

# ELEC 302 Summary

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## Diodes

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### Ideal Diodes

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Definition: The ideal diode has two regimes:

- Reverse Bias Regime ( $v \leq 0$ ):
  - Zero current flow.
  - Diode behaves like an open circuit.
- Forward Biased Regime ( $v > 0$ ):
  - Zero voltage drop across diode.
  - Diode behaves like a short circuit.

Analysis in DC Circuits:

1. Figure out whether the diode is being forward biased or reverse biased.
2. Replace the diode with either an open or short circuit and proceed as normal.

Peak Inverse Voltage: Max reverse bias voltage. You should always select a diode which can handle 1.5 times your PIV.

Conduction Angle: The conduction angle,  $2\theta$  is the total phase for which the diode is passing current. You can find this at:

$$2\theta = 2 \cos^{-1} \frac{V_S^B}{V_S^A}$$

Where  $V_S^A$  is the amplitude of your alternating voltage source and  $V_S^B$  is the voltage of the source at which the diode becomes forward biased.

### Real Diodes

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Operation Modes: Real diodes have three operation modes:

- Forward Bias ( $V > 0.7V$ )
- Reverse Bias ( $V_{ZK} \leq V \leq 0.7V$ )
- Breakdown ( $V < V_{ZK}$ )

Hence, a real diode generally has a voltage drop of  $0.7V$  across it. Real Diode Model: We can approximate the  $i - v$  characteristics of a real diode by the following expression:

$$i = I_S \left( e^{\frac{v}{nV_T}} - 1 \right)$$

Where:

- $I_S$  is the *Reverse Saturation Current*. Typically very small (pA).
- $n$  is the *Empirical Constant*, generally  $1 \leq n \leq 2$  depending on the type of the diode and its physical structure. *In this course, assume  $n = 1$ .*

- $V_T$  is the *Thermal Voltage*, defined as:

$$V_T = \frac{kT}{q}$$

Where:

- $k = 1.38 \cdot 10^{-23} \text{J/K}$  (Boltzmann Constant)
- $T$ : Absolute temperature (K)
- $q = 1.602 \cdot 10^{-19}$  (Electronic Charge)

At 20°C,  $V_T \approx 25\text{mV}$ .

More notes on the diode model:

- When  $v \ll 0$ , reduces to  $i \approx -I_S$ .

Breakdown and Zener Diodes: Breakdown occurs when  $V < -V_{ZK}$ . Generally,  $V_{ZK} \approx 30\text{V}$ .

- *Zener Diodes* are diodes designed specifically to operate in breakdown regimes.
- This is useful for providing constant voltage in breakdown.

Temperature Dependence: Just assume  $T = 300\text{K}$  in this course.

DC Circuit Analysis – Iterative Method: First, know your equations:

$$V_{DD} = f(I, V) \tag{1}$$

$$I = I_S(e^{V/V_T} - 1) \tag{2}$$

Assume for a “ $x$ -mA diode”, assume:

$$I_S = x \cdot 10^{-3} e^{-0.7/V_T}$$

And thus (1) becomes:

$$I = x \cdot 10^{-3} e^{(V-0.7)/V_T}$$

For the first iteration, assume  $V = 0.7\text{V}$ .

Iterative Steps:

1. Using your guess for  $V$ , calculate  $I$  from (1).
2. Substitute  $I$  into (2) to get a new value for  $V$ .
3. Substitute  $V$  back into (1) to get a value for  $I$ .

Continue iterating until:

$$\frac{(I_n - I_{n-1})}{I_n} \leq 1\%$$

Simpler Models:

- Ideal
- Constant Voltage Drop
- Piecewise-Linear

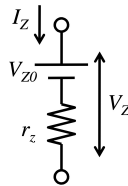
## Power Supplies

Stages of a Power Supply: A power supply consists of three stages:

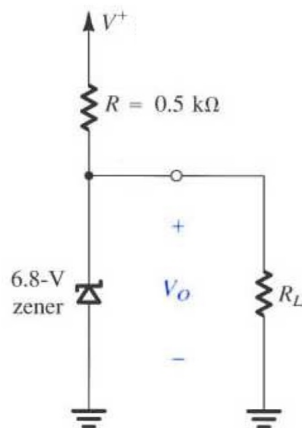
1. Rectifier
2. Filter
3. Regulator

Zener Diode Model: Since a Zener diode past breakdown is nearly linear, we can model it as:

$$V_Z = V_{Z0} + r_Z I_Z$$



Zener Voltage Regulator:



This has the following expressions:

$$V_0 = V_{Z0} \frac{\frac{R}{R_L}}{r_Z + \frac{R}{R_L}} + V^+ \frac{\frac{r_Z}{R_L}}{R + \frac{r_Z}{R_L}}$$

Figures of Merit:

- Line Regulation:

$$\left. \frac{\Delta V_0}{\Delta V^+} \right|_{I_L = \text{constant}} = \frac{r_Z}{R + r_Z}$$

- Load Regulation:

$$\left. \frac{\Delta V_0}{\Delta I_L} \right|_{V^+ = \text{constant}} = -\frac{r_Z}{R}$$

## Rectifiers

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### Types:

- Half-Wave
- Full-Wave
  - Center-Tapped
    - \* Requires only 2 diodes
    - \* Less voltage loss
  - FULL BRIDGE
    - \* Requires fewer turns in the secondary winding of the transformer for the same output voltage.
    - \* Lower PIV.

## Filters

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Purpose is to smooth out the bumpy rectified signal. Also known as a peak detector.

### Ripple Voltage:

$$V_r \cong \frac{V_p T}{2R_L C} = \frac{V_p}{2f R_L C}$$

### Input Voltage:

$$\begin{aligned} V_p \cong V_o &= V_p - \frac{1}{2} V_r \approx V_p \\ &\cong V_p \left( \frac{4f R_L C - 1}{4f R_L C} \right) \end{aligned}$$

### Load Current:

$$I_L \cong \frac{R_L}{V_L} = \frac{V_p - \frac{1}{2} V_r}{R_L} \approx \frac{V_p}{R_L}$$

### Conduction Angle:

$$\omega \Delta t = \sqrt{\frac{2V_r}{V_p}}$$

### Max Diode Current:

$$i_D^{\max} \cong I_L \left( 1 + 2\pi \sqrt{\frac{2V_p}{V_r}} \right)$$

### Average Diode Current:

$$i_D^{\text{ave}} \cong I_L \left( 1 + \pi \sqrt{\frac{2V_p}{V_r}} \right)$$

### Peak Inverse Voltage, 2-Diode Config:

$$PIV = 2V_p$$

### Peak Inverse Voltage, 4-Diode Config:

$$PIV = V_p$$

### Substitutions for Real Case, Constant Voltage Drop:

- Half-Wave

$$V_p \rightarrow V_p - V_D$$

$$PIV = 2V_p - V_D = (V_p - V_D) - (-V_p)$$

- Full-Wave

$$V_p \rightarrow V_p - V_D \quad (2\text{-diode config})$$

$$V_p \rightarrow V_p - 2V_D \quad (4\text{-diode config})$$

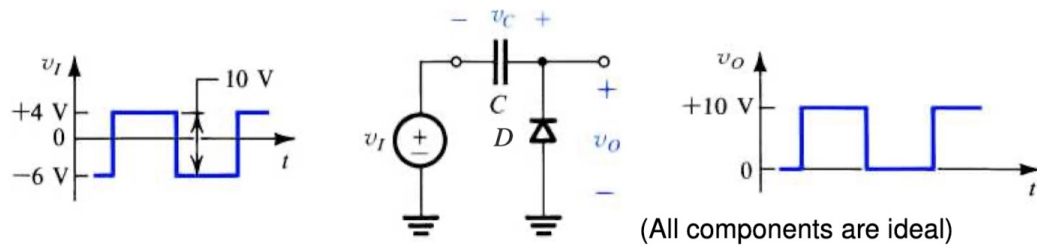
$$PIV \approx 2V_p - V_D \quad (2\text{-diode config})$$

$$PIV \approx V_p - V_D \quad (4\text{-diode config})$$

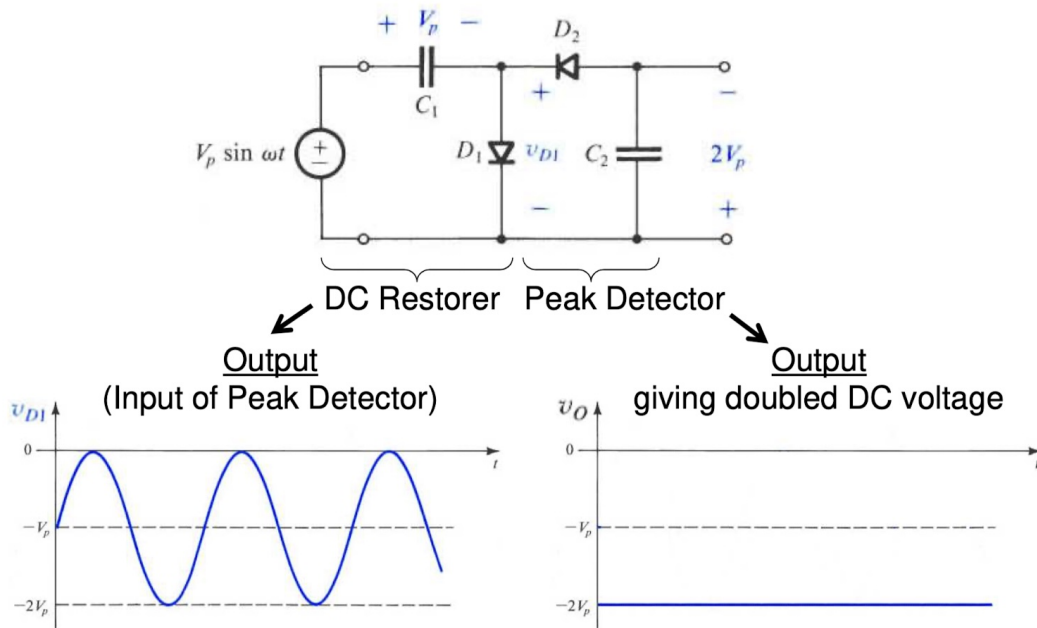
### Other Similar Circuits

#### DC Restorer:

- Reversing the polarity of  $D$  makes the signal negative.



#### Voltage Doubler: Combination of DC restorer and peak detector:



## Limiting/Clipping Circuits

Generally, a clipping circuit should have the following form:

$$v_o = \begin{cases} L_+ & v_i > \frac{L_+}{K} \\ K v_i & \frac{L_-}{K} \leq v_i \leq \frac{L_+}{K} \\ L_- & v_i < \frac{L_-}{K} \end{cases}$$

## Semiconductors

- Two types of charge carriers:
  - Electron (-ve)
  - Hole (+ve)
- Conductivity can be influenced by introducing impurities:
  - Adding electrons: n-type
  - Adding holes: p-type

This then leads to an equilibrium intrinsic carrier concentration:

### Intrinsic Carrier Concentration for Pure Silicon

$$n_i^2 = BT^3 \exp\left(-\frac{E_g}{kT}\right)$$

where:

$E_g = 1.12\text{eV}$ : band gap

$T$ : temperature

$k = 8.62 \times 10^{-5}\text{eV/K}$ : Boltzmann constant

$B = 5.4 \times 10^{31}$

## Conductivity

*Conductivity* relates the current density to the electric field:

### Conductivity

$$\sigma = q(p\mu_p + n\mu_n)$$

Where:

$p$ : Concentration of free holes

$n$ : Concentration of free electrons

$\mu_p$ : Mobility of holes (RT:  $480\text{cm}^2/\text{V}\cdot\text{s}$ )

$\mu_n$ : Mobility of electrons (RT:  $1350\text{cm}^2/\text{V}\cdot\text{s}$ )

$q = 1.6 \times 10^{-19}\text{C}$ : Electron charge

Hence, conductivity depends on the concentration of free charge carriers and the mobility  $\mu$  of the carriers in the material.

Dopants: It is possible to vary  $\sigma$  by “doping” the silicon with impurities, called *dopants*. Doped semiconductors are said to be “extrinsic”.

Donors: Some atoms, depending on their structure, can “donate” one free electron per atom.

- These donors enhance  $\sigma$  by increasing  $n \rightarrow n > p$ .

- This Si is called “n-type”.

Acceptors: Some atoms, depending on their structure, can “accept” one free electron per atom.

- Enhance  $\sigma$  by increasing  $p \rightarrow p > n$ .
- This Si is called “p-type”.

Law of Mass Action: After doping,  $n$  and  $p$  are no longer, equal, but the following relationship must hold:

Law of Mass Action
$n_i^2 = pn$
Note: $n_i$ for silicon is generally $1.45 \times 10^{10} \text{cm}^{-3}$ .

as well as:

Charge Neutrality Relationship
$N_D + p = N_A + n$
$N_D$ : Concentration of donor atoms $N_A$ : Concentration of acceptor atoms

It is also important to discuss majority and minority carriers, where some approximations can be made:

- P-type:

$$(N_D > N_A \wedge N_D \gg n_i) \implies n \approx N_D \implies p \approx \frac{n_i^2}{N_D}$$

- N-type:

$$(N_A > N_D \wedge N_A \gg n_i) \implies p \approx N_A \implies n \approx \frac{n_i^2}{N_A}$$

## Drift Current

The current produced by an electric field is called a “drift current”. In semiconductors, it consists of one due to holes and one due to electrons.

Drift Current Density
$J = \sigma E$

Drift Current
$J_{\text{drift}} = qp\mu_p E + qn\mu_n E$

## Diffusion Current

Like everything with concentration imbalances, diffusion will occur. Hence, we can quantify the current resulting from this diffusion current:

#### Diffusion Current for Holes

$$J_{p,\text{diffn}} = -qD_p \frac{dp}{dx}$$

Where  $D_p = 34\text{cm}^2/\text{s}$  at 300K

#### Diffusion Current for Electrons

$$J_{n,\text{diffn}} = -qD_n \frac{dn}{dx}$$

Where  $D_p = 12\text{cm}^2/\text{s}$  at 300K

And hence, it is also to be noted that both drift and diffusion current can exist in a semiconductor.

#### PN Junctions

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Let's say we have a bar of silicon with a doped p-type side adjacent to a doped n-type side. Hence, holes and electrons will combine to ionize and reach an equilibrium state. This only happens in a specific area called the depletion region. Hence, this region holds charges, giving rise to a built-in voltage. This built-in voltage limits further diffusion. Hence, drift current and diffusion current balance, resulting in an equilibrium condition.

It is to be noted that the  $p$  side and  $n$  side should have equal charge. That is:

$$qN_A x_p A = qN_D x_n A.$$

#### Op-Amps

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We will skip the basics.

#### Differential Amplifiers

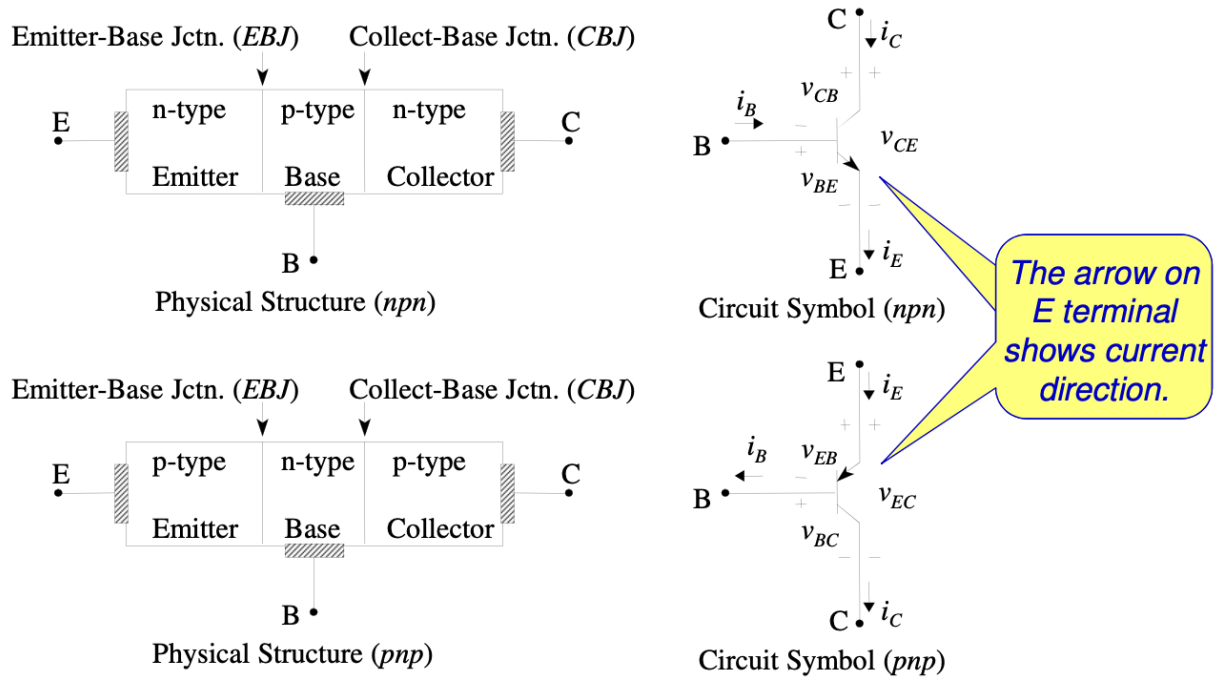
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- Amplifies the difference between signals and rejects the commonality.



## BJTs

BJTs have two varieties, NPN and PNP. Their arrangements are as follows:



Operation Modes:

Mode	Emitter-Base Junction	Collector-Base Junction
Cut-Off (Switching)	Reverse	Reverse
Active (Amplification)	Forward	Reverse
Saturation (Switching)	Forward	Forward