

DSP Additional.

$$\textcircled{1} \quad x[n] = \begin{cases} 1 + \frac{n}{3}, & -3 \leq n \leq -1 \\ 1, & 0 \leq n \leq 3 \\ 0, & \text{elsewhere} \end{cases}$$

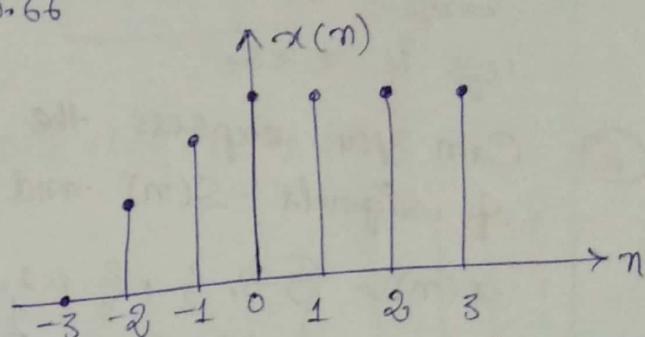
- \textcircled{a} determine its values and sketch the signal $x(n)$.

$$x(-3) = 1 + \frac{-3}{3} = 0 \quad x(1) = 1$$

$$x(-2) = 1 + \frac{-2}{3} = 0.33 \quad x(2) = 1$$

$$x(-1) = 1 + \frac{-1}{3} = 0.66 \quad x(3) = 1$$

$$x(0) = 1$$

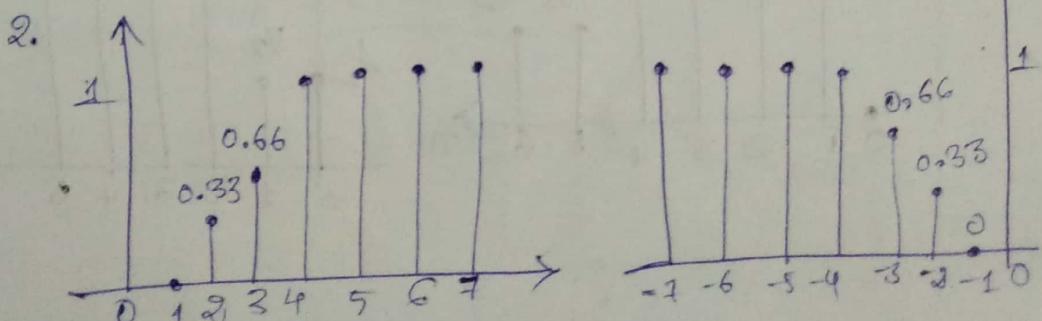
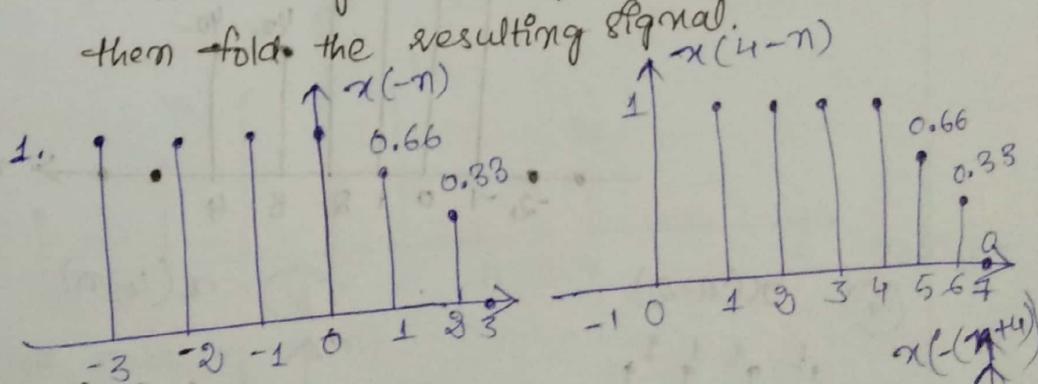


- \textcircled{b} Sketch the sig that result if we:

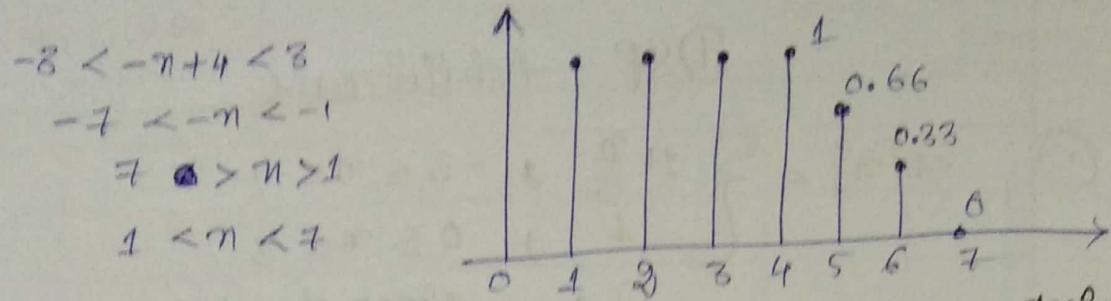
1. First fold $x(n)$ and then delay the

resulting signal by four samples.

2. First delay $x(n)$ by four samples and then fold the resulting signal.



④ Sketch the signal $x(-n+4)$



⑤ Compare the results in parts ④ & ④ and derive a rule for obtaining the signal $x(-n+k)$

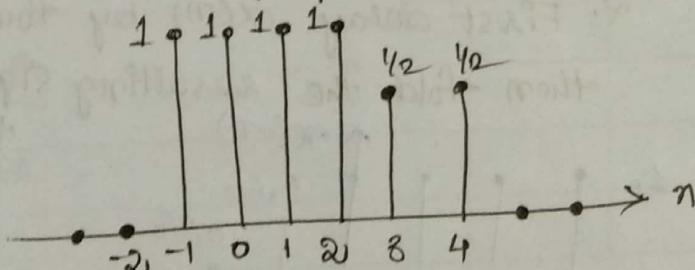
from $x(n)$.
To obtain $x(-n+k)$, first fold $x(n)$ which gives $x(-n)$. Then we have to shift by k samples to the right of $n=0$, or k samples left if $k < 0$.

⑥ Can you express the signal $x(n)$ in terms of signals $s(n)$ and $u(n)$?

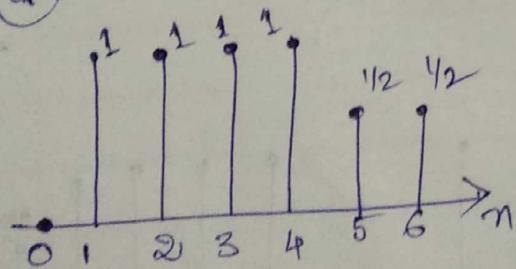
$$x(n) = \left\{ \begin{matrix} 0, \frac{1}{3}, \frac{2}{3}, 1, 1, 1, 1 \\ -3 -2 -1 0 \end{matrix} \right\}$$

$$x(n) = \frac{1}{3} s(n-2) + \frac{2}{3} s(n-1) + s(n) + 2 u(n) - u(n-4)$$

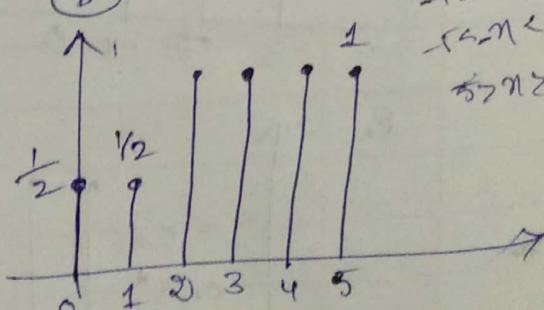
⑦ A discrete-time signal $x(n)$ is shown in fig. Sketch the label carefully each of the following sig.



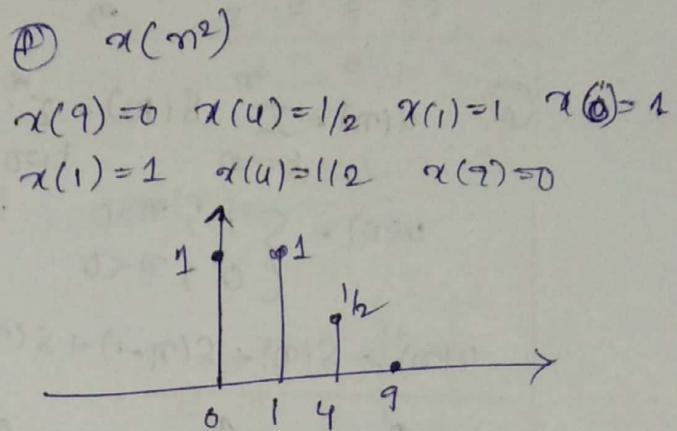
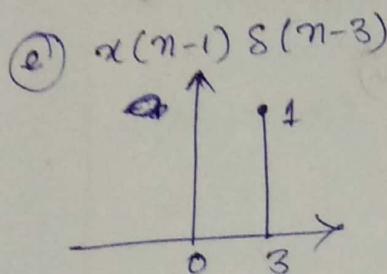
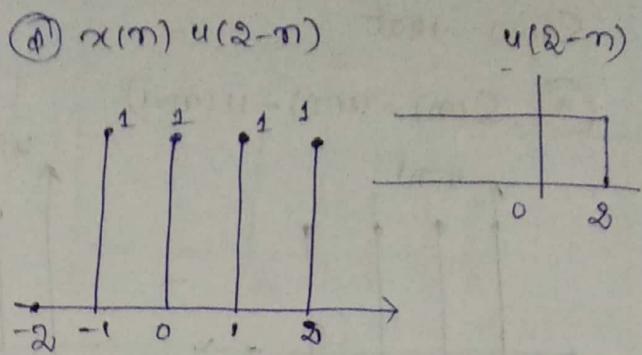
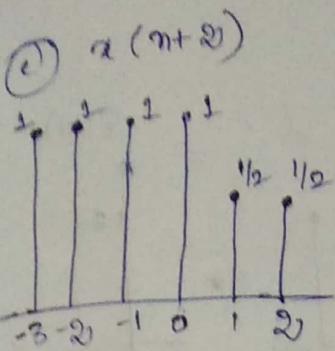
⑧ a) $x(n-2)$



b) $x(4-n)$



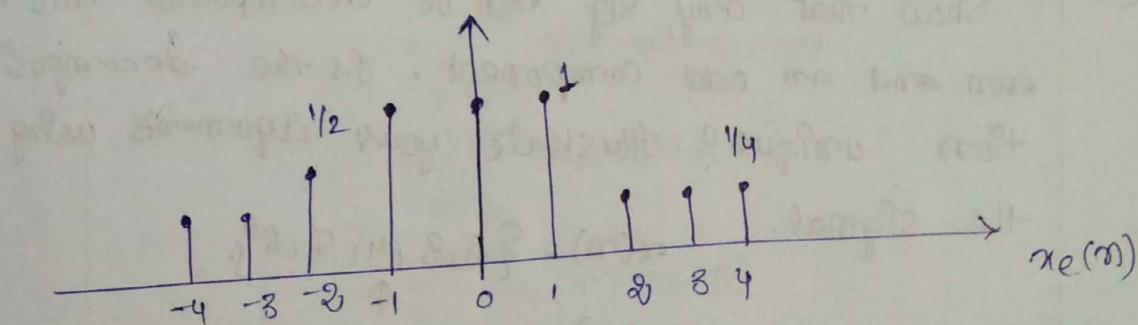
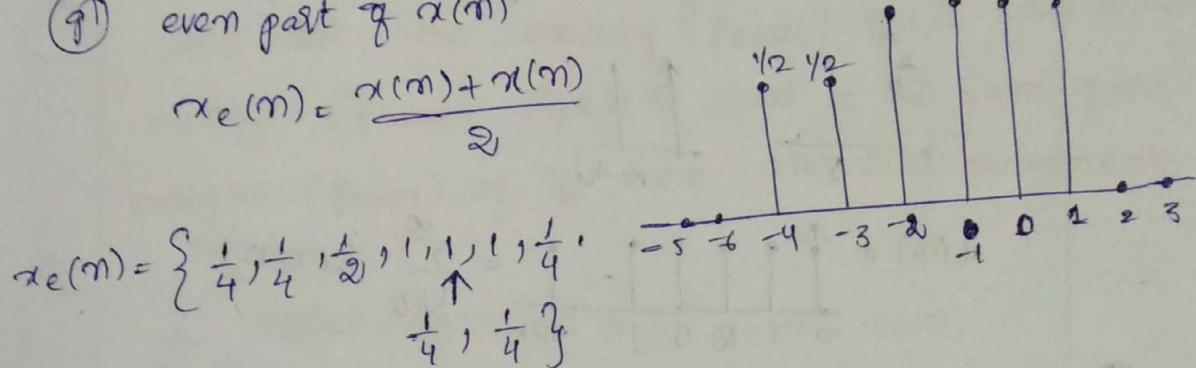
$$\begin{aligned} -1 < 4-n < 5 \\ -5 < n < 5 \\ \therefore n \geq 0 \end{aligned}$$



$$\{0, 1/2, 1, 1, 1, 1/2, 0\}$$

⑦ even part of $x(n)$

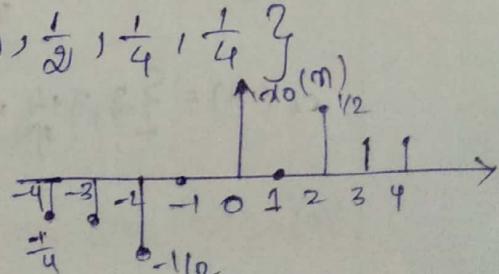
$$x_e(n) = \frac{x(n) + x(-n)}{2}$$



⑧ odd part of $x(n)$

$$x_o(n) = \frac{x(n) - x(-n)}{2}$$

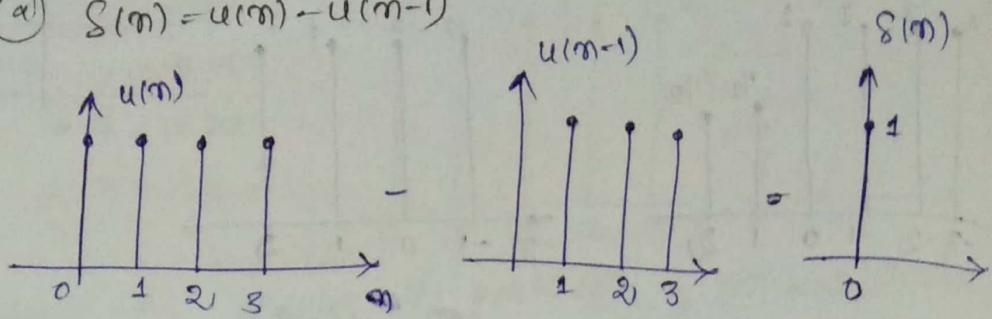
$$x_o(n) = \left\{ -\frac{1}{4}, -\frac{1}{4}, -\frac{1}{2}, 0, 0, 0, \frac{1}{2}, \frac{1}{4}, \frac{1}{4} \right\}$$



(3)

Show that

$$\text{a) } \delta(m) = u(m) - u(m-1)$$

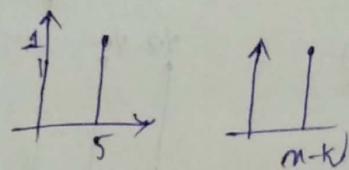
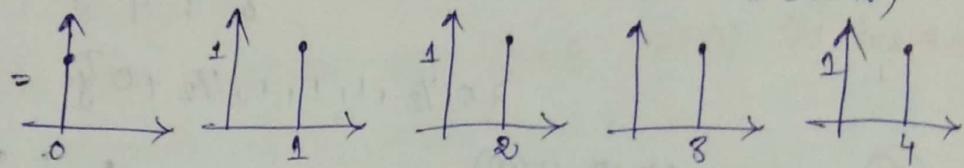


$$\text{b) } u(m) = \sum_{k=0}^m \delta(k) = \sum_{k=0}^{\infty} \delta(m-k)$$

$$u(m) = \begin{cases} 1; & m > 0 \\ 0; & m \leq 0 \end{cases}$$

$$u(m) = \delta(m) + \delta(m-1) + \delta(m-2) + \delta(m-3) + \delta(m-4) + \delta(m-5)$$

$\cdots \cdots \delta(m-k)$



$$u(m) = \underbrace{\delta(0) + \delta(1) + \delta(2) + \delta(3) + \delta(4)}_{\text{even part}} - \underbrace{\delta(m-1) - \delta(m-3)}_{\text{odd part}} = \sum_{k=0}^{\infty} \delta(m-k)$$

(4)

Show that any sig can be decomposed into an even and an odd component. Is the decomposition unique? Illustrate your arguments using the signal.

$$x(m) = \{2, 3, 4, 5, 6\}$$

Ans

$$x_e(m) = \frac{x(m) + x(-m)}{2}$$

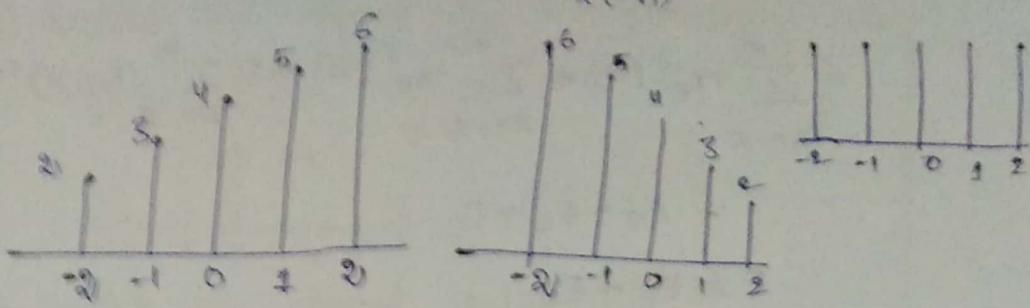
$$x_e(m) = x_e(-m)$$

$$x_o(m) = \frac{x(m) - x(-m)}{2}$$

$$x_o(m) = -x_o(-m)$$

$$x(m) = \{2, 3, 4, 5, 6\}$$

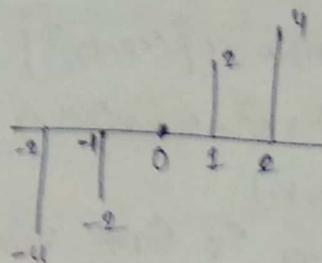
$$\Rightarrow x(m) = x_e(m) + x_o(m)$$



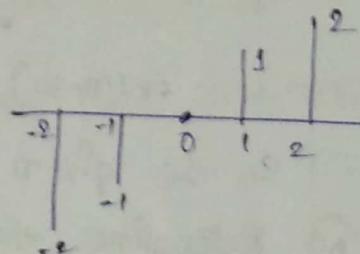
$$x_e(n) = \frac{x_e(n) + x(-n)}{2} \Rightarrow \begin{array}{|c|c|c|c|c|} \hline & 4 & 4 & 4 & 4 & 4 \\ \hline n = -2 & -1 & 0 & 1 & 2 \\ \hline \end{array}$$

$x(n) - x(-n)$

\Rightarrow



$$x_o(n) = \frac{x(n) - x(-n)}{2}$$



- ⑤ Show that the energy (Power) of a real-valued energy (Power) signal is equal to the sum of the energies (power) of its even and odd components.

First prove

$$\sum_{n=-\infty}^{\infty} x_e(n) + x_o(n) = 0 \Rightarrow \sum_{n=-\infty}^{\infty} x_e(n) + x_o(n)$$

$$= \sum_{m=-\infty}^{\infty} x_e(-m) x_o(-m)$$

$$= - \sum_{m=-\infty}^{\infty} x_e(m) x_o(m)$$

$$= \sum_{m=-\infty}^{\infty} x_e(m) x_o(m)$$

$$= \sum_{n=-\infty}^{\infty} x_e(n) x_o(n) \underline{\underline{= 0}}$$

energies (Power)

$$\sum_{n=-\infty}^{\infty} x^2(n) = \sum_{n=-\infty}^{\infty} (x_e(n) + x_o(n))^2$$

$$= \sum_{n=-\infty}^{\infty} x_e^2(n) + x_o^2(n) + 2x_e(n)x_o(n)$$

$$= \sum_{n=-\infty}^{\infty} x_e^2(n) + \sum_{n=-\infty}^{\infty} x_0^2(n) + 2 \sum_{n=-\infty}^{\infty} x_e(n)x_0(n)$$

$$= E_e + E_0 + 0$$

$$E = E_e + E_0$$

⑥ Consider the S/I $y(n) = \mathcal{P}[x(n)] = x(n^2)$

ⓐ Determine if the S/I is time invariant

$$\text{Given } y(n) = \mathcal{P}[x(n)] = x(n^2)$$

$$x(n-k) \rightarrow y_1(n) = x[(n-k)^2]$$

$$= x(n^2 + k^2 - 2nk)$$

$$x(n-k) \neq y(n-k)$$

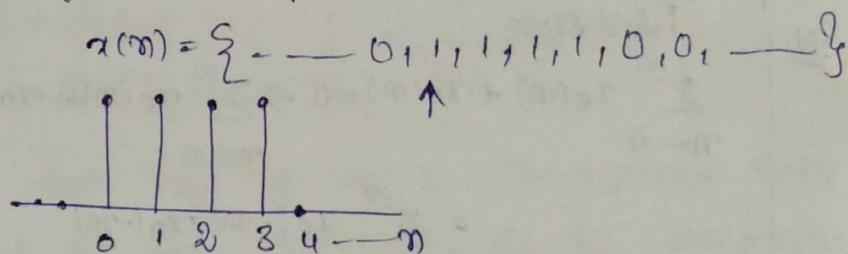
So the given system is time variant

ⓑ Determine the result in part Ⓛ assume that

$$\text{the signal } x(n) = \begin{cases} 1, & 0 \leq n \leq 3 \\ 0, & \text{elsewhere} \end{cases}$$

is applied into the S/I.

① Sketch the signal $x(n)$

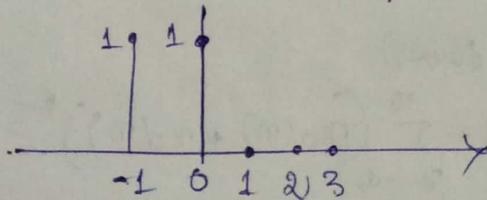


② Determine & sketch the signal $y(n) = \mathcal{P}[x(n)]$

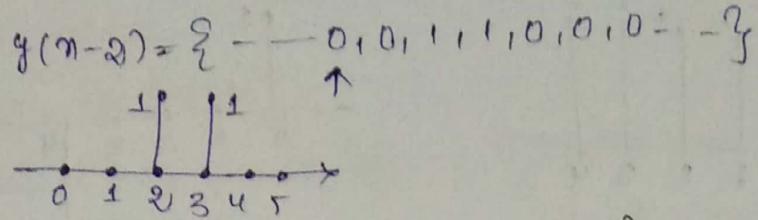
$$y(n) = \mathcal{P}[x(n)] = x(n^2) \Rightarrow \{x(0), x(1), x(4), x(9), \dots\}$$

$$\Rightarrow \{x(0), x(1), x(4), x(9), x(16), \dots\}$$

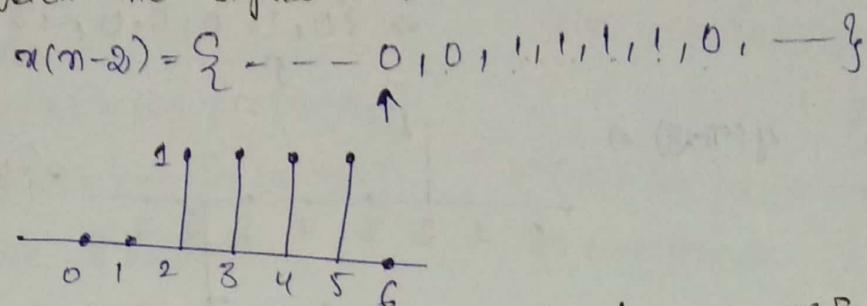
$$y(n) = x(n^2) = \begin{cases} 1, & n = 0, 1, 2, 3 \\ 0, & \text{elsewhere} \end{cases}$$



③ Sketch the signal $y_2'(n) = y(n-2)$.



④ Sketch the signal $x_2(n) = x(n-2)$.



⑤ Determine & sketch the signal $y_2(n) = P[x_2(n)]$

$$\begin{aligned} y_2(n) &= P[x_2(n)] = \{ x(0), x(1), x(2), x(3), x(4), \\ &\quad x(5), x(6) \} \\ &= \{ \dots, 0, 1, 0, 0, 0, 1, 0, \dots \} \\ &\text{Plot: } \begin{array}{ccccccc} & 1 & & & 1 & & \\ \text{---} & \downarrow & & & \downarrow & & \\ -2 & -1 & 0 & 1 & 2 & 3 & \rightarrow y_2(n) \end{array} \end{aligned}$$

⑥ compare the signal $y_2(n)$ and $y(n-2)$.
what is conclusion?

$y_2(n) \neq y(n-2) \rightarrow$ system is time variant

⑦ Repeat part ⑥ for the sm.

$$y(n) = x(n) - x(n-1)$$

Can you use this result to make any statement about the time invariance of the sm.

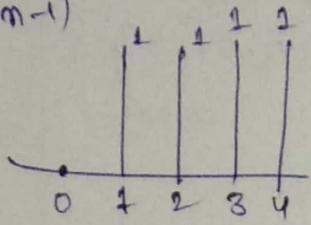
$$\textcircled{1} \quad x(n) = \begin{array}{ccccccc} & 1 & 1 & 1 & 1 & & \\ \text{---} & \downarrow & & & \downarrow & & \\ -1 & 0 & 1 & 2 & 3 & 4 & \rightarrow n \end{array} = \{ 1, 1, 1, 1 \}$$

$$\textcircled{2} \quad y(n) = x(n) - x(n-1)$$

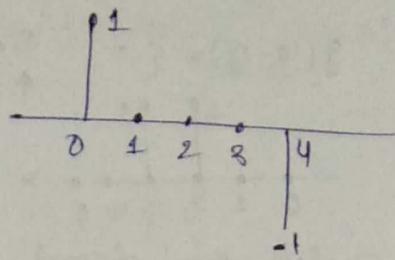
$$\begin{array}{ccccccc} x(n) & & 1 & 1 & 1 & 1 & \\ \text{---} & & \downarrow & & \downarrow & & \\ -1 & 0 & 1 & 2 & 3 & 4 & \rightarrow n \end{array}$$

$y(n)$
 $y(n) = x(n) - x(n-1)$

$x(n-1)$

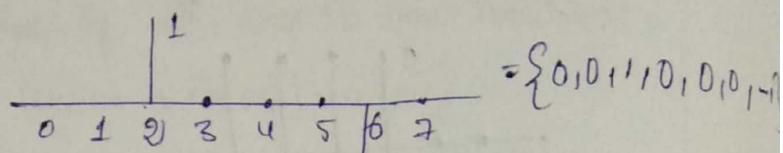


$$y(n) \Rightarrow x(n) - x(n-1)$$



$$\Rightarrow \{0, 1, 0, 0, 0, -1\}$$

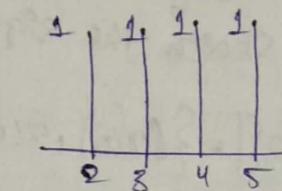
③ $y(n-2) \Rightarrow$



$$= \{0, 0, 1, 1, 0, 0, 0, -1\}$$

④

$x(n-2) \Rightarrow$



$$\Rightarrow \{0, 0, 1, 1, 1, 1, 1\}$$

⑤

$$y_2(n) = \{0, 0, 1, 1, 0, 0, 0, -1\}$$

⑥ $y_2(n) = y(n-2) \Rightarrow$ SLM is time invariant.

⑦ Repeat parts ⑥ and ⑤ for the SLM.

$$y(n) = \gamma[x(n)] = n x(n)$$

\rightarrow integer value from 0

$$i) \quad y(n) = n x(n)$$

$$x(n) = \{-1, 1, 1, 1, 1, 0, -1\}$$

$$② \quad y(n) = \{-1, 0, 1, 2, 3, 4, -1\}$$

$$③ \quad y(n-2) = \{-1, 0, 0, 1, 2, 3, 4, -1\}$$

$$④ \quad x(n-2) = \{-1, 0, 0, 0, 1, 1, 1, -1\}$$

$$⑤ \quad y_2(n) = \gamma[x(n-2)] = \{-1, 0, 1, 3, 4, 5, -1\}$$

⑥ $y_2(n) \neq y(n-2) \Rightarrow$ SLM is time variant.

⑦ A discrete-time system can be

① static or dynamic

② linear or nonlinear

③ Time invariant or time varying

④ Causal or noncausal

⑤ Stable or unstable

Examine the following systems with respect to the properties above.

(a) $y(n) = \cos[x(n)]$

i) static (only present if P)

ii) only present if P \rightarrow causal

iii) stable

iv) $y(n) = \cos[x(n-n_0)]$

$$y(n) = \cos[x(n-n_0)]$$

\Rightarrow Time variant

(b) $y(n) = \sum_{k=-\infty}^{n+1} x(k)$

Dynamic (depends on future values)

linear, time invariant, non causal (also depends on future values).

unstable.

(c) $y(n) = x(n) \cdot \cos[\omega_0(n)]$

Static, linear, Time variant, causal, stable

$$y(n) = x(n-n_0) \cos[\omega_0(n-n_0)]$$

$$y'(n) = x(n-n_0) \cos[\omega_0 n]$$

(d) $y(n) = x(-n+2)$

Dynamic

↓

at $n=0 \Rightarrow y(0)=x(2)$

future value ↪

$$y_1(n) = x_1(-n+2) + x_2(-n+2)$$

$$y_2(n) = (x + x_2(-n+2))$$

$$\Rightarrow x_1(-n+2) + x_2(-n+2)$$

→ linear

Non-causal, stable, Time invariant

(e) $y(n) = \text{Trunc}[x(n)]$

Static, non-linear, time invariant, causal, stable

(f) $y(n) = \text{Round}[x(n)]$

Static, non-linear, time invariant, causal, stable

ii) $y_1(n) = \cos[x_1(n)]$

$$y_2(n) = \cos[x_2(n)]$$

$$y(n) = \cos[x_1(n)] + \cos[x_2(n)]$$

$$y'(n) = \cos[x_1(n) + x_2(n)]$$

\Rightarrow Non-linear

$$g) y(n) = |x(n)|$$

static, non-linear, time invariant, causal, stable

$$h) y(n) = x(n) u(n)$$

static, linear, time invariant, causal, stable.

$$i) y(n) = x(n) + nx(n+1)$$

dynamic, linear, time variant, non-causal, stable

$$j) y(n) = x(2n)$$

dynamic, linear, time variant, non-causal, stable

$$k) y(n) = \begin{cases} x(n), & \text{if } x(n) \geq 0 \\ 0, & \text{if } x(n) < 0 \end{cases}$$

static, linear, time invariant, non-causal, stable

$$l) y(n) = x(-n)$$

dynamic, linear, time variant, causal, stable.

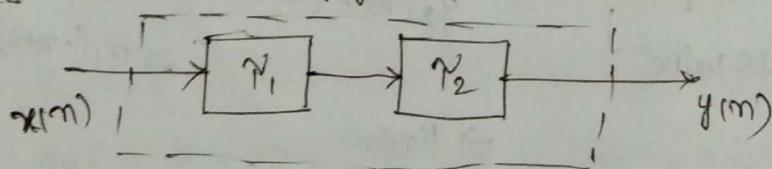
$$m) y(n) = \text{sgn}[x(n)]$$

static, non-linear, time invariant, causal, stable

$$n) x(n) = x_1(nT) \quad -\infty < n < \infty$$

static, linear, time variant, non-causal, stable

- o) Two discrete-time systems γ_1 and γ_2 are connected in cascade to form a new system γ as shown in fig. Prove or disprove the following



$$\gamma = \gamma_1 \gamma_2$$

- a) If γ_1 and γ_2 are linear, then γ is linear (i.e., the cascade connection of two linear systems is linear).

True if $y_1(n) = \gamma_1[x_1(n)]$ and $y_2(n) = \gamma_2[x_2(n)]$
then

$$\alpha_1 x_1(n) + \alpha_2 x_2(n) \text{ yields } \\ \alpha_1 v_1(n) + \alpha_2 v_2(n)$$

By the linearity property of γ_1 , similarly, if

$$y_1(n) = \gamma_1 [v_1(n)] \text{ and}$$

$$y_2(n) = \gamma_2 [v_2(n)]$$

$$\text{then } \beta_1 v_1(n) + \beta_2 v_2(n) \rightarrow y(n) = \beta_1 y_1(n) + \beta_2 y_2(n)$$

by the linearity property of γ_2 , since

$$v_1(n) = \gamma_1 [x_1(n)] \text{ and}$$

$$v_2(n) = \gamma_2 [x_2(n)]$$

$$\text{it follows that, } \alpha_1 x_1(n) + \alpha_2 x_2(n)$$

yields the output

$$\alpha_1 \gamma_1 [x_1(n)] + \alpha_2 \gamma_2 [x_2(n)]$$

where $\gamma = \gamma_1 \gamma_2$ hence γ is linear

(b) If γ_1 and γ_2 are time invariant, then γ is time invariant.

True for γ_1 , if $x(n) \rightarrow v(n)$ &
 $x(n-k) \rightarrow v(n-k)$

for γ_2 , if $v(n) \rightarrow y(n)$

and $v(n-k) \rightarrow y(n-k)$

Hence, for $\gamma_1 \gamma_2$ if $x(n) \rightarrow y(n)$
 $x(n-k) \rightarrow y(n-k)$

$\gamma = \gamma_1 \gamma_2$ is time invariant.

(c) If γ_1 and γ_2 are causal, then γ is causal.

True, γ_1 is causal $\Rightarrow v(n)$ depends only on $x(k)$ for

$k \leq n$. γ_2 is causal $\Rightarrow y(n)$ depends only on $v(k)$ for

$k \leq n$. Therefore, $y(n)$ depends only on ~~all previous~~ $x(k)$ for $k \leq n$.

Hence γ is causal.

(d) If γ_1 and γ_2 are linear and time invariant
 the same holds for γ .

True. Combine (a) and (b) then we will get
 to understand that γ_1 and γ_2 are linear and

time invariant that holds for γ .

- (c) If γ_1 and γ_2 are linear and time invariant
then interchanging their order doesn't change

the sum γ .
True this follows from $h_1(n) * h_2(n) = h_2(n) * h_1(n)$

- (d) As in part (c) except that γ_1, γ_2 are now
time varying.

false. Ex: $\gamma_1: y(n) = n x(n)$

$$\gamma_1: y(n) = n x(n)$$

$$\gamma_2: y(n) = n x(n+1)$$

$$\gamma_2 \cdot [\gamma_1 [s(n)] = \gamma_2(0) = 0$$

$$\gamma_1 \cdot [\gamma_2 [s(n)]] = \gamma_1[s(n+1)] \\ = -s(n+1)$$

- (e) If γ_1 and γ_2 are nonlinear, then $\neq 0$

γ is nonlinear.

false. Ex: $\gamma_1: y(n) = x(n) + b$ and

$$\gamma_2: y(n) = x(n) - b, \text{ where } b \neq 0$$

$$\gamma [x(n)] = \gamma_2 [\gamma_1 [x(n)]] = \gamma_2 [x(n) + b] = x(n)$$

Hence γ is linear

- (f) If γ_1 and γ_2 are stable, then γ is stable

True. γ_1 is stable \Rightarrow $y(n)$ is bounded if $x(n)$ is bounded

γ_2 is stable \Rightarrow $y(n)$ is bounded if $v(n)$ is bounded.

Hence, $y(n)$ is bounded if $x(n)$ is bounded \Rightarrow
 $\gamma = \gamma_1 \gamma_2$ is stable.

- (i) Show by an example that the inverse of
parts (c) and (f) do not hold in general.

Inverse of (c). γ_1 and γ_2 are noncausal

$\Rightarrow \gamma$ is noncausal. Ex: $\gamma_1: y(n) = x(n+1)$ &
 $\gamma_2: y(n) = x(n-2)$

which is causal, hence the inverse of (c) is false.
Inverse of (f); γ_1 and/or γ_2 is unstable,

implies γ is unstable. $\exists \epsilon > 0$: $y(n) = e^{\gamma n}$, stable
and another that η_2 : $y(n) = \ln [x(n)]$ which η_2
unstable.

But η : $y(n) = x(n)$, which η is stable. Hence the
inverse of (h) is false.

(g) Let η be an LTI, relaxed and BIBO stable
S/I with $\eta(p)$ $x(n)$ and O/P $y(n)$. Show that

(a) If $x(n)$ is periodic with period N i.e.
 $x(n) = x(n+N)$ for all $n \geq 0$ the O/P $y(n)$ tends
to a periodic sig with the same period.

$$x(n) = x(n+N) \quad \forall n \geq 0$$

$$y(n) = \sum_{k=-\infty}^{\infty} h(k) x(n-k)$$

$$y(n+N) = \sum_{k=n+1}^{n+N} h(k) x(n+N-k)$$

$$y(n+N) = \sum_{k=n+1}^{n+N} h(k) x(n+k) + \sum_{k=-\infty}^n h(k) x(n+k)$$

$$y(n+N) = y(n) + \sum_{k=n+1}^{n+N} h(k) x(n+k)$$

for BIBO S/I $\lim_{n \rightarrow \infty} |h(n)| = 0$

$$\lim_{n \rightarrow \infty} \sum_{k=n+1}^{n+N} h(k) x(n+k) = 0$$

$$\lim_{n \rightarrow \infty} y(n+N) = y(n)$$

$$\therefore y(n) = y(n+N)$$

(b) If $x(n)$ is bounded and tends to a
constant, the O/P will also tend to a constant.

$$x(n) = x_0(n) + a u(n) \quad x_0(n) \rightarrow \text{bounded with}$$

$$\lim_{n \rightarrow \infty} x_0(n) = 0$$

$$\sum_n x_0^2(n) < \infty \Rightarrow \sum_n y_0^2(n) < \infty.$$

$$\text{hence } \lim_{n \rightarrow \infty} |y_0(n)| = 0 = a \sum_{k=0}^n h(k) = \text{constant}$$

⑥ If $x(n)$ is an energy signal, the output $y(n)$ will also be an energy signal.

$$y(n) = \sum_{k=-\infty}^{\infty} h(k) x(m-k)$$

$$\sum_{n=-\infty}^{\infty} y^2(n) = \sum_{n=-\infty}^{\infty} \left[\sum_{k=-\infty}^{\infty} h(k) x(m-k) \right]^2$$

$$= \sum_{k=-\infty}^{\infty} \sum_{l=-\infty}^{\infty} h(k) h(l) \sum_{m=-\infty}^{\infty} x(m-k) x(m-l)$$

$$\text{but } \sum_{m=-\infty}^{\infty} x(m-k) x(m-l) \leq \sum_{m=-\infty}^{\infty} x^2(m) |h(l)|$$

for LTI system $\sum_{k=-\infty}^{\infty} |h(k)| < \infty$

Hence $E[y] \leq n^2 E[x]$, so that $E[y] < 0$ if $E[x] < 0$.

⑦ The following IIP-OIP pairs have been observed during the operation of a time invariant S/I/M.

$$x_1(n) = \{1, 0, 1, 2\} \leftrightarrow y_1(n) = \{0, 1, 1, 2\}$$

$$x_2(n) = \{0, 0, 1, 3\} \leftrightarrow y_2(n) = \{0, 1, 1, 0, 2\}$$

$$x_3(n) = \{0, 1, 0, 0, 1\} \leftrightarrow y_3(n) = \{1, 2, 1, 1\}$$

Can you draw any conclusions regarding the linearity of the S/I/M. What is the impulse response of the S/I/M?

As this is a time-invariant S/I/M.

Sol As this is a time-invariant S/I/M, $y_2(n)$ should have only 5 elements and

$y_3(n)$ should have 4 elements.

So, it is non linear.

⑧ The following input-output pairs have been observed during the operation of a linear S/I/M.

$$x_1(n) = \{-1, 1, 2, 1, 1\} \rightarrow y_1(n) = \{1, 1, 2, 1, -1, 0, 1, 1\}$$

$$x_2(n) = \{1, 1, -1, 1, -1\} \rightarrow y_2(n) = \{-1, 1, 1, 0, 1, 2\}$$

$$x_3(n) = \{0, 1, 1\} \rightarrow y_3(n) = \{1, 1, 2, 1, 1\}$$

Can you draw any conclusions about the time invariance of this system?

$x_1(n) + x_2(n) = s(n)$
if system is linear the impulse response
of the sum is $y_1(n) + y_2(n) = \sum_{k=0}^2 b_k x_k = 1, 2, 1$
↑
if $s(n)$ were time invariant the response of
 $x_3(n)$ would be $\{3, 2, 1, 3, 1\}$

(12) The only available information about the sys consists of N input-output pairs of slgs.

$$y_i(n) = Y[x_i(n)], i = 1, 2, \dots, N.$$

a) what is the class of input slgs for which we can determine the o/p, using the information above, if the sys is known to be linear?
linear combination of signal in the form of

$$x_i^0(n), i = 1, 2, \dots, N$$

bcoz if we take $i = 1, 2$

$$y_1(n) = x_1(n) \Rightarrow y(n) = y_1(n) + y_2(n) = x_1(n) + x_2(n)$$

linear

b) Same repeat, if the sys is known to be time invariant

Any $x_i^0(n-k)$ where k is any integer,
 $i = 1, 2, \dots, N$

1st replace $n = n - n_0 \Rightarrow x_i^0(n - n_0 = k)$

$x(n)$ by $x(n - n_0) \Rightarrow x^0(n - k - n_0)$

(time invariant)

(13) Show that the necessary and sufficient condition for a relaxed LTI system to be

$$\text{BIBO stable if } \sum_{n=-\infty}^{\infty} |h(n)| \leq M_h < \infty$$

for some constant M_m

A sim to be BIBO stable only when bounded output should produce bounded input

$$y(n) = \sum_k h(k) \cdot x(n-k)$$

$$|y(n)| = \sum_k |h(k)| \cdot |x(n-k)|$$

$$= \sum_k |x(n-k)| \leq M_m \quad \{ \text{some cons}$$

$$\text{so } |y(n)| = M_m \cdot \sum_k |h(k)|$$

$|y(n)| < \infty$ for all n , if and only if

$$\sum_k |h(k)| < \infty \quad \text{so } \sum_{n=-\infty}^{\infty} |y(n)|$$

\rightarrow A sim to be BIBO stable only when bounded input produce bounded output.

$$y(n) = \sum_{k=-\infty}^{\infty} h(k) \cdot x(n-k) ; n \leq n-k$$

$$k \geq 0$$

$$|y(n)| = \sum_{k=-\infty}^{\infty} |h(k)| \cdot |x(n-k)|$$

as $\sum_{k=-\infty}^{\infty} |x(n-k)| \leq M_m$ for some constant

$$k \geq 0$$

$$|y(n)| = M_m \cdot \sum_{k=-\infty}^{\infty} |h(k)| ; n \leq n-k$$

$$k \geq 0$$

$|y(n)|$ is $< \infty$, if & only if $\sum_{k=-\infty}^{\infty} |h(k)| < \infty$

$$\text{So } \sum_{n=-\infty}^{\infty} |h(n)| < \infty$$

14. Show that

(a) A relaxed linear sim is causal if & only if for any input $x(n)$ such that

$$x(n) = 0 \text{ for } n < n_0 \Rightarrow y(n) = 0$$

for $n < n_0$

If a system is causal output depends only on the present and past input as $x(n) = 0$ for $n < n_0$. the $y(n)$ also becomes zero for $n < n_0$

- (b) A relaxed LTI Lm is causal if & only if $h(n) = 0$ for $n < 0$

$$y(n) = \sum_{k=-\infty}^{\infty} h(k) x(n-k)$$

for finite impulse response

$$h(n) = 0, n < 0 \text{ and } n \geq m$$

$$\text{so } y(n) \text{ reduces to } y(n) = \sum_{k=n}^{n-1} h(k) x(n-k)$$

If it is infinite impulse response

$$\text{then } y(n) = \sum_{k=0}^{\infty} h(k) x(n-k)$$

(15)

- (a) Show that for any real or complex constant a , and any finite integer numbers M and N , we have

$$\sum_{n=M}^N a^n = M a^M - \frac{a^{M+1} - a^{N+1}}{1-a}, \text{ if } a \neq 1$$

$$N-M+1, \text{ if } a=1$$

$$\text{for } a=1, \sum_{m=M}^N a^m = N-M+1$$

$$\text{for } a \neq 1, \sum_{m=M}^N a^m = \frac{a^M - a^{N+1}}{1-a}$$

$$(1-a)^N \sum_{m=M}^N a^m = a^M + a^{M+1} + a^{M+2} + \dots + a^N - a^{N+1}$$

$$= a^M - a^{N+1}$$

- (b) show that if $|a| < 1$, then

$$\sum_{m=0}^{\infty} a^m = \frac{1}{1-a}$$

For $M=0$, $|a| < 1$ and $N \rightarrow \infty$

$$\sum_{n=0}^{\infty} a^n = \frac{1}{1-a}, |a| < 1$$

(16) (a) If $y(n) = x(n) * h(n)$, show that $\sum_n y(n) = \sum_n x(n) \sum_n h(n)$

$$\text{where } \sum_n x(n) = \sum_{n=-\infty}^{\infty} x(n).$$

$$y(n) = \sum_k h(k) x(n-k)$$

$$\sum_n y(n) = \sum_n \sum_k h(k) x(n-k)$$

$$\sum_n y(n) = \sum_k h(k) \cdot \sum_{n=\infty}^{\infty} x(n-k)$$

$$\sum_n y(n) = \left(\sum_k h(k) \right) \left(\sum_n x(n) \right)$$

(b) Compute the convolution $y(n) = x(n) * h(n)$ of the following signals and check the correctness of the results by using the test

(a).

~~18/14/07~~

$$(i) x(n) = \{1, 2, 4\}, h(n) = \{1, 1, 1, 1\}$$

$$y(n) = \{1, 3, 7, 7, 7, 6, 4\}$$

$$\sum_n y(n) = 35; \quad \sum_n x(n) = 7, \quad \sum_n y(n) = 35$$

$$\sum_n y(n) = \sum_n x(n) \sum_n h(n) \quad \text{Tabular Method}$$

$$35 = 7 \times 5$$

$$35 = 35$$

$x(n)$	1	2	4
$h(n)$	1	1	2 / 4
1	1	1	2 / 4
1	2	2	4
1	1	2	4
1	1	2	4

② $x(n) = \{1, 2, -1\}$, $h(n) = x(n)$

$$x(n) = \{1, 2, -1\}, h(n) = \{1, 2, -1\}$$

$$y(n) = x(n) * h(n)$$

$$y(n) = \{1, 4, 2, -4, 1\}$$

$$\sum_n y(n) = 4; \sum_n x(n) = 2, \sum_n h(n) = 2$$

$$\sum_n y(n) = \sum_n x(n) * h(n)$$

$$4 = 4$$

③ $x(n) = \{0, 1, -2, 3, -4\}$, $h(n) = \{\frac{1}{2}, \frac{1}{2}, 1, \frac{1}{2}\}$

$$y(n) = \{0, \frac{1}{2}, -\frac{1}{2}, \frac{3}{2}, -2, 0, -\frac{5}{2}, 1, 2\}$$

$$\sum_n y(n) = -5, \sum_n x(n) = -2, \sum_n h(n) = \frac{5}{2}$$

$$\sum_n y(n) = \sum_n x(n) h(n)$$

$$-5 = 5$$

④ $x(n) = \{1, 2, 3, 4, 5\}$, $h(n) = \{1\}$

$$y(n) = \{1, 2, 3, 4, 5\}$$

$$\sum_n y(n) = \sum_n x(n) \sum_n h(n)$$

$$15 = 15$$

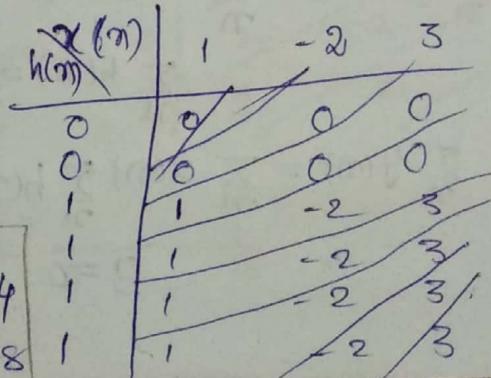
⑤ $x(n) = \{1, 2, 3\}$, $h(n) = \{0, 0, 1, 1, 1, 1\}$

$$y(n) = \{0, 0, 1, -1, 2, 2, 1, 3\}$$

$$\sum_n y(n) = 8; \sum_n x(n) = 6;$$

$$\sum_n h(n) = 4$$

$$\sum_n y(n) = \sum_n x(n) \cdot \sum_n h(n) 8 = 8$$



$$\textcircled{6} \quad x(n) = \{0, 0, 1, 1, 1, 1, 1\}, h(n) = \{1, -2, 3\}$$

$$y(n) = \{0, 0, 1, -1, 2, 2, 1, 3\}$$

$$\sum_n y(n) = 8, \sum_n x(n) = 4, \sum_n h(n) = 2$$

$$\sum_n y(n) = \sum_n x(n) \sum_n h(n)$$

$$8 = 8$$

	$x(n)$	$h(n)$
1.	0 0 + 1 1 1	
-2	0 0 -2 -2 -2	
3	0 0 3 3 3	

$$\textcircled{7} \quad x(n) = \{0, 1, 4, -3\},$$

$$h(n) = \{1, 0, -1, -1\}.$$

$$y(n) = \{0, 1, 4, -4, -5, -1, 3\}$$

$$\sum_n y(n) = -2; \sum_n x(n) = -2;$$

$$\sum_n h(n) = 1$$

$$\sum_n y(n) = \sum_n x(n) \sum_n h(n)$$

	$x(n)$	$h(n)$
1	0 + 4 -8	
0	0 1 4 -8	
-1	0 0 0 0	
-1	0 -1 -4 3	
0	-1 -4 3	

$$\textcircled{8} \quad x(n) = \{1, 1, 2\}, h(n) = 4(n)$$

$$y(n) = \{1, 2, 4, 3, 2\}$$

$$\sum_n y(n) = 12; \sum_n x(n) = 4;$$

$$\sum_n h(n) = 3.$$

$$\sum_n y(n) = \sum_n x(n) \sum_n h(n)$$

$$12 = 4 \times 3$$

	$x(n)$	$h(n)$
1	1 1 2	
1	1 1 2	
1	1 1 2	

$$\textcircled{9} \quad x(n) = \{1, 1, 0, 1, 1\}, h(n) = \{1, -2, -3, 4\}$$

$$y(n) = \{1, -1, -5, 2, 3, -5, 1, 4\}$$

$$\sum_n y(n) = 0, \sum_n x(n) = 4$$

$$\sum_n h(n) = 0$$

$$\sum_n y(n) = \sum_n x(n) \sum_n h(n)$$

$$0 = 0$$

	$x(n)$	$h(n)$
1	1 1 0 1 1	
-2	1 1 0 1 1	
-3	1 1 0 1 1	
-4	1 1 0 1 1	

(10) $x(n) = \{1, 2, 0, 2, 1\}$. $h(n) = u(n)$

 $y(n) = \{1, 4, 4, 4, 10, 4, 4, 4, 1\}$ $\sum_n y(n) = 36$ $\sum_n x(n) = 6$ $\sum_m h(m) = 6$

$$\sum_n y(n) = \sum_m x(m) \sum_n h(n)$$

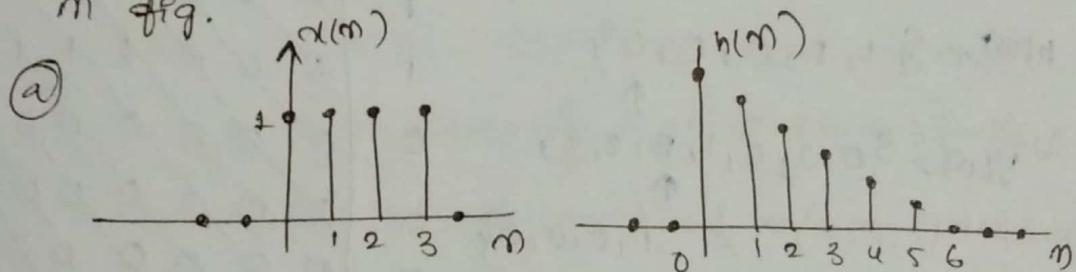
$$36 = 36$$

	$x(n)$	1	2	0	2	1
1	1	2	0	2	1	
2	2	4	0	4	2	
0	0	0	0	0	0	
2	2	4	0	4	2	
1	1	2	0	2	1	

(11) $x(n) = \left(\frac{1}{2}\right)^n u(n)$, $h(n) = \left(\frac{1}{4}\right)^n u(n)$

 $y(n) = \left(2\left(\frac{1}{2}\right)^n - \left(\frac{1}{4}\right)^n u(n)\right)$
 $\sum_n y(n) = \frac{8}{3}$; $\sum_n h(n) = \frac{4}{3}$; $\sum_n x(n) = 2$.

(17) Compute and plot convolutions $x(n) * h(n)$ and $h(n) * x(n)$ for the pairs of signals shown in fig.



$$x(n) = \{1, 1, 1, 1\}$$

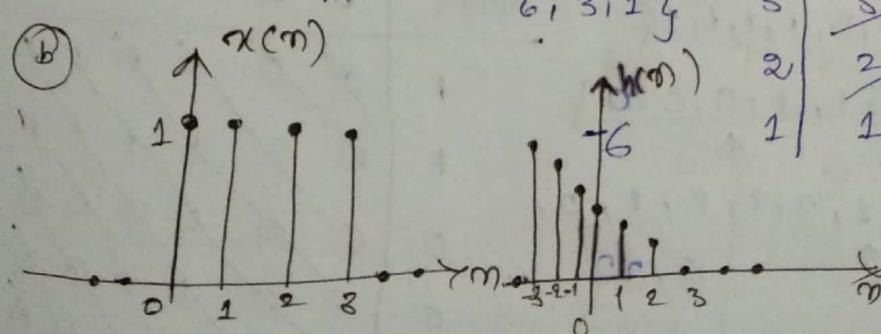
$$h(n) = \{6, 5, 4, 8, 2, 1, 0\}$$

$$y(n) = x(n) * h(n)$$

$$y(n) = \{6, 11, 15, 18, 14, 10,$$

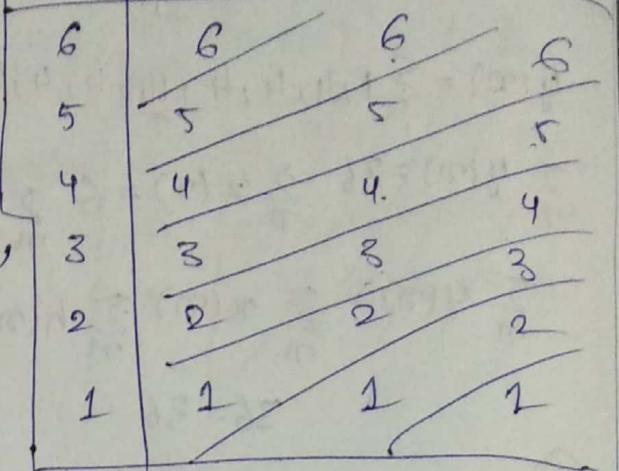
$$6, 3, 1\}$$

	$x(n)$	↓	$h(n)$	↓
6	6		6	6
5	5		5	5
4	4		4	4
3	3		3	3
2	2		2	2
1	1		1	1



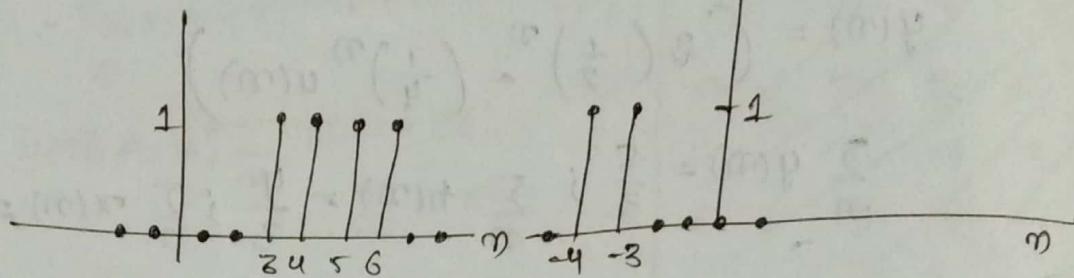
$$x(n) = \{6, 5, 4, 3, 2, 1, 0\}$$

$$y(n) = \{6, 11, 15, 18, 14, 10, \\ 6, 3, 1\}$$



③

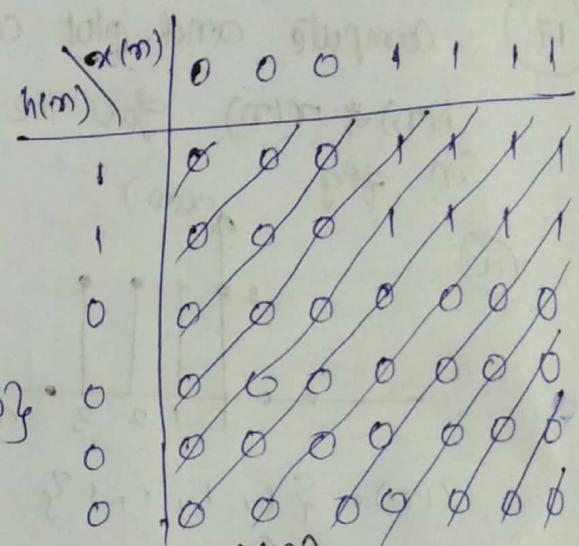
$$x(n)$$



$$x(n) = \{0, 0, 0, 1, 1, 1, 1\}$$

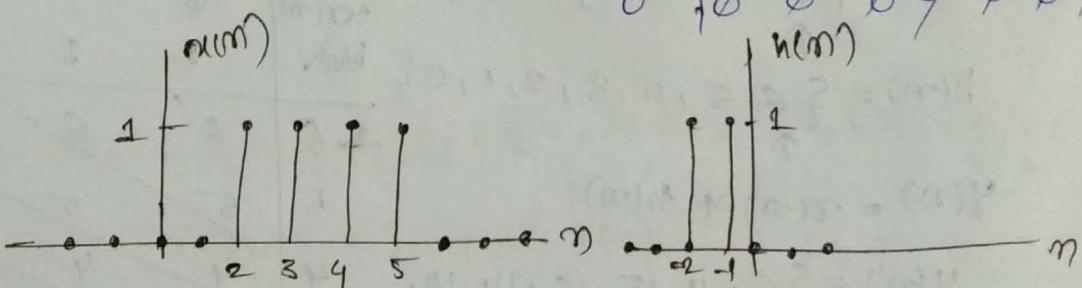
$$h(n) = \{1, 1, 0, 0, 0\}$$

$$y(n) = \{0, 0, 0, 1, 2, 2, 2, \\ 1, 0, 0, 0\}$$



④

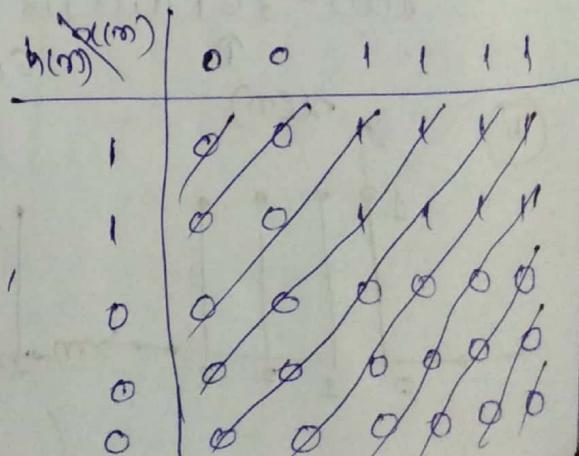
$$x(n)$$



$$x(n) = \{0, 0, 1, 1, 1, 1\}$$

$$h(n) = \{1, 1, 0, 0, 0\}$$

$$y(n) = \{0, 0, 1, 2, 2, 2, 1, 0, \\ 0, 0\}$$



(18) Determine and sketch the convolution $y(n)$ of the signals

$$x(n) = \begin{cases} \frac{1}{3}^n, & 0 \leq n \leq 6 \\ 0, & \text{elsewhere} \end{cases} \quad h(n) = \begin{cases} 1, & -2 \leq n \leq 2 \\ 0, & \text{elsewhere} \end{cases}$$

(a) Graphically (b) Analytically

(a) $x(n) = \{0, \frac{1}{3}, \frac{2}{3}, 1, \frac{4}{3}, \frac{5}{3}, 2\}$

$h(n) = \{1, 1, 1, 1, 1, 1\}$

$y(n) = \{0, \frac{1}{3}, \frac{1}{3}, 1, \frac{10}{3}, 5, \frac{20}{3}, 6, 5, \frac{14}{3}, 2\}$

$x(n)$	0	$\frac{1}{3}$	$\frac{2}{3}$	1	$\frac{4}{3}$	$\frac{5}{3}$	2	
$h(n)$	1	0	$\frac{1}{3}$	$\frac{2}{3}$	1	$\frac{4}{3}$	$\frac{5}{3}$	2
$y(n)$	1	0	$\frac{1}{3}$	$\frac{2}{3}$	1	$\frac{4}{3}$	$\frac{5}{3}$	2
$h(n)$	1	0	$\frac{1}{3}$	$\frac{2}{3}$	1	$\frac{4}{3}$	$\frac{5}{3}$	2
$y(n)$	1	0	$\frac{1}{3}$	$\frac{2}{3}$	1	$\frac{4}{3}$	$\frac{5}{3}$	2
$h(n)$	1	0	$\frac{1}{3}$	$\frac{2}{3}$	1	$\frac{4}{3}$	$\frac{5}{3}$	2

(b) $x(n) = \frac{1}{3} n [u(n) - u(n-7)]$

$$h(n) = u(n+2) - u(n-3)$$

$$\begin{aligned} y(n) &= \frac{1}{3} n [u(n) - u(n-7)] * u(n+2) - u(n-3) \\ &= \frac{1}{3} n [u(n) * u(n+2) - \frac{1}{3} n [u(n) * u(n-3)] - \cancel{\frac{1}{3} n [u(n-7) * u(n+2)] + \frac{1}{3} n [u(n-7) * u(n-3)]}] \end{aligned}$$

(20) Consider the following three operations

(a) Multiply the integer numbers ; 131 and 122

$$131 \times 122 = 15982$$

(b) Compute the convolution of signals : $\{1, 3, 1\} * \{1, 2, 2\}$.

$$y(n) = \{1, 5, 9, 8, 12\}$$

$x(n)$	1	3	1
$h(n)$	1	2	2
$y(n)$	1	2	2
$h(n)$	1	2	2
$y(n)$	1	3	6
$h(n)$	1	2	6
$y(n)$	1	2	2

c) Multiply the polynomials: $1+3z+2^2$ and $1+2z+2z^2$.

$$(z^2 + 3z + 1) \times (2z^2 + 2z + 1) \Rightarrow 2z^4 + 6z^3 + 2z^2 + 2z^3 \\ 6z^2 + 2z + z^2 + 3z + 1 \\ \Rightarrow 2z^4 + 8z^3 + 9z^2 + 5z + 1$$

d) Repeat part a) for the numbers 1.31 & 12.2

$$1.31 \times 12.2 = 15.982$$

e) Comment on your results.

These are different ways to perform convolution.

f) Compute the convolution y(m) of the following pairs of signals.

a) $x(m) = a^m u(m)$, $h(m) = b^m u(m)$ when $a \neq b$ &
when $a = b$.

$$y(m) = x(m) * h(m)$$

$$= a^m u(m) * b^m u(m)$$

$$= [a^m * b^m] u(m)$$

$$y(m) = \sum_{k=0}^m a^k u(k) b^{m-k} u(m-k)$$

$$= b^m \sum_{k=0}^m a^k u(k) b^{-k}$$

$$= b^m \cdot \sum_{k=0}^m (ab)^{-k}$$

$$\text{if } a \neq b \text{ then } y(m) = \frac{b^{m+1} - a^{m+1}}{b-a} u(m)$$

$$\text{if } a = b \Rightarrow b^m (n+1) u(m)$$

b) $x(m) = \begin{cases} 1; & m = -2, 0, 1 \\ 2; & m = -1 \\ 0; & \text{elsewhere} \end{cases}$

$$x(m) = \{1, 2, 1, 1\}$$

$$h(m) = s(m) - s(m-1) + s(m-4) \\ + s(m-5)$$

$$y(m) = \{1, 1, 1, -1, 0, 0, 3, 3, 2, 1\}$$

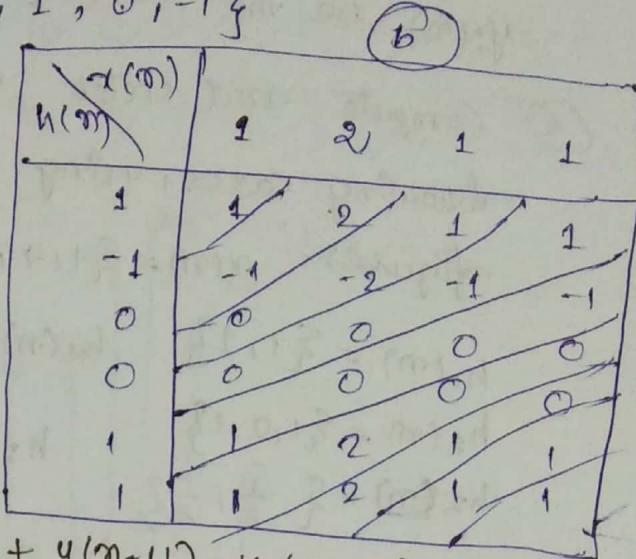
$$\textcircled{c} \quad x(n) = 4(n+1) - 4(n-4) - 8(n-5);$$

$$h(n) = [u(n+2) - u(n-3)] \cdot (3 - 1n)$$

$$x(n) = \{1, 1, 1, 1, 1, 0, -1\}$$

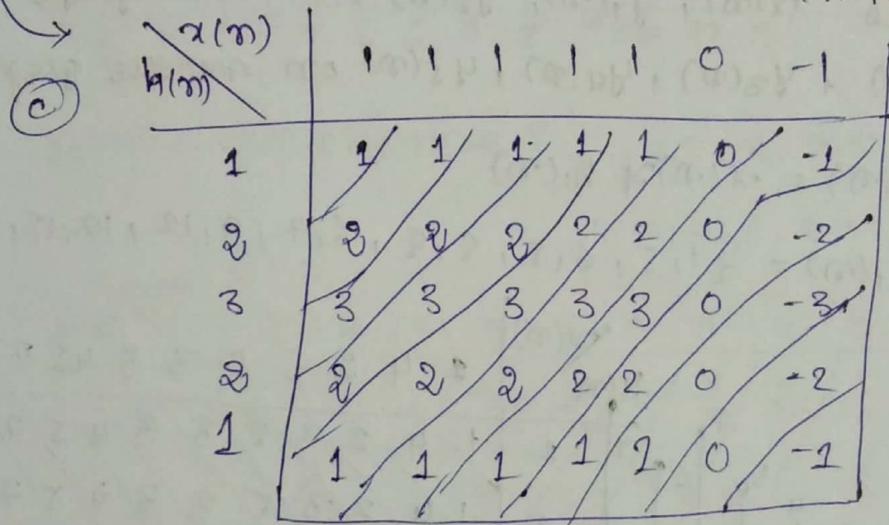
$$h(n) = \{1, 2, 3, 2, 1\}$$

$$y(n) = \{1, 3, 6, 8, 9, 8, \\ 5, 1, -2, 1, -2, 1\}$$



$$\textcircled{d} \quad x(n) = u(n) - u(n-5),$$

$$h(n) = u(n-2) - u(n-8) + u(n-11) - u(n-17)$$



(d)

$$x(n) = \{1, 1, 1, 1, 1\}$$

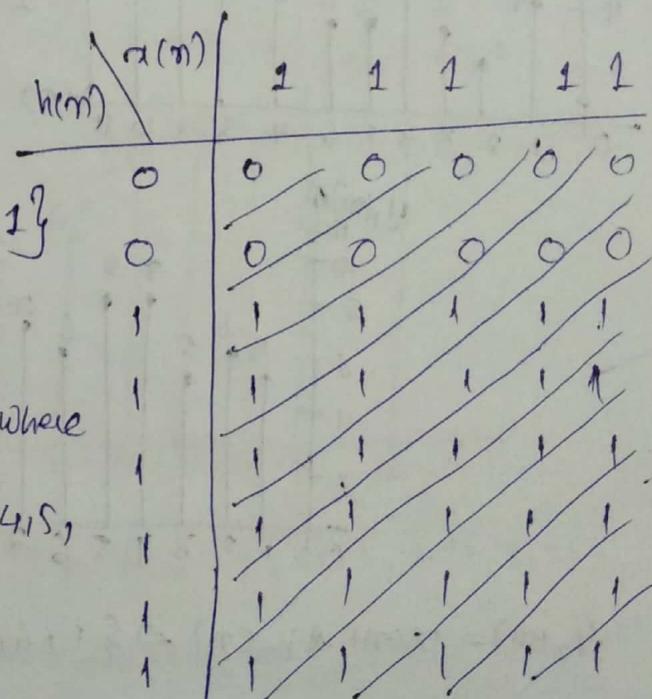
$$h'(n) = \{0, 0, 1, 1, 1, 1, 1\}$$

$$h(n) = h'(n) + h'(n-9)$$

$$y(n) = y'(n) + y'(n-9), \text{ where}$$

$$y'(n) = \{0, 0, 1, 2, 3, 4, 5, \\ 5, 4, 3, 2, 1\}$$

$$\{1, 2, 3, 4, 5, 5, 4, 3, 2, 1\}$$



(28) Let $x(n)$ be the QIP signal to a discrete-time FIR with impulse response $h_1(n)$ and let $y_1(n)$ be the corresponding output

① Compute and sketch $x(n)$ and $y_1(n)$ in the following cases, using the same scale in all figures. $x(n) = \{1, 4, 2, 8, 5, 8, 8, 4, 9, 7, 6, 9\}$

$$h_1(n) = \left\{ \frac{1}{2}, \frac{1}{2} \right\} \quad h_2(n) = \left\{ \frac{1}{4}, \frac{1}{2}, \frac{1}{4} \right\}$$

$$h_3(n) = \left\{ \frac{1}{2}, \frac{1}{2} \right\} \quad h_4(n) = \left\{ \frac{1}{4}, -\frac{1}{2}, \frac{1}{4} \right\}$$

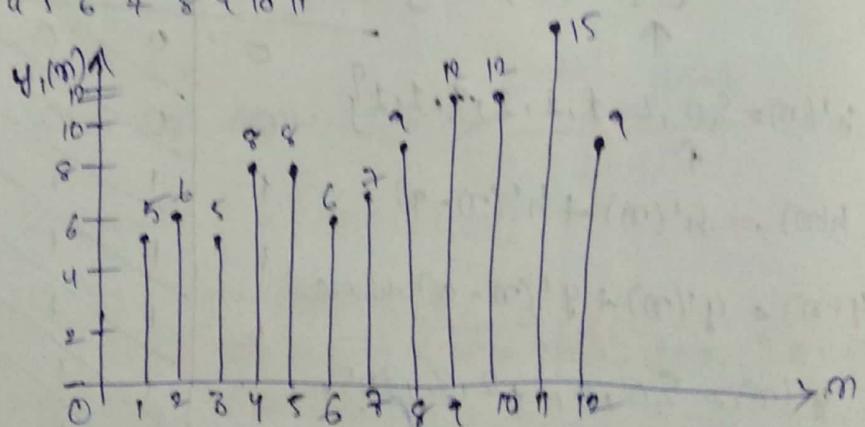
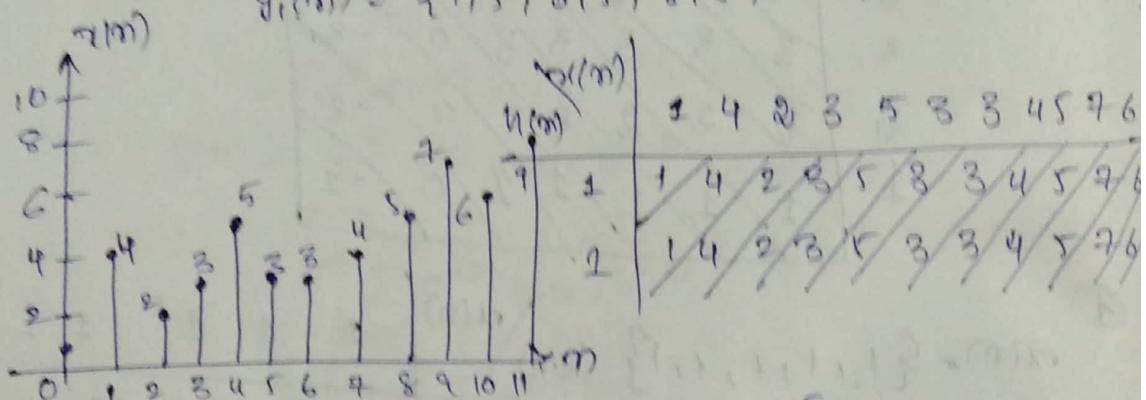
$$h_5(n) = \left\{ \frac{1}{2}, \frac{1}{2} \right\}$$

Sketch $x(n)$, $y_1(n)$, $y_2(n)$ on one graph
and $x(n) + y_3(n)$, $y_4(n) + y_5(n)$ on another graph.

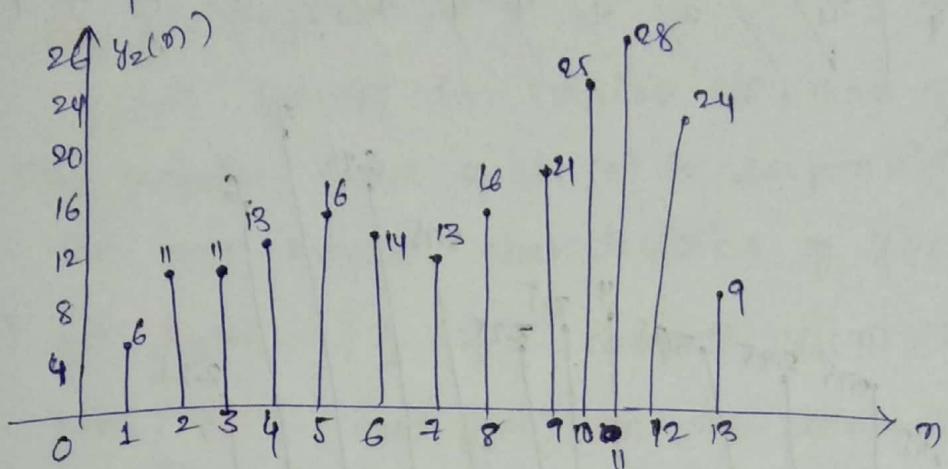
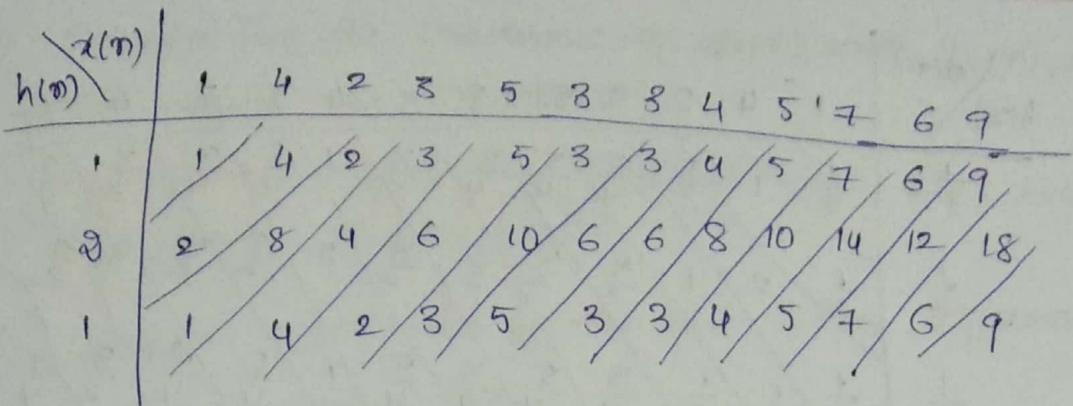
Sol

$$y_1(n) = x(n) * h_1(n)$$

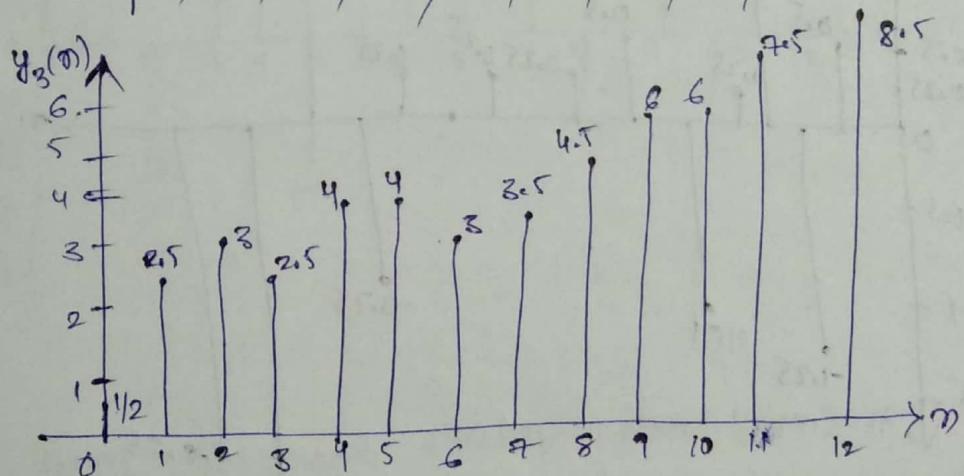
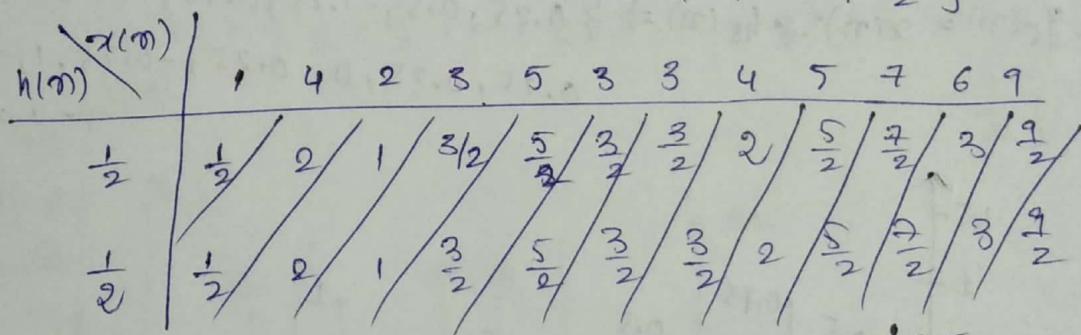
$$y_1(n) = \{1, 5, 6, 5, 8, 8, 6, 7, 9, 12, 12, 15, 9\}$$



$$y_2(n) = x(n) * h_2(n) \Rightarrow \{1, 6, 11, 11, 13, 16, 14, 18, 16, 21, 25, 28, 24, 9\}$$

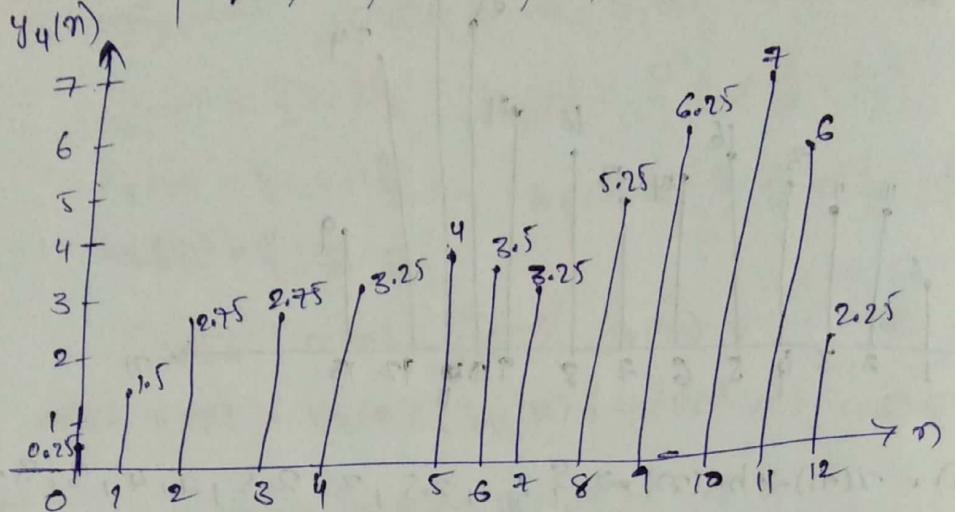


$$y_3(n) = x(n) * h_3(n) \Rightarrow \left\{ \frac{1}{2}, 2.5, 3, 2.5, 4, 4, 3, 3.5, 4.5, 6, 6, 7.5, \frac{9}{2} \right\}$$

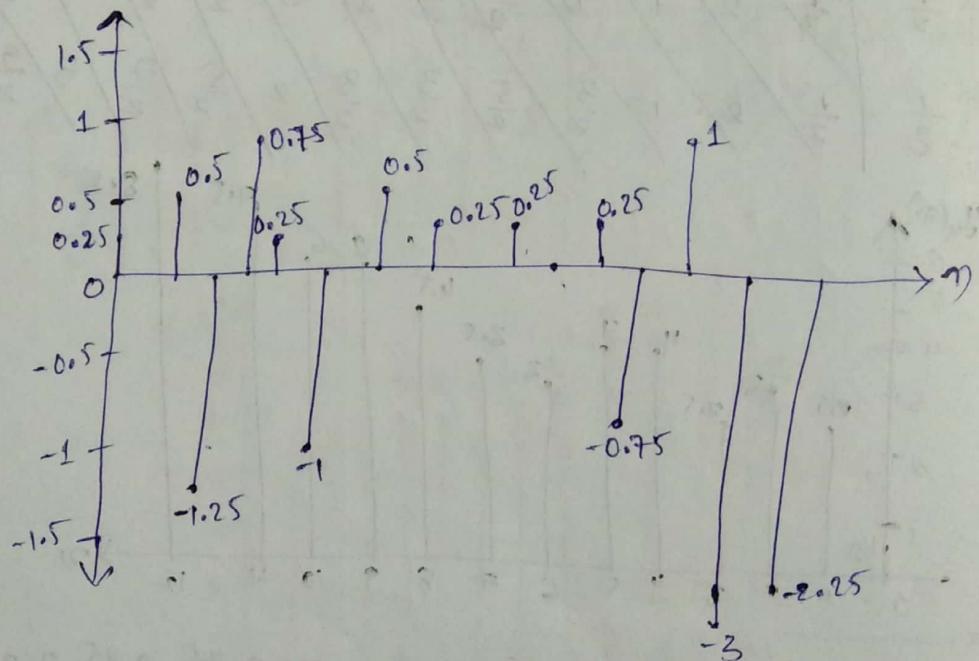


$$\Rightarrow y_4(n) = x(n) * h_4(n) \Rightarrow \left\{ 0.25, 1.5, 2.75, 2.75, 3.25, 4, 3.5, 3.25, 5.25, 6.25, 7, 6, 2.25 \right\}$$

$x(n)$	1	4	2	3	5	3	3	4	5	7	6	9
$\frac{1}{4}$	$\frac{1}{4}$	1	$\frac{1}{2}$	$\frac{3}{4}$	$\frac{5}{4}$	$\frac{3}{4}$	$\frac{3}{4}$	1	$\frac{5}{4}$	$\frac{9}{4}$	$\frac{3}{2}$	$\frac{9}{4}$
$\frac{1}{2}$	$\frac{1}{2}$	2	1	$\frac{3}{2}$	$\frac{5}{2}$	$\frac{3}{2}$	$\frac{3}{2}$	2	$\frac{5}{2}$	$\frac{7}{2}$	3	$\frac{9}{2}$
$\frac{1}{4}$	$\frac{1}{4}$	1	$\frac{1}{2}$	$\frac{3}{4}$	$\frac{5}{4}$	$\frac{3}{4}$	$\frac{3}{4}$	1	$\frac{5}{4}$	$\frac{7}{4}$	$\frac{3}{2}$	$\frac{9}{4}$



$$\Rightarrow y_5(n) = x(n) * h_5(n) \Rightarrow \{ 0.25, 0.5, -1.25, 0.75, 0.25, -1, 0.5, \\ 0.25, 0.25, 0, 0.25, -0.75, 1, -3, \\ -2.25 \}$$



(b) What is the difference b/w $y_1(n)$ & $y_2(n)$ and b/w $y_3(n)$ & $y_4(n)$?

$$y_3(n) = \frac{1}{2} y_1(n); \quad h_3(n) = \frac{1}{2} h_1(n)$$

$$y_4(n) = \frac{1}{4} y_2(n); \quad h_4(n) = \frac{1}{4} h_2(n)$$

(c) Comment on the smoothness of $y_2(n)$ and $y_4(n)$. Which factors affect the smoothness?

$y_2(n)$ and $y_4(n)$ are smoother than $y_1(n)$ becuz of smaller scalar factor.

(d) Compare $y_4(n)$ with $y_5(n)$. What is the difference? Can you explain it?

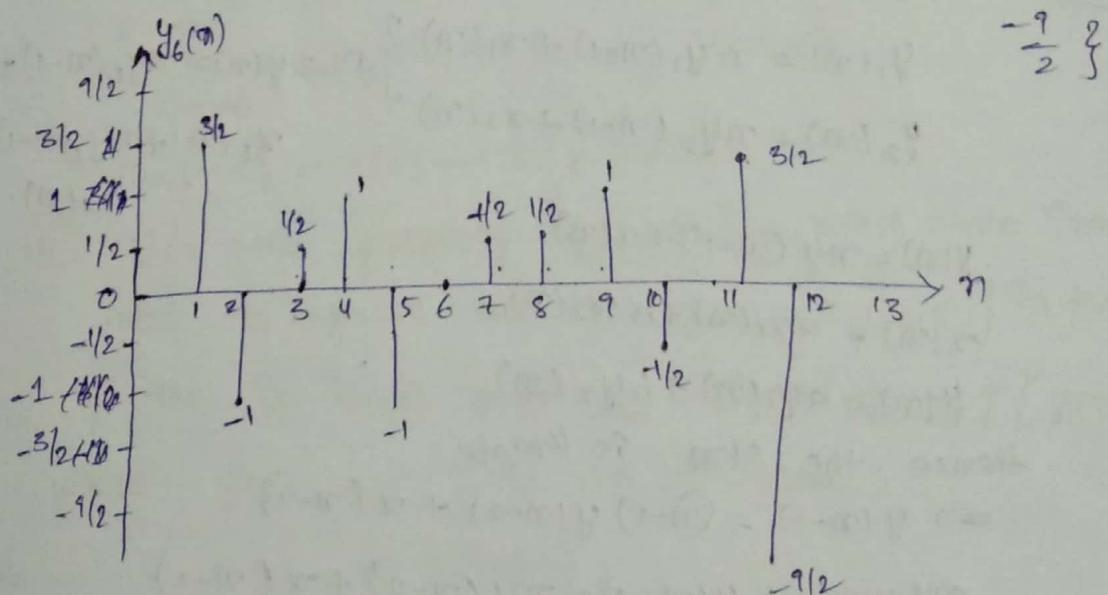
$y_4(n)$ results in smaller OIP, than $y_5(n)$.

The negative value of $h_5(0)$ is responsible for the non-smooth characteristics of $y_5(n)$.

(e) Let $h_6(n) = \left\{ \frac{1}{2}, -\frac{1}{2} \right\}$ compute $y_6(n)$. Sketch $x(n)$, $y_2(n)$ and $y_6(n)$ on the same figure and comment on the results.

$$y_6(n) = x(n) * h_6(n)$$

$$y_6(n) = \left\{ \frac{1}{2}, \frac{3}{2}, -1, \frac{1}{2}, 1, 1; -1, 0, \frac{1}{2}, \frac{1}{2}, 1, \frac{1}{2}, \frac{3}{2}, -\frac{9}{2} \right\}$$



$y_2(n)$ is more smoother than $y_6(n)$

(28) Express the OIP $y(n)$ of a linear time-invariant S/I system with impulse response $h(n)$ in terms of its step response $s(n) = h(n) * u(n)$ and the input $x(n)$.

Sol:-

$$\begin{aligned}
 s(n) &= u(n) - u(n-1) \\
 h(n) &= h(n) * s(n) \\
 &= h(n) * [u(n) - u(n-1)] \\
 &= h(n) * u(n) - h(n) * u(n-1) \\
 &= S(n) - S(n-1)
 \end{aligned}$$

then $y(n) = h(n) * x(n)$

$$\begin{aligned}
 &= [S(n) - S(n-1)] * x(n) \\
 &= S(n) * x(n) - S(n-1) * x(n)
 \end{aligned}$$

Q4) The discrete-time system

$y(n) = ny(n-1) + x(n)$, $n \geq 0$

is at rest [i.e., $y(-1) = 0$]. Check if the SLM is linear time invariant and BIBO stable.

Ans

$$\begin{aligned}
 y(n) &= ny(n-1) + x(n), n \geq 0 \\
 y_1(n) &= ny_1(n-1) + x_1(n) \\
 y_2(n) &= ny_2(n-1) + x_2(n)
 \end{aligned}
 \quad \left. \begin{array}{l} \text{④} \\ \Rightarrow y(n) = ny_1(n-1) + \\ x_1(n) + ny_2(n-1) + \\ x_2(n). \end{array} \right.$$

$$y(n) = ny(n-1) + x(n)$$

$$x(n) = ax_1(n) + bx_2(n)$$

$$y(n) = ay_1(n) + by_2(n)$$

Hence the SLM is linear.

$$\Rightarrow y(n-1) = (n-1)y(n-2) + x(n-1)$$

$$\text{delayed} \Rightarrow y(n-1) = ny(n-2) + x(n-1)$$

So the SLM is TI.

\Rightarrow If $x(n) = u(n)$, then $|x(n)| \leq 1$, for this bounded I/P, O/P is $y(0) = 0$, $y(1) = 2$, $y(2) = 5$ — — — , unbounded. So SLM is unstable

(25) Consider the signal $y(n) = a^n u(n)$, $0 < a < 1$.

- a) Show that any sequence $x(n)$ can be decomposed as $x(n) = \sum_{k=-\infty}^{\infty} c_k \cdot y(n-k)$ and express c_k in terms of $x(n)$.

$$\underline{\text{Sol}} \quad S(n) = y(n) - a y(n-1)$$

$$S(n-k) = y(n-k) - a y(n-k-1)$$

$$x(n) = \sum_{k=-\infty}^{\infty} x(k) \cdot S(n-k)$$

$$= \sum_{k=-\infty}^{+\infty} x(k) \cdot [y(n-k) - a y(n-k-1)]$$

$$x(n) = \sum_{k=-\infty}^{+\infty} x(k) y(n-k) - a \sum_{k=-\infty}^{+\infty} x(k) y(n-k-1)$$

$$x(n) = \sum_{k=-\infty}^{+\infty} x(k) y(n-k) - a \sum_{k=-\infty}^{+\infty} x(k-1) y(n-k)$$

$$= \sum_{k=-\infty}^{\infty} [x(k) - a x(k-1)] y(n-k)$$

$$\text{Thus } c_k = x(k) - a x(k-1)$$

- b) Use the property of linearity and time invariance to express the O/P $y(n) = T[x(n)]$ in terms of the I/P $x(n)$ and the signal $g(n) = T[y(n)]$,

where $\underline{\text{Sol}}$ $y[n]$ is an LTI S/I.

$$y(n) = y \cdot [x(n)]$$

$$= y \cdot \left[\sum_{k=-\infty}^{+\infty} c_k \cdot y(n-k) \right]$$

$$= \sum_{k=-\infty}^{+\infty} c_k y \cdot [y(n-k)]$$

$$= \sum_{k=-\infty}^{+\infty} c_k g(n-k)$$

(2) Express the impulse response $h(n) = \gamma \{s(n)\}$

in terms of $g(n)$.

$$h(n) = \gamma \{s(n)\}$$

$$h(n) = \gamma [\gamma(n) - a\gamma(n-1)]$$

$$= g(n) - ag(n-1)$$

(2b) Determine the zero-input response of the S/I system described by the second-order difference equation

$$x(n) + 3y(n-1) + 4y(n-2) = 0$$

Ans

$$\text{with } x(n) = 0$$

$$-3y(n-1) - 4y(n-2) = 0 \quad (\rightarrow (-3))$$

$$y(n-1) + \frac{4}{3}y(n-2) = 0$$

$$\text{at } n=0; \quad y(-1) = -\frac{4}{3}y(-2)$$

$$\text{at } n=1; \quad y(0) = \left(\frac{-4}{3}\right)y(-1) = \left(\frac{-4}{3}\right)^2 y(-2)$$

$$y(1) = \left(\frac{-4}{3}\right)^3 y(-2)$$

$$y(k) = \left(\frac{-4}{3}\right)^{k+2} y(-2)$$

↳ zero-input response

(2c) Determine the particular solution of the difference eqn

$$y(n) = \frac{5}{6}y(n-1) + \frac{1}{6}y(n-2) + x(n) \text{ when the forcing fun. is } x(n) = 2^n u(n).$$

$$y(n) = \frac{5}{6}y(n-1) + \frac{1}{6}y(n-2) + x(n)$$

$$x(n) = y(n) - \frac{5}{6}y(n-1) - \frac{1}{6}y(n-2)$$

Characteristic eqn is,

$$x^2 - \frac{5}{6}x + \frac{1}{6} = 0; \quad x = 1/2, 1/3$$

$$\text{So } y_n(n) = c_1 \left(\frac{1}{2}\right)^n + c_2 \left(\frac{1}{3}\right)^n$$

$$u(n) = 2^n u(n)$$

$$y_p(n) = k(2^n) u(n)$$

$$\text{So, } k(2^n) u(n) - k\left(\frac{5}{6}\right)(2^{n-1}) u(n-1) + k\left(\frac{1}{6}\right)(2^{n-2}) u(n-2) = 2^n u(n)$$

for $n=2$,

$$4k - \frac{5k}{3} + \frac{k}{6} = 4$$

$$k = \frac{8}{5}$$

Total solution is

$$y_p(n) + y_n(n) = y(n).$$

$$y(n) = \frac{8}{5}(2^n) u(n) + c_1 \left(\frac{1}{2}\right)^n u(n) + c_2 \left(\frac{1}{3}\right)^n u(n)$$

$$\text{Assume } y(-2) = y(-1) = 0 \quad \text{so } y(0) = 1$$

$$\text{then } y(1) = \frac{5}{6}y(0) + 2 = \frac{17}{6}$$

$$\frac{8}{5} + c_1 + c_2 = 1$$

$$c_1 + c_2 = \frac{3}{5} \quad \text{--- (1)}$$

$$\frac{16}{5} + \frac{1}{2}c_1 + \frac{1}{3}c_2 = \frac{17}{6}$$

$$3c_1 + 2c_2 = -\frac{11}{5} \quad \text{--- (2)}$$

By solving (1) & (2)

$$c_1 = -1, c_2 = \frac{2}{5}$$

So the total solution is

$$y(n) = \left[\frac{8}{5}(2^n) - \left(\frac{1}{2}\right)^n + \frac{2}{5} \left(\frac{1}{3}\right)^n \right] u(n)$$

(28)

$$\text{at } y(-1) = 1$$

$$\text{The given } e^{-n} \text{ is } y(n) = (-a)^{n+1} + \frac{(1 - (-a)^{n+1})}{1+a}$$

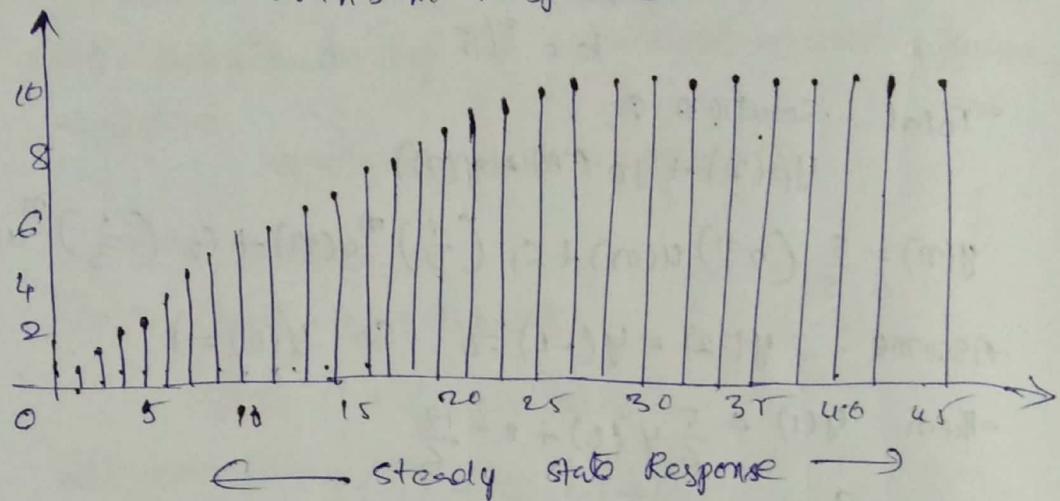
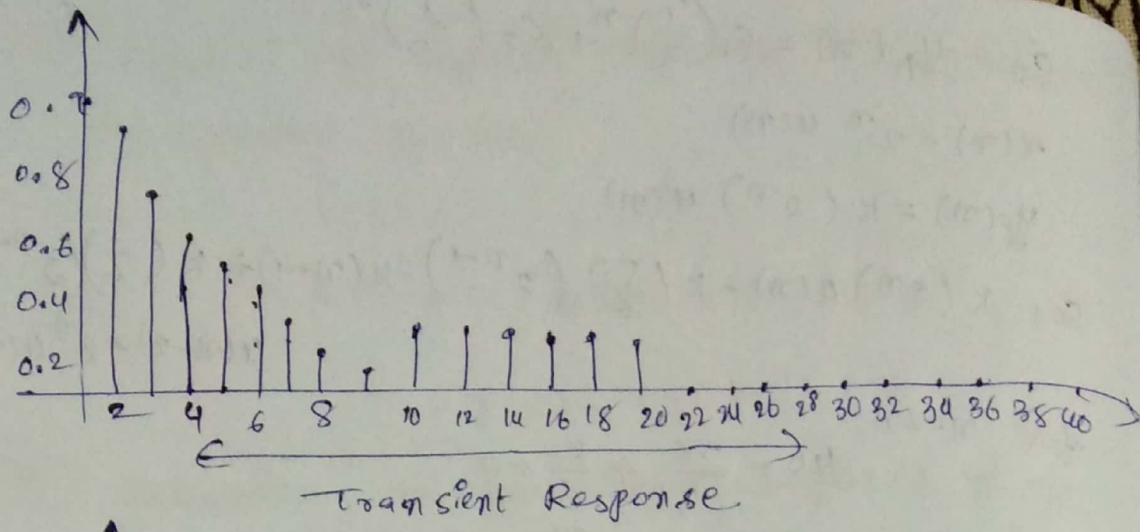
$$y(n) = y_{z1}(n) + y_{z2}(n)$$

= Transient + Steady state

$$y(0) = 0.9$$

$$y(1) = 0.87$$

$$y(2) = 0.72$$



$$\begin{aligned}
 \textcircled{29} \quad h_1(n) &= a^n [u(n) - u(n-N)] \quad \& \quad h_2(n) = [u(n) - u(n-M)] \\
 h(n) &= h_1(n) * h_2(n) \\
 &= \sum_{k=-\infty}^{\infty} a^k \cdot [u(k) u(k-N)] [u(n-k) - u(n-k-M)] \\
 &= \sum_{k=-\infty}^{\infty} a^k u(k) u(n-k) - \sum_{k=-\infty}^{\infty} a^k u(k) u(n-k-M) - \\
 &\quad \sum_{k=-\infty}^{\infty} a^k u(k-N) u(n-k) + \sum_{k=-\infty}^{\infty} a^k u(k-N) u(n-k-M) \\
 &= \left(\sum_{k=0}^n a^k - \sum_{k=0}^{n-M} a^k \right) - \left(\sum_{k=N}^n a^k - \sum_{k=N}^{n-M} a^k \right)
 \end{aligned}$$

$$u(n) = 0$$

$$\begin{aligned}
 \textcircled{30} \quad y(n) - 3y(n-1) + 4y(n-2) &= x(n) + 2x(n-1) \quad \text{to 91P} \\
 91P \quad x(n) &= 4^n u(n),
 \end{aligned}$$

Ans

$$y(n) - 3y(n-1) - 4y(n-2) = x(n) + 2x(n-1)$$

characteristic equation is

$$\lambda^2 - 3\lambda - 4 = 0$$

$$\lambda = 4, -1$$

$$\text{so } y_h(n) = c_1 4^n + c_2 (-1)^n$$

$$x(n) = 4^n u(n)$$

$$y_p(n) = k n 4^n u(n)$$

$$k n 4^n u(n) - 3k(n-1) 4^{n-1} u(n-1) - 4k(n-2) 4^{n-2} u(n-2) \\ = 4^n u(n) + 2(4)^{n-1} u(n-1)$$

$$\text{for } n=2, \quad k(32-12) = 4^2 + 8 = 24 \rightarrow k = 6/5$$

The total solution is,

$$y(n) = y_p(n) + y_h(n)$$

$$= \left[\frac{6}{5} n 4^n + c_1 4^n + c_2 (-1)^n \right] u(n)$$

to find c_1 & c_2 , let $y(-2) \equiv y(-1) = 0$ then

$$y(0) = 1$$

$$y(1) = 3y(0) + 4 + 2 = 9$$

$$c_2 + c_1 = 1 \quad \text{--- (1)}$$

$$\frac{24}{5} + 4c_1 + c_2 = 9 \Rightarrow 4c_1 + c_2 = \frac{21}{5} \quad \text{--- (2)}$$

from (1) & (2),

$$c_1 = \frac{26}{25} \quad \& \quad c_2 = -\frac{1}{25}$$

$$\text{so } y(n) = \left[\frac{6}{5} n 4^n + \frac{26}{25} 4^n - \frac{1}{25} (-1)^n \right] u(n).$$

(31)

$$y(n) - 3y(n-1) - 4y(n-2) = x(n) + 2x(n-1)$$

Sol

Characteristic eq :- $\lambda^2 - 3\lambda - 4 = 0$

$$\lambda = -4, 1$$

$$y_h(n) = c_1 4^n + c_2 (-1)^n$$

$$x(n) = S(n)$$

$$y(0) = 1 \text{ and } y(1) = 3y(0) + 2 \quad y(1) = 5$$

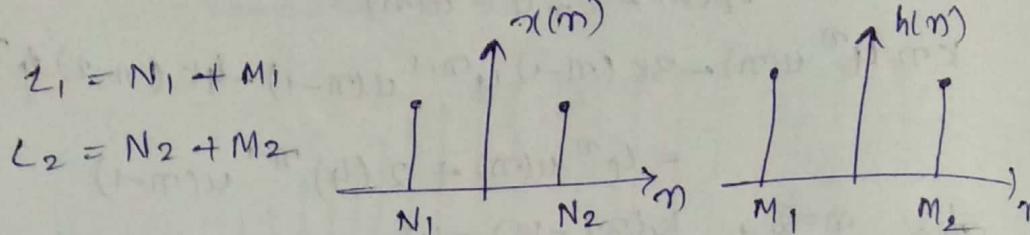
$$\text{So } c_1 + c_2 = 1 \quad \textcircled{1}$$

$$4c_1 - c_2 = 5 \quad \textcircled{2}$$

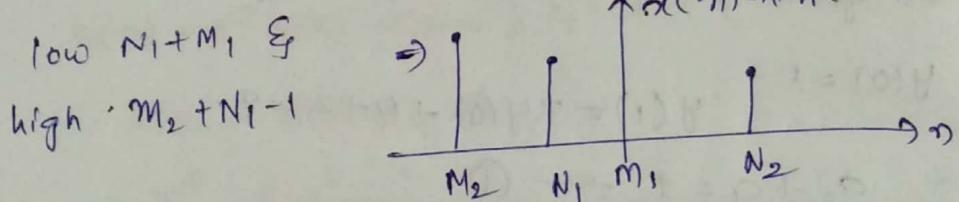
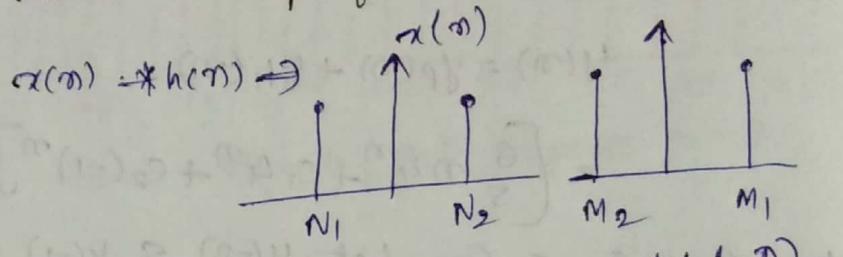
from $\textcircled{1} \& \textcircled{2}$ $c_1 = \frac{6}{5}$ & $c_2 = -\frac{1}{5}$

$$\therefore h(n) = \left[\frac{6}{5} 4^n - \frac{1}{5} (-1)^n \right] u(n)$$

(22) a) determine the range $l_1 \leq n \leq l_2$ of their convolution in terms of N_1, N_2, M_1 & M_2 .



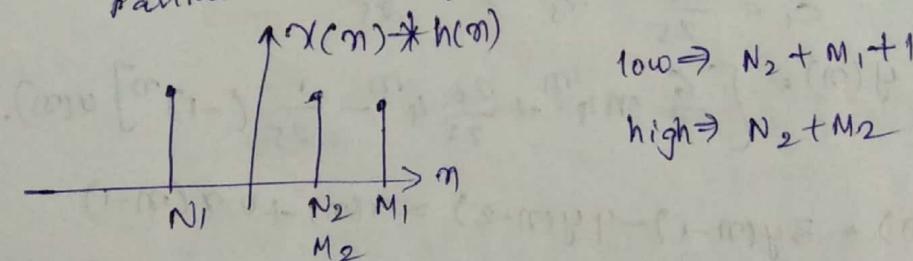
b) Partial overlap from left



If fully overlap then

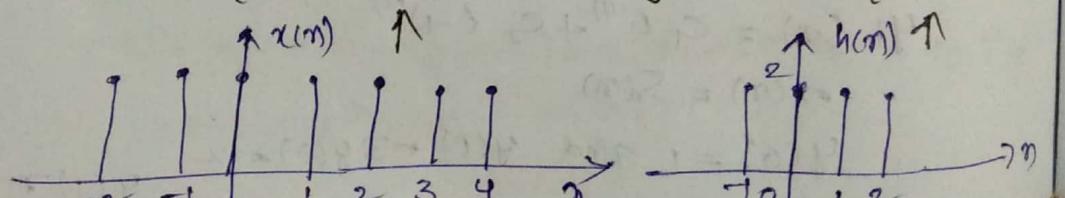
$N_1 + M_2$ (low) & high $N_2 + M_1$

Partial overlap from right



If fully overlaped high $N_2 + M_1$; low $= N_1 + M_2$

c) $x(n) = \{1, 1, 1, 1, 1, 1, 1, 1\}$ $h(n) = \{2, 2, 2, 2\}$



$$N_1 = -2 \quad N_2 = 4 \quad M_1 = -1, M_2 = 2$$

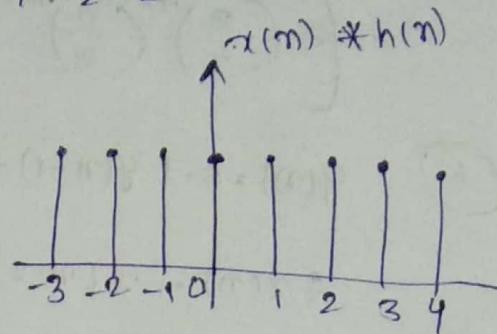
Partial overlap from left

$$\text{low } N_1 + M_1 = -3$$

$$\text{high } M_2 + N_1 - 1 = 2 - 2 - 1 = -1$$

full overlap $n=0, n=3$

partial right; $n=4, n=6, L_2 = 6$.



$$(23) \quad a) \quad y(n) = 0.6 y(n-1) - 0.08 y(n-2) + x(n)$$

$$\underline{\text{sol}} \quad x(n) = y(n) - 0.6 y(n-1) - 0.08 y(n-2)$$

characteristic equation

$$\lambda^2 - 0.6\lambda + 0.08 = 0$$

$$\lambda = \frac{1}{2}, \frac{2}{5}$$

$$y_h(n) = C_1 \left(\frac{1}{2}\right)^n + C_2 \left(\frac{2}{5}\right)^n$$

Impulse response $x(n) = \delta(n)$ with $y(0)=1$

$$y(1) = -0.6 y(0) = 0 \Rightarrow y(1) = 0.6$$

$$\text{so } C_1 + C_2 = 1 \quad \textcircled{1}$$

$$\frac{1}{2} C_1 + \frac{2}{5} C_2 = 0.6 \quad \textcircled{2}$$

$$\text{from } \textcircled{1} \& \textcircled{2} \quad C_1 = -1, C_2 = 3$$

$$h(n) = \left[-\left(\frac{1}{2}\right)^n + 2\left(\frac{2}{5}\right)^n \right] u(n)$$

step response $x(n) = u(n)$

$$s(n) = \sum_{k=0}^n y_h(n-k), \quad n \geq 0$$

$$= \sum_{k=0}^n \left[2\left(\frac{2}{5}\right)^{n-k} - \left(\frac{1}{2}\right)^{n-k} \right]$$

$$= 2\left(\frac{2}{5}\right)^n \sum_{k=0}^n \left(\frac{5}{2}\right)^k - \left(\frac{1}{2}\right)^n \sum_{k=0}^n \left(\frac{1}{5}\right)^k$$

$$= \left[2 \left(\frac{1}{2} \right)^n \left(\frac{5}{2} \right)^{n+1} - 1 \right] - \left[\left(\frac{1}{5} \right)^n \left(5^{n+1} - 1 \right) \right] u(n)$$

(b) $y(n) = 0.7y(n-1) + 0.1y(n-2) + 2x(n) - x(n-2)$

$$2x(n) - x(n-2) = y(n) - 0.7y(n-1) + 0.1y(n-2)$$

characteristic equation

$$\lambda^2 - 0.7\lambda + 0.1 = 0$$

$$\lambda = \frac{1}{2} + \frac{1}{5}$$

$$y_n(n) = c_1 \left(\frac{1}{2} \right)^n + c_2 \left(\frac{1}{5} \right)^n$$

Impulse Response $x(n) = \delta(n)$, $y(0) = 2$

$$y(1) = -0.7y(0) = 0 \Rightarrow y(1) = 1.4$$

$$c_1 + c_2 = 2$$

$$\frac{1}{2}c_1 + \frac{1}{5}c_2 = \frac{7}{5} \quad \textcircled{1}$$

$$c_1 + \frac{2}{5}c_2 = \frac{14}{5} \quad \textcircled{2}$$

Solving \textcircled{1} & \textcircled{2}

$$c_1 = \frac{10}{3}, c_2 = -\frac{4}{3}$$

$$\text{So } h(n) = \left[\frac{10}{3} \left(\frac{1}{2} \right)^n - \frac{4}{3} \left(\frac{1}{5} \right)^n \right] u(n)$$

Step Response $s(n) = \sum_{k=0}^n h(n-k)$

$$= \frac{10}{3} \sum_{k=0}^n \left(\frac{1}{2} \right)^{n+k} - \frac{4}{3} \sum_{k=0}^n \left(\frac{1}{5} \right)^{n+k}$$

$$= \frac{10}{3} \left(\frac{1}{2} \right)^n \sum_{k=0}^n 2^k - \frac{4}{3} \left(\frac{1}{5} \right)^n \sum_{k=0}^n 5^k$$

$$= \frac{10}{3} \cdot \left[\frac{1}{2}^n (2^{n+1} - 1) u(n) \right] - \frac{4}{3} \cdot \left[\frac{1}{5}^n (5^{n+1} - 1) u(n) \right]$$

(e) $h(n) = \begin{cases} \left(\frac{1}{2} \right)^n, & 0 \leq n \leq 4 \\ 0, & \text{elsewhere} \end{cases}$

$$y(n) = \begin{cases} 1, 2, 2.5, 3, 3.3, 2, 1, 0 \end{cases}$$

$$h(n) = \left\{ 1, \frac{1}{2}, \frac{1}{4}, \frac{1}{8}, \frac{1}{16}, \dots \right\}$$

$$y(n) = \left\{ 1, 2, 2.5, 3, 3, 2, 1, 0, \dots \right\}$$

$$y(0) = x(0) \cdot h(0)$$

$$y(0) = x(0) \cdot 1 \rightarrow x(0) = 1$$

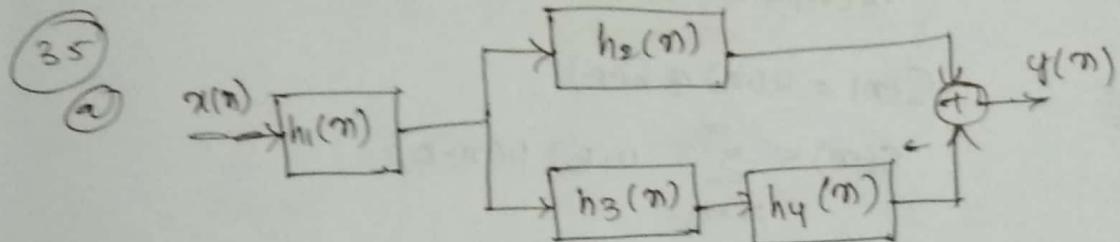
$$y(1) = x(1) + h(1) \cdot x(0)$$

$$\therefore x(1) + \frac{1}{2} (1) \Rightarrow x(1) = \frac{3}{2}$$

$$y(2) = x(2) + h(2) \cdot x(1) + h(1) \cdot x(0)$$

$$2.5 = x(2) + \frac{1}{4} \left(\frac{3}{2} \right) + \frac{1}{2} (1)$$

$$\text{so } x(n) = \left\{ 1, \frac{5}{8}, \frac{3}{4}, \frac{7}{4}, \frac{3}{2}, \dots \right\}$$



② Express the overall impulse response in terms of $h_1(n)$, $h_2(n)$, $h_3(n)$, and $h_4(n)$.

$$\underline{\text{Sol}} \quad h(n) = h_1(n) * [h_2(n) - \{h_3(n) * h_4(n)\}]$$

③ Determine $h(n)$ when $h_1(n) = \left\{ \frac{1}{2}, \frac{1}{4}, \frac{1}{8} \right\}$

$$h_2(n) = h_3(n) = (n+1) u(n)$$

$$h_4(n) = 8(n-2)$$

$$\underline{\text{Sol:}} \quad h_3(n) * h_4(n) = (n+1) u(n) * 8(n-2) \\ = (n+1) u(n-2) = (n+1) u(n-2)$$

$$h_2(n) - [h_3(n) * h_4(n)] = (n+1) u(n) - (n+1) u(n-2) \\ = 2 u(n) - 8(n)$$

$$h_1(n) = \frac{1}{2} s(n) + \frac{1}{4} \cdot s(n-1) + \frac{1}{8} \cdot s(n-2)$$

$$h(n) = \left[\frac{1}{2} \cdot s(n) + \frac{1}{4} \cdot s(n-1) + \frac{1}{8} \cdot s(n-2) \right] * [2u(n) - 8(n)] \\ = \frac{1}{2} s(n) + \frac{5}{4} s(n-1) + 2 s(n-2) + \frac{5}{2} u(n-3)$$

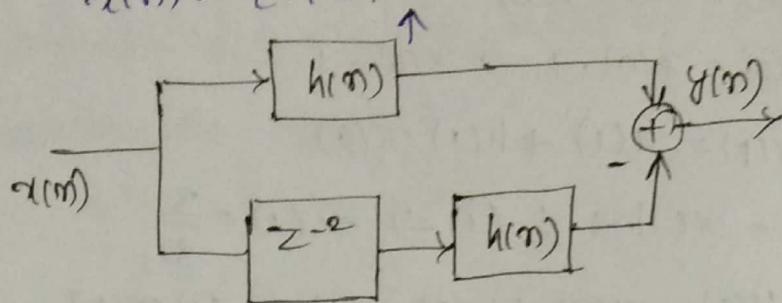
① Determine the response of the SLM in part (e)

$$\text{if } x(n) = s(n+2) + 3s(n-1) - 4s(n-3)$$

Ans

$$x(n) = \{1, 0, 1, 0, 3, 1, 0, -4\}$$

③ 36



$h(n) = a^n u(n)$, $-1 < a < 1$, determine the response $y(n)$ of the SLM to the excitation

$$x(n) = u(n+5) - u(n-10)$$

Sol

$$S(n) = u(n) * h(n)$$

$$S(n) = \sum_{k=0}^{\infty} u(k) h(n-k)$$

$$= \sum_{k=0}^{\infty} h(n-k)$$

$$k=0$$

$$= \sum_{k=0}^{\infty} a^{n-k} = \underbrace{a^{n+k}}_{a-1} ; n \geq 0$$

for $x(n) = u(n+5) - u(n-10)$ then

$$S(n+5) - S(n-10) = \underbrace{a^{n+6}-1}_{a-1} u(n+5) - \underbrace{a^{n-9}-1}_{a-1} u(n-10)$$

from given figure

$$y(n) = x(n) * h(n) - x(n-2) * h(n-2)$$

$$y(n) = \underbrace{a^{n+6}-1}_{a-1} u(n+5) - \underbrace{a^{n-9}-1}_{a-1} u(n-10) - \underbrace{a^{n+4}-1}_{a-1} u(n+3) + \underbrace{a^{n+1}-1}_{a-1} u(n+2)$$

(37)

compute & sketch the step response of the sm

$$y(n) = \frac{1}{M} \sum_{k=0}^{M-1} x(n-k)$$

Sol

$$h(n) = \left[\frac{u(n) - u(n-m)}{m} \right]$$

$$S(n) = \sum_{k=-\infty}^{\infty} u(k) h(n-k)$$

$$= \sum_{k=0}^n h(n-k) = \begin{cases} \frac{n+1}{m}, & n < m \\ 1, & n \geq m \end{cases}$$

(38)

$$h(n) = \begin{cases} a^n, & n \geq 0, n \text{ even} \\ 0, & \text{otherwise} \end{cases}$$

Sol

$$\sum_{k=-\infty}^{\infty} |h(n)| = \sum_{n=0}^{\infty} |a|^n$$

$$= \sum_{n=0}^{\infty} |a|^{2n} = \frac{1}{1-|a|^2}$$

Stable if $|a| < 1$

(39)

h(n) = $a^n u(n)$ to O/P signal $x(n) = u(n) - u(n-10)$ Sol

$$h(n) = a^n u(n)$$

$$y_1(n) = \sum_{k=0}^{\infty} u(k) h(n-k)$$

$$= \sum_{k=0}^n a^{n-k} = a^n \sum_{k=0}^n a^{-k}$$

$$= \frac{1-a^{n+1}}{1-a} u(n)$$

$$y(n) = y_1(n) - y_1(n-10)$$

$$= \frac{1}{1-a} \cdot [(1-a^{n+1}) u(n) - ((-a^{n-1}) u(n-10))]$$

$$(40) \quad x(n) = \begin{cases} 1, & 0 \leq n \leq 10 \\ 0, & \text{otherwise} \end{cases} \quad h(n) = \left(\frac{1}{2}\right)^n u(n)$$

from '86' problem with $a=1/2$

$$y(n) = 2 \left[1 - \left(\frac{1}{2}\right)^{n+1} \right] u(n) - 2 \left[1 \left(\frac{1}{2}\right)^{n-9} \right] u(n-10)$$

$$(41) \quad h(n) = \left(\frac{1}{2}\right)^n u(n)$$

Determine response for given 2/8 signals.

$$(a) \quad x(n) = 2^n u(n)$$

$$\underline{\text{Sol}} \quad x(n) = 2^n u(n)$$

$$y(n) = \sum_{k=-\infty}^{\infty} h(k) x(n-k)$$

$$= \sum_{k=0}^n \left(\frac{1}{2}\right)^k 2^{n-k} = 2^n \sum_{k=0}^n \left(\frac{1}{4}\right)^k$$

$$= 2^n \left[1 - \left(\frac{1}{4}\right)^{n+1} \right] \left(\frac{4}{3}\right) = \frac{2}{3} \left[2^{n+1} - \left(\frac{1}{2}\right)^{n+1} \right]$$

$$(b) \quad x(n) = u(-n)$$

$$y(n) = \sum_{k=-\infty}^{\infty} h(k) x(n-k)$$

$$= \sum_{k=0}^{\infty} h(k) = \sum_{k=0}^{\infty} \left(\frac{1}{2}\right)^k = 2, \quad n < 0$$

$$y(n) = \sum_{k=n}^{\infty} h(k) \Rightarrow \sum_{k=n}^{\infty} \left(\frac{1}{2}\right)^k$$

$$= \sum_{k=0}^{\infty} \left(\frac{1}{2}\right)^k - \sum_{k=0}^{n-1} \left(\frac{1}{2}\right)^k$$

$$= 2 - \left(\frac{1 - \left(\frac{1}{2}\right)^n}{\frac{1}{2}} \right)$$

$$= 2 \left(\frac{1}{2}\right)^n, \quad n \geq 0$$

(42) Three S/I/M with impulse response $h_1(n) = \delta(n) - \delta(n-1)$, $h_2(n) = h(n)$ and $h_3(n) = u(n)$ connected in cascade.

a) What is the $h_c(n)$, impulse response of overall S/I/M?

$$h_c(n) = h_1(n) * h_2(n) * h_3(n)$$

$$= [\delta(n) - \delta(n-1)] * u(n) * h(n)$$

$$= [u(n) - u(n-1)] * h(n)$$

$$= \delta(n) * h(n) = h(n)$$

b) Does the order of interconnection affect the overall S/I/M?

Sol No.

(43) a) Prove & explain graphically the difference b/w the relations

$$x(n) \delta(n-n_0) = x(n_0) \delta(n-n_0) \text{ and}$$

$$x(n) * \delta(n-n_0) = x(n-n_0)$$

Sol $\rightarrow x(n) \delta(n-n_0) = x(n_0)$. Thus only the value of $x(n)$ at $n=n_0$ is of interest.

$\rightarrow x(n) * \delta(n-n_0) = x(n-n_0)$. Thus, we obtained shifted version of $x(n)$ sequence.

$$(b) y(n) = \sum_{k=-\infty}^{\infty} h(k) x(n-k)$$

$$= h(n) * x(n)$$

$$\text{linearity: } x_1(n) \rightarrow y_1(n) = h(n) * x_1(n)$$

$$x_2(n) \rightarrow y_2(n) = h(n) * x_2(n)$$

$$= \alpha h(n) * x_1(n) + \beta h(n) * x_2(n)$$

$$= \alpha y_1(n) + \beta y_2(n)$$

Time invariance: $x(n) \rightarrow y_1(n) = h(n) \neq x(n)$

$x(n-n_0) \rightarrow y_2(n) = h(n) \neq x(n-n_0)$

$$= \sum_k h(k) x(n-n_0-k)$$

$$= y(n-n_0)$$

⑥ $h(n) = S(n-n_0)$

⑦ 45 Compute zero state response for SHM

$$y(n) + \frac{1}{2} y(n-1) = x(n) + 2x(n-2)$$

To the input $x(n) = \{1, 2, 3, 4, 2\}$

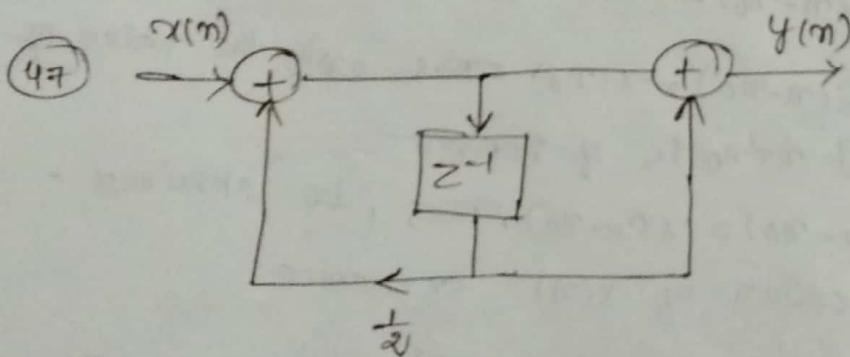
Sol $y(n) = -\frac{1}{2} y(n-1) + x(n) + 2x(n-2)$

$$\text{at } y(-2) = -\frac{1}{2} y(-3) + x(-2) + 2x(-4) = 1$$

$$y(-1) = -\frac{1}{2} y(-2) + x(-1) + 2x(-3) = \frac{9}{2}$$

$$y(0) = -\frac{1}{2} y(-1) + 2x(-2) + x(0) = \frac{17}{4}$$

$$y(1) = -\frac{1}{2} y(0) + x(1) + 2x(-1) = \frac{47}{8}$$



a) Compute the first samples of its impulse response.

Sol $x(n) = \{1, 0, 0, \dots\}$

$$y(n) = -\frac{1}{2} y(n-1) + x(n) + x(n-1)$$

$$y(0) = x(0) = 1$$

$$y(1) = -\frac{1}{2} y(0) + x(1) + x(0) = \frac{3}{2}$$

$$y(2) = -\frac{1}{2} y(1) + x(2) + x(1) = \frac{8}{4}$$
 Thus we obtain

$$y(n) = \left\{ 1, \frac{3}{2}, \frac{3}{4}, \frac{3}{8}, \frac{3}{16}, \frac{3}{32}, \dots \right\}$$

(b) find the input-output relation.

Sol $y(n) = \frac{1}{2} y(n-1) + x(n) + x(n-1)$

(c) for IIP $x(n) = \left\{ 1, 1, 1, \dots \right\}$ compute first 10 samples.

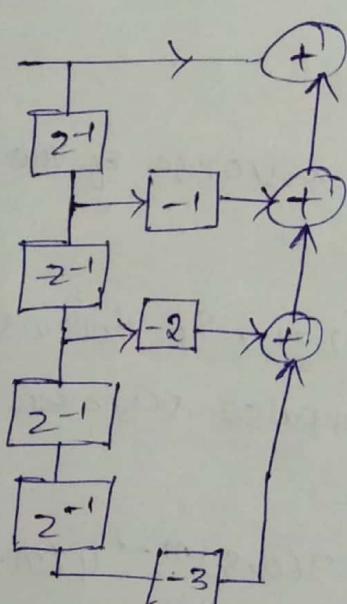
as in part (a) we obtain

$$y(n) = \left\{ 1, \frac{5}{2}, \frac{13}{4}, \frac{29}{8}, \frac{61}{16}, \dots \right\}$$

(d) Compute 10 samples for IIP given (c), by using causal convolution.

$$y(n) = u(n) * h(n)$$

$$= \sum_k u(k) h(n-k)$$



$$= \sum_{k=0}^n h(n-k)$$

$$y(0) = h(0) = 1$$

$$y(1) = h(0) + h(1) = \frac{5}{2}$$

$$y(2) = h(0) + h(1) + h(2) = \frac{13}{4} \text{ etc.}$$

(e) from part (a), $h(n)=0$ for $n < 0 \Rightarrow$ The S/I/M is causal.

$$\sum_{n=0}^{\infty} |h(n)| = 1 + \frac{3}{2} \left(1 + \frac{1}{2} + \frac{1}{4} + \dots \right) = 4$$

\Rightarrow S/I/M is stable.

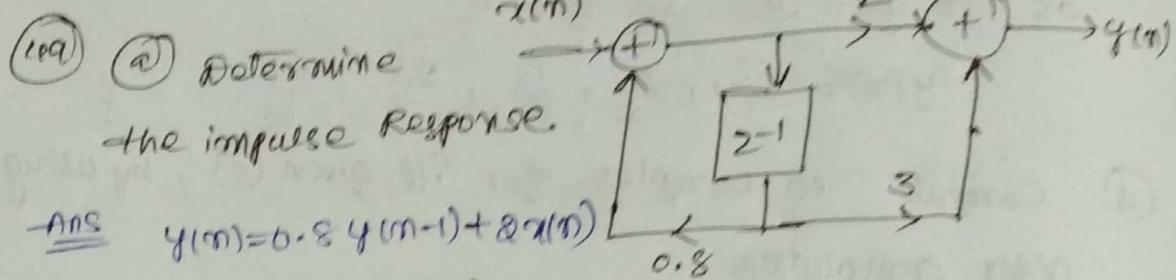
(48) $y(n) = ay(n-1) + bx(n)$

(a) determine b in terms of a so that

$$\sum_{n=-\infty}^{\infty} h(n) = 1$$

$$\begin{aligned}
 \textcircled{80} \quad y(n) &= ay(n-1) + bx(n) \\
 h(n) &= b a^n u(n) \\
 \sum_{n=0}^{\infty} h(n) &= \frac{b}{1-a} = 1 \\
 b &= 1-a \\
 \textcircled{b} \quad S(n) &= \sum_{k=0}^{n-1} h(n-k) \\
 &= b \left[\frac{1-a^{n+1}}{1-a} \right] a(n) \\
 S(\infty) &= \frac{b}{1-a} = 1 \\
 b &= 1-a
 \end{aligned}$$

(c) $b=1-a$ in both cases.



$$\begin{aligned}
 \textcircled{Ans} \quad y(n) &= 0.8y(n-1) + 2x(n) \\
 &\quad + 3x(n-1) \\
 y(n) - 0.8y(n-1) &= 2x(n) + 3x(n-1)
 \end{aligned}$$

The characteristic equation is

$$\begin{aligned}
 2 - 0.8 &= 0 \\
 r &= 0.8
 \end{aligned}$$

$y_h(n) = c(0.8)^n$
Let's first consider the response of the shu

$y(n) - 0.8y(n-1) = x(n)$
to $x(n) = S(n)$, since $y(0)=1$, it follows that
 $c=1$. Then, the impulse response of
the original shu is

$$\begin{aligned}
 u(n) &= 2(0.8)^n u(n) + 3(0.8)^{n-1} u(n-1) \\
 &= 2S(n) + 4.6(0.8)^{n-1} u(n-1)
 \end{aligned}$$

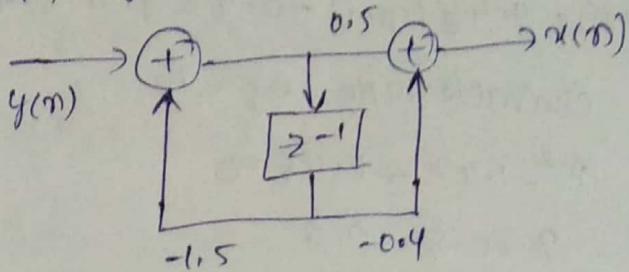
(e) The inverse shu is characterized by the difference eq

$$x(n) = -1.5u(n-1) + \frac{1}{2}y(n) - 0.4y(n-1)$$

(f) (a) $y(n) = 0.9y(n-1) + x(n) + 2x(n-1) + 3x(n-2)$
 $y(n) - 0.9y(n-1) = x(n) + 2x(n-1) + 3x(n-2)$
after $x(n) = S(n)$, we have

$$y(0) = 1, \quad y(1) = 2.9, \quad y(2) = 5.61, \quad y(3) = 5.049$$

$$y(4) = 4.544, \quad y(5) = 4.090$$



(b) $s(0) = y(0) = 1$

$$s(1) = y(0) + y(1) = 3.9$$

$$s(2) = y(0) + y(1) + y(2) = 9.51$$

$$s(3) = y(0) + y(1) + y(2) + y(3) = 14.56$$

$$s(4) = \sum_0^4 y(n) = 19.10$$

$$s(5) = \sum_0^5 y(n) = 23.19.$$

(c) $h(n) = (0.9)^n u(n) + 2(0.9)^{n-1} u(n-1) + 3(0.9)^{n-2} u(n-2)$

$$= s(n) + 2.9 \cdot s(n-1) + 5.61(0.9)^{n-2} u(n-2).$$

(51)

(2) $y(n) = \frac{1}{3}x(n) + \frac{1}{3}x(n-3) + y(n-1)$

for $x(n) = s(n)$, we have

$$h(n) = \left\{ \frac{1}{3}, \frac{1}{3}, \frac{1}{3}, \frac{2}{3}, \frac{2}{3}, \frac{2}{3}, \dots \right\}$$

(b)

$$y(n) = \frac{1}{2}y(n-1) + \frac{1}{8}y(n-2) + \frac{1}{2}x(n-2)$$

with $x(n) = s(n)$ and

$$y(-1) = y(-2) = 0, \text{ we obtain}$$

$$h(n) = \left\{ 0, 0, \frac{1}{2}, \frac{1}{4}, \frac{3}{16}, \frac{1}{8}, \frac{11}{128}, \frac{15}{256}, \frac{41}{1024}, \dots \right\}$$

(c)

$$y(n) = 1.4y(n-1) - 0.48y(n-2) + x(n)$$

with $x(n) = s(n)$ and

$$y(-1) = y(-2) = 0, \text{ we obtain}$$

$$h(n) = \left\{ 1, 1.4, 1.48, 1.4, 1.2496, 1.0774, 0.9086, \dots \right\}$$

(d) All three systems are IIR.

(e) $y(n) = 1.4y(n-1) - 0.48y(n-2) + x(n)$

The characteristic eq is

$$q^2 - 1.4q + 0.48 = 0$$

$$q = 0.8, 0.6$$

$$y_h(n) = C_1(0.8)^n + C_2(0.6)^n \text{ for } x(n) = s(n).$$

$$C_1 + C_2 = 1 \text{ and}$$

$$0.8C_1 + 0.6C_2 = 1.4$$

$$C_1 = 4, C_2 = -3$$

$$h(n) = [4(0.8)^n - 3(0.6)^n] u(n).$$

(52) (a) $h_1(n) = c_0 s(n) + c_1 s(n-1) + c_2 s(n-2)$

$$h_2(n) = b_2 s(n) + b_1 s(n-1) + b_0 s(n-2).$$

$$h_3(n) = a_0 s(n) + (a_1 + a_0 + a_2) s(n-1) + a_1 a_2 s(n-2).$$

(b) The only question is whether

$$h_3(n) = ? h_2(n) = h_1(n)$$

$$\text{let } a_0 = c_0$$

$$a_1 + a_2 c_0 = c_1, \Rightarrow a_1 + a_2 c_0 - c_1 = 0$$

$$a_2 a_1 = c_2, \Rightarrow \frac{c_2}{a_2} = a_1$$

$$\Rightarrow \frac{c_2}{a_2} + a_2 c_0 - c_1 = 0$$

$$\Rightarrow c_0 a_2^2 - a_1 a_2 + c_2 = 0$$

for $c_0 \neq 0$, the quadratic has a real solution

if & only if $a_1^2 - 4c_0 c_2 \geq 0$.

(53) (a) $y(n) = \frac{1}{2} y(n) + x(n) + x(n-1)$

$$\text{for } y(n) - \frac{1}{2} y(n-1) = x(n) + x(n-1) \quad x(n) = s(n)$$

$$h(n) = (\frac{1}{2})^n u(n) + (\frac{1}{2})^{n-1} u(n-1)$$

(b) $h_1(n) \neq [s(n) + s(n-1)] = (\frac{1}{2})^n u(n) + (\frac{1}{2})^{n-1} u(n-1)$

$$(54) \textcircled{a} \quad x_1(n) = \{1, 2, 4\} \quad h(n) = \{1, 1, 1, 1, 1\}$$

Convolution: $y_1(n) = \{1, 3, 7, 7, 7, 6, 4\}$

Correlation: $\delta_1(n) = \{1, 3, 7, 7, 7, 6, 4\}$

$$\textcircled{b} \quad x_2(n) = \{0, 1, -2, 3, -4\} \quad h_2(n) = \left\{\frac{1}{2}, 1, 2, 1, \frac{1}{2}\right\}$$

Convolution: $y_2(n) = \left\{\frac{1}{2}, 0, \frac{3}{2}, -2, \frac{1}{2}, -6, -\frac{5}{2}, -2\right\}$

Correlation: $\delta_2(n) = \left\{\frac{1}{2}, 0, \frac{3}{2}, -2, \frac{1}{2}, -6, -\frac{5}{2}, -2\right\}$

Note $y_2(n) = \delta_2(n)$, $\therefore h_2(-n) = h_2(n) \text{ (c)}$

$$\textcircled{c} \quad x_3(n) = \{1, 2, 3, 4\} \quad h_3(n) = \{4, 3, 2, 1\}$$

Convolution: $y_3(n) = \{4, 11, 20, 30, 20, 11, 4\}$

Correlation: $\delta_3(n) = \{1, 4, 10, 20, 25, 24, 16\}$

$$\textcircled{d} \quad x_4(n) = \{1, 2, 3, 4\} \quad h_4(n) = \{1, 2, 3, 4\}$$

Convolution: $y_4(n) = \{1, 4, 10, 20, 25, 24, 16\}$

Correlation: $\delta_4(n) = \{4, 11, 20, 30, 20, 11, 4\}$

Note that $h_3(-n) = h_4(n+3)$

$$\delta_3(n) = y_4(n+3)$$

$$h_4(-n) = h_3(n+3)$$

$$\delta_4(n) = y_3(n+3)$$

$$(55) \quad \textcircled{-Ans} \quad \text{length of } h(n) = 2$$

$$h(n) = \{h_0, h_1\}$$

$$h_0 = 1$$

$$3h_0 + h_1 = 4$$

$$\boxed{h_0=1, h_1=1}$$

$$(2.5.6) \quad y(n) = -\sum_{k=1}^n a_k y(n-k) + \sum_{k=0}^m b_k x(n-k)$$

$$(2.5.7) \quad w(n) = -\sum_{k=1}^n a_k w(n-k) + x(n)$$

$$(2.5.10) \quad y(n) = \sum_{k=0}^m b_k w(n-k)$$

From 2.5.9 we obtain $x(n) = y(n) + w(n)$

$$x(n) = w(n) + \sum_{k=1}^n a_k w(n-k) \quad \text{by substituting}$$

$$(2.5.10) \quad \text{for } y(n) \text{ and (6) into (2.5.6)}$$

We obtain L.H.S = R.H.S.

$$(57) \quad \underline{\text{S.D.}} \quad y(-1) = y(-2) = 0$$

$$y(n) - 4y(n-1) + 4y(n-2) = x(n) - x(n-1)$$

The characteristic equation is

$$\lambda^2 - 4\lambda + 4 = 0$$

$\lambda = 2, 2$. Hence

$$y_h(n) = C_1 2^n + C_2 n 2^n$$

The partial solution is

$$y_p(n) = k(-1)^n u(n).$$

Substituting this solution is onto the difference eq, we obtain

$$\begin{aligned} k(-1)^n u(n) - 4k(-1)^{n-1} u(n-1) + 4k(-1)^{n-2} u(n-2) \\ = (-1)^n u(n) - (-1)^{n-1} u(n). \end{aligned}$$

For $n=2$, $k(1+4+4)=2 \Rightarrow k=\frac{2}{9}$. The total solution is $y(n) = [C_1 2^n + C_2 n 2^n + \frac{2}{9} (-1)^n] u(n)$

from the initial conditions, we obtain

$$y(0)=1, y(1)=2.$$

$$c_1 + \frac{2}{9} = 1$$

$$\Rightarrow c_1 = \frac{7}{9}$$

$$2c_1 + 2c_2 - \frac{2}{9} = 2$$

$$c_2 = 1/3$$

(58)

Sol

From problem (57)

$$h(n) = [c_1 2^n + c_2 n 2^n] u(n)$$

with $y(0) = 1, y(1) = 3$ we have

$$c_1 = 1 \quad ; \quad 2c_1 + 3c_2 = 1$$

$$c_2 = 1/2$$

$$\text{Thus } h(n) = [2^n + \frac{1}{2} n 2^n] u(n)$$

(59)

Sol

$$u(n-k) = \begin{cases} 1, & n \geq k \\ 0, & \text{otherwise} \end{cases}$$

$$x(n) = x(n) * s(n)$$

$$= x(n) * [u(n) - u(n-1)]$$

$$= [x(n) - x(n-1)] * u(n)$$

$$x(n) = \sum_{k=-\infty}^{\infty} [x(k) - x(k-1)] u(n-k)$$

(60)

$$\text{Sol: } u(n-k) = \begin{cases} 1, & n \neq k \\ 0, & \text{otherwise} \end{cases}$$

$$x(n) = x(n) * g(n)$$

$$= x(n) * [u(n) - u(n-1)]$$

$$= [v(n) - v(n-1)] * u(n)$$

Let $h(n)$ be the impulse response of the

$$\text{S/I.M.} \quad S(k) = \sum_{m=-\infty}^{+\infty} h(m)$$

$$h(k) = S(k) - S(k-1)$$

$$y(n) = \sum_{k=-\infty}^{\infty} h(k) x(n-k)$$

$$= \sum_{k=-\infty}^{\infty} [S(k) - S(k-1)] x(n-k)$$

(61)

$$\underline{\text{Sol}} \quad x(n) = \begin{cases} 1, & n_0 - N \leq n \leq n_0 + N \\ 0, & \text{otherwise} \end{cases}$$

$$y(n) = \begin{cases} 1, & -N \leq n \leq N \\ 0, & \text{otherwise} \end{cases}$$

$$Y_{xx}(l) = \sum_{n=-\infty}^{\infty} x(n) x(n-l)$$

The range of non-zero values of $Y_{xx}(l)$ is

determined by $n_0 - N \leq n \leq n_0 + N$

$$n_0 - N \leq n - l \leq n_0 + N$$

which implies

$$-2N \leq l \leq 2N$$

for a given shift l , the no. of terms in the summation for which both $x(n)$ and $x(n-l)$ are non-zero is $2N+1-|l|$ and the value each term is 1. Hence,

$$Y_{xx}(l) = \begin{cases} 2N+1-|l|, & -2N \leq l \leq 2N \\ 0, & \text{otherwise} \end{cases}$$

for $Y_{xy}(l)$ we have

$$Y_{xy}(l) = \begin{cases} 2N+1-(l-n_0), & n_0 - 2N \leq l \leq n_0 + 2N \\ 0, & \text{otherwise} \end{cases}$$

(62)

$$\textcircled{a} \quad Y_{xx}(l) = \sum_{n=-\infty}^{\infty} x(n) x(n-l)$$

$$Y_{xx}(-3) = x(0) x(3) = 1$$

$$Y_{xx}(-2) = x(0) x(2) + x(1) x(3) = 3$$

$$Y_{xx}(-1) = x(0) x(1) + x(1) x(2) + x(2) x(3) = 5$$

$$Y_{xx}(0) = \sum_{n=0}^3 x^2(n) = 7$$

$$\text{also } Y_{xx}(-l) = Y_{xx}(l)$$

$$\therefore \gamma_{xx}(l) = \{1, 3, 5, 7, 5, 3, 1\}$$

$$\textcircled{5} \quad \gamma_{yy}(l) = \sum_{n=-\infty}^{\infty} y(n) y(n-l)$$

$$\text{we obtain } \gamma_{yy}(l) = \{1, 3, 5, 7, 5, 3, 1\}$$

we obtain $y(n) = x(-n+3)$. which is equivalent to reverse the sequence $x(n)$. This has NOT changed the value.

\textcircled{62}

$$\text{Sol} \quad \gamma_{xx}(l) = \sum_{n=-\infty}^{\infty} x(n) x(n-l)$$

$$= \begin{cases} 2N+1-|l|, & -2N \leq l \leq 2N \\ 0, & \text{otherwise} \end{cases}$$

$$\gamma_{xx}(0) = 2N+1$$

\because The normalized auto correlation is

$$\rho_{xx}(l) = \begin{cases} \frac{1}{2N+1} (2N+1-|l|), & -2N \leq l \leq 2N \\ 0, & \text{otherwise} \end{cases}$$

\textcircled{64}

$$\textcircled{a} \quad \gamma_{xx}(l) = \sum_{n=-\infty}^{\infty} x(n) x(n-l)$$

$$= \sum_{n=-\infty}^{\infty} [\delta(n) + \gamma_1 \delta(n-k_1) + \gamma_2 \delta(n-k_2)] * [\delta(n-1) + \gamma_1 \delta(n-1-k_1) + \gamma_2 \delta(n-1-k_2)]$$

$$= (1 + \gamma_1^2 + \gamma_2^2) \gamma_{ss}(l) + \gamma_1 [\gamma_{ss}(l+k_1) + \gamma_{ss}(l-k_1)] + \gamma_2 [\gamma_{ss}(l+k_2) + \gamma_{ss}(l-k_2)] + \gamma_1 \gamma_2 [\gamma_{ss}(l+k_1+k_2) + \gamma_{ss}(l+k_2-k_1)]$$

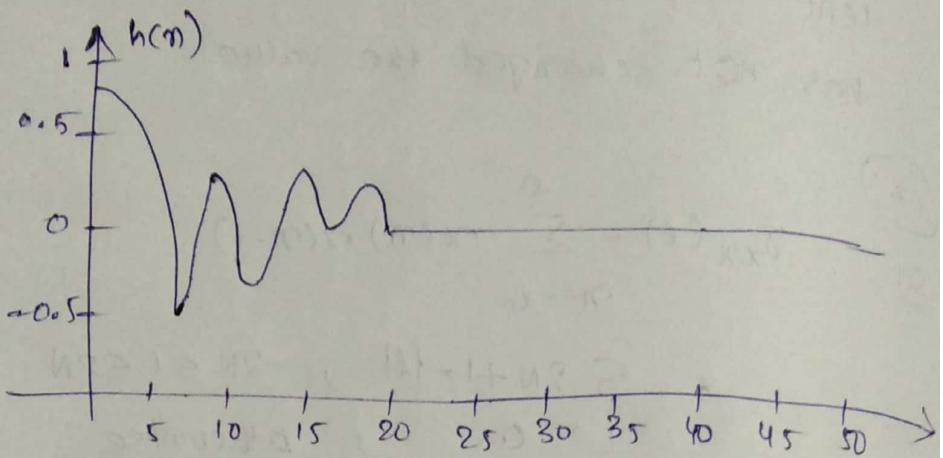
\textcircled{b} \quad $\gamma_{ss}(l)$ has peaks at $l=0, \pm k_1, \pm k_2$ and $\pm (k_1 + k_2)$ suppose that $k_1 < k_2$. Then, we

can determine γ_1 and k_1 . The problem is to determine γ_2 and k_2 from the other peaks.

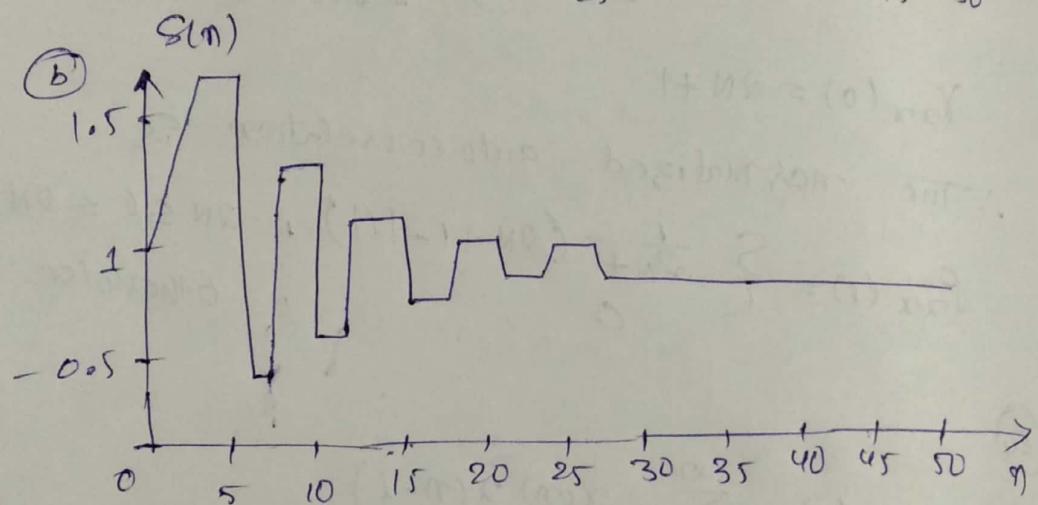
⑥ If $\gamma_2=0$, the peaks occur at $t=0$, and $t=\pm k_1$. That is easy to obtain γ_1 and k_1 .

(6B)

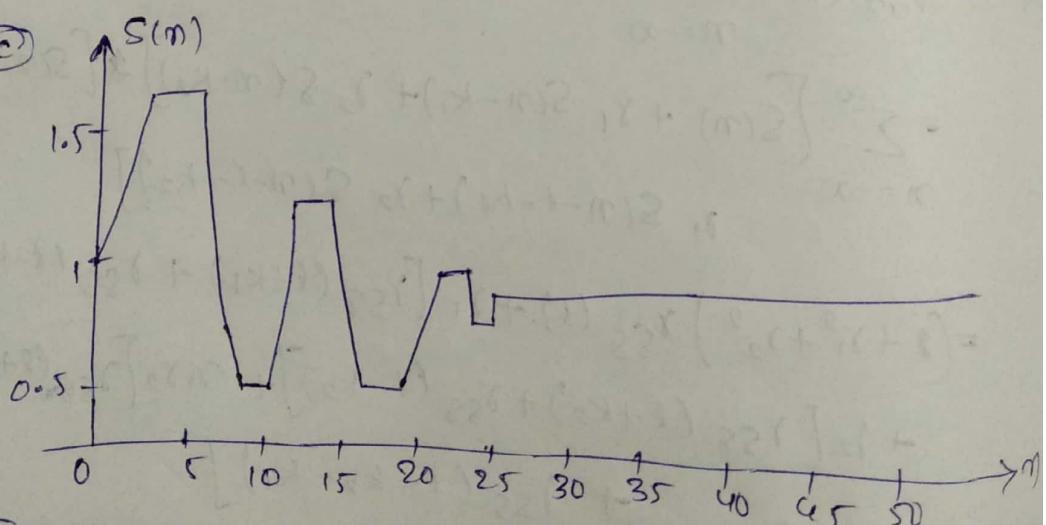
a)



b)



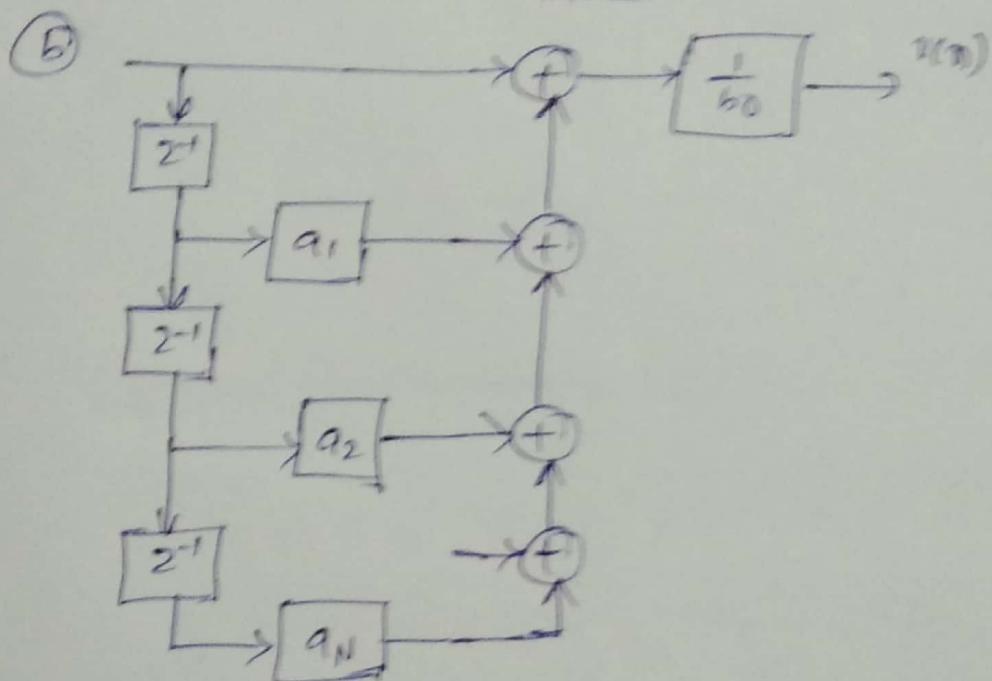
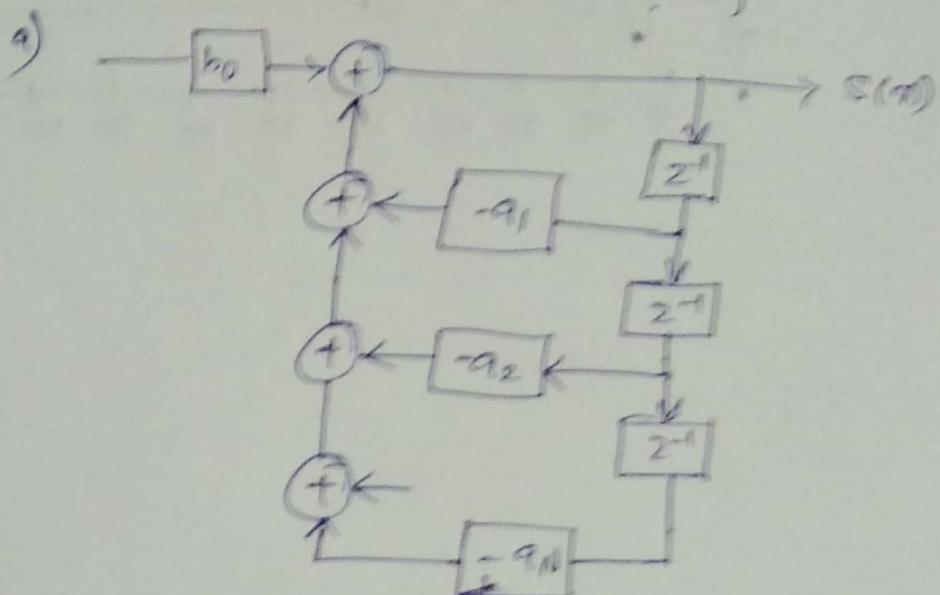
c)



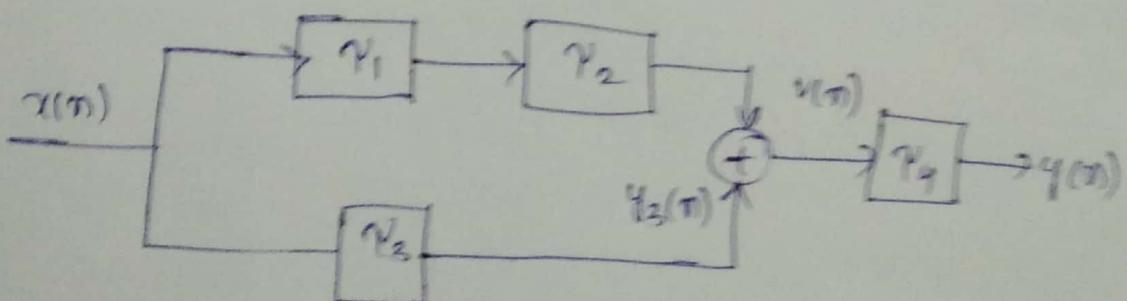
d) c, b are similar except c have steady state after $m=20$, where b have nearly at $m=20$.

(44) ④ $S(n) = -a_1 S(n-1) - a_2 S(n-2) - \dots - a_N S(n-N) + b_0 v(n)$

⑤ $v(n) = \frac{1}{b_0} [S(n) + a_1 S(n-1) + a_2 S(n-2) + \dots + a_N S(n-N)]$



⑥ Plot $h(n)$ for $0 \leq n \leq 99$.



SOL:
010

