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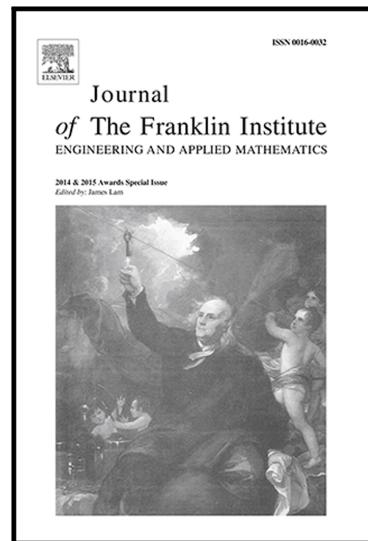
### Fault Modeling of General Momentum Exchange Devices in Spacecraft Attitude Control Systems

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# Fault Modeling of General Momentum Exchange Devices in Spacecraft Attitude Control Systems

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## Abstract

This paper investigates the mechanism of various faults in momentum exchange devices. These devices are modeled as a cascade electric motor (EM) - variable speed drive(VSD) system. Considering the mechanical part of the EM and the VSD system, potential faults are reviewed and summarized. Then, a general fault model in a cascade multiplicative structure is established for momentum exchange devices. Based on this general model, various fault scenarios can be simulated and tested, and the possible outputs affected by faults can be appropriately visualized. Specifically, six types of working condition are identified, and the corresponding fault models are constructed. Using these fault models, the control responses using reaction wheels and single gimbal control moment gyros under various fault conditions can be demonstrated. The simulation results show the severities of the faults and demonstrate that the additive fault is more serious than the multiplicative fault from the viewpoint of control accuracy. Finally, existing fault-tolerant control strategies are briefly summarized and potential approaches including both passive and active ones accommodating gimbal fault of single gimbal control moment gyro are demonstrated.

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## 1. Introduction

Momentum exchange devices (MEDs) have significant advantages of cleanliness, without the expulsion of gases, over thrusters. In addition, these devices came always with small volume and light weight. Thus they have been widely employed in spacecraft attitude determination and control system (ADCS) [1, 2, 3, 4]. Among all momentum exchange devices, the reaction wheel (RW) is the primary attitude control actuator due to its mechanical simplicity and low cost. However, most RWs only provide less than 1 N·m maximum torque that is much smaller than control moment gyros (CMGs) with 100 – 5000 N·m maximum torque [5]. Thus RWs are replaced by CMGs in the agile spacecraft for rapid maneuver, such as Pleiades [6] and Wordview-2 [7]. However, failures of these momentum exchange devices occur occasionally in practical missions. For instance, four RWs of the Far Ultraviolet Spectroscopy Explorer (FUSE) spacecraft failed in 2001, 2002, 2004, and 2007, respectively. Recently, one failed control moment gyro prevented the spacecraft WorldView-4 from pointing accurately in 2019. To accommodate MED fault in spacecraft attitude control system, it is important to understand MED fault mechanism and develop fault model.

In existing fault-tolerant researches, most of works such as [8, 9, 10, 11, 12, 13, 14, 15, 16, 17] focus on control problem itself instead of fault mechanism and modeling. In [10] and [11], four potential faults (recoverable) and/or failures (irrecoverable) of RWs are briefly introduced. However, modeling fault in additive way and multiplicative way is not clear. For the CMG-actuated spacecraft, the fault model of CMG is unclear and the fault-tolerant result is rare. In [12], the skew angle of CMG configuration is analyzed and a genetic algorithm is adopted to simultaneously tune the skew angle and controller gains to achieve fault tolerance. In [13], Yue et al. demonstrated the controllability

of the spacecraft with two parallel SGCMG and gave underactuated control strategies to spacecraft. In [14, 15] and [16], fault-tolerant control under some state constraints are investigated. In [17], the system performance during the control process is considered and fault-tolerant control is achieved. Comparing with these existing works, this paper gives insight into the fault of MEDs and explains clearly why the fault of the reaction wheel can be modeled in additive and multiplicative way. Then, we generalize the RW fault model to a wide range of MEDs, especially the single gimbal control moment gyros. The contribution of this paper paves the way of developing fault-tolerant control strategies for the CMG-actuated spacecraft. Since this work focuses on the fault modeling instead of the fault-tolerant controller design, the description of the fault-tolerant controller design is just to describe the potential implementation of our fault model.

For MED fault modeling, MEDs is considered as a cascade combination of an electric motor (EM) and its variable speed drive (VSD) from systematic point of view. More specifically, a RW is a flywheel mounted to an electric brushless DC motor (BLDC) [18, 19] and the torque is generated through wheel's acceleration or deceleration. For CMGs, a momentum wheel is mounted on one or two gimbals containing two kinds of motors: stepper motor and BLDC motor [20]. The stepper motor provides precision gimbal control of CMGs while the BLDC motor provides an efficient way of driving the momentum wheel to store the angular momentum. Thus, the RW is a one-EM-VSD-loop system and the CMG can essentially be regarded as a cascade combination of two (SGCMG) or three (DGCMG) EM-VSD loops. All potential faults of RW and CMGs would lie in the mechanical part of the EM, sensors and actuators of VSD, or the electrical part of these components.

The details of potential faults in EM-VSD are analyzed in [20, 21, 22, 23, 24, 25, 26] and [27]. For the EM, potential faults are categorized into: stator faults, rotor faults, eccentricity-related faults and bearing or gear faults. These faults belong to multiplicative faults. In the VSD, the faults can be categorized into sensor faults and actuator faults. These faults are considered as additive

faults [24, 28]. The schematic diagram of EM-VSD system containing the potential fault is shown in Fig. 1, where  $f_a(t)$ ,  $f_c(t)/a_{ij}(t)$  and  $f_s(t)$  represent the parameter errors of actuators, electric motors and sensors caused by fault and the detailed explanations can be found in Section 2 to Section 4.

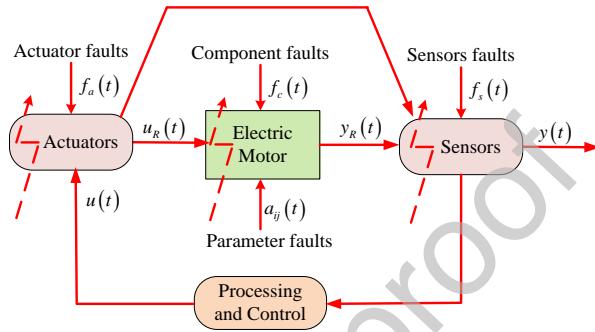


Figure 1: Fault model of EM-VSD system

Motivated by the aforementioned observations, this paper investigates the potential faults in the EM-VSD system, which are categorized into multiplicative or additive fault through analyzing an EM-VSD model. The momentum exchange devices are considered as being in a cascade mechanical structure, in which an EM-VSD system governs one degree of control freedom and works independently. Based on this model, and considering potential faults in the EM-VSD system, a general fault model for momentum exchange devices is then established. To the best of the authors' knowledge, this is the first attempt to propose a generalized fault model to a wide range of momentum exchange devices with a clear understanding of mechanical mechanism. With the utilization of this model, the contribution in our work makes it possible to describe the potential fault scenarios efficiently and effectively, and removes the existing barrier lacking of a unified fault model in developing the necessary fault-tolerant controllers for CMG-actuated spacecraft system. Simulations of the control responses of the RW and SGCMG actuated spacecraft under various fault scenarios visualize the severity of the multiplicative and additive faults

and show that the additive fault is more serious than the multiplicative fault  
 80 from the viewpoint of control accuracy. Finally, existing fault-tolerant control strategies are briefly summarized and potential fault-tolerant strategies based on the proposed fault model to accommodate the gimbal fault of SGCMGs are demonstrated to describe the potential implementation of our proposed model.

The rest of this paper is organized as follows. Section 2 presents the EM  
 85 system, the potential fault of the EM and its fault model. Section 3 addresses the potential fault of sensors and actuators, as well as the corresponding fault models. Combining fault models of the EM and VSD, the overall structure of the EM-VSD system and the fault model is established in Section 4. Based on the fault model of the EM-VSD, a general fault frame of momentum exchange  
 90 devices is established in Section 5. Various fault scenarios in different momentum exchange devices are then modeled through choosing different parameters in fault model. In Section 6, simulations are conducted to demonstrate system performance in the presence of different faults, illustrating the appropriate suitability and applicability of this general fault frame. In Section 7, the fault-tolerant control strategies are summarized and potential approaches to handle  
 95 the gimbal faults of the SGCMGs are given. Finally, conclusions are noted in Section 8.

## 2. EM Fault Modeling

### 2.1. Mathematical Model of EM

For the momentum exchange devices, permanent magnet (PM) brushless DC motors (BLDC) are widely employed due to their advantages of high power density, high efficiency, long operating life, noiseless operation, high speed ranges and etc. The BLDC motors come in single-phase, two-phase, and three phase configurations. Among these, the three-phase motor is the most popular one and widely used in industry. The 3-phase electronically commutated BLDC motor drive system is shown in Fig. 2a and the one phase equivalent circuit of BLDC motor is illustrated in Fig. 2b[29]. Without loss of generality and

considering the phase A, the control input is denoted as  $V_a$  and the phase current is denoted as  $I_a$ . The electrical equivalent of the armature coil can be described by a resistance  $R_a$ , a self-inductance  $L_a$  and an induced voltage referring to the back electromotive force (emf)  $E_{ma}$  which opposes the voltage source. The relationship can be modeled as:

$$V_a = E_{ma} + R_a I_a + L_a \frac{dI_a}{dt}, \quad (1a)$$

$$E_{ma} = K_E \omega_r, \quad (1b)$$

where  $K_E$  is the back-emf constant and  $\omega_r$  is the angular velocity of the rotor.

Using the similar method as in [26] and taking the mean value, the electrical subsystem of the BLDC motor can be described as:

$$V = k_E \omega_r + RI + L \frac{dI}{dt}. \quad (2)$$

Performing an energy balance on the system, the sum of the torques of the motor must equal zero. Therefore, we have

$$J_m \ddot{\omega}_r + \sigma \omega_r = T_e - T_l, \quad (3)$$

where  $J_m$  is the inertia of the rotor and the equivalent mechanical load,  $\sigma$  is the viscous friction coefficient, and  $T_l$  is the load torque.  $T_e$  is the electromagnetic torque and expressed as:

$$T_e = K_t I, \quad (4)$$

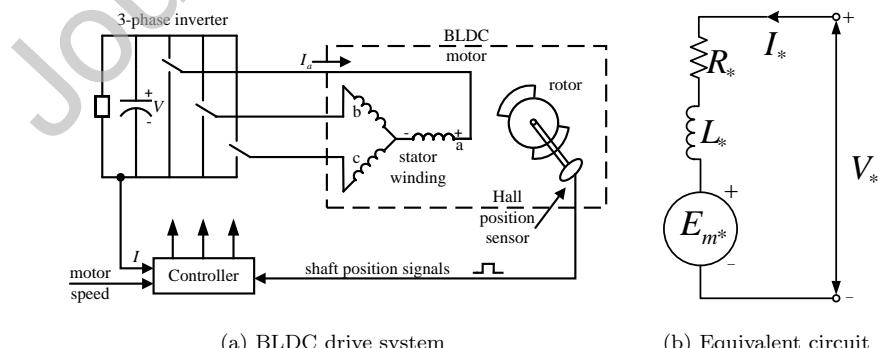


Figure 2: BLDC motor

with  $K_t$  being the torque constant depending on the flux density of the fixed magnets, the reluctance of the iron core, and the number of turns in the armature winding.

Denoting  $x = [I \ \omega_r]^T$  as the state variable and  $V$  as the control input, the state space model of a BLDC motor is obtained as:

$$\begin{cases} \begin{bmatrix} \dot{I} \\ \dot{\omega}_r \end{bmatrix} = \begin{bmatrix} -\frac{R}{L} & -\frac{K_E}{L} \\ \frac{K_t}{J_m} & -\frac{\sigma}{J_m} \end{bmatrix} \begin{bmatrix} I \\ \omega_r \end{bmatrix} + \begin{bmatrix} \frac{1}{L} \\ 0 \end{bmatrix} V + \begin{bmatrix} 0 \\ -\frac{1}{J_m} \end{bmatrix} T_l \\ y = C \begin{bmatrix} I \\ \omega_r \end{bmatrix} \end{cases}, \quad (5)$$

with  $y$  being the measurements.  $C$  is the output matrix that decides the measurement output. For example, the matrix  $C$  could be the identity matrix to measure both current  $I$  and the angular velocity of the rotor  $\omega_r$ , or it could be matrix  $[0 \ 1]$  to measure the angular velocity of the rotor  $\omega_r$  only. For EM-VSD system,  $\omega_r$  is more important, hence we choose matrix  $C$  to be  $[0 \ 1]$ .

Denoting

$$A = \begin{bmatrix} -\frac{R}{L} & -\frac{K_E}{L} \\ \frac{K_t}{J_m} & -\frac{\sigma}{J_m} \end{bmatrix}, \quad B = \begin{bmatrix} \frac{1}{L} \\ 0 \end{bmatrix}, \quad D = \begin{bmatrix} 0 \\ -\frac{1}{J_m} \end{bmatrix},$$

and  $u = V$ , equation (5) can be further written into a compact form as:

$$\begin{cases} \dot{x} = Ax + Bu + DT_l \\ y = Cx \end{cases}. \quad (6)$$

Other motors such as the stepper motors have the similar compact form as (6) [30, 31].

## 2.2. Potential Fault of EM

In an EM, there may exist mechanical and electrical faults or failures, or a combination of these mechanical and electrical faults. Specifically, the potential faults may be [20, 23, 27, 32]:

- Stator faults. As stated in [27], the most common stator fault is the breakdown of the winding insulation in the position where the end windings

120

enter the stator slots. It may be due to large electrical voltage stresses, electro-dynamic forces generated by winding currents, thermal aging in multiple heating and cooling cycles, and mechanical vibrations from external and internal sources. This winding insulation breakdown can lead to turn-to-turn faults, and eventually give rise to short circuits to ground.

125

- Rotor faults. The rotor of the BLDC are PM, whose major fault is the damaged rotor magnet [27]. Some permanent magnets corrode [33] and cracks formed during manufacturing [34] can lead to disintegration. The partial demagnetization of the magnets may also influence the magnetic flux density distribution [35]. Other faults may be the broken rotor bar or cracked rotor end-rings [20]. These faults are mainly caused by excess stresses.
- Eccentricity-related faults can be categorized as static and/or dynamic air-gap irregularities. The dynamic eccentricity is the character describing the displacement between the center of the rotor and the center of the rotation. The possible reasons are bent rotor shaft, bearing wear or misalignment, and mechanical resonance. Static eccentricity may be caused by the ovality of the stator core or incorrect positioning of the rotor or stator. When the eccentricity becomes large, it can result in damage of stator and rotor.
- Bearing and gearbox faults or failures. Bearing failures account for the vast majority of the recorded motor failure [36]. The bearing failures are caused by continued stress, inherent eccentricity, fatigue and other external causes, such as unbalanced load, improper installation, contamination and corrosion, and improper lubrication. This kind of fault may lead to excessive noises and vibrations.

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### 2.3. Fault Model of EM

As addressed in [28, 37], the fault of EM can be modeled as component faults  $f_c(t)$  or parameter faults  $a_{ij}(t)$  as shown in Fig. 3.

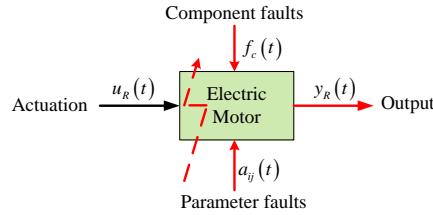


Figure 3: Fault model of electric motor

The component fault occurs when some condition changes in the system. In some other cases, the faults can be expressed as a change in the system parameter. Then the mathematical fault model can be constructed as:

$$\dot{x} = Ax + Bu + DT_l + f_c(t) \quad (7)$$

<sup>145</sup> OR

$$\begin{aligned} \dot{x} &= Ax + Bu + DT_l + \sum \left( \sum a_{ij} x_j \right) e_i \\ &= (A + \Delta A)x + Bu + DT_l, \end{aligned} \quad (8)$$

where  $e_i$  is the  $i$ th basis vector, and  $\Delta A$  is the discrepancy of the state-transition matrix caused by parameter faults. Indexes  $i, j$  are related to the EM system and can be determined by fault diagnosis. Fault models (7) and (8) are equivalent in describing the component fault and the system matrix  $A$  is influenced by the component faults.

Solving the foregoing equation (8) from the time instant  $k$  to  $k+1$  with a constant control input  $u(k)$ , we obtain

$$x_R(k+1) = e^{(A+\Delta A)h} x_R(k) + e^{(A+\Delta A)h} \left\{ \int_0^h e^{-(A+\Delta A)t} dt \right\} (Bu(k) + DT_l), \quad (9)$$

where  $h$  is the step size,  $x_R(k+1)$  and  $x_R(k)$  are real states at the time instants  $k+1$  and  $k$ , respectively, and the subscript “R” represents real value rather than the measured one. Solving  $\int_0^h e^{-(A+\Delta A)t} dt$  and using the first order approximation

$e^{-(A+\Delta A)h} \approx I - (A + \Delta A)h$ , we have:

$$\int_0^h e^{-(A+\Delta A)t} dt \approx -(A + \Delta A)^{-1} [e^{-(A+\Delta A)h} - I] \approx h. \quad (10)$$

Moreover, substituting (10) into (9), it follows that

$$x_R(k+1) \approx e^{\Delta Ah} \{ e^{Ah} [x_R(k) + (Bu(k) + DT_l)h] \}. \quad (11)$$

To compensate the load torque and cancel the influence of the system state  $x_R(k)$ , we assume the control torque  $u(k)$  can be designed as:

$$u(k) = \frac{1}{h} u_0(k) - \frac{1}{h} B^+ (x_R(k) + DT_l h), \quad (12)$$

where  $B^+ = B^T(BB^T + \varsigma I)^{-1}$  is the pseudo-inverse of the matrix  $B$  and  $\varsigma$  is a small positive number.

Then, substituting (12) into (11), we obtain

$$x_R(k+1) \approx e^{\Delta Ah} e^{Ah} B u_0(k). \quad (13)$$

When the motor is an one-dimensional speed or torque control system, we have:

$$x_R(k+1) \approx e^{\Delta ah} (e^{ah} b) u_0(k) = \eta g u_0(k) \quad (14a)$$

$$y_R(k+1) = c x_R(k+1) \approx \eta c g u_0(k) = \eta u^+(k) \quad (14b)$$

with  $g = e^{ah} b$  being the transfer function from input  $u_0(k)$  to state  $x_R(k+1)$  without component faults, and  $u^+(k) = c g u_0(k)$  being the equivalent input. In  
155 the following section, the control input  $u(k)$  will refer to this equivalent input  $u^+(k)$ . The term  $e^{\Delta ah}$  is denoted as the effectiveness factor  $\eta$  describing the component faults, and it is constrained in the interval  $\eta \in [0, 1]$  practically.  
160 Different  $\eta$  values correspond to different scenarios. For example, a)  $\eta = 1$ , EM works normally and no fault occurs; b)  $0 < \eta < 1$  refers to malfunctions, in which EM partially loses effectiveness, but not fail totally; c)  $\eta = 0$  denotes a complete failure.

It is clearly observed from (14b) that component fault of EM can be represented by a multiplicative effectiveness factor  $\eta$ . The result in this section

establishes the mathematical foundation to describe the component fault in a  
 165 multiplicative way.

### 3. VSD System Fault Modeling

#### 3.1. Introduction of VSD and Potential Faults

To effectively and precisely drive the EM, VSD system is essential. It consists of the EM as the plant, sensors to measure the system state and output signal,  
 170 processing and control component to generate control command, and actuators to drive the EM.

In the VSD system, sensors are adopted to measure the motor velocity, position and the output voltage of the inverter. Sensors can be mechanical or electrical, and the reliability of mechanical sensor is lower than the electrical  
 175 one due to mechanical complexity [24]. For the BLDC, the sensors may be the Hall position sensor and the electrical tachometer. The loss of a Hall sensor results in torque pulsations when the rotor is moving.

Actuators in VSD are inverters which convert DC electricity to AC electricity since almost all PM motors are inverter-fed. Various faults can occur in the  
 180 inverter, such as the loss of one or more switches of a phase, the short circuit of a switch, or the opening of one of the lines to the machine [20].

#### 3.2. Fault Model of Sensors and Actuators in VSD

Fault of sensors and actuators are often modeled as the additive fault as shown in Fig. 4. As described in Fig. 4a, the parameter  $f_s(t)$  denotes the sensors faults. Then the output of a faulty sensor is obtained as:

$$y(t) = y_R(t) + f_s(t). \quad (15)$$

All sensors' faults can be described by choosing a proper  $f_s(t)$  vector. For example, when the output gets stuck at a particular value  $a$ ,  $f_s(t)$  can be chosen  
 185 as  $a - y_R(t)$ .

Similar to the sensors fault, the real actuation  $u_R(t)$  and the actuator command  $u(t)$  are connected via the actuator fault vector  $f_a(t)$  as:

$$u_R(t) = u(t) + f_a(t). \quad (16)$$

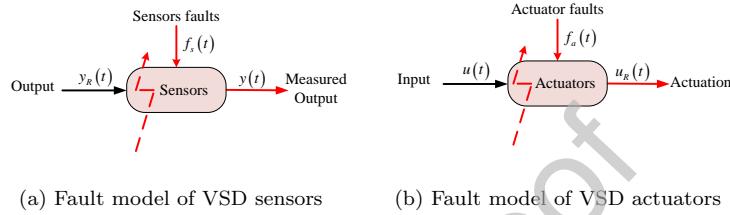


Figure 4: Fault model of sensors and actuators

It is observed that actuator will generate an additional control command to feed the plant after fault occurs. When the sensor is faulty, the output will have a direct bias affecting nominal measurement. Thus, the fault in actuators and sensors belongs to the additive category. This section establishes the foundation 190 of modeling the sensor and actuator fault in an additive way.

#### 4. Fault Model of EM-VSD System

Considering all potential faults in EM, sensors and actuators, the schematic diagram of the overall EM-VSD system is shown in Fig. 1. Based on the analysis in previous two sections, the EM suffers the multiplicative fault represented by  $\Delta A$ , while the sensors and actuators in VSD system may occur the additive fault  $f_s(t)$  and  $f_a(t)$ , respectively. Then the state space model of the EM-VSD system in the presence of faults/failures can be constructed as:

$$\begin{cases} \dot{x} = (A + \Delta A)x + B[u + f_a(t)] + DT_l \\ y = Cx + f_s(t) \end{cases}. \quad (17)$$

In view of the output relationship (14), and substituting the potential fault model (15) and (16) into (17), we get the output of the overall system as:

$$y = \eta(u^+ + f_a(t)) + f_s(t). \quad (18)$$

Denoting the offset as  $y_o = \eta f_a(t) + f_s(t)$ , which is on the basis of precise output  $y$  under the nominal control signal  $u^+$ , equation (18) is further written as:

$$y = \eta u^+ + y_o. \quad (19)$$

Now, we have developed the fault model of an EM-VSD system as presented in (19). Considering the fact that the EM-VSD system is working either in the speed control mode or in the torque control mode, a general fault frame of an EM-VSD system in speed control mode or torque control mode can be obtained as:

$$\begin{cases} \omega = \eta_\omega \omega_c + \omega_o, & \text{Speed control mode} \\ T = \eta_T T_c + T_o, & \text{Torque control mode} \end{cases} \quad (20)$$

where  $\omega_c$  and  $T_c$  are the angular velocity control command and the torque control command corresponding to the speed control mode and the torque control mode, respectively,  $\eta_\omega$  and  $\eta_T$  are the effectiveness factors representing the component fault of the EM in speed control mode and torque control mode, similar to the loss-of-effectiveness faults as in [38],  $\omega_o$  and  $T_o$  are the additive measurement offset as a combination of the actuator fault and the sensor fault in VSD system, similar to the actuator bias faults as in [39], and  $\omega$  and  $T$  are the measured angular velocity and output torque.

## 200 5. General Fault Model of Momentum Exchange Devices

The most commonly used momentum exchange devices in aerospace mission are RWs and CMGs. CMGs can be further divided into SGCMG, DGCMG and VSCMG. The differences among these categories are the number of degree of control freedom and the working mode of flywheels. Specifically, the RW and SGCMG are single degree-of-freedom devices, while DGCMG and VSCMG have two and three degree-of-freedom, respectively. From the working mode of flywheels perspective, SGCMGs and DGCMGs hold their flywheel speed at a constant value [40], whereas the the rotor speed of RWs and VSCMG are time-varying.

Although RWs and CMGs have different working principles, these momentum exchange devices can be modeled by a series of cascade EM-VSD systems driving the wheel and the gimbal frame, separately. As stated in [41], the dynamics of the CMG gimbal is independent of rotor momentum for the case of a very stiff gimbal. In parallel, we assume that dynamics of gimbal are working independent. In each control loop, the potential fault and the corresponding model are given in Section 2 and 3. When they are integrated as a cascade instrument, there will be a high dependence among each system, and the overall fault model will be in a multiplicative form. Consequently, the general fault model of a momentum exchange device is given by:

$$y = \sum_{j=1}^m \left\{ \prod_{i_j=1}^{n_j} [\eta^{i_j} u_c^{i_j} + y_o^{i_j}] \right\}, \quad (21)$$

where the superscript  $i_j$  means the  $i$ th element in series of the  $j$ th parallel term,  $m$  and  $n_j$  represents the total number of parallel terms and the total number of term in series of the  $j$ th parallel term,  $u_c^{i_j}$  is the nominal control command,  $y_o^{i_j}$  is the offset caused by fault, failure and malfunctions, and  $\eta^{i_j} \in [0, 1]$  is the control effectiveness factor. In the following, by using the general fault model of the momentum exchange devices proposed in (21), we give the specific fault model of the RW, SGCMG, DGCMG and VSCMG, respectively.

### 5.1. RW Fault Model

RW contains a rotating flywheel, an internal BLDC motor as well as associated electronics [40]. RW can be regarded as a single loop EM-VSD system. It is always fed by control commands in order to generate the desired control torque via acceleration or deceleration. Therefore, the whole loop is considered as the acceleration control loop. When the RW works in the torque control mode, the fault model can be written as:

$$\tau_{rw} = \eta_{rw} \tau_c + \tau_o. \quad (22)$$

where  $\tau_c$  is the command torque,  $\eta_{rw}$  is the effectiveness of RW,  $\tau_o$  is the output offset due to fault and  $\tau_{rw}$  is the real output. This equation is consistent with

- <sup>220</sup> the fault model of RW-actuated spacecraft system in existing literature, such as [11], [42] and [43].

### 5.2. SGCMG Fault Model

SGCMG contains a spinning rotor mounted on a gimbal. In nominal condition, the rotor holds a constant speed using a BLDC motor, while the gimbal is manipulated to change the direction of angular momentum by a stepper motor. As a result, a gyroscopic reaction torque orthogonal to both the rotor spin and gimbal axes is generated. With a small input of the gimbal, a larger control torque is produced to act on the spacecraft, which is the so-called torque amplification characteristic. More specifically, the torque is proportional to both the angular momentum and gimbal angular rate, and can be calculated as:

$$\tau = -h_0 \dot{\delta} \hat{t}, \quad (23)$$

- where  $h_0$  is the constant angular momentum of the spinning rotor determined by the moment inertia of the wheel  $J_{sc}$  and motor speed  $\omega_{sc}$  as  $h_0 = J_{sc}\omega_{sc}$ ,  $\delta$   
<sup>225</sup> is the gimbal angle, and  $\hat{t}$  is a unit vector in the direction of output torque. The symbol “-” in (23) means the output torque lies in the opposite direction of  $\hat{t}$ .

The SGCMG is considered as a combination of two EM-VSD systems. The potential fault of SGCMG may exist in the rotor control loop and the gimbal frame control loop. We denote the rotor's EM-VSD control loop as the first degree-of-freedom that is marked with a superscript “r”, while the control loop of gimbal frame is considered as the second degree-of-freedom with a superscript “g”. According to the independence assumption, the dynamics of rotor and gimbal will not influence each other in their own control loop. Then for the rotor, the angular momentum is the product of moment of inertia and the spindle speed, i.e.  $h_0 = J_{sc}\omega_{sc}$ , and the motor speed is controlled by the VSD system. Considering possible faults in the rotor control loop, the angular momentum can be computed as:

$$h_0 = \eta^r J_{sc}\omega_c + h_o, \quad (24)$$

where  $h_o$  is the output offset. When the flywheel works normally, the output  $h_0$  equals to the command angular momentum  $h_c = J_{sc}\omega_c$ . Considering the control loop of gimbal, its fault model is expressed as:

$$\dot{\delta} = \eta^g \dot{\delta}_c + \dot{\delta}_o \quad (25)$$

with  $\dot{\delta}_c$  and  $\dot{\delta}_o$  being the commanded gimbal rate and the gimbal rate offset caused by fault. Substituting equations (24) and (25) into (23), we obtain the fault model of a SGCMG as

$$\tau = -[\eta^r J_{sc}\omega_c + h_o] [\eta^g \dot{\delta}_c + \dot{\delta}_o] t. \quad (26)$$

Noting that the saturation is not treated as fault since it is a kind of actuator physical constraints. Based on (26), it is clear that different combinations of  $\eta^r, \eta^g, h_o$  and  $\dot{\delta}_o$  represent the different fault scenarios in a SGCMG. To illustrate 230 all potential faults and fault-free situation of a SGCMG, the potential fault scenarios in each single control loop (either rotor control loop or gimbal frame control loop) are defined as:

$N$  : Nominal working condition;

$F_a$  : Partially lose effect, without offset;

235  $F_b$  : Totally fail, without offset;

$F_c$  : Partially lose effect and have offset;

$F_d$  : Totally fail and have offset;

$F_e$  : Pure offset without losing effect.

Combining these different work conditions, the fault model list of SGCMG 240 is obtained in Table 1. This model is consistent with that in [44] that only takes the gimbal fault into account. To the best knowledge of authors, this is the first attempt to establish a systematic fault model of a SGCMG in literature.

Table 1: Work condition and fault model of SGCMG

Rotor	Gimbal	Model
$N, F_d$	$N$	$-h_* \dot{\delta}_c \hat{t}$
	$F_a$	$-h_* [\eta^g \dot{\delta}_c] \hat{t}$
	$F_b$	0
	$F_c$	$-h_* [\eta^g \dot{\delta}_c + \dot{\delta}_o] \hat{t}$
	$F_d$	$-h_* \dot{\delta}_o \hat{t}$
	$F_e$	$-h_* [\dot{\delta}_c + \dot{\delta}_o] \hat{t}$
$F_a$	$N$	$-[\eta^r J_{sc} \omega_c] \dot{\delta}_c \hat{t}$
	$F_a$	$-[\eta^r J_{sc} \omega_c] [\eta^g \dot{\delta}_c] \hat{t}$
	$F_b$	0
	$F_c$	$-[\eta^r J_{sc} \omega_c] [\eta^g \dot{\delta}_c + \dot{\delta}_o] \hat{t}$
	$F_d$	$-[\eta^r J_{sc} \omega_c] \dot{\delta}_o \hat{t}$
	$F_e$	$-[\eta^r J_{sc} \omega_c] [\dot{\delta}_c + \dot{\delta}_o] \hat{t}$
$F_b$	$N, F_a, F_b$ $F_c, F_d, F_d$	0
$F_c$	$N$	$-[\eta^r J_{sc} \omega_c + h_o] \dot{\delta}_c \hat{t}$
	$F_a$	$-[J_{sc} \text{sat}_{\omega_m} (\eta^r \omega_c) + h_o] [\eta^g \dot{\delta}_c] \hat{t}$
	$F_b$	0
	$F_c$	$-[\eta^r J_{sc} \omega_c + h_o] [\eta^g \dot{\delta}_c + \dot{\delta}_o] \hat{t}$
	$F_d$	$-[\eta^r J_{sc} \omega_c + h_o] \dot{\delta}_o \hat{t}$
	$F_e$	$-[\eta^r J_{sc} \omega_c + h_o] [\dot{\delta}_c + \dot{\delta}_o] \hat{t}$
$F_e$	$N$	$-[J_{sc} \omega_c + h_o] \dot{\delta}_c \hat{t}$
	$F_a$	$-[J_{sc} \omega_c + h_o] [\eta^g \dot{\delta}_c] \hat{t}$
	$F_b$	0
	$F_c$	$-[J_{sc} \omega_c + h_o] [\eta^g \dot{\delta}_c + \dot{\delta}_o] \hat{t}$
	$F_d$	$-[J_{sc} \omega_c + h_o] \dot{\delta}_o \hat{t}$
	$F_e$	$-[J_{sc} \omega_c + h_o] [\dot{\delta}_c + \dot{\delta}_o] \hat{t}$

Note:  $\eta^r, \eta^g \in (0, 1)$  in this table;  $h_* = h_c$ , command angular momentum for working condition  $N$ ;  $h_* = h_o$ , pure angular momentum offset for working condition  $F_d$ .

### 5.3. DGCMG Fault Model

Double gimbal control moment gyro is a flywheel with a constant angular speed mounted on two orthogonally installed gimbal frames. Different from the SGCMG, a DGCMG has two gimbals including the inner gimbal  $\hat{g}_i$  and outer gimbal  $\hat{g}_o$ . The two gimbals axes together with the direction of angular momentum  $\hat{h}$  form the CMG frame, denoted as  $\mathbb{G} = \{\hat{h}, \hat{g}_i, \hat{g}_o\}$ . Therefore, the DGCMG can be modeled by three EM-VSD systems working independently. Similar to the mechanism of SGCMG, the output torque is generated by rotating gimbals [41]. Based on the independence assumption, the torques generated by inner gimbal and outer gimbal are independent. Then the nominal overall control torque of a DGCMG can be expressed as [45]:

$$\tau = \tau_i + \tau_o = -h_0 \dot{\delta}_i \hat{t}_i - h_0 \dot{\delta}_o \hat{t}_o, \quad (27)$$

where  $\tau_i$  and  $\tau_o$  are the torques generated by the inner gimbal loop and outer gimbal loop,  $\dot{\delta}_i$  is the gimbal rate of the inner gimbal,  $\dot{\delta}_o$  is the gimbal rate of the outer gimbal, and  $\hat{t}_i$  and  $\hat{t}_o$  are the directions of the inner and outer output torque. Then, for the inner gimbal and outer gimbal, their fault models are given by:

$$\tau_i = -[\eta^r J_{dc} \omega_c + h_o] [\eta^{g_i} \dot{\delta}_{c_i} + \dot{\delta}_{o_i}] \hat{t}_i \quad (28)$$

and

$$\tau_o = -[\eta^r J_{dc} \omega_c + h_o] [\eta^{g_o} \dot{\delta}_{c_o} + \dot{\delta}_{o_o}] \hat{t}_o, \quad (29)$$

respectively. The subscript “dc” represents DGCMG, “*i*” represents inner gimbal and “*o*” represents outer gimbal.  
245

### 5.4. VSCMG Fault Model

VSCMG can be considered as a combination of RW and CMGs. It has two major categories: single gimbal VSCMG (SGVSCMG) and double gimbal VSCMG (DGVSCMG). The fault model of these two types of VSCMG are addressed in the following.  
250

#### a). SGVSCMG

This SGVSCMG is a combination of a RW and a SGCMG, so it has two working modes, i.e., RW working mode and SGCMG working mode. The output torque is generated by the acceleration in the direction of angular momentum working as the RW and the gimbal rotation working as the SGCMG [46, 47]. The nominal output torque of a SGVSCMG can be expressed as:

$$\tau = \tau_r + \tau_c = -J_{svs}\dot{\omega}_{svs}\hat{t}_r - h(\omega_{svs})\dot{\hat{t}}_g, \quad (30)$$

where  $\tau_r$  and  $\tau_c$  are output torque generated in the RW working mode and the SGCMG working mode, and the subscript “svs” represents SGVSCMG. Considering the potential fault in RW, the fault model of the RW working mode is:

$$\tau_r = -[\eta^{r_{svs}} J_{svs}\dot{\omega}_{c_{svs}} + \tau_{o_{svs}}]\hat{t}_r. \quad (31)$$

When the SGVSCMG is in the SGCMG working mode, only the measurement of angular momentum, or the angular velocity equivalently, is used to design the gimbal rate command. Thus we just need to consider the additive fault caused by the sensors in the rotor control loop when SGVSCMG is in SGCMG working mode. Then the fault model of the SGCMG working mode can be established as:

$$\tau_c = -[J_{svs}\omega_{c_{svs}} + h_{o_{svs}}]\left[\eta^g\dot{\delta}_c + \dot{\delta}_o\right]\hat{t}_g. \quad (32)$$

#### b). DGSCMG

The DGSCMG can be regarded as a combination of the RW and the DGCMG. The output torque contains three parts: 1)  $\tau_r$  caused by rotor acceleration; 2)  $\tau_{g_i}$  generated by inner gimbal rotation; and 3)  $\tau_{g_o}$  caused by outer gimbal rotation. The output torque can be expressed as[45, 48]:

$$\tau = \tau_r + \tau_{g_i} + \tau_{g_o} = -J_{dvs}\dot{\omega}_{dvs}\hat{t}_r - h(\omega_{dvs})\dot{\delta}_{g_i}\hat{t}_{g_i} - h(\omega_{dvs})\dot{\delta}_{g_o}\hat{t}_{g_o}, \quad (33)$$

where the subscript “dvs” represents DGVSCMG. The fault model of RW working mode is same as equation (31). The faults relating to the inner and outer gimbals of DGCMG can be described as

$$\tau_{g_i} = -[J_{dvs}\omega_{c_{dvs}} + h_{o_{dvs}}]\left[\eta^{g_i}\dot{\delta}_{i_c} + \dot{\delta}_{i_o}\right]\hat{t}_{g_i} \quad (34)$$

and

$$\tau_{g_o} = -[J_{dvs}\omega_{c_{dvs}} + h_{o_{dvs}}] \left[ \eta^{g_o} \dot{\delta}_{o_c} + \dot{\delta}_{o_o} \right] \hat{t}_{g_o}. \quad (35)$$

## 6. Simulation Results

This section demonstrates attitude control results of RWs or SGCMGs actuated spacecraft under different fault scenarios. The severities of faults are qualitatively analyzed to give a guideline for fault-tolerant control system design.

### 6.1. Fault Effects in EM-VSD

According to the failure analysis reported in [49], 32% of spacecraft failure come from the Attitude and Orbit Control System (AOCS). Among the failures of AOCS, 44% of them are related to actuators including Thrusters, RWs, CMG and XIPS. For the failure type, more than half of the failures (54%) are mechanical, and a relative small portion (20%) are electrical. The detailed analysis are shown in Fig. 5.

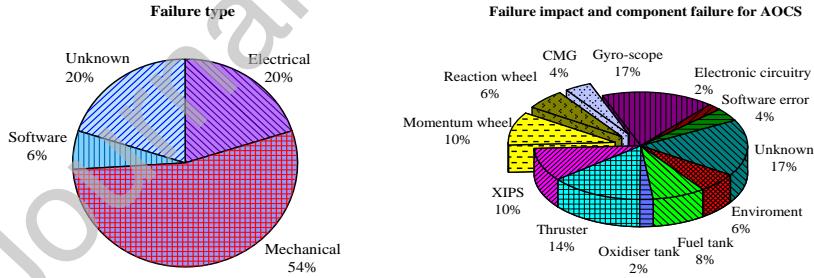


Figure 5: AOCS failure distributions [49]

To demonstrate the EM-VSD performance under different fault conditions, all the listed potential faults  $F_a$  to  $F_e$  in Section 5.2 are compared with the nominal condition. Taking the practical variation of possible fault and failure into consideration, an exponential function  $u_{out} = \eta + (1 - \eta)e^{-t_a(t - t_c)}$  is adopted to

describe the dynamic characteristic of the multiplicative fault, and this exponential function is also used in the SGCMG-actuated spacecraft simulations. In this equation,  $\eta$  represents the effectiveness factor of the motor after the occurrence of fault or failure,  $t_a$  represents the time constant for the fault or failure, and  $t_c$  is the time instant at which fault or failure happens. In the simulation, the time constant  $t_a$  is chosen to be 2 and 1 for fault and failure, respectively.

In the nominal condition  $N$ , the output torque is set as 0.4 Nm. The actuator is assumed to lose its effectiveness at  $t = 5$  s and the bias offset is added at  $t = 15$  s. As shown in Fig. 6, the effectiveness factor  $\eta$  is chosen as 0.75 and the offset is chosen as 0.04, which is 10% of the nominal command. These parameters are

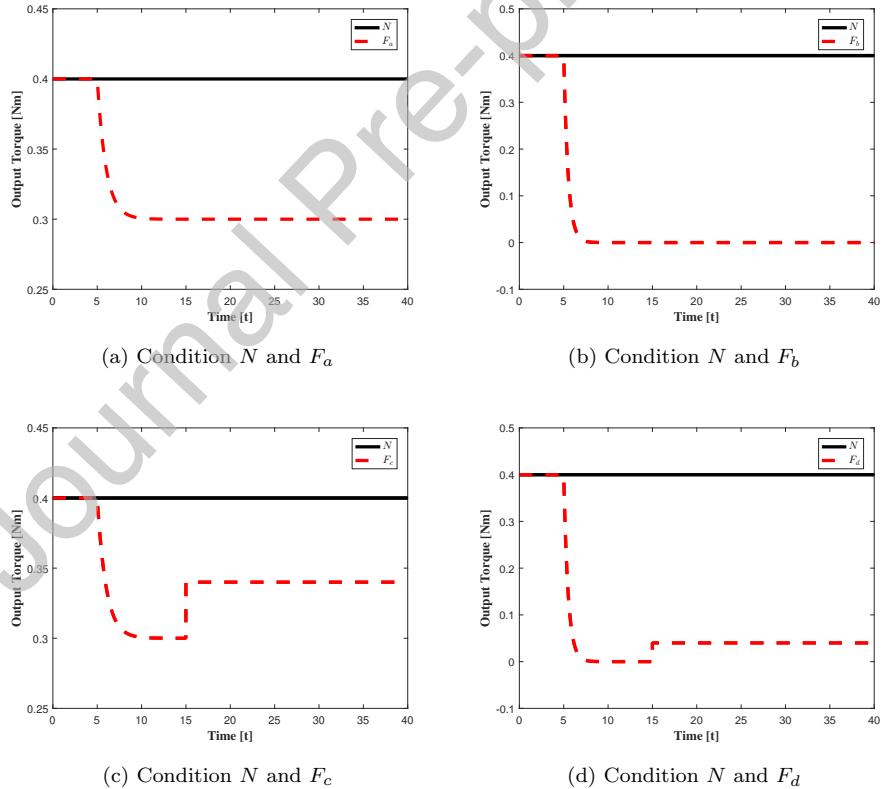


Figure 6: Fault scenarios of RWs

also used in the RW-actuated spacecraft simulation. For the working condition  
 280  $F_e$ , it is similar to part of the Fig. 6c and Fig. 6d when the offset is added, so  
 the demonstration is omitted here.

### 6.2. Attitude Control Results of RW-Actuated Spacecraft

Before we move forward to the demonstration of the control performance of  
 RW-actuated and CMG-actuated spacecraft under actuator faults, the dynamics  
 and kinematics of the attitude control are [50]

$$\begin{cases} \dot{\mathbf{q}} = \frac{1}{2} \begin{bmatrix} -\mathbf{q}_v^T \\ q_o \mathbf{I}_3 + \mathbf{q}_v^\times \end{bmatrix} \boldsymbol{\omega} \\ \mathbf{J} \dot{\boldsymbol{\omega}} + \boldsymbol{\omega}^\times (\mathbf{J} \boldsymbol{\omega} + \mathbf{H}) = \boldsymbol{\tau}_e + \mathbf{d} \end{cases} \quad (36)$$

where  $\mathbf{q} = [q_0, \mathbf{q}_v^T]^T$  is the unit quaternion,  $\boldsymbol{\omega}$  is the angular velocity of the  
 spacecraft,  $\mathbf{J} = \mathbf{J}^T$  is moment of inertia of spacecraft,  $\mathbf{H}$  is the angular mo-  
 285 mentum of actuators and  $\mathbf{d}$  is external disturbance.  $\boldsymbol{\tau}_e = -\dot{\mathbf{H}}$  is the output of  
 actuators.

To have a better demonstration the influence of the RW fault/failure to  
 attitude control system, we do not consider RW redundancy in the simulation.  
 As a result, the installation matrix of the RWs is identity. Considering the fault  
 model of the RW given by (22), the output of the RW cluster is given by:

$$\boldsymbol{\tau} = \mathbf{E}_\eta \mathbf{u} + \mathbf{E}_o \quad (37)$$

where  $\mathbf{u}$  is the control command calculated by the controller,  $\mathbf{E}_\eta = \text{diag}[\eta_1, \eta_2, \eta_3]$   
 is the actuator effectiveness matrix and  $\mathbf{E}_o = \text{diag}[u_{o1}, u_{o2}, u_{o3}]$  is the offset  
 caused by RW faults.

The objective of the attitude control of this simulation is to stabilize the  
 attitude and angular velocity such that the spacecraft can change its orientation  
 from the initial value to the target. The initial three-axis Euler angle are set  
 to be 60 deg, -20 deg and 30 deg, and the initial angular velocity is zero. The  
 attitude controller in the simulation is the widely used cascade PD control in  
 [3], which is in the form of:

$$\mathbf{u} = -\mathbf{J} \left\{ 2k \underset{L_i}{\text{sat}}(\mathbf{q}_e) + c\boldsymbol{\omega} \right\}, \quad (38)$$

with

$$L_i = (c/2k) \min \left\{ \sqrt{4a_i |q_{ei}|}, |\omega_i|_{\max} \right\}, \quad (39)$$

where  $\mathbf{q}_e$  is the quaternion-error vector obtained by quaternion product of the current attitude  $\mathbf{q}$  and the desired attitude  $\mathbf{q}_d$ ,  $c$  and  $k$  are natural frequency and damping ratio related control gains. The parameters  $a_i = u_{\max}/J_{ii}$  and  $|\omega_i|_{\max}$  are the maximum acceleration and angular velocity along the  $i$ th axis,  $|\omega_i|_{\max}$  is set to be 4 deg/s, and the PD gains are  $k = 9.54$  and  $c = 5.5$ .

The simulation results of the RW-actuated spacecraft under different fault scenarios including the fault-free situation are shown in Figs. 7 to 11. In Fig. 7, the trajectories of Euler angle, angular velocity, control torque and angular

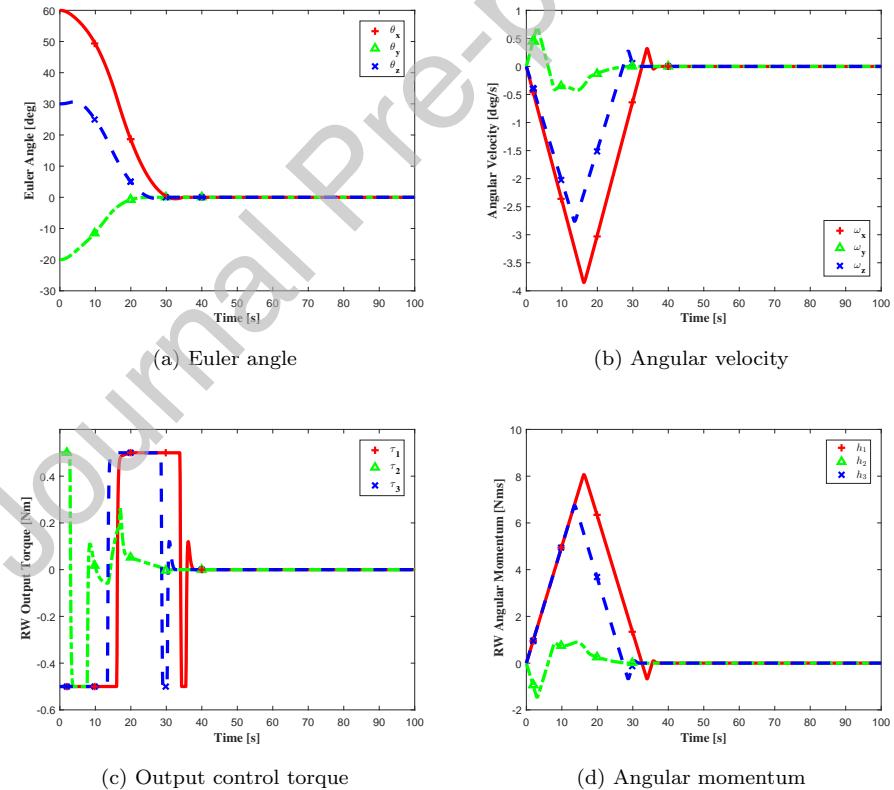
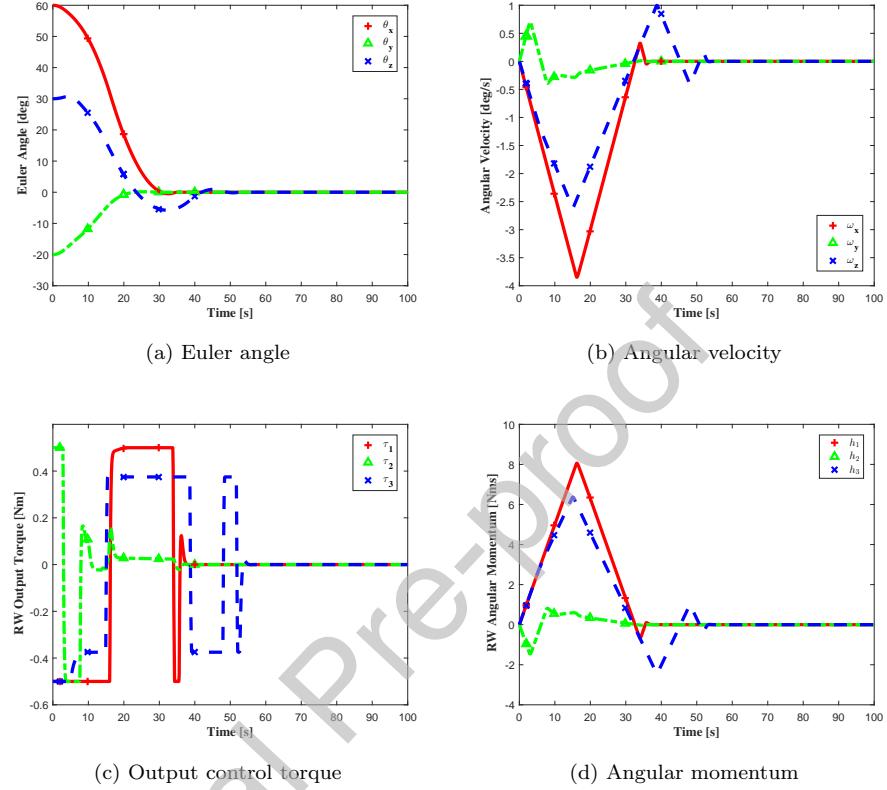
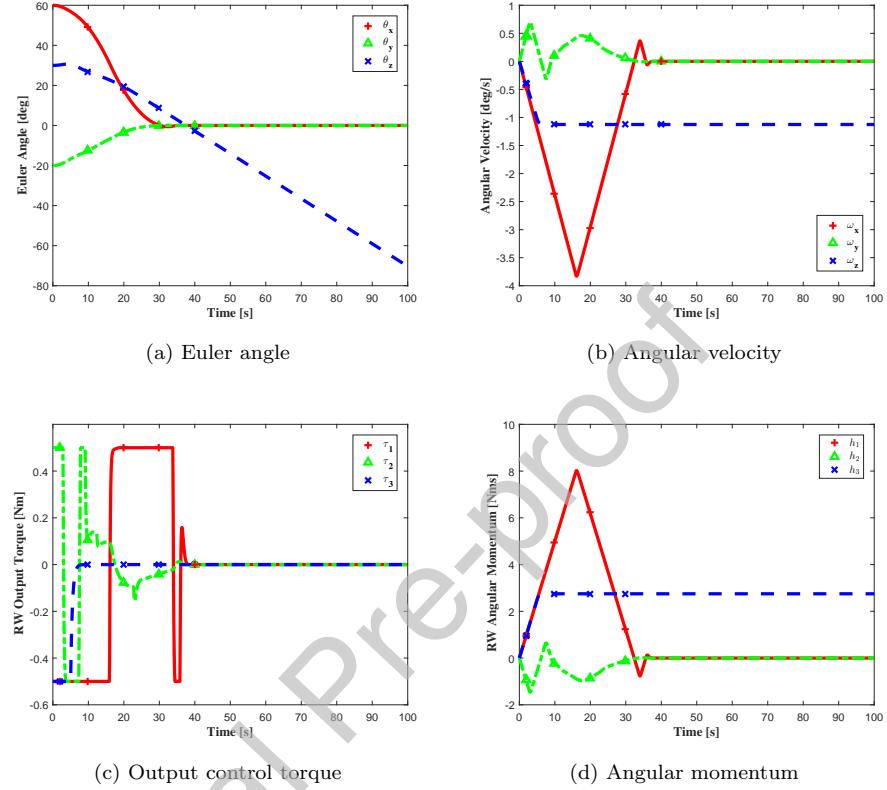


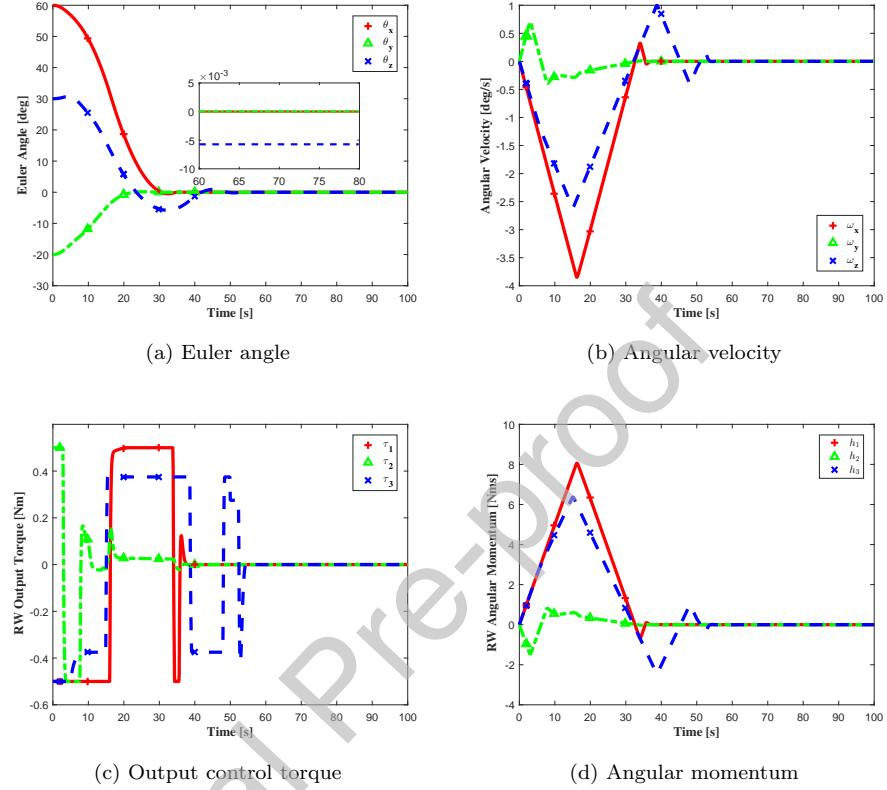
Figure 7: RW-actuated attitude control result under nominal condition

Figure 8: RW-actuated attitude control result under  $F_a$ 

momentum of RWs in the nominal condition are presented. It can be seen from Figs. 7a to 7d that the required maneuver is completed in 40 s. The scenario that the wheel along Z axis partially loses its effectiveness at 5 s (the fault  $F_a$ ) is demonstrated in Fig. 8. Compared to Fig. 7, the whole system is controllable, but the stabilization is achieved in a longer period. It should be mentioned from Fig. 8c that the maximum output torque drops to  $\eta$  times of the command gradually after 5 s. Fig. 9 shows the control result when the third wheel totally fails at 5 s (the failure  $F_b$ ). It is clear that the axis about which the failed RW is installed becomes uncontrollable and the Euler angle diverges as shown in Fig. 9a. Fig. 9b states the angular velocity in Z axis keeps at a

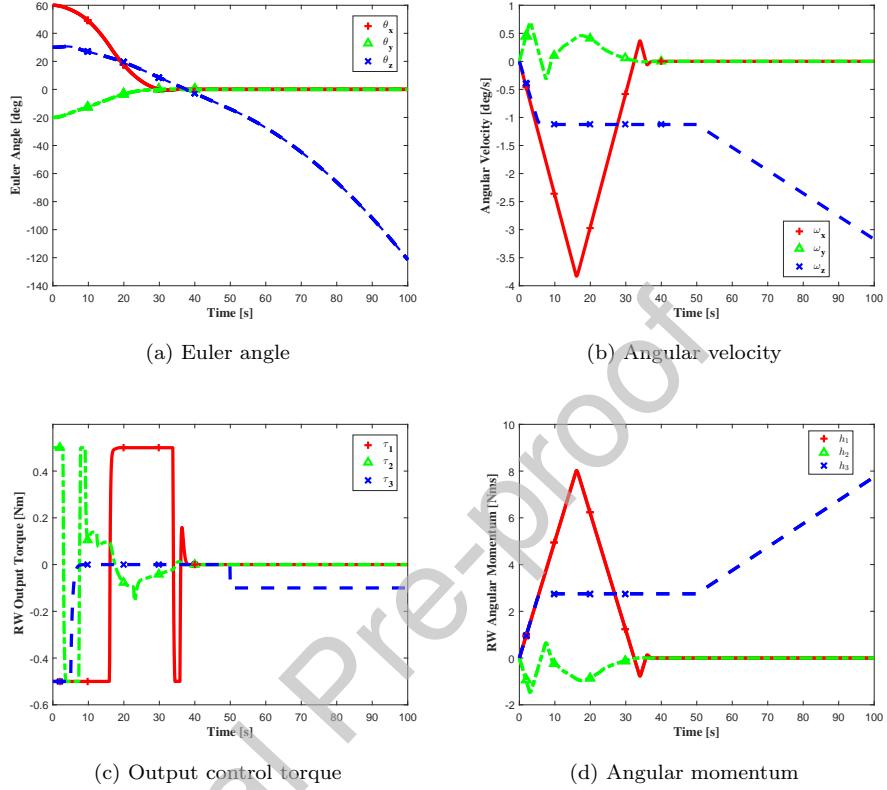
Figure 9: RW-actuated attitude control result under  $F_b$ 

constant after the wheel fails. This is because the speed of the third RW does not change after failure happens as shown in Fig. 9d. Fig. 10 shows the control result in the condition that the third wheel partially loses its effectiveness at 5 s and experiences the additive fault in 50 s (the fault  $F_c$ ). Comparing with Fig. 8, the transient process is almost the same, but a larger steady-state error is observed in Fig. 10a. The abrupt bias after 50 s is clearly observed in Fig. 10c. As shown in Fig. 11c, the wheel suffers from failure at 5 s and offset at 50 s (the failure  $F_d$ ). Under bias fault, angular velocities of the spacecraft and the wheel speed diverge as depicted in Fig. 11b and Fig. 11d. Same as in Fig. 9, the whole system is uncontrollable under the RW failure.

Figure 10: RW-actuated attitude control result under  $F_c$ 

From the simulation results in Figs. 7 to 11, we can obtain the following qualitative conclusions:

- The RW failure has more serious consequences than the RW fault in attitude control, and will directly make the system uncontrollable and unstable if there is no RW redundancy.
- The additive fault is more serious than the multiplicative fault, and it can result in steady-state error;
- The partial loss of effectiveness fault can be regarded as the case that the wheel is replaced by another one with a smaller control capacity. So, it

Figure 11: RW-actuated attitude control result under  $F_d$ 

does not affect system's controllability.

### 6.3. Attitude Control Results of SGCMG-Actuated Spacecraft

Simulation of SGCMG-actuated spacecraft system is conducted in this section. A pyramid configuration of four SGCMGs as shown in Fig. 12 is adopted to control the spacecraft. The attitude controller, the initial states and target are the same as they are in Section 6.2. The generalized singular robust inverse method is used to steer the gimbal as in [3]. Thus the gimbal rate command can be calculated as follows:

$$\dot{\delta}_c = \frac{1}{h_0} \mathbf{A}^\# \mathbf{u} \quad \text{and} \quad \mathbf{A}^\# = \mathbf{A}^T [\mathbf{A} \mathbf{A}^T + \lambda \mathbf{E}]^{-1}, \quad (40)$$

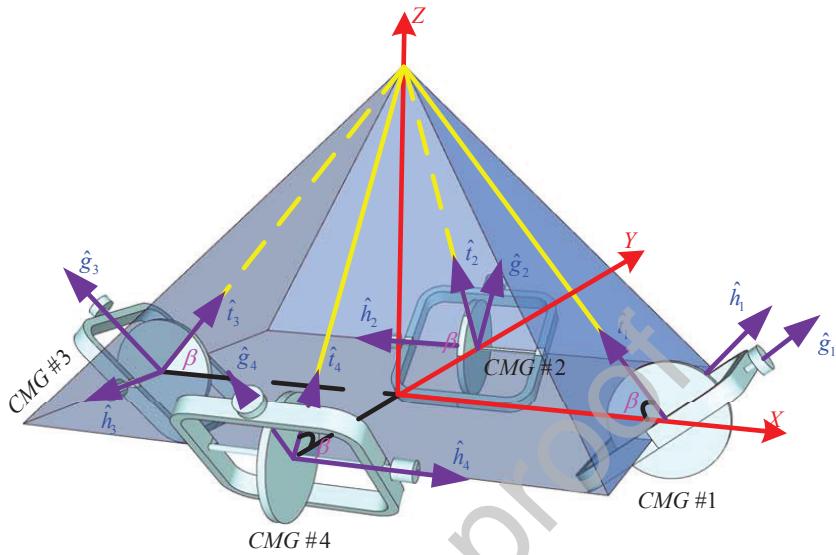


Figure 12: Pyramid configuration of the SGCMGs

where  $\mathbf{A}$  is the Jacobian matrix,  $\lambda = 0.01 \exp [-10 \det (\mathbf{AA}^T)]$ , and the matrix  $\mathbf{E}$  is expressed as:

$$\mathbf{E} = \begin{bmatrix} 1 & \varepsilon_3 & \varepsilon_2 \\ \varepsilon_3 & 1 & \varepsilon_1 \\ \varepsilon_2 & \varepsilon_1 & 1 \end{bmatrix} > 0$$

with  $\varepsilon_i = 0.01 \sin (0.5\pi t + \phi_i)$ ,  $\phi_1 = 0$ ,  $\phi_2 = \pi/2$  and  $\phi_3 = \pi$ .

Since we only consider the gimbal fault, using the fault model described by (26), the actual gimbal rate can be obtained as

$$\dot{\delta} = \mathbf{E}_{\eta^g} \dot{\delta}_c + \mathbf{E}_{ga} \quad (41)$$

330 where  $\mathbf{E}_{\eta^g} = \text{diag}[\eta_1^g, \eta_2^g, \eta_3^g, \eta_4^g]$  is the effectiveness matrix of the gimbal loop and  $\mathbf{E}_{ga} = \text{diag}[\dot{\delta}_{o_1}, \dot{\delta}_{o_2}, \dot{\delta}_{o_3}, \dot{\delta}_{o_4}]$  is the additive bias of the gimbal loop caused by faults.

The nominal angular momentum of the rotor is set as 10 Nm and the maximum gimbal angular velocity is set as 100 deg/s. The initial gimbal angle is 335  $[0, 0, 0, 0]^T$  deg. Considering the one degree of redundancy in the pyramid configuration, the SGCMG pair in the  $x-z$  plane, i.e. the 1st and the 3rd SGCMGs

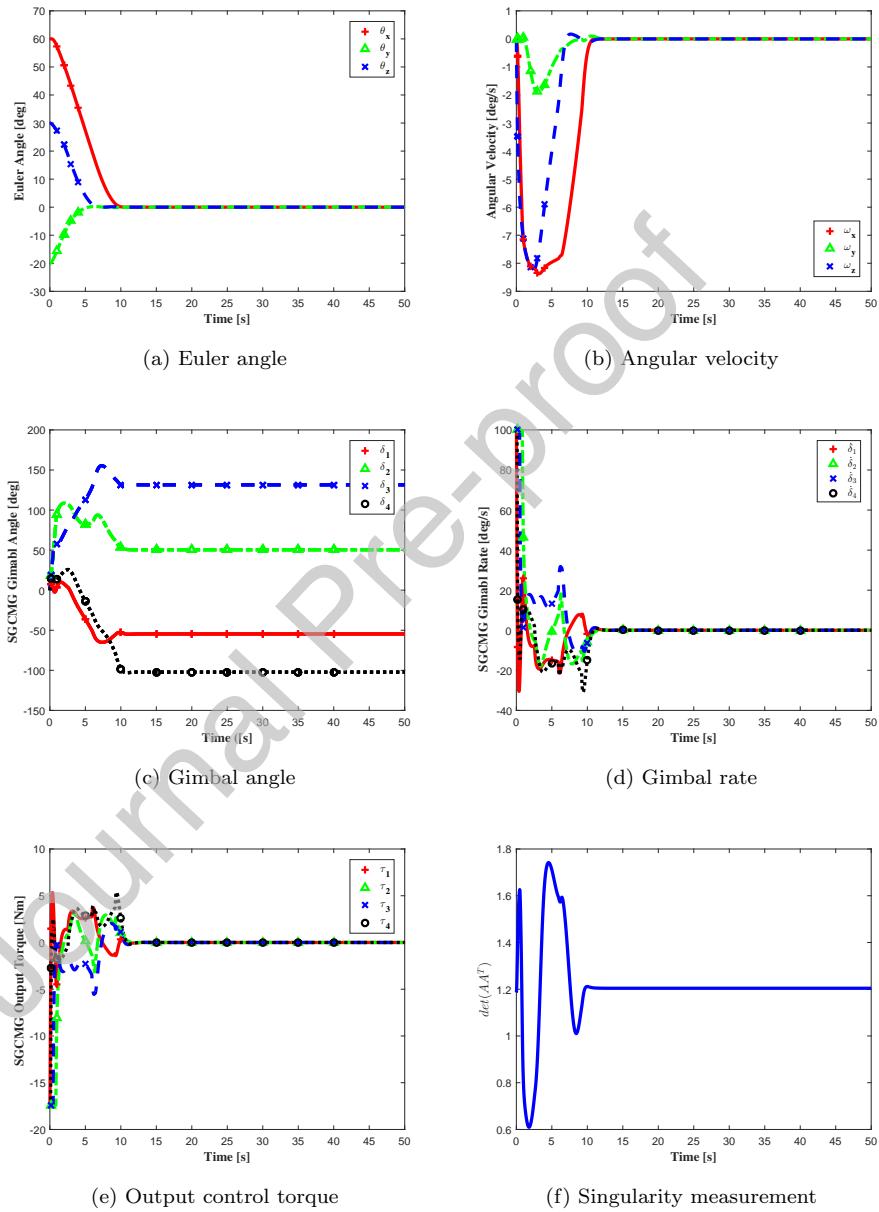


Figure 13: SGCMG-actuated attitude control result under nominal condition

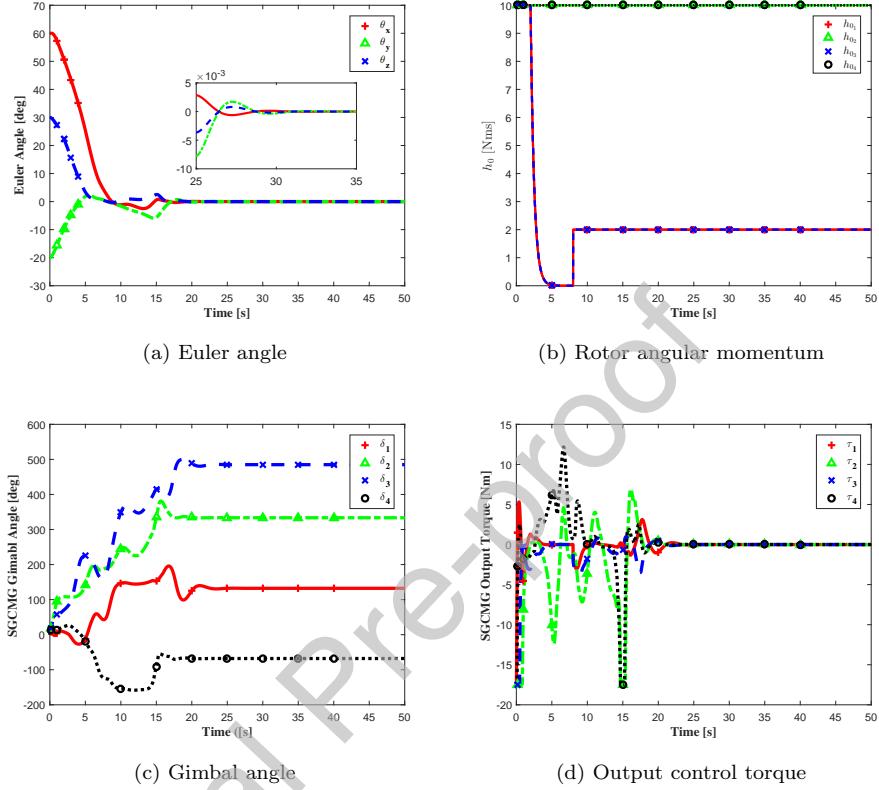


Figure 14: SGCMG-actuated attitude control result in the presence of only rotor fault

are assumed to be faulty simultaneously. To simplify the simulation, we only implement some typical scenarios in Table 1 in the simulation, which include failures with offset on the 1st and 3ed rotors (Fig. 14), failures with offset on the 1st and 3ed gimbal (Fig. 15) and both the rotor and gimbals failure (Fig. 340 17). In the following simulations, the effectiveness factors and bias of the rotor control loop are set as 0.5 and 2 Nms, while 0.75 and 20 deg/s for the gimbal control loop.

Fig. 13 shows the attitude control results under nominal condition. It can 345 be seen from Fig. 13 that the maneuver is completed in about 10 s. The maximum output torque is the product of rotor's maximum angular momen-

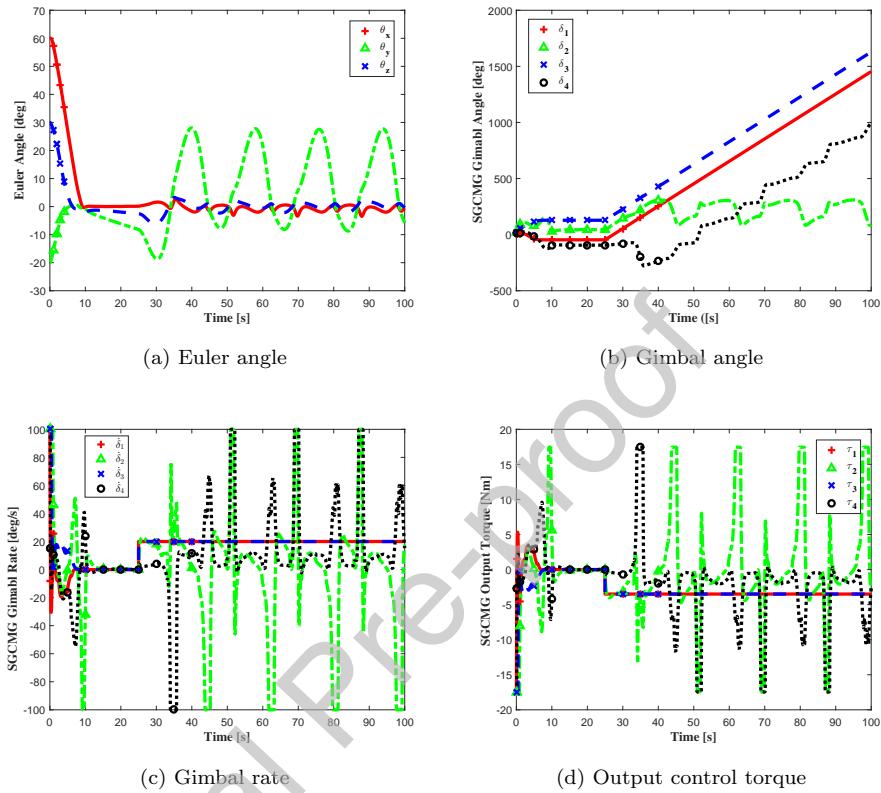


Figure 15: SGCMG-actuated attitude control result in the presence of only gimbal fault

tum and gimbal's maximum angular speed. During the whole process, there is no singularity. Fig. 14 presents the Euler angle trajectory, rotor's angular momentum, gimbal angle and output torque in the presence of rotor failures.

350 Fig. 14b shows that the rotor fails at 2 s and the angular momentum drops to zero rapidly. Consequently, the output torque also becomes zero along with the variation of rotor's angular momentum, which is shown in Fig. 14d. When this pair of CMGs totally fails, the system will turn to be uncontrollable. When the bias is added at 8 s, there is a jump in the output torque. After the bias, the system goes to be stable state in about 20 s as in Fig. 14a. Fig. 15 shows the control result when the EM-VSD system of gimbal fails totally. It is observed

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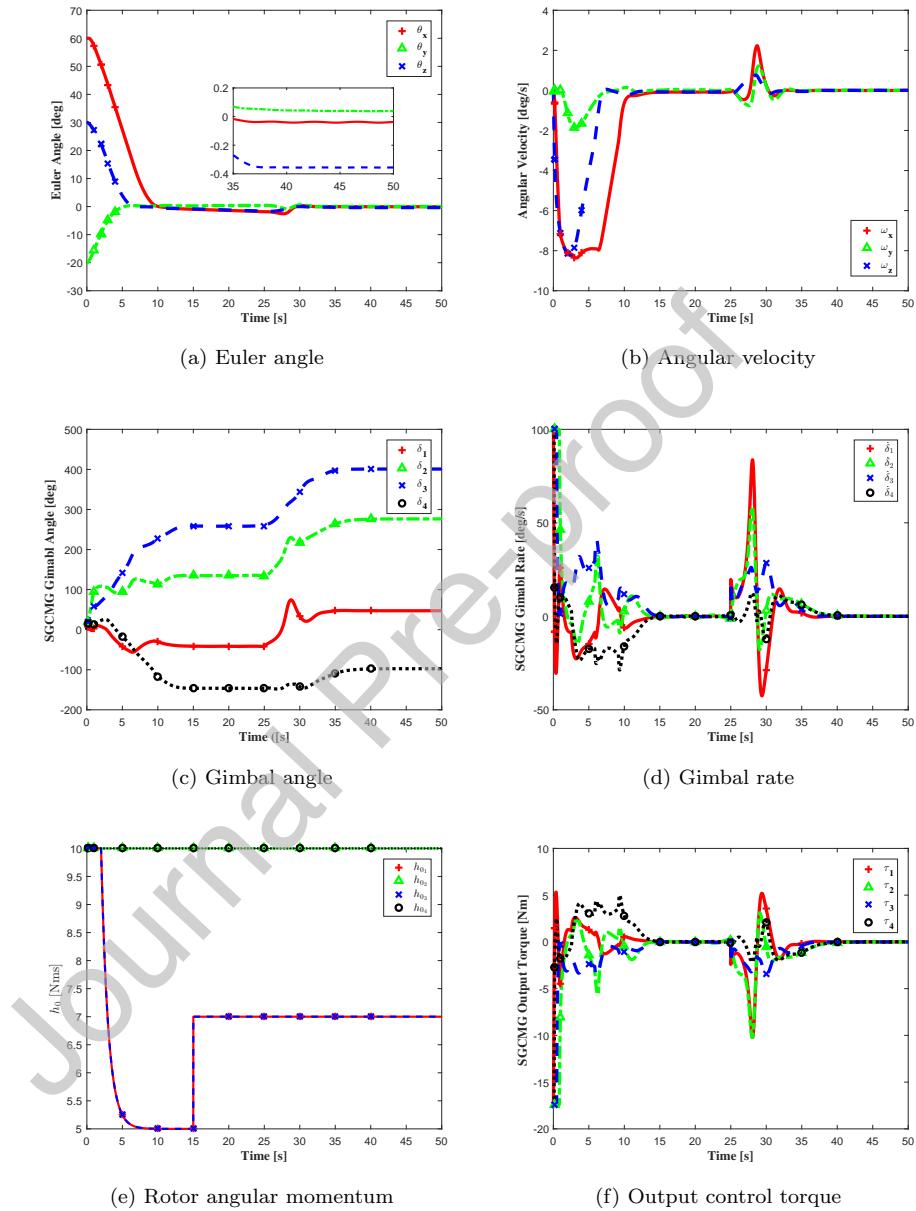


Figure 16: SGCMG-actuated attitude control result in the presence of rotor fault and gimbal fault

that the system becomes unstable until the additive fault occurs. After the additive fault, the system approaches to stable state as shown in Fig. 15a. Fig. 17 shows the simulation results when both the variable speed drive system of 360 rotor and gimbal experience fault. We can see obvious steady-state error in Fig. 17a and output jump due to the fault.

Parallel to the simulation results of RW-actuated system, we can obtain the similar qualitative conclusions:

- The gimbal fault is more serious than the rotor fault when the rotor's 365 angular momentum is greater than zero;
- The angular momentum of rotor is the amplification factor of the SGCMG. The fault in rotor can be regarded as the replacement of another SGCMG with a smaller control capacity.
- The additive fault may result in steady-state error.

## 370 7. Fault-Tolerant Control Strategies

Various fault-tolerant controllers (FTCs) have been developed, as summarized in the review of [8] and [51]. Generally speaking, the existing FTCs can be divided into the passive FTCs, active FTCs and the hybrid of passive and active 375 FTCs. A comparative study between the active and passive approaches is given in [9]. For these FTCs, the redundancy of the system, including but not limited to the hardware redundancy, is the key factor and the significant difference lies in how these redundancies are used. For the passive FTCs, a single fixed controller is designed among all admissible solution sets within the overlapped region considering both the nominal conditions and design basis faults. This 380 kind of method emphasises on robustness for all identified cases rather than achieving optimal performance. The passive FTCs are more conservative when compared with active FTCs, which contain the fault detection and diagnosis (FDD) scheme, reconfigurable controller, and the decision making or redundancy management scheme. The performance of active FTCs highly depends

385 on the FDD results. More results can be found in [52] and [53].

In this section, we focus on how to incorporate the proposed fault model in fault-tolerant strategies design to accommodate SGCMG gimbal fault. Different from the previous section where the control torque generated by the SGCMGs expressed as  $\tau_c = -h_0 \mathbf{A} \dot{\boldsymbol{\delta}}$ , the internal torque  $\tau = -h_0 \mathbf{A} \dot{\boldsymbol{\delta}} - \boldsymbol{\omega}^\times \mathbf{H}$  390 is adopted as the output of the SGCMG cluster, which makes the SGCMG to be a replaceable unit. Then the spacecraft dynamic can be expressed as  $\mathbf{J} \ddot{\boldsymbol{\omega}} + \boldsymbol{\omega}^\times \mathbf{J} \boldsymbol{\omega} = \boldsymbol{\tau} + \mathbf{d}$ , which is independent with actuators. The schematic diagram of two potential fault-tolerant strategies to accommodate SGCMG gimbal fault are demonstrated in Fig. 17.

395 In Fig. 17a, an additive equivalent strategy is adopted and the real gimbal rate  $\dot{\boldsymbol{\delta}}$  is expressed by the gimbal rate command  $\dot{\boldsymbol{\delta}}_c$  and a lumped bias  $\mathbf{f}$ , i.e.  $\dot{\boldsymbol{\delta}} = \dot{\boldsymbol{\delta}}_c + \mathbf{f}$ . To estimate the lumped bias, local estimators (LE) can be designed to estimate each component of  $\mathbf{f}$  corresponding to each SGCMG. Then the gimbal rate command after compensation becomes  $\dot{\boldsymbol{\delta}}_c = \dot{\boldsymbol{\delta}}'_c - \hat{\mathbf{f}}$ . As a consequence, the 400 actual gimbal rate is  $\dot{\boldsymbol{\delta}} = \dot{\boldsymbol{\delta}}'_c - \hat{\mathbf{f}} + \mathbf{f} \approx \dot{\boldsymbol{\delta}}'_c$ . That is to say, the fault-tolerant target is achieved. Since the FDD scheme and the reconfiguration are not used, above additive equivalent fault-tolerant strategy can be regarded as a passive fault-tolerant control strategy.

In Fig. 17b, the fault effect is described in a multiplicative way, i.e.  $\dot{\boldsymbol{\delta}} = 405 \mathbf{E}_{\eta^g} \dot{\boldsymbol{\delta}}_c$ . To complete the fault-tolerant steering law design, some strategies are required to identify the fault and evaluate how serious the fault is, namely to estimate the effectiveness matrix  $\mathbf{E}_{\eta^g}$ . Using the estimated effectiveness matrix  $\hat{\mathbf{E}}_{\eta^g}$ , new efficient SGCMG steering law using the estimated effectiveness matrix can be developed to minimize the use of the faulty SGCMG. Once the estimation 410 is completed, the steering law is re-configured from the traditional one to the effectiveness weighted one. This reallocates the control command and the faulty SGCMGs can potentially be isolated. This strategy can be regarded as an active strategy. Together with the method shown in Fig. 17a, the gimbal fault of the SGCMGs can be handled. How to design the local estimator to estimate the 415 equivalent effectiveness matrix will be one of our future works.

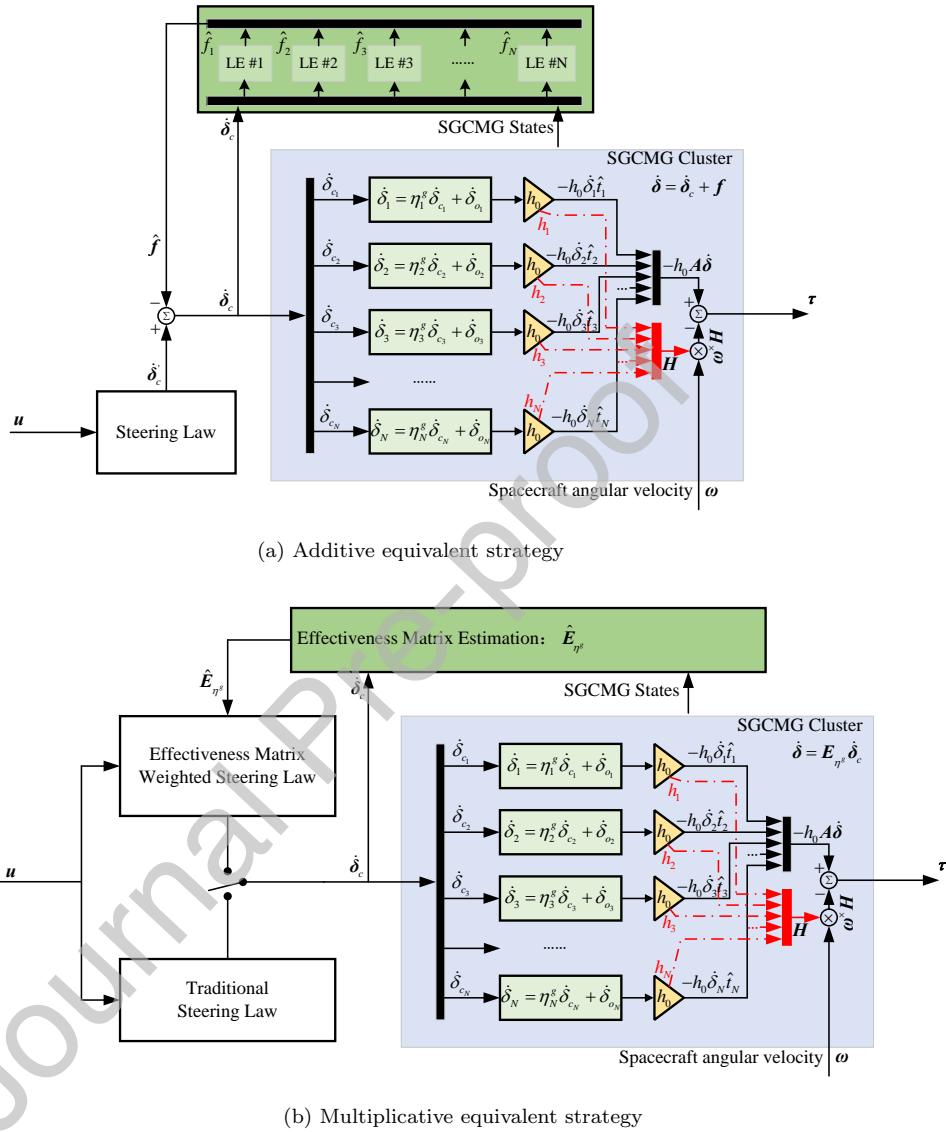


Figure 17: Potential fault-tolerant control strategies to handle gimbal fault in attitude control systems

## 8. Conclusion

This paper investigated the fault modeling in momentum exchange devices. A general fault model is established by considering the momentum exchange devices as the cascade-connected EM-VSD systems. The potential faults of the EM-VSD system are identified, and the reason why these faults can be categorized into multiplicative fault and additive fault is given. Based on the developed EM-VSD fault model, we further develop fault models of momentum exchange devices such as RW, SGCMG, SGCMG and VSCMG. Through simulations of spacecraft attitude control using RWs and SGCMGs as actuators, the potential faults in RW and SGCMG as well as their influences on the control performance are analyzed in detail. Moreover, we also obtain the qualitative conclusion that the additive fault has more serious influence than the multiplicative fault from the viewpoint of control accuracy. This observation can be a guideline in developing fault-tolerant control system for the momentum exchange devices actuated spacecraft. The future works may focus on verifying the proposed fault model using reaction wheel and control moment gyro testbed. Moreover, to accommodate the actuator fault and recover the nominal control performance, analysis and design of fault diagnosis and fault-tolerant control by leveraging the developed fault model should also be addressed.

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