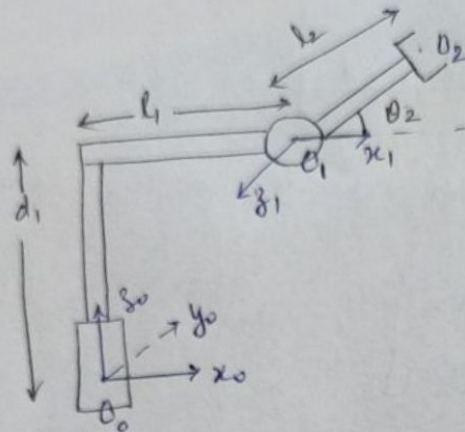


Problem 1



$${}^0D_2 = \begin{bmatrix} x_{ef} \\ y_{ef} \end{bmatrix} = \begin{bmatrix} l_1 + l_2 \cos \theta_2 \\ d_1 + l_2 \sin \theta_2 \end{bmatrix}$$

Alternatively, you can also take derivative of given D_2^0 matrix.

$$J = \begin{bmatrix} \frac{dx}{da_1} & \frac{dx}{da_2} \\ \frac{dy}{da_1} & \frac{dy}{da_2} \end{bmatrix} \quad \begin{matrix} a_1: d_1 \\ a_2: \theta_2 \end{matrix}$$

$$J = \begin{bmatrix} 0 & -l_2 \sin \theta_2 \\ 1 & l_2 \cos \theta_2 \end{bmatrix}$$

a) Velocity kinematics

	Linear component	Angular component
Revolute	$Jv_i = Z_{i-1}^0 \times (\dot{\theta}_2 - \dot{\theta}_1)$	$Jw_i = Z_{i-1}^0$
Prismatic	$Jv_i = Z_{i-1}^0$	$Jw_i = 0$

$$Jv_1 = Z_0, \quad Jw_1 = 0$$

$$Jv_2 = Z_1 \times (\dot{\theta}_2 - \dot{\theta}_1), \quad Jw_2 = Z_1$$

Z_1 and $\dot{\theta}_1$ can be obtained from observation simply.

$$Z_1 = [0 \ 1 \ 0]^T \quad \text{and} \quad \dot{\theta}_1 = [l_1 \ 0 \ d_1]^T$$

$$\text{Jacobian} = \begin{bmatrix} Jv_1 & Jv_2 \\ Jw_1 & Jw_2 \end{bmatrix}$$

$$= \begin{bmatrix} 0 & \begin{bmatrix} 0 \\ -1 \\ 0 \end{bmatrix} \times \begin{bmatrix} l_1 + l_2 \cos \theta_2 - l_1 \\ 0 \\ d_1 + l_2 \sin \theta_2 - d_1 \end{bmatrix} \\ 1 & \begin{bmatrix} 0 \\ 0 \\ 0 \end{bmatrix} \\ 0 & \begin{bmatrix} 0 \\ 1 \\ 0 \end{bmatrix} \\ 0 & \begin{bmatrix} 0 \\ 0 \\ 0 \end{bmatrix} \end{bmatrix} = \begin{bmatrix} 0 & -l_2 \sin \theta_2 \\ 0 & 0 \\ 1 & l_2 \cos \theta_2 \\ 0 & 0 \\ 0 & -1 \\ 0 & 0 \end{bmatrix}$$

2x2 jacobian that relates joint velocities to the linear velocity of EF:

$$\begin{bmatrix} 0 & -l_2 \sin \theta_2 \\ 1 & l_2 \cos \theta_2 \end{bmatrix}$$

(b) Singular configurations?

These exist when the inverse of the jacobian matrix doesn't exist.

When it is not possible to deduce the joint velocities corresponding to given EF velocities.

$$\text{determinant}(J) = 0$$

$$0(l_2 \cos \theta_2) - (-l_2 \sin \theta_2(1)) = 0$$

$$\boxed{l_2 \sin \theta_2 = 0}$$

$$\begin{bmatrix} v_x \\ v_z \end{bmatrix} = \begin{bmatrix} 0 & 0 \\ 1 & l_2 \cos \theta_2 \end{bmatrix} \begin{bmatrix} \dot{q}_1 \\ \dot{q}_2 \end{bmatrix}$$

We can see that if $\theta_2 = 0$ or $\theta_2 = \pi$, no matter what the joint velocities \dot{q}_1 & \dot{q}_2 are, we cannot obtain velocities v_x ~~and~~ v_z for the end effector.

$\therefore \theta_2 = 0, \pi$ restrict motion in 'x'

Problem 2

$$(a) \quad J(\theta) = \begin{bmatrix} 0 & 0 & 0 & l_3 \\ 0 & -l_1 & -l_1 - l_2 & -l_1 - l_2 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \\ 1 & 1 & 1 & 1 \end{bmatrix}$$

$$f = [10 \ 10 \ 0]^T, \quad n = [0 \ 0 \ 10]^T$$

we know that,

$$\tau = J^T F, \quad \text{where } F = [F_x \ F_y \ F_z \ n_x \ n_y \ n_z]^T$$

$$\therefore \begin{bmatrix} \tau_1 \\ \tau_2 \\ \tau_3 \\ \tau_4 \end{bmatrix} = \begin{bmatrix} 0 & 0 & 0 & 0 & 0 & 1 \\ 0 & -l_1 & 0 & 0 & 0 & 1 \\ 0 & -(l_1 + l_2) & 0 & 0 & 0 & 1 \\ l_3 & -(l_1 + l_2) & 0 & 0 & 0 & 1 \end{bmatrix} \begin{bmatrix} 10 \\ 10 \\ 0 \\ 0 \\ 0 \\ 10 \end{bmatrix}$$

$$\tau = \begin{bmatrix} 10 \\ -10l_1 + 10 \\ -10(l_1 + l_2) + 10 \\ -10(l_1 + l_2 - l_3) + 10 \end{bmatrix} \quad \text{OR} \quad \begin{bmatrix} 10 \\ 10(1 - l_1) \\ 10(1 - l_1 - l_2) \\ 10(1 + l_3 - l_1 - l_2) \end{bmatrix}$$

$$b) \quad J(\theta) = \begin{bmatrix} l_3 s_4 + l_2 s_{34} + l_1 s_{234} & l_3 s_4 + l_2 s_{34} & l_3 s_4 & 0 \\ l_4 + l_3 c_4 + l_2 c_{34} + l_1 c_{234} & l_4 + l_3 c_4 + l_2 c_{34} & l_4 + l_3 c_4 & l_4 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \\ 1 & 1 & 1 & 1 \end{bmatrix}$$

This jacobian matrix is equivalent to the following matrix after applying the column operations:

$$\begin{aligned} c_1 &\rightarrow c_1 - c_2 \\ c_2 &\rightarrow c_2 - c_3 \\ c_3 &\rightarrow c_3 - c_4 \end{aligned}$$

$$\begin{bmatrix} l_1 s_{234} & l_2 s_{34} & l_3 s_{4} & 0 \\ l_1 c_{234} & l_2 c_{34} & l_3 c_{4} & l_1 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 1 \end{bmatrix}$$

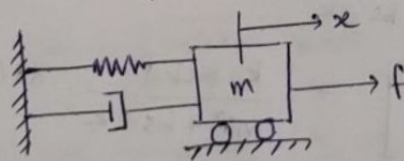
we can see that

- (i) $c_2 = \frac{l_2}{l_1} c_1$ when $\theta_2 = 0$
- (ii) $c_3 = \frac{l_3}{l_1} c_1$ when $\theta_2 = \theta_3 = 0$
- (iii) $c_3 = \frac{l_3}{l_2} c_2$ when $\theta_3 = 0$
- (iv) $c_3 = \frac{l_3}{2l_1} c_1 + \frac{l_3}{2l_2} c_2$ when $\theta_2 = \theta_3 = 0$.

Hence the singularities of the 4R robot for the given configuration occur when $\theta_2 = 0$, $\theta_3 = 0$ and $\theta_2 = \theta_3 = 0$ together.

Problem 3

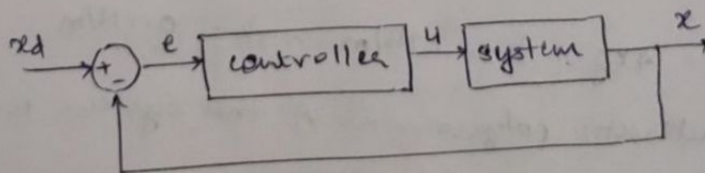
1DOF mass-spring damper system.



x : horizontal pos.
 f : control force.

$$m\ddot{x} + b\dot{x} + kx = f$$

controller: $f = k_p(e) + k_d(\dot{e})$



(a) system model in the Laplace domain

$$m\ddot{x} + b\dot{x} + kx = f$$

$$ms^2 X(s) + bsX(s) + kX(s) = F(s) \quad \text{--- (1)}$$

controller: $f = k_p e + k_d \dot{e}$

$$F(s) = k_p E(s) + k_d s E(s) \quad \text{--- (2)}$$

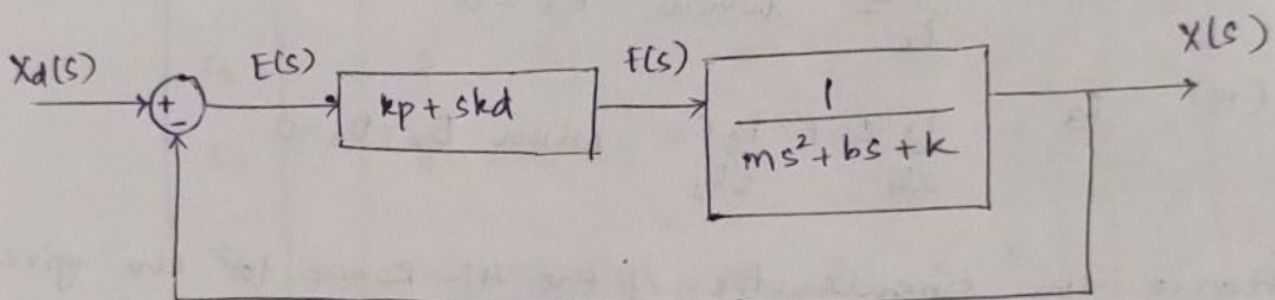
(b) Transfer functions for $\frac{F(s)}{E(s)}$, $\frac{X(s)}{F(s)}$, $\frac{X(s)}{E(s)}$

from (2) we have $\frac{F(s)}{E(s)} = k_p + s k_d$ ——— (3)

from (1) we have $\frac{X(s)}{F(s)} = \frac{1}{ms^2 + bs + k}$ ——— (4)

from (3) & (4), we have $\frac{X(s)}{E(s)} = \frac{F(s)}{E(s)} \times \frac{X(s)}{F(s)}$
 $= \frac{k_p + s k_d}{ms^2 + bs + k}$

(c) Closed-loop transfer function
 (i.e) $\frac{X(s)}{X_d(s)}$



$$F(s) = (k_p + s k_d) E(s) \quad \& \quad E(s) = X_d(s) - X(s)$$

$$X(s) = F(s) \cdot \frac{1}{ms^2 + bs + k}$$

$$X(s) = \frac{(k_p + s k_d)(X_d(s) - X(s))}{ms^2 + bs + k}$$

$$X(s) \left[1 + \frac{k_p + s k_d}{ms^2 + bs + k} \right] = \frac{k_p + s k_d}{ms^2 + bs + k} \cdot X_d(s)$$

$$\therefore \frac{X(s)}{X_d(s)} = \frac{k_p + s k_d}{ms^2 + (b + k_d)s + (k_p + k)}$$

Given $m = 4 \text{ kg}$, $b = 2 \text{ Ns/m}$, $k = 0.1 \text{ N/m}$

d) For the uncontrolled system,

$$\text{natural freq} = \omega_n = \sqrt{\frac{k}{m}} = \sqrt{\frac{0.1}{4}} = \frac{\sqrt{0.1}}{2} = 0.158$$

$$\text{damping ratio} = \zeta = \frac{b}{2\sqrt{km}} = \frac{2}{2\sqrt{0.1(4)}} = \frac{1}{2\sqrt{0.1}} = 1.58$$

Since the damping ratio is > 1 , the uncontrolled system is overdamped

$$2\% \text{ settling time } T_s = \frac{4}{\zeta \omega_n} = \frac{4}{(1.58)(0.158)} = 16.02 \text{ s}$$

e) Overall closed loop transfer function for the system in question as calculated in (c) is

$$\text{CLTF} : \frac{k_p + s k_d}{ms^2 + (b + k_d)s + (k_p + k)}$$

For a P (proportional) controller, $k_d = 0$

$$\therefore \text{CLTF} = \frac{k_p}{ms^2 + bs + (k_p + k)}$$

$$\text{CP} = ms^2 + bs + (k_p + k)$$

$$\text{CP (std. form)} : s^2 + \frac{b}{m}s + \left(\frac{k_p + k}{m}\right)$$

$$\therefore \omega_n = \sqrt{\frac{k_p + k}{m}} \quad \text{and} \quad \zeta = \frac{b}{2\sqrt{(k_p + k)m}}$$

For critical damping, $\zeta = 1$

$$\therefore 1 = \frac{b}{2\sqrt{(k_p + k)m}}$$

$$2\sqrt{(k_p + k)m} = b$$

$$2\sqrt{(k_p + 0.1)4} = 2$$

$$(k_p + 0.1)4 = 1$$

$$k_p = 0.25 - 0.1 = 0.15$$

f) Overall CLTF =
$$\frac{k_p + s k_d}{m s^2 + (b + k_d) s + (k_p + k)}$$

for a D (derivative) controller, $k_p = 0$

$$\therefore \text{CLTF} = \frac{s k_d}{m s^2 + (b + k_d) s + k}$$

$$CP = m s^2 + (b + k_d) s + k$$

$$CP(\text{std. form}) : s^2 + \left(\frac{b + k_d}{m} \right) s + \frac{k}{m}$$

$$\omega_n = \sqrt{\frac{k}{m}} \quad \text{and} \quad \zeta = \frac{b + k_d}{2 \sqrt{k m}}$$

for critical damping, $\zeta = 1$

$$2 \sqrt{k m} = b + k_d$$

$$2 \sqrt{0.1(4)} = 2 + k_d$$

$$4(0.1)(4) = k_d^2 + 4k_d + (k_d + 2)^2$$

$$\neq \sqrt{1.6} = k_d + 2$$

$$\therefore k_d = \frac{-2 + \sqrt{1.6}}{1} = -0.735$$

g) Overall CLTF for PD controller:

$$\frac{k_p + s k_d}{m s^2 + (b + k_d) s + (k_p + k)}$$

$$CP = m s^2 + (b + k_d) s + (k_p + k)$$

$$CP(\text{std. form}) = s^2 + \left(\frac{b + k_d}{m} \right) s + \left(\frac{k_p + k}{m} \right)$$

$$\omega_n = \sqrt{\frac{k_p + k}{m}} \quad , \quad \zeta = \frac{b + k_d}{2 \sqrt{(k_p + k) m}}$$

for critical damping, $\zeta = 1$

$$b + k_d = 2 \sqrt{(k_p + k) m} \quad \text{--- (1)}$$

for 2% settling time of 0.01,

$$\frac{4}{\zeta \omega_n} = 0.01$$

$$\omega_n = \frac{4}{0.01(1)} = 400$$

$$\omega_n = \sqrt{\frac{k_p + k}{m}}$$

$$(400)^2 = \frac{k_p + 0.1}{4}$$

$$\therefore k_p = 640000 - 0.1$$

$$\boxed{k_p = 639999.9}$$

from ①, $b + k_d = 2 \sqrt{(k_p + k)m}$

$$= 2 \sqrt{\frac{k_p + k}{m} \cdot m}$$

$$= 2 \omega_n \cdot m$$

$$= 2(400)(4)$$

$$k_d = \frac{1600}{2} = 3200, -2$$

$$\boxed{k_d = 3198}$$

Problem 4 :

(a) $M\ddot{\theta}(t) + b\dot{\theta} = u(t) + d(t)$

Given: disturbance $\rightarrow 0$, $M=10$, $b=1$

Taking Laplace Transform for robot model :

$$Ms^2\theta(s) + bs\theta(s) = U(s)$$

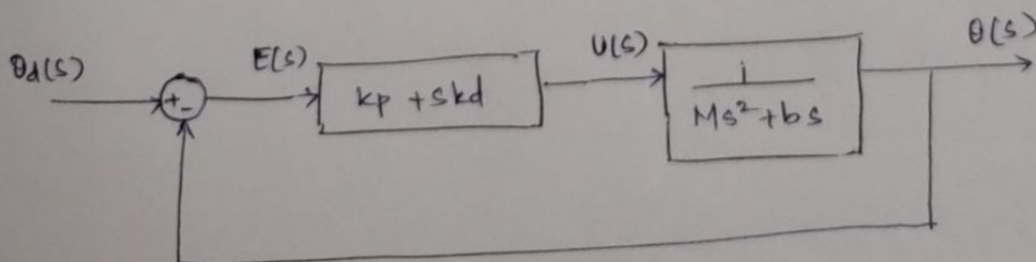
$$\therefore \frac{\theta(s)}{U(s)} = \frac{1}{Ms^2 + bs}$$

PD controller : $(e)k_p + k_d(\dot{e}) = U$ (Input to system)

Taking Laplace transform :

$$k_p E(s) + sk_d E(s) = U(s)$$

$$\frac{U(s)}{E(s)} = (k_p + sk_d)$$



$$E(s) = \theta_d(s) - \theta(s)$$

$$U(s) = E(s)(k_p + sk_d)$$

$$\Theta(s) = U(s) \cdot \frac{1}{Ms^2 + bs}$$

$$\Theta(s) = \frac{(\Theta_d(s) - \Theta(s))(k_p + s k_d)}{Ms^2 + bs}$$

$$CLTF = \frac{\Theta(s)}{\Theta_d(s)} = \frac{k_p + s k_d}{Ms^2 + (b + k_d)s + k_p}$$

$$\text{standardized CP} = s^2 + \left(\frac{b + k_d}{M}\right)s + \left(\frac{k_p}{M}\right)$$

$$\therefore \omega_n = \sqrt{\frac{k_p}{m}}, \quad \zeta = \frac{b + k_d}{2\sqrt{k_p m}}$$

Given that $T_s = 2$ and $\zeta = 1$ (critically damped)

$$\frac{4}{\zeta \omega_n} = 2$$

$$\omega_n = \frac{4}{2} = 2 \Rightarrow \frac{k_p}{M} = 4 \Rightarrow k_p = 4(10) = 40$$

$$b + k_d = 2\sqrt{k_p \cdot m}$$

$$1 + k_d = 2\sqrt{40(10)}$$

$$= 2\sqrt{400}$$

$$= 2(20) = 40$$

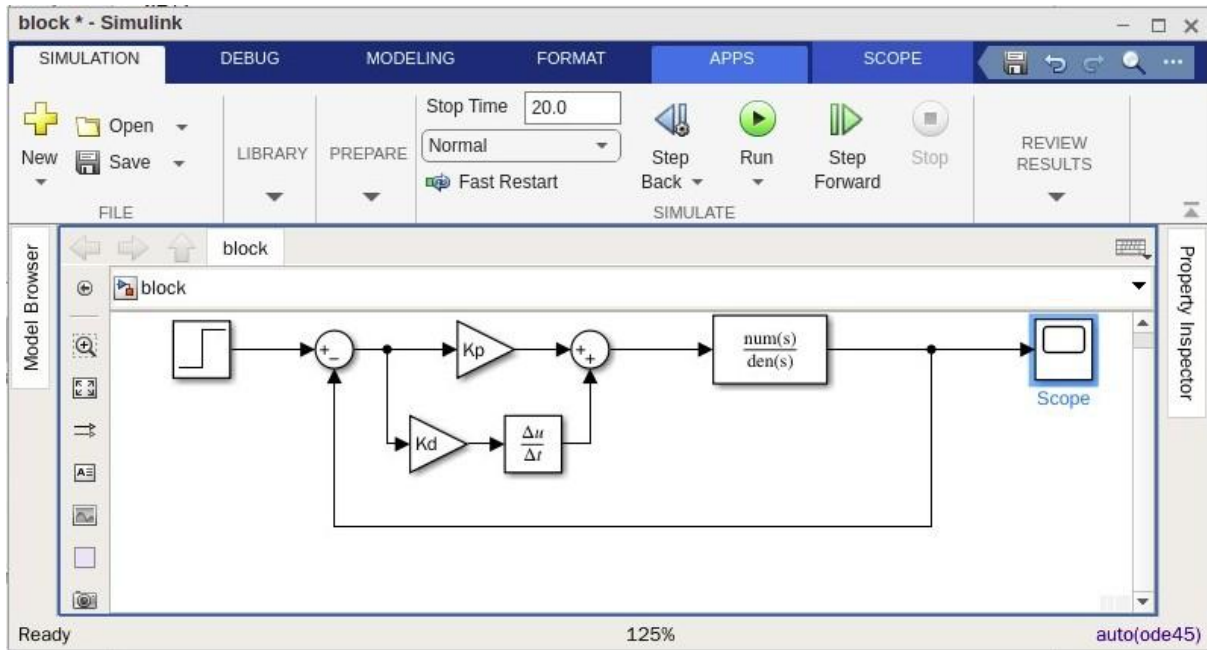
$$\therefore k_d = 40 - 1 = 39$$

RBE500- Foundations of Robotics

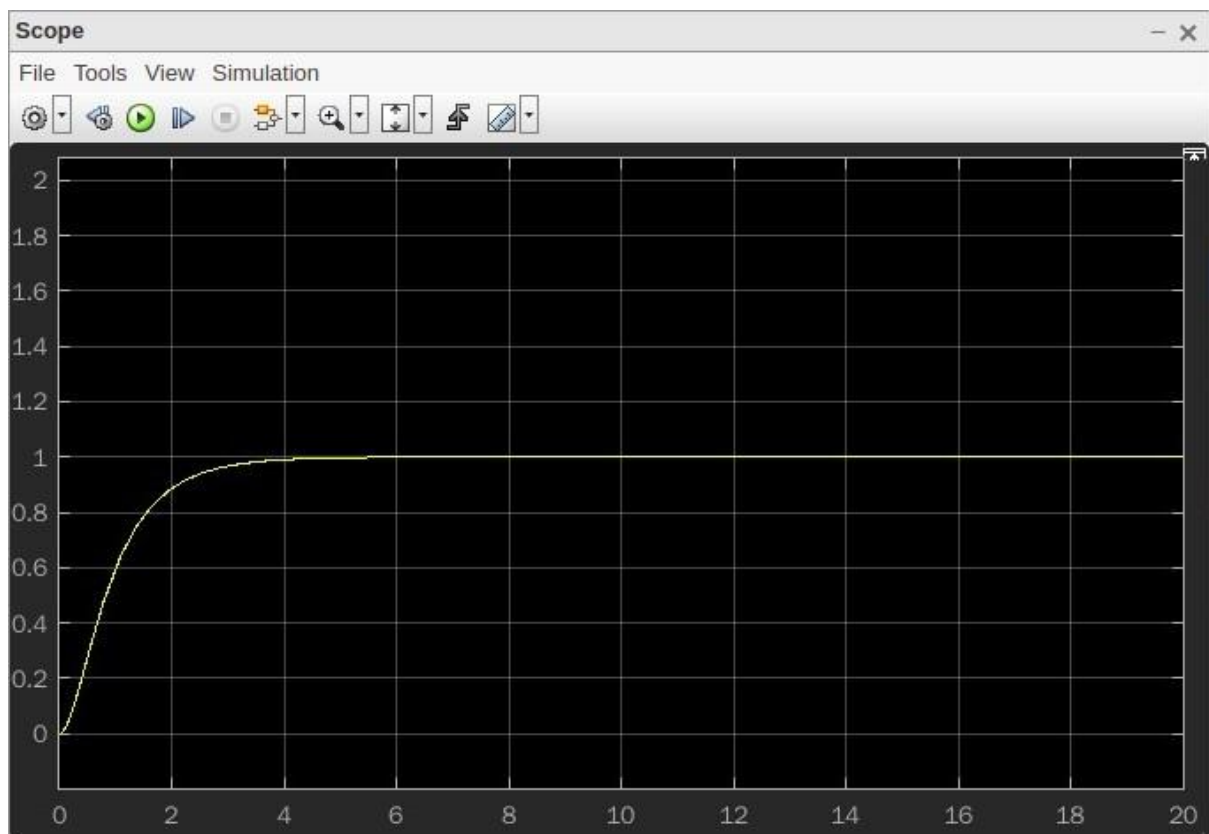
Problem Set 3 Report

Problem 4 (Part b): Given- $M = 10$, $b = 1$

Block diagram:



Scope output:



PD Controller: $K_p = 40$, $K_d = 39$

The controller gains K_p and K_d as calculated from the requirement of a critically damped system with settling time of 2s did not result in the expected system response.

The settling time was found to be 3.4s rather than its intended value of 2s. It did come close to representing a critically damped system, though.

The deviation from the expected response was because of the fact that the system was untuned and the expressions for ω_n and ζ were only approximations.

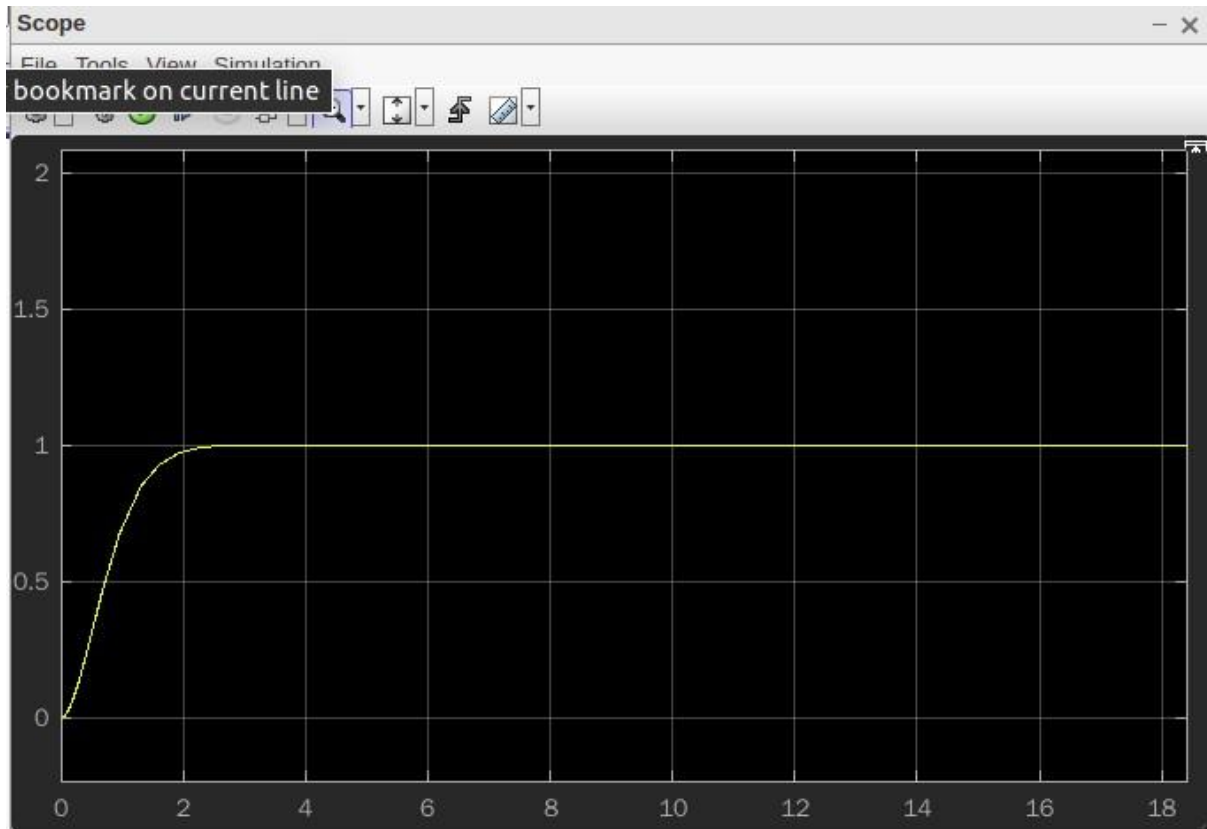
Tuning process (Part c) :

Initially, K_p was kept constant and K_d was varied since derivative gain helps with reducing the settling time. On increasing K_d , it was found that the settling time was increasing still. Hence, now K_d was decreased from its original value of 39 to 35 to 30 and so on gradually to bring the settling time as close as possible to 2s. Decreasing it to 28 caused a slight overshoot which was undesirable. Finally, the sweet spot of 29.5 was set where the settling time was close to 2 without any overshoot.

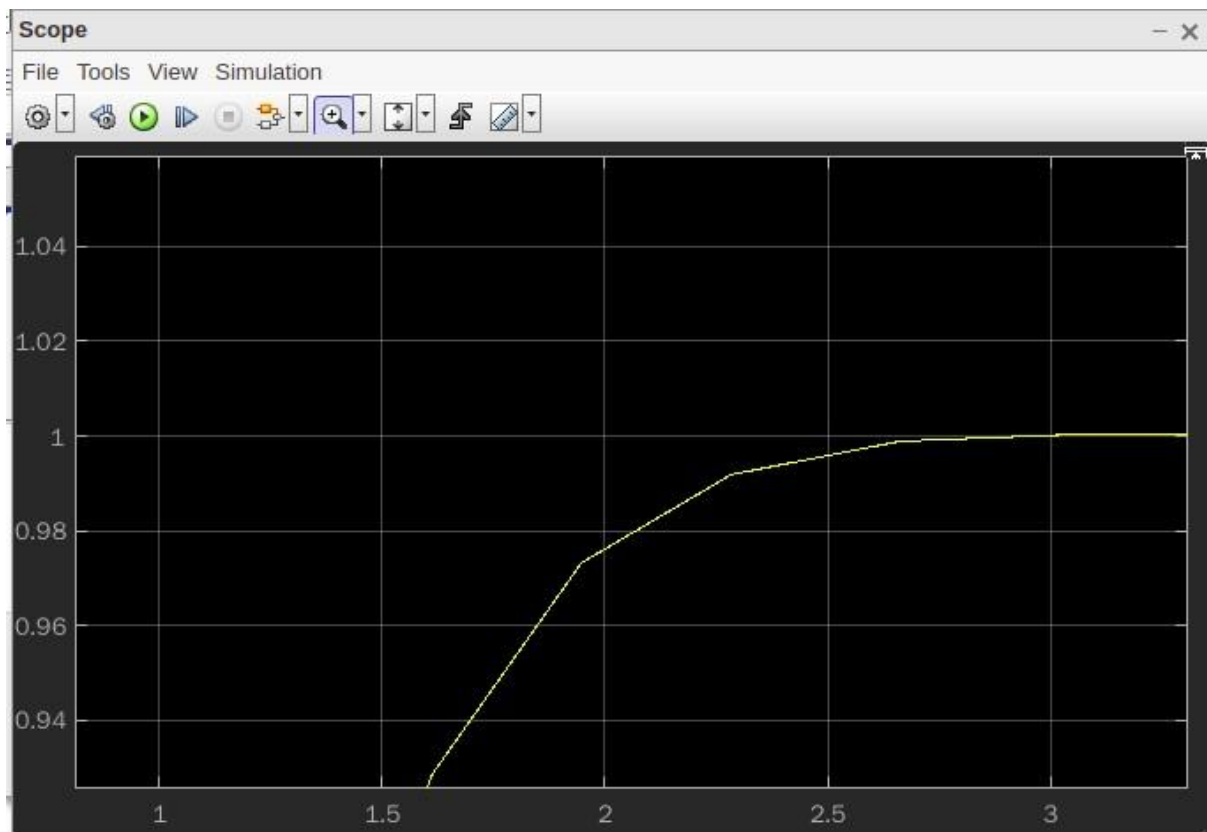
To check the effect of K_p on the system response, it was initially increased which resulted in increasing the overshoot and slight decrease of settling time. Reducing the K_p from the original value increased the settling time drastically.

Parameter changed	Value	Overshoot	Settling time	Stability
K_d	45	No effect	4.1s	Stable
K_d	50	No effect	4.7s	Stable
K_d	35	No effect	2.9s	Stable
K_d	30	No effect	2.15s	Stable
K_d	28	Slight overshoot	1.85s	Stable
K_d	29	Slight overshoot	~2s	Stable
K_d	29.5	No overshoot	2.05s	Stable
K_p	50	Slight overshoot	1.48s	Stable
K_p	60	Overshoot increases	1.17s	Stable
K_p	30	No overshoot	3.2s	Stable
K_p	15	No overshoot	7.3s	Stable

Results post tuning:

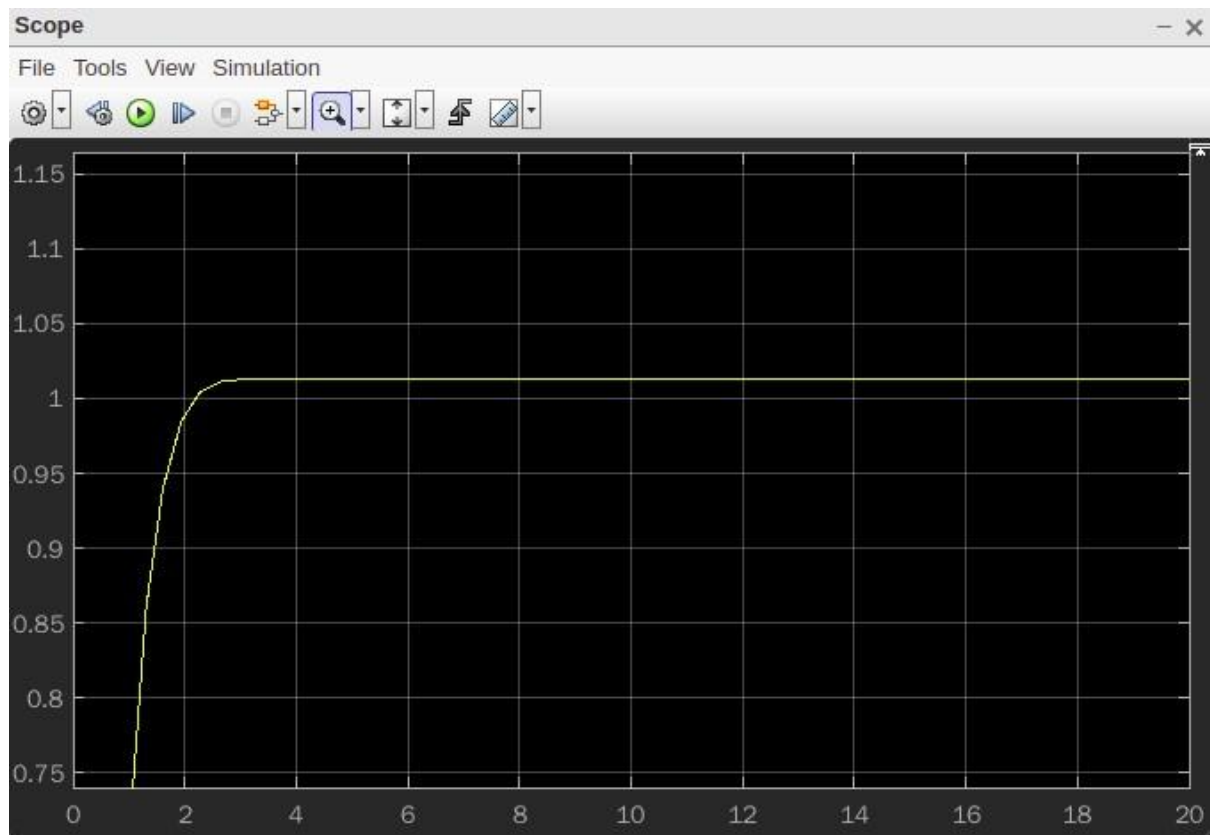


Settling time ~2s at $K_d = 29.5$:



Part d:

The steady state error can be seen in the response plot below. Its value is 0.0125.



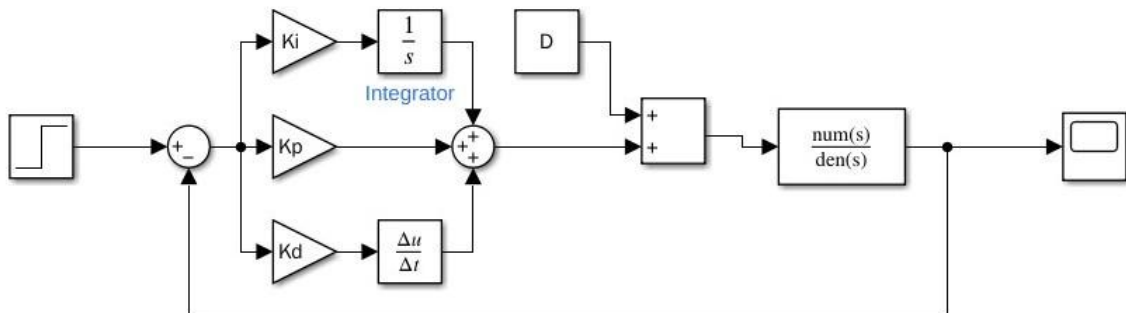
Part e:

Following table shows the tuning stages.

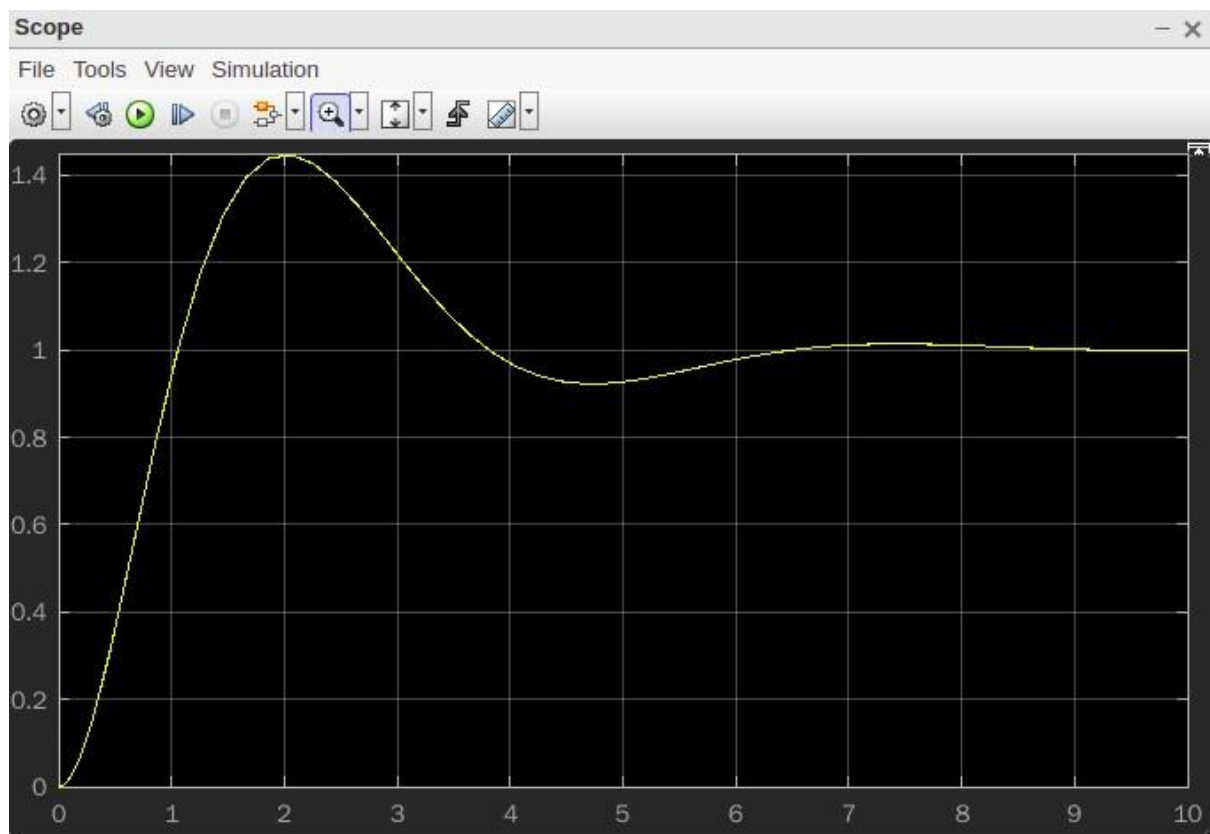
The parameters K_p and K_d were also varied but the performance was best for the already set values from previous tuning.

Parameter changed	Value	Overshoot	Settling time	Steady-state error
K_i	0	No effect	1.9s	0.0125
K_i	1	Increases	-	-
K_i	5	Increases	5.87s	0
K_i	15	Same as before	5.87s	0
K_i	35	Slight decrease	6.01s	0

Block Diagram



Response



The final values of K_p , K_d and K_i were found out to be **40**, **29.5** and **35** respectively.