

## Revised Schedule

- 9:00 ~ 10:30 **session 1**
- 10:30 ~ 11:00 coffee
- 11:00 ~ 12:30 **session 2**
- 12:30 ~ 13:30 lunch
- 13:30 ~ 15:00 **session 3**
- 15:00 ~ 15:30 coffee
- 15:30 ~ 17:30 **session 4**
- 17:30 ~ 9:30 beer & homework

## ME1: Measurement Equation Of a (Polarized) Point Source

Objectives:

- **Aligning our terminology!**
- Mapping your existing intuition about interferometry onto ME concepts
- Implementing some MEs using MeqTrees.

## The Measurement Equation: Putting the “Meq” into MeqTrees!

- The Measurement Equation tells you what you can expect to observe with an interferometer, given a sky and the properties of your instrument.
- Absolutely crucial for simulating and calibrating the next generation of radio telescopes; everything literally revolves around it.

**Therefore:** no-one gets any beer tonight until we achieve full harmony and understanding!

## Survey Results...

### 1. Radio interferometry...

- [ 1] heard of it
- [11] basic knowledge
- [ 6] do it all the time
- [ 3] I was doing it when Oleg was in diapers

Therefore...

...you know practically everything about the Measurement Equation!

## On The Other Hand...

### 3. The Measurement Equation...

- [ 1] never heard of it
- [ 2] heard of it
- [10] know what it looks like, never actually used it
- [ 6] know it pretty well
- [ 2] I use Jones matrices to do my taxes

- Conclusion:

The ME is no longer one of these unknown knowns -- the things we don't know we know -- that Donald Rumsfeld didn't know he knew.

## A Wafer-Thin Slice of Physics: EM Field Propagation

Pick an  $xyz$  frame with  $z$  along the direction of propagation.

The EM field can be described by the complex vector  $\vec{e} = \begin{pmatrix} e_x \\ e_y \end{pmatrix}$

The fundamental assumption is **LINEARITY**:

1. Propagation through a medium is linear

⇒ can be fully described by a 2x2 complex matrix:

$$\vec{e}' = \mathbf{J}\vec{e} \quad \text{i.e.} \quad \begin{pmatrix} e'_x \\ e'_y \end{pmatrix} = \begin{pmatrix} \square & \square \\ \square & \square \end{pmatrix} \begin{pmatrix} e_x \\ e_y \end{pmatrix}$$

2. Receptor voltages  $\vec{v} = \begin{pmatrix} v_x \\ v_y \end{pmatrix}$  are also linear w.r.t.  $\vec{e}$

$$\Rightarrow \vec{v} = \mathbf{J}\vec{e}$$

## A Wafer-Thin Slice of Physics: Jones Matrices

\*  $\mathbf{J}$  is called a *Jones matrix*.

\*  $\mathbf{J}$  is obviously cumulative:

$$\vec{v} = \mathbf{J}_n(\mathbf{J}_{n-1}(\dots \mathbf{J}_1 \vec{e})) = \left( \prod_{i=1}^n \mathbf{J}_i \right) \vec{e} = \mathbf{J} \vec{e}$$

where  $\mathbf{J}_1 \dots \mathbf{J}_n$  describes the full signal path.

\* Do remember that matrices, in general, do not commute.

NB: What if something is non-linear?..

...we can also write down an equation:  $\vec{v} = \mathcal{J}(\vec{e})$

...but this is to be avoided if at all possible.

## A Wafer-Thin Slice of Physics: Correlations & Visibilities

An interferometer measures *correlations* btw voltages  $\vec{v}_p, \vec{v}_q$ :

$$v_{xx} = \langle v_{px} v_{qx}^* \rangle, v_{xy} = \langle v_{px} v_{qy}^* \rangle, v_{yx} = \langle v_{py} v_{qx}^* \rangle, v_{yy} = \langle v_{py} v_{qy}^* \rangle$$

It is convenient to represent these as a matrix product:

$$\mathbf{V}_{pq} = \langle \vec{v}_p \vec{v}_q^\dagger \rangle = \begin{pmatrix} v_{px} \\ v_{py} \end{pmatrix} \begin{pmatrix} v_{qx}^* & v_{qy}^* \end{pmatrix} = \begin{pmatrix} v_{xx} & v_{xy} \\ v_{yx} & v_{yy} \end{pmatrix}$$

( $\langle \rangle$ ): time/freq averaging;  $\dagger$ : conjugate-and-transpose)

$\mathbf{V}_{pq}$  is also called the *visibility matrix*.

Now let's assume that all radiation arrives from a single point, and designate the "source" E.M. vector by  $\vec{e}$ .

## A Wafer-Thin Slice of Physics: The M.E. Emerges

Antennas  $p, q$  then measure:  $\vec{v}_p = \mathbf{J}_p \vec{e}$ ,  $\vec{v}_q = \mathbf{J}_q \vec{e}$   
where  $\mathbf{J}_p, \mathbf{J}_q$  are Jones matrices describing the signal paths  
from the source to the antennas.

Then  $\mathbf{V}_{pq} = \langle (\mathbf{J}_p \vec{e})(\mathbf{J}_q \vec{e})^\dagger \rangle = \langle \mathbf{J}_p (\vec{e} \vec{e}^\dagger) \mathbf{J}_q^\dagger \rangle = \mathbf{J}_p \langle \vec{e} \vec{e}^\dagger \rangle \mathbf{J}_q^\dagger$   
(making use of  $(\mathbf{AB})^\dagger = \mathbf{B}^\dagger \mathbf{A}^\dagger$ , and assuming  $\mathbf{J}_p$  is constant over  $\langle \rangle$ )

The inner quantity is known as the *source coherency*:

$$\mathbf{B} = \langle \vec{e} \vec{e}^\dagger \rangle = \frac{1}{2} \begin{pmatrix} I+Q & U+iV \\ U-iV & I-Q \end{pmatrix} \leftrightarrow (I, Q, U, V)$$

which we can also call the *source brightness*. Thus:

$$\mathbf{V}_{pq} = \mathbf{J}_p \mathbf{B} \mathbf{J}_q^\dagger$$

## And That's The Measurement Equation!

$$\mathbf{V}_{pq} = \mathbf{J}_p \mathbf{B} \mathbf{J}_q^\dagger$$

- Or in more pragmatic terms:

$$\begin{pmatrix} XX & XY \\ YX & YY \end{pmatrix}^{\text{measured}} = \begin{pmatrix} j_{xx(p)} & j_{xy(p)} \\ j_{yx(p)} & j_{yy(p)} \end{pmatrix}^{\mathbf{J}_p} \frac{1}{2} \begin{pmatrix} I+Q & U+iV \\ U-iV & I-Q \end{pmatrix}^{\text{source}} \begin{pmatrix} j_{xx(q)}^* & j_{yx(q)}^* \\ j_{xy(q)}^* & j_{yy(q)}^* \end{pmatrix}^{\mathbf{J}_q^\dagger}$$

- NB: it is also possible to write the ME with a circular polarization basis (RR, LL, etc.) We'll use linear polarization throughout.

## Accumulating Jones Terms

If  $\mathbf{J}_p, \mathbf{J}_q$  are products of Jones matrices:

$$\mathbf{J}_p = \mathbf{J}_{pn} \cdots \mathbf{J}_{p1}, \quad \mathbf{J}_q = \mathbf{J}_{qm} \cdots \mathbf{J}_{q1}$$

Since  $(\mathbf{AB})^\dagger = \mathbf{B}^\dagger \mathbf{A}^\dagger$ , the M.E. becomes:

$$\mathbf{V}_{pq} = \mathbf{J}_{pn} \cdots \mathbf{J}_{p2} \mathbf{J}_{p1} \mathbf{B} \mathbf{J}_{q1}^\dagger \mathbf{J}_{q2}^\dagger \cdots \mathbf{J}_{qm}^\dagger$$

or in the "onion form":

$$\mathbf{V}_{pq} = \mathbf{J}_{pn} (\cdots (\mathbf{J}_{p2} (\mathbf{J}_{p1} \mathbf{B} \mathbf{J}_{q1}^\dagger) \mathbf{J}_{q2}^\dagger) \cdots) \mathbf{J}_{qm}^\dagger$$

## Jones' Anatomy

- $\mathbf{J}_p$  is "cumulative": more effects correspond to additional multiplicative Jones terms.
- Therefore, the "total"  $\mathbf{J}_p$  is a matrix product of a "Jones chain" of individual effects:

$$\mathbf{J}_p = \mathbf{J}_{pn} \mathbf{J}_{pn-1} \cdots \mathbf{J}_{p1}$$

- The order of the  $\mathbf{J}$  terms corresponds to the physical order of the effects. In general, the matrices don't commute!

## Why is this great?

- A complete and mathematically elegant framework for describing all kinds of signal propagation effects.
- ...including those at the antenna, e.g.:
  - beam & receiver gain
  - dipole rotation
  - receptor cross-leakage
- Effortlessly incorporates polarization:
  - think in terms of a **B** matrix and never worry about polarization again.
- Applies with equal ease to heterogeneous arrays, by using different Jones chains.

## Why is this even greater?

- Most effects have a very simple Jones representation:

$$\text{gain: } \mathbf{G} = \overbrace{\begin{pmatrix} g_x & 0 \\ 0 & g_y \end{pmatrix}}^{\text{diagonal matrix}} \quad \text{phase delay: } \overbrace{\begin{pmatrix} e^{-i\phi} & 0 \\ 0 & e^{-i\phi} \end{pmatrix}}^{\text{scalar matrix}} \equiv e^{-i\phi}$$

$$\text{rotation: } \begin{pmatrix} \cos y & -\sin y \\ \sin y & \cos y \end{pmatrix} \equiv \text{Rot}(y) \quad (\text{rotation matrix})$$

[ e.g. Faraday rotation:  $\mathbf{F} = \text{Rot}\left(\frac{RM}{v^2}\right)$  ]

$$\text{receptor cross-leakage: } \mathbf{D} = \begin{pmatrix} 1 & d \\ -d & 1 \end{pmatrix} \quad (\text{or Rot}(d)?)$$

## Three Layers Of Intuition

- **Physical:** e.g. beam gain, parallactic angle
  - beam pattern of X and Y dipoles different, causes polarization of off-center sources
  - P.A. rotates polarization angle
- **Geometrical:** stretching, rotation
  - do not commute...
- **Mathematical:** matrix properties

$$\mathbf{G} = \begin{pmatrix} g_x & 0 \\ 0 & g_y \end{pmatrix} \quad \text{and} \quad \mathbf{P} = \begin{pmatrix} \cos y & -\sin y \\ \sin y & \cos y \end{pmatrix} \quad \text{do not commute;}$$

$$\text{m.e. is: } \mathbf{V}_{pq} = \mathbf{G}_p \mathbf{P}_p \mathbf{B} \mathbf{P}_q^\dagger \mathbf{G}_q^\dagger$$

## ME ME ME

- The general formulation above is "The Measurement Equation" (of a generic radio interferometer...)
- When we want to simulate a specific instrument, we put specific Jones terms into the ME, and derive a measurement equation for that instrument.
- We then *implement* that specific m.e. in software (e.g. with MeqTrees)
- Existing packages implicitly use specific m.e.'s of their own.

## Observing a point source with a perfect instrument

Even w/o instrumental effects, we still have geometry, so:

$$\mathbf{V}_{pq} = \mathbf{K}_p \mathbf{B} \mathbf{K}_q^\dagger$$

$\mathbf{K}_p$  is the *phase shift* term, a **scalar** Jones matrix:

$$\mathbf{K}_p = \begin{pmatrix} e^{-i\phi_p} & 0 \\ 0 & e^{-i\phi_p} \end{pmatrix} \equiv e^{-i\phi_p}$$

Antenna phase  $\phi_p$  accounts for the pathlength difference:

$$\phi_p = 2\pi(u_p l + v_p m + w_p(n-1))$$

where  $u_p, v_p, w_p$  are *antenna coordinates* (in wavelengths), and  $l, m, n$  are the direction cosines for the source.

$$(n = \sqrt{1 - l^2 - m^2}, \text{ for "small" fields } n \rightarrow 1.)$$

## The (familiar?) Scalar Case

'Classic' (scalar) visibility of a source:

$$V_{pq} = I e^{-i\phi_{pq}}$$

where  $\phi_{pq}$  is the *interferometer phase difference*:

$$\phi_{pq} = 2\pi(u_{pq}l + v_{pq}m + w_{pq}(n-1))$$

Baseline coordinates  $\vec{u}_{pq} = (u_{pq}, v_{pq}, w_{pq})$

have a very simple relationship to antenna coordinates.

## Antenna UVWs and Antenna Phase

Pick an arbitrary reference point  $O$ .

$$\vec{u}_p \equiv \vec{OP}, \quad \vec{u}_q \equiv \vec{OQ}, \quad \vec{u}_{pq} \equiv \vec{QP}$$

Then regardless of which  $O$  we picked,

$$\vec{u}_{pq} = \vec{u}_p - \vec{u}_q, \text{ i.e.}$$

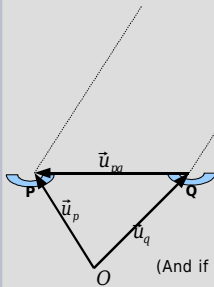
$$u_{pq} = u_p - u_q, \quad v_{pq} = v_p - v_q, \quad w_{pq} = w_p - w_q$$

and for phases:  $\phi_{pq} = \phi_p - \phi_q$ .

$$\Rightarrow e^{-i\phi_{pq}} = e^{-i(\phi_p - \phi_q)} = e^{-i\phi_p} e^{i\phi_q} = e^{-i\phi_p} (e^{-i\phi_q})^*$$

(And if you're used to thinking in terms of closure phases:

$$\phi_{pq} + \phi_{qr} + \phi_{rp} = \phi_p - \phi_q + \phi_q - \phi_r + \phi_r - \phi_p = 0)$$



## The (familiar?) Scalar Case

We can decompose each *interferometer* phase term into a pair of *antenna* phases:

$$V_{pq} = e^{-i\phi_p} I (e^{-i\phi_q})^*$$

compare this to:

$$\mathbf{V}_{pq} = \mathbf{K}_p \mathbf{B} \mathbf{K}_q^\dagger,$$

$$\text{with } \mathbf{B} = \frac{1}{2} \begin{pmatrix} I & 0 \\ 0 & I \end{pmatrix}$$

## K: Breaking The Coherency Barrier

- The interferometer phase term is traditionally considered separately; however, it fits the Jones formalism like a glove.
- **K** combines two physical effects:
  - pathlength difference
  - time delays (i.e. fringe stopping)
- Scalar matrices commute with everything, so we're allowed to "merge" these two effects and shift the resulting **K**'s around.
- **K** is scalar only for co-located receivers:
  - moral: keep your dipoles together!
- Forget about the Fourier Transform for now...

## Building a Tree: Matrix Multiplication

$$V_{pq} = K_p B K_q^\dagger$$

- See ME1/demo1-predict-ps.py
- The **Meq.MatrixMultiply** node implements matrix multiplication.
- We repeat this for all interferometers (all  $p$ - $q$  pairs), in a **for** loop.

## Building a Tree: Creating a B Matrix

$$B = \frac{1}{2} \begin{pmatrix} I+Q & U+iV \\ U-iV & I-Q \end{pmatrix}$$

- The **Meq.Matrix22** shortcut creates a matrix from four children (using a **Meq.Composer** node).
- $IQUV$ 's will be hardwired constants (for now)

## Building a Tree: VisPhaseShift

- The **Meq.VisPhaseShift** node computes the a phase term as follows:

$$\Phi(u, v, w, l, m, n; \nu) = \exp\left(-\frac{2\pi i \nu}{c}(ul + vm + wn)\right)$$

- Takes two "vector" children for  $uvw$ 's (in meters) and  $lmn$ 's.
- The **Meq.ConjTranspose** shortcut implements the "†" operation.
- We can explicitly form up an  $(l, m, n-1)$  vector using a **Meq.Composer** node.
- But where do we get the  $uvw$ 's?

## Building a Tree: Where to get meta-data?

Where do *uvw*'s come from?

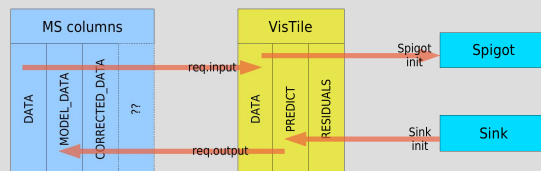
- *uvw*'s can be computed by a **Meq.UVW** node.
- This requires antenna positions, time, and the phase centre RA/Dec:
  - time: comes from the MS grid (via the request)
  - phase centre RA/Dec: from MS sub-tables
  - antenna positions: from MS sub-tables
- Don't want to hard-code this stuff, else our script will be tied to a particular MS.
- Need a way to get observational parameters from the MS and put them into the tree.

## Building a Tree: Attaching an init-script

- An *init-script* is a Python script that is executed *on the kernel side* by the **VisDataMux** node. The script name is specified as part of the I/O request.
- It can provide a *MS header handler* (among other things).
- The MS header contains all observational parameters.
- The header handler can put the required values into "placeholder" nodes (finding them by name).
- The same script can be used for all trees, as long as our placeholder nodes follow the same naming convention.
  - i.e. "ra", "dec" for phase center
  - "x0", "y0", "z0" for array center
  - "x:p", "y:p", "z:p" for position of antenna #*p*.
- Placeholders are created as constants, e.g. "ns . ra<<0".

## Building a Tree: Writing the data

- **VisTile** is an intermediate *uv* data layer (don't want to be locked into MSs forever...)
- The I/O record maps MS columns to VisTile
- Sinks & Spigots map tile columns to trees



## Building a Tree: Ready!

- Load up the "K Jones" and "Inspector" bookmarks.
- Run the tree by selecting "test forest"
- Now we want to make an image.

## We'll use the AIPS++ imager...

### 2. AIPS++...

[5] heard of it  
 [2] tried to run it once  
 [9] succeeded in running it once  
 [5] have used it in anger  
 [0] invented it

- We have enough expertise in this room...
- ...no-one to blame this time though.

## TDL Jobs: Doing other useful stuff

- `_test_forest()` is a "TDL job".
- More jobs can be added by defining functions called `_tdl_job_foo()`, all these will be automatically placed into the "Exec" menu.
- Jobs can contain arbitrary Python code...
  - ...including calling the shell...
  - ...e.g. to run Glish and call the AIPS++ imager.
- See ME1/demo2-predict-ps-image.py

## Introducing (Complex) Gain Errors

$$\mathbf{V}_{pq} = \mathbf{G}_p \mathbf{K}_p \mathbf{B} \mathbf{K}_q^t \mathbf{G}_q^t$$

$$\mathbf{G}_p = \begin{pmatrix} g_{x,p} & 0 \\ 0 & g_{y,p} \end{pmatrix} \quad (g_x, g_y \text{ may be complex})$$

or in scalar form:

$$\begin{aligned} V_{xx,pq} &= g_{x,p} \dot{g}_{x,q} e^{-i\phi_{pq}(I+Q)/2} \\ V_{yy,pq} &= g_{y,p} \dot{g}_{y,q} e^{-i\phi_{pq}(I-Q)/2} \\ V_{xy,pq} &= g_{x,p} \dot{g}_{y,q} e^{-i\phi_{pq}(U+iV)/2} \\ V_{yx,pq} &= g_{y,p} \dot{g}_{x,q} e^{-i\phi_{pq}(U-iV)/2} \end{aligned}$$

## Building a Tree With Gains

- See ME1/demo3-predict-ps-gain.py
- It's trivial to add extra Jones terms to **Meq.MatrixMultiply**.
- The biggest effort is actually figuring out what numbers to plug in.
  - depends on your simulation objectives
  - we'll set up subtrees to compute these
  - could also come from a parameter DB, or from FITS images/tables.
- For now, let's assign a random, time-variable gain-phase to each antenna.



## Time Variability On the Cheap

Something like:  $g = e^{iA \sin(Bt+C)}$

- Remember that we get a “time grid” with each request.
- The **Meq.Time** node returns  $f(t) = t$ , combine it with a **Meq.Sin** node to compute  $A \sin(Bt+C)$
- Generate random  $A, B, C$ 's per dipole (using Python's **random** module)
- Meq.Polar** builds  $x \exp(iy)$
- Run the tree (load up the bookmarks).
- Make a map.
  - note that we can now select a column to image, the new simulation is in DATA, the old one is in MODEL\_DATA.

## Some Performance Considerations

$$\mathbf{V}_{pq} = \mathbf{G}_p \mathbf{K}_p \mathbf{B} \mathbf{K}_q^t \mathbf{G}_q^t$$

- This does  $N_{time} \times N_{freq} \times 4$  individual matrix multiplications.
- $\mathbf{B}$  is constant,  $\mathbf{G}$ 's are variable in time, and  $\mathbf{K}$ 's are variable in time-freq.
- If we reorder the terms as follows ( $\mathbf{K}$  commutes):

$$\mathbf{V}_{pq} = \mathbf{K}_p (\mathbf{G}_p \mathbf{B} \mathbf{G}_q^t) \mathbf{K}_q^t$$

- we end up with  $N_{time} \times 2 + N_{time} \times N_{freq} \times 2$  ops.
- Is this a good idea?

## To Quote Ancient Wisdom...

*“Premature optimization is the root of all evil.”*

-- Donald Knuth

*“More computing sins are committed in the name of efficiency (without necessarily achieving it) than for any other single reason – including blind stupidity.”*

-- W.A. Wulf

- Don't worry about optimizing for calculations until you establish that performance is a problem...
- ...and where the problem lies.
- Jones terms don't always commute, so think carefully before moving them around.
- But if you can reshuffle them, significant CPU savings may in fact result.

## Exercise 1: Instrumental Polarization

- Use ME1/demo3-predict-ps-gain.py as a starting point.
- Make the source unpolarized, with  $l=1jy$ .
- Make the gain amplitudes frequency-dependent (and reset the gain phases to 0):

$$\mathbf{G}_p = \begin{pmatrix} 1 + a_p(\nu - \nu_0) & 0 \\ 0 & 1 - a_p(\nu - \nu_0) \end{pmatrix}$$

- pick random  $a_p$ 's in the range  $[1e-10, 1e-9]$
- For  $\nu_0$ , create placeholder ( $ns.freq0 << 0$ )
- Produce a per-channel map, look at  $Q$  fluxes.

## Exercise 2: Alt-Az Mounts

- A “perfect” instrument has an equatorial mount, i.e. stationary sky.
- With an alt-az mount, the sky rotates relative to each antenna, so we must add a rotation term to our M.E.

$$\mathbf{V}_{pq} = \mathbf{G}_p \mathbf{P}_p \mathbf{K}_p \mathbf{B} \mathbf{K}_q^\dagger \mathbf{P}_q^\dagger \mathbf{G}_q^\dagger$$

$$\mathbf{P}_p = \begin{pmatrix} \cos \gamma_p & -\sin \gamma_p \\ \sin \gamma_p & \cos \gamma_p \end{pmatrix} \equiv \text{Rot} \gamma_p$$

$\gamma_p$ : parallactic angle for antenna  $p$

## Exercise 2: Alt-Az Mounts

- Use ME1/demo3-predict-ps-gain.py as a starting point.
- Use  $l=1$  Jy,  $Q=.2$  Jy
- Insert the gain terms from Exercise 1.
- Add a  $\mathbf{P}$  term to model sky rotation.
- Produce MFS and per-channel maps of  $IQUV$  flux.
- Hint: **Meq.ParAngle** computes the P.A. as a function of time. It expects a **radec** child (phase center), and an **xyz** child (station position).

$$\mathbf{V}_{pq} = \mathbf{G}_p \mathbf{P}_p \mathbf{K}_p \mathbf{B} \mathbf{K}_q^\dagger \mathbf{P}_q^\dagger \mathbf{G}_q^\dagger$$

$$\mathbf{P}_p = \begin{pmatrix} \cos \gamma_p & -\sin \gamma_p \\ \sin \gamma_p & \cos \gamma_p \end{pmatrix}$$

## Exercise 3: UFO

- Use ME1/demo2-predict-ps-image.py as a starting point.
- Make the source move:

$$\begin{pmatrix} l \\ m \end{pmatrix} = \begin{pmatrix} l_0 \\ m_0 \end{pmatrix} + \begin{pmatrix} \dot{l} \\ \dot{m} \end{pmatrix} \frac{v - v_0}{v_1 - v_0} (t - t_0)$$

$$l_0 = m_0 = 0, \quad \dot{l} = \dot{m} = .5' / \text{hour}$$

- For  $t_0$   $v_0$   $v_1$  use placeholder constants: (ns.time0<<0); (ns.freq0<<0); (ns.freq1<<0); the header script will initialize them for you.
- Produce a per-channel map.

## Exercise 4: Amoebas

- Use ME1/demo2-predict-ps-image.py as a starting point, put source at  $l=m=0$ .
- Insert ionospheric phase (**Z** jones) that we produced in Intro1/example5 and exercise 3.
  - use one TID (ampl=.1, 50km, 200km/h)
  - for the x,y ionospheric “positions”, use the station x,y (use a Meq.Composer to form up the xy vector...)
- Bonus points: make inspectors for TECs and **Z**.
- Run script and make a per-channel map.
- Make a time-slice movie:
 

```
glish -l ~/Workshop2007/make_movie.g
MODEL_DATA ms=demo.MS
```

## Bonus Exercise 5: Simulating a Transient

- Simulate a transient source, make an MFS and a per-channel map.
- For bonus points, make the source narrow-band.

