AER 1217 – Autonomy for UAVs

Lecture 2 - Quadrotor Dynamics and Control

Prof. Steven Waslander (slides by Prof. Angela Schoellig)



CONTEXT

Today's lecture will be used in the lab next week!



AER1217: Winter 2021 Lab 2 February 3, 2022

AER1217: Development of Unmanned Aerial Vehicles

Lab 2: Quadrotor Simulation and Control Design

1 Introduction

This is the second lab in a series designed to complement the lecture material and help you with the course project. This lab requires the design of a quadrotor position controller, and implementation in ROS. It is set up to work with Gazebo, which simulates your interface with the Parrot AR.Drone UAV and the Vicon motion capture system for position and attitude measurements. To be able to complete this lab within your dedicated time slot, you should have preliminary knowledge of ROS, and Python programming.





MOTIVATION

1. Why quadrotors?

2. Why control?



MOTIVATION

1. Why quadrotors?



WHY QUADROTORS?

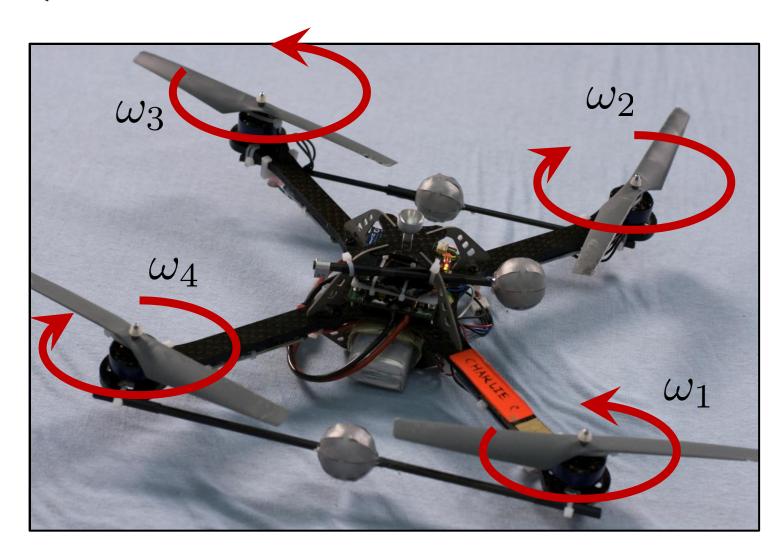


- 1. Rigid frame
- 2. Four independently controlled motors
- 3. Move by changing the motor speeds
- 4. Vertical take-off and landing

- Mechanically simple
- Highly maneuverable



QUADROTOR VEHICLE



Vary the speeds of the rotors to control:

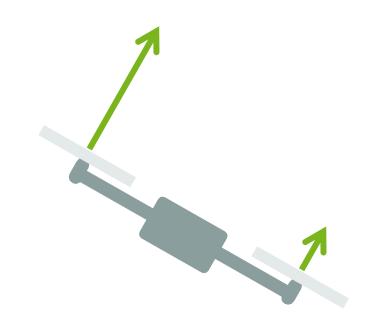
- the position
- and orientation of the robot



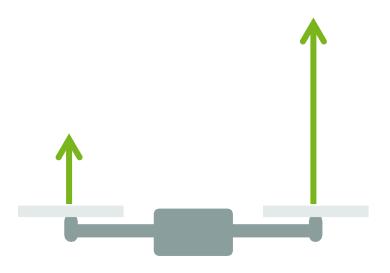
Green arrows represent the force produced by the motor, which is roughly proportional to the squared rotor speed



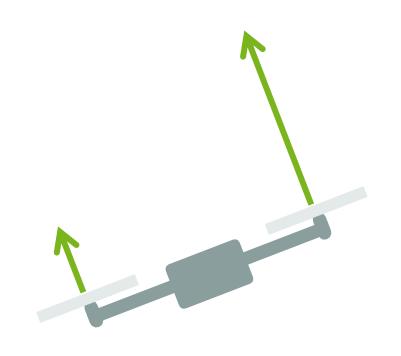














DANCE





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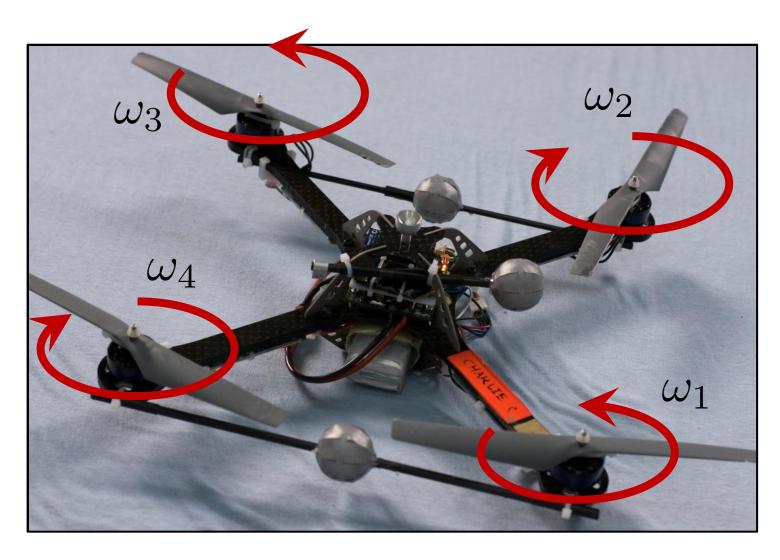
FLIP—up to 1800 degrees/second





youtu.be/bWExDW9J9sA

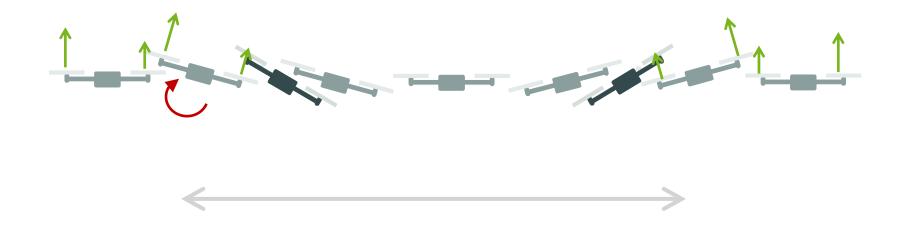
YAW



How do you get the robot to steer/yaw?



TRANSLATION

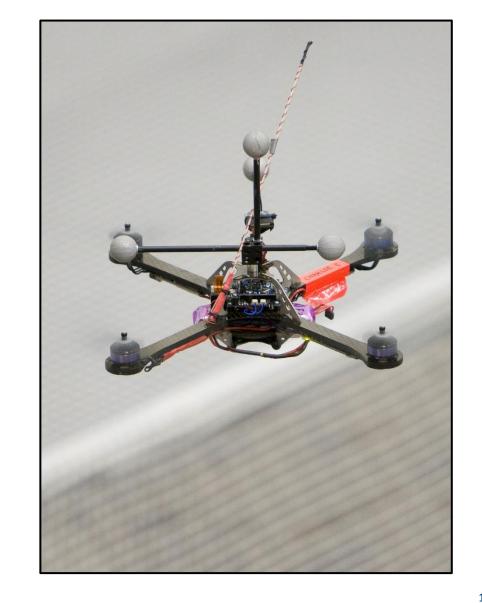




QUESTION

The robot has **6 degrees of freedom**!

- 1. How many **different ways** can you **rotate or translate** the robot?
- 2. How many of them are **independent given** that the robot has **4 motors**?





MOTIVATION

1. Why quadrotors?

2. Why control?



MOTIVATION

2. Why control?



WHAT IS CONTROLS?

Controls enables self-regulating systems

Despite disturbances

That is, machines to achieve a task on their own



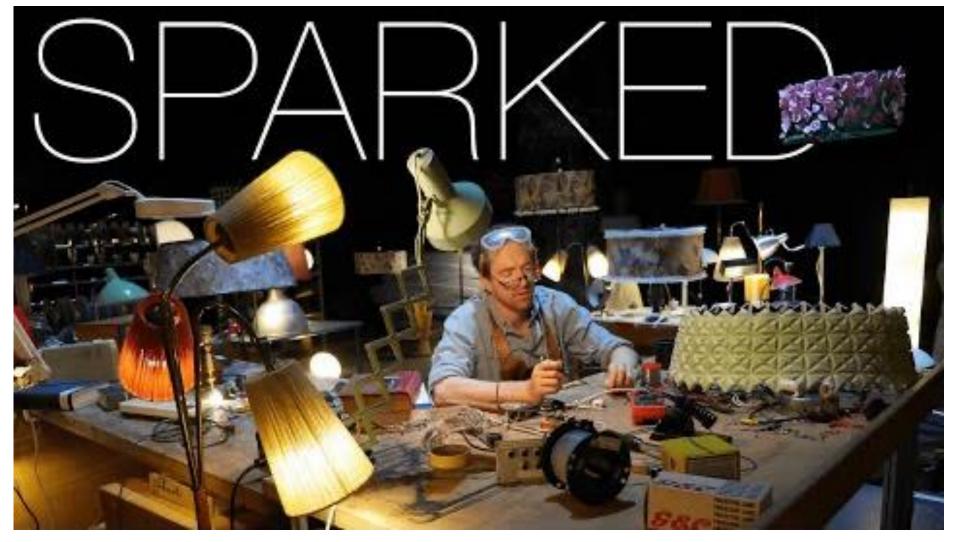
DYNAMIC RESPONSE





youtu.be/nQ2ziVW6kts

SPARKED



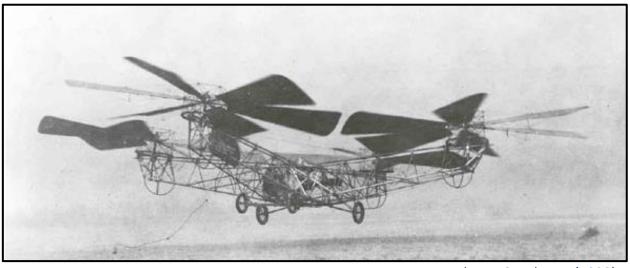


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QUADROTOR HISTORY

First successful flight in 1923

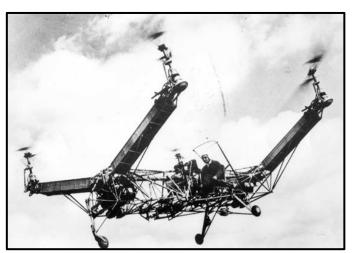
- Design by **George De Bothezat**
- Vertical take-off and landing (VTOL)
- Poor stability, control



De Bothezat Quadrotor (1923)

Later attempts

• Used simple feedback control



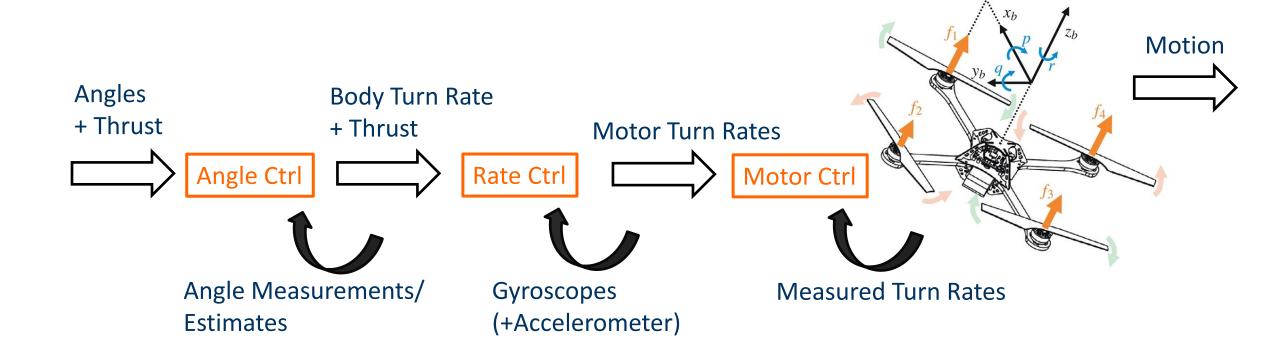
Convertawings Model "A" (1956)



Curtiss-Wright X-19A (1960)

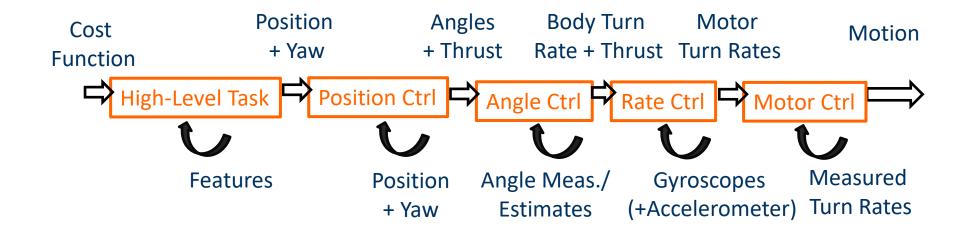


QUADROTOR CONTROL (1 of 2)





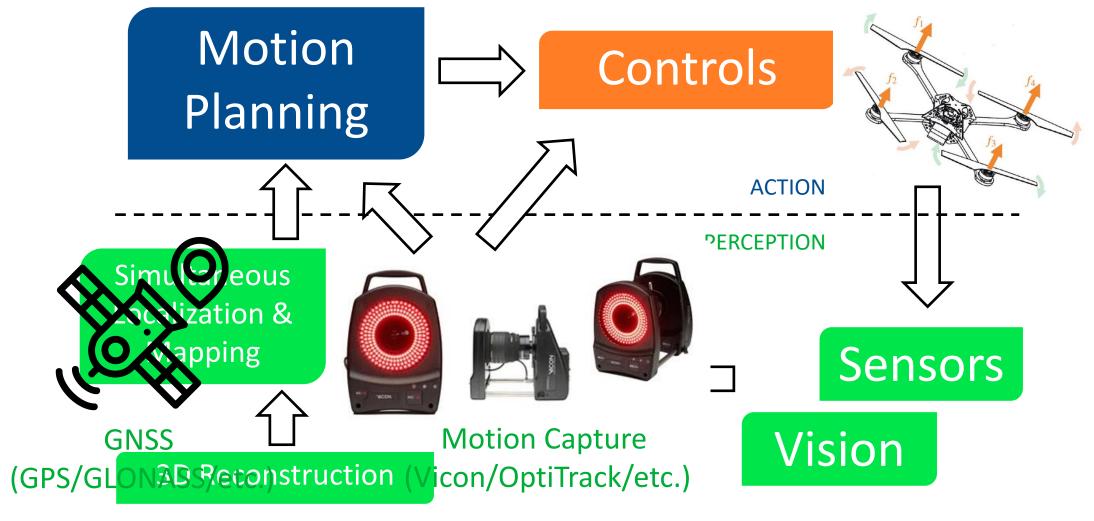
QUADROTOR CONTROL (2 of 2)



Control allows us to focus on the high-level task



HOW DOES IT ALL FIT TOGHETER? AN OVERVIEW





OBJECTIVE

My goal for today:

Prepare you to design your own quadrotor controllers!



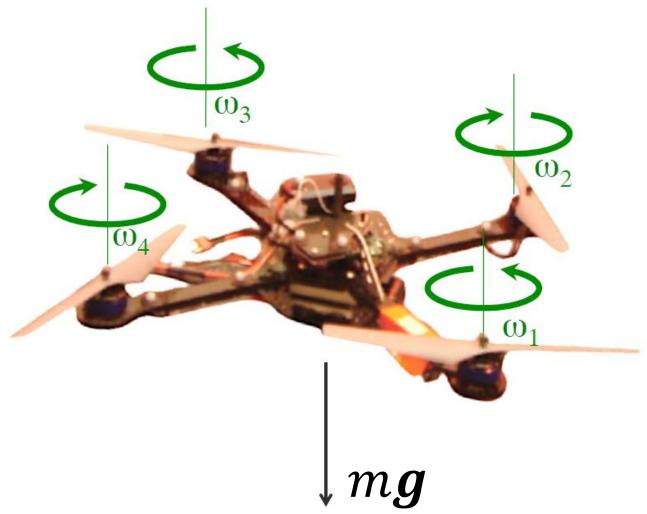
OUTLINE

- 1. Basic Mechanics
- 2. Dynamics & Control of the Vertical Direction (1D)
- 3. Dynamics & Control in the Vertical Plane (2D)
- 4. Trajectory Tracking Control (3D)
- 5. Summary
- 6. What Can Go Wrong?
- 7. Learning-Enabled Control

(Several figures/slides are taken from Vijay Kumar's excellent Coursera course on Aerial Robotics)

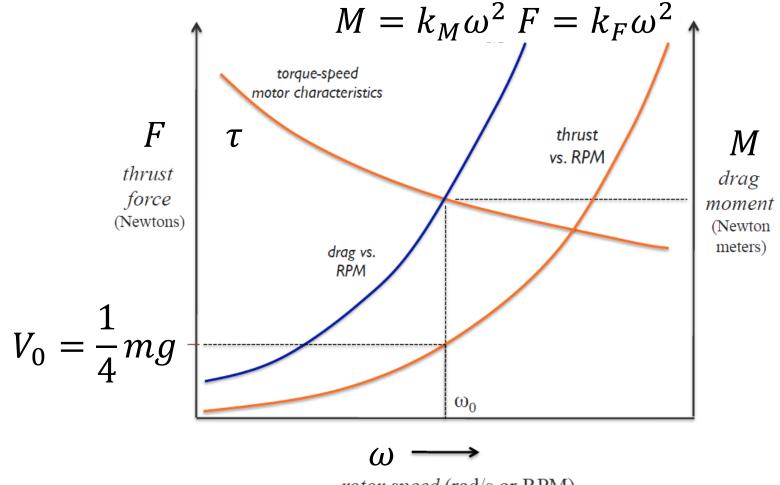


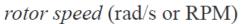
1. BASIC MECHANICS





1. BASIC MECHANICS: ROTOR PHYSICS







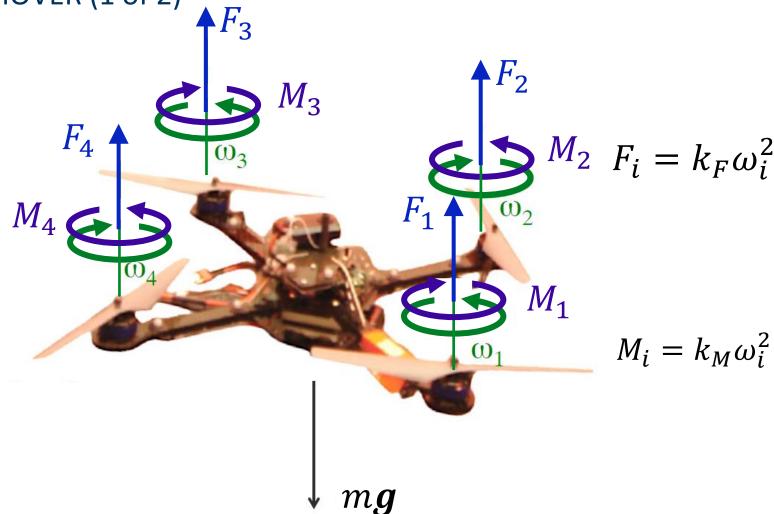


Motor Speeds

$$k_F \omega_i^2 = \frac{1}{4} mg$$

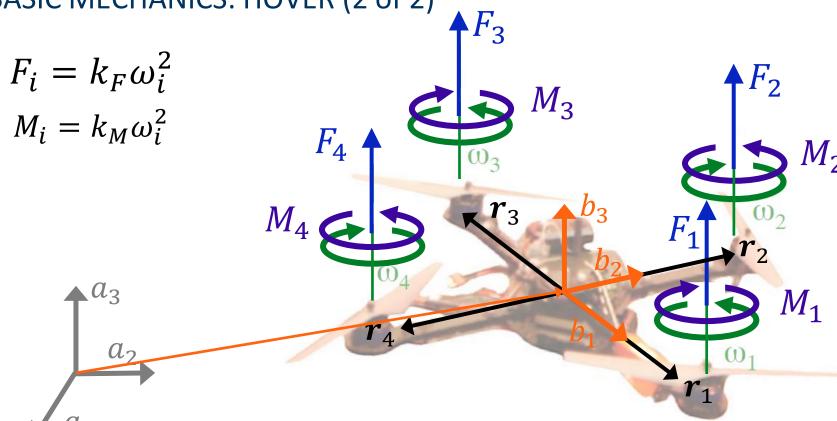
Motor Torques

$$\tau_i = k_M \omega_i^2$$









Resultant Force

$$\mathbf{F} = \mathbf{F}_1 + \mathbf{F}_2 + \mathbf{F}_3 + \mathbf{F}_4 - mg\mathbf{a}_3$$

Resultant Moment

$$M = r_1 \times F_1 + r_2 \times F_2 + r_3 \times F_3 + r_4 \times F_4 + M_1 + M_2 + M_3 + M_4$$



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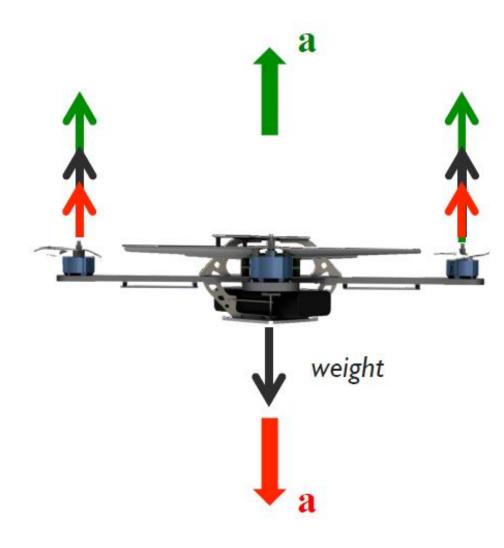
2. DYNAMICS IN THE VERTICAL DIRECTION

Hover:

$$\sum_{i=1}^4 k_F \omega_i^2 + mg = 0$$

Move:

$$\sum_{i=1}^{4} k_F \omega_i^2 + mg = ma$$



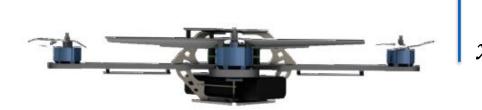
increase motor speeds

motor thrusts

decrease motor speeds



2. CONTROL IN THE VERTICAL DIRECTION (HEIGHT)



vertical displacement

$$\sum_{i=1}^{4} k_F \omega_i^2 + mg = ma \to a = \frac{d^2 x}{dt^2} = \ddot{x}$$

Define the control input as

$$u = \frac{1}{m} \left[\sum_{i=1}^{4} k_F \omega_i^2 + mg \right]$$

to obtain the second order dynamic system $u = \ddot{x}$.

What input drives the robot to the desired position?



2. CONTROL OF A LINEAR, SECOND-ORDER SYSTEM

Problem:

• State, Input $x, u \in \mathbb{R}$

• Plant model $\ddot{x} = u$

Want x to follow the desired trajectory $x^{des}(t)$

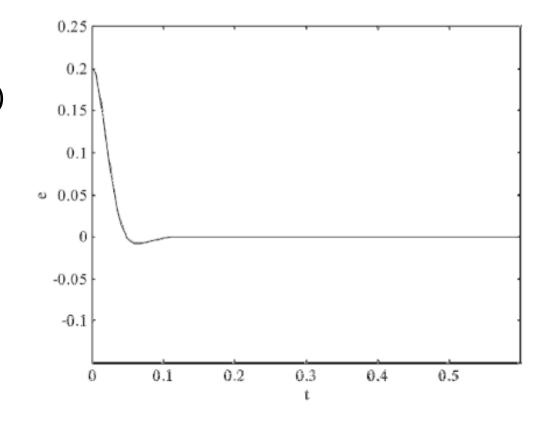
General Approach:

Define error $e(t) = x^{des}(t) - x(t)$ Want e(t) to converge exponentially to zero

Strategy:

Find u such that the error converges exponentially to 0

$$\ddot{e} + K_v \dot{e} + K_p e = 0 \quad K_p, K_v > 0$$



$$u(t) = \ddot{x}^{des}(t) + K_v \dot{e}(t) + K_p e(t)$$
Feedforward Derivative Proportional



2. TRAJECTORY TRACKING CONTROL IN 1D

PD Control

$$u(t) = \ddot{x}^{des}(t) + K_v \dot{e}(t) + K_p e(t)$$

Proportional control acts like a spring (capacitance) response.

Derivative control is a viscous dashpot (resistance) response.

Large derivative gain makes the system overdamped (slow convergence).

PID Control

Often advantageous in the presence of disturbances (e.g. wind) or modeling errors (e.g. unknown mass).

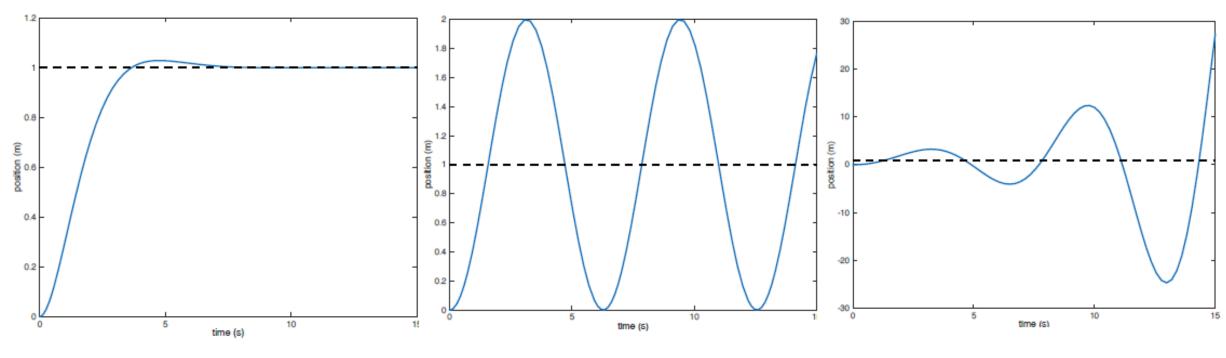
$$u(t) = \ddot{x}^{des}(t) + K_v \dot{e}(t) + K_p e(t) + K_i \int_0^t e(\tau) d\tau$$

Integral control makes the steady-state error go to zero. PID control generates a third-order closed-loop system.





2. EFFECTS OF DIFFERENT GAINS



Stable

$$K_p, K_v > 0$$

Marginally Stable

$$K_p > 0$$
, $K_v = 0$

$$K_v = 0$$

Unstable

$$K_p$$
 or $K_v < 0$



2. DIFFERENT PARAMETRIZATIONS

$$\ddot{x} = u, u(t) \coloneqq K_p(x^{des} - x) + K_v(\dot{x}^{des} - \dot{x}) + \ddot{x}^{des}$$

$$\Rightarrow \ddot{e} + K_v \dot{e} + K_p e = 0$$

$$\Rightarrow \qquad \ddot{e} + 2\zeta\omega_n\dot{e} + \omega_n^2e = 0$$





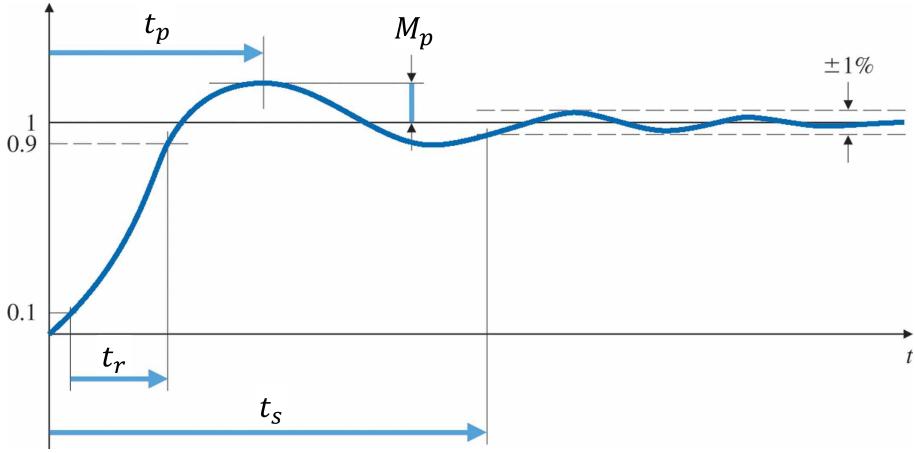
Intuitive parameterization:

- Damping ratio: $\zeta \in [0.7..1]$
- Natural frequency, related to rise time (10-90%): $\omega_n \approx \frac{t_r}{1.8}$
- Settling time: $t_S \approx \frac{4.6}{\zeta \omega_n}$



2. SECOND-ORDER CHARACTERISTICS (1 of 3)

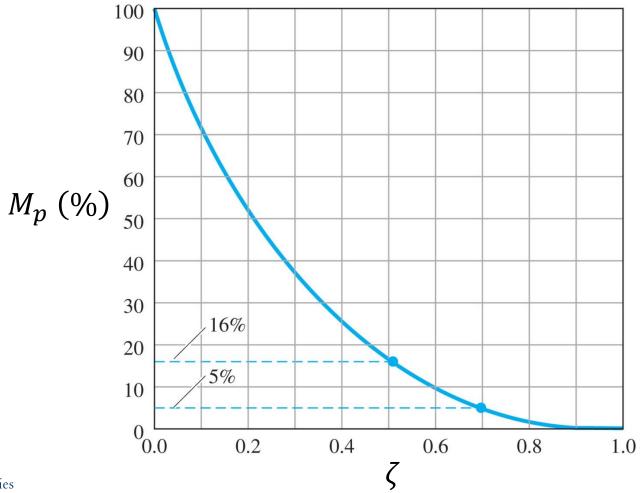
Definition of <u>rise time</u> t_p , <u>peak time</u> t_p , <u>settling time</u> t_s , and <u>overshoot</u> M_p :





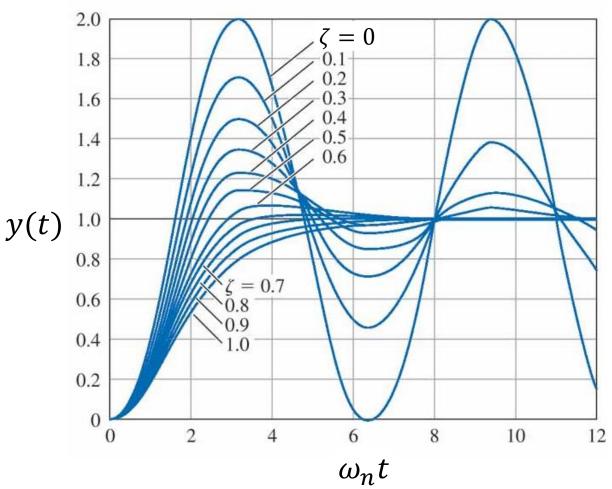
2. SECOND-ORDER CHARACTERISTICS (2 of 3)

Overshoot versus damping ratio for the second-order system:



2. SECOND-ORDER CHARACTERISTICS (3 of 3)

Step responses of second-order systems versus ζ



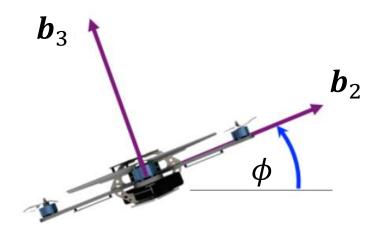


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3. PLANAR (2D) QUADROTOR MODEL



$$\begin{bmatrix} \ddot{y} \\ \ddot{z} \\ \ddot{\phi} \end{bmatrix} = \begin{bmatrix} 0 \\ -g \\ 0 \end{bmatrix} + \begin{bmatrix} -\frac{1}{m}\sin\phi & 0 \\ \frac{1}{m}\sin\phi & 0 \\ 0 & \frac{1}{I_{zz}} \end{bmatrix} \begin{bmatrix} u_1 \\ u_2 \end{bmatrix}$$

State space:
$$x = \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} = \begin{bmatrix} z \\ \phi \\ \dot{y} \\ \dot{z} \\ \dot{\phi} \end{bmatrix}$$

$$\dot{x} = \begin{bmatrix} \dot{y} \\ \dot{z} \\ \dot{\phi} \\ 0 \\ -g \\ 0 \end{bmatrix} + \begin{bmatrix} 0 & 0 \\ 0 & 0 \\ 0 & 0 \\ -\frac{1}{m}\sin\phi & 0 \\ \frac{1}{m}\sin\phi & 0 \\ 0 & \frac{1}{I_{77}} \end{bmatrix} \begin{bmatrix} u_1 \\ u_2 \end{bmatrix}$$



3. LINEARIZED DYNAMIC MODEL

Equations of motion:

$$\ddot{y} = -\frac{u_1}{m}\sin(\phi)$$
 Dynan $\ddot{z} = -g + \frac{u_1}{m}\cos(\phi)$ non $\ddot{\phi} = \frac{u_2}{I_{ZZ}}$

Dynamics are nonlinear

Equilibrium hover configuration:

$$y_0, z_0, \phi_0 = 0, \qquad u_{1,0} = mg,$$

$$u_{1,0} = mg$$

$$u_{2,0} = 0$$

Linearized dynamics around hover:

$$\ddot{y} = -g\phi$$

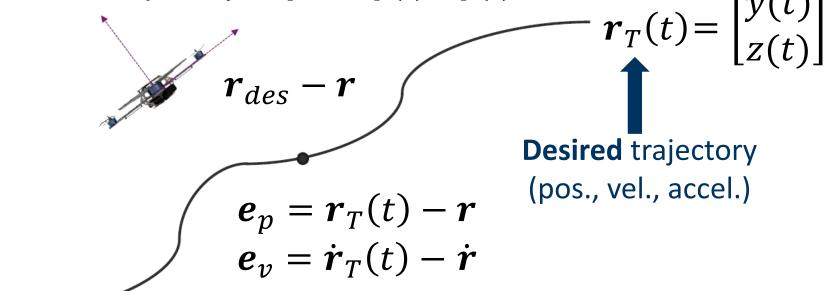
$$\ddot{z} = -g + \frac{u_1}{m}$$

$$\ddot{\phi} = \frac{u_2}{I_{aa}}$$



3. 2D TRAJECTORY TRACKING

Given the differentiable trajectory $r_T(t)$, $\dot{r}_T(t)$, $\dot{r}_T(t)$



Want error to decay exponentially to 0:

$$(\ddot{\boldsymbol{r}}_T(t) - \ddot{\boldsymbol{r}}_c) + K_{d,x}\boldsymbol{e}_v + K_{p,x}\boldsymbol{e}_p = 0$$

Commanded acceleration, calculated by the controller



3. NESTED CONTROL STRUCTURE

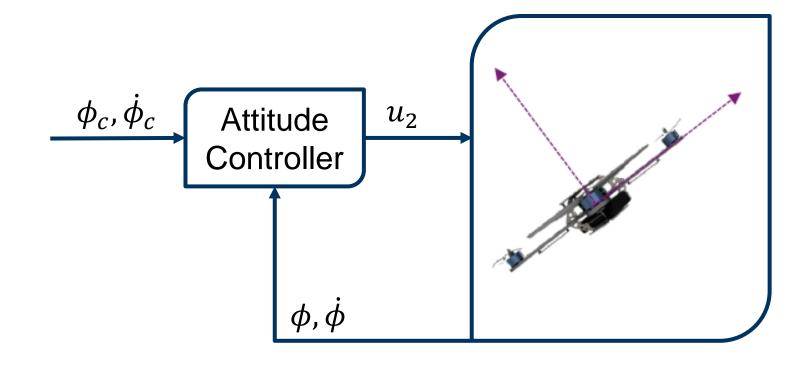
Want error to decay exponentially to 0:

$$(\ddot{\phi_c} - \ddot{\phi}) + k_{p,\phi}(\phi_c - \phi) + k_{d,\phi}(\dot{\phi}_c - \dot{\phi}) = 0$$

Linearized dynamics:

$$\ddot{\phi} = \frac{u_2}{I_{xx}}$$





$$u_2 = I_{xx}(\ddot{\phi}_c + k_{p,\phi}(\phi_c - \phi) + k_{d,\phi}(\dot{\phi}_c - \dot{\phi}))$$

3. NESTED CONTROL STRUCTURE

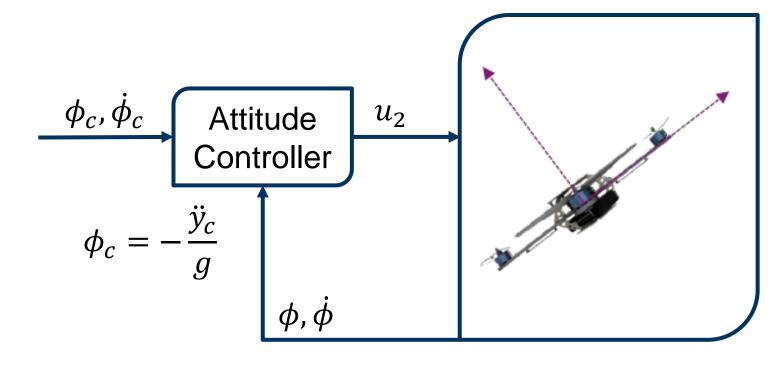
$$\phi_c = -\frac{1}{g}(\ddot{y}_{des} + k_{d,y}(\dot{y}_{des} - \dot{y}) + k_{p,y}(y_{des} - y)$$

Linearized dynamics:

$$\ddot{y} = -g\phi$$

$$\ddot{\phi} = \frac{u_2}{I_{xx}}$$

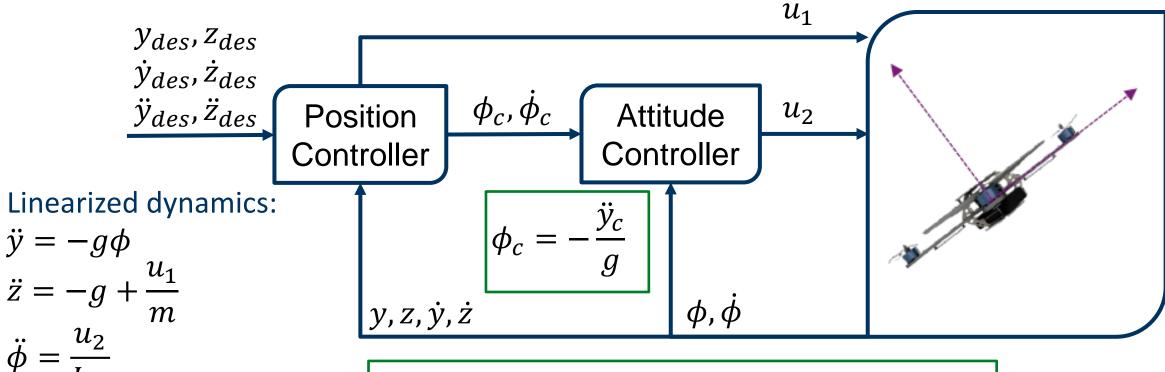




$$u_2 = I_{xx}(\ddot{\phi}_c + k_{p,\phi}(\phi_c - \phi) + k_{d,\phi}(\dot{\phi}_c - \dot{\phi}))$$

3. NESTED CONTROL STRUCTURE

$$u_1 = m(g + \ddot{z}_{des} + k_{d,z}(\dot{z}_{des} - \dot{z}) + k_{p,z}(z_{des} - z))$$



$$u_2 = I_{xx}(\ddot{\phi}_c + k_{p,\phi}(\phi_c - \phi) + k_{d,\phi}(\dot{\phi}_c - \dot{\phi}))$$

3. CONTROL EQUATIONS

$$u_1 = m(g + \ddot{z}_{des} + k_{d,z})(\dot{z}_{des} - \dot{z}) + k_{p,z}(z_{des} - z))$$

Simplification with no performance reduction

$$u_{2} = I_{xx}(\ddot{\phi}_{c} + k_{p,\phi})(\phi_{c} - \phi) + k_{d,\phi}(\dot{\phi}_{c} - \dot{\phi})$$

$$u_{2} = k_{p,\phi}(\phi_{c} - \phi) + k_{d,\phi}(\dot{\phi}_{c} - \dot{\phi})$$

$$\phi_c = -\frac{1}{g} (\ddot{y}_{des} + k_{d,y}) (\dot{y}_{des} - \dot{y}) + k_{p,y} (y_{des} - y)$$



3. CONTROL EQUATIONS

$$u_{1} = m(g + \ddot{z}_{des} + k_{d,z})(\dot{z}_{des} - \dot{z}) + k_{p,z}(z_{des} - z))$$

$$u_{2} = k_{p,\phi}(\phi_{c} - \phi) + k_{d,\phi}(\dot{\phi}_{c} - \dot{\phi})$$

$$\phi_{c} = -\frac{1}{g}(\ddot{y}_{des} + k_{d,y})(\dot{y}_{des} - \dot{y}) + k_{p,y}(y_{des} - y)$$



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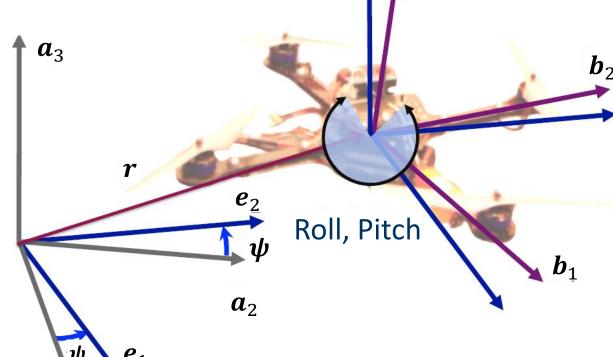


4. 3D VEHICLE MODEL

$q = \begin{bmatrix} x \\ y \\ z \\ \phi \\ \theta \\ \psi \end{bmatrix} \begin{array}{c} r \\ \text{Pitch} , x = \begin{bmatrix} q \\ \dot{q} \end{bmatrix} \\ \text{Roll} \\ \text{Yaw} \end{array}$

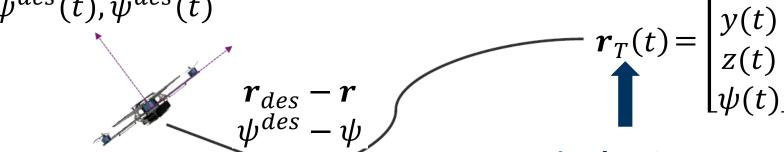
Angular velocities components in b

$$\begin{bmatrix} p \\ q \\ r \end{bmatrix} = \begin{bmatrix} c\theta & 0 & -c\phi s\theta \\ 0 & 1 & s\phi \\ s\theta & 0 & c\phi c\theta \end{bmatrix} \begin{bmatrix} \dot{\phi} \\ \dot{\theta} \\ \dot{\psi} \end{bmatrix}$$



4. CONTROL TASK

Given $r_T(t)$, $\dot{r}_T(t)$, $\ddot{r}_T(t)$ (differentiable) $\mathbf{r}^{des}(t), \dot{\mathbf{r}}^{des}(t), \ddot{\mathbf{r}}^{des}(t)$ $\psi^{des}(t), \dot{\psi}^{des}(t), \ddot{\psi}^{des}(t)$



 $e_p = r_T(t) - r$ $e_v = \dot{r}_T(t) - \dot{r}$

$$\mathbf{e}_{v} = \dot{\mathbf{r}}_{T}(t) - \dot{\mathbf{r}}$$

Desired trajectory (pos., yaw)

Want error to decay exponentially to 0:

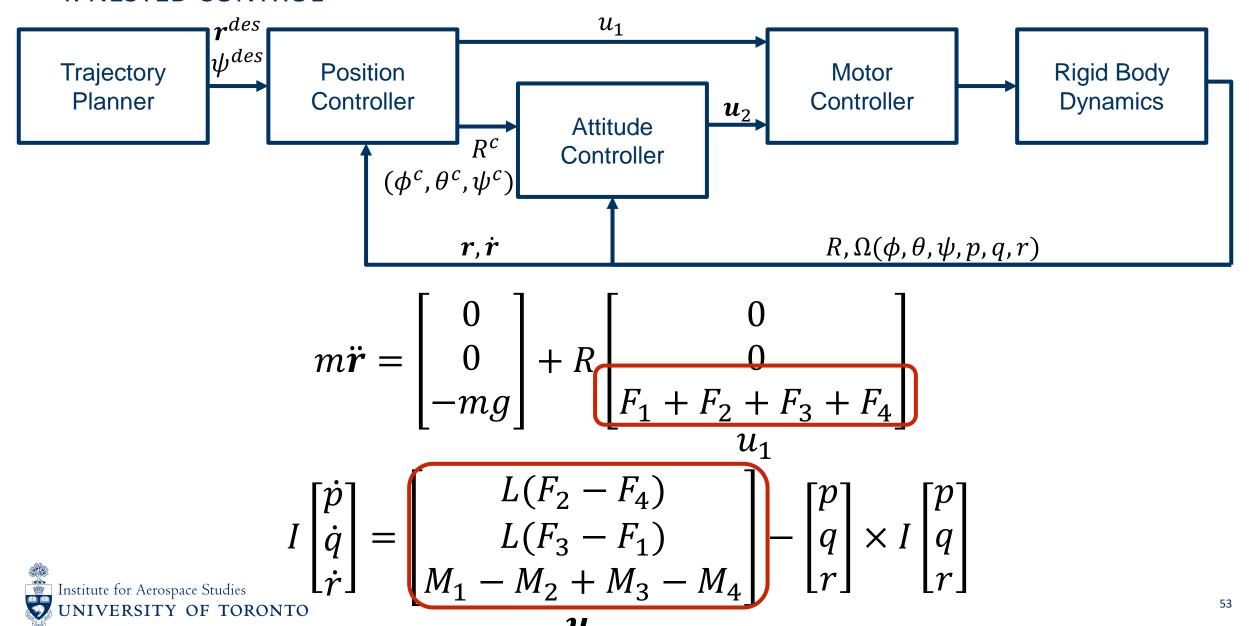
$$(\ddot{\boldsymbol{r}}_T(t) - \ddot{\boldsymbol{r}}_c) + K_{d,x}\boldsymbol{e}_v + K_{p,x}\boldsymbol{e}_p = 0$$



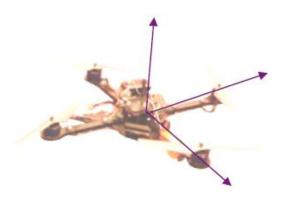
Commanded acceleration, calculated by the controller



4. NESTED CONTROL



4. HOVER CONTROL (1 of 5)



Linearize the dynamics at the hover configuration

$$(u_1 \sim mg, \theta \sim 0, \phi \sim 0, \psi \sim \psi_0)$$

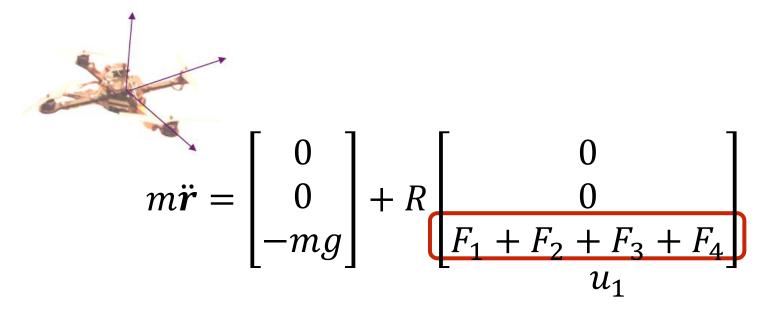
 $(u_2 \sim 0, p \sim 0, q \sim 0, r \sim 0)$

$$m\ddot{r} = \begin{bmatrix} 0 \\ 0 \\ -mg \end{bmatrix} + R \begin{bmatrix} 0 \\ 0 \\ F_1 + F_2 + F_3 + F_4 \end{bmatrix}$$

$$I \begin{bmatrix} \dot{p} \\ \dot{q} \\ \dot{r} \end{bmatrix} = \begin{bmatrix} L(F_2 - F_4) \\ L(F_3 - F_1) \\ M_1 - M_2 + M_3 - M_4 \end{bmatrix} - \begin{bmatrix} p \\ q \\ r \end{bmatrix} \times I \begin{bmatrix} p \\ q \\ r \end{bmatrix}$$



4. HOVER CONTROL (2 of 5)



Linearization

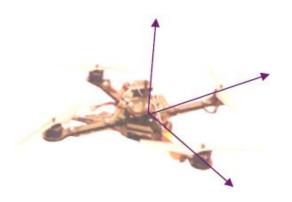
$$(u_1 \sim mg, \theta \sim 0, \phi \sim 0, \psi \sim \psi_0)$$

$$\ddot{r}_1 = \ddot{x} = g(\theta \cos \psi + \phi \sin \psi)$$

$$\ddot{r}_2 = \ddot{y} = g(\theta \sin \psi - \phi \cos \psi)$$

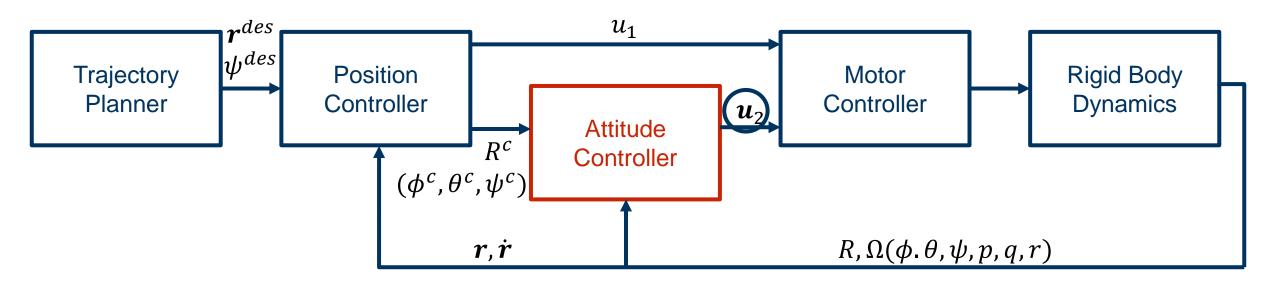


4. HOVER CONTROL (3 of 5)





4. HOVER CONTROL (4 of 5)



$$\mathbf{u}_{2} = \begin{bmatrix} k_{p,\phi}(\phi_{c} - \phi) + k_{d,\phi}(p_{c} - p) \\ k_{p,\theta}(\theta_{c} - \theta) + k_{d,\theta}(q_{c} - q) \\ k_{p,\psi}(\psi_{c} - \psi) + k_{d,\psi}(r_{c} - r) \end{bmatrix}$$

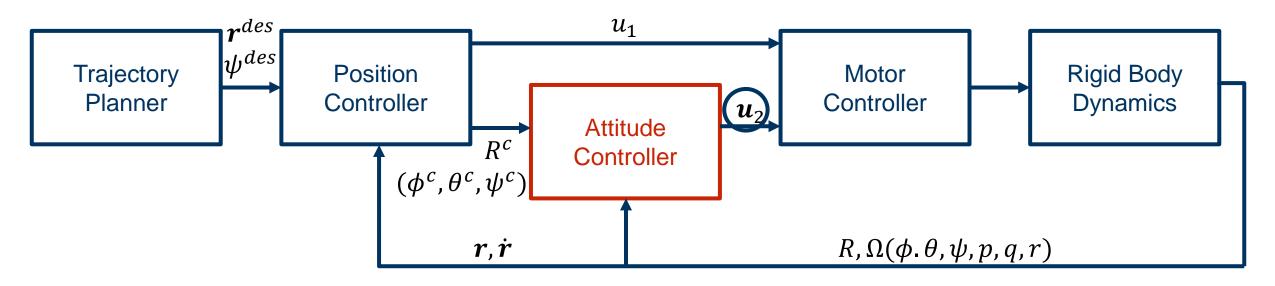
Using linearized dynamics:

$$\ddot{r}_1 = \ddot{x} = g(\theta \cos \psi + \phi \sin \psi)$$

$$\ddot{r}_2 = \ddot{y} = g(\theta \sin \psi - \phi \cos \psi)$$



4. HOVER CONTROL (4 of 5)



$$\mathbf{u}_{2} = \begin{bmatrix} k_{p,\phi}(\phi_{c} - \phi) + k_{d,\phi}(p_{c} - p) \\ k_{p,\theta}(\theta_{c} - \theta) + k_{d,\theta}(q_{c} - q) \\ k_{p,\psi}(\psi_{c} - \psi) + k_{d,\psi}(r_{c} - r) \end{bmatrix}$$

Using linearized dynamics

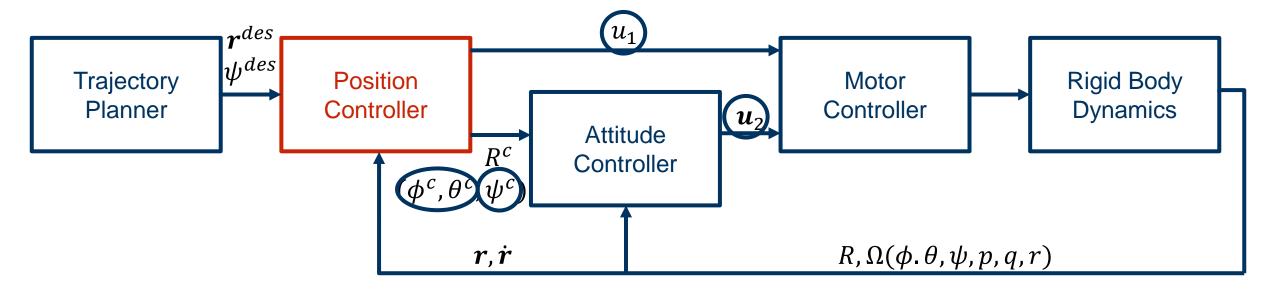
$$\phi_c = \frac{1}{g} (\ddot{r}_{1,c} \sin \psi_{des} - \ddot{r}_{2,c} \cos \psi_{des})$$

$$\theta_c = \frac{1}{g} (\ddot{r}_{1,c} \cos \psi_{des} - \ddot{r}_{2,c} \sin \psi_{des})$$

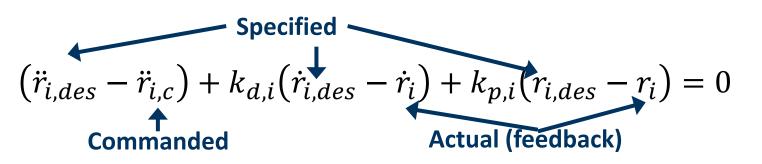
$$\psi_c = \psi^{des}$$



4. HOVER CONTROL (5 of 5)



$$u_1 = m(g + \ddot{r}_{3,c})$$

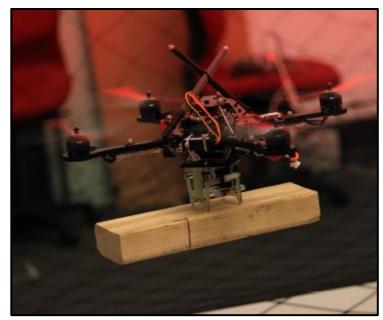




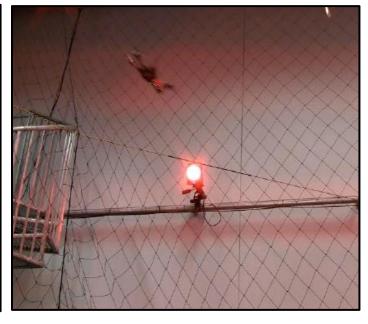
Lupashin, Sergei, et al. "A platform for aerial robotics research and demonstration: The flying machine arena." *Mechatronics* 24.1 (2014): 41-54.

4. LIMITATIONS OF LINEAR CONTROL

Assumptions: roll and pitch angles, and all velocities are close to zero









4. NONLINEAR MODEL (1 of 2)

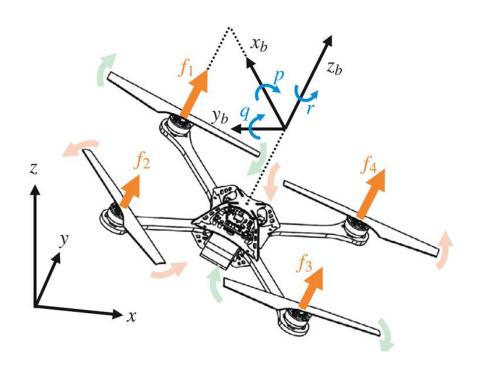
Newton's equations of motion

$$\begin{bmatrix} \ddot{x} \\ \ddot{y} \\ \ddot{z} \end{bmatrix} = \begin{bmatrix} 0 \\ 0 \\ -g \end{bmatrix} + R \begin{bmatrix} 0 \\ 0 \\ c \end{bmatrix}$$

where

$$c = \frac{(F_1 + F_2 + F_3 + F_4)}{m}$$

is mass-normalized.



Maps from body to inertial frame



4. NONLINEAR MODEL (2 of 2)

Newton's equations of motion

$$\begin{bmatrix} \ddot{x} \\ \ddot{y} \\ \ddot{z} \end{bmatrix} = \begin{bmatrix} 0 \\ 0 \\ -g \end{bmatrix} + R \begin{bmatrix} 0 \\ 0 \\ c \end{bmatrix}$$

Euler's equations of motion

$$\begin{bmatrix} \ddot{x} \\ \ddot{y} \\ \ddot{z} \end{bmatrix} = \begin{bmatrix} 0 \\ 0 \\ -g \end{bmatrix} + R \begin{bmatrix} 0 \\ 0 \\ c \end{bmatrix} \qquad I \begin{bmatrix} \dot{p} \\ \dot{q} \\ \dot{r} \end{bmatrix} = \begin{bmatrix} L(F_2 - F_4) \\ L(F_3 - F_1) \\ \kappa(F_1 - F_2 + F_3 - F_4) \end{bmatrix} - \begin{bmatrix} p \\ q \\ r \end{bmatrix} \times I \begin{bmatrix} p \\ q \\ r \end{bmatrix}$$

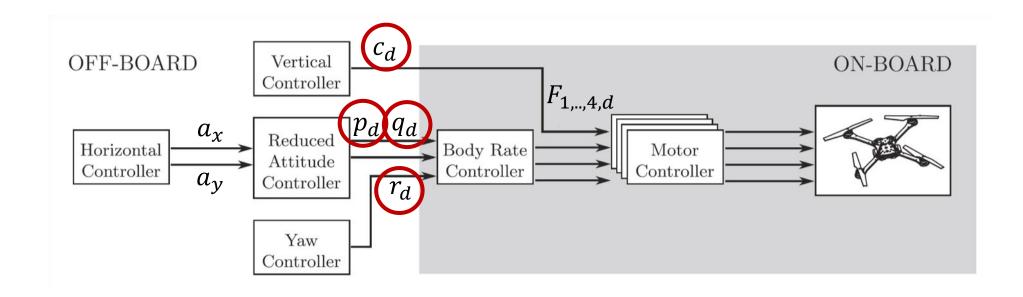
where

$$c = \frac{(F_1 + F_2 + F_3 + F_4)}{m}$$

is mass-normalized.



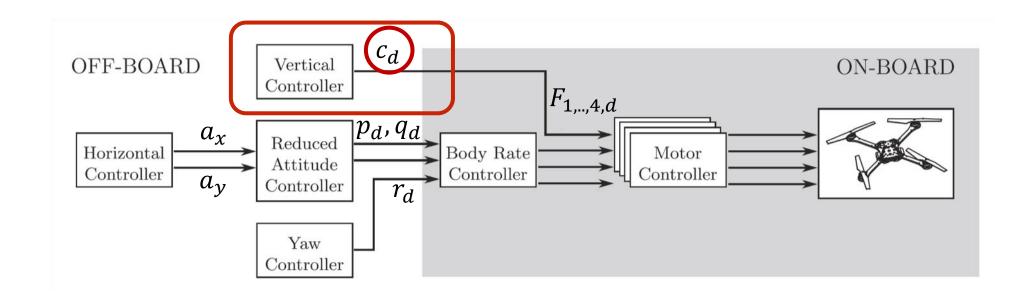
4. CONTROL DESIGN OVERVIEW



We need to calculate four inputs for the on-board controllers



4. PART 1: VERTICAL CONTROL





4. PART 1: VERTICAL CONTROL

Recall the equation of motion:

$$\begin{bmatrix} \ddot{x} \\ \ddot{y} \\ \ddot{z} \end{bmatrix} = \begin{bmatrix} 0 \\ 0 \\ -g \end{bmatrix} + R \begin{bmatrix} 0 \\ 0 \\ c \end{bmatrix}$$

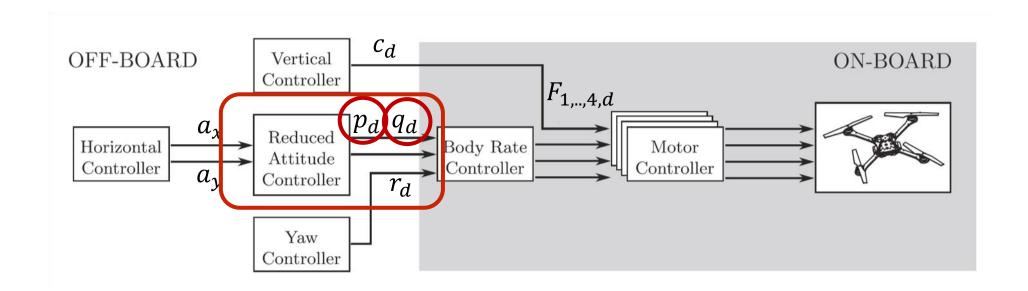
Solve for *c*:

$$c = \frac{1}{R_{33}} (\ddot{z} + g)$$

Substitute controller for exponentially decaying error:

$$c_d = \frac{1}{R_{33}} \left(\omega_{n,z}^2 (z_d - z) + 2\zeta_z \omega_{n,z} (\dot{z}_d - \dot{z}) + \ddot{z}_d + g \right)$$

4. PART 2: LATERAL CONTROL





4. PART 2: LATERAL CONTROL

Horizontal controller to determine accelerations:

$$\ddot{x} = \omega_{n,x}^{2}(x_{d} - x) + 2\zeta_{x}\omega_{n,x}(\dot{x}_{d} - \dot{x}) + \ddot{x}_{d}$$

$$\ddot{y} = \omega_{n,y}^{2}(y_{d} - y) + 2\zeta_{y}\omega_{n,y}(\dot{y}_{d} - \dot{y}) + \ddot{y}_{d}$$

From the equations of motion:

$$\begin{bmatrix} \ddot{x} \\ \ddot{y} \end{bmatrix} = \begin{bmatrix} R_{13,d} \\ R_{23,d} \end{bmatrix} c_d$$

Rotational kinematics:

$$\dot{\mathbf{R}} = \mathbf{R} \begin{bmatrix} 0 & -r & q \\ r & 0 & -p \\ -q & p & 0 \end{bmatrix}$$

Shape reaction of matrix entries as a first-order system:

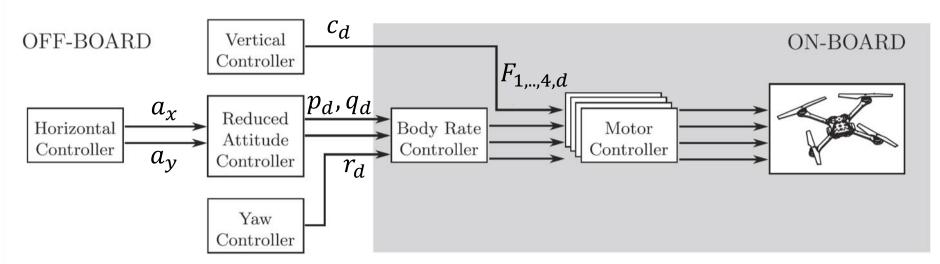
$$\dot{R}_{13,d} = \frac{1}{\tau_{13,d}} \left(R_{13,d} - R_{13} \right) \qquad \dot{R}_{23,d} = \frac{1}{\tau_{23,d}} \left(R_{23,d} - R_{23} \right)$$

Use rotational kinematics and rate of change calculated above:

$$\begin{bmatrix} p_d \\ q_d \end{bmatrix} = \frac{1}{R_{33}} \begin{bmatrix} R_{21} - R_{11} \\ R_{22} - R_{12} \end{bmatrix} \begin{bmatrix} \dot{R}_{13,d} \\ \dot{R}_{23,d} \end{bmatrix}$$



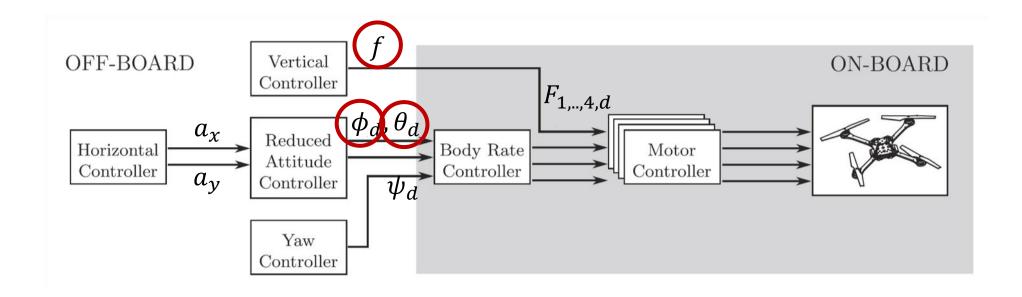
4. OVERALL CONTROL



Body rate controller:
$$I \begin{bmatrix} \frac{1}{\tau_{p}} (p_{d} - p) \\ \frac{1}{\tau_{q}} (q_{d} - q) \\ \frac{1}{\tau_{r}} (r_{d} - r) \end{bmatrix} + \begin{bmatrix} p \\ q \\ r \end{bmatrix} \times I \begin{bmatrix} p \\ q \\ r \end{bmatrix} = \begin{bmatrix} l(F_{2,d} - F_{4,d}) \\ l(F_{3,d} - F_{1,d}) \\ \kappa(F_{1,d} - F_{2,d} + F_{3,d} - F_{4,d}) \end{bmatrix}$$



$$(F_{1,d} - F_{2,d} + F_{3,d} - F_{4,d}) = mc_d$$



We need to calculate three inputs for the on-board controllers (yaw controller is optional)



Newton's equations of motion

Euler's equations of motion

$$\begin{bmatrix} \ddot{x} \\ \ddot{y} \\ \ddot{z} \end{bmatrix} = \begin{bmatrix} 0 \\ 0 \\ -g \end{bmatrix} + R \begin{bmatrix} 0 \\ 0 \\ c \end{bmatrix} \qquad \mathbf{I} \begin{bmatrix} \dot{p} \\ \dot{q} \\ \dot{r} \end{bmatrix} = \begin{bmatrix} L(F_2 - F_4) \\ L(F_3 - F_1) \\ \kappa(F_1 - F_2 + F_3 - F_4) \end{bmatrix} - \begin{bmatrix} p \\ q \\ r \end{bmatrix} \times \mathbf{I} \begin{bmatrix} p \\ q \\ r \end{bmatrix}$$

Using $\psi = 0$ to simplify the rotation matrix:

$$R = \begin{bmatrix} . & . & \cos \phi \sin \theta \\ . & . & \sin \phi \\ . & . & \cos \theta \cos \phi \end{bmatrix}$$

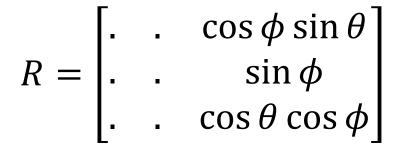


Newton's equations of motion

$$\begin{bmatrix} \ddot{x} \\ \ddot{y} \\ \ddot{z} \end{bmatrix} = \begin{bmatrix} 0 \\ 0 \\ -g \end{bmatrix} + R \begin{bmatrix} 0 \\ 0 \\ c \end{bmatrix}$$

Hence:

$$\ddot{x} = f \cos \phi \sin \theta$$
$$\ddot{y} = -f \sin \phi$$
$$\ddot{z} = \cos \theta \cos \phi$$



Solve for angles and force:

$$\phi = \sin^{-1}\left(\frac{-\ddot{y}}{f}\right)$$

$$\theta = \sin^{-1}\left(\frac{-\ddot{x}}{f\cos\phi}\right)$$

$$f = \frac{\ddot{z} + g}{\cos\theta\cos\phi}$$



Use measurements to calculate the current mass-normalized force:

$$f = \frac{\ddot{z} + g}{\cos\theta\cos\phi}$$

Horizontal controller to determine accelerations:

$$\ddot{x} = \omega_{n,x}^{2}(x_{d} - x) + 2\zeta_{x}\omega_{n,x}(\dot{x}_{d} - \dot{x})$$

$$\ddot{y} = \omega_{n,y}^{2}(y_{d} - y) + 2\zeta_{y}\omega_{n,y}(\dot{y}_{d} - \dot{y})$$

Calculate commanded roll and pitch:

$$\phi_c = \sin^{-1}\left(\frac{-\ddot{y}}{f}\right)$$

$$\theta_c = \sin^{-1}\left(\frac{-\ddot{x}}{f\cos\phi_c}\right)$$



4. OTHER APPROACHES

Linear controller based on linearized system

- Hoffmann, Gabriel, et al. "Quadrotor helicopter flight dynamics and control: Theory and experiment." AIAA guidance, navigation and control conference and exhibit. 2007.
- Gurdan, Daniel, et al. "Energy-efficient autonomous four-rotor flying robot controlled at 1 kHz." Proceedings 2007 IEEE International Conference on Robotics and Automation. IEEE, 2007.
- Bouabdallah, Samir. Design and control of quadrotors with application to autonomous flying. No. THESIS. Epfl, 2007.

LQR

• Cowling, Ian D., James F. Whidborne, and Alastair K. Cooke. "Optimal trajectory planning and LQR control for a quadrotor UAV." International Conference on Control. 2006.

Feedback Linearization / Backstepping

- Madani, Tarek, and Abdelaziz Benallegue. "Backstepping control for a quadrotor helicopter." 2006 IEEE/RSJ International Conference on Intelligent Robots and Systems. IEEE, 2006.
- Lee, Daewon, H. Jin Kim, and Shankar Sastry. "Feedback linearization vs. adaptive sliding mode control for a quadrotor helicopter." *International Journal of control, Automation and systems* 7 (2009): 419-428.

• Exact linearization, differentially flat system

 Mellinger, Daniel, and Vijay Kumar. "Minimum snap trajectory generation and control for quadrotors." 2011 IEEE international conference on robotics and automation. IEEE, 2011.

• L1 adaptive control

Hovakimyan, Naira, and Chengyu Cao. £1 Adaptive Control Theory: Guaranteed Robustness with Fast Adaptation. Society for Industrial
and Applied Mathematics, 2010.



OUTLINE

- 1. Basic Mechanics
- 2. Dynamics & Control of the Vertical Direction (1D)
- 3. Dynamics & Control in the Vertical Plane (2D)
- 4. Trajectory Tracking Control (3D)
- 5. Summary
- 6. What Can Go Wrong?
- 7. Learning-Enabled Control



6. SUMMARY

- 1. Quadrotor dynamics are nonlinear and coupled
- 2. Four independent degrees of freedom
- 3. Controller approach:
 - A. Nested controller
 - B. "Transform" individual loops into double integrators
 - C. Design the closed-loop behavior to resemble a second-order system behavior
 - D. Two intuitive parameters to choose for each second-order system
- 4. There are many more advanced controllers including adaptive and learning controls





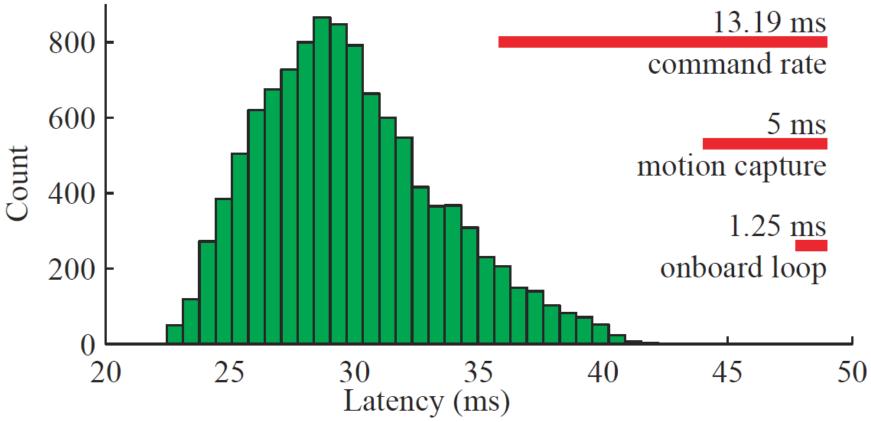
OUTLINE

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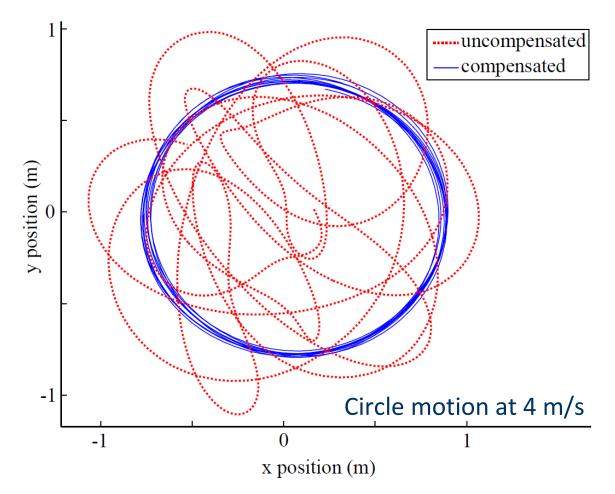
4. LIMITATIONS (from doi.org/10.1016/j.mechatronics.2013.11.006)

• Latency (1 of 3)





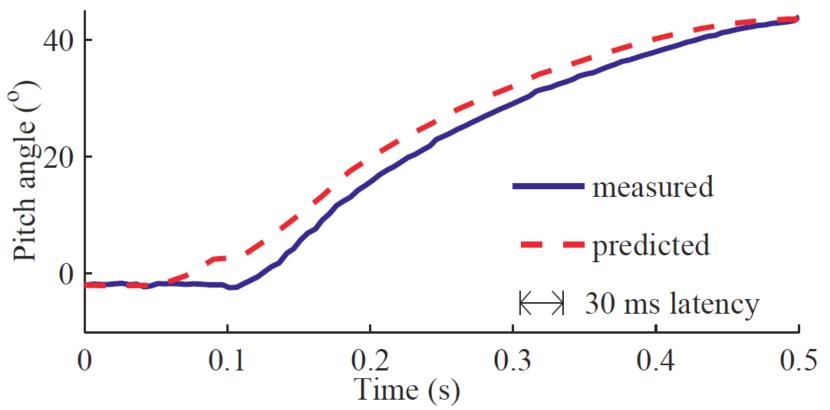
• Latency (2 of 3)





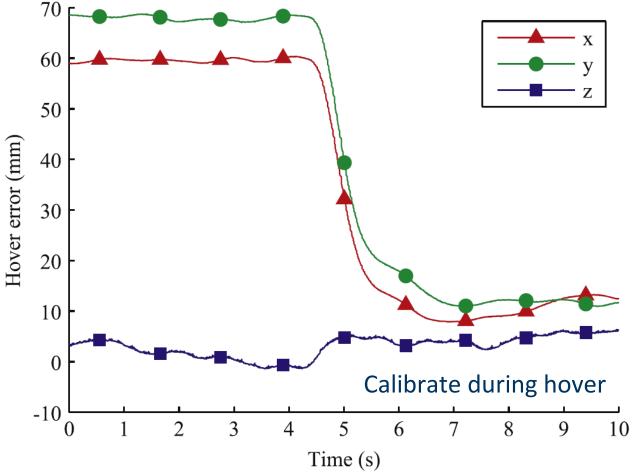
• Latency (3 of 3)

Predict the vehicle position at the time the input arrives at the vehicle





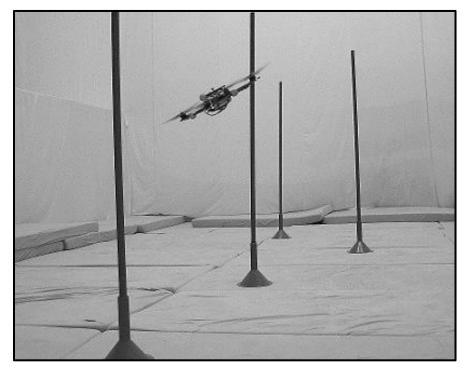
Offsets

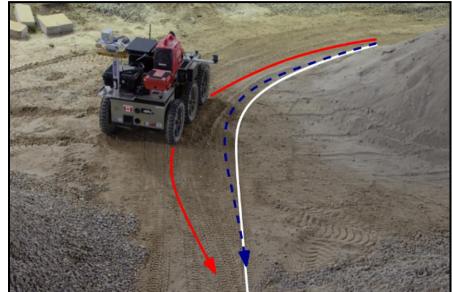




Aggressive Maneuvers

- 1. Triple flip with a quadrotor
- 2. Time-optimized slalom
- 3. Fast path following with a ground vehicle
- 4. ..







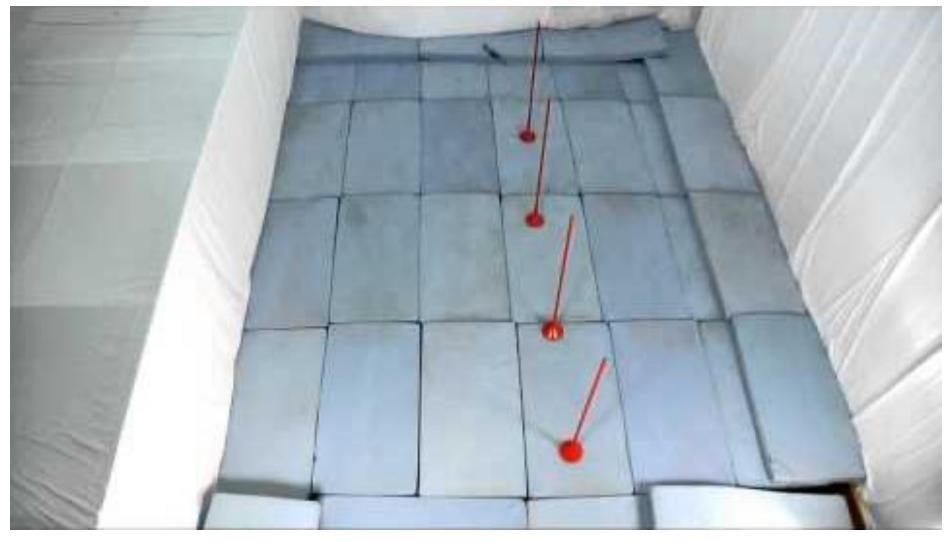
4. TRIPLE ADAPTIVE FLIPS





youtu.be/bWExDW9J9sA

4. SLALOM LEARNING





youtu.be/zHTCsSkmADo

4. VISUAL TEACH AND REPEAT





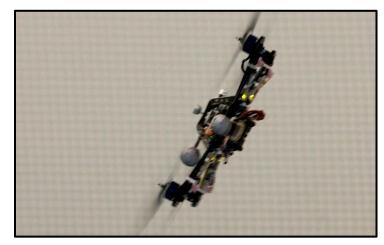
youtu.be/08_d1HSPADA

4. EXPLANATION

- 1. Unmodelled dynamics
- 2. Unknown external disturbances
 - (e.g., environment conditions such as wind, unknown payload, topography or weather)

Model inaccuracies limit achievable performance!









OUTLINE

- 1. Basic Mechanics
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- 4. Trajectory Tracking Control (3D)
 - What Can Go Wrong?
- 5. <u>Learning-Enabled Control</u>
- 6. Summary



5. LEARNING-ENABLE CONTROL: RATIONALE

Learning and adaptation enable

- safe, high-performance motions
- in uncontrolled, unknown or changing environments



5. RESEARCH FOCUS

DYNAMICS CONTROLS MACHINE LEARNING ROBOTICS

Prior information

- Which motions are feasible?
- How to plan collision-free motions?

- + Current sensor measurement
- How to guide the vehicle along a desired path?
- + Past experiment data
- Can the performance be improved by leveraging past data?

Towards robotics applications



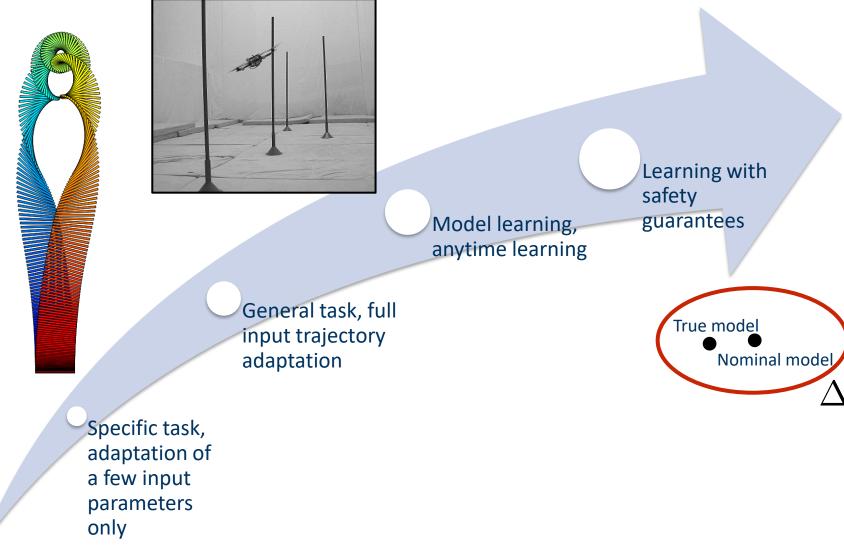
4. LEARNING-BASED ROBUST CONTROL





youtu.be/YqhLnCm0KXY

5. DEVELOPMENT PROCESS





5. CONCLUDING REMARKS

- Safe, reliable, high-performance flight
- of <u>single</u> or <u>multiple aerial vehicles</u>
- in changing and <u>unknown conditions</u>
 using a combination of a-priori <u>models and data</u>





UPCOMING LECTURES

The next two lectures will be labs:

- 1. Quadrotor simulation and position control design
- 2. From quadrotor simulation to experiment.







THANK YOU, Angela!

For follow-up discussions, please contact:

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• email: <u>schoellig@utias.utoronto.ca</u>

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