

## ④ Representation of Discrete Time Signals

Signals are represented in following way

1. Functional Representation

2. Graphical Representation

3. Tabular Representation

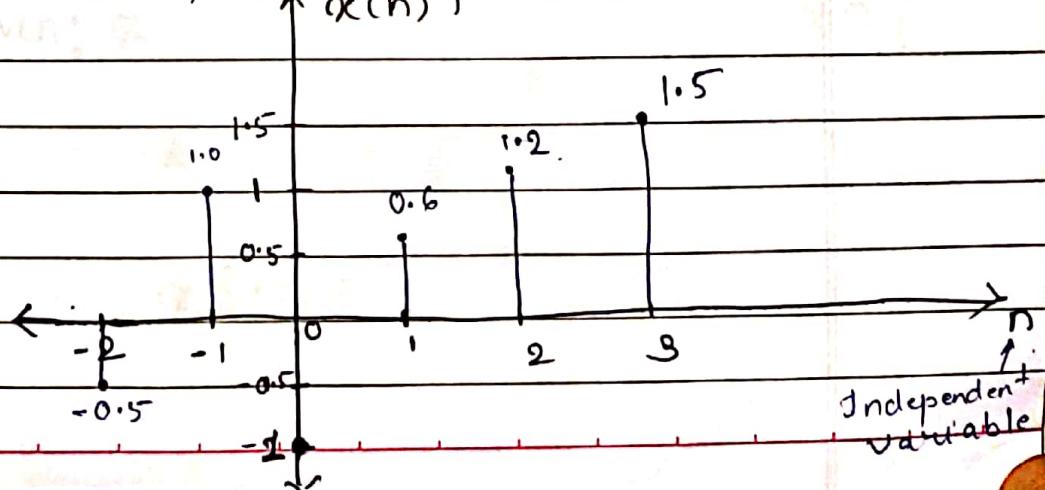
4. Sequence Representation

~~Functional Representation~~

→ Represented in Mathematical Equation

$$\text{Eg: } x(n) = \begin{cases} -0.5 & ; n = -2 \\ 1.0 & ; n = -1 \\ -1.0 & ; n = 0 \\ 0.6 & ; n = 1 \\ 1.2 & ; n = 2 \\ 1.5 & ; n = 3 \\ 0 & ; \text{other } n. \end{cases}$$

2. Graphical Representation :



### 3. Tabular Representation

$n$	.....	-2	-1	0	1	2	3	...
$x(n)$	.....	-0.5	1.0	-1.0	0.6	1.2	1.5	...

→ Independent variable

~~No. of Signal~~ ~~Value~~ ~~for each variable~~

minifacets of solution 18

~~minister for foreign affairs~~ Mr. Edward Norton.

## 4<sup>o</sup> Sequence representation

→ Represented in one dimensional

Position and array behavior -

matou p<sup>3</sup>

$$x(n) = \begin{cases} \dots, -0.5, 1.0, -1.0, 0.6, 1.2, \dots \end{cases}$$

$$S = \{x \in \mathbb{R}^n : x(m) > 0\}$$

$$1 - \alpha = 0.1 =$$

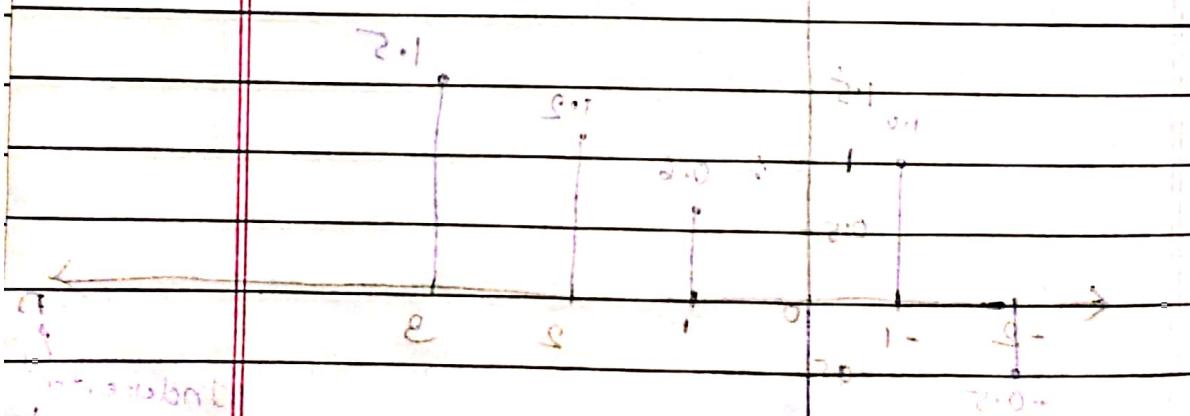
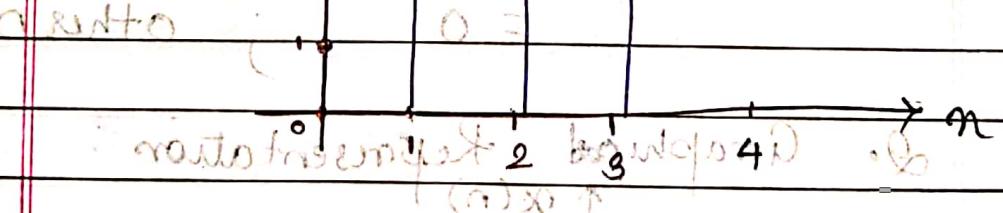
$$\text{Eq: } \{a(n) \mid n \in \mathbb{N}\} = \{0, 12, 3, 4\}$$

二〇一九年六月

$$t = a$$

~~1~~ 3 ~~1~~ ~~1~~

2



# Mathematical operations on Discrete Time Signals.

$$0.2 = 0.1 \times 2.0 = (0.2)_D \quad 0 = 0$$

$$2.0 = 1.0 \times 2.0 = (1.0)_D \quad 1 = 1$$

$$0.8 = 0.8 \times 1.0 = (0.8)_D \quad 0.8 = 0.8$$

↳ Amplitude scaling

↳ Time scaling

$$\{ 0.1 \quad 0.8 \} \rightarrow (1.0)_D$$

$$3. Shifting$$

↳ Right shift (advance)

↳ Left shift (delay)

$$4. Addition$$

$$15. Multiplication$$

## 1. (a) Amplitude Scaling

↳ Multiply value of every signal sample by A (constant).

$$\text{Eg: } y(n) = A \cdot x(n)$$

amplitude scaled      constant

Signal

Eg Given:  $x(n) = \{10, 16, 20\}$

$$\text{Eg: } A = 0.2$$

$$x(n) \times 0.2$$

$$y(n) = A \alpha(n)$$

$$n=0; y(0) = 0.2 \times 10 = 2.0$$

$$n=1; y(1) = 0.2 \times 16 = 3.2$$

$$n=2; y(2) = 0.2 \times 20 = 4.0$$

$$y(n) = \{2.0, 3.2, 4.0\}$$

(Example 1) Find the filter  $A$

Example 2

$$\alpha(n) = \{1, 2, 1, 0, 1, 2, 1\}$$

$$A = \text{pril 2} \rightarrow \text{shifting } A(n)$$

$\rightarrow$

$$y(n) = A \alpha(n) \leftarrow$$

$$y(n) = \{4, 1, 4, 1\}$$

$$(a) \rightarrow A = \{1, 2, 1, 0, 1, 2, 1\}$$

~~Frontend~~, ~~backend~~

1(b) Timescaling (Downsampling or Upsampling)  
 $\rightarrow$  decreasing or increasing the signal, 8 times

~~for~~ ~~down~~  $\rightarrow$  ~~up~~  $\rightarrow$  ~~Downsampling~~

Eg:

$$\alpha(n) = \{1, 0, 1, 1, 1, 2\}$$

Fnd  $\alpha(2n) = ?$

$$n = -1 \quad ; \quad x(2n) = x(2 \cdot -1) = x(-2) = 0$$

$$n=0 \quad ; \quad x^{(2n)} = x^{(0)} \quad (1)$$

$$n=1 \quad ; \quad x(2n) = x(2 \cdot 1) = x(2) = 1$$

$$n=2 \quad ; \quad x(2n)=x(4)=0$$

$$\alpha(2n) = \{0, 0, 1, 1, 0\}$$

↑  $z - z = 4$

(\*) Upsampling =  $x(n/I)$   
 $(nE)^p = (n)_p$

$$A = -1, x(-1/2) = 0$$

$$\phi = (\mu -) \times \psi = (\mu -) \mu^{-1} \circ \phi = \phi$$

E.g.:  $\beta(\alpha \cdot (n)) = \{ \beta(p) : |p| = |\alpha|, 2 \}$

$$D = (\varphi) \times = (\varphi) p : \rightarrow 1 = C$$

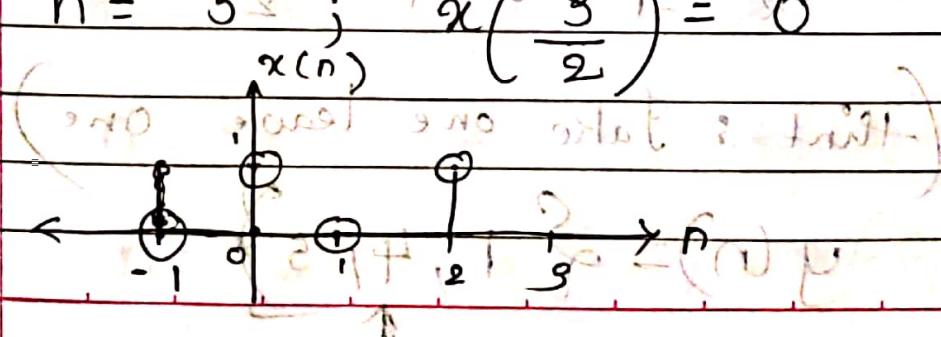
$$n = -1 \in \left( ; \right] x \in \left( \frac{n}{2} \right) = \left( x \left( \frac{-1}{2} \right) \right)$$

$$n=6 \quad ; \quad x\left(\frac{n}{2}\right) = x(0) = 1$$

$$n = 1 \quad ; \quad x\left(\frac{n}{2}\right) = x\left(\frac{1}{2}\right) = 0$$

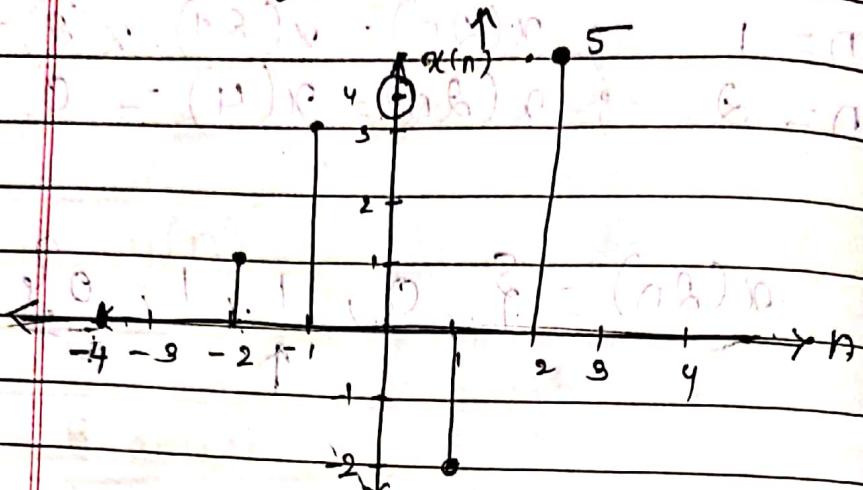
$$n = 2 \quad ; \quad x\left(\frac{2}{2}\right) = x(1) = 1$$

$$n = 3 \text{ bei } x \left( \frac{3}{2} \right) = 0$$



## Additional Examples

$$\alpha(n) = \{1, 3, 4, -2, 5\}$$



② Downscaling = subsampling

$$y(n) = \alpha(2n)$$

$$n = -2 ; y(n) = \alpha(-4) = 0$$

$$n = -1 ; y(n) = \alpha(-2) = 3$$

$$n = 0 ; y(n) = \alpha(0) = 5$$

$$n = 1 ; y(n) = \alpha(2) = -2$$

$$n = 2 ; y(n) = \alpha(4) = 0$$

$$n = 3 ; y(n) = \alpha(6) = 4$$

$$n = 4 ; y(n) = \alpha(8) = 0$$

$$n = 5 ; y(n) = \alpha(10) = 3$$

$$n = 6 ; y(n) = \alpha(12) = -2$$

$$n = 7 ; y(n) = \alpha(14) = 5$$

$$n = 8 ; y(n) = \alpha(16) = 0$$

$$n = 9 ; y(n) = \alpha(18) = 3$$

(Hint : Take one, leave one)

$$y(n) = \{1, 4, 5\}$$

$$y(n) = x(-n)$$

$$n = -2 \quad y(-2) = x(2) = -1.6$$

$$n = -1 \quad y(-1) = x(1) = 0.8$$

$$n = 0 \quad y(0) = x(0) = 0$$

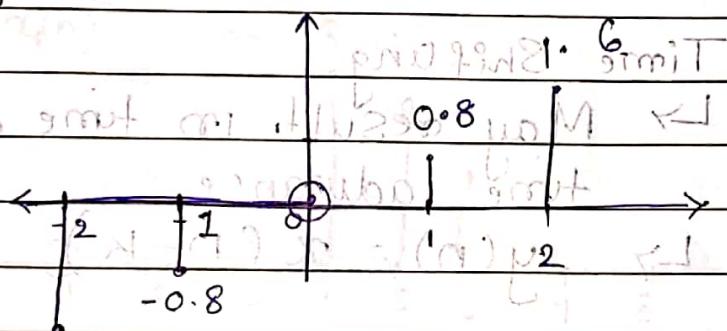
$$n = 1 \quad y(1) = x(-1) = -0.8$$

$$n = 2 \quad y(2) = x(-2) = -1.6$$

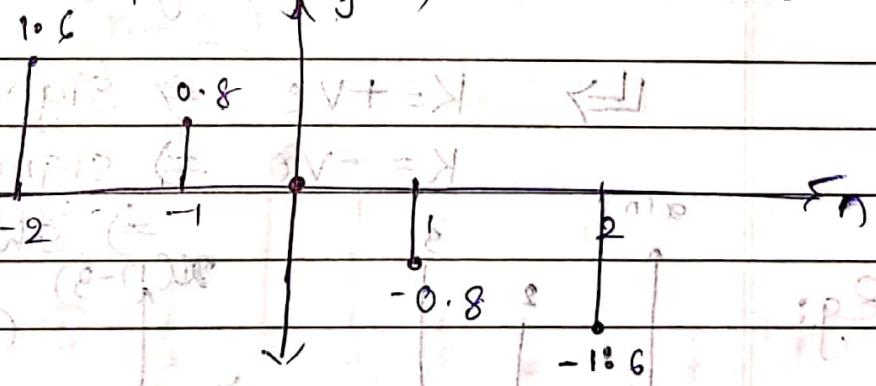
$$y(n) = \{ 1.6, 0.8, 0, -0.8, -1.6 \}$$

↑ (2)      ↑ (0)

Before  $\alpha(n)$



After folding by  $y(n)$



e.g.:

$$x(n) = \{ 0.5, 1, 2, 0.5, 1, 1, 1, 0.5 \}$$

$$y(n) = \{ 1, 1, 1, 0.5, 2, 1, 0.5 \}$$

↑ (2)

[Process of increasing the signal strength by inserting zeros between samples]

### Downsampling

b)

$$x(n) = \{1, 3, 4, -2, 5\}$$

$$y(n) = x\left(\frac{n}{3}\right)$$

$$n = -1, 0, 1, 2, 3, y(-2) = x\left(-2/3\right) = 0$$

$$n = -1 \quad y(-1) = x\left(-1/3\right) = 0$$

$$\rightarrow n=0 \quad y(0) = x(0) = 1$$

$$n = -1 \quad y(-1) = x\left(1/3\right) = 0$$

$$n = -2 \quad y(-2) = x\left(2/3\right) = 0$$

$$n = 3 \quad y(3) = x\left(3/3\right) = -2$$

$$n = 4 \quad y(4) = x\left(4/3\right) = 0$$

$$n = 5 \quad y(5) = x\left(5/3\right) = 0$$

$$n = 6 \quad y(6) = x\left(6/3\right) = 5$$

$$n = -3 \quad y(-3) = x\left(-3/3\right) = 3$$

$$n = -4 \quad y(-4) = x\left(-4/3\right) = 0$$

$$n = -5 \quad y(-5) = x\left(-5/3\right) = 0$$

$$n = -6 \quad y(-6) = x\left(-6/3\right) = 1$$

Original Signal  $x(n) = \{1, 3, 4, -2, 5\}$

Upscaled Signal  $y(n) = ?$

$$y(n) = (2)y(n) = \{1, 0, 1, 3, 0, 0, 4, -2, 0, 5\}$$

$$y(n) = (2|1-)x(n) = (1-)y(n) \quad 1- = a$$

Q. Folding (= Reflection or Transpose or Time reversal)

$$(1) \quad x(n) = x(-n) \quad n = a$$

Let  $x(n)$  be original signal

Then  $y(n) = x(-n)$  is the folded signal.

$$x(n) = (0.8)n \quad n-2 \leq n \leq 2$$

Eg:

$$x(n) = (0.8)n \quad n-2 \leq n \leq 2$$

$$x(-1) = -1(0.8) \quad -1 = n$$

$$x(-1) = -0.8$$

$$x(1) = (0.8)1 \quad 1 = n$$

$$x(1) = 0.8$$

$$x(2) = (0.8)2 \quad 2 = n$$

$$x(n) = \{0.8, -0.8, 0, 0.8, 1.6\}$$

$$x(n) = (-x(-n)) \quad -n = a$$

$$x(n) = (x(n)) \quad n = a$$

$$y(n) = x(-n)$$

$$n = -2 \quad y(-2) = x(2) = -1.6$$

$$n = -1 \quad y(-1) = x(1) = 0.8$$

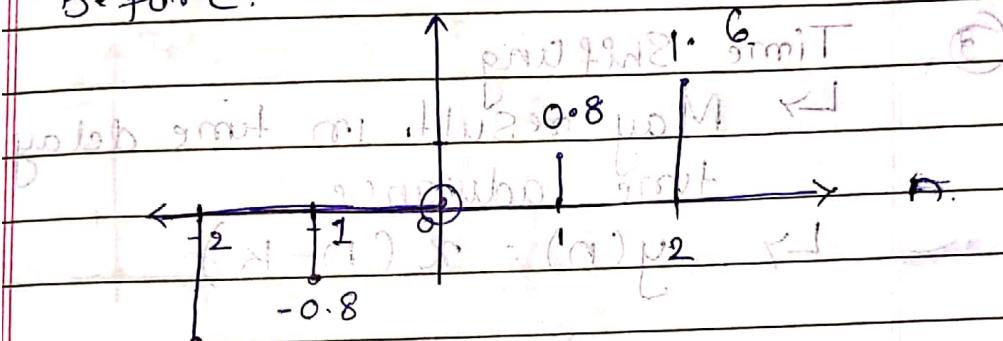
$$n = 0 \quad y(0) = x(0) = 0$$

$$n = 1 \quad y(1) = x(-1) = -0.8$$

$$n = 2 \quad y(2) = x(-2) = -1.6$$

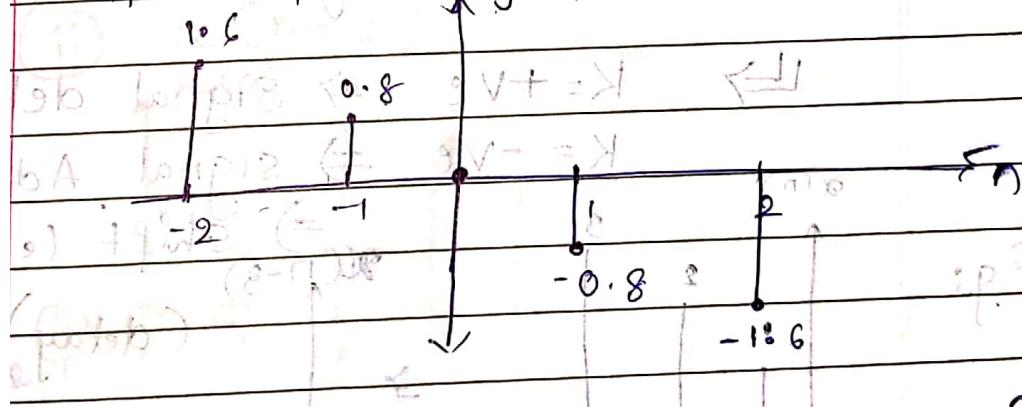
$$y(n) = \{ 1.6, 0.8, 0, -0.8, -1.6 \}$$

Before  $x(n)$



No pd - 1.6 Fartdown  $x(n)$

After folding  $y(n)$



$$x_c(n) = \{ 0.5, 1, 2, 0.8, 1, 1, 1, 1, 1, 0.5 \}$$

$$y(n) = \{ 1, 1, 1, 0.8, 2, 1, 0.5 \}$$

Example: Time scaling

$$x(n) = \{1, 2, 8, 4, 6\}$$

Downscaling

$$x(2n) = \{1, 8, 6\}$$

Upscaling

$$x\left(\frac{n}{2}\right) = \{1, 0, 2, 0, 8, 0, 4, 0, 6\}$$

③

Time shifting

May result in time delay or time advance.

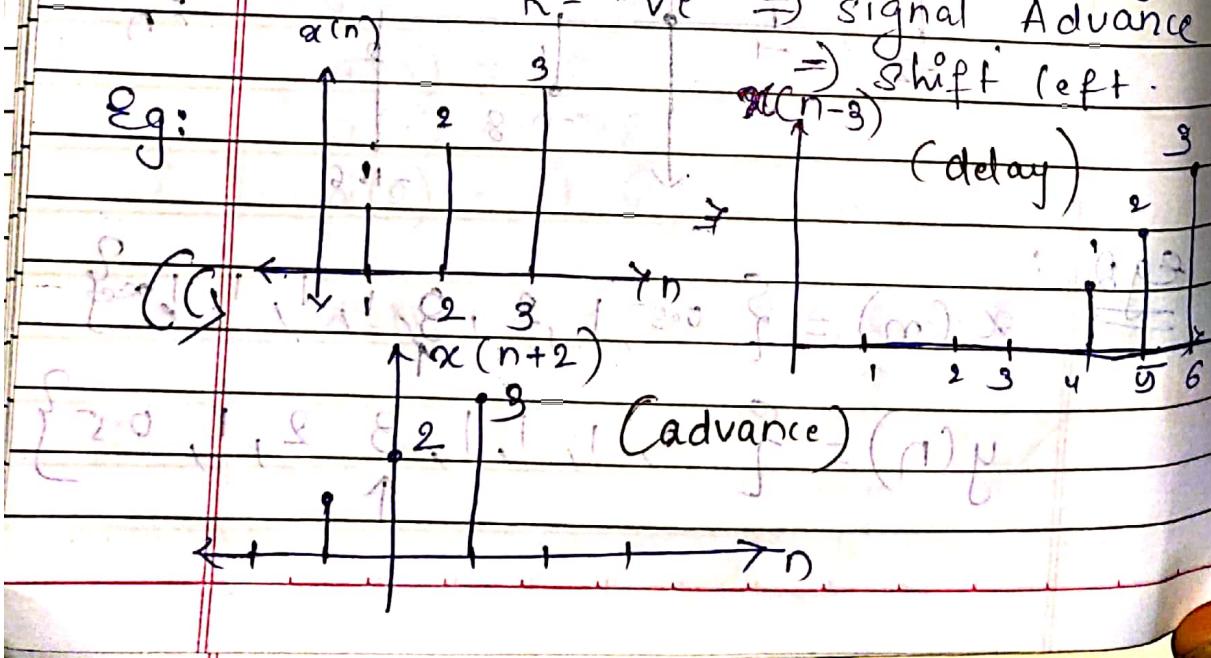
$$\Rightarrow y(n) = x(n-k)$$

$y(n)$  is obtained by shifting the  $x(n)$  by ' $k$ ' units

$\Rightarrow k = +ve \Rightarrow$  signal delay  $\Rightarrow$  shift right

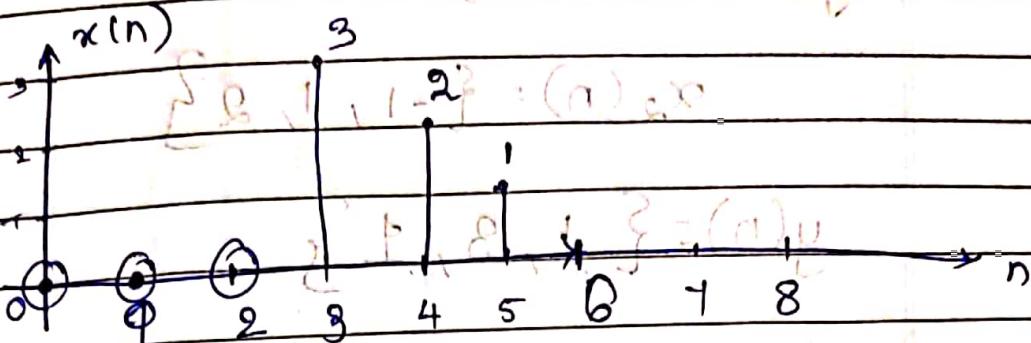
$k = -ve \Rightarrow$  signal Advance  $\Rightarrow$  shift left.

Eg:



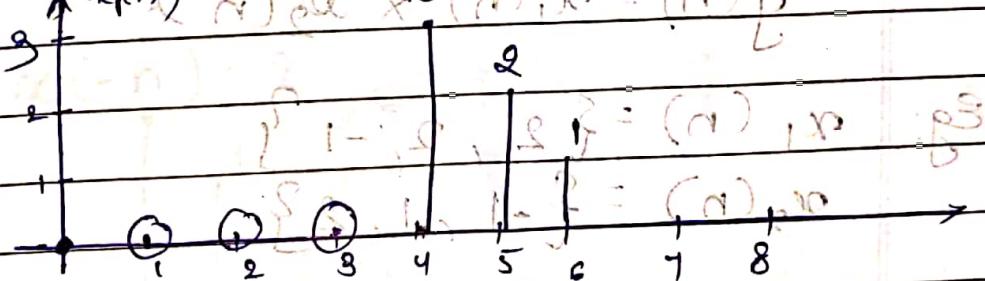
Eg 2

$$\alpha(n) = \{0; 0, 0, 3, 2, 1, 0, 0, \dots\}$$



$$\textcircled{1} \quad x(n-1)$$

$$\alpha'(n)(\alpha) \in x^3(\alpha), \text{ so } \alpha' = f(\alpha).$$

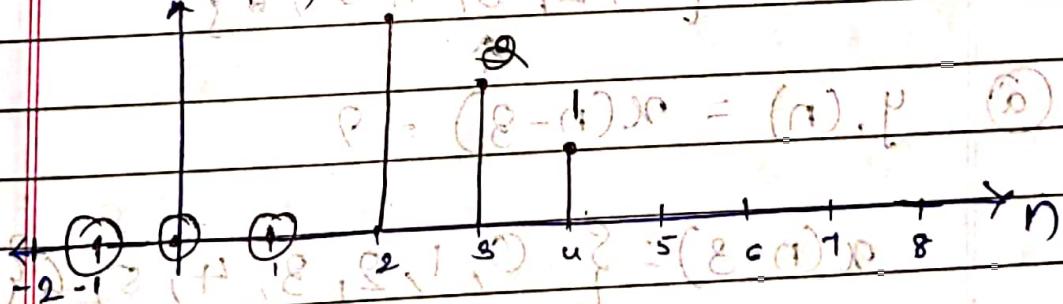


## Delay the Signal

1

$$\alpha(n+2)$$

$$\{x_i(n) \in \mathbb{R}^d, i=1, \dots, n\} = (\alpha)\Omega$$



#### 4. Addition

$$y(n) = x_1(n) + x_2(n)$$

Eg:  $x_1(n) = \{2, 2, -1\}$

$$x_2(n) = \{-1, 1, 2\}$$

$$y(n) = \{1, 3, 1\}$$

#### 5. Multiplication

$$y(n) = x_1(n) \times x_2(n)$$

Eg:  $x_1(n) = \{2, 2, -1\}$

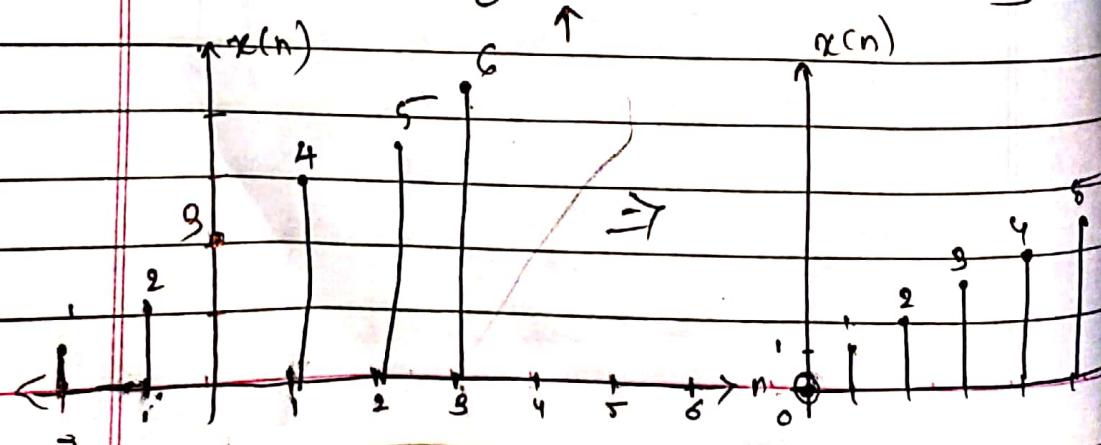
$$x_2(n) = \{-1, 1, 2\}$$

$$y(n) = \{-2, 2, -2\}$$

Eg.  $\alpha(n) = \{1, 2, 3, 4, 5, 6\}$

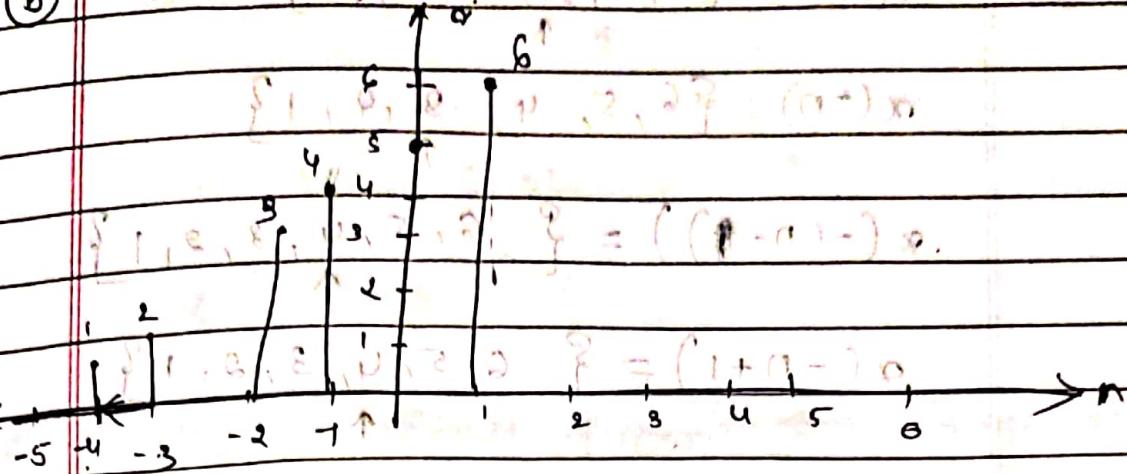
(a)  $y_1(n) = \alpha(n-3) = 9$

$$\alpha(n-3) = \{0, 1, 2, 3, 4, 5, 6\}$$



(b)

$$\alpha(n+2) = ?$$

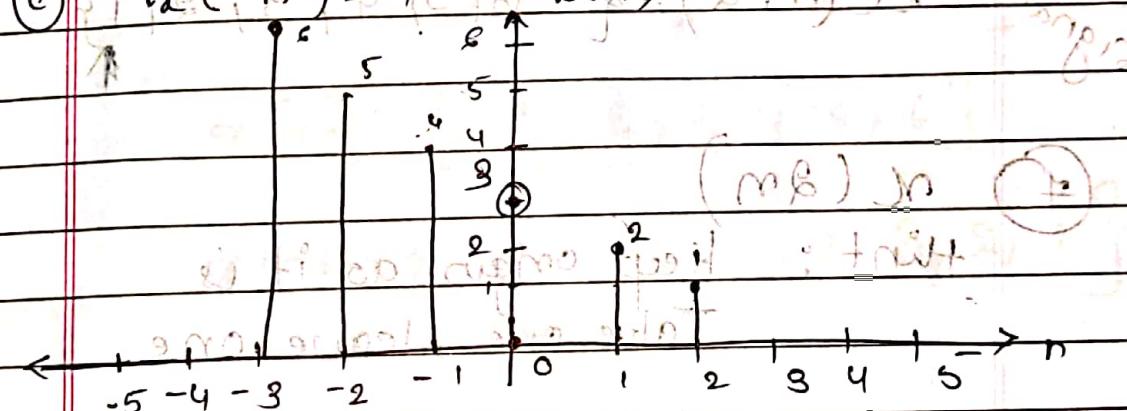


$$\alpha(n+2) = \{1, 2, 3, 4, 5, 6\} \quad (3)$$

$$f = ((n+1)-4) *$$

(c)

$$\alpha(-n) = ? \quad (\alpha(n))$$



$$\alpha(-n) = \{1, 2, 3, 4, 5, 6\}$$

$$\alpha(-n) = \{6, 5, 4, 3, 2, 1\}$$

(d)

$$\alpha(-n+1) \quad (\text{P.T. for alternate solution})$$

$$= \alpha(-(n+1)) \Rightarrow \text{fold & delay operations}$$

$$\alpha(n) \xrightarrow{\text{Fold}} \alpha(-n) \xrightarrow{\text{delay}} \alpha(-(n+k))$$

Ex: positive numbers  $\Rightarrow$  delay

$$\{1, 2, 3, 4, 5, 6, 7, 8, 9, 10\} \xrightarrow{\text{delay}} \alpha(-n+k)$$

$$\{9, 8, 7, 6, 5, 4, 3, 2, 1\} \xrightarrow{\text{delay}} \alpha(-n+k)$$

$$x(n) = \{1, 2, 3, 4, 5, 6\}$$

$$x(-n) = \{6, 5, 4, 3, 2, 1\}$$

$$x(-(n-1)) = \{6, 5, 4, 3, 2, 1\}$$

$$x(-n+1) = \{6, 5, 4, 3, 2, 1\}$$

\* ⑦  $x(-n-2) = x(-(n+2))$

Fold advance  
2 advance  
signal

$$x(-n) = \{6, 5, 4, 3, 2, 1\}$$

$$x(-(n+2)) = \{6, 5, 4, 3, 2, 1\}$$

⑧  $x(2n)$

hint: keep origin as it is

Take one, leave one

$$x(2n) = \{1, 3, 5\}$$

\* ⑨  $x(3n-1)$  (1+n-1) (3)

hint: Right (to def) after

((1+n-1) delay by 1)  $\leftrightarrow (n)$

2. Down scaling by 3

$$x(n) = \{1, 2, 3, 4, 5, 6\}$$

$$x(n-1) = \{1, 2, 3, 4, 5, 6\}$$

$$x(3n-1) = \{1, 2, 5\}$$

Alternate Solution

$$x(-n+2) = ?$$

Hint: Start from Right to Left

1. Delay by 2

2. Fold.

$$x(n) = \{1, 2, 3, 4, 5, 6\}$$

$$x(n-2) = \{1, 2, 3, 4, 5, 6\}$$

Long Method.

$$x(-n+2) = \{6, 5, 4, 3, 2, 1\}$$

Alternate Solution

$$x(-n+1)$$

1. Advance signal by 1

2. Fold.

$$x(n+1) = \{1, 2, 3, 4, 5, 6\}$$

$$x(-n+1) = \{1, 6, 5, 4, 3, 2, 1\}$$

~~$x(n) = \{1, 2, 3\}$~~

### ① Amplitude scaling

$A = 2$  multiplied to each term

$$y(n) = A x(n)$$

$$\Rightarrow 2 \cdot \{1, 2, 3\}$$

$$y(n) = \{2, 4, 6\}$$

### ② Time scaling

#### a) Downscaling

$\Rightarrow$  Decreasing the signal strength

$$x(n) = \{1, 2, 3\}$$

$$y(n) = 0.5 x(2n)$$

Hint: 1. Take origin 2. Take 1 leave 1

$$\Rightarrow y(n) = \{1, 3\}$$

#### Long Method:

$$y(-1) = x(2 \times -1) = 0$$

$$y(0) = x(0) = 1/07$$

$$y(1) = x(2 \times 1) = 3$$

$$y(2) = x(2 \times 2) = (0+1)$$

$$y(n) = \{1, 3\}$$

(b)

UpSampling  
 ⇒ Increase the signal strength  
 by inserting zeros

Eg:  $y(n) = x\left(\frac{n}{A}\right)$

Example:

$$x(n) = \{1, 2, 3\}; A=3$$

Short Method: Insert 2 zeros  
 between each sample.

$$x\left(\frac{n}{2}\right) = \{1, 0, 2, 0, 3\}$$

Long Method:

$$(n-a)x = (a)x$$

$$y(-1) = x\left(\frac{-1}{2}\right) = 0$$

$$y(0) = x\left(\frac{0}{2}\right) = 1$$

$$y(1) = x\left(\frac{1}{2}\right) = 0$$

$$y(2) = x\left(\frac{2}{2}\right) = 2$$

$$y(3) = x\left(\frac{3}{2}\right) = 0$$

$$y(4) = x\left(\frac{4}{2}\right) = 3$$

$$y(n) = x\left(\frac{n}{2}\right) = \{1, 0, 2, 0, 3\}$$

### 3. Folding

Eg:  $x(n) = \{1, 2, 3\}$

$$\uparrow$$

$$x(-n) = \{x(3), 2, 1\}$$

$$\uparrow$$

Eg 2:  $\{1, 2, 3, 4, 5\}$

$$\uparrow \quad \uparrow$$

$$x(-n) = \{5, 4, 3, 2, 1\}$$

$$\text{square } \uparrow \text{dine } \uparrow \text{repeat}$$

$$\{8, 0, 0, 0, 1, 2\} = (n)$$

### 4. Time Shifting

$\Rightarrow$  Result in delay or advance  
in signal

$$y(n) = x(n-k)$$

$(n-k) \rightarrow$  Delay  $\rightarrow$  Shift Right  $\rightarrow$   
 $(n+k) \rightarrow$  Advance  $\rightarrow$  Shift Left  $\leftarrow$

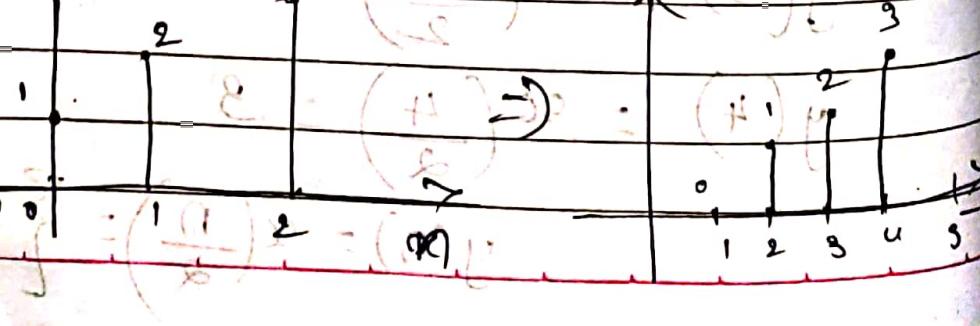
$$x(n) = \{1, 2, 3\}$$

$$\uparrow$$

$$x(n-k) \Rightarrow (\text{delay})$$

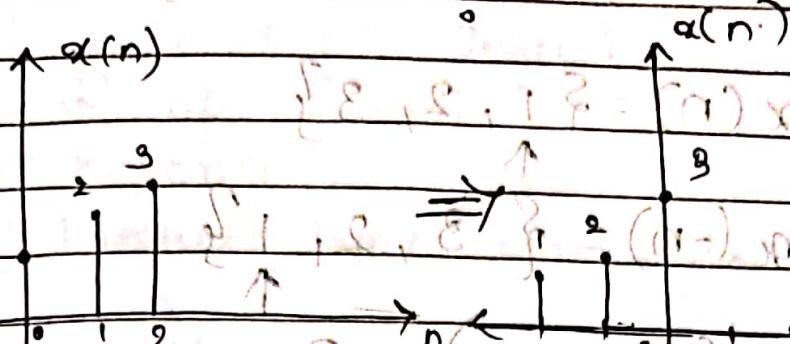
$$x(n+k) \Rightarrow \{0, 1, 2, 3\}$$

$$\uparrow \quad \uparrow \quad \uparrow \quad \uparrow$$



$x(n+k) \rightarrow$  Advance

$$x(n+2) = ?$$



$$x(n+2) = \{1, 2, 3\}$$

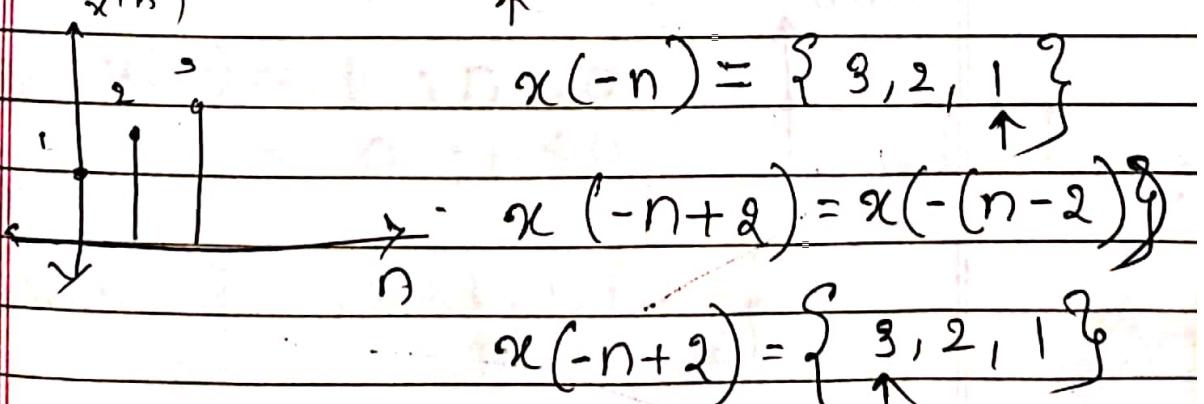
5. Fold and delay  
 $\downarrow$  [↓, shift]  $\rightarrow$  (Right)

replacement by  $-n$  (im)  $x(n)$

$$\begin{cases} x(n) & \xrightarrow{\text{Fold}} x(-n) \\ x(n) & \xrightarrow{\text{Delay}} x(n-k) \end{cases}$$

$$\therefore x(-n) \rightarrow x(-(n-k)) = x(-n+k)$$

$$\text{eg: } x(n) = \{1, 2, 3\} \quad x(n+2) = ?$$



$$x(-n+2) = \{3, 2, 1\}$$

7)

Fold & advance

↓

$x(-n)$

↑

$x(n+2)$

$x(n) = \{1, 2, 3\}$

$x(-n) = \{3, 2, 1\}$

$x(-n-2) = 9$

$x(-(n+2))$

By Fold.

By Advance bao blof

$(d) - x(-n) = \{3, 2, 1\}$

$x(-(n+2)) = \{9, 2, 1\}$

$(k+a-)x = ((k-a)-)x \leftarrow (a-)x$

$8-(g+a-)x = \{8, 2, 1\} \leftarrow (a-)x$

$\{1, 3, 8\} = (a-)x$

$8-(g-a-)x = (8+a-)x$

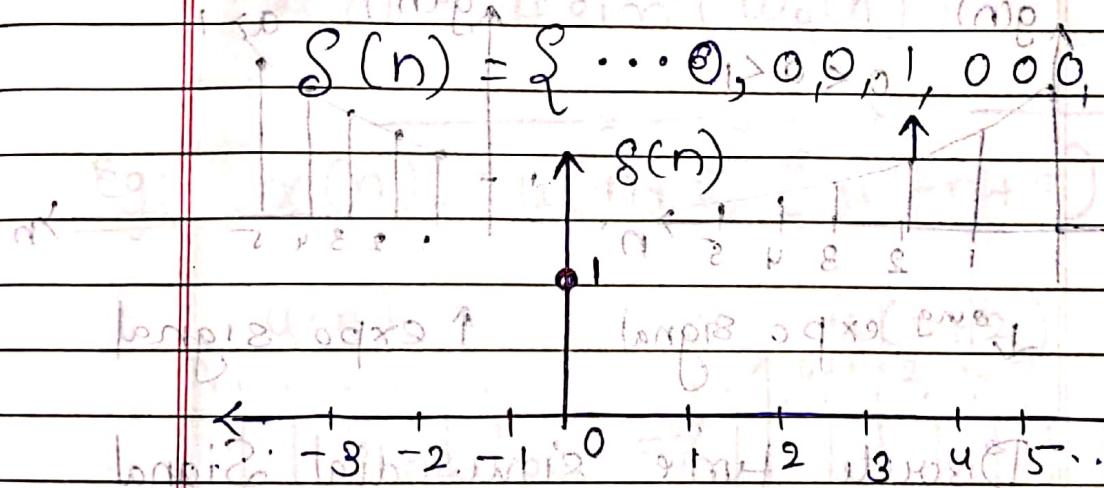
## \* Standard DT signals:

- ① Digital Impulse signal / unit sample sequence:  $s(n) = \delta(n)$
- ② Unit Step Signal
- ③ Ramp Signal
- ④ Exponential Signal
- ⑤ Discrete time sinusoidal signal.

### ① Digital Impulse signal / unit sample sequence:

$$s(n) = 1; n=0 \\ 0; n \neq 0$$

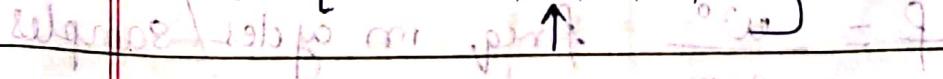
$$s(n) = \{ \dots, 0, 0, 0, 1, 0, 0, \dots \}$$



### ② Unit Step Signal: $u(n) = (1; n \geq 0, 0; n < 0)$

$$u(n) = \begin{cases} 1; n \geq 0 \\ 0; n < 0 \end{cases}$$

$$u(n) = \{ \dots, 0, 0, 1, 1, 1, 1, \dots \}$$

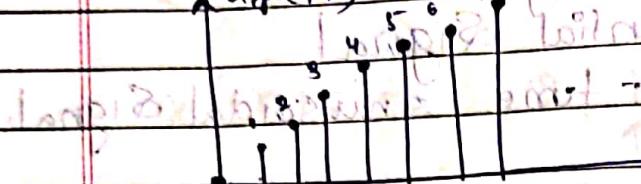


### 3. Ramp signal

$$u(n) = n ; n \geq 0$$

$$= 0 ; n < 0$$

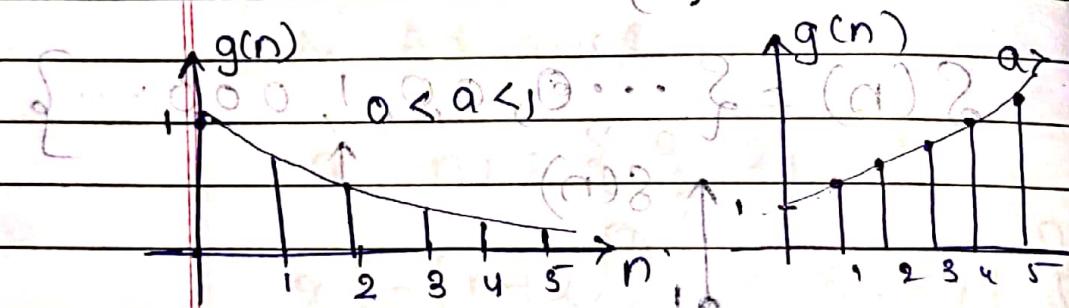
$u_r(n)$



### 4. Exponential Signal

$$g(n) = a^n ; n \geq 0$$

$$= 0 ; n < 0$$



$\downarrow$  dec exp signal

$\uparrow$  exp signal

### 5. Discrete time sinusoidal Signal

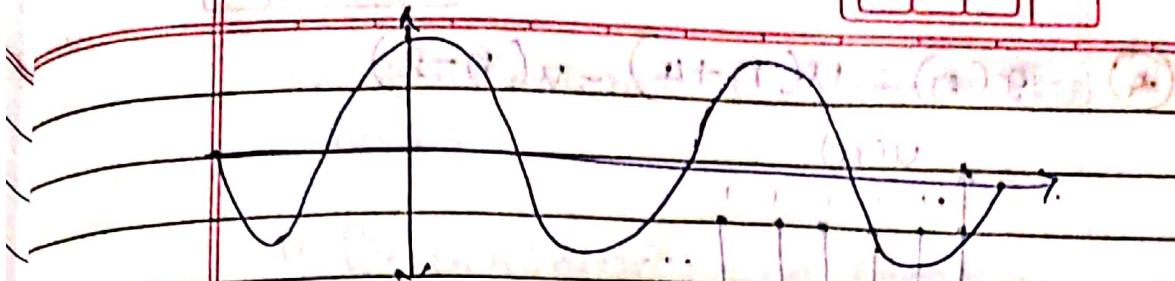
$$x(n) = A \cos(\omega_0 n + \theta) \quad -\infty < n < +\infty$$

$$x(n) = A \sin(\omega_0 n + \theta) \quad -\infty < n < +\infty$$

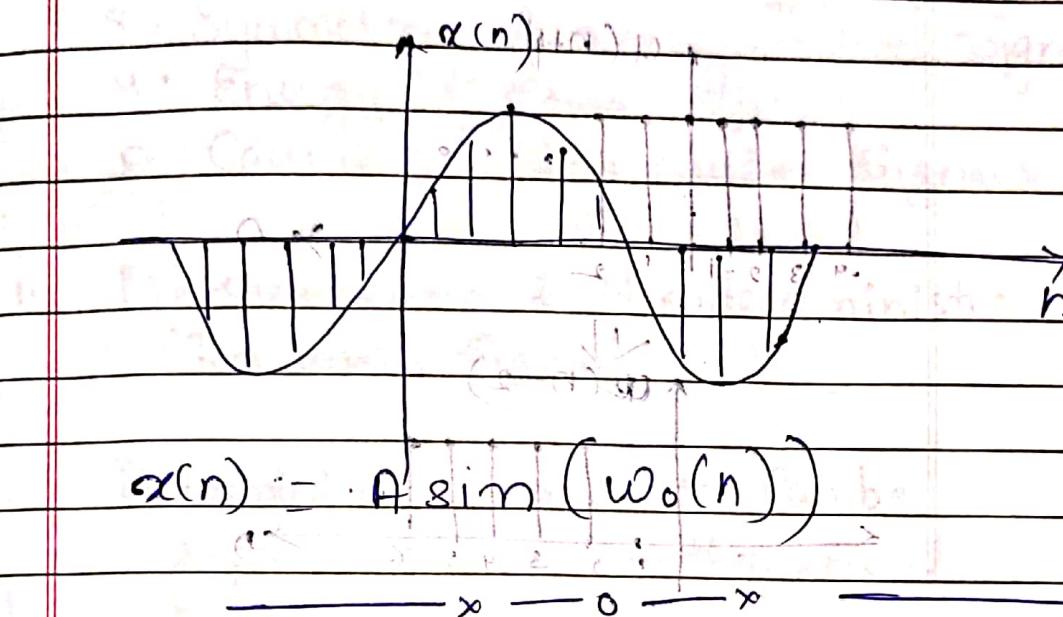
$\omega_0$  = freq in radians/sample

$\theta$  = Phase, in radians - (n)  $\pi$

$$f = \frac{\omega_0}{2\pi} \text{ freq in cycles/sample}$$

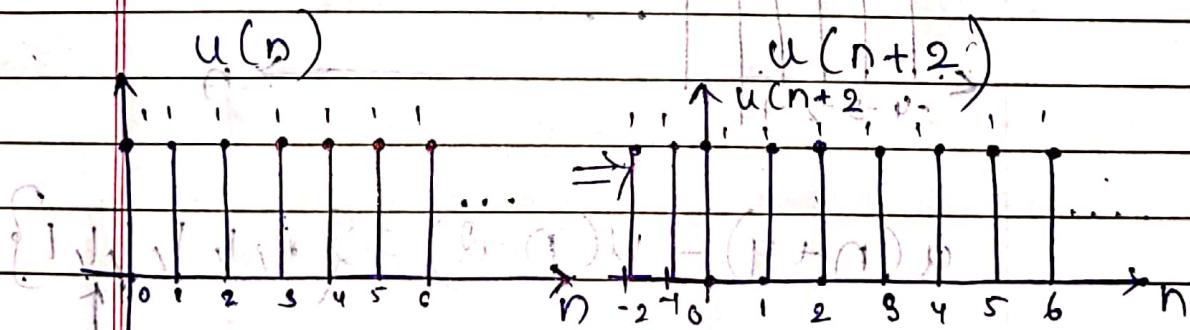


$$x(n) = A \cos(\omega_0 n)$$

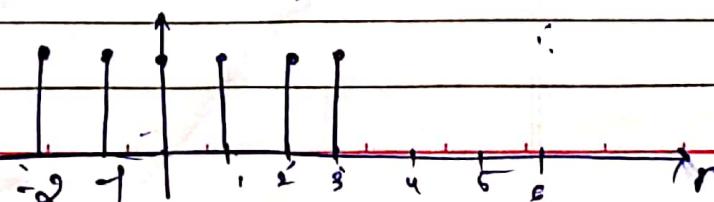
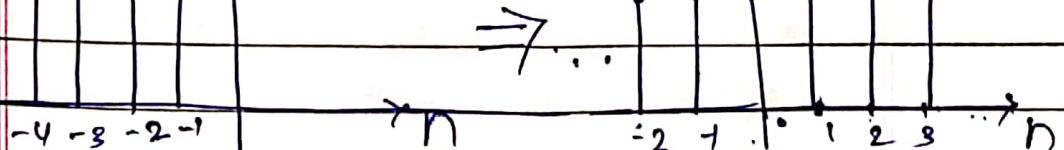
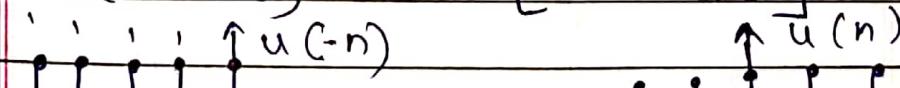


$$x(n) = A \sin(\omega_0 n)$$

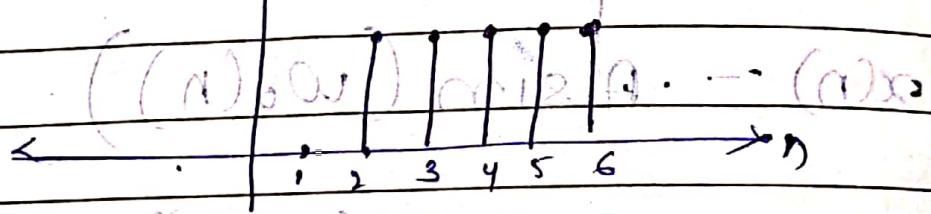
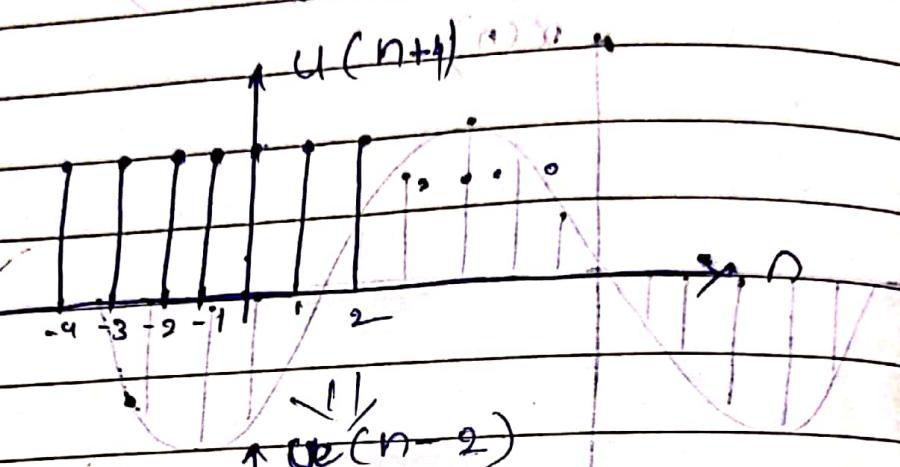
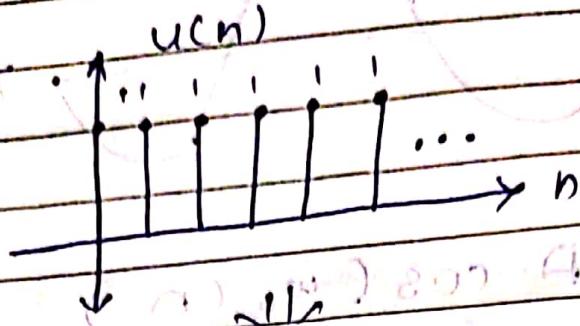
Eg:  $x(n) = u(n+2) - u(-n+3)$



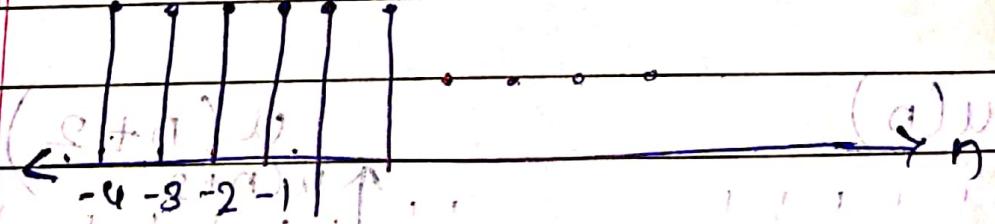
$$u(-n+3) \Rightarrow u[-(n-3)]$$



$$x(n) = u(n+4) - u(n-2)$$



$$(8+a-1)u \quad u(n+4) - u(n-2) \quad (a)u$$



$$u(n+4) - u(n-2) = \{ 1, 1, 1, 1, 0, 0, 0, 0 \}$$

$$(8-a-1)u \leq (8+a-1)u$$

## Classification of Discrete Time Signals

1. Deterministic and Non-deterministic Signals

2. Periodic & Aperiodic Signals

3. Symmetric (<sub>even</sub>) & antisymmetric (<sub>odd</sub>) Signals

4. Energy & Power Signals

Causal & Non-causal Signals

Memoryless & Non-memoryless

5. Deterministic & Non-deterministic (Random) Signal: (a) (b)

Deterministic Signal  $\rightarrow$  Can be represented in mathematical Equations showing math

Eg: Sinusoidal signal  $x(n) = \cos \omega n$   
Exponential signal, ramp signal

Random Signal  $\rightarrow$  Cannot be represented by any mathematical equation

Eg: Noise signal

$$x(n) = (\alpha \pi / 10.0) \sin(\alpha \pi n)$$

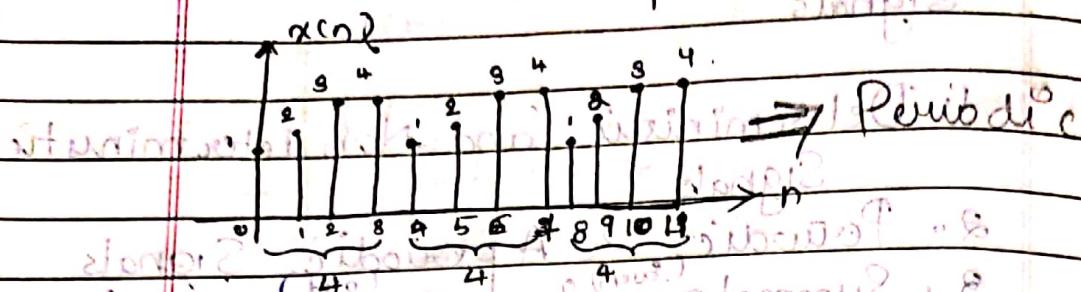
$$\alpha \pi / 10.0 = \omega \Rightarrow \omega = (\alpha \pi) \times$$

$$\pi / 10.0 = \alpha$$

$\pi / 10.0$  is random noise

$\therefore$   $\pi / 10.0$  is a random signal

## 2. Periodic & Aperiodic Signal



The above signal is said to be periodic as the signal is getting repeated after 4 intervals

$$x(n) = x(n+N)$$

$\leftarrow$  Period of Signal.

Non Periodic Signal:

$$\left[ x(n) \neq x(n+N) \right]$$

Eg: Determine whether given signal

is periodic or Aperiodic.

If periodic then find fundamental period

✓ ①  $\cos(0.01\pi n)$

$$x(n) = \cos(0.01\pi n)$$

$$x(n) = \cos \omega n$$

$$\therefore \omega = 0.01\pi$$

$$\therefore 2\pi f = 0.01\pi$$

$$f = \frac{0.01\pi}{2\pi} = \frac{1}{200} \text{ cycles}$$

$$\therefore N = 800$$

$\therefore f = \frac{1}{N} \rightarrow$  Condition for periodicity

(ii)

$$\cos(8\pi n)$$

$$T = \frac{1}{8}$$

$$x(n) = \cos(8\pi n)$$

$$T = \frac{1}{8}$$

$$x(n) = \cos(\omega n)$$

$$T = \frac{1}{8}$$

$$\omega = 8\pi$$

$$2\pi f = 8\pi$$

$$f = \frac{8\pi}{2\pi} = 4$$

$\therefore f$  is ratio of rational nos.

∴  $\cos(8\pi n)$  is a periodic signal

By condition of periodicity

$$800 = NK \Rightarrow T = \frac{800}{N}$$

$\therefore T = 200 \Rightarrow$  fundamental Period

(iii)

$$x(n) = \sin(3\pi n)$$

✓

$$x(n) = \sin(\omega n)$$

$$\omega = 3$$

$$2\pi f = 3$$

$$\therefore f = \frac{3}{2\pi}$$

$\therefore f$  is ratio of irrational nos.

$\therefore \sin(3\pi n)$  is an aperiodic signal

$$\text{iv) } \cos\left(\frac{n}{8}\right) \cdot \cos\left(\frac{\pi n}{8}\right)$$

$$y(n) = \cos(\omega_1 n) \cdot \cos(\omega_2 n)$$

$$\omega_1 = \frac{1}{8} \quad \text{&} \quad \omega_2 = \frac{\pi}{8}$$

$$\Delta \pi f_1 = \frac{1}{8} \quad \text{for } \Delta \pi f_2 = \frac{\pi}{8}$$

$$f_1 = \frac{1}{16\pi} \quad (\text{as } \Delta \pi f_1 = \frac{1}{8})$$

$$f_2 = \frac{\pi}{16\pi} = \frac{\pi}{16}$$

$f_1$  is Aperiodic

$f_2$  is periodic in  $n$

Multiplication of periodic & Aperiodic is Aperiodic

$$\text{v) } y(n) = \sin \frac{\pi}{5} n + \sqrt{2} \cos \frac{\pi}{5} n$$

$$y(n) = y_1(n) + y_2(n)$$

$$y_1(n) = \sin \frac{\pi}{5} n$$

$$(\text{as } \Delta \pi f_1 = \frac{\pi}{5})$$

$$= \sin \omega_1 n = (a)_{10}$$

$$\therefore \omega_1 = \frac{\pi}{5} \quad 8 = 0.1$$

$$\Delta \pi f_1 = \frac{\pi}{5} = \frac{\pi}{5}$$

$$2\pi \text{ period} = \frac{\pi}{\Delta \pi f_1} = \frac{\pi}{0.1} = 10$$

$$\text{length of signal } N = \frac{10}{\Delta \pi f_1} = 10$$

Periodic signal  $\Rightarrow N_1 = 10$

$$y_2(n) = 2 \cos \frac{\pi}{\sqrt{2}} n \quad (a) \text{ or } (\text{ir})$$

~~$$y_2(n) = A \cos \frac{\omega}{\sqrt{2}} n$$~~

$$\therefore y_2(n) = 2 A \cos \omega n$$

$$\omega = \frac{2\pi}{N} \cdot 2\sqrt{2}$$

$$\frac{2\pi}{\sqrt{2}} = \omega$$

$$2\pi f = \frac{\pi}{\sqrt{2}} \Rightarrow \pi f = \frac{\pi}{\sqrt{2}}$$

$$f = \frac{1}{\pi} \cdot \frac{\pi}{\sqrt{2}} = \frac{1}{\sqrt{2}}$$

$$f = \frac{1}{2\sqrt{2}}$$

$$\therefore f = \frac{1}{2\sqrt{2}}$$

$$N_2 = 2\sqrt{2}$$

$$\therefore 8 = 1 \times 8$$

$y(n)$  is Periodic if  $N_1/N_2$  is  
ratio of (2 integers).  $\text{or } (iir)$

$$\therefore \frac{N_1}{N_2} = \frac{10}{2\sqrt{2}} = \frac{5}{\sqrt{2}}$$

$\therefore y(n)$  is non  $\neq$  Periodic in nature

$$\frac{1}{\pi f} = \frac{1}{\pi \cdot \frac{1}{2\sqrt{2}}} = 2\sqrt{2}$$

on condition to make it periodic

$$\text{vi) } x_1(n) = e^{j\left(\frac{\pi}{4}n\right)} \quad (\text{a.p})$$

(v) For discrete time, complex exponential equation is

$$x(n) = e^{j\omega n}$$

$$\omega = \frac{\pi f}{T}$$

$$2\pi f = \frac{\pi}{4} + \pi k$$

$$f = \frac{k}{4}$$

$$k = 1, 2, 3, \dots$$

$$= \frac{1}{8} = f$$

$$f = \frac{R}{N} = \frac{1}{8}$$

$$\therefore N = 8$$

$$\text{vii) } x_2(n) = 8 \sin\left(\frac{\pi}{8}n\right)$$

$$\therefore x(n) = 0.1 A \sin \omega n$$

$$\omega = \frac{2\pi f}{T} = \frac{\pi}{8}$$

$$\therefore f = \frac{1}{16} \text{ Hz}$$

$\therefore f$  is ratio of irrational no  
 $\therefore$  Non-periodic

viii)  $x(n) = \cos\left(\frac{\pi}{3} + 0.3n\right)$

$x(n) = \cos(1\theta + \omega_n)$

$\therefore \omega = 0.3 = (f) \text{ & } \theta = \frac{\pi}{3}$

$\omega T = 0.3 = 0, (\text{Phase})$

$\omega T P = 10.3, \omega T \text{ non rational}$

001 &  $\omega T$

$x(t) = \cos\frac{3}{2\pi t}$

which is not rational  
 $\therefore x(n)$  is non-periodic.

Continuous Signal:

$x(t) \triangleq x(t + T_0)$

↑

(Period)

Note : Discrete Time signal is represented in 'n'.  
Continuous time signal is represented in 't'.

continuous time signal is represented in 't'.

$T_0 = \frac{2\pi}{\omega}$

Eg: (1)  $\alpha(t) = 2 \cos 100\pi t + 5 \sin 50t$

$$\alpha(t) = \alpha_1(t) + \alpha_2(t)$$

$$\alpha_1(t) = 2 \cos 100\pi t$$

$$\alpha_1(t) = A \cos(\omega_1 t)$$

$$(1) \omega = 100\pi$$

$$2\pi f_1 = 100\pi$$

$$f_1 = 50$$

or  $\omega$

~~100~~

But  $f_1 = \frac{1}{T_1}$  (freq. is inverse of time period)

∴  $T_1 = \frac{1}{f_1}$

$$\therefore \frac{1}{T_1} = \frac{100}{2\pi}$$

$$\therefore T_1 = \frac{2\pi}{100} = \frac{\pi}{50}$$

↑

(2)  $\alpha_2(t) = 5 \sin 50t$

$$\therefore \alpha_2(t) = A \sin \omega_2 t$$

in L.H.P.D. write  $50\pi t$  :  $50t$

$$\therefore \omega_2 = 50$$

$$2\pi f_2 = 50$$

in L.H.P.D. write  $50t$  :  $50t$

$$\therefore f_2 = \frac{50}{2\pi}$$

$$T_2 = \frac{1}{f_2} = \frac{2\pi}{50}$$

$$\frac{T_1}{T_2} = \frac{1}{50} \div \frac{2\pi}{50}$$

$$= \frac{1}{50} \times \frac{50}{2\pi} \\ = \frac{1}{2\pi}$$

$x(t)$  is non periodic.

(ii)  $x(t) = 3 \sin 4t + A$  (v)

$$\omega = 4$$

$$2\pi f = 4 \Rightarrow f = 2$$

$$(4t + 2\pi) : (A)$$

∴ Non-Periodic Signal

(iii)  $x(t) = 10 \sin(12\pi t) + 4 \sin(18\pi t)$

$$x_1(t) = 10 \sin(12\pi t) + x_2(t)$$

$$x_1 = 10 \sin(12\pi t) \quad x_2 = 4 \sin(18\pi t)$$

$$x_1 = A \sin \omega_1 t$$

$$\omega_1 = 12\pi$$

$$2\pi f_1 = 12\pi$$

$$f_1 = 6$$

$$\frac{1}{T_1} = 6$$

$$\frac{1}{T_1} = 6$$

$$\therefore T_1 = 1/6$$

$$\frac{1}{T_2} = 9$$

$$\therefore T_2 = 1/9$$

$$\frac{T_1}{T_2} = \frac{1/1}{1/6} \div \frac{1/9}{1/1}$$

$$\frac{T_1}{T_2} = \frac{1}{1} \times \frac{9}{1}$$

$$= \frac{1}{1} \times \frac{9}{1}$$

∴ continuous Hence periodic signal

(iv)  $x(t) = 3 + t + t^2 \cdot 8 = \text{(i)} + \text{(ii)}$

$\therefore$  A continuous signal is said to be periodic of  $T_0 = \frac{2\pi}{\omega}$

$$x(t) = x(t + T_0)$$

Replace  $t$  with  $t + T_0$  in

$$(FT81) \quad 3 + (t + T_0) + t^2 \cdot 8 = (t) + (t + T_0)^2$$

$$3 + t + T_0 + t^2 + 2tT_0 + T_0^2$$

$$(FT81) \quad 3 + t + T_0 + t^2 + 2tT_0 + T_0^2 = 3 + t + t^2 \cdot 8$$

$$\therefore 3 + t + T_0 + t^2 + 2tT_0 + T_0^2 = 3 + t + t^2 \cdot 8$$

$$(FT81) \quad \therefore x(t) \neq x(t + T_0)$$

$$(FT81) \quad \therefore T_0 = 9\pi$$

$\therefore$  A periodic signal if

$$P = \frac{1}{\omega T}$$

$$P = \frac{1}{\omega T}$$

$$P = \frac{1}{\omega T} = \frac{1}{\omega} \cdot \frac{1}{T}$$

$$P = \frac{1}{T}$$

## Even & Odd Signal :

\* Even signal : (Symmetric signal)  
 $x(n) = x(-n)$  for all  $n$ .

→ Symmetrical across vertical axis

→ Eg: cosine signal.

1. Function  $x(n)$



↳  $x(n) = \cos(\omega n)$

$$x(n) = (\cos(\omega n))_n + (\sin(\omega n))_n = \cos(\omega n)$$

$x(n)$

$$(\cos(\omega n))_n - (\sin(\omega n))_n = \cos(\omega n)$$



\* \*

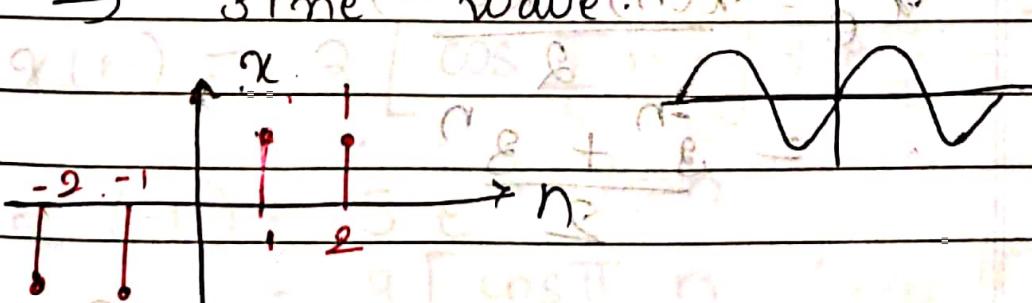
Odd Signals: (antiSymmetric Signal)

$$\cancel{x(-n)} = -\cancel{x(n)} = (\alpha)_n$$

$$x(n) = \cancel{\alpha} - \cancel{x(-n)} \text{ for all } n.$$

→ Antisymmetric across vertical axis

→ Sine wave



$$x(n) = \sin(\omega n) = (\sin(\omega n))_n + (\cos(\omega n))_n = \sin(\omega n)$$

$\sin(\omega n)$

Any signal is sum of even + odd components

$$\text{Signal} = (\text{even}) + (\text{odd})$$

Signal = even + odd

$$\text{Signal} = x(n) = x_e(n) + x_o(n)$$

~~Odd component~~

Even component

Odd component

$$x_e = \frac{x(n) + x(-n)}{2}$$

$$x_o = \frac{x(n) - x(-n)}{2}$$

Eg: ① Find whether given signal is even or odd

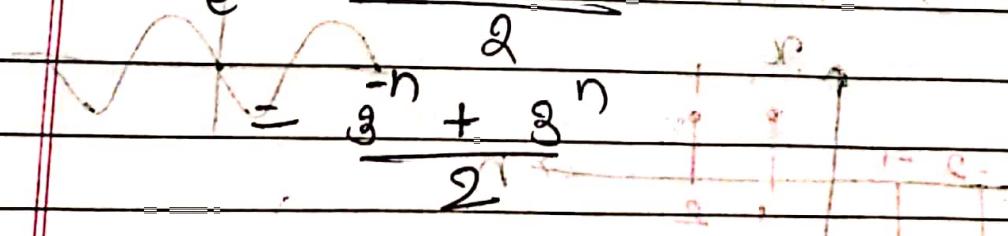
$$x(n) = 3^{-n}$$

$$\text{or } x(-n) = 3^{-(-n)} = 3^n$$

$$x(-n) = 3^n$$

Even function

$$x_e = x(n) + x(-n)$$



$$x_o = \frac{x(n) - x(-n)}{2} = \frac{3^{-n} - 3^n}{2}$$

Eg (ii)  $x(n) + (q(n)) = e^{jn}$

$$x(n) = \cos \theta n + j \sin \theta n$$

$$x(n) = x_e(n) + x_o(n)$$

$$e^{jn} = x_e(n) + x_o(n)$$

$$\therefore x_e(n) = \cos \theta n$$

$$x_o(n) = j \sin \theta n$$

$$x_e(n) = x(n) + x(-n)$$

$$[(a-1)x - (a)x]_1 = (a)_e x$$

$$= e^{jn} + e^{-jn}$$

$$[(a\frac{\pi}{5}) \cos \theta n + (a\frac{\pi}{5}) \sin \theta n]_1 = \cos n.$$

$$x_o(n) = x(n) - x(-n)$$

$$x_o(n) = \frac{j}{2} (e^{jn} - e^{-jn})$$

$$= j \sin \theta n$$

Eg (iii)  $x(n) = 3 e^{j\pi/5 n}$ .

$$x(n) = 3 \left[ \cos \frac{\pi}{5} n + j \sin \frac{\pi}{5} n \right]$$

$$x(-n) = 3 e^{-j\pi/5 n}$$

$$= 3 \left[ \cos \frac{\pi}{5} n - j \sin \frac{\pi}{5} n \right]$$

$$\text{Even Part } x_e(n) = \frac{1}{2} [x(n) + x(-n)]$$

$$= \frac{1}{2} [3 \cos \frac{\pi}{5} n + 3j \sin \frac{\pi}{5} n]$$

$$= \frac{1}{2} [3 \cos \frac{\pi}{5} n - 3j \sin \frac{\pi}{5} n]$$

$$= \frac{1}{2} [6 \cos \frac{\pi}{5} n]$$

$$= (a) \cos \frac{\pi}{5} n$$

$$= 3 \cos \frac{\pi}{5} n$$

$$(a)_e = (a) \cos \frac{\pi}{5} n$$

$$\text{Odd part } x_o(n) = \frac{1}{2} [x(n) - x(-n)]$$

$$= \frac{1}{2} [3 \cos \frac{\pi}{5} n + 3j \sin \frac{\pi}{5} n] -$$

$$= \frac{1}{2} [3 \cos \frac{\pi}{5} n - 3j \sin \frac{\pi}{5} n]$$

$$= \frac{1}{2} [6j \sin \frac{\pi}{5} n]$$

$$= (3j) \sin \frac{\pi}{5} n$$

Eg:  $x(n) = \{1, 2, 1, 3\}$

$$x(-n) = \{3, 2, 1, 1\}$$

$$x_e(n) = x(n) + x(-n)$$

$$= \{1, 2, 1, 3\} + \{3, 1, 2, 1\}$$

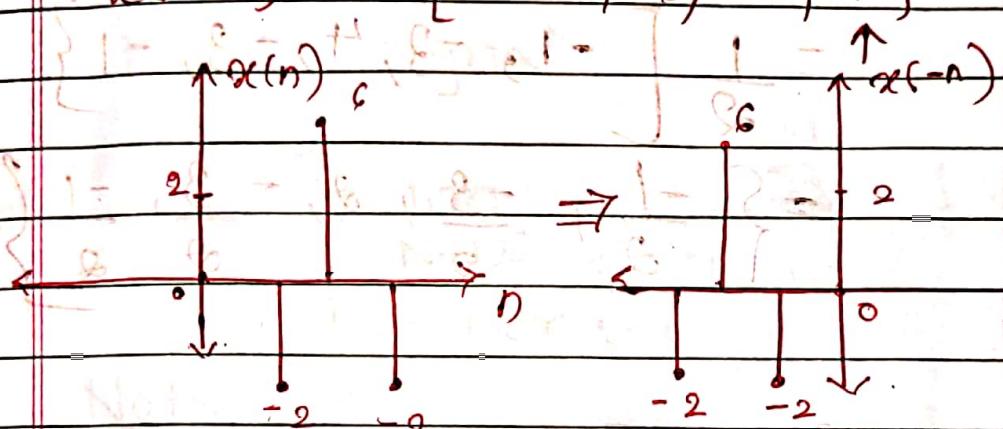
$$x_0(n) = \left\{ \begin{array}{l} 3, 1, 2, 8, 0, 1, 3 \end{array} \right\}$$

$$= \left\{ \frac{3}{2}, \frac{1}{2}, 1, 1, 1, 1, 3 \right\}$$

$$[(a-1) \oplus b] \oplus [(a \oplus b)] = (a \oplus b)$$

$$\textcircled{x}: x(n) = \left\{ \begin{array}{l} 2, -2, 6, -2 \end{array} \right\}$$

$$x(-n) = \left\{ \begin{array}{l} -2, 6, -2, 2 \end{array} \right\}$$



$$[x(n-1) \oplus x(n)] \oplus [x(n)] = (n)_0$$

$$x_e(n) = x(n) + x(-n)$$

$$(B-E), (A+1), (Q-E)$$

$$\frac{1}{2} \left[ (-2+0), (6+0), (-2+0), (2+2) \right]$$

$$= \frac{1}{2} \left[ 0, -2, 6, -2, 4, -2, 6, -2 \right]$$

$$= \left\{ \begin{array}{l} -1, 3, -1, 2, -1, 3, -1 \end{array} \right\}$$

$$x_o(n) = \frac{1}{2} (x(n) - x(-n))$$

$$\left\{ \begin{array}{l} 1, -3, 1, 0, 1, 3, -1 \end{array} \right\} = \frac{1}{2} \left\{ \begin{array}{l} 2, -6, 2, 0, -2, 6, -2 \end{array} \right\}$$

$$\text{iii) } \alpha(n) = \{-3, 1, 2, -4, 2\}$$

$$\alpha(-n) = \{2, -4, 1, -3\}$$

$$\alpha_e(n) = \frac{1}{2} [\alpha(n) + \alpha(-n)]$$

$$= \frac{1}{2} [(-3+2), (1-4), (2+1), (-4+1)]$$

$$= \frac{1}{2} [-1, -3, 4, -3, -1]$$

$$= \left\{ -\frac{1}{2}, -\frac{3}{2}, 2, -\frac{3}{2}, -\frac{1}{2} \right\}$$

$$\alpha_o(n) = \frac{1}{2} [\alpha(n) - \alpha(-n)]$$

$$= \frac{1}{2} [(-3-2), (1+4), (2-2)]$$

$$= \frac{1}{2} [-5, 5, 0, -5, 5]$$

$$= \left\{ -\frac{5}{2}, \frac{5}{2}, 0, -\frac{5}{2}, \frac{5}{2} \right\}$$

$$= \{1, 8, 1, 8, 1, 8, 1, 8\} =$$

## \* Energy & Power Signal

Energy signal:

$$E = \sum_{n=-\infty}^{\infty} |x(n)|^2$$

$E = \text{non-zero \& finite.}$

then it is energy signal

Power signal

$$P = \lim_{N \rightarrow \infty} \frac{1}{2N+1} \sum_{n=-N}^{N} |x(n)|^2$$

$P = \text{non-zero \& finite.}$

Note:

| For Energy signal  $\Rightarrow P \neq 0$  |

| For Power signal  $\Rightarrow E = \emptyset$  |

| No signal is both Energy & Power signal. |

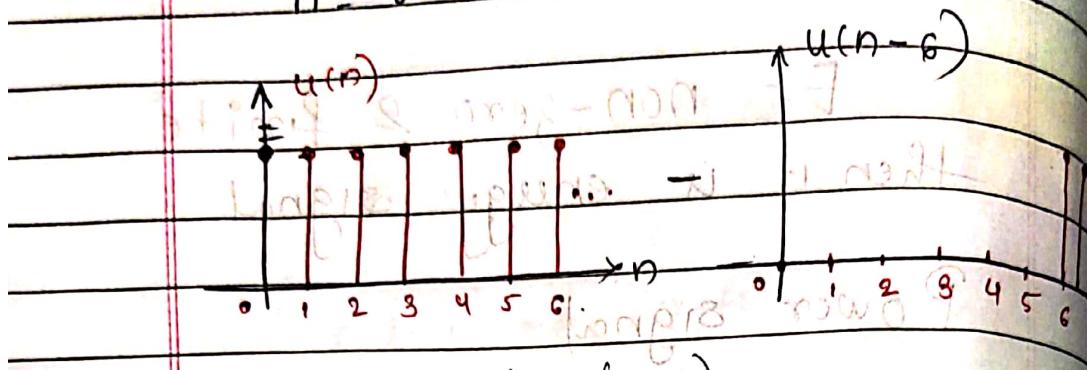
| Every bounded & periodic signal  $\Rightarrow$  Power Signal |

| Every bounded & non-periodic signal  $\Rightarrow$  Energy. |

$$\text{Eg: } x(n) = u(n) - u(n-6)$$

$$E = \sum_{n=-\infty}^{\infty} |x(n)|^2$$

$$= \sum_{n=-\infty}^{\infty} |u(n) - u(n-6)|^2$$



$$|x(n)|^2 = \sum_{n=0}^{14} |u(n) - u(n-6)|^2$$

~~$$E = \sum_{n=-1}^{\infty} |x(n)|^2$$~~

~~$$E = 0 + 1 + 1 + 1 + 1 + 1 + 1$$~~

$$E = 6 \text{ joules}$$

$$\text{Eg: } x(n) = n u(n)$$

~~$$E = \sum_{n=-\infty}^{\infty} |x(n)|^2$$~~

~~$$= \sum_{n=-\infty}^{\infty} |nu(n)|^2$$~~

$$= \sum_{n=-1}^{\infty} n^2 (u(n))^2 + \sum_{n=0}^{\infty} n^2 (u(n))^2$$

$$= \cdot 0 + \sum_{n=0}^{\infty} n^2 \cdot (-1)^2$$

$$= \sum_{n=0}^{\infty} n^2$$

$$E = \infty$$

Power Signal

$$P = \lim_{N \rightarrow \infty} \frac{1}{2N+1} \sum_{n=-N}^{N} |\alpha(n)|^2$$

$$= \lim_{N \rightarrow \infty} \frac{1}{2N+1} \sum_{n=0}^{N} |n \cdot u(n)|^2$$

$$= \lim_{N \rightarrow \infty} \frac{1}{2N+1} \sum_{n=0}^{N} n^2$$

$$= \lim_{N \rightarrow \infty} \frac{1}{2N+1} [0^2 + 1^2 + 2^2 + 3^2 + \dots + N^2]$$

$$= \lim_{N \rightarrow \infty} \frac{1}{2N+1} [0 + 1 + 2 + \dots + \infty]$$

$$= \infty$$

Energy infinite &  $P = 0$

$\therefore$  Energy signal.

$$2) * x(n) = (0.5)^n u(n)$$

$$\begin{aligned} E &= \sum_{n=-\infty}^{\infty} |x(n)|^2 \\ &= \sum_{n=-\infty}^{\infty} |(0.5^n u(n))|^2 \end{aligned}$$

$$\therefore \sum_{n=-\infty}^{-\infty} u(n) = 0$$

$$|x(n)|^2 = \sum_{n=0}^{\infty} |0.5^n u(n)|^2$$

$$(n) = \left| \sum_{n=0}^{\infty} (0.25)^n \right|^2$$

Geometric Series: sum

$$\sum_{n=0}^{\infty} A^n = \frac{1}{1-A}$$

$$\left[ 0 + 0.25 + 0.0625 + \dots \right] = \frac{1}{1-0.25}$$

$$= \frac{1}{0.75}$$

$$= 4/3$$

$\Rightarrow$  Energy signal

$$P = \lim_{N \rightarrow \infty} \frac{1}{QN+1} \sum_{n=-N}^N |x(n)|^2.$$

$$= \lim_{N \rightarrow \infty} \frac{1}{QN+1} \sum_{n=-N}^N |e^{j\omega n} u(n)|^2$$

$$= \lim_{N \rightarrow \infty} \frac{1}{QN+1} \sum_{n=0}^{\infty} 0.25^n \cdot 1$$

Geometric Series

$$\sum_{n=0}^N A^n = A^{N+1} - 1 = q$$

$$\lim_{N \rightarrow \infty} \frac{A^{N+1} - 1}{QN+1} = \frac{0.25^{N+1} - 1}{0.25 - 1}$$

$$\lim_{N \rightarrow \infty} a^n = 0 \text{ if } a < 1$$

$$\therefore 0.25^{N+1} = 0$$

$$= (1) \cdot 0 \quad \lim_{N \rightarrow \infty} 1 = 0$$

$$\lim_{N \rightarrow \infty} \frac{1}{QN+1} = 0$$

$$\lim_{N \rightarrow \infty} \frac{1}{QN+1} = 0$$

$$x(n) = n u(n) \quad P = 0$$

$$E = \sum_{n=-\infty}^{\infty} |x(n)|^2$$

$$E = \sum_{n=-\infty}^{\infty} |x(n)|^2$$

$$= \sum_{n=0}^{\infty} n^2$$

$$= 1^2 + 2^2 + 3^2 + 4^2 + \dots \infty$$

Not an Energy signal

$$P = \lim_{N \rightarrow \infty} \frac{1}{2N+1} \sum_{n=-N}^{N} n^2 x(n)$$

$$\begin{aligned} P &= \lim_{N \rightarrow \infty} \frac{1}{2N+1} \sum_{n=0}^{N} n^2 \\ &\stackrel{(1)}{=} \lim_{N \rightarrow \infty} \frac{1}{2N+1} \sum_{n=0}^{N} n(n+1) \\ &\stackrel{(2)}{=} \lim_{N \rightarrow \infty} \frac{1}{2N+1} \sum_{n=0}^{N} (n^2 + n) \\ &\stackrel{(3)}{=} \lim_{N \rightarrow \infty} \frac{1}{2N+1} \left( \frac{N(N+1)(2N+1)}{6} + \frac{N(N+1)}{2} \right) \end{aligned}$$

$$\therefore \text{Power} = 18 \infty$$

∴ Not a Power signal

$$x(n) = \begin{cases} n^2 & 0 \leq n \leq 3 \\ 10-n & 4 \leq n \leq 6 \\ 0 & 7 \leq n \leq 9 \\ 0 & \text{otherwise} \end{cases}$$

$$\begin{aligned} E &= \sum_{n=-\infty}^{\infty} |x(n)|^2 \\ &= \sum_{n=-\infty}^{-1} 0 + \sum_{n=0}^3 (n^2)^2 + \sum_{n=4}^6 (10-n)^2 + \sum_{n=7}^9 0 \\ &= 0^2 + 1^2 + 2^2 + 3^2 + (10-4)^2 + (10-5)^2 + (10-6)^2 \end{aligned}$$

$$= 0^4 + 1^4 + 2^4 + 3^4 + \sum_{n=4}^6 (100 - 20n + n^2)$$

$$\begin{aligned}
 &= 0 + 1 + 16 + 81 + 100 - 20 \times 4 + 4^2 + \\
 &\quad 100 - 20 \times 5 + 5^2 + 100 - 20 \times 6 + 6^2 \\
 &\quad + 49 + 64 + 81 \\
 &= 369 \text{ J}
 \end{aligned}$$

So, Energy of signal

$$P = \lim_{N \rightarrow \infty} \frac{1}{2N+1} \sum_{n=-N}^N |x(n)|^2$$

$$= \lim_{N \rightarrow \infty} \frac{1}{2N+1} \left[ \sum_{n=-1}^{N-1} 0 + \sum_{n=0}^{N-1} 6^4 + \sum_{n=0}^{N-1} (10^{-4})^2 + \dots \right]$$

For large N, terms at  $\pm 10^{-4}$  are negligible.

$$= \lim_{N \rightarrow \infty} \frac{1}{2N+1} [869] \quad (a) \rightarrow C$$

$$= \frac{1}{2\infty+1} [869] \quad (b) \rightarrow C$$

$$\underset{\infty}{=} 0 \cdot 869 = 0$$

Power is zero. Energy is

## \* Causal & Non Causal Signal.

Causal  $x(n) = 0 ; n < 0$

⇒ Positive signal.

e.g.  $u(n)$

Non-Causal

Anti-Causal  $x(n) = 0 ; n > 0$

⇒ Negative signal

e.g.  $u(-n)$

Non-causal

⇒ A signal which exist from +ve & -ve time.

Convolution

→ Mathematical operation, just like multiplication

Addition, to combine 2 signals

to form a third signal.

→ denoted by \*

→  $y(n) = x(n) * h(n)$

→ e.g.

Types of Convolution

(1) Linear Convolution

(2) Circular Convolution

(3) Linear using Circular Convolution

## ① Linear Convolution

→ Calculates O/P for any LTI  
(Linear Time Invariant System)

given its input & its impulse response

### Method 1: Matrix Method

$$x(n) = \{4, 2, 1, 3\}$$

$$h(n) = \{1, 2, 1, 3\}$$

$$y(n) = x(n) * h(n)$$

$$\{4, 2, 1, 3\} = (a)d$$

$$\begin{array}{c|ccccc} 4 & -4 & 8 & 8 & 4 \\ 2 & (a)2 & +4(a)4 & 2 & (a)2 \\ 1 & (a-1)d & +1(a)2 & 2 & 1 \\ 3 & 3 & 6 & 6 & 3 \end{array}$$

$$y(n) = \{4, 10, 13, 13, 10, 4, 9\}$$

$$d = f(a) \times b^m$$

How to decide origin of  $y(n)$

lower index of  $x(n)$  + lower index of  $h(n)$

$$-2 - 1 = -3$$

$\therefore$   $y(n)$  starts at  $n = -3$

### Method 2: Graphical Method

Procedure:

1. Replace  $n$  with  $m$ .

$\therefore x(n) * h(n)$  is replaced to  $x_+(m) * h_-(m)$

2. Sketch  $x_+(m)$  &  $h_-(m)$

3. Fold  $h(m)$  to  $h(-m)$

4. Shift  $h(-m)$  to left &

5. Multiply  $x(m) \& h(-m)$ .

6. Sum up all samples of O/P sequence

7. To get next sample of O/P sequence shift  $h(-m)$  of previous step.

to one position right & multiply & sum.

8. Repeat steps.

Eg:

$$x(n) = \{1, 2, 0.5, 1\}$$

$$h(n) = \{-1, 2, 1, -1\}$$

$$\begin{aligned}y(n) &= x(n) * h(n) \\&= \sum_{m=-\infty}^{\infty} x(m) * h(n-m)\end{aligned}$$

lower index  $[x(n)] = 0$ .

lower index  $[h(n)] = -1$

O/P starts at -1.

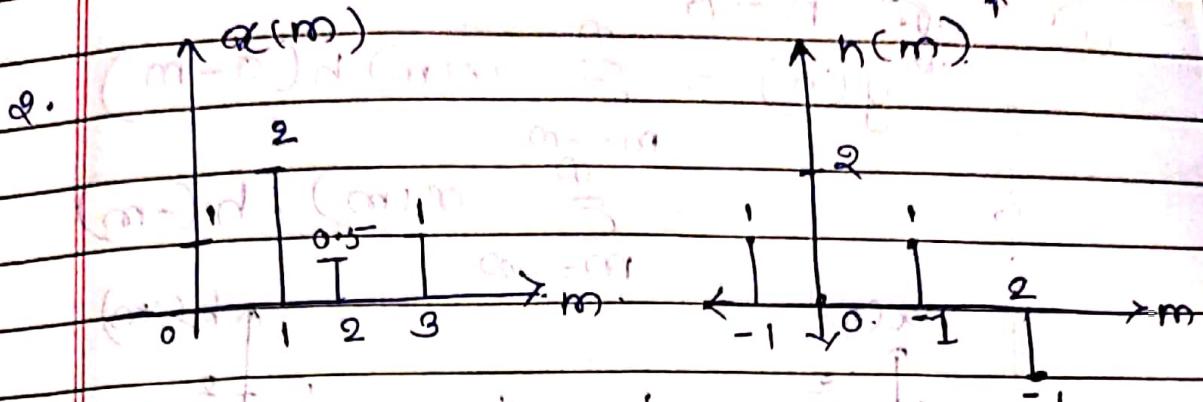
$$\text{len}[x(n)] = 4$$

$$\text{len}[h(n)] = 4$$

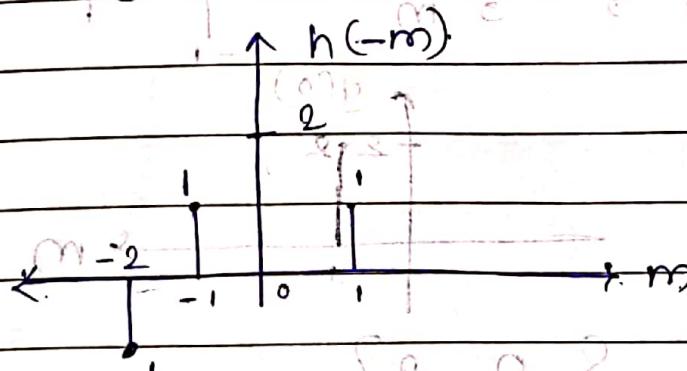
$$4+4-1 = 7$$

O/P sequence consists of 7 samples

1. Replace  $x(n) \rightarrow x(m)$  = {1, 2, 0.5, 1, 2}   
 $h(n) \rightarrow h(m)$  = {1, 2, 1, -1}



2. Fold  $h(m)$  to  $h(-m)$



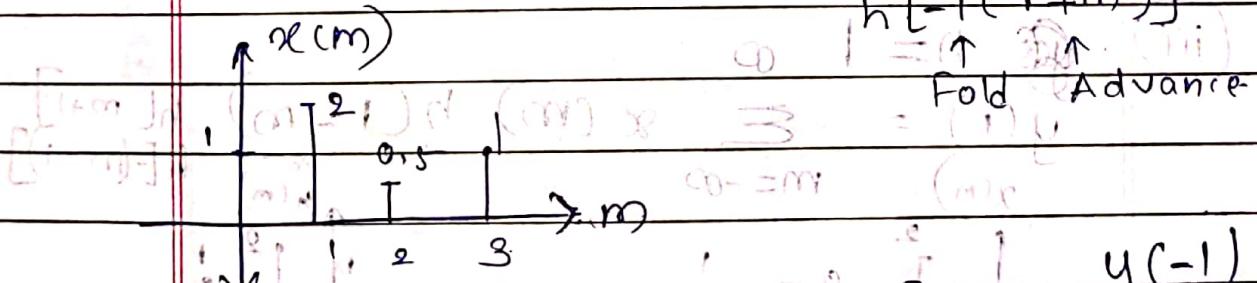
$$\{1, 2, 1, -1\} = (0)p$$

3. ~~Integrate~~

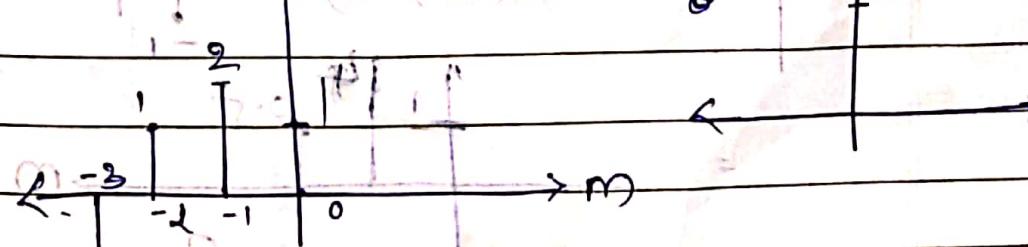
$$y(-1) = \sum_{m=-\infty}^{\infty} x(m) h(-1-m) = (0)p$$

$$y(-1) = \sum_{m=-\infty}^{\infty} x(m) h[-1(-1+m)]$$

$\downarrow$   $l = 1$   $\uparrow$  Advance



$$y(-1) = \sum_{m=-\infty}^{\infty} x(m) h(-1-m) = \{0, 0, 0, 1, 0, 0, 0\}$$



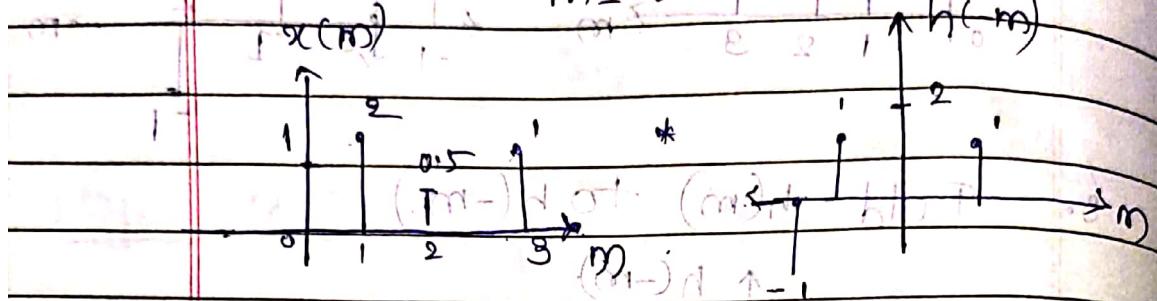
$$y(-1) = \{0, 0, 0, 1, 0, 0, 0\}$$

$$y(-1) = x_1 \cdot h(-1) + x_2 \cdot h(0)$$

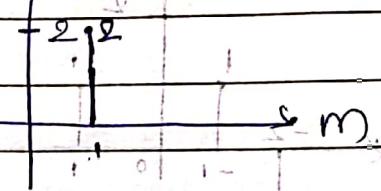
(ii)

$$y(0) = \sum_{m=-\infty}^{\infty} x(m) h(0-m)$$

$$= \sum_{m=-\infty}^{\infty} x(m) \cdot h(-m)$$



$$y(0)$$



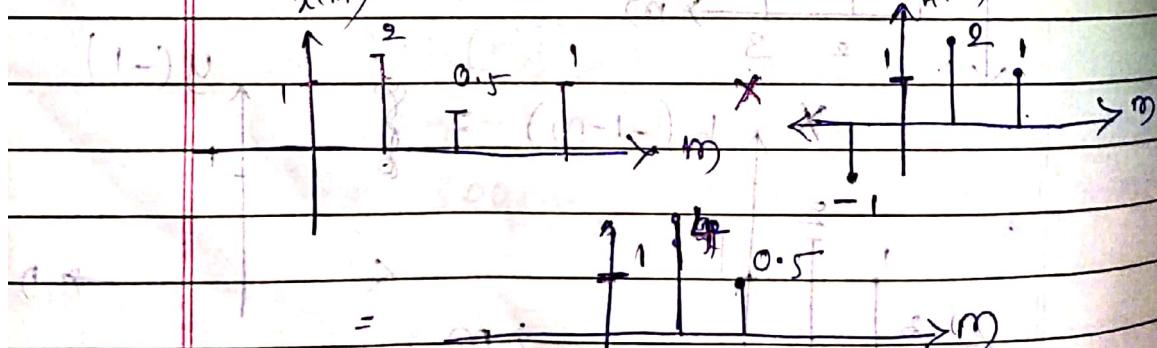
$$y(0) = \{2, 2\}$$

$$y(0) = 2 + 2 = 4$$

(iii)

$$n=1$$

$$y(1) = \sum_{m=-\infty}^{\infty} x(m) h(1-m), h[-m+1]$$



$$y(1) = \{1, 0.5, 0.5\} = -1 + 0.5 + 0.5 = 0.5$$

iv

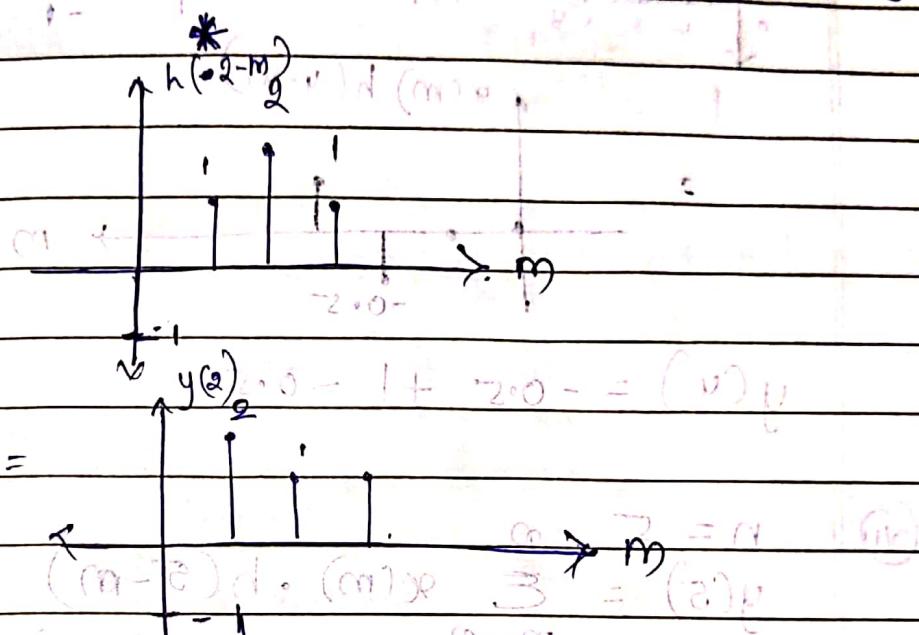
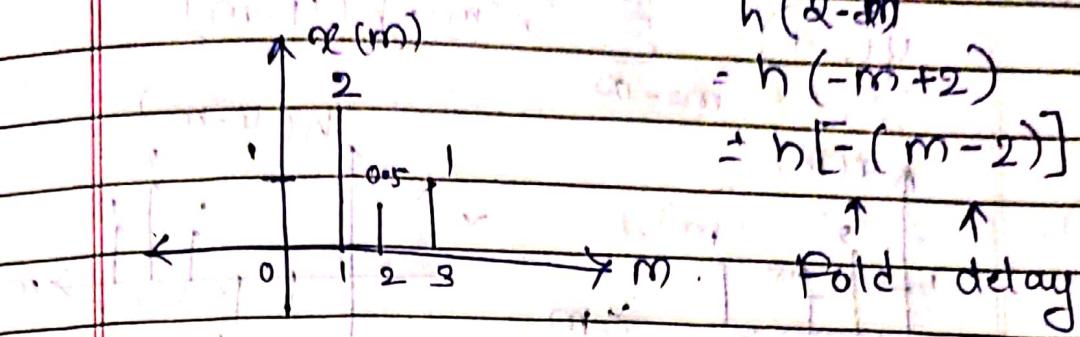
$$n=2$$

$$y(2) = \sum_{m=-\infty}^{\infty} x(m) \cdot h(2-m)$$

$$h(2-m)$$

$$= h(-m+2)$$

$$= h[-(m-2)]$$

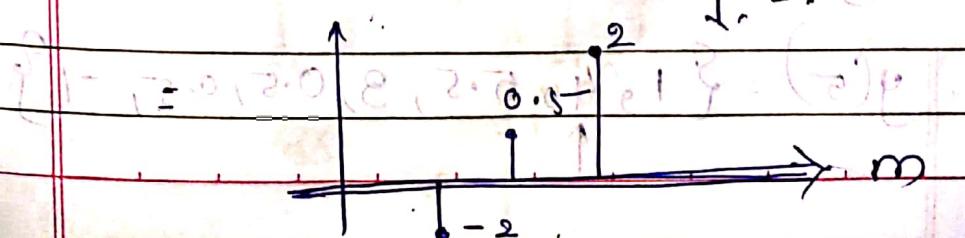
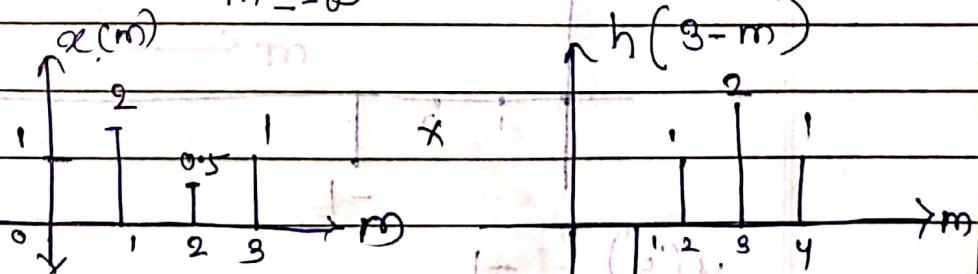


$$\therefore y(2) = -1 + 2 + 0.5 = 1.5$$

v

When  $n=3$

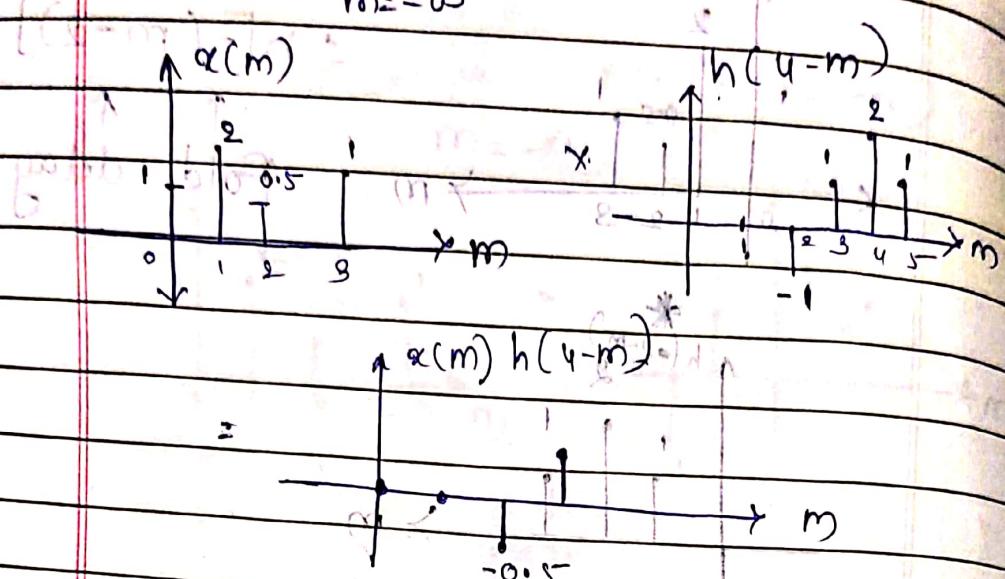
$$y(3) = \sum_{m=-\infty}^{\infty} x(m) \cdot h(3-m)$$



$$y(3) = -2 + 0.5 + 2 = 0.5$$

(ii) When  $n=4$

$$y(4) = \sum_{m=-\infty}^{\infty} x(m) \cdot h(4-m)$$

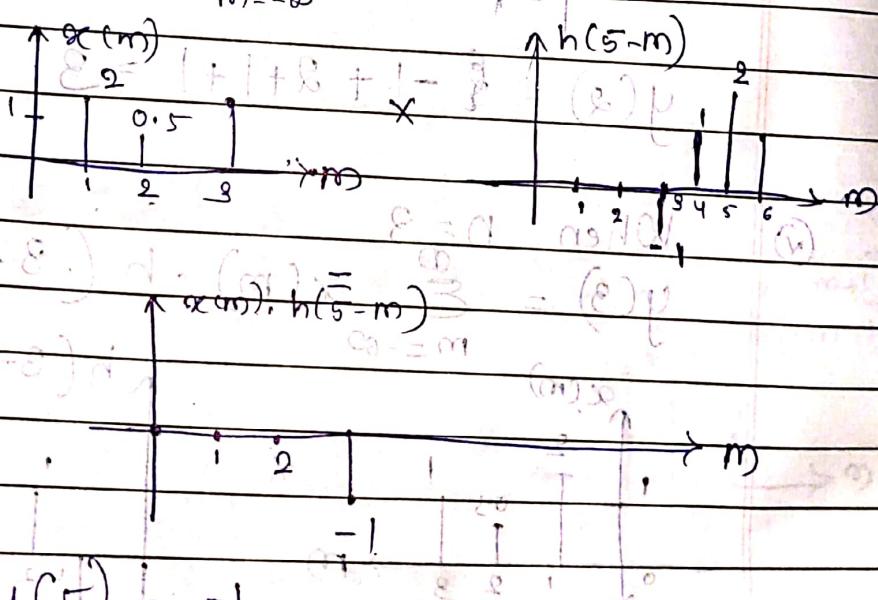


$$y(4) = -0.5 + 1 = 0.5$$

(vi)

$$n=5$$

$$y(5) = \sum_{m=-\infty}^{\infty} x(m) \cdot h(5-m)$$



$$y(5) = -1$$

$$y(5) = \{1, 4, 5.5, 3, 0.5, 0.5, -1\}$$

$$x(n) = \{1, 2, 3, 1\}$$

$$h(n) = \{1, 2, 1, -1\}$$

Replace  $x(n) \rightarrow x(m) = \{1, 2, 3, 1\}$

$$h(n) \rightarrow h(m) = \{1, 2, 1, -1\}$$

When lower index  $[x(n)] = 0$ ,  
 $[h(n)] = -1$

$$0 - 1 = -1$$

O/P sequence starts at -1

$$\text{Len}[x(n)] = 4$$

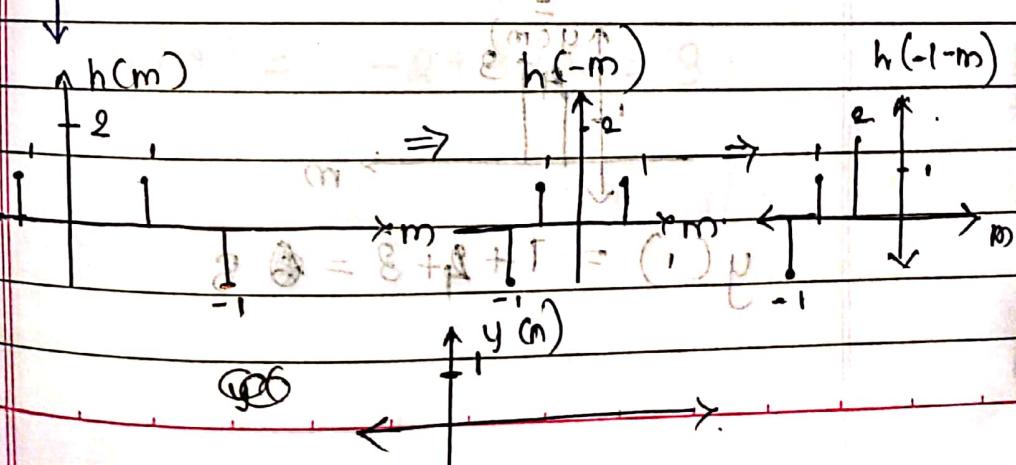
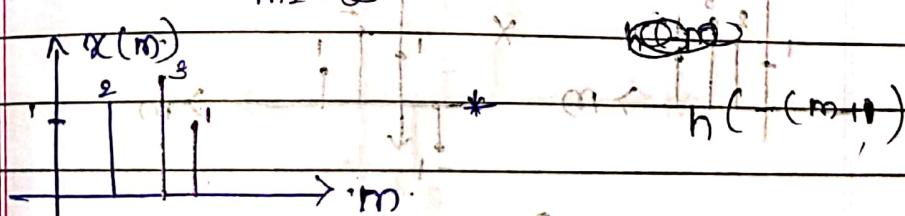
$$\text{Len}[h(n)] = 4$$

$$4 + 4 - 1 = 7$$

O/P sequence = 7 samples

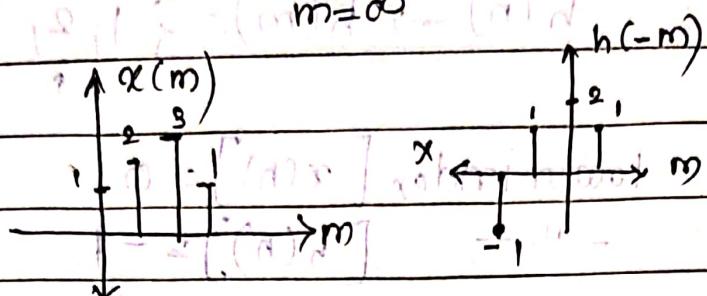
i). When  $n = -1$   $m = -(m) \Rightarrow m = 1$

$$y(-1) = \sum_{m=-\infty}^{\infty} x(n) * h(-1-m)$$



$$\boxed{y(-9) = 1}$$

$$\text{ii) } y(0) = \sum_{m=-\infty}^{\infty} x(m) \cdot h(0-m)$$

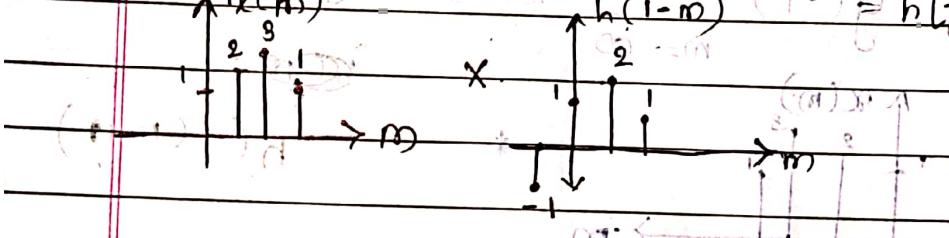


$$I^- = y(m)$$

$$y(0) = 2 + 2 - 4 \neq 1 + 3$$

iii)  $n = 1$

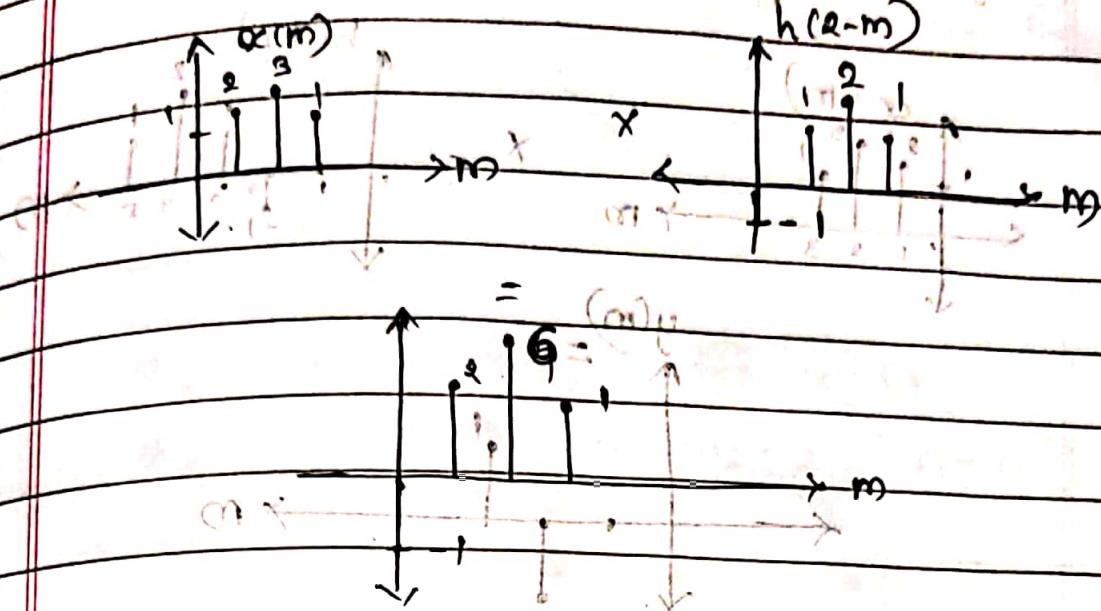
$$y(1) = \sum_{m=-\infty}^{\infty} x(m) \cdot h(1-m)$$



$$y(-1) = 1 + 4 + 3 = 8$$

$$\textcircled{iv} \quad n = 8$$

$$y(8) = \sum_{m=-\infty}^{\infty} x(m) \cdot h(8-m)$$

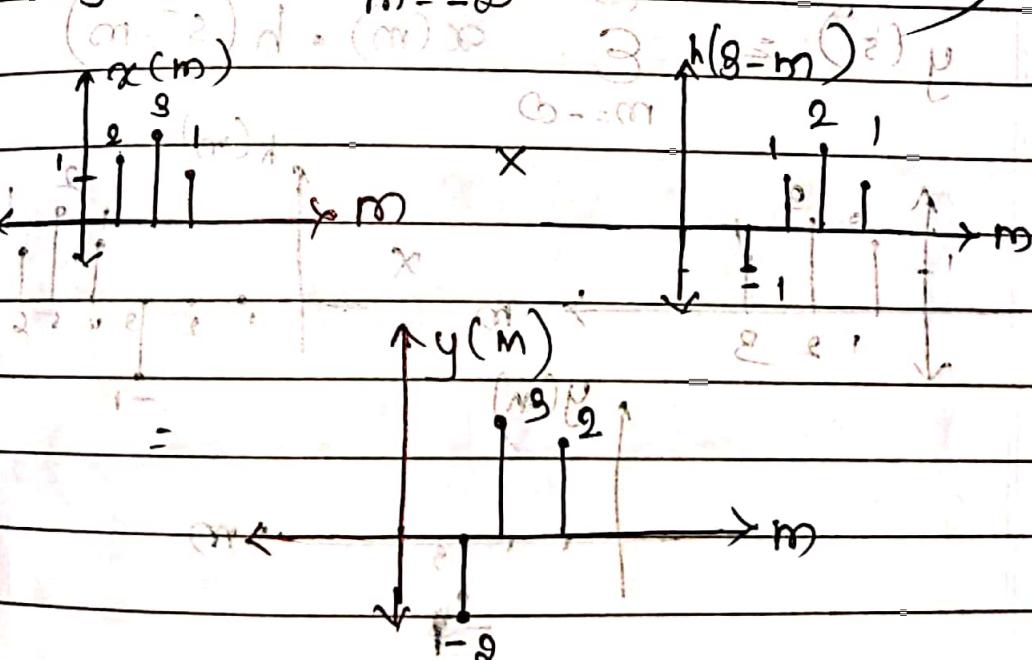


$$y(8) = -1 + 2 + 6 + 1 = 8$$

$$8 = -1 + 8 = 7 \text{ (p)}_u$$

$$\textcircled{v} \quad n = 3$$

$$y(3) = \sum_{m=-\infty}^{\infty} x(m) \cdot h(3-m)$$

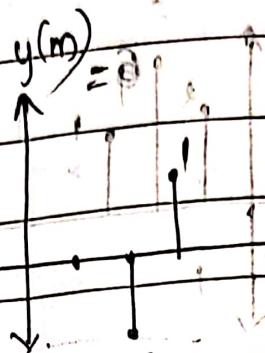
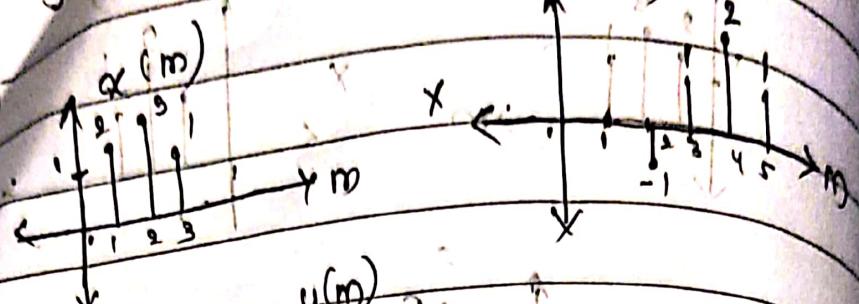


$$y(3) = -2 + 3 + 1 = 2 \text{ (p)}_u$$

$$\{1, 2, 1, 1, 1, 1, 1, 1\} = (n)_u$$

$$\textcircled{v_i} \quad n=4 \quad \alpha(m) \cdot h(4-m)$$

$$y(n) = \sum_{m=-\infty}^{\infty} \alpha(m) \cdot h(4-m)$$

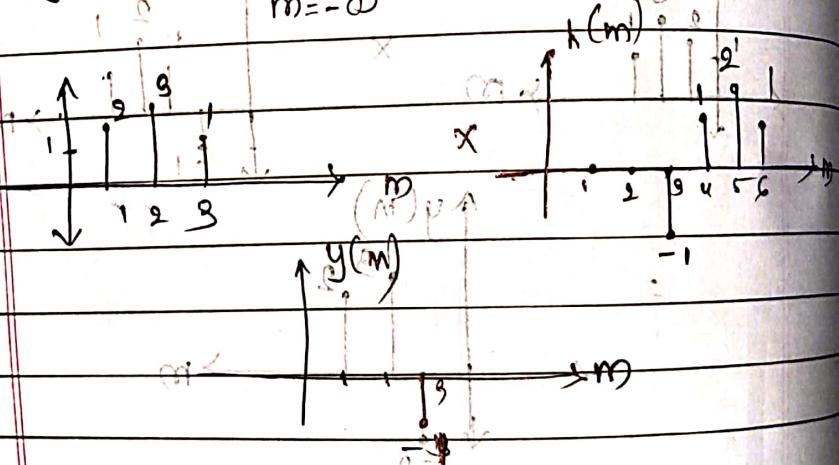


$$y(4) = 1 + 3 + 1 = 5$$

$$y(4) = -3 + 1 = -2$$

$$\textcircled{v_i} \quad (n=8) \quad \alpha(m) \cdot h(8-m) = (8)\mu$$

$$y(8) \underset{m=-\infty}{=} \sum_{m=-\infty}^{\infty} \alpha(m) \cdot h(8-m)$$



$$y(8) = 1 + 2 + 3 + 1 = 7$$

$$y(n) = \{1, 4, 8, 8, 3, -2, -1\}$$

## Circular Convolution

→ Circular Convolution is performed on periodic sequences

Formula:

$$x_g(n) = (x_1(n) \oplus x_2(n))$$

$$(x_g(n)) = \sum_{m=0}^{N-1} x_1(m) \cdot x_2(n-m)$$

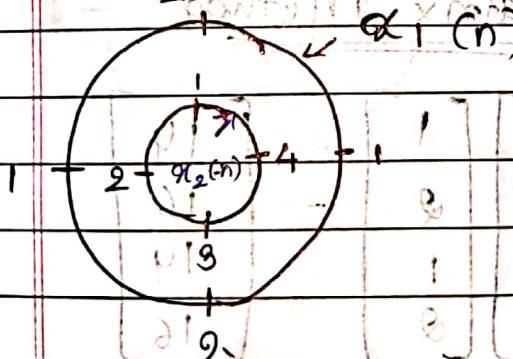
Method 1 : Graphical Method

(Concentric Circle Method)

$$\text{Eg: } x_1(n) = \{1, 2, 1, 2\}$$

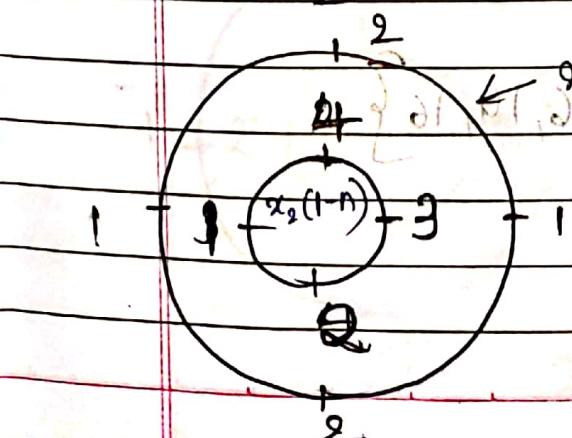
$$x_2(n) = \{4, 3, 2, 1\}$$

$$x_g(n) = x_1(n) \oplus x_2(n)$$



$$x_g(0) = 1 \times 4 + 2 \times 1 + 2 \times 1$$

$$+ 3 \times 2 = 14$$



$$x_g(1) = 1 \times 3 + 4 \times 2 + 1 \times 1$$

$$+ 2 \times 2$$

$$= 16$$

$$\begin{array}{cccc|c} & 1 & 2 & 3 & \\ \text{Row 1} & 1 & 1 & 2 & 1 \\ \text{Row 2} & 1 & 2 & 3 & 2 \\ \text{Row 3} & 1 & 3 & 4 & 3 \\ \text{Row 4} & 1 & 4 & 5 & 4 \end{array}$$

$$\alpha_3(2) = 1 \times 1 + 2 \times 2 + 3 \times 4 = 14$$

$$\begin{array}{cccc|c} & 1 & 2 & 3 & \\ \text{Row 1} & 1 & 1 & 2 & 1 \\ \text{Row 2} & 1 & 2 & 3 & 2 \\ \text{Row 3} & 1 & 3 & 4 & 3 \\ \text{Row 4} & 1 & 4 & 5 & 4 \end{array}$$

$$\alpha_4(2) = 1 \times 1 + 2 \times 2 + 3 \times 4 + 4 \times 2$$

horizontal (row 2) : 14 + 16

(transferring to standard)

$$\begin{array}{cccc|c} & 1 & 2 & 3 & \\ \text{Row 1} & 1 & 1 & 2 & 1 \\ \text{Row 2} & 1 & 2 & 3 & 2 \\ \text{Row 3} & 1 & 3 & 4 & 3 \\ \text{Row 4} & 1 & 4 & 5 & 4 \end{array}$$

$$\alpha_3(2) = \{14, 16, 14, 16\}$$

$$(a)_s p + (a)_r p = (a)_{s+r} p$$

Method 2: Matrix Method

$$\left[ \begin{array}{cccc|c} 1 & 2 & 3 & 4 & 1 \\ 1 & 3 & 4 & 5 & 2 \\ 1 & 4 & 5 & 6 & 3 \\ 1 & 5 & 6 & 7 & 4 \end{array} \right] \xrightarrow{\text{Row 2} - \text{Row 1}, \text{Row 3} - \text{Row 1}, \text{Row 4} - \text{Row 1}} \left[ \begin{array}{cccc|c} 1 & 2 & 3 & 4 & 1 \\ 0 & 1 & 2 & 3 & 1 \\ 0 & 2 & 3 & 4 & 2 \\ 0 & 3 & 4 & 5 & 3 \end{array} \right] \xrightarrow{\text{Row 3} - 2 \times \text{Row 2}, \text{Row 4} - 3 \times \text{Row 2}} \left[ \begin{array}{cccc|c} 1 & 2 & 3 & 4 & 1 \\ 0 & 1 & 2 & 3 & 1 \\ 0 & 0 & 1 & 1 & 1 \\ 0 & 0 & 1 & 2 & 2 \end{array} \right] \xrightarrow{\text{Row 4} - \text{Row 3}} \left[ \begin{array}{cccc|c} 1 & 2 & 3 & 4 & 1 \\ 0 & 1 & 2 & 3 & 1 \\ 0 & 0 & 1 & 1 & 1 \\ 0 & 0 & 0 & 1 & 1 \end{array} \right]$$

$$\alpha_3(n) = \{14, 16, 14, 16\}$$

$$1x1 + 2x2 + 3x3 + 4x4 = 1$$

$$1x1 + 2x2 + 3x3 + 4x4 = 1$$

$$1x1 + 2x2 + 3x3 + 4x4 = 1$$

$$1x1 + 2x2 + 3x3 + 4x4 = 1$$

$$\alpha_1(n) = \{1, -1, 1, -1\}$$

$$\alpha_2(n) = \{1, 2, 3, 4\}$$

$$\alpha_3(n) = \alpha_1(n) \oplus \alpha_2(n)$$

$$\begin{array}{c} \text{Diagram of } \alpha_1(n) \\ \text{Diagram of } \alpha_2(n) \end{array}$$

$$\alpha_3(0) = 1 - 4 + 3 - 2$$

$$= -2$$

$$\begin{array}{c} \text{Diagram of } \alpha_1(n) \\ \text{Diagram of } \alpha_2(n) \end{array}$$

$$\alpha_3(1) = 2 - 1 + 4 - 3$$

$$= 0$$

$$\begin{array}{c} \text{Diagram of } \alpha_1(n) \\ \text{Diagram of } \alpha_2(n) \end{array}$$

$$\alpha_3(2) = -2 + 3 + (-4) + 1$$

$$= -4$$

$$\begin{array}{c} \text{Diagram of } \alpha_1(n) \\ \text{Diagram of } \alpha_2(n) \end{array}$$

$$\alpha_3(3) = 4 - 3 + 2 - 1$$

$$= 2$$

$$\alpha_3(n) = \{-2, 2, -2, 2\}$$

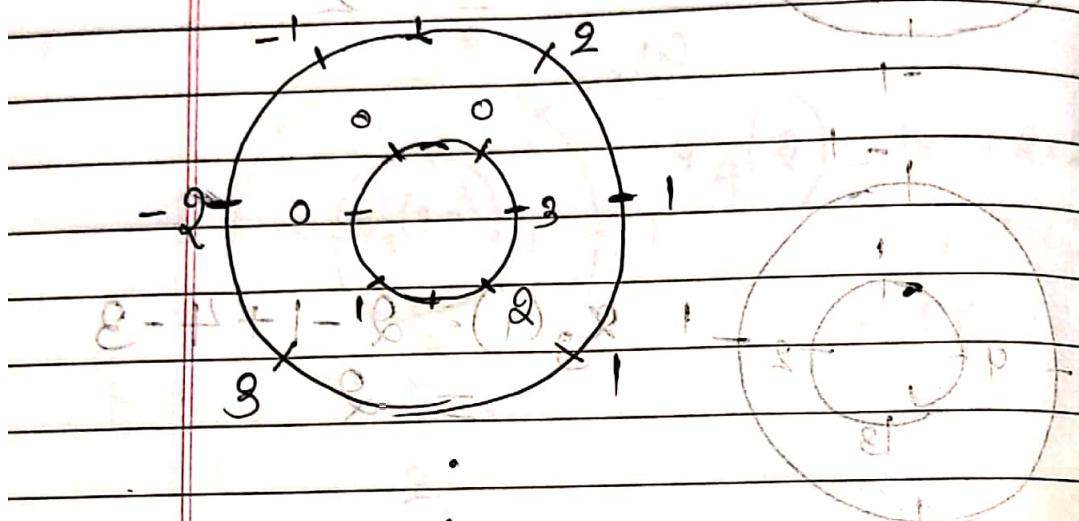
Eg.

$$x_1(n) = \{1, 2, -1, -2, 3, 1\}$$

$$x_2(n) = \{3, 2, 1, 0, 0, 0\}$$

Pad zeros to  $x_2(n)$ :

$$x_2(n) = \{3, 2, 1, 0, 0, 0\}$$



Ans:  $x_3(n) = \{8, 9, 2, -6, 4, 7\}$

\* \* Linear convolution using  
Circular Convolution (Periodic  
Convolution).

Eg:  $x(n) = \{3, -2, 1, 4\}$

$h(n) = \{2, 5, 3\}$

$M=4 \quad N=3$

$\therefore M+N-1 = 4+3-1 = 6$

Add zeros to  $x(n) \& h(n)$

$\therefore x(n) = \{3, -2, 1, 4, 0, 0\}$

$h(n) = \{2, 5, 3, 0, 0, 0\}$

3	0	0	4	1	8	-2	8	8
-2	8	0	0	1	2	4	8	1
1	-2	3	{ 10	0	0	M	9	1
4	1	-2	3	0	0	0	0	-1
0	(4, -1)	11	* 2	(0, 8)	0	(0, 0)	23	
0	0	4	1	-2	3	0	0	12

$$\begin{aligned}
 y(n) &= \{ 3, -2, 1, 4 \} * \{ 2, 5, 3 \} \\
 &= \{ 3, -2, 1, 4, 0, 0, 0 \} + \{ 2, 5, 3, 0, 0, 0 \} \\
 &= \{ 6, 11, 1, 7, 23, 12 \}
 \end{aligned}$$

① Correlation

→ Mathematical operation used to compare 2 signals

→ Used in radar & sonar systems to find location of target by comparing the transmitted & reflected signals

→ Other applications are:

- Image processing

→ Control Engineering

→ Separate Noise from signals

2 Types of Correlation

① Cross - Correlation ② Auto - Correlation

$$R_{xy}(n) = x(n) * h(-n) \quad R_{xx}(n) = x(n) * x(-n)$$

$$R_{xy}(n) = \sum_{k=-\infty}^{\infty} x(k) \cdot y(k-n)$$

$$R_{xx}(n) = \sum_{k=-\infty}^{\infty} x(k) \cdot x(k-n)$$

## (i) Cross Correlation

$$x(n) = \{2, 3, 1, 4\}$$

$$y(n) = \{1, 3, 2, 1\}$$

$$y(-n) = \{1, 2, 3, 1\}$$

$$R_{xy}(n) = x(n) * y(-n)$$

	1	2	3	1
1	2	3	6	2
2	3	6	9	3
3	6	9	12	6
4	2	3	12	4

$$R_{xy}(n) = \{-2, 8+4, 1+6+6, 4+2+9+2\}$$

or both numbers  $8+3+3, 12+1, 4^2$

$$\text{Ans} = \{2, 7, 13, 17, 14, 13, 4\}$$

front of natural half of

left important left rounded zero

(i)

Auto correlation

$$x(n) = \{2, 3, 1, 4\}$$

$$x(n) = \{4, 1, 3, 2\}$$

$$R_{xx}(n) = x(n) * x(-n)$$

	1	2	3	4
1	2	8	2	6
2	3	12	3	9
3	6	18	6	12
4	4	16	4	12

$$R_{xx}(n) = \{8, 14, 13, 30, 13, 14, 8\}$$

$$\text{eg: } \alpha(n) = \{3, 5, 1, 2\}$$

$$h(n) = \{1, 4, 3\}$$

$$\text{Ans: } h(-n) = \{3, 4, 1\}$$

3 4 1

	9	9	12	9	9
5	15	20	15		
1	3	4	1		
2	6	8	2		

$$R_{xy}(n) = \alpha(n) * h(n)$$

$$\{9, 27, 26, 15, 9, 2\}$$

$$\text{eg: } \cancel{\text{How}} \quad \alpha(n) = \{2, 5, -4\}$$

$$\text{Ans: } R_{xx}(n) = \{-8, -10, 45, -10, -8\}$$

$$\{ -8, -10, 45, -10, -8 \}$$

$$\{ -8, -10, 45, -10, -8 \}$$

$$\{ -8, -10, 45, -10, -8 \}$$

$$\{ -8, -10, 45, -10, -8 \}$$

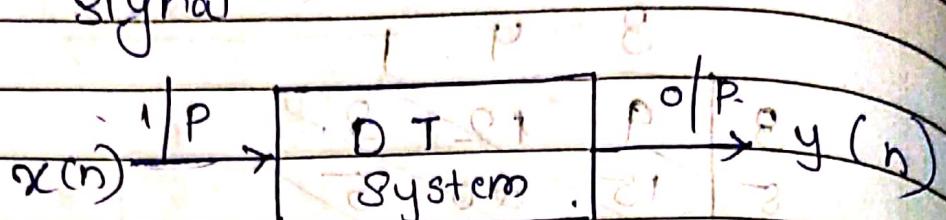
$$\{ -8, -10, 45, -10, -8 \}$$

$$\{ -8, -10, 45, -10, -8 \}$$



## Discrete Time Systems

System acts on input signal & transforms it into an o/p signal



### Classification of DT systems:

1. Static / Dynamic - Systems
2. Causal / Non-causal Systems
3. Linear / non-Linear - "
4. Time invariant / time variant
5. Stable / Unstable

6. Static (Memory less) & Dynamic (Memory) Systems

Static (Memory less)  $\rightarrow$  if o/p depends only on present i/p

Dynamic (Memory)  $\rightarrow$  if o/p depends on past & future i/p's.

Eg: Static Systems

(i)  $y(n) = x(n)$

(ii)  $y(n) = 2x^2(n)$

Eg: Dynamic Systems

(i)  $y(n) = x(2n)$

(ii)  $y(n) = \alpha x(n) + x(n-2)$

(iii)  $y(n) = x(n) + 4y(n-1) + 4y(n-2)$

Eg: Find whether following system is static or dynamic.

(i)  $y(n) = x(n+2)$

O/P depends on future i/Ps

∴ Dynamic System

(ii)  $y(n) = x^2(n)$

O/P depends on present i/P

∴ Static System

(iii)  $y(n) = \alpha x(n-2) + x(n) \quad (i)$

O/P depends on past i/P

∴ Dynamic System

$$(B-1)x + (a)x = (a)p \quad a = 0$$

$$(a)x + (B)x = (a)p \quad B = 1$$

## (\*) Causal & Non Causal Systems

- Causal  $\rightarrow$  O/P of system depends on present & past I/P
- $\rightarrow$  Causal systems are real time systems
- $\rightarrow$  They are physically realizable

$$\text{eg: } y(n) = n x(n) \quad (i)$$

$$(z-a)^{-1} y(n) = x(n-1) + x(n) \quad (ii)$$

Non causal  $\rightarrow$  O/P depends on future I/P.

- $\rightarrow$  O/P is produced even before input is given
- $\rightarrow$  Non-existent in real time
- $\rightarrow$  Not physically realizable

$$\text{eg: } y(n) = x(n) + x(2n) \quad (iii)$$

$$y(n) = 2x^2(n) + x(n+2)$$

$$\text{eg: } (i) y(n) = x(n) + x(n-2) \quad (iii)$$

$$n = -2 \quad y(-2) = x(-2) + x(0)$$

$$n = 0 \quad y(0) = x(0) + x(-2)$$

$$n = +2 \quad y(2) = x(2) + x(0)$$

$\therefore$  O/P depends on present & past  
 $\Rightarrow$  Causal Systems

$$(i) \quad y(n) = \alpha x(n)$$

$$n=0 \quad y(0) = \alpha x(0) \rightarrow \text{Present}$$

$$n=-2 \quad y(-2) = \alpha x(-4) \rightarrow \text{past}$$

$$n=2 \quad y(2) = \alpha x(4) \rightarrow \text{future}$$

$$(1-\alpha)x - (\alpha)x =$$

for positive values of  $n$ , the O/P depends

on future values of I/P

$\Rightarrow$  Non-causal

$$(ii) \quad -x - y(n) = \lim_{k \rightarrow -\infty} [x(k)] \quad \text{Non causal}$$

$\Rightarrow$  Non causal

time  $\rightarrow$  past (a) P prob work

$$(iv) \quad y(n) = x(-n)$$

$$n=-2 \quad \rightarrow -y(0) = x(-2) \Rightarrow x(-2) \leftarrow \text{future}$$

$$n=0 \quad y(0) = x(0) \rightarrow \text{present}$$

$$n=2 \quad y(2) = x(-2) \rightarrow \text{past}$$

$$(1-\alpha)x = (1-\alpha)y$$

For -ve 'n' O/P depends on

future values of I/P

∴ system is noncausal

$$(a)x_a = (a)y$$

$$[(a)x]T = (a)y$$

$$(a)x_a =$$

initialization at  $n=0$  and  $n \rightarrow -\infty$

\* Time invariant & Time variant

System : A system is time invariant if its I/P/O/P characteristics do not change with time.

$$y(n) = x(n) - x(n-1)$$

$$y(n) = T[x(n)]$$

$$= x(n) - x(n-1)$$

Delay i/p by k units & apply to system

$$\therefore y(n, k) = x(n-k) - x(n-k-1)$$

$$\text{from part } \Rightarrow \text{I}$$

Now delay  $y(n)$  by k units in time

$$y(n-k) = x(n-k) - x(n-k-1)$$

$$\text{from } \Rightarrow (a) x = (a) u \quad \text{II} = a$$

$$\text{from } \Rightarrow (a) x = (a) u \quad \text{II} = a$$

$$y(n, k) = y(n-k)$$

$\Rightarrow$  Time Inv. Invariant

2)

$$y(n) = n x(n)$$

$$y(n) = T[x(n)]$$

$$= n x(n)$$

Delay i/p  $x(n)$  by k units in time

$$y(n, k) = n x(n-k) - \text{I}$$

Now, delay o/p  $y(n)$  by k units

$$y(n-k) = (n-k)x(n-k)$$

$$= n \alpha(n-k) - k \alpha(n-k) \quad \text{--- (I)}$$

From (I) & (II)

$$y(n, k) \neq y(n-k) \quad \text{--- (II)}$$

$\therefore$  System is time variant

System is time variant

$$(I) \rightarrow (1-t) \circ f = (1, t) u$$

$$(3) \quad y(n) = x(-n)$$

$$y(n) = T[x(n)] = x(-n)$$

Delay o/p by  $k$  units

$$\text{Delay o/p } y(n, k) = x(-n-k) \quad \text{--- (I)}$$

Delay o/p  $y(n, k)$  by  $k$  units.

$$y(n-k) = x(-n+k) \quad \text{--- (II)}$$

$$y(n, k) \neq y(n+k) u$$

$\therefore$  System is time variant

$$(4) \quad y(n) = x(n) \cos \omega_0 n$$

$$y(n) = T[x(n)] = x(n) \cos \omega_0 n$$

$$= x(n) \cdot \cos \omega_0 n$$

Delay input by  $k$  units

$$\therefore y(n, k) = x(n-k) \cdot \cos \omega_0 n \quad \text{--- (I)}$$

Delay o/p by  $k$  units

$$y(n-k) = x(n-k) \cdot \cos \omega_0(n-k) \quad \text{--- (II)}$$

$$\text{From (I) & (II)} \quad y(n, k) \neq y(n-k)$$

System is time variant

$$(II) \quad \text{From } (I) \& (II) \quad x(t) = y(t-k) + y(t+k)$$

$$(5) \quad y(t) = t \alpha(t) + (a-1)y(t)$$

$$y(t) = T[\alpha(t)]$$

$$\text{Delay: Input by } k \text{ units}$$

Delay i/p by  $k$  units

$$y(t, k) = t \alpha(t-k) \quad (I)$$

$$(a-1)x = (a)p$$

Delay: [O/p by  $k$  units]

$$y(t-k) = (t-k) \alpha(t-k) \quad (II)$$

(I)  $\rightarrow$  System time variant.

$$\text{eflow pd } \alpha(t) \quad q/o \quad \text{Delay}$$

$$(6) \quad (y(t)-e)x = (k-a)v$$

$$(II) \quad (y(t)) = T[\alpha(t)]$$

$$y(t, k) = e^{\alpha(t-k)} \quad (I)$$

$$y(t-k) = e^{\alpha(t-k)} \quad (II)$$

From (I) & (II)  $x = (a)v$

$$y(t, k) = y(t-k) \quad (a)v$$

$$(a-1)x =$$

∴ System is time invariant

$$(I) \quad a_0 \cdot 20 \cdot (k-1)v = (k-1)v$$

$$(II) \quad \text{eflow pd q/o result}$$

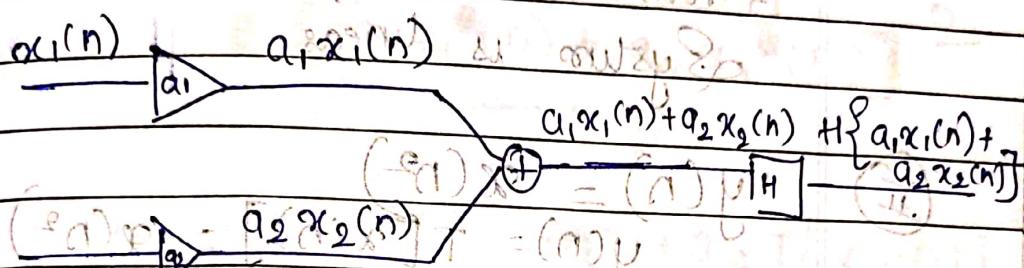
~~Linear & Non-Linear Systems~~

Linear  $\rightarrow$  static for superposition principle

$$(1) \quad y(n) = n x(n)$$

~~Superposition Principle: H~~

$$H\{a_1 x_1(n) + a_2 x_2(n)\} = a_1 H\{x_1(n)\} + a_2 H\{x_2(n)\}$$



$$y(n) = a_1 x_1(n) + a_2 x_2(n)$$

$$(a_1 x_1(n)) + (a_2 x_2(n)) = a_1 H[x_1(n)] + a_2 H[x_2(n)]$$

$$(a_1 x_1(n)) + (a_2 x_2(n)) = a_1 H[x_1(n)] + a_2 H[x_2(n)]$$

Eg:  $y(n) = n x(n), n$

$$x(n) \rightarrow H \rightarrow y(n) = H\{x(n)\} = n x(n)$$

Consider 2 signals,  $x_1(n)$  &  $x_2(n)$

$$x_1(n) \rightarrow H \rightarrow y_1(n) = H\{x_1(n)\} = n x_1(n)$$

$$x_2(n) \rightarrow H \rightarrow y_2(n) = H\{x_2(n)\} = n x_2(n)$$

$$a_1 H\{x_1(n)\} + a_2 H\{x_2(n)\} = a_1 n x_1(n) + a_2 n x_2(n)$$

$$\text{Now } \alpha_1 x_1(n) + \alpha_2 x_2(n) = H[x_1(n)] + H[x_2(n)]$$

$\alpha_1$

$H$

$$H[x_1(n)] = f(a_1 x_1(n))$$

$$H[x_2(n)] = f(a_2 x_2(n))$$

$$H[x_1(n)] + H[x_2(n)] = H[\alpha_1 x_1(n) + \alpha_2 x_2(n)]$$

$$= H[\alpha_1 x_1(n) + \alpha_2 x_2(n)]$$

$$= H[\alpha_1 n x_1(n) + \alpha_2 n x_2(n)]$$

$$f(a_1 x_1(n)) + f(a_2 x_2(n)) = f(\alpha_1 x_1(n) + \alpha_2 x_2(n)) \quad (1)$$

From I & (1)

System is linear

$$(II) \quad y(n) = x(n^2)$$

$$y(n) = T[x(n)] = T[x(n^2)]$$

Superposition principle,

$$T[a_1 x_1(n) + a_2 x_2(n)] = a_1 T[x_1(n)] + a_2 T[x_2(n)]$$

$$y_1(n) = x_1(n^2)$$

$$y_2(n) = x_2(n^2)$$

$$L \cdot H \cdot S = T[a_1 x_1(n) + a_2 x_2(n)]$$

$$= a_1 T[x_1(n)] + a_2 T[x_2(n)]$$

$$\text{Final } L \cdot H \cdot S = a_1 y_1(n) + a_2 y_2(n) \quad (1)$$

$$(a) \quad R \cdot H \cdot S = a_1 T[x_1(n)] + a_2 T[x_2(n)]$$

$$= a_1 x_1(n^2) + a_2 x_2(n^2)$$

$$R \cdot H \cdot S = a_1 x_1(n^2) + a_2 x_2(n^2) \quad (1)$$

$$L \cdot H \cdot S = R \cdot H \cdot S$$

∴ System is Linear

$$\textcircled{c} \quad y(n) = \alpha^2(n)$$

$$y(n) = T[\alpha(n)] = \alpha^2(n)$$

Superposition Principle,

$$T[a_1x_1(n) + a_2x_2(n)] = a_1T[x_1(n)] + a_2T[x_2(n)]$$

$$\text{L.H.S} = T[a_1x_1(n) + a_2x_2(n)]$$

$$= T[a_1x_1^2(n) + a_2x_2^2(n)]$$

$$\text{R.H.S} = a_1T[x_1(n)] + a_2T[x_2(n)]$$

$$= a_1x_1^2(n) + a_2x_2^2(n)$$

$$\text{L.H.S} \neq \text{R.H.S}$$

$\therefore$  Non Linear

$$\textcircled{d} \quad y(n) = (B)x(n) + C$$

$$y(n) = T[x(n)] = Bx(n) + C$$

Superposition principle

$$T[a_1x_1(n) + a_2x_2(n)] = a_1T[x_1(n)] + a_2T[x_2(n)]$$

$$\text{L.H.S} = T[a_1x_1(n) + a_2x_2(n)]$$

$$= B[a_1x_1(n) + a_2x_2(n)] + C$$

$$R.H.S = a_1 T[x_1(n)] + a_2 T[x_2(n)]$$

$$(a)x(n) = a_1 [Bx_1(n) + c] + a_2 [Bx_2(n) + c]$$

$$= a_1 Bx_1(n) + a_1 c + a_2 Bx_2(n) + a_2 c$$

$$R.H.S \neq L.H.S$$

$+ [(a)x(n)]^T$  System is non-linear

$[a_1 x_1(n)]^T$  &  $[a_2 x_2(n)]^T$

\* Stable & Unstable Systems

$$y(n) = x(n)$$

Now,  $n \rightarrow \infty$  can have infinite value

$$[(a)x(n)]^T + [(a)x(n)]^T = 2 \cdot H.B.$$

(a)  $x(n)$  But,  $(x(n))^\infty$  will produce finite value

∴ Stable System

$$2 \cdot H.B. \neq 2 \cdot H.I.$$

Case 1:  $y(n) = n x(n)$

$$n = \infty (x(\infty)) = (a)p \quad (b)$$

$$+ [(a)x(n)]^T = [\infty x(n)]^T = (a)p$$

∴ stable

∴ Unstable Systems

$$+ [(a)x(n)]^T, \infty = + [(a)x(n)]^T + [(a)x(n)]^T$$

$$\text{Case 2: } y(n) = n \cdot x(n)$$

$$= 0 \cdot x(\infty)$$

$$+ [(a)x(n)]^T = 2 \cdot H.I.$$

$$+ [(a)x(n)]^T = 0$$

∴ stable

$$y(n) = \cos[x(n)]$$

$\cos \theta$  lies between -1 to +1 for any value of  $\theta$ .  
 $\therefore$  o/p  $y(n)$  is bounded  
 $\therefore$  Stable system

$$y(n) = \cos(-n - \theta) \text{ mlt}$$

$\therefore$  finite values  
 $\therefore$   $y(n)$  is stable

$$y(n) = \cos[x(n)]$$

check whether these are

- i) static or dynamic
- ii) linear or nonlinear
- iii) shift invariant or shift variant
- iv) causal (or non-causal)
- v) stable (or unstable)

$$y(n) = \cos[x(n)]$$

i) Static system  $\Rightarrow$  O/P depends on present i/p

ii) Linear & Non Linear

$$y(n) = T[x(n)] = \cos[x(n)]$$

Superposition principle

$$T[a_1x_1(n) + a_2x_2(n)] = a_1T[x_1(n)] + a_2T[x_2(n)]$$

$$\text{L.H.S} = T[a_1x_1(n) + a_2x_2(n)]$$

$$= \cos[a_1x_1(n) + a_2x_2(n)]$$

$$\text{R.H.S} = a_1T[x_1(n)] + a_2T[x_2(n)]$$

$$= a_1 \cdot \cos[x_1(n)] + a_2 \cdot \cos[x_2(n)]$$

$$\text{L.H.S} \neq \text{R.H.S}$$

$\therefore$  Nonlinear System

3) Shift Invariant  $\Rightarrow$  Shift variant

$$y(n) = \cos[x(n)]$$

$$= T[x(n)] \neq \cos[x(n)]$$

$$y(n, k) = \cos[x(n-k)] \quad (1)$$

$$y(n-k) = \cos[x(n-k)] \quad (2)$$

$$y(n, k) = T[y(n-k)] \quad (1)$$

$$\Rightarrow \text{Time variant system}$$

4) Causal & Non Causal

$$y(n) = \cos[x(n)]$$

Causal System: o/p depends on present i/p

3) Stable & Unstable

$$y(n) = \cos[x(n)]$$

For bounded value of  $x(n)$ , the cosine function has.

Bounded o/p  $\Rightarrow$  Stable system

Thus the system is Static

Non linear

Time variant

Causal

Unstable

$$y(n) = \sum_{k=-\infty}^{n+1} \alpha(k) u(k)$$

$$y(n) = (\sum_{k=0}^{\infty} \alpha(k)) u(n)$$

i) Static or Dynamic

System's o/p for  $n^{\text{th}}$  Sample

is equal to sum of past i/p,

present & one next sample

$\Rightarrow$  Causal Dynamic

$$0 \leq k \leq n \quad y(n) = \sum_{k=0}^n \alpha(k) u(k)$$

ii) Linear & Non-Linear System

$$y(n) = \sum_{k=-\infty}^{n+1} \alpha(k) u(k)$$

$$= \alpha(-\infty) + \dots + \alpha(-2) + \alpha(-1) + \alpha(0) +$$

$$(1) + \dots + 1 + 1 + \dots + \alpha(n) + \alpha(n+1)$$

$\Rightarrow$  System is Linear since it is  
summation of individual i/p s-

Summation operations is a linear

operation. Hence linear system

is also bounded

iii) Shift variant & Invariant

$y(n)$  is linear sum. of i/p's.

If we shift the i/p's, then there  
will be corresponding shift in o/p

Hence shift invariant

#### iv) Causal & Non Causal

$O/P y(n)$  can be expressed as  $y(n) = x(-\infty) + \dots + x(0) + \dots + x(n) + \dots$

$O/P y(n)$  depends on past i/p like  $x(-\infty)$ , present  $x(n)$  & future  $x(n+1) = \dots$

$\therefore$  Non Causal

Q5) ~~Stable / Unstable~~ has  
Consider the system  $x(k)$  as unit step seq  $u(k)$ .

$$u(k) = \begin{cases} 1 & \text{for } k \geq 0 \\ 0 & \text{otherwise} \end{cases}$$

$\therefore$  System equation is  $y(n)$

$$y(n) = \sum_{k=0}^{n+1} u(k)$$

$$+ (0)x + (-1)x + (2)x + \dots + (n)x$$

$$(1+\alpha)x + (\alpha)x + \dots + 1 + 1 + 1 + \dots + (n+1)$$

i. If  $n \rightarrow \infty$  then  $y(n) \rightarrow \infty$

ii.  $y(n) = \infty$  for all  $n$ .

Thus  $u(k)$  is unit sample seq.

it is bounded. But  $y(n)$  is unbounded as  $n \rightarrow \infty$

Hence Unstable system.

So have 2 marks b. (Ans)

and 2 marks for finding out

diminishes as  $n \rightarrow \infty$ . 3rd 1 mark

as  $x(n) \rightarrow 0$  as  $n \rightarrow \infty$ .

$$y(n) = x(n) \cos(\omega_0 n)$$

i) static / Dynamic

$\Rightarrow$  O/P depends on present i/p  
 $\therefore$  Static System

ii) Linear / Nonlinear

Superposition principle.

$$\begin{aligned} L.H.S &= T[a_1 x_1(n) + a_2 x_2(n)] \\ &= [a_1 x_1(n) + a_2 x_2(n)] \cos(\omega_0 n) \\ &= a_1 x_1(n) \cos(\omega_0 n) + a_2 x_2(n) \cdot \cos(\omega_0 n) \end{aligned}$$

$$\text{R.H.S} = a_1 T[x_1(n)] + a_2 T[x_2(n)]$$

$$= a_1 x_1(n) \cos(\omega_0 n) + a_2 x_2(n) \cos(\omega_0 n)$$

$$L.H.S = R.H.S$$

Linear System

iii) Time variant & Time invariant

$$y(n) = x(n) \cos(\omega_0 n)$$

$$y(n-k) = x(n-k) \cos(\omega_0 n - \omega_0 k) \quad \textcircled{1}$$

$$y(n-k) = x(n-k) \cos(\omega_0(n-k)) \quad \textcircled{2}$$

$$\textcircled{1} \neq \textcircled{2}$$

$\Rightarrow$  Non linear Time invariant

(N)

Causal & Non causal

$$y(n) = x(n) \cos(\omega_0 n)$$

$\Rightarrow$  O/P depends on present i/p  
 $\therefore$  Causal.

iv) Stable / Unstable

$\cos(\omega_0 n) \Rightarrow$  Bounded &  $x(n) \Rightarrow$  Bounded

$\Rightarrow$  Stable

## Module 8:



### Discrete Fourier Transform

$$\cos \pi = -1$$

$$\cos \pi = 0 \quad \sin \pi = 0$$

$$\cos \pi/2 = 0$$

$$\sin \pi/2 = 1$$

$$\cos 3\pi/2 = 0$$

$$\sin 3\pi/2 = -1$$

3 Types of DFT

→ 2 point

→ 4 point

→ 8 point

$x(n) = 1/P$  Sequence

$X(k) = 0/P$  sequence = im DFT

$$X(k) = \sum_{n=0}^{N-1} x(n) e^{-j \frac{2\pi}{N} nk}$$

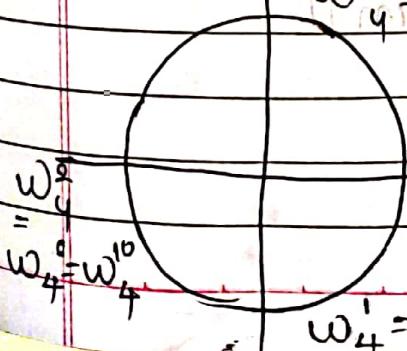
$$X(k) = \sum_{n=0}^{N-1} x(n) \cdot \omega_N^{kn}$$

$$= \begin{bmatrix} \omega_N^{k \cdot 0} & x(0) \\ \omega_N^{k \cdot 1} & x(1) \\ \vdots & \vdots \\ \omega_N^{k \cdot N-1} & x(N-1) \end{bmatrix}$$

How to construct this twiddle factor

Matrix 4

$$\omega_4^3 = \omega_4^4 = \omega_4^8$$



$$\omega_4^0 = \omega_4^4 = \omega_4^8$$

$$\omega_4^1 = \omega_4^5 = \omega_4^9$$

$$N=4 \text{ (number of samples)}$$

i.e.  $n = 0 \text{ to } 3$

$K=0 \text{ to } 3$

	$n=0$	$n=1$	$n=2$	$n=3$
$k=0$	$\omega_4^0$	$\omega_4^1$	$\omega_4^2$	$\omega_4^3$
$k=1$	$\omega_4^0$	$\omega_4^1$	$\omega_4^2$	$\omega_4^3$
$k=2$	$\omega_4^0$	$\omega_4^1$	$\omega_4^2$	$\omega_4^3$
$k=3$	$\omega_4^0$	$\omega_4^1$	$\omega_4^2$	$\omega_4^3$

$$\omega_4^0 = e^{-j\frac{2\pi n k}{N}}$$

$$= e^{-j\frac{2\pi(0)}{4}}$$

$$= e^{j0} = 1 = (1)x$$

$$\omega_4^1 = (1)x$$

$$\omega_4^1 = e^{-j\frac{2\pi \cdot 1 \cdot k}{4}}$$

$$= e^{-j\frac{2\pi}{4}}$$

$$= e^{-j\frac{\pi}{2}}$$

$$\omega_4^1 = \cos \frac{\pi}{2} - j \sin \frac{\pi}{2}$$

$$= 0 - j = -j$$

$$\omega_4^2 = e^{-j\frac{2\pi \cdot 2 \cdot k}{4}}$$

similarly  $= e^{-j\frac{2\pi \cdot 2}{4}} = e^{-j\pi}$  or  $\omega_4^2 = -1$

$$= e^{-j\pi} = -1$$

$$\therefore \cos \pi = -j \sin \pi$$

$$= -1$$

$$P_{01} + P_{02} + P_{03}$$

$$w_4^8 = e^{-j \frac{8\pi}{N} nk}$$

$$= e^{-j \frac{8\pi}{N} \frac{3}{4}}$$

$$= e^{-j \frac{3\pi}{2}}$$

$$w_4^8 = \cos \frac{8\pi}{N} - j \sin \frac{8\pi}{N}$$

$$w_4^{eig} = \begin{vmatrix} 1 & 1 & 1 & 1 \\ 1 & -j & +j & j \\ 1 & -1 & 1 & -1 \\ 1 & j & -1 & -j \end{vmatrix} = \begin{vmatrix} 1 & 1 & 1 & 1 \\ 1 & -j & +j & j \\ 1 & -1 & 1 & -1 \\ 1 & j & -1 & -j \end{vmatrix}$$

$$j^2 - 1 = -1 - j^2 + 1 = 1$$

$$\begin{vmatrix} 1 & 1 & 1 & 1 \\ 1 & -j & +j & j \\ 1 & -1 & 1 & -1 \\ 1 & j & -1 & -j \end{vmatrix} = \begin{vmatrix} 1 & 1 & 1 & 1 \\ 1 & -j & +j & j \\ 1 & -1 & 1 & -1 \\ 1 & j & -1 & -j \end{vmatrix}$$

$$\begin{array}{|c|c|c|c|} \hline & 0 & 1 & 2 & 3 \\ \hline 0 & 0+1+2+3 & 1 & 6 \\ \hline 1 & 0-j-2+3j & 1 & -2+2j \\ \hline 2 & 0-1+2-3j & 1 & -2 \\ \hline 3 & 0+j-2-3j & 1 & -2-2j \\ \hline \end{array}$$

$\Rightarrow$  IDFT  $\rightarrow$  Inverse Discrete Fourier

Transform:

$$(a) x(n) = \frac{1}{N} \sum_{k=0}^{N-1} X(k) \cdot w_N^{-nk}$$

$$(b) X(k) = \frac{1}{N} \sum_{n=0}^{N-1} w_N^{nk} \cdot x(n)$$

$$(c) X \cdot X_N = DFT \text{ Matrix}$$

\* = complex conjugate  
= change sign of  $j$

$$[W_4^*] = \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & j & -1 & -j \\ 1 & -1 & 1 & -1 \\ 1 & -j & -1 & j \end{bmatrix}$$

$$\text{IDF} = x_N = \frac{1}{N} [W_N^*] x_N$$

$$= \frac{1}{4} \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & j & -1 & -j \\ 1 & -1 & 1 & -1 \\ 1 & -j & -1 & j \end{bmatrix} \begin{bmatrix} 1 \\ 1 \\ 1 \\ 1 \end{bmatrix}$$

$$= \frac{1}{4} \begin{bmatrix} 6 - 2 + 2j - 2 - 2j \\ 6 - 2j + 2j^2 + 2 + 2j + 2j^2 \\ 6 + 2 - 2j - 2 + 2 + 2j \\ 6 + 2j - 2j^2 + 2 - 2j - 2j^2 \end{bmatrix}$$

$$\begin{bmatrix} 2 \\ 4 \\ i^2 + 8 - i \\ i^2 - 8 + i \end{bmatrix} = \begin{bmatrix} 0 \\ 1 \\ 2 \\ 8 \end{bmatrix}$$

Eg: ② Find DFT of the Sequence

$$x(n) = \{1, 2, 1, 0\} \leftarrow \text{TDI}$$

$$\begin{array}{l} W_4^* W_4 (W_4^* W_4) \begin{bmatrix} 1 \\ 2 \\ 1 \\ 0 \end{bmatrix} = (1) x(0) \\ W_4^* x W_4^* W_4^* W_4^* \begin{bmatrix} 1 \\ 2 \\ 1 \\ 0 \end{bmatrix} = 2(1) x(1) \\ W_4^* W_4^* W_4^* W_4^* \begin{bmatrix} 1 \\ 2 \\ 1 \\ 0 \end{bmatrix} = x(2) \\ W_4^* W_4^* W_4^* W_4^* \begin{bmatrix} 1 \\ 2 \\ 1 \\ 0 \end{bmatrix} = 0(1) x(3) \end{array}$$

stop now  
if apie serials

$$\begin{bmatrix} 1 & 1 & -j & j \\ 1 & -j & -1 & -j \\ 1 & -1 & 1 & -1 \\ 1 & j & -1 & -j \end{bmatrix} \begin{bmatrix} 1 \\ 2 \\ 1 \\ 0 \end{bmatrix}$$

$$= \begin{bmatrix} 1+2+j+0 \\ 1+2j-1+0 \\ 1-2+1-0 \\ 1+2j-1+0 \end{bmatrix} \begin{bmatrix} 4 \\ -2j \\ 0 \\ 2j \end{bmatrix}$$

$$\text{DFT } \{x(n)\} = \{x(k)\} = \{4, -2j, 0, 2j\}$$

$$\text{IDF}(x(n)) = 1 \cdot \left[ \frac{w}{N} \right]^{nk} x(k)$$

$$\begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & -j & -1 & -j \\ 1 & -1 & 1 & -1 \\ 1 & j & -1 & j \end{bmatrix} \begin{bmatrix} 4 \\ -2j \\ 0 \\ 2j \end{bmatrix}$$

$$\text{Im}[2j] = \text{Im}[2j] =$$

$$1 - 1 = 1 \begin{bmatrix} 4 - 2j + 0 + 2j \\ 4 - 2j^2 - 0 - 2j^2 \\ 4 + 2j + 0 - 2j \\ -4 + 2j^2 + 0 + 2j^2 \end{bmatrix}$$

$$\text{GOI } \sum T(x) = b \cdot \frac{1}{4} \begin{bmatrix} 4 & 1 & 1 & 0 \\ 1 & 8 & 1 & 0 \\ 1 & 1 & 4 & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix}$$

$$\begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & -j & -1 & -j \\ 1 & -1 & 1 & -1 \\ 1 & j & -1 & j \end{bmatrix} \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & -j & -1 & -j \\ 1 & -1 & 1 & -1 \\ 1 & j & -1 & j \end{bmatrix} \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & -j & -1 & -j \\ 1 & -1 & 1 & -1 \\ 1 & j & -1 & j \end{bmatrix}$$

$$[Eg 3] \quad x(n) = \{1, 1, -1, 2, 1, -2\}$$

$$\text{Ans: } x(k) = \{0, -1-j, 6, -1+j\}$$

Q Point: DFT

$n=0$	$n=1$
$x(0)$	$x(1)$

$$N=2$$

$$\begin{bmatrix} i & 0 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} 1 & 1 \\ 1 & -1 \end{bmatrix} = \begin{bmatrix} 0 & 4+2j \\ 0 & 0-1+2j \end{bmatrix}$$

$$W_N^0 = e^{-j\frac{2\pi}{N} \cdot 0} = e^0 = 1$$

$$W_N^1 = e^{-j\frac{2\pi}{N} \cdot 1} = e^{-j\frac{\pi}{2}}$$

$$\begin{bmatrix} 1 & 1 \\ 1 & -1 \end{bmatrix} = e^{-j\frac{\pi}{2}} = \cos \pi - j \sin \pi$$

$$e^{j\pi} + 0 + j(e^{-j\pi} - 1) = -1$$

$$W_2^{kn} = e^{-j\frac{2\pi}{2} \cdot k} = e^{-jk\pi}$$

Eg:  $x(n) = \{1, 1\}$  Find DFT & IDFT

$$x(k) = [W_N^{nk}] x(n)$$

$$= \begin{bmatrix} 1 & 1 \\ 1 & -1 \end{bmatrix} \begin{bmatrix} 1 \\ 1 \end{bmatrix} = \begin{bmatrix} 1+1 \\ 1-1 \end{bmatrix} = \begin{bmatrix} 2 \\ 0 \end{bmatrix}$$

∴ 2 point DFT is

$$x(k) = \{2, 0\}$$

$$\text{IDFT: } x(k) = \frac{1}{N} \sum_{n=0}^{N-1} w_N^{kn} \cdot x(n)$$

$$i = 1 - 2 \quad i = 1 - 1 \quad i = 0$$

$$= \frac{1}{2} [2 + 0] - \frac{1}{2} [2] = [1]$$

$$j = 1 - 0 \quad j = 1 - 1 \quad j = 0$$

$$\alpha = \text{for off-Pole} - \pi/2 - \text{for pole} - \pi/3 - \text{for pole} - \pi/4 - \text{for pole} - \pi/6$$

$$\sin \alpha = 0 - 1 - 0 - 1 - \sqrt{3}/2 - 1/\sqrt{2} - 1/\sqrt{2}$$

$$\cos \alpha = 1 - 0 - 1 - \text{for pole} - 1/\sqrt{2} - \text{for pole} - 1/\sqrt{2} - 1/\sqrt{3}/2$$

$$W_8^6 = W_8^{14} \quad W_8^9 = W_8^{15}$$

$$\{0, 0, 0, 0, 1, 2, 1, 2, 1\} = (\alpha) x$$

$$W_8^9 = W_8^{12} \quad W_8^0 = W_8^8$$

$$W_8^0 = W_8^4 \quad W_8^8 = W_8^4$$

$$W_8^2 = W_8^6 \quad W_8^2 = W_8^{10}$$

$$W_N^{Kn} = \begin{bmatrix} 1 & 1 & 1 & 1 & 1 & 1 & 1 \\ 1 & 0.707 - j0.707 & -j & -0.707 - j0.707 \\ 1 & -j & -1 & j \\ -0.707 + j0.707 & 1 & j & 0.707 - j0.707 \\ 1 & -1 & 1 & -1 \\ -0.707 + j0.707 & -j & 0.707 + j0.707 & -j \\ 1 & 0.707 + j0.707 & j & -j & 0.707 + j0.707 \\ 1 & 0 & 0 & 0 & 0 & 0 & 0 \end{bmatrix}$$

$$W_B^{Kn} = \begin{bmatrix} 1 & 1 & 1 & 1 & 1 & 1 & 1 \\ 1 & 0.707 - j0.707 & -j & -0.707 - j0.707 & -1 & -0.707 + j0.707 & j \\ 1 & -j & -1 & j & 1 & -j & -j \\ -0.707 - j0.707 & j & 0.707 - j0.707 & -1 & 0.707 + j0.707 & j & -j \\ 1 & -1 & 1 & -1 & 1 & -1 & 1 \\ -0.707 + j0.707 & -j & 0.707 + j0.707 & -1 & -0.707 - j0.707 & j & -j \\ 1 & j & -j & j & 1 & j & -1 \\ 0.707 + j0.707 & j & -0.707 + j0.707 & -1 & -0.707 - j0.707 & j & -j \end{bmatrix}$$

Eg:  $\alpha(n) = \{1, 2, 1, 2, 1, 0, 0, 0, 0\}$

$$X(k) = \{6, 1 - j2.414, 0, 1 - j1.828, -2, 1 + j1.828, 0, 1 + j9.828\}$$

$$\omega = 8\pi$$

$$x(n) = e^{-jn\omega} \sum_{k=0}^n \alpha_k e^{jk\omega}$$