A}) \subset \mathbb{C}^-$  is called **asymptotically stable**. Then, all state trajectories decay exponentially as  $t \to \infty$  and

a) the infinite controllability and observability Gramians exist:

$$\begin{split} \mathbf{P} &= \int_0^\infty \mathrm{e}^{\mathbf{E}^{-1}\mathbf{A}t} \mathbf{E}^{-1} \mathbf{B} \mathbf{B}^\mathsf{T} \mathbf{E}^{-\mathsf{T}} \mathrm{e}^{\mathbf{A}^\mathsf{T} \mathbf{E}^{-\mathsf{T}}t} \mathrm{d}t \\ \mathbf{E}^\mathsf{T} \mathbf{Q} \mathbf{E} &= \int_0^\infty \mathrm{e}^{\mathbf{A}^\mathsf{T} \mathbf{E}^{-\mathsf{T}}t} \mathbf{C}^\mathsf{T} \mathbf{C} \mathrm{e}^{\mathbf{E}^{-1}\mathbf{A}t} \mathrm{d}t. \end{split}$$

b) P, Q solve the two Lyapunov equations

$$\mathbf{A}\mathbf{P}\mathbf{E}^\mathsf{T} + \mathbf{E}\mathbf{P}\mathbf{A}^\mathsf{T} = -\mathbf{B}\mathbf{B}^\mathsf{T}, \qquad \mathbf{A}^\mathsf{T}\mathbf{Q}\mathbf{E} + \mathbf{E}^\mathsf{T}\mathbf{Q}\mathbf{A} = -\mathbf{C}^\mathsf{T}\mathbf{C}$$

c) the  $\mathcal{H}_2$ -norm can be expressed as

$$\left\|\mathbf{H}\right\|_{\mathcal{H}_2}^2 = \mathrm{tr} \left(\mathbf{CPC}^{\mathsf{T}}\right) = \mathrm{tr} \left(\mathbf{B}^{\mathsf{T}} \mathbf{QB}\right).$$

# System Analysis | Frequency-Domain Analysis

#### **Bode Plots**

The Bode plot for **H** consists of a magnitude plot and a phase plot.

#### Bode magnitude plot

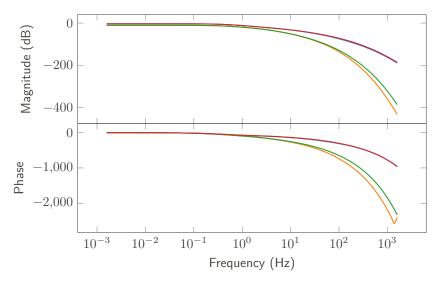
- component-wise graph of the function  $|\mathbf{H}(\mathrm{i}\omega)|$  for frequencies  $\omega \in [\omega_{\min}, \omega_{\max}] \subset \mathbb{R}.$
- $\omega$ -axis is logarithmic.
- magnitude is given in decibels, i.e.,  $|\mathbf{H}(i.)|$  is plotted as  $20\log_{10}(|\mathbf{H}(i.)|)$ .

### Bode phase plot

- component-wise graph of the function  $\arg \mathbf{H}(\mathrm{i}\omega)$  for frequencies  $\omega \in [\omega_{\min}, \omega_{\max}] \subset \mathbb{R}.$
- $\omega$ -axis is logarithmic.
- phase is given in degrees on a linear scale.

# **System Analysis** | Frequency-Domain Analysis

### **Bode Plot for the Thermal Block Example**



# **System Analysis** | Frequency-Domain Analysis

### (Sigma) Magnitude Plots

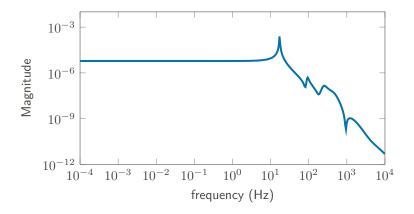
### Sigma magnitude plot

- 2-norm-wise graph of the function  $\mathbf{H}(\mathrm{i}\omega)$  for frequencies  $\omega \in [\omega_{\min}, \omega_{\max}] \subset \mathbb{R}.$
- $\omega$ -axis is logarithmic.

The name is due to the fact that for a given matrix  ${\bf M}$  the norm  $\|{\bf M}\|_2$  is given by its largest singular value.

The real sigma magnitude plot depicts all singular values as functions of  $\omega$ .

### Sigma Magnitude Plot for the Artificial Fishtail



# $\mathbb{P}/MOR \mid A$ Selection of MOR Methods

- 1 Linear Time-Invariant (LTI) Systems
- 2 Transfer Function and Realizations
- 3 System Analysis
- 4 A Selection of MOR Methods
  - Modal Methods
  - Balancing Based MOR
  - Moments and Interpolation

# MOR | A Selection of MOR Methods | Modal Methods

#### **Modal Coordinates**

Assume that the pair (E,A), respectively the triple (M,E,K), is simultaneously diagonalizable in  $\mathbb{C}^{n\times n}$ .

#### Classic Modal Truncation

- Compute diagonal realization from an eigendecomposition.
- State-space transformation matrices contain eigenvectors (modes).
- Use W = V.
- Populate V with modes corresponding to eigenvalues closest to  $i\mathbb{R}$ .
- Add a few domain specific or "anxiety" modes.

#### **Problem**

- Does not take inputs and outputs into account!
- How many "anxiety" modes are necessary?

### **Dominant Poles Approximation**

(PMOR: MTReductor)

Recall the pole residue form of the transfer function

$$\mathbf{H}(s) = \sum_{i=1}^{n} \frac{R_i}{s - \lambda_i},$$

where  $R_i = (\mathbf{C}v_i)(w_i^{\mathsf{H}}\mathbf{B})$ , assuming  $w_i^{\mathsf{H}}v_i = 1$ .

Sort and select modes by the magnitude of the  $||R_i|| / \text{Re}(\lambda_i)$ . Then

#### Error bound

$$\left\|\mathbf{H} - \hat{\mathbf{H}}\right\|_{\infty} \leq \sum_{i=r+1}^{n} \frac{\left\|R_{i}\right\|}{\left|\operatorname{Re}(\lambda_{i})\right|}$$

Computation is feasible via *subspace accelerated MIMO dominant pole algorithm* (SAMDP).

#### Idea:

• The system ( $\Sigma$ ), in realization ( $\mathbf{E} = \mathbf{I}, \mathbf{A}, \mathbf{B}, \mathbf{C}$ ), is called balanced, if the solutions  $\mathbf{P}$ ,  $\mathbf{Q}$  of the Lyapunov equations

$$AP + PA^{T} + BB^{T} = 0,$$
  $A^{T}Q + QA + C^{T}C = 0,$ 

satisfy: 
$$\mathbf{P} = \mathbf{Q} = \operatorname{diag}(\sigma_1, \dots, \sigma_n)$$
 where  $\sigma_1 \geq \sigma_2 \geq \dots \geq \sigma_n > 0$ .

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- $\{\sigma_1, \dots, \sigma_n\}$  are the Hankel singular values (HSVs) of  $\Sigma$ .
- A balanced realization is computed via state space transformation

$$\begin{split} \boldsymbol{\mathcal{T}} : (\textbf{I}, \textbf{A}, \textbf{B}, \textbf{C}) &\mapsto (\textbf{I}, \textbf{TAT}^{-1}, \textbf{TB}, \textbf{CT}^{-1}) \\ &= \left( \left[ \begin{array}{cc} \textbf{A}_{11} & \textbf{A}_{12} \\ \textbf{A}_{21} & \textbf{A}_{22} \end{array} \right], \left[ \begin{array}{cc} \textbf{B}_{1} \\ \textbf{B}_{2} \end{array} \right], \left[ \begin{array}{cc} \textbf{C}_{1} & \textbf{C}_{2} \end{array} \right] \right). \end{split}$$

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• Truncation  $\rightsquigarrow$  reduced order model:  $(I, \hat{A}, \hat{B}, \hat{C}) = (I, A_{11}, B_1, C_1)$ .

### Implementation: The Square Root Method

### The SR Method

(EMOR: BTReductor)

1. Compute (Cholesky) factors of the solutions to the Lyapunov equation,

$$P = S^T S$$
,  $Q = R^T R$ .

### Implementation: The Square Root Method

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2. Compute singular value decomposition

$$\mathbf{SR}^\mathsf{T} = \left[ egin{array}{ccc} \mathbf{U}_1, \ \mathbf{U}_2 \end{array} 
ight] \left[ egin{array}{ccc} oldsymbol{\Sigma}_1 & & \ & oldsymbol{\Sigma}_2 \end{array} 
ight] \left[ egin{array}{ccc} \mathbf{V}_1^\mathsf{T} \ \mathbf{V}_2^\mathsf{T} \end{array} 
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### Implementation: The Square Root Method

#### The SR Method

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1. Compute (Cholesky) factors of the solutions to the Lyapunov equation,

$$\mathbf{P} = \mathbf{S}^{\mathsf{T}}\mathbf{S}, \quad \mathbf{Q} = \mathbf{R}^{\mathsf{T}}\mathbf{R}.$$

2. Compute singular value decomposition

$$\mathbf{SR}^{\mathsf{T}} = \left[ \, \mathbf{U}_1, \, \mathbf{U}_2 \, \right] \left[ \begin{array}{cc} \Sigma_1 & \\ & \Sigma_2 \end{array} \right] \left[ \begin{array}{c} \mathbf{V}_1^{\mathsf{T}} \\ \mathbf{V}_2^{\mathsf{T}} \end{array} \right].$$

3. Define

$$\mathbf{W} := \mathbf{R}^{\mathsf{T}} \mathbf{V}_1 \Sigma_1^{-1/2}, \qquad \mathbf{V} := \mathbf{S}^{\mathsf{T}} \mathbf{U}_1 \Sigma_1^{-1/2}.$$

4. Then the reduced order model is (**W**<sup>T</sup>**AV**, **W**<sup>T</sup>**B**, **CV**).

# PYMOR | A Selection of MOR Methods | Balancing Based MOR

#### **Properties**

- Lyapunov balancing preserves asymptotic stability.
- We have the a priori error bound:  $\|\mathbf{H} \hat{\mathbf{H}}\|_{\mathcal{H}_{\infty}} \leq 2\sum_{k=r+1}^{n} \sigma_{k}$

#### **Variants**

### (PYNOR: BRBTReductor, LQGBTReductor)

Other versions for special classes of systems or applications exist, such as

- positive-real balancing, (passivity-preserving)
- bounded-real balancing, (contractivity-preserving)
- linear-quadratic Gaussian balancing. (stability preserving)
   (aims at low-order output feedback controllers)

The given ones all compute P, Q as solutions of algebraic Riccati equations of the form:

$$\begin{split} 0 &= \tilde{\mathbf{A}} \mathbf{P} \tilde{\mathbf{E}}^\mathsf{T} + \tilde{\mathbf{E}} \mathbf{P} \tilde{\mathbf{A}}^\mathsf{T} + \tilde{\mathbf{B}} \tilde{\mathbf{B}}^\mathsf{T} \pm \tilde{\mathbf{E}} \mathbf{P} \tilde{\mathbf{C}}^\mathsf{T} \tilde{\mathbf{C}} \mathbf{P} \tilde{\mathbf{E}}^\mathsf{T} \\ 0 &= \tilde{\mathbf{A}}^\mathsf{T} \mathbf{Q} \tilde{\mathbf{E}} + \tilde{\mathbf{E}}^\mathsf{T} \mathbf{Q} \tilde{\mathbf{A}} + \tilde{\mathbf{C}}^\mathsf{T} \tilde{\mathbf{C}} \pm \tilde{\mathbf{E}}^\mathsf{T} \mathbf{Q} \tilde{\mathbf{B}} \tilde{\mathbf{B}}^\mathsf{T} \mathbf{Q} \tilde{\mathbf{E}}. \end{split}$$

#### Tools I

### Lemma (Neumann series)

Let  $\mathbf{A} \in \mathbb{C}^{n \times n}$  with spectral radius  $\rho(\mathbf{A}) < 1$  be given. Then  $\mathbf{I} - \mathbf{A}$  is invertible and it holds that

$$(\mathbf{I} - \mathbf{A})^{-1} = \sum_{k=0}^{\infty} \mathbf{A}^k.$$

Will be important to identify the actual shape of Markov parameters and system moments.

#### Tools II

### **Definition** ((polynomial) Krylov subpace)

Given an invertible matrix  $\mathbf{A} \in \mathbb{R}^{n \times n}$  and a vector  $\mathbf{b} \in \mathbb{R}^n$  the k-dimensional (polynomial) Krylov subspace is defined as

$$\mathcal{K}_{k}(\mathbf{A},\mathbf{b}) := \operatorname{span}\{\mathbf{b},\mathbf{Ab},\mathbf{A}^{2}\mathbf{b},\ldots,\mathbf{A}^{k-1}\mathbf{b}\}$$
.

### Definition (rational Krylov subpace)

Given an invertible matrix  $\mathbf{A} \in \mathbb{R}^{n \times n}$  a vector  $\mathbf{b} \in \mathbb{R}^n$  and a vector of shifts  $s \in \mathbb{R}^k$  the k-dimensional rational Krylov subspace is defined as

$$\mathcal{K}_k(\mathbf{A}, \mathbf{b}, s) := \operatorname{span}\left\{ \left(s_1 \mathbf{I} - \mathbf{A}\right)^{-1} \mathbf{b}, \left(s_2 \mathbf{I} - \mathbf{A}\right)^{-1} \mathbf{b}, \dots, \left(s_k \mathbf{I} - \mathbf{A}\right)^{-1} \mathbf{b} \right\}.$$

Orthonormal bases of these spaces should be computed via the **Arnoldi** iteration.

### Padé-type approximations

#### Goal

Match the coefficients  $\mathbf{M}_k(s_0)$  or  $\mathbf{M}_k(\infty)$  in

$$\mathbf{H}(s) = \sum_{k=0}^{\infty} (s - s_0)^k M_k(s_0)$$
  $\mathbf{H}(s) = \sum_{k=0}^{\infty} s^{-k} M_k(\infty)$ 

#### Motivation

(assume: m = p = 1, s large enough)

$$\begin{split} \mathbf{H}(s) &= \mathbf{C}(s\mathbf{E} - \mathbf{A})^{-1}\mathbf{B} = \frac{1}{s}\mathbf{C}\underbrace{\left(\mathbf{I} - \frac{1}{s}\mathbf{E}^{-1}\mathbf{A}\right)^{-1}}_{=\sum_{k=0}^{\infty} \frac{1}{s^k}(\mathbf{E}^{-1}\mathbf{A})^k} \mathbf{E}^{-1}\mathbf{B} \\ &= \sum_{k=1}^{\infty} \mathbf{C}(\mathbf{E}^{-1}\mathbf{A})^{k-1}\mathbf{E}^{-1}\mathbf{B}\frac{1}{s^k}. \end{split}$$

### Padé-type approximations

### Motivation

(assume: m = p = 1, s large enough)

$$\begin{split} \mathbf{H}(s) &= \mathbf{C}(s\mathbf{E} - \mathbf{A})^{-1}\mathbf{B} = \frac{1}{s}\mathbf{C}\underbrace{\left(\mathbf{I} - \frac{1}{s}\mathbf{E}^{-1}\mathbf{A}\right)^{-1}}_{=\sum_{k=0}^{\infty} \frac{1}{s^k}(\mathbf{E}^{-1}\mathbf{A})^k} \mathbf{E}^{-1}\mathbf{B} \\ &= \sum_{k=1}^{\infty} \mathbf{C}(\mathbf{E}^{-1}\mathbf{A})^{k-1}\mathbf{E}^{-1}\mathbf{B}\frac{1}{s^k}. \end{split}$$

Therefore, we have

$$M_k(\infty) = \begin{cases} 0, & \text{if } k = 0, \\ \mathbf{C}(\mathbf{E}^{-1}\mathbf{A})^{k-1}\mathbf{E}^{-1}\mathbf{B}, & \text{if } k \geq 1. \end{cases} \implies \text{use } \mathbf{V} = \mathcal{K}_r(\mathbf{E}^{-1}\mathbf{A}, \mathbf{E}^{-1}\mathbf{B})$$

### Padé-type approximations

#### Approximation at $\infty$

$$\mathbf{V} = \mathcal{K}_r(\mathbf{E}^{-1}\mathbf{A}, \mathbf{E}^{-1}\mathbf{B}), \qquad \mathbf{W} = \mathbf{V} \text{ or } \mathbf{W} = \mathcal{K}_r(\mathbf{A}^\mathsf{T}\mathbf{E}^{-\mathsf{T}}, \mathbf{C}^\mathsf{T})$$

### **Approximation at** $s_0 = 0$

$$\mathbf{V} = \mathcal{K}_r(\mathbf{A}^{-1}\mathbf{E}, \mathbf{A}^{-1}\mathbf{B}), \qquad \mathbf{W} = \mathbf{V} \text{ or } \mathbf{W} = \mathcal{K}_r(\mathbf{E}^T\mathbf{A}^{-T}, \mathbf{C}^T)$$

### **Approximation at** $s_0 \in (0, \infty)$

$$\mathbf{V} = \mathcal{K}_r((s_0\mathbf{E} - \mathbf{A})^{-1}\mathbf{E}, (s_0\mathbf{E} - \mathbf{A})^{-1}\mathbf{B}), \qquad \mathbf{W} = \mathbf{V}$$

or

$$\mathbf{W} = \mathcal{K}_r(\mathbf{E}^\mathsf{T}(s_0\mathbf{E}^\mathsf{T} - \mathbf{A}^\mathsf{T})^{-1}, \mathbf{C}^\mathsf{T})$$

### Multi-point Moment Matching, Interpolation and IRKA/TSIA

### **Approximation at** $s_1, \ldots, s_r$

$$\mathbf{V} = \mathcal{K}_r(\mathbf{s}, \mathbf{E}^{-1} \mathbf{A}, \mathbf{E}^{-1} \mathbf{B}), \qquad \mathbf{W} = \mathbf{V} \text{ or } \mathbf{W} = \mathcal{K}_r(\mathbf{s}, \mathbf{A}^\mathsf{T} \mathbf{E}^{-\mathsf{T}}, \mathbf{C}^\mathsf{T}).$$

- $\mathbf{W} = \mathbf{V}$  as above matches first r moments of  $(\Sigma)$ .
- $\mathbf{W} \neq \mathbf{V}$  as above matches first 2r moments of  $(\Sigma)$ .
- $\mathbf{W} \neq \mathbf{V}$  as above actually achieves Hermite interpolation of  $\mathbf{H}$ , see, e.g., [ABG20].

How do we choose  $s_1, \ldots, s_r$ ?

#### $\mathcal{H}_2$ -optimal MOR

Find  $\mathbf{s} = [s_1, \dots, s_r]^\mathsf{T}$ , such that  $\|\mathbf{H} - \hat{\mathbf{H}}\|_{\mathcal{H}_2}$  is minimized.

IRKA iterative improvement of **s** using  $\Lambda(\hat{\mathbf{E}}_j, \hat{\mathbf{A}}_j)$ .

(PYNOR: IRKAReductor)

TSIA run a fixed point iteration on the first order necessary conditions. (FYMOR: TSIAReductor)

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