

# EN1020 Circuits, Signals, and Systems: Introduction to Fourier Transform

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## Section 1

# Continuous-Time Fourier Transform

# Outline

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### Introduction

Development of the Fourier Transform Representation

The Fourier Transform for Periodic Signals

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- In EN2063, we will do a more rigorous study of Fourier techniques.

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- The resulting spectrum of coefficients in this representation is called the **Fourier transform**.
- The synthesis integral itself, which uses the Fourier transform to represent the signal as a linear combination (integral) of complex exponentials, is called the **inverse Fourier transform**.

# Outline

## Continuous-Time Fourier Transform

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# Fourier Series Representation for Square Wave

The continuous-time periodic square wave, sketched below, is defined over one period as

$$x(t) = \begin{cases} 1, & |t| < T_1, \\ 0, & T_1 < |t| < T/2, \end{cases}$$

This signal periodically repeats with the fundamental period  $T$  and the fundamental frequency  $\omega_0 = 2\pi/T$ .

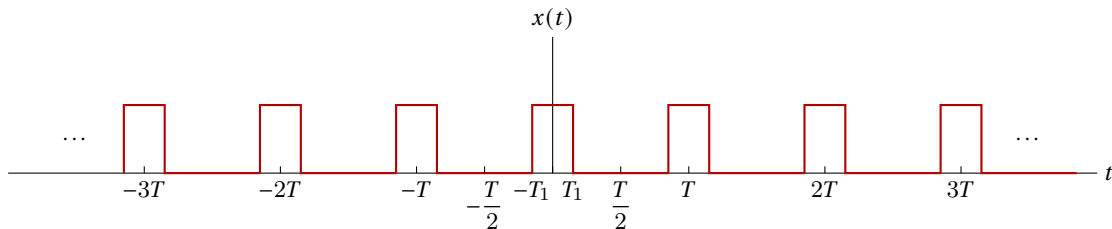


Figure: Periodic square wave

The Fourier series coefficients  $a_k$  of this wave are

$$a_k = \frac{2 \sin(k \omega_0 T_1)}{k \omega_0 T}. \quad (1)$$

We plotted this for a fixed value of  $T_1$  and several values of  $T$  (shown in the next slide). An alternative way of interpreting Eq. 1 is as samples of an envelope function:

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With  $\omega$  thought of as a continuous variable, the function  $\frac{2 \sin(\omega T_1)}{\omega}$  represents the envelope of  $T a_k$ , and the coefficients  $a_k$  are simple equally spaced samples of this envelope. For fixed  $T_1$ , the envelope of  $T a_k$  is independent of  $T$ .

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Plots of scaled  
Fourier series  
coefficients  $a_k$  for  
the periodic square  
wave with  $T_1$  fixed  
and for several  
values of  $T$ :  $T = 4T_1$ ,  
 $T = 8T_1$ ,  $T = 16T_1$ .

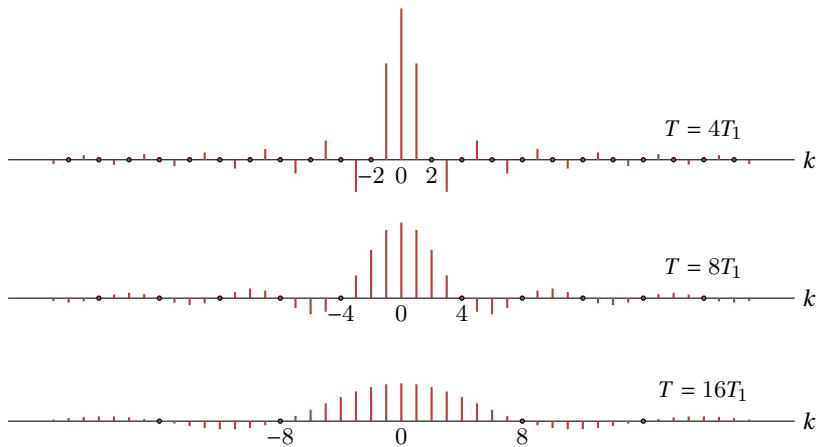
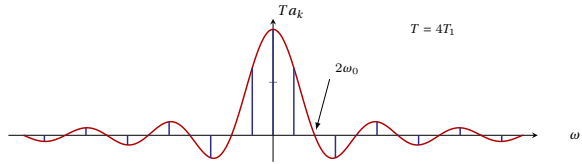


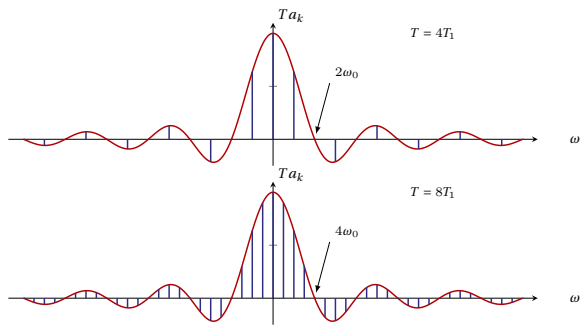
Figure: Plots of scaled Fourier series coefficients  $a_k$



The Fourier series coefficients and their envelope for periodic square wave for several values of  $T$  (with  $T_1$  fixed):  $T = 4T_1$ ,  $T = 8T_1$ ,  $T = 16T_1$ . The coefficients are regularly-spaced samples of the envelope  $(2 \sin \omega T_1) / \omega$ , where the spacing between samples,  $2\pi / T$ , decreases as  $T$  increases.

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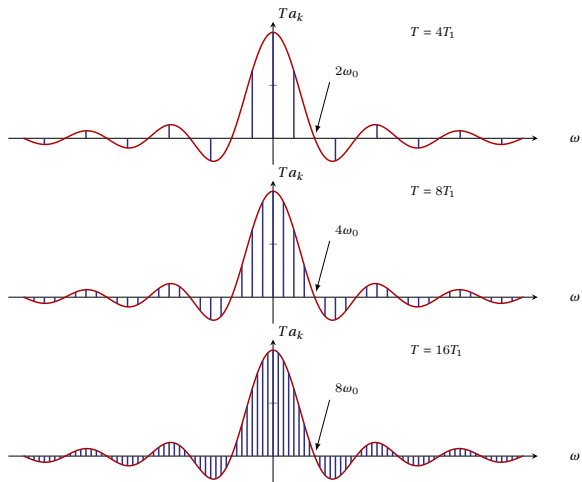
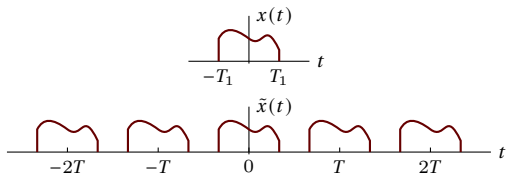


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As  $T$  increases, or equivalently, as the fundamental frequency  $\omega_0 = 2\pi/T$  decreases, the envelope is sampled with a close and closer spacing. As  $T$  becomes arbitrarily large, the original periodic square waveform approaches the rectangular pulse. Also, the Fourier series coefficients, multiplied by  $T$ , become more and more closely spaced samples of the envelope. So, in some sense, the set of Fourier series coefficients approaches the envelope function as  $T \rightarrow \infty$ .



$$\tilde{x}(t) = \sum_{k=-\infty}^{\infty} a_k e^{jk\omega_0 t}, \quad \omega_0 = \frac{2\pi}{T}$$

$$a_k = \frac{1}{T} \int_{-T/2}^{T/2} \tilde{x}(t) e^{-jk\omega_0 t} dt.$$

As  $\tilde{x}(t) = x(t)$  for  $|t| < T/2$ , and also, as  $x(t) = 0$  outside this interval,

$$a_k = \frac{1}{T} \int_{-T/2}^{T/2} x(t) e^{-jk\omega_0 t} dt = \frac{1}{T} \int_{-\infty}^{\infty} x(t) e^{-jk\omega_0 t} dt.$$

Defining the envelope  $X(j\omega)$  of  $Ta_k$  as

$$X(j\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt,$$

we have, for the coefficients  $a_k$ ,

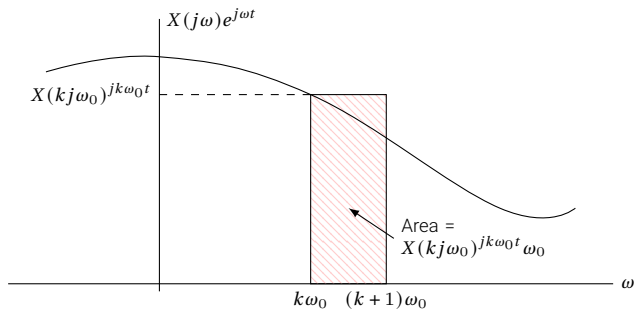
$$a_k = \frac{1}{T} X(jk\omega_0).$$

Combining and expressing  $\tilde{x}(t)$  in terms of  $X(j\omega)$

$$\tilde{x}(t) = \sum_{k=-\infty}^{\infty} \frac{1}{T} X(jk\omega_0) e^{jk\omega_0 t},$$

or, as  $\omega_0 = 2\pi/T$

$$\tilde{x}(t) = \frac{1}{2\pi} \sum_{k=-\infty}^{\infty} X(jk\omega_0) e^{jk\omega_0 t} \omega_0. \quad (2)$$



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Inverse Fourier transform:

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega.$$

Fourier transform or Fourier integral:

$$X(j\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt.$$

As  $T \rightarrow \infty$ ,  $\tilde{x}(t)$  approaches  $x(t)$ , and consequently, Eq. 2 becomes a representation of  $x(t)$ . Furthermore, as  $\omega_0 \rightarrow 0$  as  $T \rightarrow \infty$ , and the right-hand side of Eq. 2 passes to an integral. As  $\omega_0 \rightarrow 0$ , the summation converges to the integral of  $X(j\omega) e^{j\omega t}$ .

# Fourier Transform: Synthesis and Analysis Equations

Inverse Fourier transform:

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega. \quad (3)$$

Fourier transform or Fourier integral:

$$X(j\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt. \quad (4)$$

The transform  $X(j\omega)$  of an aperiodic signal  $x(t)$  is referred to as the spectrum of  $x(t)$ .

FS synthesis and analysis equations:

$$x(t) = \sum_{k=-\infty}^{+\infty} a_k e^{jk\omega_0 t}$$
$$a_k = \frac{1}{T} \int_T x(t) e^{-jk\omega_0 t} dt$$

## Relation with $a_k$

Assume that the Fourier transform of  $x(t)$  is  $X(j\omega)$ .

If we construct a periodic signal  $\tilde{x}(t)$  by repeating the aperiodic signals  $x(t)$  with period  $T$ , its Fourier series coefficients are

$$a_k = \frac{1}{T} X(j\omega) \Big|_{\omega=k\omega_0} \quad (5)$$

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$$X(j\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt$$

# Convergence of Fourier Transform

Assume that we evaluated  $X(j\omega)$  according to eq. 4, and let  $\hat{x}(t)$  denote the signal obtained by using  $X(j\omega)$  in 3:

$$\hat{x}(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega.$$

When is  $\hat{x}(t)$  a valid representation of the original signal  $x(t)$ ? We define the error between  $\hat{x}(t)$  and  $x(t)$  as

$$e(t) = \hat{x}(t) - x(t).$$

If  $x(t)$  has finite energy (square integrable), i.e.,

$$\int_{-\infty}^{\infty} |x(t)|^2 dt < \infty, \quad (6)$$

$X(j\omega)$  is finite, and

$$\int_{-\infty}^{\infty} |e(t)|^2 dt = 0, \quad (7)$$

If  $x(t)$  has finite energy, then, although  $x(t)$  and its Fourier representation  $\hat{x}(t)$  may differ significantly at individual values of  $t$ , there is no energy in their difference.



# Convergence of Fourier Transform: Dirichlet Conditions

There are alternative conditions sufficient to ensure that  $\hat{x}(t)$  is equal to  $x(t)$  for any  $t$  except at a discontinuity, where it is equal to the average of the values on either side of the discontinuity.

1.  $x(t)$  is absolutely integrable, i.e.,

$$\int_{-\infty}^{\infty} |x(t)| dt < \infty, \quad (8)$$

2.  $x(t)$  has a finite number of maxima and minima within any finite interval.
3.  $x(t)$  has a finite number of discontinuities within any finite interval. Furthermore, each of these discontinuities must be finite.

Therefore, absolutely integrable signals that are continuous or that have finite number of discontinuities have a Fourier transform.

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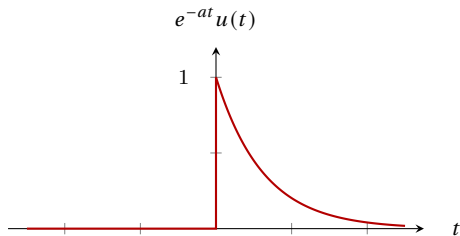


Figure:  $e^{-at}u(t)$ ,  $a > 0$ .

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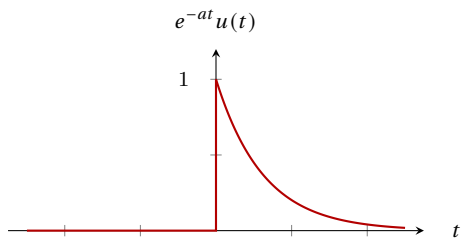


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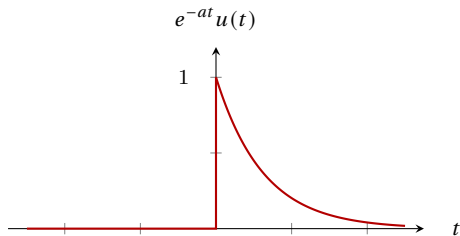


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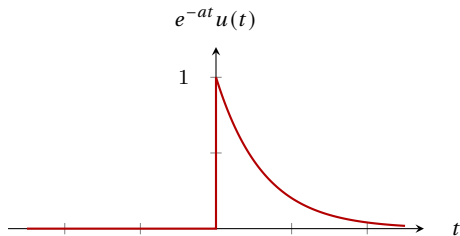


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Example Cntd. FT of  $e^{-at}u(t)$ ,  $a > 0$

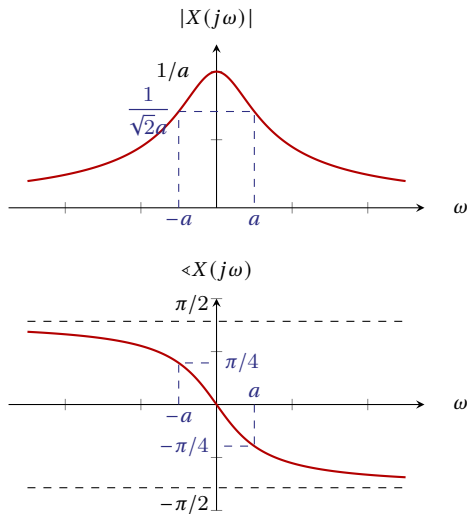


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$$X(j\omega) = \frac{1}{a - j\omega} + \frac{1}{a + j\omega},$$

$$= \frac{2a}{a^2 + \omega^2}.$$

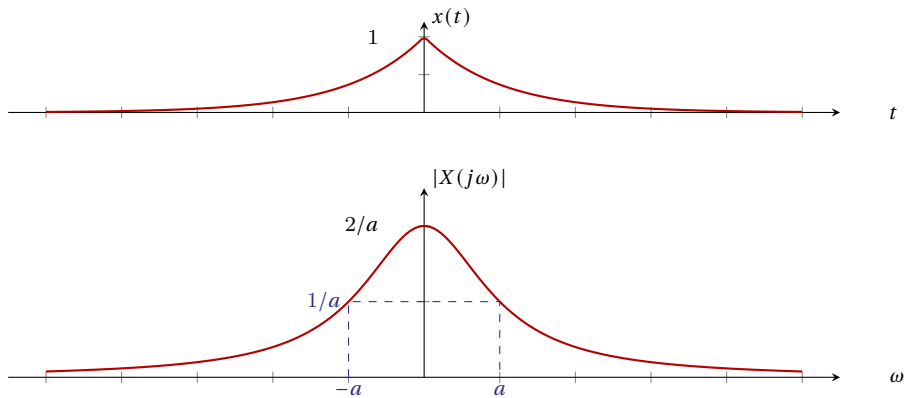


Figure: Fourier transform of the signal  $e^{-a|t|}$ ,  $a > 0$ .

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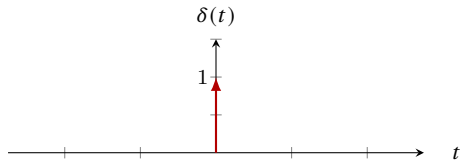


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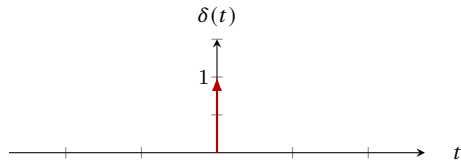


Figure:  $\delta(t)$

$$X(j\omega) = \int_{-\infty}^{\infty} \delta(t) e^{-j\omega t} dt = 1.$$

The unit impulse has a Fourier transform consisting of equal contributions at all frequencies.



# Rectangular Pulse

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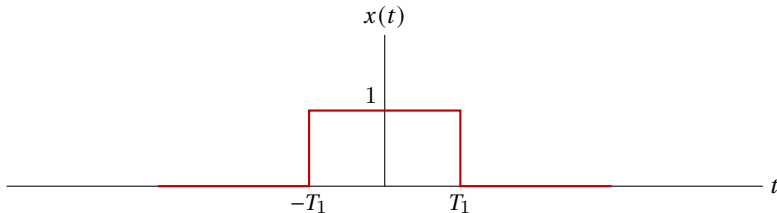


Figure: Rectangular pulse and the Fourier transform.

$$X(j\omega) = \int_{-\infty}^{\infty} x(t)e^{-j\omega t} dt.$$

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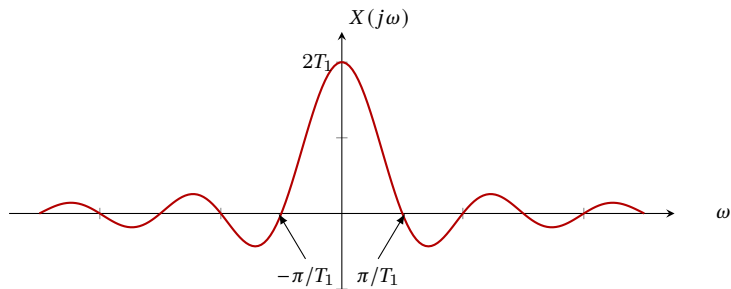


Figure: Fourier transform of the rectangular pulse.

## Example

Consider the signal  $x(t)$  whose Fourier transform is

$$X(j\omega) = \begin{cases} 1, & |\omega| < W, \\ 0, & |\omega| > W. \end{cases}$$

Determine  $x(t)$ .

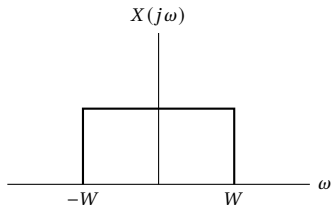


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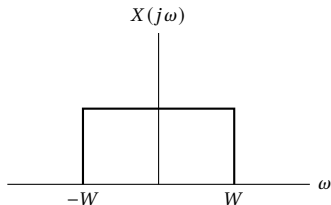


Figure: Fourier transform for  $x(t)$ .

Using the synthesis equation:

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega.$$

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$$\begin{aligned}x(t) &= \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega. \\&= \frac{1}{2\pi} \int_{-W}^W e^{j\omega t} d\omega.\end{aligned}$$

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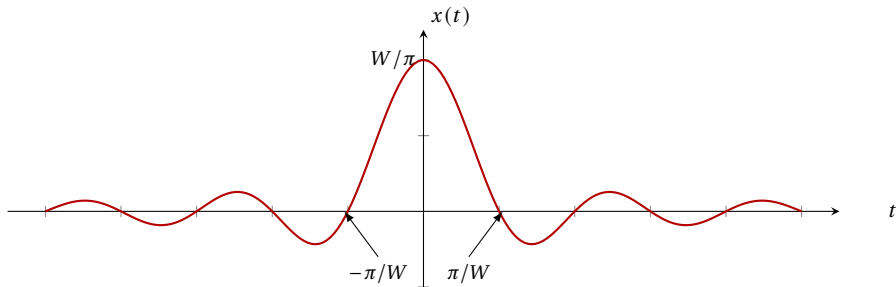


Figure: Time function.

# The sinc Function

$$\text{sinc}(\theta) = \frac{\sin \pi \theta}{\pi \theta}. \quad (9)$$

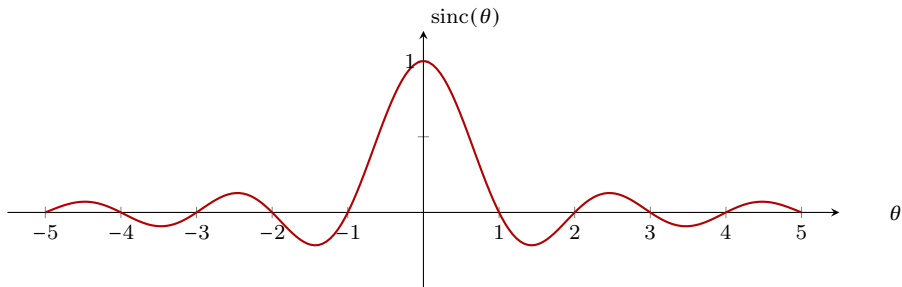


Figure: Fourier transform for  $x(t)$ .

Express

$$\frac{2 \sin \omega T_1}{\omega}$$

and

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$$\begin{aligned}\frac{2 \sin \omega T_1}{\omega} &= 2T_1 \operatorname{sinc}\left(\frac{\omega T_1}{\pi}\right) \\ \frac{\sin W t}{\pi t} &= \frac{W}{\pi} \operatorname{sinc}\left(\frac{W t}{\pi}\right)\end{aligned}$$

## What Happens when $W$ Increases?

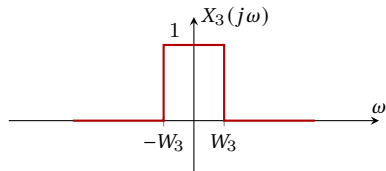
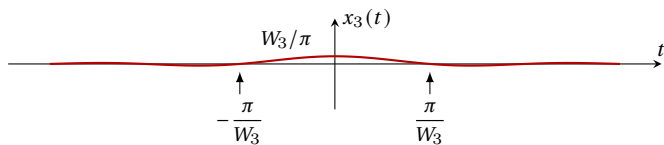
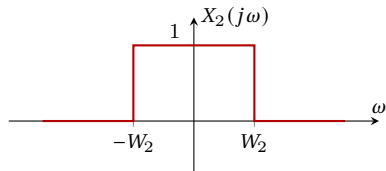
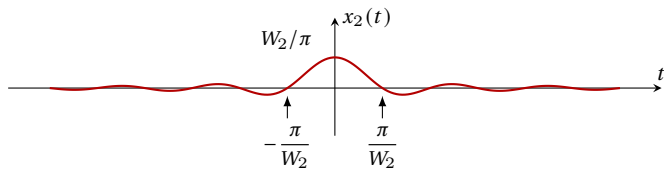
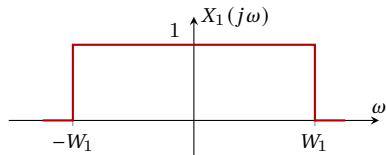
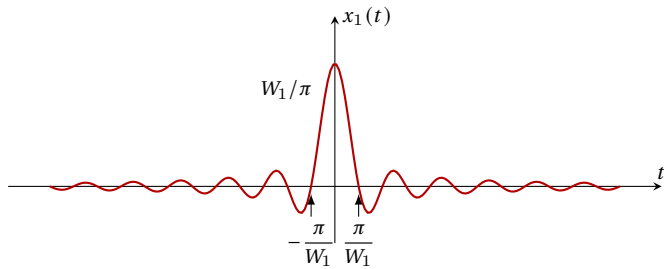
- As  $W$  increases,  $X(j\omega)$  becomes broader, while the main peak of  $x(t)$  at  $t = 0$  becomes higher and the width of the first lobe of this signal (i.e., the part of the signal for  $|t| < \pi/W$ ) becomes narrower.

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- The behavior is an example of the inverse relationship that exists between the time and frequency domains.



# Outline

## Continuous-Time Fourier Transform

Introduction

Development of the Fourier Transform Representation

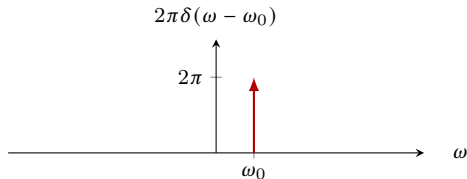
The Fourier Transform for Periodic Signals

# The Fourier Transform for Periodic Signals: Introduction

In the previous section, we studied the Fourier transform representation, paying attention to aperiodic signals. We can also develop Fourier transform representations for periodic signals. This allows us to consider periodic and aperiodic signals in a unified context. We can construct the Fourier transform of a periodic signal directly from its Fourier series representation.

Consider a signal  $x(t)$  with the Fourier transform  $X(j\omega)$  that is a single impulse of area  $2\pi$  at  $\omega = \omega_0$ , i.e.,

$$X(j\omega) = 2\pi\delta(\omega - \omega_0) \quad (10)$$



Let's determine the signal  $x(t)$ :



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More generally, if  $X(j\omega)$  is of the form of a linear combination of impulses equally spaced in frequency, i.e.,

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then

$$x(t) = \sum_{k=-\infty}^{\infty} a_k e^{jk\omega_0 t}. \quad (12)$$

which is exactly the Fourier series representation of a periodic signal.

Thus, the Fourier transform of a periodic signal with Fourier series coefficients  $\{a_k\}$  can be interpreted as a train of impulses occurring at the harmonically related frequencies and for which the area of the impulse at the  $k$ th harmonic frequency  $k\omega_0$  is  $2\pi$  times the  $k$ th Fourier series coefficient  $a_k$ .

## Example

Find the Fourier transform of the square wave signal whose Fourier series coefficients are

$$a_k = \frac{\sin k\omega_0 T_1}{\pi k}.$$

Method: Multiply the Fourier series coefficients  $a_k$  by  $2\pi$ , place them using the impulse function  $\delta(\omega - k\omega_0)$ , and sum.

$$X(j\omega) = \sum_{k=-\infty}^{\infty} \frac{2 \sin k\omega_0 T_1}{k} \delta(\omega - k\omega_0).$$

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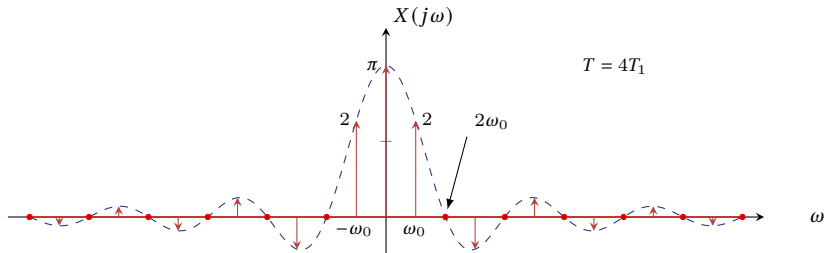


Figure: Fourier transform of a symmetric periodic square wave.

## Example

Find the Fourier transform of

$$x(t) = \sin \omega_0 t,$$

and

$$x(t) = \cos \omega_0 t.$$

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The Fourier series coefficients for this signal are

$$a_1 = \frac{1}{2j}, \quad a_{-1} = -\frac{1}{2j},$$

$$a_k = 0, k \neq 1 \text{ or } -1$$



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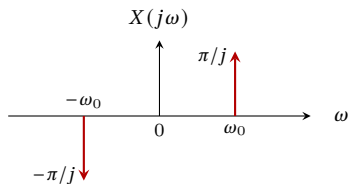


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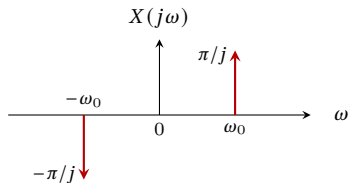


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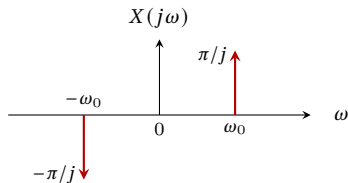


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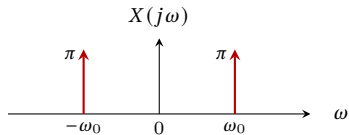


Figure: Fourier transform of the  $x(t) = \cos \omega_0 t$ .

## Example

Find the Fourier transform of the impulse train

$$x(t) = \sum_{k=-\infty}^{\infty} \delta(t - kT).$$

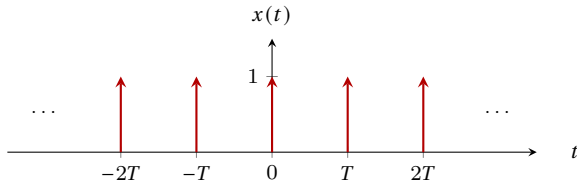


Figure: Pulse train.

The Fourier series coefficients for this signal:

$$a_k = \frac{1}{T} \int_{-T/2}^{T/2} \delta(t) e^{-jk\omega_0 t} dt = \frac{1}{T}.$$

That is, every Fourier coefficient of the periodic impulse train has the same value,  $1/T$ .  
Substituting this value for  $a_k$

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Thus, the Fourier transform of a periodic impulse train in the time domain with period  $T$  is a periodic impulse train in the frequency domain with period  $2\pi/T$ ,



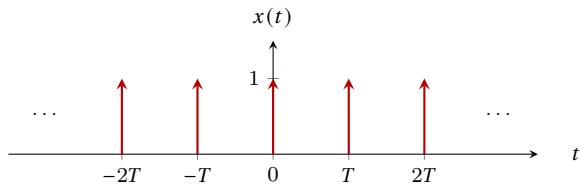


Figure: Periodic impulse train and its Fourier transform.

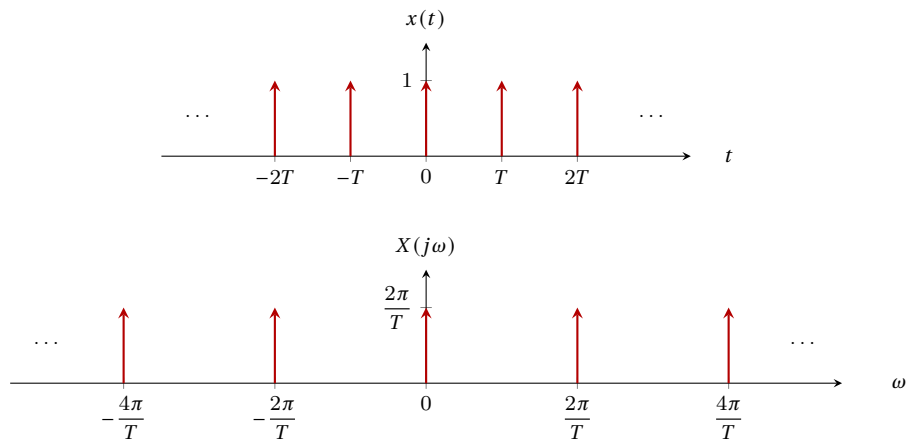


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