## State-feedback control systems

University of Ulsan Prof. KIM

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## 6 Sliding mode control

- Sliding mode control (SMC) is a nonlinear control method to regulate the behavior of dynamic systems subject to disturbances or uncertainties.
- The core idea behind SMC is to create a sliding surface in the state space of the system, such that

"Once the system reaches this sliding surface, it ideally remains there regardless of disturbances or uncertainties."

- In other words, this control law is designed such that it drives the system toward the sliding surface and then maintains it there.
- One of prominent features is that this control involves the use of a discontinuous control signal which switches based on the system's position relative to the sliding surface.
- Sliding mode control has applications in various fields including aerospace, automotive, robotics, and power systems.

Let us consider the following linear systems with uncertainties and external disturbances:

$$\begin{cases} \dot{x}(t) = (A + \Delta A)x(t) + Bu(t) + N_1 w(t). \\ z(t) = Cx(t) + N_2 w(t) \end{cases}$$
(44)

Assumption: In general, the matrix B is of full column rank.

In this study, we employ the following switching function s(t):

$$s(t) = Gx(t) - \int_0^t G(A + BK)x(\tau)d\tau \tag{45}$$

where  $G \in \mathbb{R}^{m \times n}$  is chosen so that  $GB \in \mathbb{R}^{m \times m}$  is invertible and K denote the linear control gain to be designed later.

From (45), it follows that

$$\dot{s}(t) = G\dot{x}(t) - G(A + BK)x(t). \tag{46}$$

Additionally, according to (44), it is also available that

$$\dot{s}(t) = GBu(t) + G\Delta Ax(t) + GN_1w(t) - GBKx(t). \tag{47}$$

**Theorem 5.** The trajectories of system (44) can be driven onto the surface s(t) = 0 if the control input is given as

$$u(t) = Kx(t) + (GB)^{-1}\psi(t)$$
(48)

where

$$\psi(t) = -\operatorname{sgn}(s(t)) \cdot (\alpha + \epsilon(x(t)))$$

$$\epsilon(x(t)) = \epsilon_A \cdot ||G|| \cdot ||x(t)|| + \epsilon_w \cdot ||GN_1||.$$
(49)

**Remark 2.** The decision variables K and G will be designed from the stability condition of closed-loop system given that  $\dot{s}(t) = 0$ .

**Proof:** Consider the Lyapunov function:

$$V_s(t) = \frac{1}{2}s^T(t)s(t).$$

Then, we have

$$\dot{V}_s(t) = s^T(t)\dot{s}(t)$$

$$= s^T(t)\Big(GBu(t) + G\Delta Ax(t) + GN_1w(t) - GBKx(t)\Big).$$

Using (48), it is given that

$$\dot{V}_{s}(t) = s^{T}(t) \Big( \psi(t) + G\Delta Ax(t) + GN_{1}w(t) \Big)$$

$$\leq s^{T}(t) \psi(t) + ||s(t)|| \cdot ||G\Delta Ax(t)|| + ||s(t)|| \cdot ||GN_{1}w(t)||.$$

Noting that

$$||G\Delta Ax(t)|| + ||GN_1w(t)||$$

$$\leq ||G|| \cdot ||\Delta A|| \cdot ||x(t)|| + ||GN_1|| \cdot ||w(t)||$$

$$\leq \underline{\epsilon_A \cdot ||G|| \cdot ||x(t)|| + \epsilon_w \cdot ||GN_1||}_{=:\epsilon(x(t))}$$

we have

$$\dot{V}_s(t) \le s^T(t)\psi(t) + ||s(t)||\epsilon(x(t)).$$

Finally, using (49) implies

$$\dot{V}_s(t) \le -\alpha ||s(t)|| < 0, \ \forall s(t) \ne 0$$

which means that

$$\lim_{t \to \infty} s(t) \to 0.$$

According to sliding mode control theory, when the sliding motion takes place, it follows that s(t) = 0 and  $\dot{s}(t) = 0$ .

That is, from (47), the equivalent control be given as

$$u_{\rm eq}(t) = Kx(t) - (GB)^{-1}G(\Delta Ax(t) + N_1w(t)).$$

Thus, the sliding motion is described as

$$\dot{x}(t) = Ax(t) + Bu_{eq}(t) + \Delta Ax(t) + N_1 w(t)$$

$$= \left(A + BK\right) x(t) + \left(I - B(GB)^{-1}G\right) \left(\Delta Ax(t) + N_1 w(t)\right)$$

$$= \left(\bar{A} + \bar{G}\Delta A\right) x(t) + \bar{G}N_1 w(t)$$
(50)

where

$$\bar{A} = A + BK, \ \bar{G} = I - B(GB)^{-1}G.$$

**Theorem 6.** System (50) is stable under the condition of s(t) = 0 if there exits matrices  $\bar{P} = \bar{P}^T \in \mathbb{R}^{n \times n}$ ,  $\bar{K} \in \mathbb{R}^{m \times n}$  and scalars  $\epsilon > 0$ ,  $\gamma > 0$ , such that

$$\begin{split} \bar{P} > 0 \\ \begin{bmatrix} -I & C\bar{P} & N_2 & 0 \\ (*) & He\{A\bar{P} + B\bar{K}\} + \epsilon \bar{G}EE^T\bar{G}^T & \bar{G}N_1 & \bar{P}H^T \\ (*) & (*) & (*) & -\gamma^2 I & 0 \\ 0 & (*) & (*) & 0 & -\epsilon I \end{bmatrix} < 0. \end{split}$$

Furthermore, the control gain can be reconstructed as follows:

$$K = \bar{K}\bar{P}^{-1}.$$

**Proof:** Let us choose

$$V(t) = x^{T}(t)Px(t).$$

Then we have

$$\dot{V}(t) + z^{T}(t)z(t) - \gamma^{2}w^{T}(t)w(t) = \begin{bmatrix} x(t) \\ w(t) \end{bmatrix}^{T} \Psi \begin{bmatrix} x(t) \\ w(t) \end{bmatrix}$$

where

By the Schur complement, the robust  $\mathcal{H}_{\infty}$  stability condition  $\Psi < 0$  is transformed into

$$\begin{bmatrix} -I & C & N_2 \\ (*) & \text{He}\{P\bar{A} + P\bar{G}\Delta A\} & P\bar{G}N_1 \\ (*) & (*) & -\gamma^2 I \end{bmatrix} < 0.$$
 (51)

Furthermore, pre- and post-multiplying (51) by  $\operatorname{diag}(I, \bar{P} := P^{-1}, I)$  and its transpose yields

$$\begin{bmatrix} -I & C\bar{P} & N_2 \\ (*) & \text{He}\{\bar{A}\bar{P} + \bar{G}\Delta A\bar{P}\} & \bar{G}N_1 \\ (*) & (*) & -\gamma^2 I \end{bmatrix} < 0.$$
 (52)

Noting that

$$\operatorname{He}\{\bar{G}\Delta A\bar{P}\} = \operatorname{He}\{\bar{G}E\Upsilon H\bar{P}\} \le \epsilon\bar{G}EE^T\bar{G}^T + \epsilon^{-1}\bar{P}H^TH\bar{P}$$

and by apply the Schur complement, we can see that the following condition implies (52):

where  $\bar{K} = K\bar{P}$ .