

Lab9

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1 Lab de Circuitos - Preparatório 9

1.1 Rafael Rubim Cabral - 1511068

1.1.1 1)

Cálculo da função de transferência $H_Z(s)$:

$$Z_1(s) = \frac{1}{sC} // (Ls + Z) = \frac{1}{\frac{1}{Ls + Z} + sC} = \frac{Ls + Z}{1 + LCs^2 + ZCs}$$

$$V_2(s) = V(s) \frac{Z_1(s)}{R + Z_1(s)}$$

$$V_Z(s) = V_2(s) \frac{Z}{Ls + Z}$$

$$H_Z(s) = \frac{V_Z(s)}{V(s)} = \frac{Z_1(s)}{R + Z_1(s)} \frac{Z}{Ls + Z} = \frac{Z_1 Z}{RLs + ZR + Z_1 Ls + Z_1 Z} = \frac{Z}{\frac{RLs}{Z_1} + \frac{ZR}{Z_1} + Ls + Z}$$

$$H_Z(s) = \frac{Z}{\frac{RLs + RL^2Cs^3 + ZRLCs^2}{Ls + Z} + \frac{ZR + ZRLCs^2 + Z^2RCs}{Ls + Z} + Ls + Z}$$

$$H_Z(s) = \frac{ZLs + Z^2}{RLs + RL^2Cs^3 + ZRLCs^2 + ZR + ZRLCs^2 + Z^2RCs + (Ls + Z)^2}$$

$$H_Z(s) = \frac{ZLs + Z^2}{RL^2Cs^3 + ZRLCs^2 + ZRLCs^2 + L^2s^2 + RLs + Z^2RCs + 2ZLs + ZR + Z^2}$$

$$H_Z(s) = \frac{ZLs + Z^2}{RL^2Cs^3 + (2ZRLC + L^2)s^2 + (RL + Z^2RC + 2ZL)s + (ZR + Z^2)}$$

Vamos considerar a fase da fonte $\phi = 0$. Logo temos o fasor $\dot{V} = Ae^{j0}$. O fasor da saída \dot{V}_Z será:

$$\dot{V}_Z = H_Z(j\omega_0)\dot{V}$$

$$H_Z(j\omega_0) = \frac{jZL\omega_0 + Z^2}{-jRL^2C\omega_0^3 - (2ZRLC + L^2)\omega_0^2 + (RL + Z^2RC + 2ZL)j\omega_0 + (ZR + Z^2)}$$

$$H_Z(j\omega_0) = \frac{Z^2 + jZL\omega_0}{[ZR + Z^2 - (2ZRLC + L^2)\omega_0^2] + j[(RL + Z^2RC + 2ZL)\omega_0 - RL^2C\omega_0^3]}$$

1.1.2 a)

$Z = Z_0 e^{j\phi}$ puramente resistivo: $Z = R + j.0 \rightarrow \phi = 0 \rightarrow Z = Z_0$

$$H_Z(j\omega_0) = \frac{Z_0^2 + jZ_0L\omega_0}{[Z_0R + Z_0^2 - (2RLZ_0C + L^2)\omega_0^2] + j[(RL + Z_0^2RC + 2LZ_0)\omega_0 - RL^2C\omega_0^3]}$$

Logo $\dot{V}_Z = H_Z(j\omega_0)\dot{V} = H_Z(j\omega_0)A$ e então $V_Z(t) = \Re(H_Z(j\omega_0)Ae^{j\omega_0t})$

$$V_Z(t) = \Re\left(\frac{(Z_0^2 + jZ_0L\omega_0)Ae^{j\omega_0t}}{[Z_0R + Z_0^2 - (2RLZ_0C + L^2)\omega_0^2] + j[(RL + Z_0^2RC + 2LZ_0)\omega_0 - RL^2C\omega_0^3]}\right)$$

1.1.3 b)

$Z = Z_0e^{j\phi}$ puramente capacitivo: $Z = 0 - j.X \rightarrow \phi = \frac{3\pi}{2} \rightarrow Z = -jZ_0$

$$H_Z(j\omega_0) = \frac{-Z_0^2 + Z_0L\omega_0}{[2LZ_0\omega_0 - Z_0^2 - L^2\omega_0^2] + j[(RL - Z_0^2RC)\omega_0 - Z_0R - RL^2C\omega_0^3 + 2RLZ_0C\omega_0^2]}$$

Logo $\dot{V}_Z = H_Z(j\omega_0)\dot{V} = H_Z(j\omega_0)A$ e então $V_Z(t) = H_Z(j\omega_0)Ae^{j\omega_0t}$

$$V_Z(t) = \Re\left(\frac{(-Z_0^2 + Z_0L\omega_0)Ae^{j\omega_0t}}{[2LZ_0\omega_0 - Z_0^2 - L^2\omega_0^2] + j[(RL - Z_0^2RC)\omega_0 - Z_0R - RL^2C\omega_0^3 + 2RLZ_0C\omega_0^2]}\right)$$

1.1.4 c)

$Z = Z_0e^{j\phi}$ puramente indutivo: $Z = 0 + j.X \rightarrow \phi = \frac{\pi}{2} \rightarrow Z = jZ_0$

$$H_Z(j\omega_0) = \frac{-Z_0^2 - Z_0L\omega_0}{[-2LZ_0\omega_0 - Z_0^2 - L^2\omega_0^2] + j[(RL - Z_0^2RC)\omega_0 + Z_0R - RL^2C\omega_0^3 - 2RLZ_0C\omega_0^2]}$$

Logo $\dot{V}_Z = H_Z(j\omega_0)\dot{V} = H_Z(j\omega_0)A$ e então $V_Z(t) = H_Z(j\omega_0)Ae^{j\omega_0t}$

$$V_Z(t) = \Re\left(\frac{(-Z_0^2 - Z_0L\omega_0)Ae^{j\omega_0t}}{[-2LZ_0\omega_0 - Z_0^2 - L^2\omega_0^2] + j[(RL - Z_0^2RC)\omega_0 + Z_0R - RL^2C\omega_0^3 - 2RLZ_0C\omega_0^2]}\right)$$

2) As simulações não puderam ser realizadas pois a licença da PUC do circuitlab não está funcionando e não encontrei outro simulador que gere gráficos com desenho de escala. Seguem as contas:

Adotando nos 3 casos:

$$R = 100\Omega$$

$$L = 1\mu H$$

$$C = 1\mu F$$

$$\omega_0 = 100\text{rad/s}$$

$$f = \frac{\omega_0}{2\pi} = 15,92\text{Hz}$$

$$T = \frac{1}{f} = 62,66\text{ms}$$

$$A = 10V$$

1.1.5 a)

$$Z_R = 100\Omega$$

$$Z = Z_0 = 100\Omega$$

$$V_Z(t) = \Re\left(\frac{(Z_0^2 + jZ_0L\omega_0)Ae^{j\omega_0t}}{[Z_0R + Z_0^2 - (2RLZ_0C + L^2)\omega_0^2] + j[(RL + Z_0^2RC + 2LZ_0)\omega_0 - RL^2C\omega_0^3]}\right)$$

$$\begin{aligned}
V_Z(t) &= \mathbb{R} \left(\frac{(10^4 + j10^{-2})10e^{j10^2t}}{[10^4 + 10^4 - (2.10^{-8} + 10^{-12})10^4] + j[(10^{-4} + 1 + 2.10^{-4})10^2 - 10^{-16}10^6]} \right) \\
V_Z(t) &= \mathbb{R} \left(\frac{10^4 e^{j.0} 10e^{j10^2t}}{[-10^{-8}] + j[(10^2 + 3.10^{-2}) - 10^{-10}]} \right) \\
V_Z(t) &= \mathbb{R} \left(\frac{10^5 e^{j10^2t}}{[-10^{-8}] + j[10^2]} \right) \\
V_Z(t) &= \mathbb{R} \left(\frac{10^5 e^{j10^2t}}{10^2 e^{j\frac{\pi}{2}}} \right) \\
V_Z(t) &= 10^3 \sin(10^2t - \frac{\pi}{2})
\end{aligned}$$

1.1.6 b)

$$\begin{aligned}
Z_C &= 1\mu F \\
Z &= -jZ_0 = \frac{-j}{\omega_0 Z_C} = \frac{-j}{10^2 10^{-6}} \\
Z_0 &= 10^4 \\
V_Z(t) &= \mathbb{R} \left(\frac{(-Z_0^2 + Z_0 L \omega_0) A e^{j\omega_0 t}}{[2LZ_0 \omega_0 - Z_0^2 - L^2 \omega_0^2] + j[(RL - Z_0^2 RC) \omega_0 - Z_0 R - RL^2 C \omega_0^3 + 2RLZ_0 C \omega_0^2]} \right) \\
V_Z(t) &= \mathbb{R} \left(\frac{(-10^8 + 1)10e^{j10^2t}}{[2 - 10^8 - 10^{-8}] + j[(10^{-4} - 10^4)10^2 - 10^6 - 10^{-16}10^6 + 2.10^{-2}]} \right) \\
V_Z(t) &= \mathbb{R} \left(\frac{10e^{j10^2t}}{[-10^8] + j[-2.10^6]} \right) \\
V_Z(t) &= \mathbb{R} \left(\frac{10e^{j10^2t}}{10^8 e^{j.0}} \right) \\
V_Z(t) &= 10^{-7} \sin(10^2t)
\end{aligned}$$

1.1.7 c)

$$\begin{aligned}
Z_L &= 1\mu H \\
Z &= jZ_0 = j\omega_0 Z_L = j10^2 10^{-6} \\
Z_0 &= 10^{-4} \\
V_Z(t) &= \mathbb{R} \left(\frac{(-Z_0^2 - Z_0 L \omega_0) A e^{j\omega_0 t}}{[-2LZ_0 \omega_0 - Z_0^2 - L^2 \omega_0^2] + j[(RL - Z_0^2 RC) \omega_0 + Z_0 R - RL^2 C \omega_0^3 - 2RLZ_0 C \omega_0^2]} \right) \\
V_Z(t) &= \mathbb{R} \left(\frac{(-10^{-8} - 10^{-8})10e^{j10^2t}}{[-2.10^{-8} - 10^{-8} - 10^{-8}] + j[(10^{-4} - 10^{-12})10^2 + 10^{-2} - 10^{-16}10^6 - 2.10^{-14}10^4]} \right) \\
V_Z(t) &= \mathbb{R} \left(\frac{-2.10^{-7} e^{j10^2t}}{[-4.10^{-8}] + j[2.10^{-2} - 4.10^{-10}]} \right) \\
V_Z(t) &= \mathbb{R} \left(\frac{-10^{-7} e^{j10^2t}}{[-2.10^{-8}] + j[10^{-2} - 210^{-10}]} \right) \\
V_Z(t) &= \mathbb{R} \left(\frac{-10^{-7} e^{j10^2t}}{10^{-2} e^{j\frac{\pi}{2}}} \right) \\
V_Z(t) &= -10^{-5} \sin(10^2t - \frac{\pi}{2})
\end{aligned}$$

1.1.8 3)

1.1.9 a)

A frequência de ressonância é igual a $\omega_0 = \frac{1}{\sqrt{LC}} = \frac{1}{\sqrt{3 \cdot 10^4 \cdot 10^{-6} \cdot 10^{-9}}} = \frac{1}{\sqrt{30 \cdot 10^{-6}}} = \frac{\sqrt{30} \cdot 10^6}{30}$ rad/s

1.1.10 b)

A frequência de ressonância é a frequência para a qual a soma das impedâncias do capacitor e do indutor é igual a 0. Como a impedância equivalente do circuito RLC é $Z_R + Z_L + Z_C$, então a impedância equivalente do circuito valerá $Z_R = R$. Isso significa que no regime senoidal permanente a corrente estará em fase com a fonte e o circuito pode ser enxergado como um circuito puramente resistivo.

In []: