Topic 1: A Better UDP

Carson Clanton Science and Engineering The University of Texas at Arlington Arlington, Texas 76019 clanton@uta.edu

Kevin Hayes Department of Computer Department of Computer Science and Engineering The University of Texas at Arlington Arlington, Texas 76019 kevin.hayes@mavs.uta.edu

William Hunter Department of Computer Science and Engineering The University of Texas at Arlington Arlington, Texas 76019 whunter@mavs.uta.edu

Richard Meth Department of Computer Science and Engineering The University of Texas at Arlington Arlington, Texas 76019

Ryan Nordeen Department of Computer Science and Engineering The University of Texas at Arlington Arlington, Texas 76019 richard.meth@mavs.uta.edu ryan.nordeen@mavs.uta.edu

Abstract—This article describes an approach to improving the reliability of UDP. Our approach uses sequence numbers for datagrams to force in-order delivery to the destination network application layer and error-correcting codes to help reproduce lost datagrams without requesting retransmission of these missing datagrams. The level of data reconstruction will be largely dependent on the error-correcting code chosen and the selection of its associated redundancy parameters.

I. INTRODUCTION

When a software engineer begins designing a network application, he or she has to make a decision between two transport layer options: TCP or UDP [1]. There are advantages and disadvantages to both. With UDP comes nimbleness and low overhead, along with dropped and possibly out-of-sequence datagrams. With TCP comes reliable connection-oriented stream-based communication at the expense of throughput [1]. There is not much middle ground between the two. The approach discussed in this paper attempts to change that.

We do not attempt or intend to redesign either protocol. Instead, we seek to enhance the reliability of UDP, so that it is a little more like TCP, while remaining quick and connectionless. To do this, we will require two added features to be built "on top" of UDP: Sequence Numbers and Error-Correcting Codes.

The introduction of sequence numbers should be somewhat obvious; on the receiving end of the UDP pathway, there should be someway to put datagrams back in order. The subtlety, however, of requiring sequence numbers is that in order for our error-correcting codes approach to work for the code we experimented with at least 75% of the datagram payload data must be available and properly sequenced. So, the inclusion of sequence numbers is really twofold.

The second feature to be added, error-correcting codes or ECC, is a little more complicated and requires more of an introduction, if for no other reason than to get acclimated to the terminology in the literature.

Given the no-guaranty and low-reliability characteristics of the networking layer and the little added value of UDP, a virtual communication link through a UDP socket represents an unreliable, and potentially noisy, channel, making it a viable candidate for a solution based on error-correcting codes.

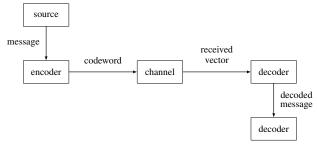


Fig. 1: Block diagram of a general communication link

A. Error-Correcting Codes

Above in Figure 1 is a block diagram of a general communication link for sending messages [2]. Here the message is first sent by the source to the encoder where the message is assigned a codeword, i.e. a string of characters from some chosen alphabet. The encoded message is sent through a *channel*, which is certain to have some level of noise that may possibly alter the codeword. This possibly-altered codeword is then received by the decoder where it is matched with its most likely candidate codeword, which is in turn translated into a message resembling the original message and passed on to the receiver. The degree of resemblance to the original message depends on how appropriate the code is in relation to the channel.

Example I.1. Suppose we are using an alphabet containing only two symbols, say 0 and 1 and we wish to send a message to a friend of either "YES" or "NO." We first might consider encoding 1 for "YES" and 0 for "NO"; however, if the codeword 0 is sent and then altered by the channel to be received as 1, the decoder will incorrectly pass the message "YES" to the receiver. One way of improving the situation is to add redundancy by repeating the symbol. For instance, we might encode "YES" as the codeword 11 and "NO" as 00. Now if "NO" is sent, two errors would have to occur before the decoder returns an incorrect message. If one error occurs, then the received vector will either be 10 or 01, neither of which is a codeword. At this point, the receiver might request a retransmission. This is an example of a code in which one error may be detected. To strengthen the code

to be error-correcting, we might increase the redundancy by repeating the message five times. Thus we encode our message "NO" as 00000. Perhaps the channel interferes to cause the decoder to receive the vector 00110. Using the method of nearest neighbor decoding, the decoder assesses the message and decides that of the two possible codewords (i.e. 00000 and 11111) 00000 is more likely the one originally sent and hence is correctly decoded as "NO."

Definition I.2. Let **F** be a finite set, or **alphabet**, of q elements. A q-**ary code C** is a set of finite sequences of symbols of **F**, called **codewords** and written $x_1x_2\cdots x_n$, or (x_1, x_2, \ldots, x_n) , where $x_i \in \mathbf{F}$ for $i = 1, \ldots, n$. If all the sequences have the same length n, then **C** is a **block code** of **block length** n.

Given an alphabet \mathbf{F} , it is consistent with terminology for vector spaces when \mathbf{F} is a field to denote the set of all sequences of length n of elements of \mathbf{F} by \mathbf{F}^n and to call these sequences vectors. The member of \mathbf{F} at the ith position of a vector is known as the coordinate at i.

Definition I.3. Let $v = (v_1, v_2, ..., v_n)$ and $w = (w_1, w_2, ..., w_n)$ be two vectors in \mathbf{F}^n . The **Hamming distance**, d(v, w), between v and w is the number of coordinate places in which they differ:

$$d(v, w) = \{i | v_i \neq w_i\}$$

We will usually refer to the Hamming distance as simply the **distance** between two vectors.

Nearest-neighbor decoding is a method in which the received vector is translated as the codeword of smallest distance, whenever it is uniquely determined. This method maximizes the likelihood of correcting errors provide the two assumptions mentioned above hold (i.e. probability of an error <1/2 and each symbol is equally likely to be transmitted).

Definition I.4. The **minimum distance** $d(\mathbf{C})$ of a code \mathbf{C} is the smallest of the distances between distinct codewords; i.e.

$$d(\mathbf{C}) = \min\{d(v, w)|v, w \in \mathbf{C}, v \neq w\}$$

Example I.5. The error-correcting code in Example I.1 is a binary code of block length 5. The distance between the codeword 00000 and the received vector 00110 is 2. The distance between the same received vector and the codeword 11111 is 3. Thus, 00000 is the nearest neighboring codeword. The minimum distance of this code is 5.

Definition I.6. A code C over F of length n is **linear** if C is a subspace of the vector space $V = F^n$. If $\dim(C) = k$ and d(C) = d, then we write [n, k, d] for the q-ary code C; if the minimum distance is not specified we simply write [n, k]. The **information rate** is k/n and the **redundancy** is n - k.

Since C is a subspace, it follows that the **all-zero** vector must be a codeword in C. Also, the sum of any two codewords must also be a codeword in C. We will also see after the next

definition that the distance d is also easy to calculate for linear codes

Definition I.7. Let $V = \mathbf{F}^n$. For any vector $v = (v_1, v_2, \dots, v_n) \in V$ set

$$S = \{i | v_i \neq 0\}$$

Then S is called the support of v and the weight of v is |S|. The **minimum weight** of a code is the minimum of the weights of the nonzero codewords.

Example I.8. Listed below are all 16 codewords in the linear [7, 4, 3] binary code. They are listed in order of increasing weight.

0000000	0111000	1000111	0101101
1100100	0001110	1101010	0110110
1010010	0010101	1110001	0011011
1001001	0100011	1011100	1111111

Notice that the minimum weight is 3. By Theorem 1.1 it follows that $d(\mathbf{C}) = 3$. It can be shown that these 16 codewords form a subspace of \mathbf{F}^7 with basis $\{(1,0,0,0,1,1,1), (0,1,0,0,0,1,1), (0,0,1,0,1), (0,0,0,1,1,1,0)\}$. Thus $\dim(\mathbf{C}) = 4$ as conveyed in the notation and the redundancy is 3 since n - k = 7 - 4 = 3.

Theorem I.9. Let C be a code of minimum distance d. If $d \ge 2t + 1$, then C can be used to correct up to t errors.

II. DESIGN

A Better UDP is "wedged in" as another IP stack layer sitting between the transport layer and the application layer, see Figure 2. We did not modify any kernel source and A Better UDP is not a standard yet, so to use it an application developer or software engineer will have to include this functionality as a library. Instead of using a socket library to interface with the transport layer, the engineer will link with A Better UDP library to interface with the remainder of the protocol stack which sits below the application layer. Henceforth we will refer to A Better UDP as BUDP.

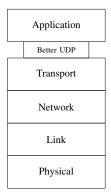


Fig. 2: Five-layer Internet Protocol Stack with a Better UDP Wedged in

Messages, from the application layer at the source, are passed to BUDP, where, currently, four at a time are buffered, assigned sequence numbers, encoded to produce an ECC, and then sent as datagrams via standard UDP to their destination address. At the destination, the datagrams are collected and stored in an accumulation buffer, where they are placed in their proper sequence (see Figure 3). The purpose of buffering at the source is to support the process of encoding the ECC. At the destination, it is a requisite part of the process since datagrams, presumably, arrive one-at-a-time and the decoding process cannot begin until a sufficient amount of the block data has been received. Here, we say block data, because as we will soon see datagrams have to be encoded in groups which we refer to, at various places in the source code, as blocks, for "blocks of datagrams."

One of three conditions prompts the delivery of messages to the application layer at the destination:

- 1) All four datagrams are received intact.
- 2) Three datagrams are received as well as the ECC, intact.
- 3) A countdown timer expires.

The first of these conditions is obvious, while the other two require further explanation. It is easier to explain 3) first, so we will do that now. As stated previously in our introduction, we do not intend to rewrite TCP, and we would like to keep our UDP implementation connectionless, so instead of requesting retransmission of missing datagrams, for example, when the one of the first two conditions above is not met, our implementation will simply give what it has to the application layer at the destination. Our justification for doing this is that it is what a standard UDP implementation would do. Actually, a standard implementation of UDP would do less, because our implementation will at least give the application layer in-order messages.

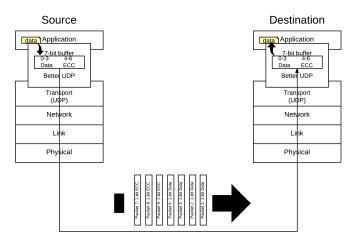


Fig. 3: A Better UDP Overview Diagram

The second condition which can prompt message delivery to the application layer requires the use of error-correcting codes. For the purposes of prototyping our design, we used the single error-correcting Hamming code introduced above which allowed us to benchmark our implementation against both standard TCP and standard UDP. Our purposes for using this code over the Reed-Solomon code suggested in RFC5510 [3] were four-fold:

- It is simple to implement and suits our purposes very well.
- One of our co-author was very familiar with it and its connection to Finite Geometry.
- 3) The [7,4,3] Hamming code is a perfect single errorcorrecting code, which happens to be the smallest perfect code, and is generated by a finite geometry known as the Fano Plane which is the smallest projective plane [2].
- 4) Looking long term, we perhaps want to move our implementation to the newer so-called "Low-Density Parity-Check Codes" which have been beating some of the more recent ECC frontrunners such as turbo codes in competitions [4]. Some of these codes can be generated from Finite Geometries, and those that are "can be decoded in various ways, ranging from low to high decoding complexity and from reasonably good to very good performance" [4].

A. Encoding and Decoding

Encoding the message stream, using four messages at a time, produces three ECC messages which are sent seperately as follow-on datagrams. So, at an overly simplified level, the message stream, grouped in blocks of seven, flows like so:

$$\cdots ddddpppddddppp \cdots$$

where d is a data symbol from the sender application, and p is a parity-check symbol associated with the four most recent d symbols prior to it. This represents an increase of 75% in the number of datagrams transmitted. These ECC or parity-check datagrams also contain sequence numbers, which are a continuation of the sequence numbers assigned to application data messages. These sequence numbers are used in the identification of these datagrams as containing parity-check bytes. They are also used to identify with which application data containing datagrams they are associated.

To use the [7,4,3] Hamming code, there are two very important matrices required: the *generator matrix* G

$$G = \begin{pmatrix} 1 & 0 & 0 & 0 & 1 & 1 & 0 \\ 0 & 1 & 0 & 0 & 1 & 0 & 1 \\ 0 & 0 & 1 & 0 & 0 & 1 & 1 \\ 0 & 0 & 0 & 1 & 1 & 1 & 1 \end{pmatrix}$$

and the parity-check matrix H

$$H = \begin{pmatrix} 1 & 1 & 0 & 1 & 1 & 0 & 0 \\ 1 & 0 & 1 & 1 & 0 & 1 & 0 \\ 0 & 1 & 1 & 1 & 0 & 0 & 1 \end{pmatrix}$$

At the source, to encode a message m, which for purposes of the math is represented as a row vector in \mathbf{F}_2^4 , multiply it on the right by G to get mG, which is a row vector in \mathbf{F}_2^7 , where

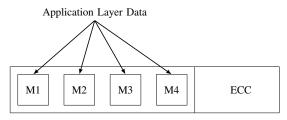


Fig. 4: Simplified ECC Encoding Scheme

the first four coordinates are m and the last 3 coordinates are the parity-check bits, or (as we have been colloquially referring to them) the "ECC." In RFC 5510 [3], these bits are called repair symbols.

At the destination, to decode the received vector $r \in \mathbf{F}_2^7$, mutliply it on the right by H^{T} to get rH^{T} , which is a row vector in \mathbf{F}_2^3 known as the syndrome. Next, syndrome decoding is performed whereby a *coset leader* e is matched with the syndrome rH^{T} . This coset leader represents the error vector, which is the error between the received vector r and the nearest neighboring codeword c to it. To determine c simply add r and e modulo 2. That is,

$$c = r + e \mod 2$$

B. Encoding Scheme

Figure 4 shows a simplified diagram of the ECC encoding scheme we implemented, where 'M' stands for message and the number is the datagram sequence number.

There are actually two levels of ECCs that we considered using, but decided against this due to the unnecessary redundancy we felt it added. This two-level scheme is shown in Figure 5. The level we implemented in the one-level approach turns out to be an outer ECC in the two-level approach as shown in Figure 5. This two level scheme is very similar to the way CDs and DVDs are encoded for error recovery [4].

The second level is the so-called inner or message level ECC encoding. This inner ECC would allow for checking and restoring datagram integrity. That is, if a datagram were received at the destination and found to be corrupted, it could be "fixed." We felt this redundancy would never fulfill its purpose since stock UDP (and, potentially, layers below it such as Ethernet and 802.11 [1]) already provides a checksumming mechanism that potentially discard corrupted datagrams.

To make a scheme like this work, at least for the approach in Figure 5, the inner ECC would have to have almost as many parity bits as there are data bits in the application layer message. The reason for this is that in the case of a lost datagram the outer ECC would have to regenerate the missing inner ECC, which, in turn, would have to regenerate the application layer message data. Our initial thought on this is that it will force a fixed-size payload, so that the size of the inner ECC can be pre-determined.

With all the header overhead already inherent in the Internet Protocol plus adding a sequence number and stream-level error-correction codes, the actual application data might end up only accounting for 20-40% of the data transmitted, depending on how the application data is fed to the DUBP layer. Adding a second level of ECCs would only exacerbate this condition and provide little added benefit.

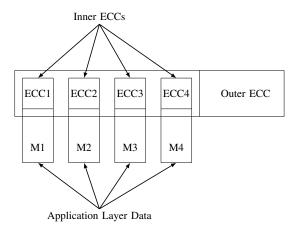


Fig. 5: Two-level ECC Encoding Scheme

C. Encapsulation

Our BUDP datagram will be encapsulated in an ordinary UDP datagram. More specifically, BUDP will contain a header including a field for the sequence number. This BUDP datagram (BUDP header + BUDP payload), will form the payload of a regular UDP datagram. See Figure 6 below.

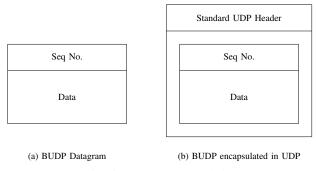


Fig. 6: BUDP Encapsulation

III. TESTING

A. Method

We have a fully functional implementation of A Better UDP protocol which was written in standard C. It was developed on a few different Linux workstations and an Apple laptop. Our code was version-controlled through the whole process with tools offered through http://bitbucket.org. The source code along with wireshark traces is included with this submittal. The code was written entirely from scratch. This proved challenging at times, but we are glad we did it this way, because we learned about a lot of "gotchas" in our sequencing and error-correcting design.

The receiving side of the protocol was implemented as a state machine, which we like for its utility in exposing the interworkings of what is happening in the various cases of receiving, buffering, reordering, and error-correcting.

Currently, the state machine is not bug-free, but it is working well enough to confirm some of our theory and predictions, which we detail here.

Firstly, we planned to compare our BUDP benchmark against TCP, but unfortunately the clock ran out on us for getting this done in an apples-to-apples sort of way. The test environment to do this was a little more challenging to get going due to the fact that we would have to standup some sort of WAN emulator requiring three PCs networked together. We have those resources in the lab, but it takes a fair bit of coordination to pull off. So, we benchmarked BUDP vs stock UDP and kept the test case simple and controlled. We programmatically forced every 4^{th} datagram to be lost. That is, we simply did not send it. For BUDP, we still sent the ECC bytes that trail the 4^{th} datagram.

Secondly, for timing each transmit session we used Wireshark and did the transmission over the loopback address. This turned out to be a very test friendly way of doing things, besides recording packets in sufficient detail.

Lastly, for transmission data we used text file, which contained the text to the ebook "A Christmas Carol" by Charles Dickens as it seemed appropriate given the season. This file contained 189067 and is also included with our submission. Hopefully, this passes the fair use clause of copyright law.

B. Results

With BUDP, for every 4 data bytes we are sending 3 parity-check bytes, whereas for stock UDP only the 4 data bytes are sent. Therefore, we expected the transmission time for BUDP to be, roughly, 1.75 times greater than the transmission time of stock UDP. This is exactly, with in tolerances, what we saw. Table I summarizes the timing outputs for both BUDP and stock UDP Wireshark sessions. The increase in BUDP over UDP represents a 1.765 multiple. The difference between that and the predicted 1.75 could be due to the added bytes in the BUDP datagram for sequence numbers.

TABLE I: Wireshark Trace Timings

Protocol	Timing (sec.)	
UDP	11.15	
BUDP	19.68	

Also, the [7,4,3] Hamming code provided the error-correction capability we anticipated. On the receiver side of the dataflow, we wrote the received application data (post error recovery) to an output file named "budp.out". We used a tool known as meld to "diff" the BUDP output with the original text and found they were identical. Remember, this includes forcing the sender to drop every 4^{th} data byte. Obviously, a similar "diff" between the stock UDP output and the original file showed every 4^{th} text character was missing. This proves our assertion, in the beginning, that BUDP improves upon UDP.

IV. CONCLUSION

We have presented a design for a better UDP implementation where datagrams arrive in-order to the application layer at the destination and when a datagram is lost it can be reconstructed, provided 75% of the datagrams in its same encoding block and their associated parity-check bytes arrive intact. If these latter two conditions are not met, then BUDP simply hands over what it has available to the application layer after a countdown clock expires. In this case, it essentially falls back to a standard UDP implementation with the added benefit of in-order datagrams. This implementation requires the maintaining of buffers at both the sender and receiver sides of the communication channel for purposes of computing the parity-check bytes, datagram re-ordering, and recovering missing datagrams.

Clearly, we need to do some more work on our implementation as the thoughput timings compared to stock UDP are 75% higher, though this was anticipated for the code we chose to implement. We believe this design has a roadmap to better timings while maintaining the same error-recovery characteristics, with newer codes generated by finite geometries promising to provide this correctability at optimal speeds [4].

Also, we find it interesting that finite geometry keeps finding its way back into the discussion of optimal error-correcting codes [4] [5].

Finally, our view is that in order to deal with the unreliability inherent in the Internet Protocol, a designer has to either build in redunduncy to help clean up the communication channel or has to build in a retransmission infastructure, as is the case with TCP. We prefer the error-correcting approach as it is more elegant and mathematically interesting and offers a path to an optimal speed protocol. If in the future such happens, we do not see why a designer would use TCP, or at least we feel TCP would be used less frequently.

ACKNOWLEDGMENT

The Error-Correct Codes subsection of the Introduction was taken, in part, from a presentation one of the co-authors presented at a Mathematical Association of America Texas Section conference in April 2005 [2] focusing on its connection to Finite Geometries. It was included here to help demystify some of the math related to the [7,4,3] Hamming code used in our prototype. All other sections and subsections are original. This is the first time any of these authors have used Error-Correcting Codes to improve a transport protocol.

REFERENCES

- [1] J. F. Kurose and K. W. Ross, *Computer networking*. Pearson Education, 2012
- [2] C. B. Clanton, "Use of finite geometries in error-correcting codes." Mathematical Association of America Texas Section, 2005.
- [3] J. Lacan, V. Roca, J. Peltotalo, and S. Peltotalo, "Reed-solomon forward error correction (fec) schemes." Institute of Electrical and Electronics Engineers, 2009.
- [4] Y. Kou, S. Lin, and M. P. Fossorier, "Low-density parity-check codes based on finite geometries: a rediscovery and new results," *Information Theory, IEEE Transactions on*, vol. 47, no. 7, pp. 2711–2736, 2001.
- [5] E. F. Assmus and J. D. Key, *Designs and their Codes*. Cambridge University Press, 1992, vol. 103.