# Electrical Science-2 (15B11EC211)

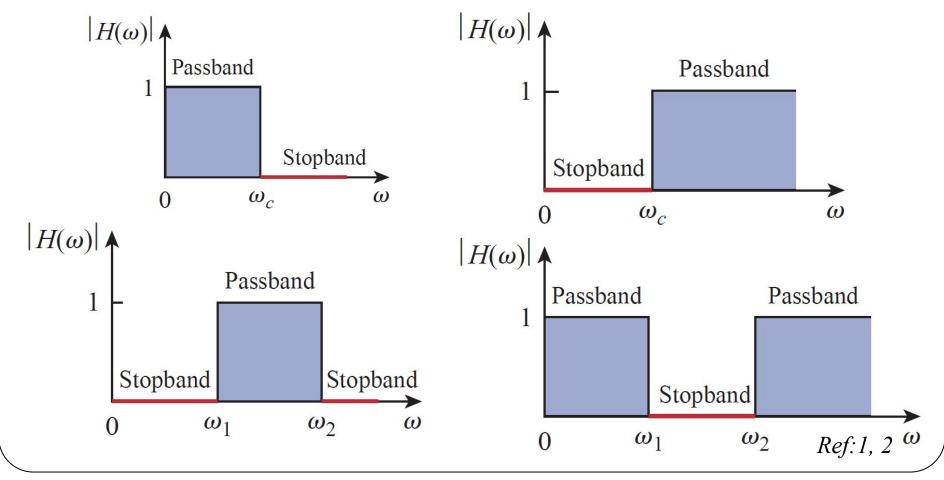
# Unit-3 Operational Amplifier and Filters Lecture-4

#### Contents

- Passive filters and its types
- Characteristics of ideal filters
- Design of lowpass filter
- Design of highpass filter
- Design of bandpass filter
- Design of bandstop filter
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#### **Passive Filters**

- A filter is a circuit that is designed to pass signals with desired frequencies and reject or attenuate others.
- A filter is a passive filter if it consists of only passive elements R, L, and
- There are four types of filters: Lowpass, Highpass, Bandpass, Bandstop



#### Summary of the characteristics of ideal filters

Table 1: Summary of the characteristics of the filters.

<b>Type of Filter</b>	H (0)	<b>H</b> (∞)	$H(\omega_c)$ or $H(\omega_0)$
Lowpass	1	0	1/√2
Highpass	0	1	<b>1</b> /√ <b>2</b>
Bandpass	0	0	1
Bandstop	1	1	0

Note  $\mathbb{D}\omega_c$  is the cutoff frequency for lowpass and highpass filters;  $\omega_0$  is the center frequency for bandpass and bandstop filters.

☐ Be aware that the characteristics in Table are only valid for first- or second-order filter

#### **Lowpass Filter**

- A lowpass filter is designed to pass only frequencies from dc up to the cutoff frequency  $\omega_c$
- A typical lowpass filter is formed when the output of an RC circuit is taken off the capacitor as shown in Figure.

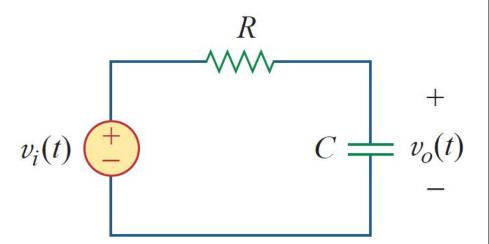
#### **Transfer function**

$$H(\omega)$$

$$\mathbf{H}(\omega) = \frac{\mathbf{V}_o}{\mathbf{V}_i} = \frac{1/j\omega C}{R + 1/j\omega C}$$

$$\mathbf{H}(\omega) = \frac{1}{1 + j\omega RC}$$

Note that 
$$H(0) = 1$$
,  $H(\infty) = 0$ 



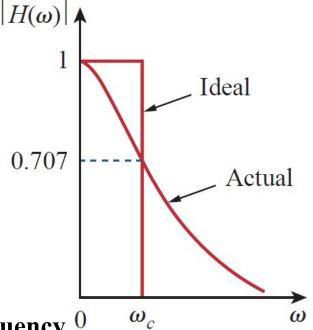
#### **Lowpass Filter**

The cutoff frequency is the frequency at which the transfer function H drops in magnitude to 70.71% of its maximum value.

Cutoff frequency 
$$\omega_c$$
 is obtained by putting 
$$|\mathbf{H}(\omega)| = 1/\sqrt{2}$$

$$H(\omega_c) = \frac{1}{\sqrt{1 + \omega_c^2 R^2 C^2}} = \frac{1}{\sqrt{2}}$$

$$\omega_c = 1/RC$$



Note: Cutoff frequency is also called the rolloff frequency.

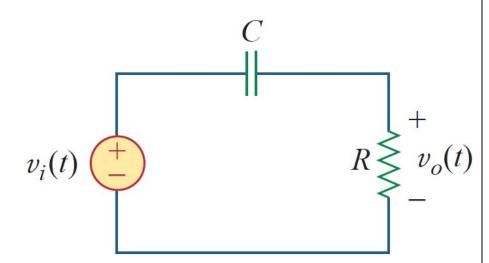
- It is also regarded as the frequency at which the power dissipated in a circuit is half of its maximum value.
- A lowpass filter can also be formed when the output of an RL circuit is taken off the resistor. Ref:1, 2

#### **Highpass Filter**

- A highpass filter is designed to pass all frequencies above its cutoff frequency  $\omega_{\text{c}}$
- A highpass filter is formed when the output of an RC circuit is taken off the resistor as shown in Figure

Transfer function
$$\mathbf{H}(\omega) = \frac{\mathbf{V}_o}{\mathbf{V}_i} = \frac{R}{R + 1/j\omega C}$$

$$\mathbf{H}(\omega) = \frac{j\omega RC}{1 + j\omega RC}$$
 $v_i$ 



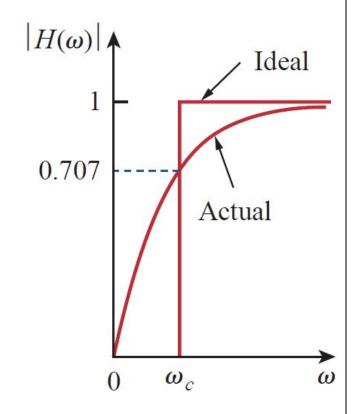
Note that 
$$H(0) = 0$$
,  $H(\infty) = 1$ 

# **Highpass Filter**

Corner or cutoff frequency is

$$\omega_c = 1/RC$$

Note: A highpass filter can also be formed when the output of an RL circuit is taken off the inductor.



- A bandpass filter is designed to pass all frequencies within a band of frequencies,  $\omega_1 < \omega < \omega_2$
- ☐ The RLC series resonant circuit provides a bandpass filter when the output is taken off the resistor as shown in Figure

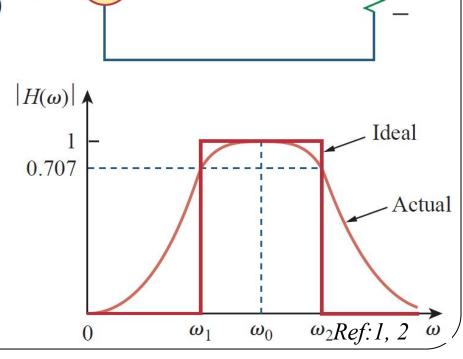
**Transfer function** 

$$\mathbf{H}(\boldsymbol{\omega}) = \frac{\mathbf{V}_o}{\mathbf{V}_i} = \frac{R}{R + j(\omega L - 1/\omega C)} v_i(t)$$

We observed that H(0) = 0,  $H(\infty) = 0$ 

 $\Leftrightarrow$  Center frequency  $\omega_0$ 

$$\rho_0 = \frac{1}{\sqrt{LC}}$$



П Since the bandpass filter is a series resonant circuit, the half-power frequencies, the bandwidth, and the quality factor can be determined by analysis of series resonant circuit

Input impedance, Z

Input impedance, Z
$$\mathbf{V}_{s} = V_{m} \angle \theta$$

$$\mathbf{Z} = \mathbf{H}(\omega) = \frac{\mathbf{V}_{s}}{\mathbf{I}} = R + j\omega L + \frac{1}{j\omega C}$$

$$\mathbf{Z} = R + j\left(\omega L - \frac{1}{\omega C}\right)$$
Series resonant circuit

Resonance results when the imaginary part of the transfer function is zero

$$Im(\mathbf{Z}) = \omega L - \frac{1}{\omega C} = 0$$

• The value of  $\omega$  that satisfies the resonance condition is called the resonant frequency  $\omega_n$  .

$$\omega_0 L = \frac{1}{\omega_0 C}$$
  $\omega_0 = \frac{1}{\sqrt{LC}} \text{ rad/s}$ 

$$\Box \quad \text{Since } \omega_0 = 2\Pi f_0 \qquad \qquad f_0 = \frac{1}{2\pi \sqrt{LC}} \text{ Hz}$$

The half-power frequencies ( $\omega_1$  and  $\omega_2$  are called the half-power frequencies) are obtained by setting  $Z = (\sqrt{2})R$ 

$$\sqrt{R^2 + \left(\omega L - \frac{1}{\omega C}\right)^2} = \sqrt{2}R$$
Solving for  $\omega$ , we obtain

Solving for 
$$\boldsymbol{\omega}$$
 , we obtain

$$\omega_1 = -\frac{R}{2L} + \sqrt{\left(\frac{R}{2L}\right)^2 + \frac{1}{LC}}$$

$$\omega_2 = \frac{R}{2L} + \sqrt{\left(\frac{R}{2L}\right)^2 + \frac{1}{LC}}$$

$$\alpha_2 = \frac{R}{2L} + \sqrt{\left(\frac{R}{2L}\right)^2 + \frac{1}{LC}}$$
Ref:1, 2

☐ Resonant frequency is the geometric mean of the half-power frequencies

$$\omega_0 = \sqrt{\omega_1 \omega_2}$$

☐ Half-power Bandwidth B

$$B = \omega_2 - \omega_1$$

- Quality factor Q:
   The "sharpness" of the resonance in a resonant circuit is measured quantitatively by the quality factor Q.
- The quality factor relates the maximum or peak energy stored to the energy dissipated in the circuit per cycle of oscillation:

$$Q = 2\pi \frac{\text{Peak energy stored in the circuit}}{\text{Energy dissipated by the circuit}}$$
in one period at resonance

$$Q = 2\pi \frac{\frac{1}{2}LI^2}{\frac{1}{2}I^2R(1/f_0)} = \frac{2\pi f_0 L}{R}$$

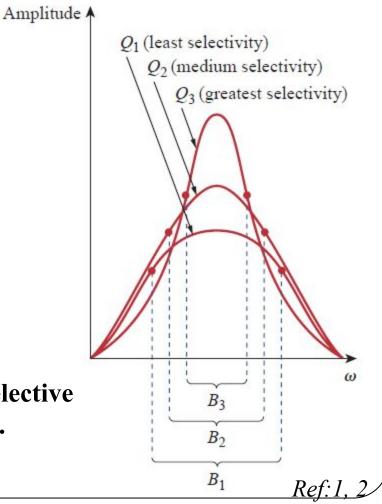
$$Q = \frac{\omega_0 L}{R} = \frac{1}{\omega_0 CR}$$

The relationship between the bandwidth B and the quality factor Q R  $\omega_0$ 

$$B = \frac{R}{L} = \frac{\omega_0}{Q}$$

• Hence, the quality factor of a resonant circuit is the ratio of its resonant frequency to its bandwidth

Figure □ higher the value of Q, the more selective the circuit is but the smaller the bandwidth.



- A resonant circuit is designed to operate at or near its resonant frequency.
- It is said to be a high-Q circuit when  $Q \ge 10$
- For high-Q circuits (Q  $\geq$  10) the half-power frequencies can be approximated as

$$\omega_1 \simeq \omega_0 - \frac{B}{2}, \qquad \omega_2 \simeq \omega_0 + \frac{B}{2}$$

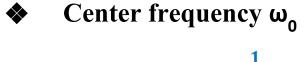
# **Bandstop Filter**

- A bandstop filter is designed to stop or eliminate all frequencies within a band of frequencies,  $\omega_1 < \omega < \omega_2$ .
- $\square$  A bandstop filter is formed when the output RLC series resonant circuit is taken off the LC series combination.

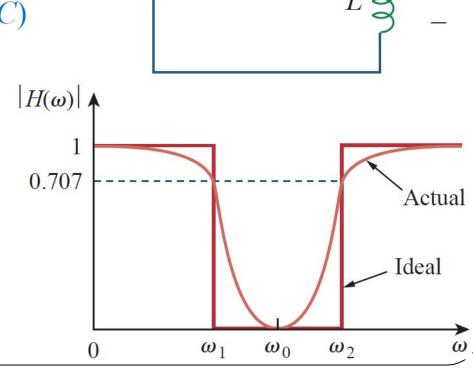
**Transfer function** 

$$\mathbf{H}(\omega) = \frac{\mathbf{V}_o}{\mathbf{V}_i} = \frac{j(\omega L - 1/\omega C)}{R + j(\omega L - 1/\omega C)}$$

Notice that H(0) = 1,  $H(\infty) = 1$ 



$$\omega_0 = \frac{1}{\sqrt{LC}}$$



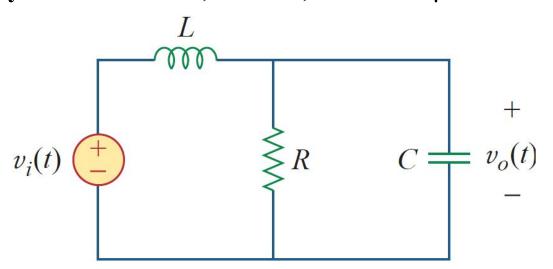
# **Bandstop Filter**

The half-power frequencies, the bandwidth, and the quality factor are calculated using the formulas discussed for a series resonant circuit.

Here,  $\omega_0$  is called the frequency of rejection, while the corresponding bandwidth (B =  $\omega_2 - \omega_1$ ) is known as the bandwidth of rejection.

#### **Example**

Determine what type of filter is shown in Figure. Calculate the corner or cutoff frequency. Take R=2 k $\Omega$ , L=2 H, and C=2  $\mu F$ .



#### **Solution:**

The transfer function is

$$\mathbf{H}(s) = \frac{\mathbf{V}_o}{\mathbf{V}_i} = \frac{R \parallel 1/sC}{sL + R \parallel 1/sC}, \qquad s = j\omega \qquad \mathbf{Eq. (1)}$$

$$R \left\| \frac{1}{sC} = \frac{R/sC}{R + 1/sC} = \frac{R}{1 + sRC} \right\|$$

#### **Solution Cont....**

Substituting Eq. (2) into Eq. (1) gives

$$\mathbf{H}(s) = \frac{R/(1+sRC)}{sL+R/(1+sRC)} = \frac{R}{s^2RLC+sL+R}, \qquad s=j\omega$$

Or

$$\mathbf{H}(\omega) = \frac{R}{-\omega^2 R L C + j\omega L + R}$$

Since H(0) = 1 and  $H(\infty) = 0$ , we conclude from Table 1 that the circuit is a second-order lowpass filter.

The magnitude of H is

$$H = \frac{R}{\sqrt{(R - \omega^2 R L C)^2 + \omega^2 L^2}} \qquad \text{Eq. (3)}$$

#### **Solution Cont....**

The corner frequency is the same as the half-power frequency, i.e., where H is reduced by a factor of  $1/\sqrt{2}$ 

Eq. (3) becomes after squaring

$$H^{2} = \frac{1}{2} = \frac{R^{2}}{(R - \omega_{c}^{2}RLC)^{2} + \omega_{c}^{2}L^{2}}$$

Or 
$$2 = (1 - \omega_c^2 LC)^2 + \left(\frac{\omega_c L}{R}\right)^2$$

Substituting the values of R, L, and C, we obtain

$$2 = (1 - \omega_c^2 \, 4 \times 10^{-6})^2 + (\omega_c \, 10^{-3})^2$$

#### **Solution Cont....**

Assuming that  $\omega_c$  is in krad/s,

$$2 = (1 - 4\omega_c^2)^2 + \omega_c^2$$

Or

$$16\omega_c^4 - 7\omega_c^2 - 1 = 0$$

Solving the quadratic equation in  $\omega_c^2$ 

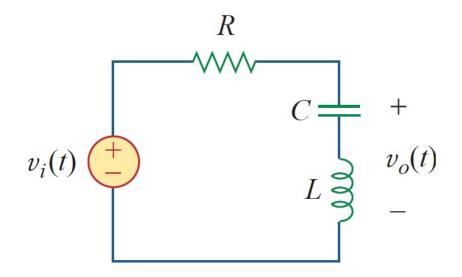
We get 
$$\omega_c^2 = 0.5509$$
 and  $-0.1134$ 

Since  $\omega_c$  is real,

$$\omega_c = 0.742 \text{ krad/s} = 742 \text{ rad/s}$$

#### **Practice Example**

If the bandstop filter shown in Figure is to reject a 200-Hz sinusoid while passing other frequencies, calculate the values of L and C. Take  $R = 150 \Omega$  and the bandwidth as 100 Hz.



**Answer:** 

$$L = 0.2387 \text{ H}, C = 2.653 \mu\text{F}$$

#### References

- 1. R.C. Dorf and James A. Svoboda, "Introduction to Electric Circuits", Chapter 17, 9th ed, John Wiley & Sons, 2013.
- 2. Charles K. Alexander and Matthew N. O. Sadiku, "Fundamentals of Electric Circuits", Chapter 19, 4th ed, Mcgraw Hill, 2009.