

# ELECENG 3CL4 Lab 2 Pre-lab

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# 1 Description of Laboratory Equipment

## Pre-lab Question 1

The step response of the model  $G(s)$  is in Equation 1. The derivation of the step response is shown below.

$$\begin{aligned}
 \mathcal{L}^{-1} \left\{ G(s) \frac{1}{s} \right\} &= \mathcal{L}^{-1} \left\{ \frac{A}{s(s\tau_m + 1)} \frac{1}{s} \right\} \\
 &= A \mathcal{L}^{-1} \left\{ \frac{1}{s^2(s\tau_m + 1)} \right\} \\
 &= A \mathcal{L}^{-1} \left\{ \frac{\tau_m^2}{s\tau_m + 1} - \frac{\tau_m}{s} + \frac{1}{s^2} \right\} \\
 &= A \left[ \mathcal{L}^{-1} \left\{ \frac{\tau_m^2}{s\tau_m + 1} \right\} - \mathcal{L}^{-1} \left\{ \frac{\tau_m}{s} \right\} + \mathcal{L}^{-1} \left\{ \frac{1}{s^2} \right\} \right] \\
 &= A \left[ \tau_m e^{-\frac{t}{\tau_m}} - \tau_m u(t) + t \right]
 \end{aligned} \tag{1}$$

## Pre-lab Question 2

The step response is not bounded because the term  $t$  is not bounded.

# 2 Closed Loop System Identification

## Pre-lab Question 3

(i) The output signal of the model can be expressed as:

$$\begin{aligned}
 Y(s) &= \frac{G_c(s)G(s)}{1 + H(s)G_c(s)G(s)} R(s) + \frac{G(s)}{1 + H(s)G_c(s)G(s)} T_d(s) \\
 &\quad + \frac{H(s)G_c(s)G(s)}{1 + H(s)G_c(s)G(s)} N(s)
 \end{aligned}$$

If we neglect the effects of the disturbance  $T_d(s)$  and the noise  $N(s)$ , we are left with:

$$Y(s) = \frac{G_c(s)G(s)}{1 + H(s)G_c(s)G(s)} R(s)$$

When  $H(s) = 1$  and we let  $\Theta(s) = Y(s)$ , we can find the transfer function  $T(s)$  in Equation 2 (idk why this isn't working), which is equal to Equation 4a from the lab document.

$$\begin{aligned}
 T(s) &= \frac{\Theta(s)}{R(s)} \\
 &= \frac{\frac{G_c(s)G(s)}{1+H(s)G_c(s)G(s)} R(s)}{R(s)} \\
 &= \frac{G(s)G_c(s)}{1 + G(s)G_c(s)}
 \end{aligned} \tag{2}$$

- (ii) When we substitute  $G(s) = \frac{A}{s(s\tau_m+1)}$  and  $G_c(s) = K$  into Equation 2, and then substituting  $\sqrt{\frac{KA}{\tau_m}} = \omega_n$  and  $\frac{1}{2\omega_n\tau_m}$  into the result, we can derive Equation 3 which is equal to Equation 4b from the lab document:

$$\begin{aligned}
T(s) &= \frac{G(s)G_c(s)}{1 + G(s)G_c(s)} \\
&= \frac{\frac{A}{s(s\tau_m+1)}K}{1 + \frac{A}{s(s\tau_m+1)}K} \\
&= \frac{KA}{s(s\tau_m+1)(1 + \frac{A}{s(s\tau_m+1)}K)} \\
&= \frac{KA}{s(s\tau_m+1 + \frac{KA}{s})} \\
&= \frac{KA}{s^2\tau_m + s + KA} \\
&= \frac{KA/\tau_m}{s^2 + (1/\tau_m)s + KA/\tau_m} \\
&= \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}
\end{aligned} \tag{3}$$

#### Pre-lab Question 4

When given  $\zeta$  and  $\omega_n$ , we can derive Equation 4 for  $\tau_m$ :

$$\begin{aligned}
\zeta &= \frac{1}{2\omega_n\tau_m} \\
\tau_m &= \frac{1}{2\omega_n\zeta}
\end{aligned} \tag{4}$$

By substituting Equation 4 into the equation for  $\omega_n$ , we can derive Equation 5 for  $A$ :

$$\begin{aligned}
\omega_n &= \sqrt{\frac{KA}{\tau_m}} \\
\omega_n^2 &= \frac{KA}{\tau_m} \\
A &= \frac{\omega_n^2\tau_m}{K} \\
A &= \frac{\omega_n/2\zeta}{K}
\end{aligned} \tag{5}$$

## 2.1 Closed-loop System Identification from the Step Response

#### Pre-lab Question 5

The percent overshoot is determined at the peak time,  $T_p$ , of the step response. For the provided step response, this peak time occurs the first time  $d\theta_{\text{step}}(t)/dt = 0$ , which occurs

when  $\omega_n \sqrt{1 - \zeta^2} t = \pi$ . Therefore, the peak time  $T_p$  is equal to  $\frac{\pi}{\omega_n \sqrt{1 - \zeta^2}}$ . At the peak time, the peak value of the step response is  $1 + \exp\left(-\zeta \pi / \sqrt{1 - \zeta^2}\right)$ . Therefore, the percent overshoot can be defined as  $P.O. = 100 \frac{(1 + \exp(-\zeta \pi / \sqrt{1 - \zeta^2})) - 1}{1} = 100 \exp\left(\frac{-\zeta \pi}{\sqrt{1 - \zeta^2}}\right)$ .

### Pre-lab Question 6

As discussed in the solution to Pre-lab Question 5, the peak time occurs the first time  $d\theta_{\text{step}}(t)/dt = 0$ , which occurs when  $\omega_n \sqrt{1 - \zeta^2} t = \pi$ . Therefore, the peak time  $T_p$  is equal to  $\frac{\pi}{\omega_n \sqrt{1 - \zeta^2}}$ .

### Pre-lab Question 7

From the equation for P.O., you can calculate  $\zeta$  as:

$$\begin{aligned}
 P.O. &= 100 \exp\left(\frac{-\zeta \pi}{\sqrt{1 - \zeta^2}}\right) \\
 \frac{P.O.}{100} &= \exp\left(\frac{-\zeta \pi}{\sqrt{1 - \zeta^2}}\right) \\
 \ln\left(\frac{P.O.}{100}\right) &= \frac{-\zeta \pi}{\sqrt{1 - \zeta^2}} \\
 \frac{(\ln(\frac{P.O.}{100}))^2}{\pi^2} &= \frac{\zeta^2}{1 - \zeta^2} \\
 \frac{(\ln(\frac{P.O.}{100}))^2}{\pi^2} &= 1 - \frac{1}{1 + \zeta^2} \\
 1 - \frac{(\ln(\frac{P.O.}{100}))^2}{\pi^2} &= \frac{1}{1 + \zeta^2} \\
 1 + \zeta^2 &= \frac{1}{\left(1 - \frac{(\ln(\frac{P.O.}{100}))^2}{\pi^2}\right)} \\
 \zeta^2 &= \frac{1}{\left(1 - \frac{(\ln(\frac{P.O.}{100}))^2}{\pi^2}\right)} - 1 \\
 \zeta &= \sqrt{\frac{1}{\left(1 - \frac{(\ln(\frac{P.O.}{100}))^2}{\pi^2}\right)} - 1} \tag{6}
 \end{aligned}$$

ok elts try something thats probably a bit easier

$$\begin{aligned}
 T_p &= \frac{\pi}{\omega_n \sqrt{1 - \zeta^2}} \\
 \omega_n &= \frac{\pi}{T_p \sqrt{1 - \zeta^2}} \tag{7}
 \end{aligned}$$

## 2.2 Closed-Loop System Identification using the Frequency Response

Pre-lab Question 8

Pre-lab Question 9

Pre-lab Question 10