

DEMYSTIFYING NON-LINEAR PA BEHAVIOR SEMINAR

In cooperation with Hi-Tech, Maury Microwave and
Amcad

ROHDE & SCHWARZ
Make ideas real



AGENDA

Time	Session
8:30 – 9:00	Welcome reception
9:00 – 10:30	Load-Pull measurements to improve efficiency of power amplifiers
10:45 – 11:00	Coffee / Tea break
11:00 – 12:00	Transistor / Amplifier Characterization by doing load pull measurements to find best efficiency behavior
13:00 – 14:00	Lunch
14:00 – 15:00	From Modelling to Simulation
15:15 – 16:15	Understand linearity improvement possibilities on a physical amplifier
16:15 – 17:00	Behavioural Modelling based on Amcad 3 tone Method plus verification on measurements

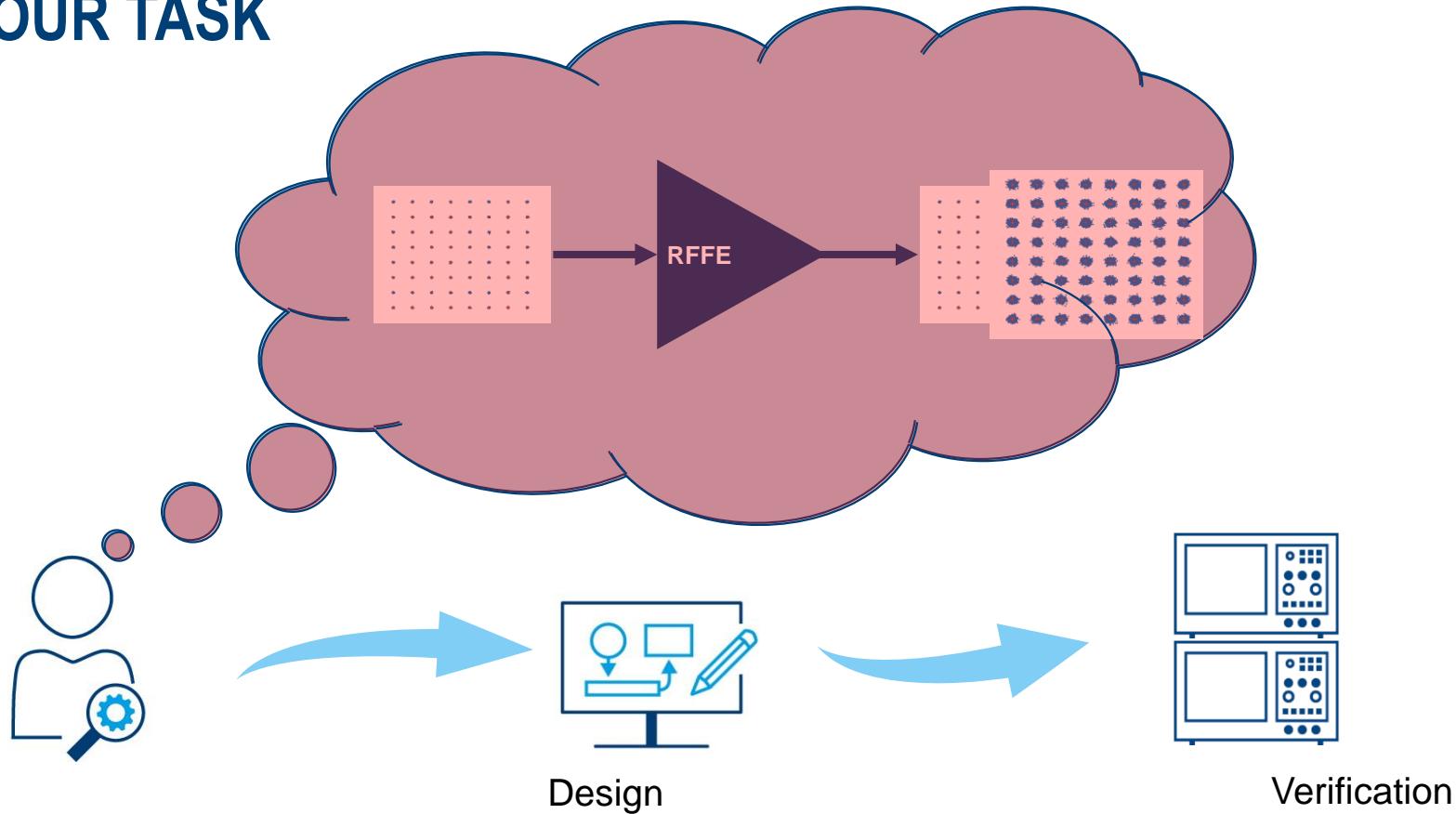


UNDERSTAND LINEARITY IMPROVEMENT POSSIBILITIES ON A PHYSICAL AMPLIFIER

ROHDE & SCHWARZ
Make ideas real



YOUR TASK

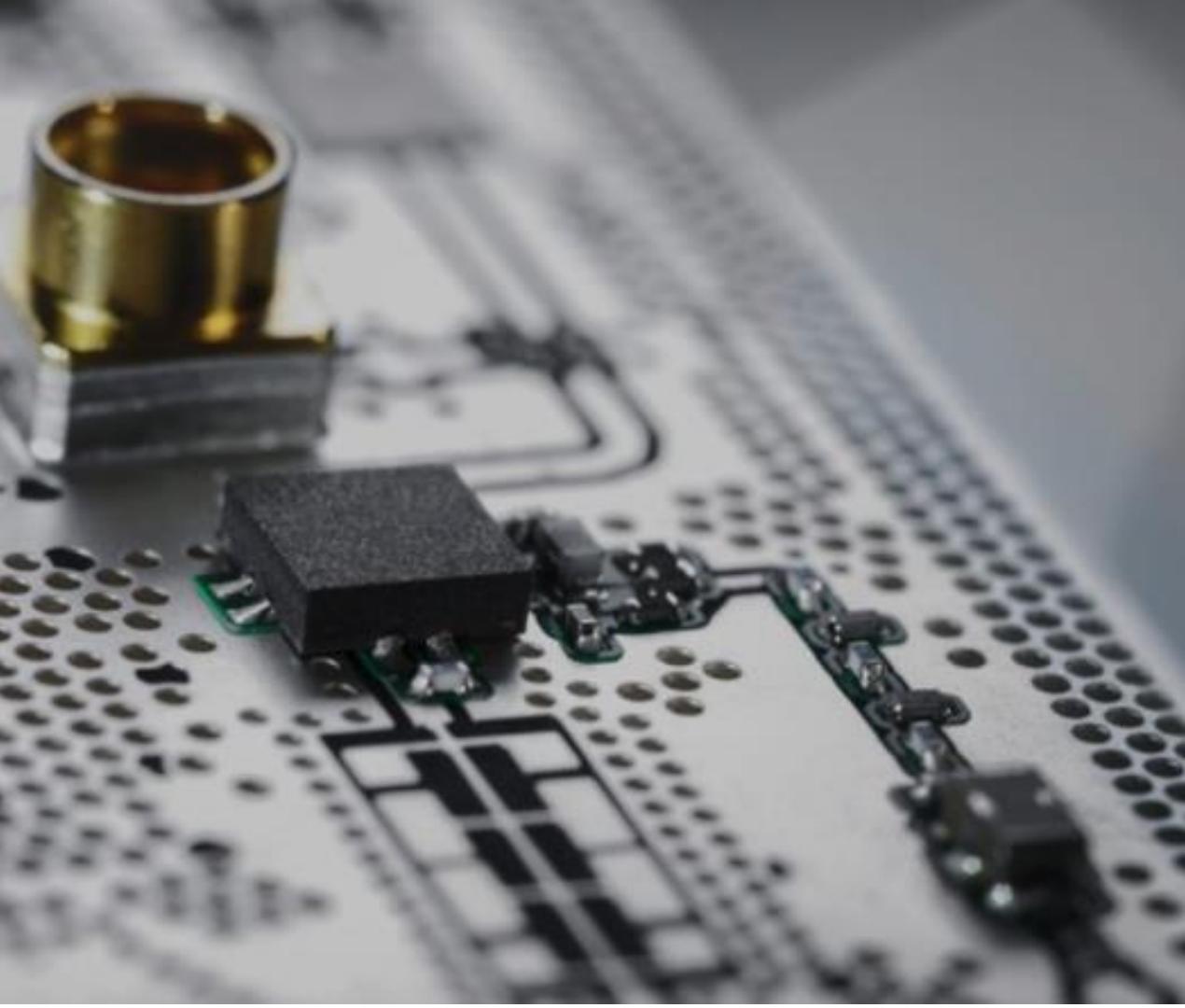


OUTLINE

- ▶ Intro and motivation
- ▶ Effect of linearization
- ▶ Why modulated signals

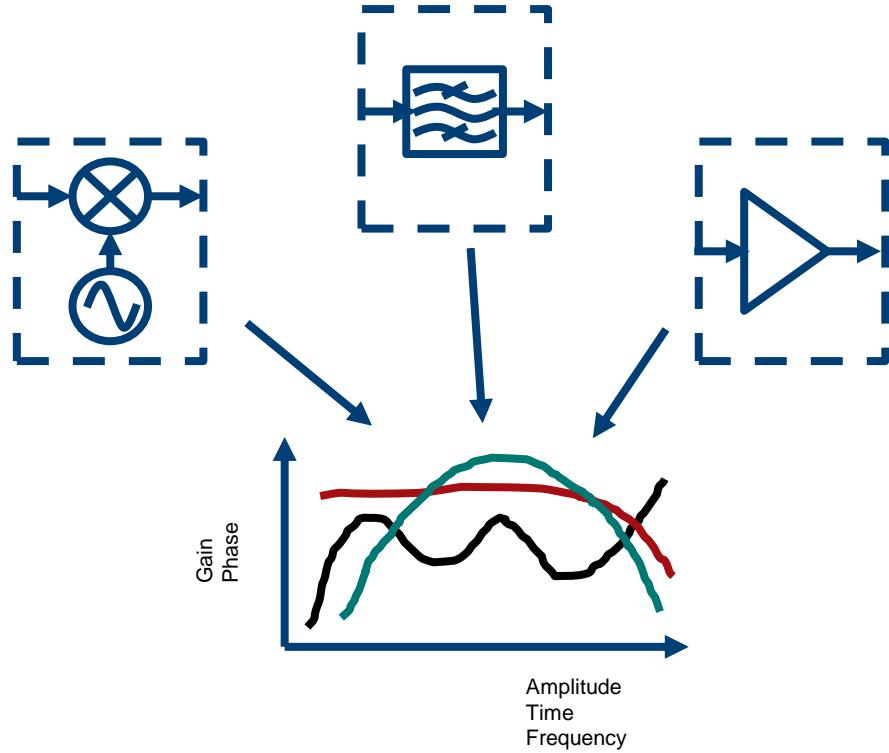
- ▶ Getting to the best possible PA performance
- ▶ Deriving models for linearization

- ▶ Conclusion



UNDERSTANDING DISTORTIONS

- ▶ Distortion limits RFFE performance
- ▶ Distortions might generally be defined as variations in complex gain (amplitude and phase) in three domains:
 - Amplitude (e.g. non-linear distortion)
 - Frequency (e.g. linear distortions)
 - Time (e.g. memory effects)
- ▶ All RFFE components demonstrate all the distortions, in varying proportions:
 - **Mixers** and **Amplifiers** often contribute most to **non-linear** and **memory effect** distortions
 - **Filters** often contribute the most **linear** distortion
- ▶ Distortion reduction is called **Linearization**

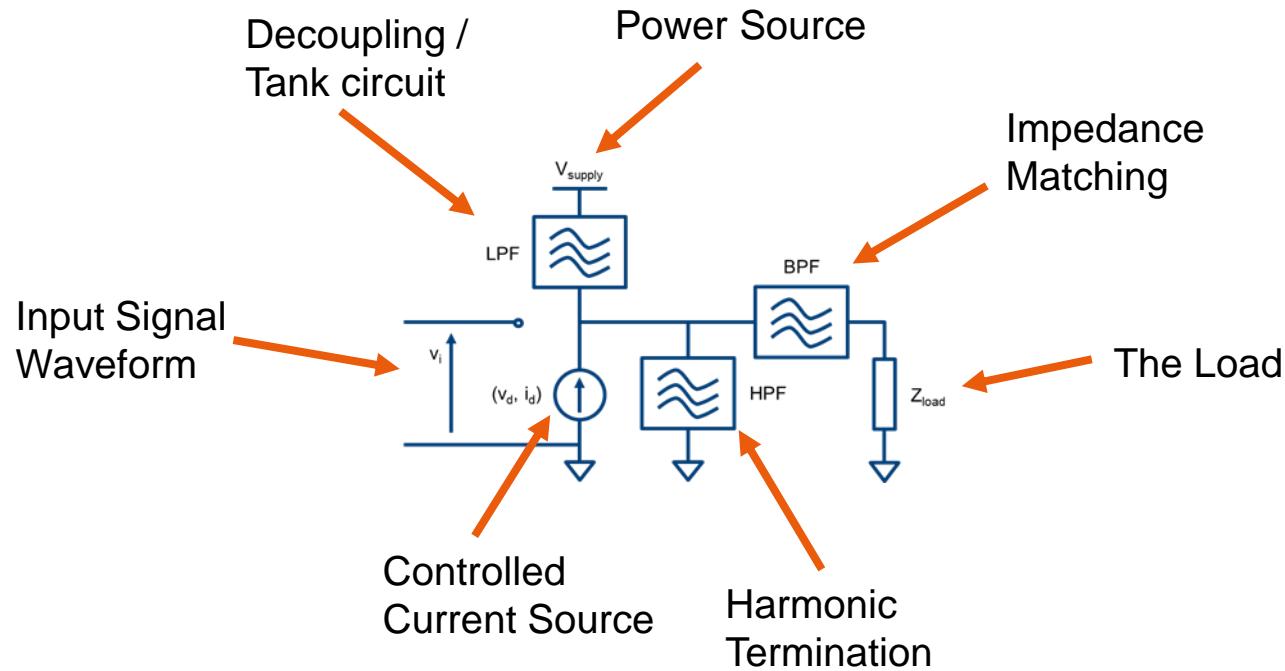


WHY LINEARIZATION?

- ▶ Challenging RF signals on RF frontends
 - 5G in mmWave and RF, mMIMO, beamforming, increasing bandwidth, higher order modulations, digital payloads, wideband Electronic Warfare (EW)
- ▶ Significant power consumption is in the RF Front-End (RFFE)
 - Operating close to saturation offers best energy efficiency
 - Technologies such as GaN absolutely require digital predistortion for linear operation
- ▶ Various PA topologies studied
 - Doherty, Load Modulated Balanced Amplifier (LMBA), Outphasing, ...
- ▶ PA gains in efficiency but is highly non-linear
→ Linearization is a *MUST*

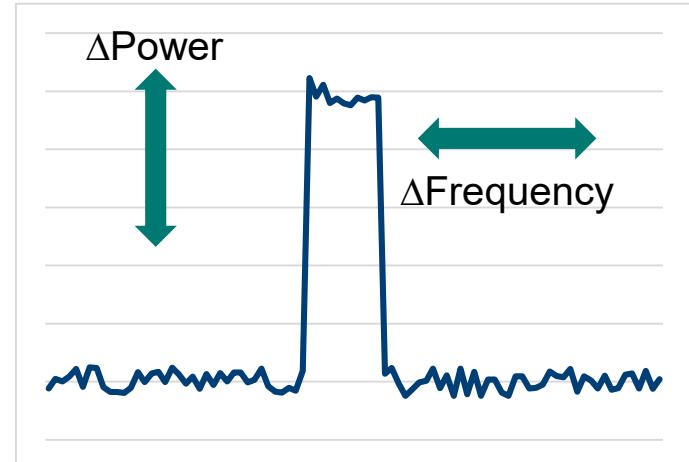
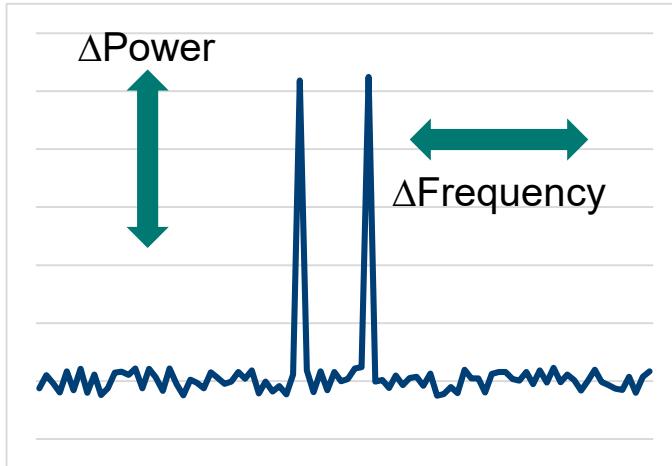


THE RF AMPLIFIER BUILDING BLOCK



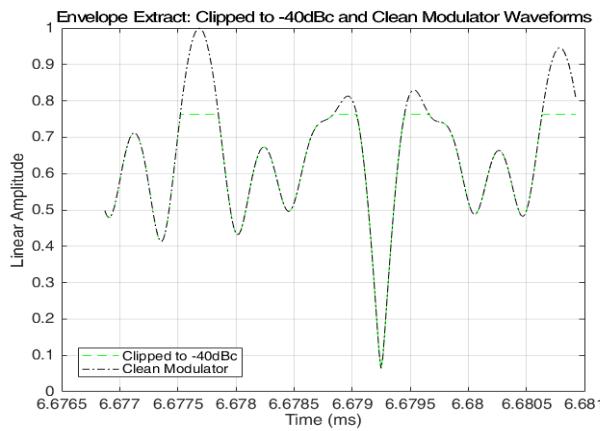
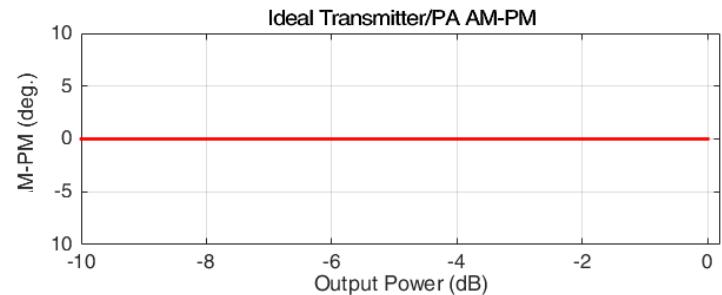
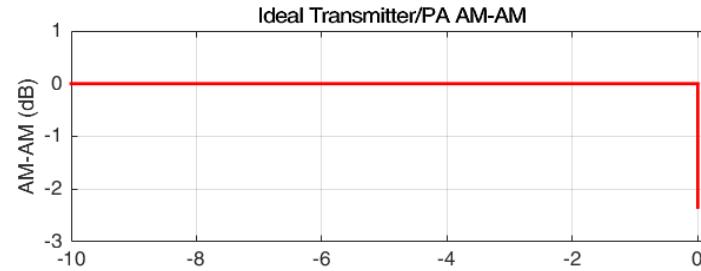
WHY MODULATED MEASUREMENTS

- ▶ Traditional approach – VNA with CW measurements
- ▶ CW signals do not accurately represent modern signals
- ▶ Alternative approach – use the same signal that will be amplified
- ▶ Modern signals are wider BW's, higher crest factors
 - Consider integrated power in channel, overshoot in symbol transitions, etc.



WHY LINEARIZATION

- ▶ Dreamland: Ideal amplifier with “hard limiter” response
 - AM/AM: Brick-wall behavior
 - AM/PM: eliminated – No distortion
 - linear system, just cutting of peaks through hard limiter



WHY LINEARIZATION

- ▶ Two areas of interest:
 - compression
 - memory effect

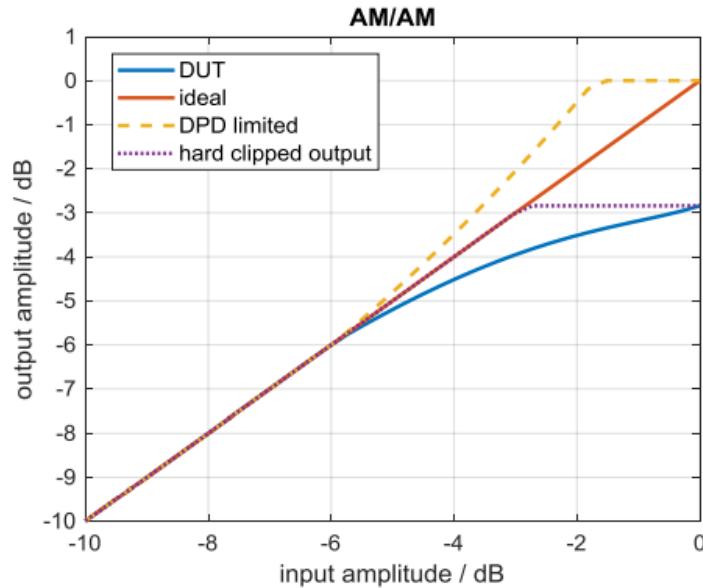
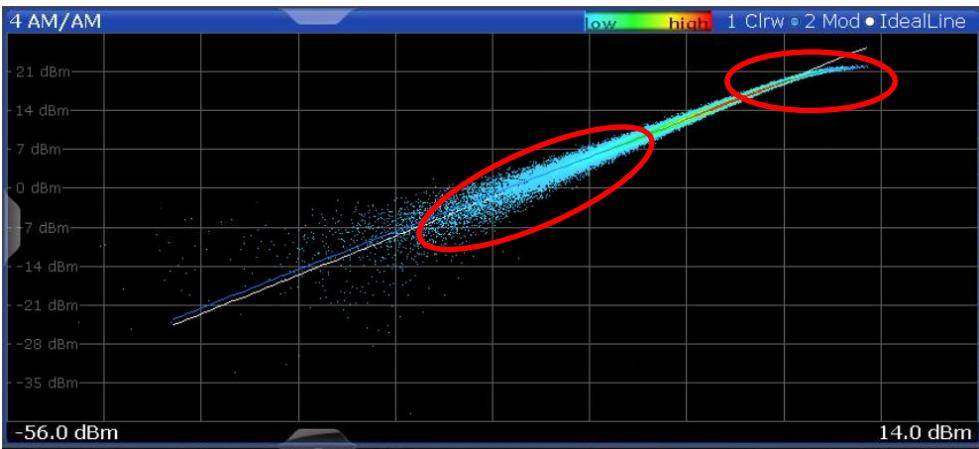
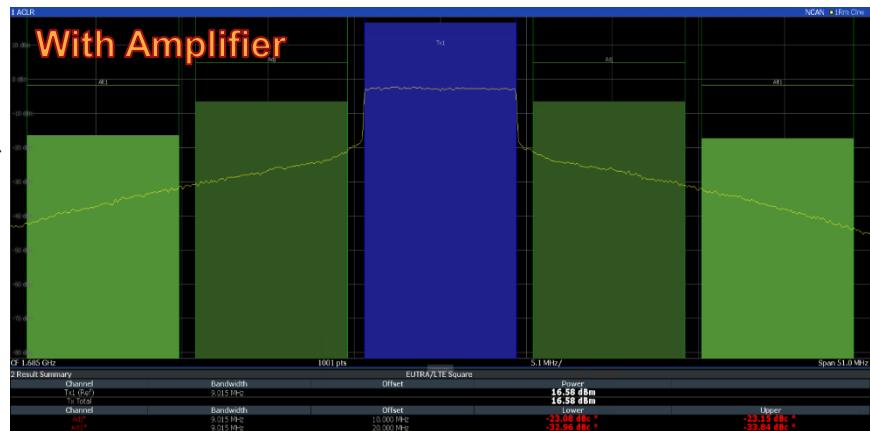
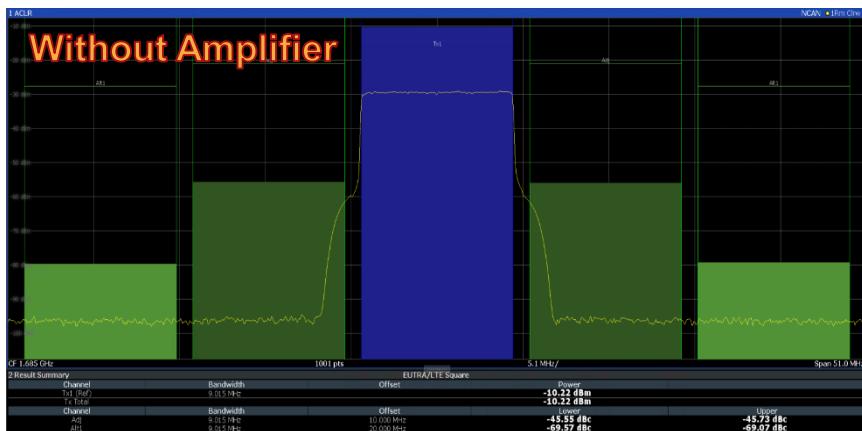


Figure 4 Overview plot: measured AM/AM, ideal output, pre-distorted input signal, and target output signal (hard clipped)

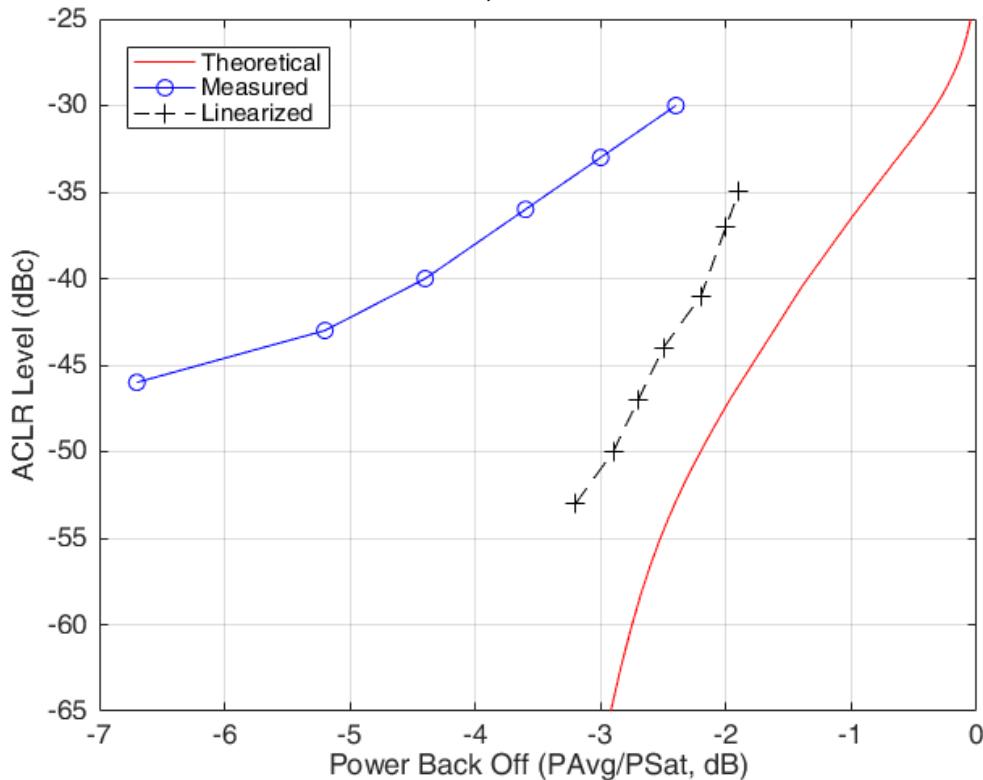
WHY LINEARIZATION

- ▶ ACLR measurements determine the channel power and adjacent channel power
- ▶ Amplifiers can cause spectral regrowth to occur in adjacent channels resulting in more power



LINEARIZATION: MEASUREMENT EXAMPLE

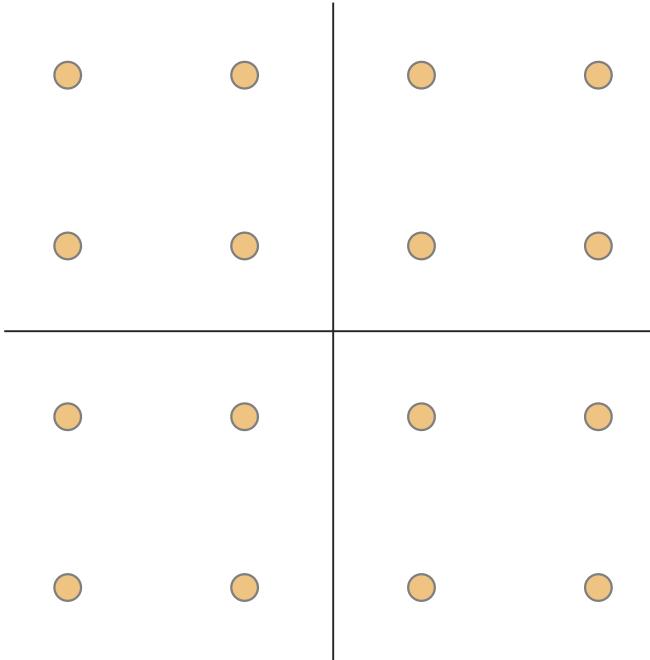
ACLR vs PAR: Measured, linearized and theoretical limit



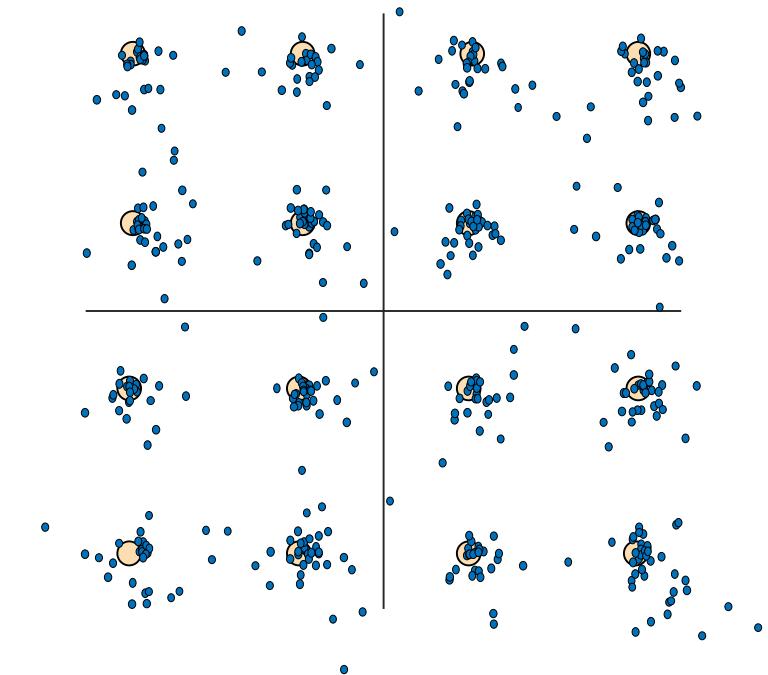
- ▶ In this specific example, digital predistortion (DPD, memoryless, polynomial) is applied to a device used a mobile front-end, with a standard test signal.

ERROR VECTOR MAGNITUDE

- Ideal constellation diagram

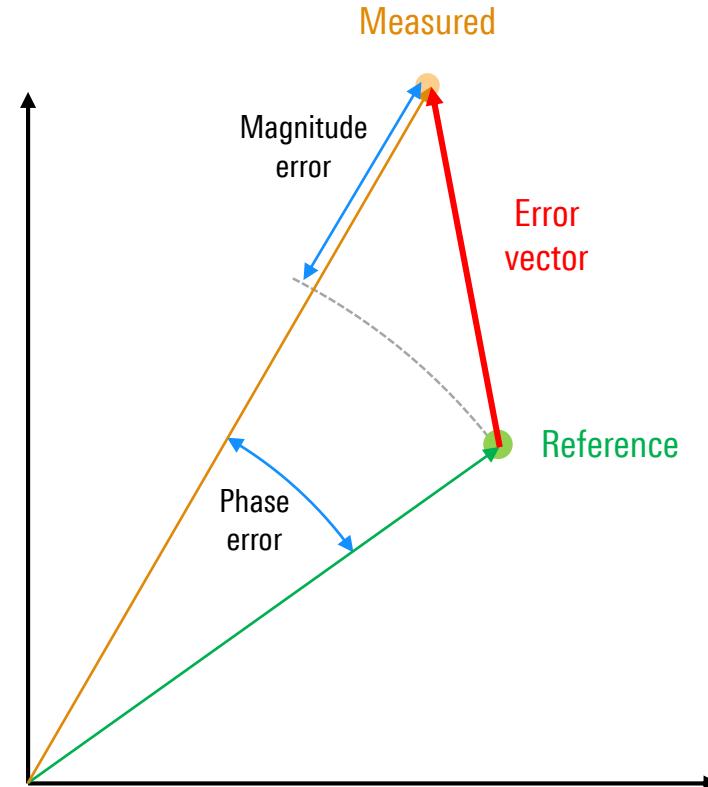


- It might look different after going through the PA

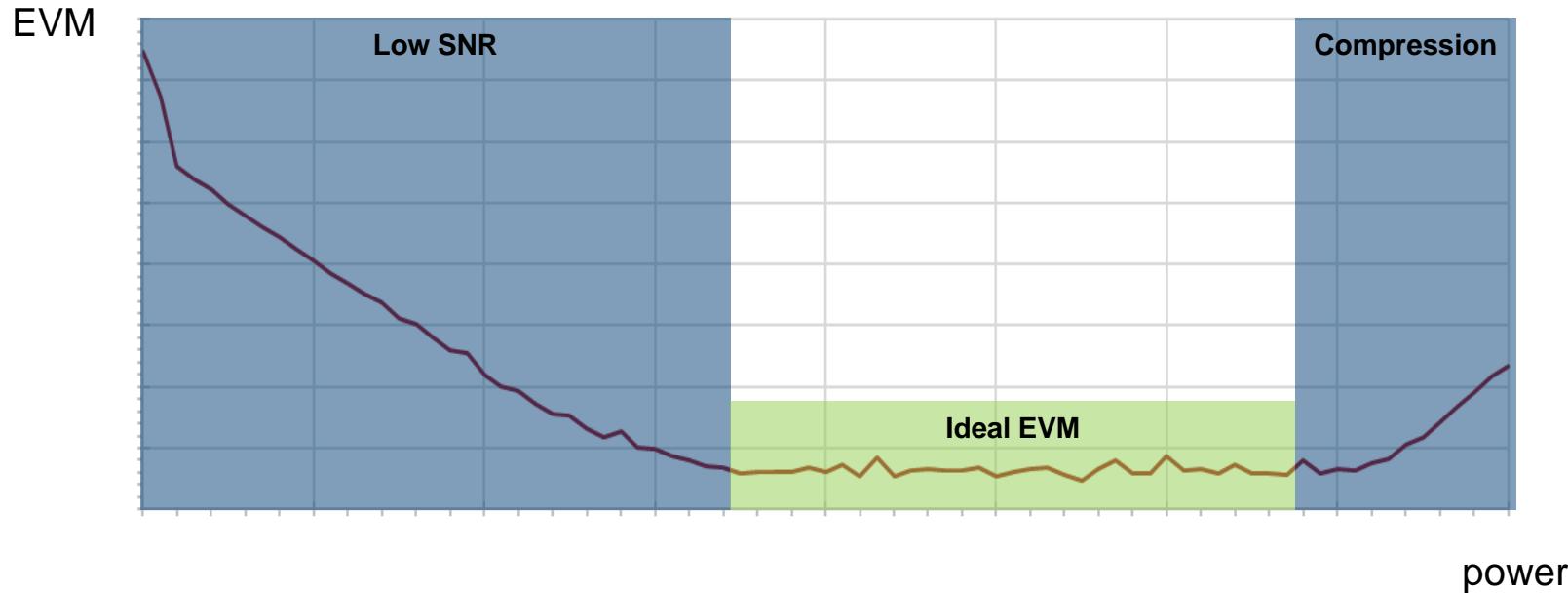


ERROR VECTOR MAGNITUDE

- ▶ Error vector: difference between ideal constellation point and actual sample
- ▶ EVM too high → BER is increasing
 - Higher modulation scheme → lower EVM required
- ▶ EVM: FOM for inband signal performance
 - Compression, non-linearity
 - Noise (low SNR)
 - Frequency response
 - Inter-symbol interference



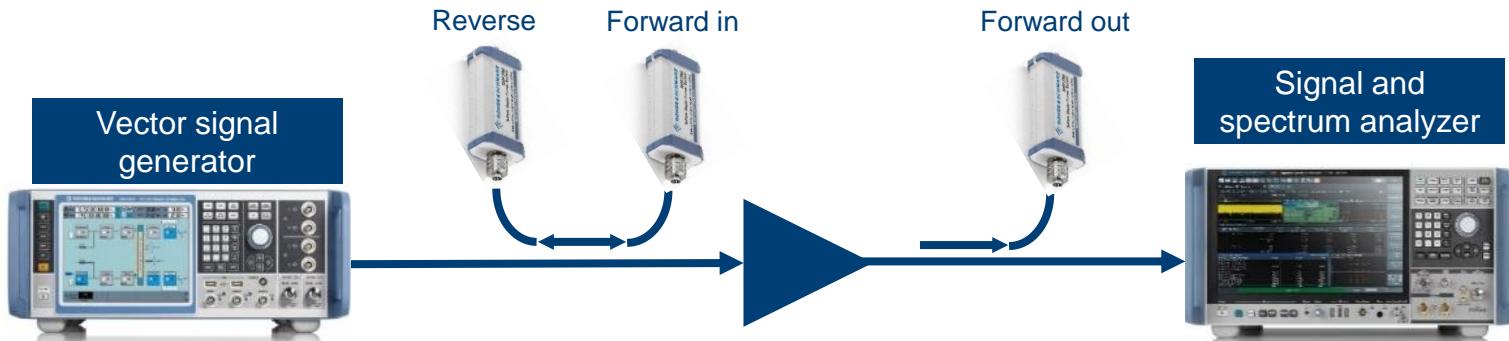
ERROR VECTOR MAGNITUDE



HARDWARE VERIFICATION

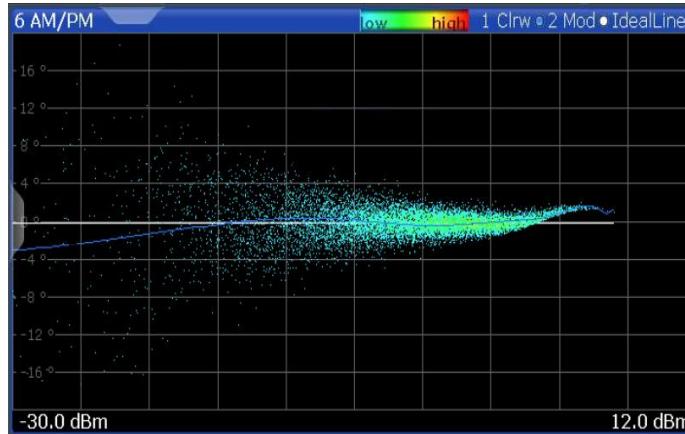
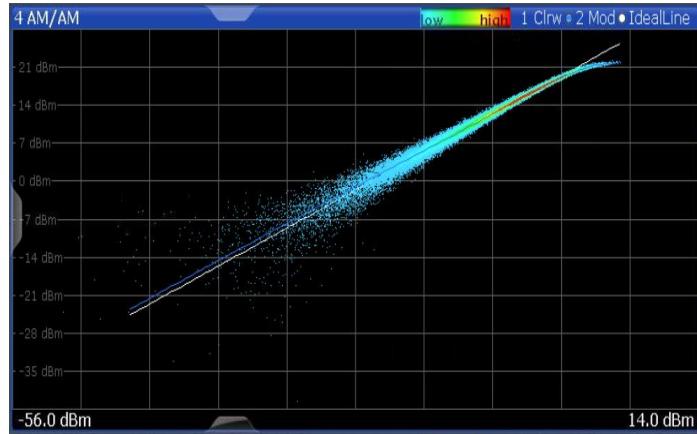
MODULATION TESTS

- ▶ Channel power
- ▶ Adjacent Channel leakage ratio, ACLR / ACP
- ▶ Modulation performance: EVM
- ▶ Distortion: AM/AM and AM/PM



HARDWARE VERIFICATION DISTORTION

- ▶ Linearity of amplitude and phase through the amplifier
- ▶ Using signal fidelity from modulation
- ▶ Variance gives indication on memory effect based distortion



GOAL: HOW GOOD CAN A PA BE?

- ▶ DPD is used in real systems to optimize the PA performance
- ▶ DPD is a specialty of each system manufacturer and the “secret sauce” in between vendors
- ▶ PA manufacturer has no access to these sometimes significant size DPD teams
- ▶ Looking for an easy way to understand how good their devices can be
- ▶ **Direct DPD** is offering this capability
 - Iterative approach
 - Compares ideal input signal to received distorted signal and calculates a new pre-distorted signal on a sample-by-sample base
 - Takes care of non-linearity, memory effect, distortion
 - Provides insight to what can be reached



AMPLIFIER OPTIMIZATION

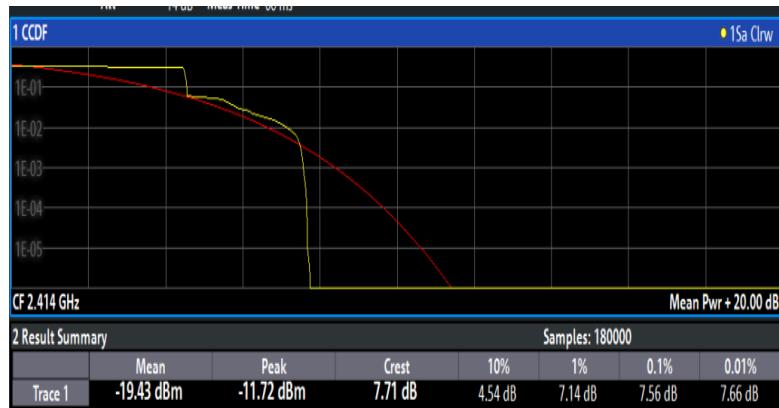
OPTIMIZATION OPTIONS

- ▶ Optimization is very much dependent on application
 - Maximum output power for CW or pulses
 - Linearity for best modulation capabilities
 - Efficiency for lowering power consumption and thermal management
- ▶ Different use cases ask for different optimization
- ▶ Optimization: mixture between simulation and measurement aided development
- ▶ Waveform engineering with different classes of amplifiers: class A, AB, B, C, ...
 - ▶ User-defined CFR
 - ▶ Different topologies
 - Envelope tracking
 - Doherty amplifier
 - Load Modulated Balanced amplifier
 - ▶ Linearization
 - Digital predistortion

OPTIMIZATION OPTIONS

CFR

- ▶ Crest factor: Ratio peak to average
- ▶ Rare peaks ask for large back-off
- ▶ Issues:
 - Compression creates intermodulation expanding spectrum of signal
 - In DAC: back-off limits resolution of RMS values → quantization noise
 - both increase ACP
- ▶ CFR:
 - Defined clipping **and** filtering
 - Manage harmonic distortion



OPTIMIZATION THROUGH DPD

- ▶ Pre-distort signal to compensate DUT characteristics
- ▶ Close to compression: Efficiency ↑ but non-linearity ↑
→ Linearization is a MUST
- ▶ PA designer: need understanding of system level performance with ideal predistortion on EVM and ACLR
 - Iterative Direct DPD provides this information

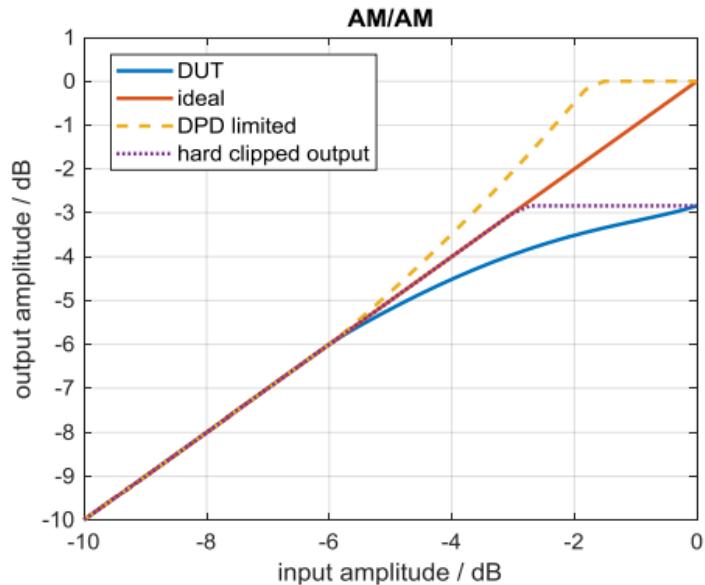
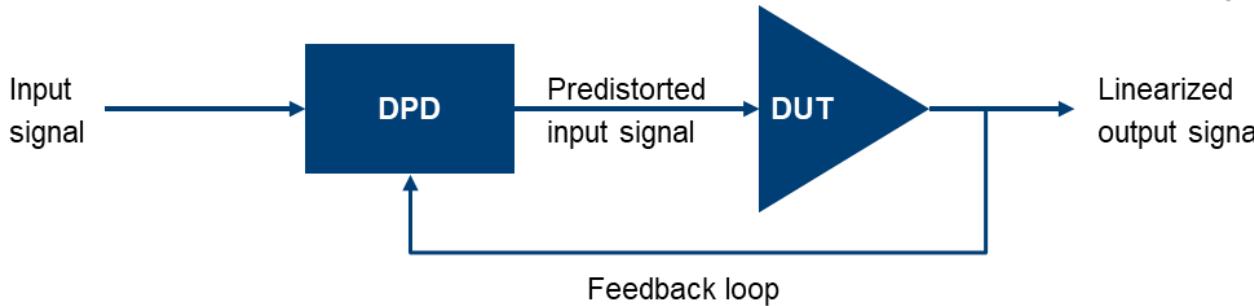
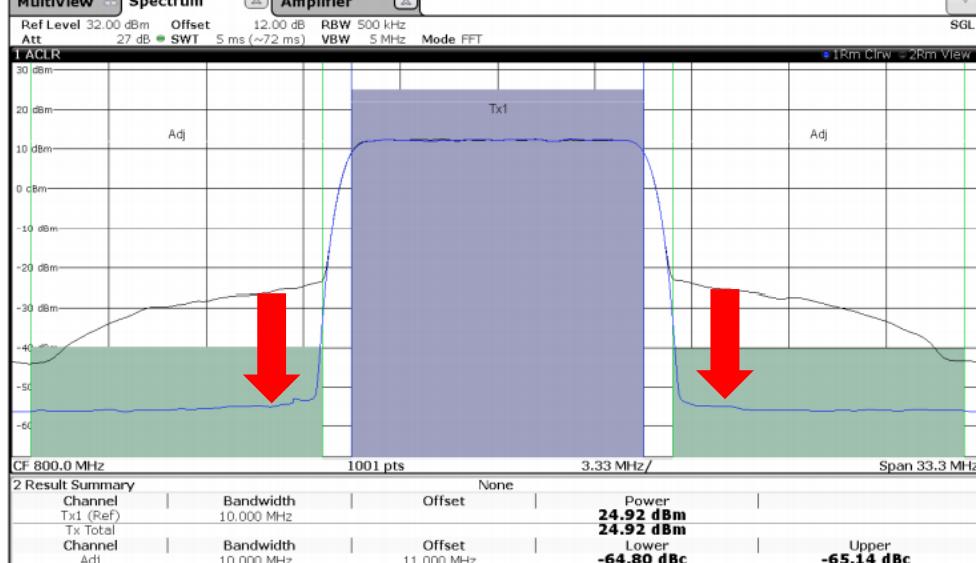
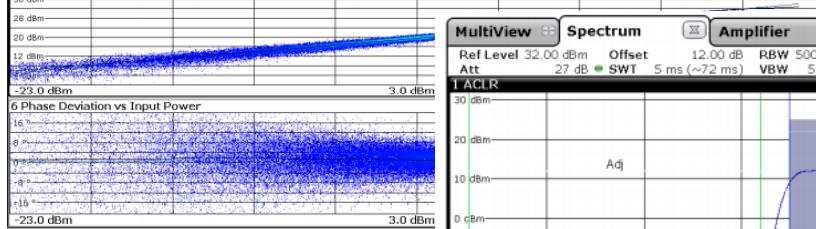
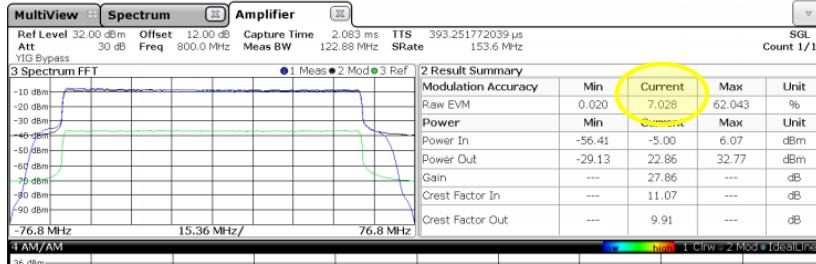


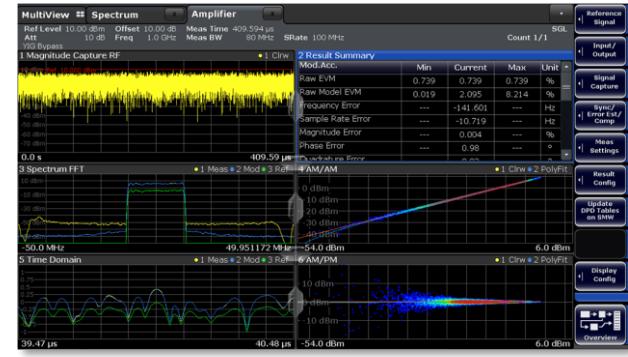
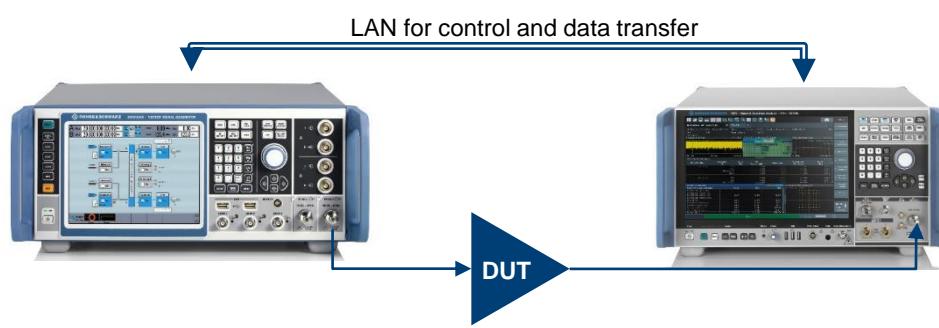
Figure 4 Overview plot: measured AM/AM, ideal output, pre-distorted input signal, and target output signal (hard clipped)

OPTIMIZATION EXAMPLE THROUGH DPD



Understand linearity improvement possibilities on a physical amplifier

DIGITAL PREDISTORTION ON A REAL POWER AMPLIFIER

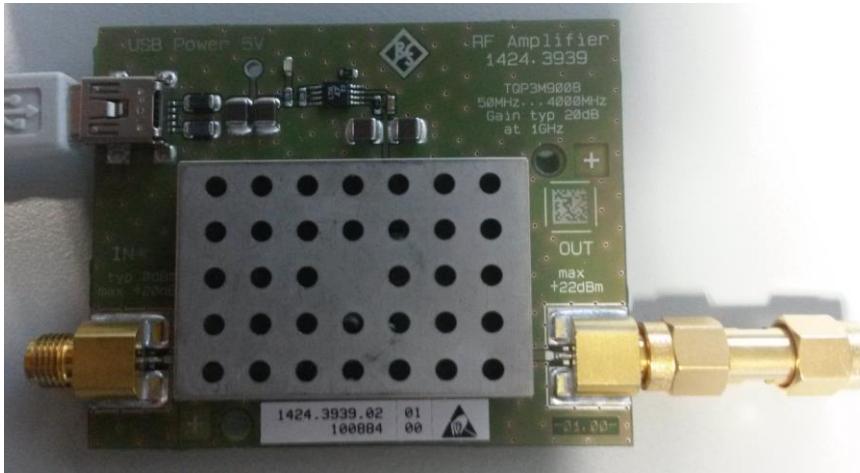


R&S®FSW-K18D Direct DPD

- Iterative approach
- Compensates for memory effects
- Excellent performance especially for amplifiers with memory effects

- Reference for best possible
 - Suppliers typically do not have access to DPD algorithms used by system integrators

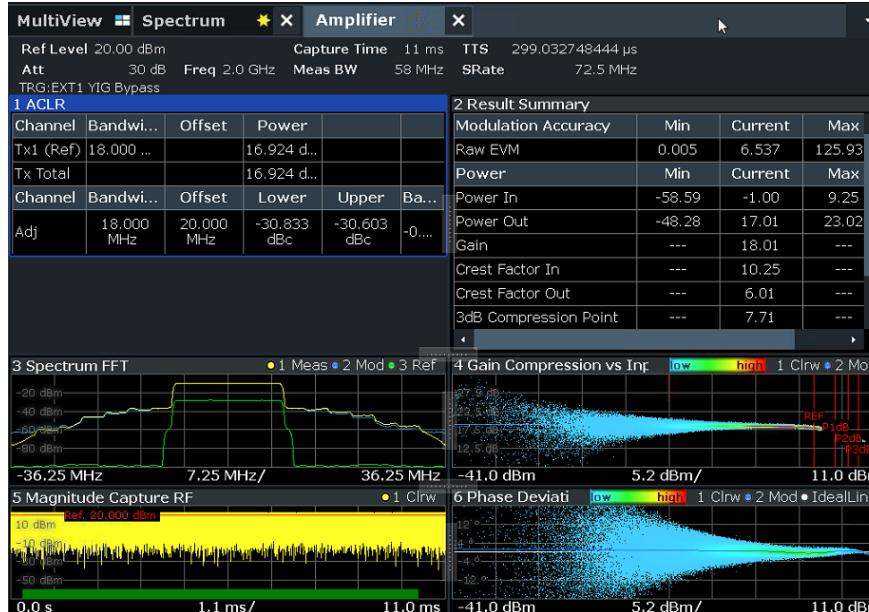
DEMO PA



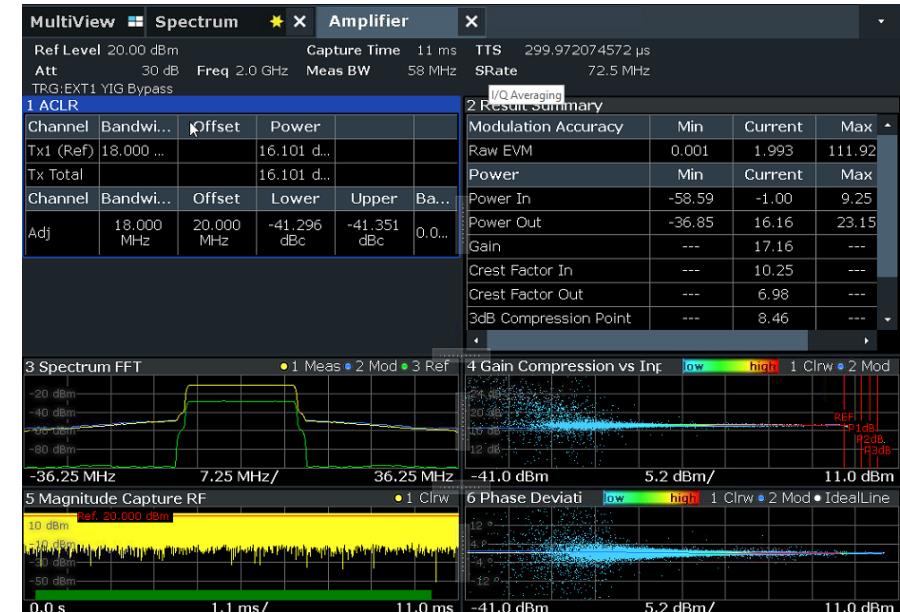
- ▶ 20 MHz OFDM signal
- ▶ 2 GHz
- ▶ Generator power: -1 dBm

RUNNING DPD

Before DPD

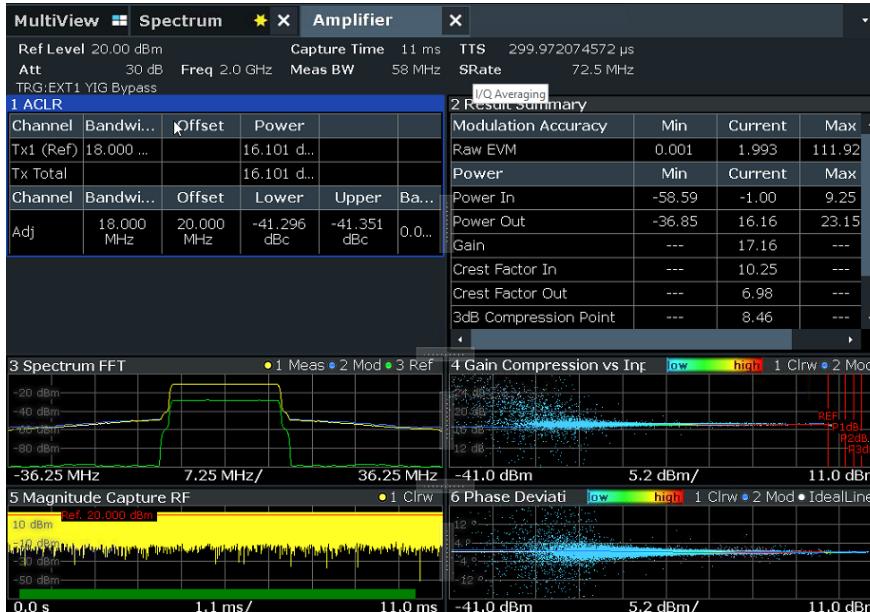


Run K18D, 0 dB Gain Expansion, 10 iterations (default setting)



GAIN EXPANSION

Run K18D, 0 dB Gain Expansion, 10 iterations (default setting)



5 dB Gain Expansion, 2 adjacent channels in ACLR

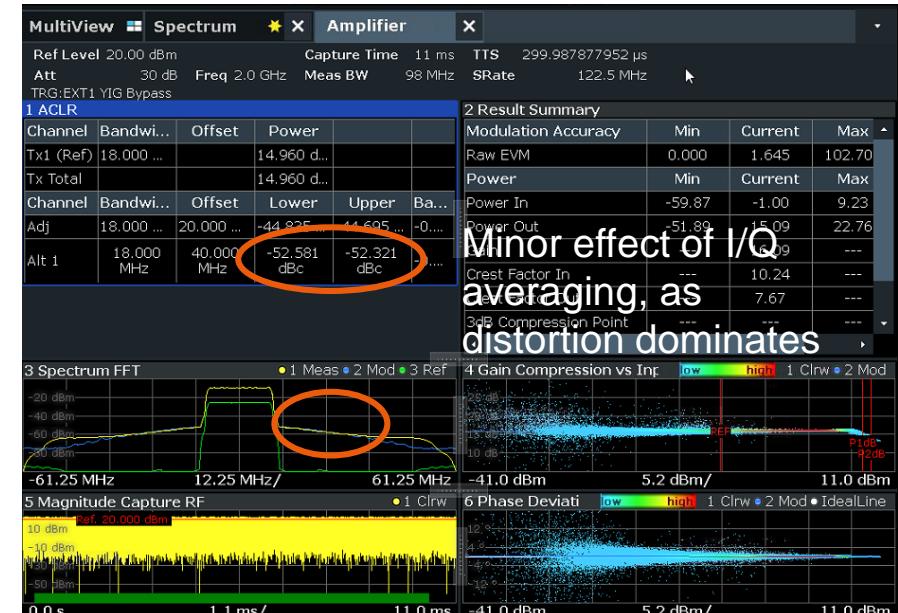


GAIN EXPANSION PLUS I/Q AVERAGING

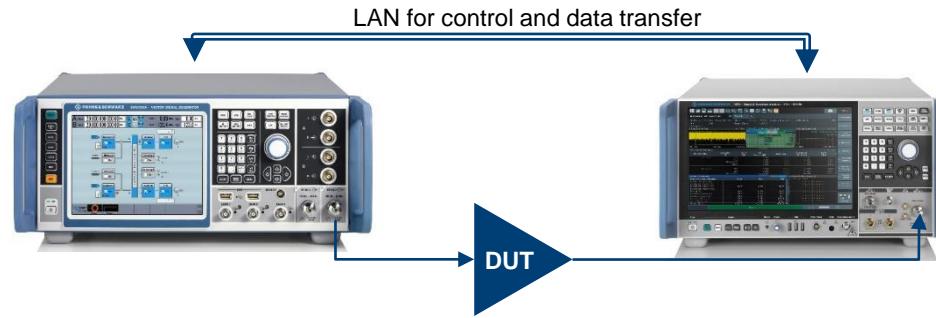
5 dB Gain Expansion, 2 adjacent channels in ACLR



5 dB Gain Expansion, 10 I/Q Averages, 2 adjacent channels in ACLR

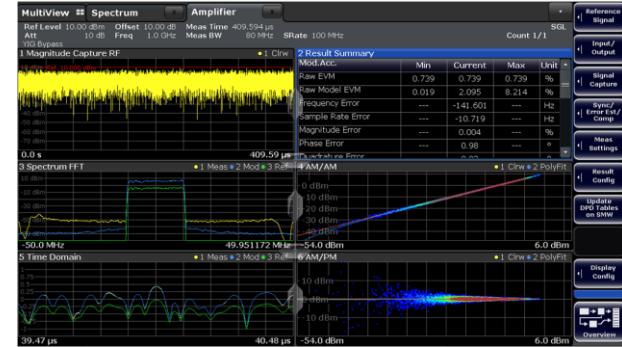


CREATING A DPD MODELL



R&S®FSW-K18D Direct DPD

- Iterative approach
- Compensates for memory effects
- Excellent performance especially for amplifiers with memory effects
- Reference for best possible
 - Suppliers typically do not have access to DPD algorithms used by system integrators



R&S®FSW-K18M memory polynomial

- Memory polynomial model or Hammerstein model based on Direct DPD result
- Modeling can be adopted in order and memory depth
- Model verification on DUT
- Proves easy linearization of RFFE solution

MEMORY POLYNOMIAL MODEL

- ▶ Derive an algorithm based memory DPD, as described in Application Note [1EF105](#)
- ▶ We use a memory polynomial DPD

$$\tilde{P}(nT) = \sum_{p=1}^P \sum_{m=1}^M k_{p,m} A(nT - \tau_m) |A(nT - \tau_m)|^{p-1}$$

- ▶ We use the result of K18D to directly derive the coefficients, rather than modeling the DUT and inverting the model

HAMMERSTEIN MODEL

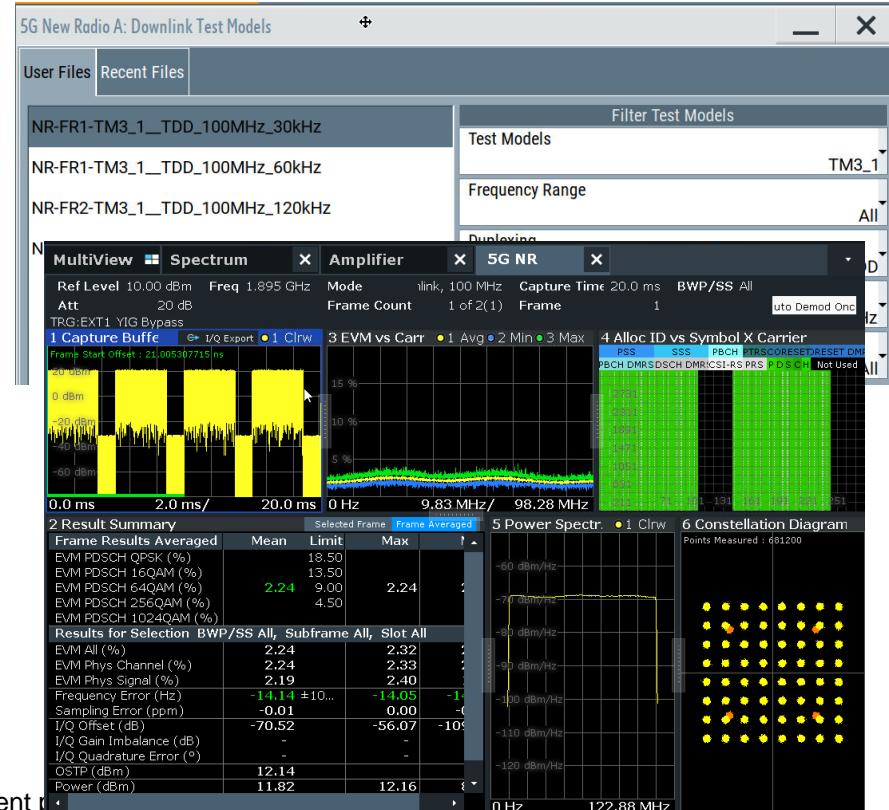
- Predistortion according to the Hammerstein model, is applied to the IQ sample stream by first applying a non-linear polynomial, followed by a convolution



- Easier to be applied in real-time to any IQ stream
- Much less complex → less power needed to apply
- But a bit less efficient in EVM & ACLR improvement

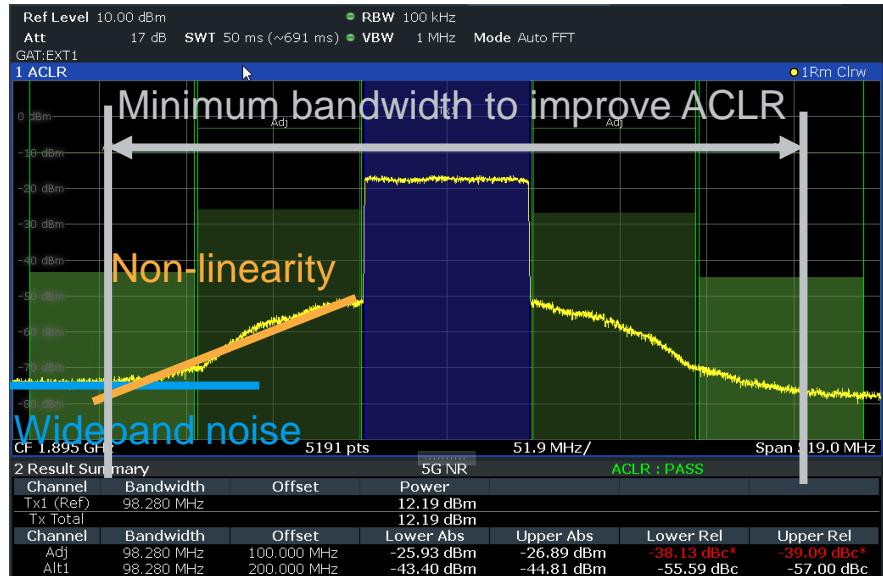
CHARACTERIZE DUT AND DERIVE DIRECT DPD

- Either use the real-world signal or
- Or use „Generate Own Signal“ with a bandwidth and crest factor similar to your real-world signal
- Example: 5G FR1, TDD, 100 MHz, 30 kHz



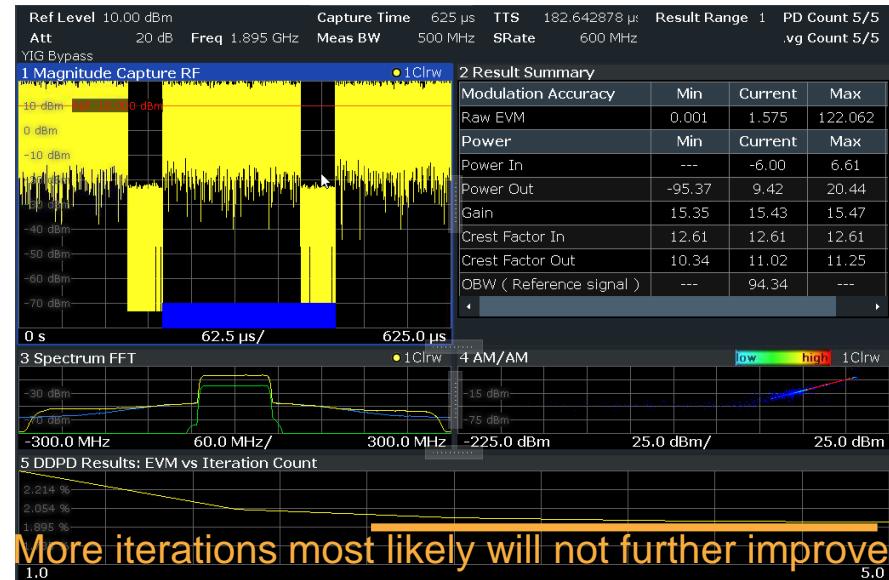
REQUIRED DPD BANDWIDTH

- ▶ Significant ACLR that we need to correct
- ▶ We'll need 4-5 x TX bandwidth for DPD
- Huge capture to process – maybe one want to use a much shorter representative signal



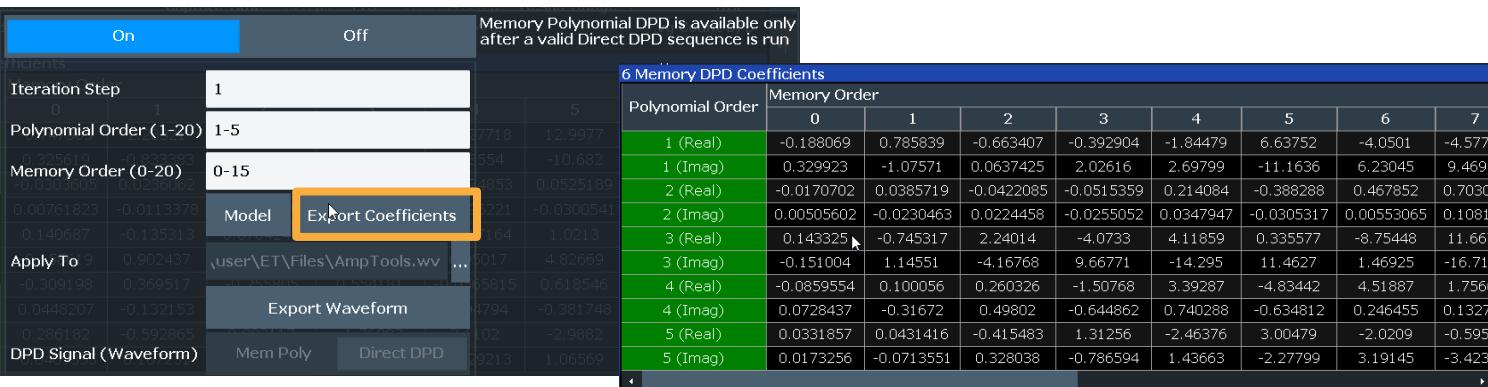
EVALUATE DIRECT DPD RESULT

- ▶ Have a look at the DDPD Results window – if it flattens out to the right, more iterations will most likely not make it better
- ▶ Double check if noise affects the K18D result, if so – increase I/Q averaging count

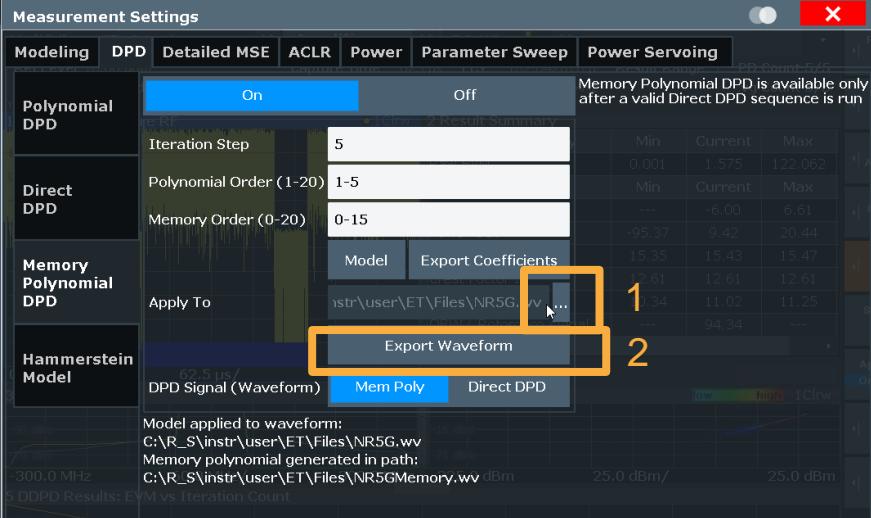


MODELLING PROCESS

- ▶ Iteration step: last iteration is typically best, but check Direct DPD Result window for best iteration
- ▶ Memory Depth: may require a bit of experimenting, high oversampling typically requires more memory depth

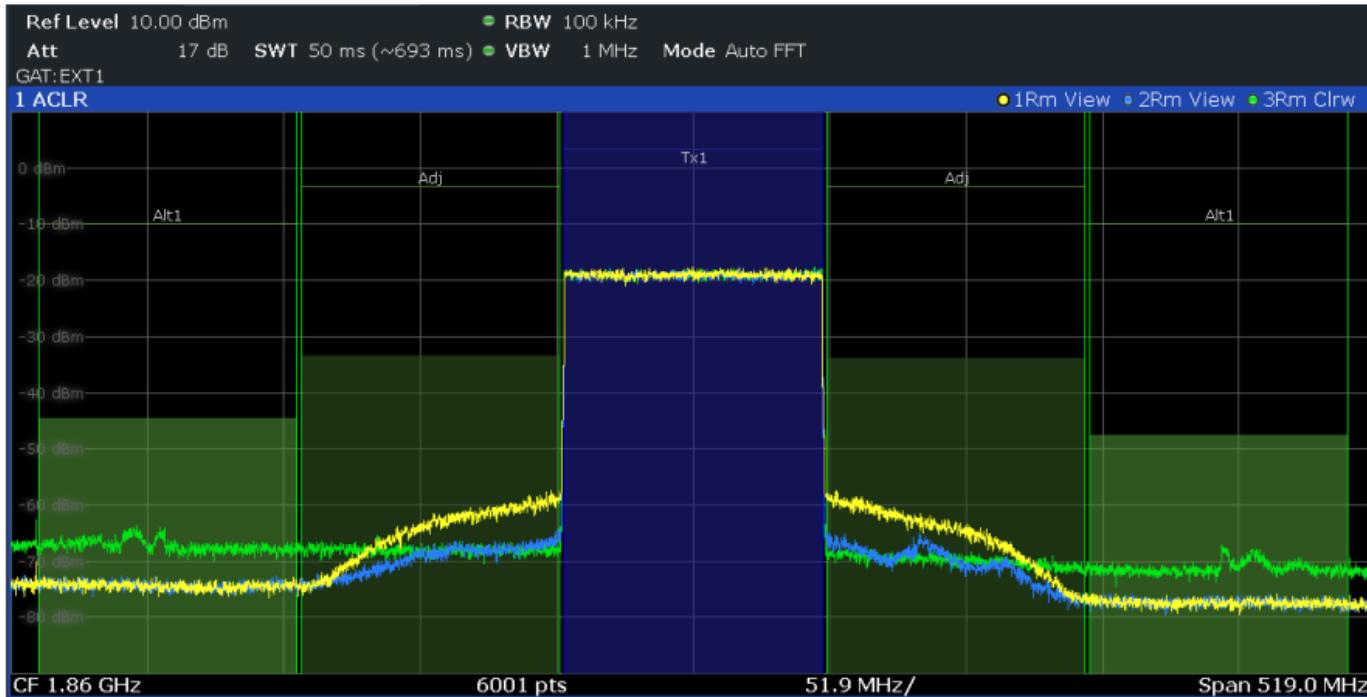


REALITY CHECK: APPLY CREATED MODEL TO ORIGINAL SIGNAL



COMPARISON OF MODELS

- Green: Memory Polynomial
- Blue: Hammerstein Model
- Yellow: reference w/o any DPD

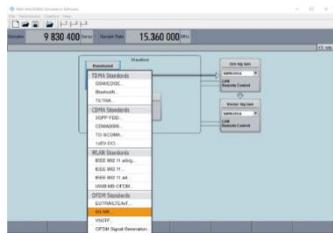


COMPARISON OF MODELS

Predistortion Approach	Measurement Time	EVM Improvement (In-band)	ACLR Improvement (Out-of-band)
Polynomial Model	✓ ✓ ✓	✓	✓
Direct DPD (with Meas Bandwidth = Signal Bandwidth)	✓ ✓ ✓	✓ ✓	✓
Direct DPD (with increased Meas Bandwidth)	✓ ✓	✓ ✓	✓ ✓
Direct DPD (with increased Meas Bandwidth <u>and IQ Averaging</u>)	✓	✓ ✓ ✓	✓ ✓ ✓
Memory Polynomial Model	✓ ✓	✓ ✓	✓ ✓
Hammerstein Model	✓ ✓	✓ ✓	✓

DESIGN: USING EDA TO PIN OUT EXPECTED PERFORMANCE WITH DPD

- ▶ Simulate as close to reality for risk mitigation
- ▶ Use Microwave Office for DUT with E-PHD or Cadence TDNN approach



R&S WinIQSIM2
Signal Generation

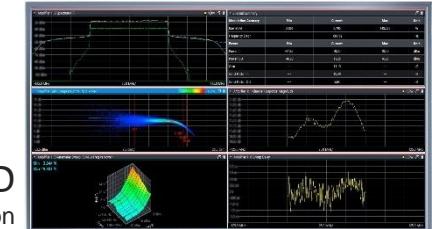


Cadence Visual System
Simulator (VSS)
RF Design/Analysis

R&S VSE
Signal Analysis

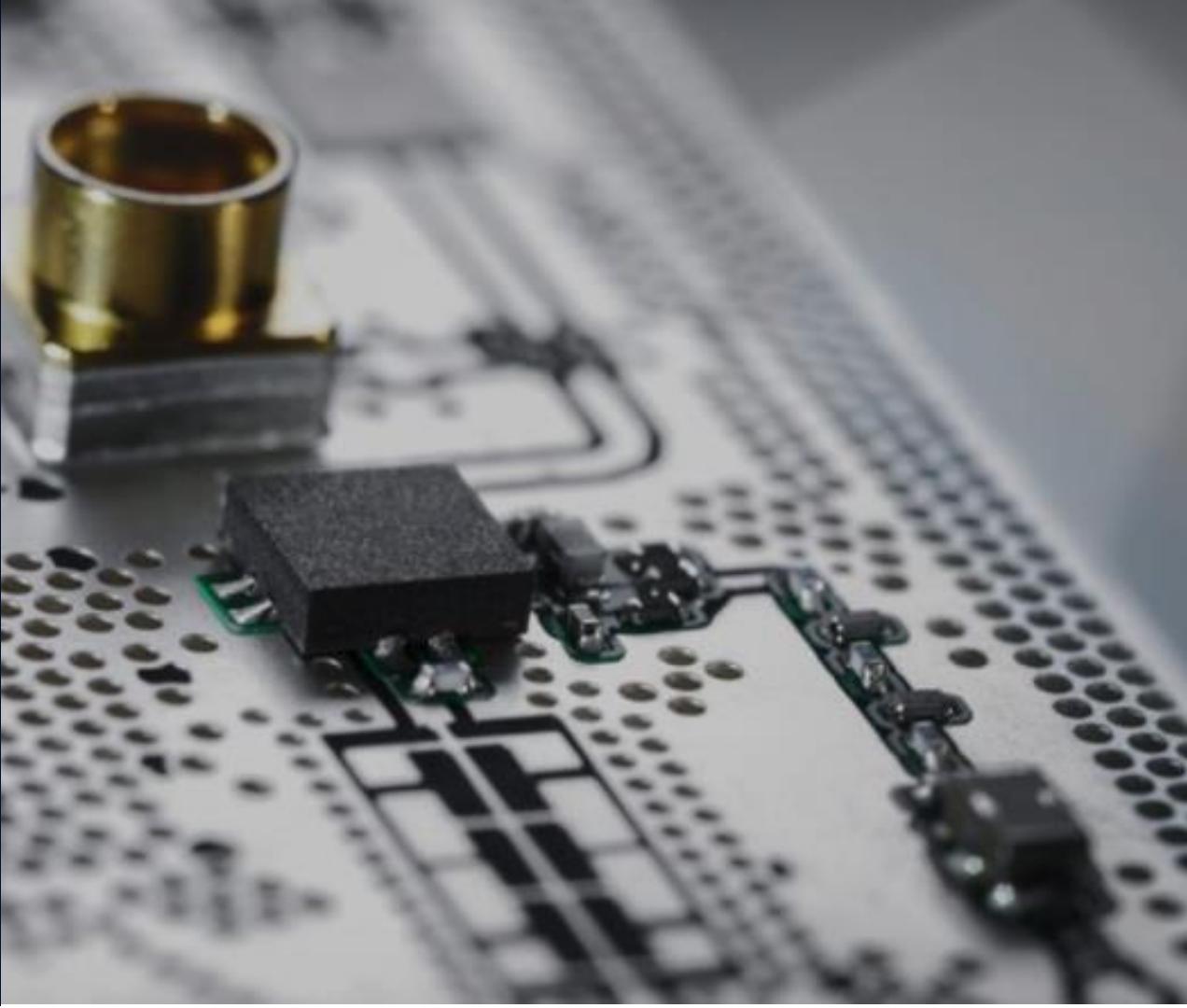


Direct DPD
Linearization



CONCLUSION

- ▶ There is an easy way to understand what is possible
- ▶ Works with any non-linear device and any signal
- ▶ Various models can be derived
- ▶ Works with physical hardware and even in EDA while design



Find out more

RF POWER AMPLIFIER TESTING | ROHDE & SCHWARZ (ROHDE- SCHWARZ.COM)

ROHDE & SCHWARZ

Make ideas real

