

0.5071 - 0.29651

Applications of DSP

Project 2

Question 1:

Part 1 a

Given,

- I. $H(j\omega)$ is a 4th order low pass Chebyshev function
- II. Ap=0.5 dB
- III. f1=2khz
- IV. fs=10kHz

obtain $H_D(z)$.

Solution: Using low pass Chebyshev prototype that is available in MATLAB (cheb1ap) in the impulse invariant method we obtain the following output in the console window:

```
-----Answer to question 1-----
1a) Hd(z)when f1=2KHz fs=10KHz
ans =
  -9.095e-17 s^3 + 0.09261 s^2 + 0.2321 s + 0.04396
   s^4 - 1.479 s^3 + 1.46 s^2 - 0.8131 s + 0.2221
Continuous-time transfer function.
poles(1)
z1 =
  1.0e+15 *
   1.0183
   -0.0000
   -0.0000
p1 =
  0.2323 + 0.7679i
  0.2323 - 0.76791
  0.5071 + 0.2965i
```

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Part 1 b:

Poles of $H_D(z)$

In Impulse invariant method, the transfer function in the analog domain is given by:

$$HA(s) = \sum_{i \models 1}^{N} (\frac{Ai}{s - pi})$$

$$HA(s) \leftrightarrow hA(t)$$

Now, $h_D(t)=h_A(nT)$, where, T is the sampling period.

$$HD(z) = \sum_{i=1}^{N} Ai * \frac{z}{z - e^{PiT}}$$

We also know that, If our poles are inside the unit circle then our system is stable. From the console of the MATLAB we get The Magnitude of the poles:

______1b____ the magnitude of each pole

0.8022

0.8022

0.5874

0.5874

Also we get the Angles of the poles:

the angle of each pole

A1_1 =

73.1702

A1_2 =

-73.1702

A1_3 =

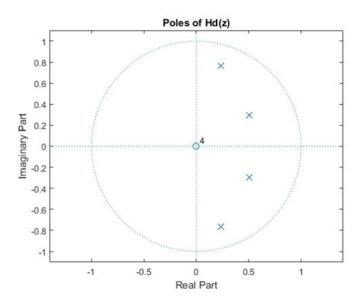
30.3081

A1_4 =

-30.3081

Here, |Zp| max is 0.8022.

Pole Zero locations are mapped below:

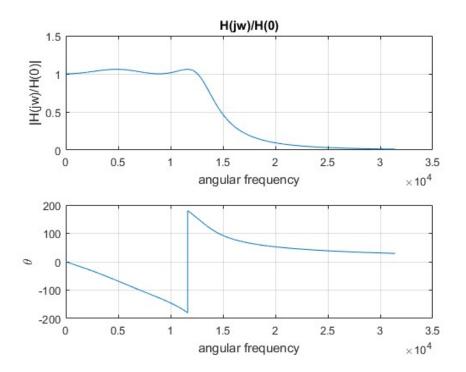


As all the poles are inside the unit circle, the system is stable.

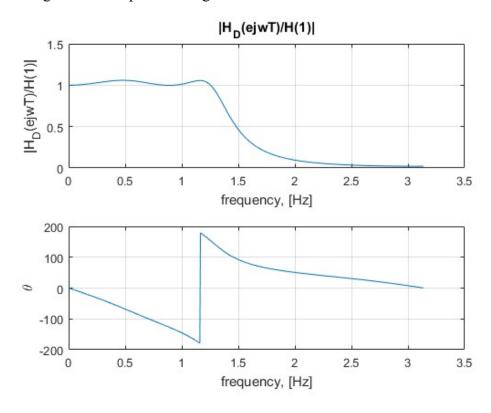
Part 1 c:

Here we need to plot The magnitude and phase of $\frac{H(jw)}{H(0)}$.

The Analog filter response is as given below:



The digital filter response is as given below:



Question 2:

Here, we repeat question 1 for f1=2kHz and fs=20KHz. Then we compare |Zp|max in (2) and (1) as well as the approximation accuracy of $\frac{Hde^{jwt}}{Hd(1)}$ relative to $\frac{H(jw)}{H(0)}$, $|w| \le \frac{ws}{2}$

```
Console output:
------Answer to question 2-----
2a____Hd(z)when f1=2KHz fs=20KHz____
ans =
 -4.547e-17 s^3 + 0.007521 s^2 + 0.02436 s + 0.005167
    s^4 - 2.918 s^3 + 3.518 s^2 - 2.032 s + 0.4713
Continuous-time transfer function.
poles(2)
z3 =
  1.0e+14 *
   1.6538
  -0.0000
  -0.0000
p3 =
  0.7192 + 0.5338i
  0.7192 - 0.5338i
  0.7398 + 0.2004i
  0.7398 - 0.2004i
```

Part 2 b:

The Magnitude of the poles found from the console window:

The Angle of the poles found from the console window:

```
the angle of each pole

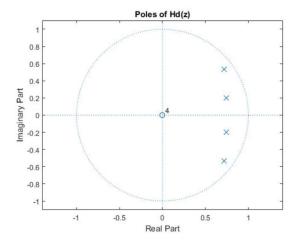
A2_1 =
    36.5851

A2_2 =
    -36.5851

A2_3 =
    15.1541

A2_4 =
    -15.1541
```

The pole-zero plot is as follows:



As all the poles are inside the unit circle and hence system is stable.

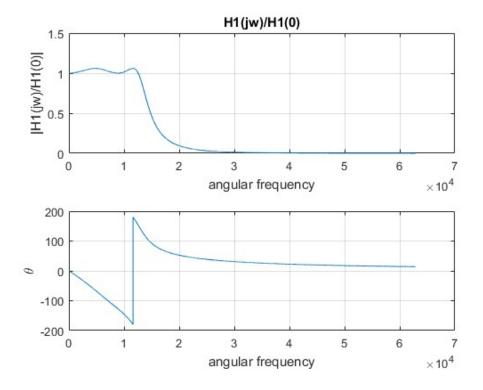
Part 2 c:

|Zp|max valus from the console window:

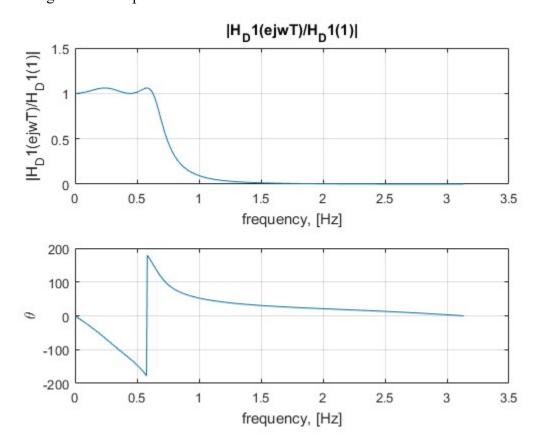
Now comparing the |Zp|max in both the questions, we see that we have P1=0.8022 and P2=0.8957. So the |Zp|max in part 2 is larger.

Coming to the part where we have to compare the approximation accuracy of $H_D(e^{j\omega t})/H_D(1)$ relative to $H(j\omega)/H(0)$.

The Analog filter response is:



The Digital filter response is:



Question 3:

Here, we repeat question 1 with f1=3KHz and fs=15KHz. The MATLAB console output showing the poles and zeros for this question is as follows:

```
----- 3-----Answer to question 3-----
3a_____ Hd(z)when f1=3KHz fs=15KHz _____
ans =
 -6.063e-17 s^3 + 0.09261 s^2 + 0.2321 s + 0.04396
  s^4 - 1.479 s^3 + 1.46 s^2 - 0.8131 s + 0.2221
Continuous-time transfer function.
poles(3)
z5 =
  1.0e+15 *
   1.5274
  -0.0000
  -0.0000
p5 =
  0.2323 + 0.7679i
  0.2323 - 0.76791
  0.5071 + 0.2965i
  0.5071 - 0.2965i
```

The Magnitude of the poles found from the console window:

```
the magnitude of each pole

p3_1 =
     0.8022

p3_2 =
     0.8022

p3_3 =
     0.5874

p3_4 =
     0.5874
```

The Angle of the poles found from the console window:

```
the angle of each pole

A3_1 =

73.1702

A3_2 =

-73.1702

A3_3 =

30.3081

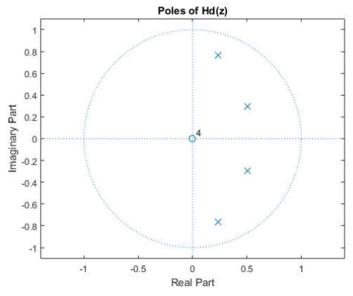
A3_4 =

-30.3081
```

|Zp|max valus from the console window:

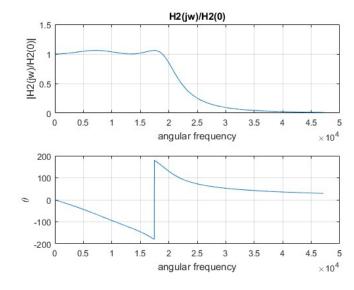
Zplmax =
 0.8022
Zp3max =
 0.8022

Comparing problem 1.b and this one, we get |Zp|max for 1.b = |Zp|max for 3. Because that the ratio of the passband frequency is equal to that of the sampling frequency, pole locations in question 1 and question 3 are same.

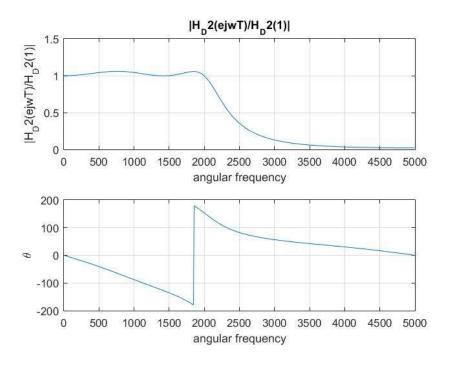


Since all the poles are inside the unit circle, the system is stable.

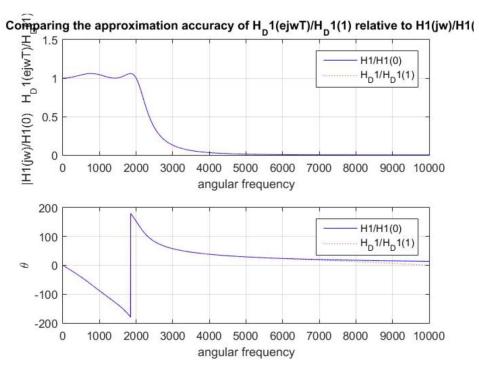
The Analog filter response is as follows:



The Digital filter response is as follows:



Comparing the approximation accuracy



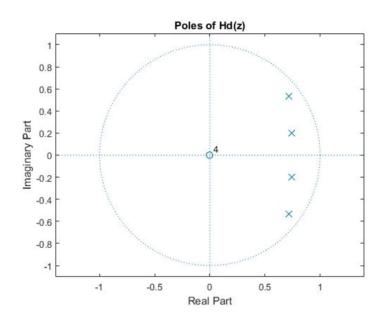
Question 4:

Here, we repeat question 2 with f1=4KHz and fs=40KHz.

The console output is as follows:

```
-----Answer to question 4-----
         Hd(z)for f1=4KHz fs=40KHz
ans =
 2.274e-17 s^3 + 0.007521 s^2 + 0.02436 s + 0.005167
   s^4 - 2.918 \ s^3 + 3.518 \ s^2 - 2.032 \ s + 0.4713
Continuous-time transfer function.
poles(4)
z7 =
  1.0e+14 *
  -3.3076
  -0.0000
  -0.0000
  0.7192 + 0.5338i
  0.7192 - 0.5338i
  0.7398 + 0.2004i
  0.7398 - 0.2004i
```

The pole locations are following:



Since all the poles are inside the unit circle, the system is stable.

The Magnitude of the poles found from the console window:

```
the magnitude of each pole

p4_1 =
      0.8957

p4_2 =
      0.8957

p4_3 =
      0.7664

p4_4 =
      0.7664
```

The Angle of the poles found from the console window:

```
the angle of each pole

A4_1 =

36.5851

A4_2 =

-36.5851

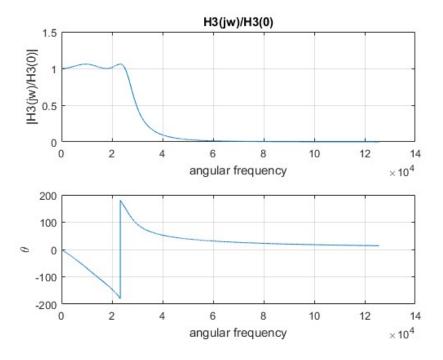
A4_3 =

15.1541

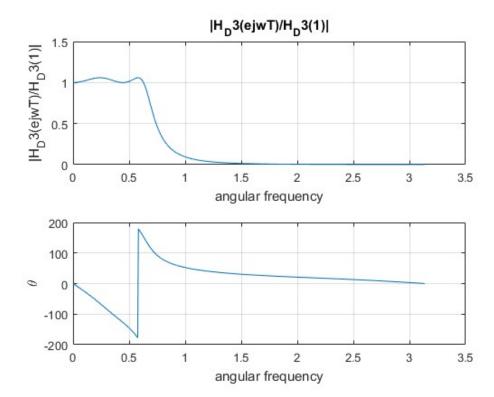
A4_4 =

-15.1541
```

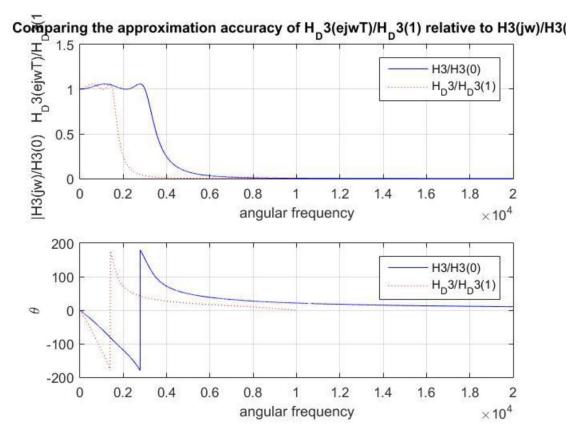
The Analog filter response is as follows:



The Digital filter response is as follows:



Comparing the approximation accuracy of $HDe^{j\omega t}/H_D(1)$ relative to $H(J\omega)/H(0)$:



Question 5:

Here, we repeat question 1 using bilinear transformation method. We compute the new value of fl of HD(z) analytically.

Bilinear Transformation is used to map a transfer function from the analog to the digital domain. This maps from s-plane to z-plane by using the following transformation of s.

$$s = \frac{2}{T} \frac{(z-1)}{(z+1)}$$
$$\omega c = 2\pi f s = 1.4 \times 10^4 rad/sec$$

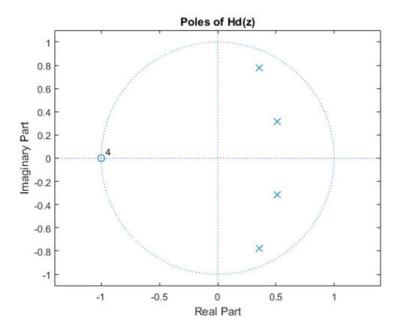
$$\Omega c = \frac{2}{T} tan^{-1} \left(\frac{\omega cT}{2} \right)$$
$$= 1.2 \times 10^{4} \, rad/sec$$

So
$$f_1=1.9 \text{ kHz}$$

From MATLAB Console:

Continuous-time transfer function.

The pole locations are following:



The Magnitude of the poles found from the console window:

```
the magnitude of each pole

p5_1 =
     0.8552

p5_2 =
     0.8552

p5_3 =
     0.6033

p5_4 =
     0.6033
```

The Angle of the poles found from the console window:

```
The angle of each pole

A5_1 =

65.5688

A5_2 =

-65.5688

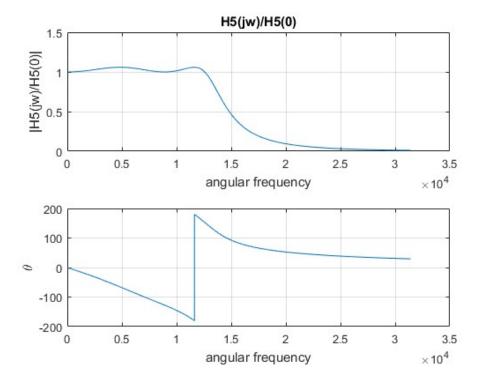
A5_3 =

31.6162

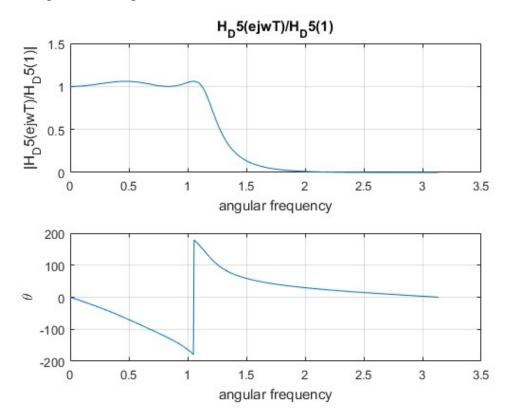
A5_4 =

-31.6162
```

The Analog filter response is as follows:



The Digital filter response is as follows:



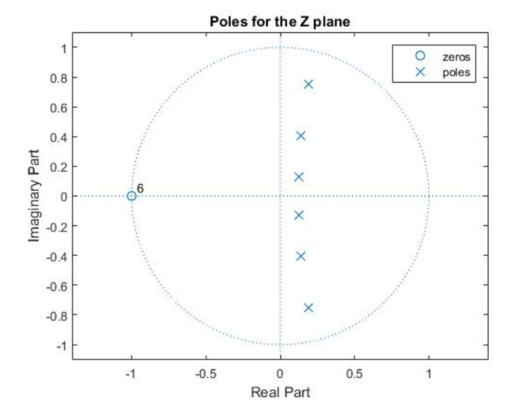
Question 6:

Here, We repeat question 1 for a 6th order HP Butterworth filter $H(j\omega)$ where $\omega s = 4\omega_{3dB}$. We compute 3dB frequency for digital filter and compare with graph.

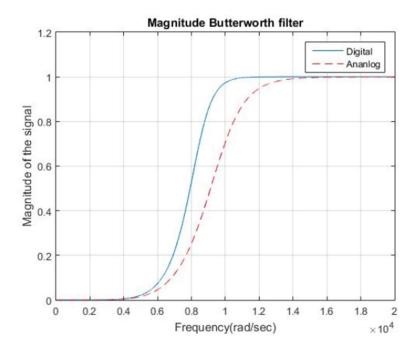
The console output is as follows:

```
 \begin{array}{l} \text{H\_bilinear} = \\ & (-4.524\text{e-}7 + 1.524\text{i}) / ((6366.0 * (2.0 * z - 2.0)) / (z + 1.0) + (7071.0 + 7071.0\text{i})) + \\ & (-2.285 - 1.319\text{i}) / ((6366.0 * (2.0 * z - 2.0)) / (z + 1.0) + (9659.0 - 2588.0\text{i})) + \\ & (-4.524\text{e-}7 - 1.524\text{i}) / ((6366.0 * (2.0 * z - 2.0)) / (z + 1.0) + (7071.0 - 7071.0\text{i})) + \\ & (0.3536 - 0.2041\text{i}) / ((6366.0 * (2.0 * z - 2.0)) / (z + 1.0) + (2588.0 - 9659.0\text{i})) + \\ & (-2.285 + 1.319\text{i}) / ((6366.0 * (2.0 * z - 2.0)) / (z + 1.0) + (9659.0 + 2588.0\text{i})) + \\ & (0.3536 + 0.2041\text{i}) / ((6366.0 * (2.0 * z - 2.0)) / (z + 1.0) + (2588.0 + 9659.0\text{i})) \\ & \text{cutoff frequency: } 8477.00 \text{ rad/seconds} \\ & \text{analytical cutoff frequency: } 8476.89 \text{ rad/seconds} \\ \end{array}
```

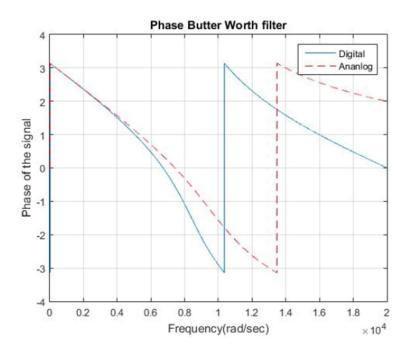
The pole zero plot is like the following:



Plot of the magnitude of the Butterworth filter:



Plot of the phase of the Butterworth filter:

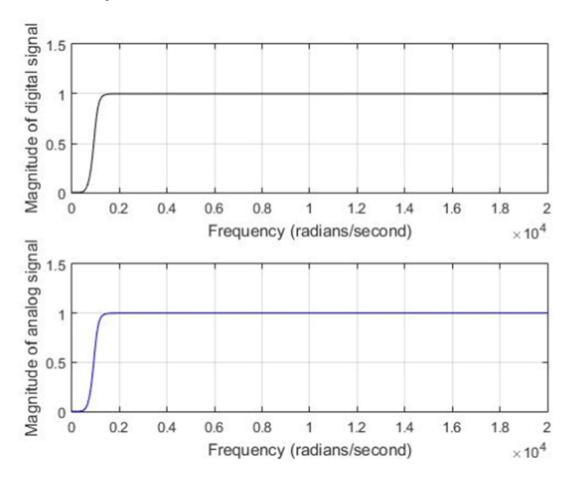


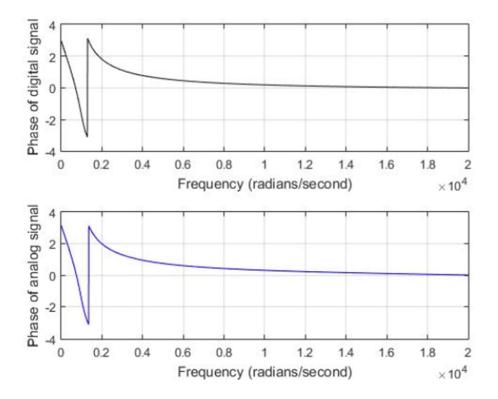
Question 7:

Here, we repeat 6 by prewarping the prototype to get the exact 3-dB frequency of the digital filter.

The continuous time filter frequency $\omega a=0$ corresponds to the discrete time filter frequency $\omega d=0$. The continuous time filter frequency $\omega a=\pm\infty$ corresponds to the discrete time frequency $\omega d=\pm(\pi T)$. This is called *frequency warping*. The continuous-time filter can be designed to compensate for this frequency warping by setting $\omega a=2T\tan\omega dT2$ for every frequency specification that the designer has control over (such as corner frequency or center frequency). This is called *pre-warping* the filter design.

The MATLAB output is as follows:





Conclusion:

In this project we have worked on Chebyshev and Butterworth filters using MATLAB. Using Bilinear transformation method, we calculated the new value of f1. Then we pewarped to gettttt 3-db frequency of the digital filter.

Reference:

- 1. https://en.wikipedia.org/wiki/Butterworth filter
- 2. https://www.mathworks.com/
- 3. https://en.wikipedia.org/wiki/Chebyshev filter
- 4. https://en.wikipedia.org/wiki/Bilinear transform
- 5. John G. Proakis, Dimitri K. Manolakis, Digital Signal Processing, Pearson, 4th edition

The following is the codes and plots of the entire project in MATLAB.

MATLAB Code

```
% ----- Applications of DSP -----
% -----Project 2 ------
%-----Submitted by-----
%-----Rajib Dey -----
clc
close all
clear all
f1=2000;
fs=10000;
fsp=0:0.1:fs/2;
w=2*pi*fsp;
[z,p,k]=cheb1ap(4,0.5);% low pass filter prototype
[b,a] = zp2tf(z,p,k); % Convert to transfer function form
[b,a] = lp2lp(b,a,2*pi*f1);% analog low pass filter to low pass filter of
specified frequency
[bz,az]=impinvar(b,a,fs); % using impulse invarient method
disp('-----')
disp ('1a) Hd(z)when f1=2KHz fs=10KHz ')
tf(bz,az)
disp ('poles(1)')
[z1,p1,k1] = tf2zp(bz,az)
disp ('
                   1b
disp ('the magnitude of each pole')
p1 1 = abs(p1(1))
p1^{2} = abs(p1(2))
p1_3 = abs(p1(3))
p1 \ 4 = abs(p1(4))
zplane(z,p)
disp ('the angle of each pole')
A1 1= angle(p1(1))*57.2958
A1^{2} = angle (p1(2)) *57.2958
A1_3 = angle(p1(3)) *57.2958
A1 4=angle(p1(4))*57.2958
%1(c) magnitude and phase plot for |H(jw)/H(0)|
H = freqs(b,a,w);
figure(11);
subplot(2,1,1);
plot(w, abs(H/H(1)));
grid on;
xlabel('angular frequency')
ylabel('|H(jw)/H(0)|')
title('H(jw)/H(0)')
```

```
subplot(2,1,2);
plot(w, angle(H/H(1))*180/pi);
xlabel('angular frequency')
ylabel('\theta')
grid on;
% magnitude and phase plot for |(H\ D)/(H\ D(1)))
[H D, w1] = freqz(bz, az);
figure (12);
subplot(2,1,1);
plot(w1, abs((H D)/(H D(1))));
xlabel('frequency, [Hz]')
ylabel('|H D(ejwT)/H(1)|')
title('|H D(ejwT)/H(1)|')
grid on;
subplot(2,1,2);
plot(w1, angle((H D)/(H D(1)))*180/pi);
xlabel('frequency, [Hz]')
ylabel('\theta')
grid on;
%comparing the approximation accuracy of H D(ejwT)/H(0) and H(jw)/H(0)
figure (100);
subplot(2,1,1);
plot(w, abs(H/H(1)), 'b-');
ylabel('|H(jw)/H(0)
                    H D(ejwT)/H D(1)|')
xlabel('angular frequency')
title('Comparing the approximation accuracy of H D(ejwT)/H D(1) relative to
H(jw)/H(0)')
hold on;
plot(w1, abs(H D/H D(1)), 'r:');
legend('H/H(0)','HD/HD(1)')
grid on;
subplot(2,1,2);
plot(w, angle(H/H(1))*180/pi, 'b-');
hold on;
plot(w1, angle(H D/H D(1))*180/pi, 'r:');
ylabel('\theta')
xlabel('angular frequency')
legend('H/H(0)','HD/HD(1)')
grid on;
fs1=20000;
fsp1=0:1:fs1/2;
w1=2*pi*fsp1;
[z2,p2,k2] = cheb1ap(4,0.5); % low pass filter prototype
[b1,a1] = zp2tf(z2,p2,k2); % % Convert to transfer function form
[b1,a1] = lp2lp(b1,a1,2*pi*f1); % analog low pass filter to low pass filter of
specified frequency
[bz1,az1]=impinvar(b1,a1,fs1); wsing impulse invarient method
```

```
disp('-----')
disp ('2a Hd(z) when f1=2KHz fs=20KHz ')
tf(bz1,az1)
disp ('poles(2)')
[z3,p3,k3] = tf2zp(bz1,az1)
disp (' 2b
                              ')
disp ('the magnitude of each pole')
p2 1 = abs(p3(1))
p2 2 = abs(p3(2))
p2_3 = abs(p3(3))
p2 \ 4 = abs(p3(4))
zplane(z,p)
disp ('the angle of each pole')
A2_1 = angle(p3(1))*57.2958
A2 2=angle(p3(2))*57.2958
A2 3=angle(p3(3))*57.2958
A2 4=angle(p3(4))*57.2958
%%%comparing the Zpmax at (1) and (2)
Zp1max = max(abs(p1))
Zp2max = max(abs(p3))
%%comparing the approximation accuracy of H D(ejwT)/H(0) and H(jw)/H(0)
H1 = freqs(b1,a1,w1);
figure(21);
subplot(2,1,1);
plot(w1, abs(H1/H1(1)));
grid on;
xlabel('angular frequency')
ylabel('|H1(jw)/H1(0)|')
title('H1(jw)/H1(0)')
subplot(2,1,2);
plot(w1, angle(H1/H1(1))*180/pi);
xlabel('angular frequency')
ylabel('\theta')
grid on;
[H D1, w2] = freqz(bz1, az1);
figure (22);
subplot(2,1,1);
plot(w2, abs((H D1)/(H D1(1))));
xlabel('frequency, [Hz]')
ylabel('|H_D1(ejwT)/H_D1(1)|')
title('|H D1(ejwT)/H D1(1)|')
grid on;
subplot(2,1,2);
plot(w2,angle((H D1)/(H D1(1)))*180/pi);
xlabel('frequency, [Hz]')
ylabel('\theta')
grid on;
figure(3);
subplot(2,1,1);
plot(w1, abs(H1/H1(1)), 'b-');
```

```
xlabel('angular frequency')
title('Comparing the approximation accuracy of H D1(ejwT)/H D1(1) relative to
H1(jw)/H1(0)')
hold on;
plot(w2, abs(H D1/H D1(1)), 'r:');
legend('H1/H1(0)','H D1/H D1(1)')
grid on;
subplot(2,1,2);
plot(w1, angle(H1/H1(1))*180/pi, 'b-');
hold on;
plot(w2, angle(H D1/H D1(1))*180/pi, 'r:');
ylabel('\theta')
xlabel('angular frequency')
legend('H1/H1(0)','H D1/H D1(1)')
grid on;
f2=3000;
W2=2*pi*f2;
fs2=15000;
fsp2=0:1:fs2/2;
w3=2*pi*fsp2;
[z4,p4,k4]=cheb1ap(4,0.5);% low pass filter prototype
[b2,a2] = zp2tf(z4,p4,k4); % % Convert to transfer function form
[b2,a2] = lp2lp(b2,a2,2*pi*f2); analog low pass filter to low pass filter of
specified frequency
[bz2,az2]=impinvar(b2,a2,fs2);% using impulse invarient method
disp('-----')
disp ('3a
          _____ Hd(z)when f1=3KHz fs=15KHz ____')
tf(bz2,az2)
disp ('poles(3)')
[z5,p5,k5] = tf2zp(bz2,az2)
disp ('the magnitude of each pole')
p3_1 = abs(p5(1))
p3 2 = abs(p5(2))
p3 \ 3 = abs(p5(3))
p3 \ 4 = abs(p5(4))
zplane(z,p)
disp ('the angle of each pole')
A3 1 = angle(p5(1))*57.2958
A3 2=angle(p5(2))*57.2958
A3 3=angle(p5(3))*57.2958
A3 4=angle(p5(4))*57.2958
Zp1max = max(abs(p1))
Zp3max = max(abs(p5))
H2 = freqs(b2, a2, w3);
```

```
figure (31);
subplot(2,1,1);
plot(w3, abs(H2/H2(1)));
grid on;
xlabel('angular frequency')
ylabel('|H2(jw)/H2(0)|')
title('H2(jw)/H2(0)')
subplot(2,1,2);
plot(w3, angle(H2/H2(1))*180/pi);
xlabel('angular frequency')
ylabel('\theta')
grid on;
[H D2,w4] = freqz(bz2,az2);
figure(32);
subplot(2,1,1);
plot(w4, abs((H D2)/(H D2(1))));
xlabel('angular frequency')
ylabel('|H D2(ejwT)/H D2(1)|')
title('|H D2(ejwT)/H D2(1)|')
grid on;
subplot(2,1,2);
plot(w4, angle((H D2)/(H D2(1)))*180/pi);
xlabel('angular frequency')
ylabel('\theta')
grid on;
f3=4000;
W3=2*pi*f3;
fs3=40000;
df=1;
fsp3=0:df:fs3/2;
w4=2*pi*fsp3;
[z6,p6,k6]=cheblap(4,0.5);% low pass filter prototype
[b3,a3] = zp2tf(z6,p6,k6); % % Convert to transfer function form
[b3,a3] = lp2lp(b3,a3,2*pi*f3); % analog low pass filter to low pass filter of
specified frequency
[bz3,az3]=impinvar(b3,a3,fs3); % using impulse invarient method
% poles of Hd4(z)
disp('-----')
tf(bz3,az3)
disp ('poles(4)')
[z7,p7,k7] = tf2zp(bz3,az3)
disp ('the magnitude of each pole')
p4 1 = abs(p7(1))
```

```
p4 2 = abs(p7(2))
p4_3 = abs(p7(3))
p4 \ 4 = abs(p7(4))
zplane(z,p)
disp ('the angle of each pole')
A4 1 = angle(p7(1))*57.2958
A4 2=angle(p7(2))*57.2958
A4 3=angle(p7(3))*57.2958
A4 4=angle(p7(4))*57.2958
%%%comparing the Zpmax at (2) and (4)
Zp2max = max(abs(p3))
Zp4max = max(abs(p7))
H3 = freqs(b3,a3,w4);
figure (41);
subplot(2,1,1);
plot(w4, abs(H3/H3(1)));
grid on;
xlabel('angular frequency')
ylabel('|H3(jw)/H3(0)|')
title('H3(jw)/H3(0)')
subplot(2,1,2);
plot(w4, angle(H3/H3(1))*180/pi);
xlabel('angular frequency')
ylabel('\theta')
grid on;
[H D3,w5] = freqz(bz3,az3);
figure (42);
subplot(2,1,1);
plot(w5, abs((H_D3)/(H_D3(1))));
xlabel('angular frequency')
ylabel('|H D3(ejwT)/H D3(1)|')
title('|H D3(ejwT)/H D3(1)|')
grid on;
subplot(2,1,2);
plot(w5, angle((H D3)/(H D3(1)))*180/pi);
xlabel('angular frequency')
ylabel('\theta')
grid on;
figure (43)
subplot(2,1,1);
plot(w4/2/pi, abs(H3/H3(1)), 'b-');
xlabel('frequency, [Hz]')
 \begin{tabular}{ll} \textbf{title('Comparing the approximation accuracy of HD3(ejwT)/HD3(1) relative to} \\ \end{tabular} 
H3(jw)/H3(0)')
hold on;
plot(w5/2/pi*fs1, abs(H_D3/H_D3(1)), 'r:');
legend('H3/H3(0)','H_D3/H_D3(1)')
grid on;
subplot(2,1,2);
plot(w4/pi/2, angle(H3/H3(1))*180/pi, 'b-');
hold on;
plot(w5/pi*fs1/2,angle(H D3/H D3(1))*180/pi,'r:');
```

```
ylabel('\theta')
xlabel('frequency, [Hz]')
legend('H3/H3(0)','H D3/H D3(1)')
grid on;
[z,p,k] = cheb1ap(4,0.5);
[A,B,C,D] = zp2ss(z,p,k);
[A1,B1,C1,D1] = lp2lp(A,B,C,D,2*pi*f1);
[b4,a4] = ss2tf(A1,B1,C1,D1);
[bz4,az4] = bilinear(b,a,fs);
응(b)
disp('-----')
disp ('_____ Hd(z) when fs=40KHz_____')
tf(bz4,az4)
disp ('poles(5)')
[z8,p8,k8] = tf2zp(bz4,az4);
disp ('the magnitude of each pole')
p5_1 = abs(p8(1))
p5 2 = abs(p8(2))
p5 \ 3 = abs(p8(3))
p5 \ 4 = abs(p8(4))
zplane(z,p)
disp ('the angle of each pole')
A5 1 = angle(p8(1))*57.2958
A5_2=angle(p8(2))*57.2958
A5_3=angle(p8(3))*57.2958
A5 4=angle(p8(4))*57.2958
용(C)
H4 = freqs(b,a,w);
figure (51);
subplot(2,1,1);
plot(w, abs(H4/H4(1)));
grid on;
xlabel('angular frequency')
ylabel('|H5(jw)/H5(0)|')
title('H5(jw)/H5(0)')
subplot(2,1,2);
plot(w,angle(H4/H4(1))*180/pi);
xlabel('angular frequency')
ylabel('\theta')
grid on;
[H D4,w6] = freqz(bz4,az4);
figure (52);
subplot(2,1,1);
plot(w6, abs(H D4/H D4(1)));
xlabel('angular frequency')
ylabel('|H D5(ejwT)/H D5(1)|')
```

```
\mathtt{title('H\_D5(ejwT)/H\_D5(1)')}
grid on;
subplot(2,1,2);
plot(w6, angle(H D4/H D4(1))*180/pi);
xlabel('angular frequency')
ylabel('\theta')
grid on;
f1 5 = atan(f1*2*pi/fs/2)*2*fs/2/pi % new value of cuttoff frequency
%% Code for (6)
W3db=10e4;
ws=4*W3db;
fs4=(ws/2/pi);
df=1;
fsp4=0:df:fs4/2;
w4=2*pi*fsp4;
num=[0 0 0 0 0 0 1];
den=[1 3.8637033 7.4641016 9.1416202 7.4641016 3.8637033 1];
[b5,a5] = 1p2hp(num,den,W3db);
[bz5,az5]=bilinear(b5,a5,fs4);
H5 = freqs(b5,a5,w4);
figure (61);
subplot(2,1,1);
plot(w4/2/pi, abs(H5));
xlabel('frequency, [Hz]')
ylabel('|H5(jw)|')
title('normalized | H5(jw)/H5(3184)|')
grid on;
subplot(2,1,2);
plot(w4/2/pi,angle((H5)*180/pi));
xlabel('frequency, [Hz]')
ylabel('\theta')
grid on;
[H D5, w7] = freqz(bz5, az5);
figure (62);
subplot(2,1,1);
plot(w7/pi/2*fs4, abs(H D5/H D5(512)));
xlabel('frequency, [Hz]')
ylabel('|H D5(ejwT)|')
title('|H D5(ejwT)/H D5(512)|')
grid on;
subplot(2,1,2);
plot(w7/pi/2*fs4,angle((H D5/H D5(512))*180/pi));
xlabel('frequency, [Hz]')
ylabel('\theta')
grid on;
disp ('poles(6)')
[z9,p9,k9] = tf2zp(bz5,az5);
disp ('the magnitude of each pole')
```

```
p6 1 = abs(p9(1))
p6^{-2} = abs(p9(2))
p6_3 = abs(p9(3))
p6 \ 4 = abs(p9(4))
zplane(z,p)
% new value of cutoff for digital filter calculated analytically
w3db=atan(W3db/fs4/2)*2*fs4
%(7) Prewarping
T=1;
ws=4*10e4;
Ts=1/fs4;
num=[0 0 0 0 0 0 1];
den=[1 3.8637033 7.4641016 9.1416202 7.4641016 3.8637033 1];
[b5,a5] = lp2hp(num,den,W3db);
w8=2*fs4*tan(W3db/2/fs4);
fp=(w8/2/pi);
[bz6,az6]=bilinear(b5,a5,fs4,fp);
H5 = freqs(b5, a5, w4);
figure (71);
subplot(2,1,1);
plot(w4/2/pi, abs(H5));
xlabel('frequency, [Hz]')
ylabel('|H6(jw)|')
title('|H6(jw)/H6(3184)|')
grid on;
subplot(2,1,2);
plot(w4/2/pi,angle((H5)*180/pi));
xlabel('frequency, [Hz]')
ylabel('\theta')
grid on;
[H D6,w8] = freqz(bz6,az6);
figure(72);
subplot(2,1,1);
plot(w8/pi*fs4/2, abs(H_D6/H_D6(512)));
xlabel('frequency, [Hz]')
ylabel('|H D6(ejwT)|')
title('|H D6(ejwT)/H D6(512)|')
grid on;
subplot(2,1,2);
plot(w8/pi*fs4/2,angle((H D6/H D6(512))*180/pi));
xlabel('frequency, [Hz]')
ylabel('\theta')
grid on;
```