
IpGBT Documentation

Release

IpGBT Design Team

Jul 26, 2019

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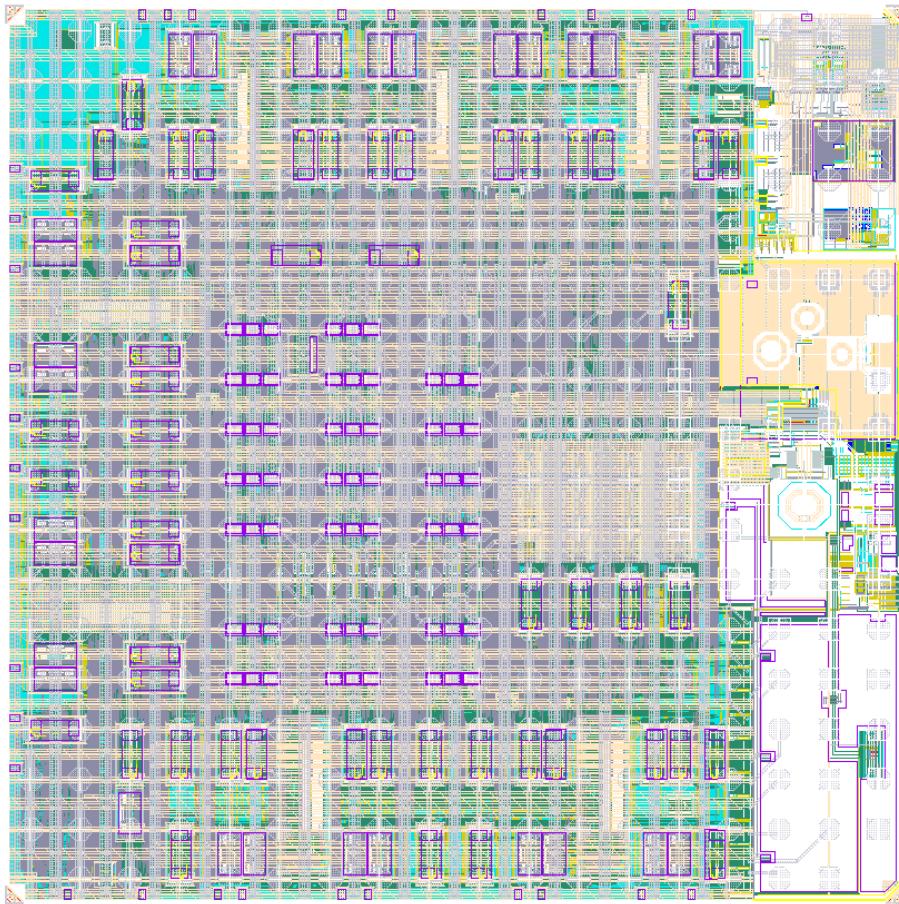
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Note: This is a working document and is therefore neither final nor complete. It is made available to potential IpGBT users to provide early information and to allow them to provide feedback to the IpGBT design and development teams.

Warning: The IpGBT is a **highly flexible device**, it has many setting related to Transceiver modes, locking modes, uplink data rate, FEC coding, clock frequencies, clock phases, number of active eLinks, bit rate of eLinks, phase-aligner modes, pre-emphasis, equalization, driving strengths, and many many more ...

IpGBT will not simply work by just having it installed in the final system! It needs to be configured!

There are **11 configuration pins** to be **hardwired** and more than **320 registers** to be **programmed**.

Quick links:

- IpGBT project on espace (<https://espace.cern.ch/GBT-Project/LpGBT/default.aspx>)

INTRODUCTION

The Low Power Giga Bit Transceiver (IpGBT) is a radiation tolerant ASIC that can be used to implement multipurpose high speed bidirectional optical links for high-energy physics experiments. The ASIC supports 2.56 Gb/s links in the direction from the counting room to the detectors (downlink) and 5.12 or 10.24 Gb/s links in the direction of the detectors to the counting room (uplink), depending on the selected operation mode.

Logically the link provides three “distinct” data paths for Timing and Trigger Control (TTC), Data Acquisition (DAQ) and Slow Control (SC) information. In practice, the three logical paths do not need to be physically separated and are merged on a single optical link as indicated in Fig. 1.1. The aim of such architecture is to allow a single bidirectional link to be used simultaneously for data readout, trigger data, timing, experiment control and monitoring. Such an Architecture establishes a point-to-point bidirectional optical link (two fibres) with constant latency that can function with very high reliability in the harsh radiation environment typical of high energy physics experiments at LHC.

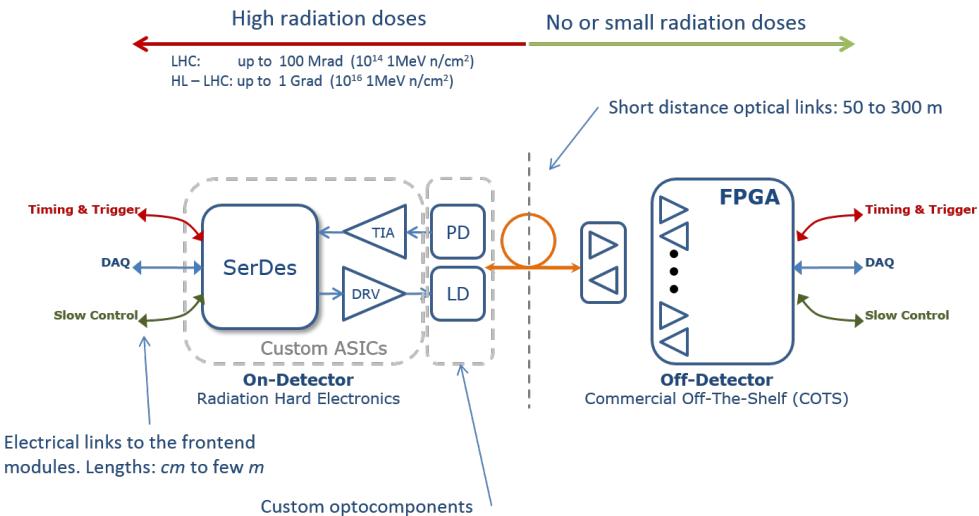


Fig. 1.1: Typical HEP Link Architecture

The development of the proposed link is conceptually divided into two distinct but complementary parts: the IpGBT link chips (IpGBT chipset) and the Versatile Link PLUS (VL+) optoelectronic components. The VL+ project selects and qualifies appropriate fibres and optoelectronic components for use in radiation and develops optical assemblies. The IpGBT develops and qualifies the required radiation hard ASICs (Transceivers, Laser Drivers and PIN Receivers).

The link is implemented by a combination of custom developed and Commercial-Off-The-Shelf (COTS) components (Fig. 1.1). In the counting room the receiver and transmitters are implemented using COTS components and FPGAs. Embedded in the experiments, the receivers and transmitters are implemented by the IpGBT chipset and the VL+ optoelectronic components. This architecture clearly distinguishes between the counting room and front-end electronics because of the very different radiation environments. The on-detector front-end electronics works in a hostile radi-

ation environment requiring custom made components. The counting room components operate in a radiation free environment and can be implemented with COTS components. The use of COTS components in the counting house allows this part of the link to take full advantage of the latest commercial technologies and components (e.g. FPGAs with many link interfaces [[intelWeb](#)] (page 325), [[latticeWeb](#)] (page 325), [[microsemiWeb](#)] (page 325) and [[xilinxWeb](#)] (page 325)) enabling efficient data concentration and data processing from many front-end sources to be implemented in very compact and cost efficient trigger and DAQ interface systems.

The IpGBT ASIC is part of the IpGBT chipset composed of the following chips: a Trans-Impedance Amplifier for the optical receiver (GBTIA/IpGBTIA - under development), a Quad Laser Driver (LDQ10) [[LDQ10_2017](#)] (page 325) and the IpGBT itself, a link ASIC that implements all the needed functions of the data and timing transceiver plus a set of functions needed for experiment control.

The IpGBT is a highly flexible link interface chip with a large number of programmable options to enable its efficient use in a large variety of front-end applications:

- Can be configured to be a bidirectional transceiver, a simplex transmitter or a simplex receiver;
- Several front-end interface modes and options;
- Extensive features for precise timing control;
- Several features for experiment control and monitoring;
- Robust operation against SEUs.

1.1 Radiation Environment

Due to the very high beam luminosity planned for the HL-LHC machine upgrade, the radiation levels for the innermost layers of vertex detectors in the LHC experiments may exceed 1 Grad and 10^{16} 1 MeV n/cm² over the ~10 year lifetime of the experiments. These extremely high levels of radiation pose significant challenges to the electronics and optoelectronics components installed in the detectors, due to Total Ionizing Dose (TID), Non Ionizing Energy Loss (NIEL) radiation and Single-Event Upsets (SEU). TID and NIEL effects are mitigated in the IpGBT chipset since it uses an extensively radiation qualified commercial 65 nm CMOS technology and special layout techniques. SEUs are a major impairment to error free data transmission in HEP applications. The IpGBT uses a particular robust line coding and error correction scheme, capable of correcting single bit and bursts errors caused by SEUs and transmission errors. The IpGBT chip also uses dedicated design methodologies to resolve SEUs in internal logic and registers.

1.2 Architecture and Functionality Overview

The general architecture of the IpGBT ASIC and its main external connections are displayed in Fig. 1.2. In its generic configuration the IpGBT connects to a laser driver ASIC and a trans-impedance amplifier ASIC.

The Clock and Data Recovery and Phase-Locked Loop (CDR/PLL) circuit receives high speed serial data (2.56 Gb/s) from the downlink. From it, recovers and generates an appropriate high speed clock to correctly sample the incoming data stream. The serial data is then de-serialized (DeSER), that is, converted from serial to parallel form and then decoded (DEC), with appropriate error corrections, and finally de-scrambled (DSCR).

In the transmitter part (uplink) the data to be transmitted is scrambled (SCR), to obtain DC balanced signal, and then encoded with a Forward Error Correction (FEC) code before being serialized and sent to the laser driver. The configuration of the laser driver can be performed via an “I2C” connection from the IpGBT.

A clock manager circuit takes care of generating and managing the different high speed and low speed clocks needed in the different parts of the ASIC. A programmable phase-shifter is available to generate 4 external user clocks with programmable frequency (40 MHz to 1.28 GHz) and phase (full 360 deg rotation in 50 ps phase increments). Depending on the Transceiver Mode, an external clock must and/or can be used for operation or during start-up as a reference clock for the serializer or as a locking aid for the CDR circuit.

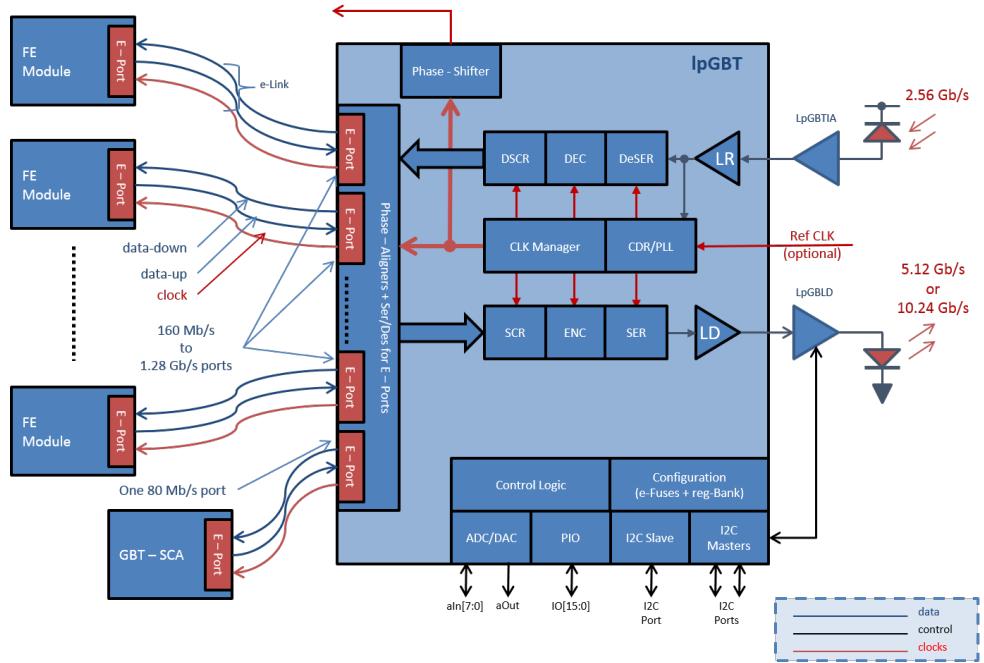


Fig. 1.2: IpGBT architecture

General control and monitoring logic takes care of controlling the different parts of the chip according to the operation mode selected and the ASIC configuration information. Initial configuration information is taken from the on chip e-Fuses that can then be modified via the optical link itself or via an I2C slave interface.

Connections to the front-end modules or ASICs are made through sets of local Input/output Electrical Links (eLinks). Depending on the data rate and transmission media used, eLinks allow connections that can extend up to a few meters. eLinks use the CERN Low Power Signaling (CLPS), with signal amplitudes that are programmable to suit different requirements in terms of transmission distances, bit rate and power consumption (see [Section 18](#) for further details). The eLinks are driven by a series of ePorts on the IpGBT and are associated with eLink ports in the front-end modules. The number of active eLinks and their data rate are programmable (see [Section 7](#) for further details).

Receiving ePorts (ePortRx) are associated with the uplink and de-serialize the data received from the frontend modules or ASICs so that it can be scrambled, coded and assembled in the uplink frame before it is transmitted to the counting room. Each ePortRx has associated a Phase-Aligner (PA) that is used to ensure that the serial data received from the frontend devices is properly sampled in the middle of the eye-diagram. Conversely, transmitter ePorts (ePortTx) are associated with the downlink and serialize the parallel data contained in the downlink frame received from the counting room and transmit it to the front end devices using eLinks. Finally, ePorts also have associated eClocks. These clocks are programmable in frequency but have fixed phase.

The IpGBT main function is that of a data Transceiver (full duplex, simplex Rx or simplex Tx). However it includes as well functionality to aid the implementation of experiment control and monitoring systems. These functions include:

- An I2C master dedicated to control and monitor the laser driver;
- Two generic I2C masters;
- A 16-bit Programmable I/O port (PIO);
- A 10-bit ADC with 8 multiplexed inputs (low analogue bandwidth);
- On chip temperature monitoring;
- Programmable current sources to drive external temperature sensors (PT100 or PT1000), up to 8 multiplexed;

- An 8-bit Voltage DAC (one port).

1.2.1 Down and Up Optical Links

These are the links that connect the IpGBT with the counting room via optical fibers.

The **downlink** transmits data from the counting room to the IpGBT. The data is transmitted at 2.56 Gb/s in a 64-bit frame. The frame uses Forward Error Correction (FEC) coding that is capable of correcting up to 12 consecutive bit errors. From the user's perspective, the frame consists of three fields:

- The **IC-field (IC)** conveys ASIC control information from the counting room to the IpGBT. The bandwidth of the IC channel is 80 Mb/s; (Please note that, alternatively the ASIC operation can be controlled through the I2C-slave interface);
- The **EC-field (EC)** is dedicated to the External Control (EC) ePort with a data rate of 80 Mb/s. This link is designed to be compatible with the GBT-SCA [[GBT-SCA](#)] (page 325) but can be used as general purpose link;
- The **D-field (D)** is dedicated to transmit user's data (D) to the frontend device eLinks. The user available bandwidth is 1.28 Gb/s.

All fields in the downlink frame are protected against transmission errors induced by noise or SEUs by a FEC code.

Two features are available to verify the quality of the data transmission over the downlink:

- **Bit Error Monitoring (BEM):** If enabled, every time the FEC decoder detects an error, an error counter is incremented. This counter can be read either through the IC channel or the I2C port. Although this is not a true Bit Error Rate (BER) count, if read frequently, it can be used to monitor the quality of the data reception (see [Section 14](#));
- **Eye Opening Monitor (EOM):** The EOM circuit allows to "reconstruct" the downlink eye-diagram at the input of the ASIC by a process similar to a "equivalent time" oscilloscope. A scan is made in amplitude and time allowing to estimate the vertical and horizontal opening of the eye. The process is under control of the user through the I2C interface (see [Section 14.4](#));

The **uplink** transmit data from the IpGBT to the counting room. The uplink data rate is programmable and can be either 5.12 Gb/s or 10.24 Gb/s. Additionally, two FEC codes can be used: FEC5 allowing to correct up to 5 consecutive wrong bits and FEC12 with a correction capacity of up to 12 bits. The effective data rate is thus determined by the selected data rate (5.12 or 10.24 Gb/s) and the FEC code used (FEC5 or FEC12) as follows:

- When **operating at 5.12 Gb/s** the IpGBT transmits a 128-bit frame to the counting room that is subdivided (from the user's perspective) in three fields:
 - The **IC-field (IC)** conveys ASIC status information to the counting room. The bandwidth of the IC channel is 80 Mb/s; Data is transmitted over the uplink IC-field as a result a command previously received on the downlink IC-field. (As is the case for the downlink, alternatively the ASIC operation can be controlled through the I2C-slave interface);
 - The **EC-field** carries the data received by EC ePort at 80 Mb/s;
 - The **D-field** carries the data from the data input ePorts (except the EC ePort). The D-field bandwidth depends on the FEC encoding being used, being **3.84 Gb/s** when the FEC12 code is used or **4.48 Gb/s** when the FEC5 is used.
- When **operating at 10.24 Gb/s** the IC and EC fields operation is the same as for operation at 5.12 Gb/s but the bandwidth of the D-field doubles to **7.68 Gb/s** when the FEC12 code is used or **8.96 Gb/s** when the FEC5 is used.

The uplink driver (Line Driver) implements pre-emphasis allowing to minimize (within limits) bandwidth limitations of the transmission line between the IpGBT and the associated line driver (see [Section 5](#)).

1.2.2 Electrical Links (eLinks)

Electrical links (eLinks) interconnect the IpGBT with the front-end electronics (detector modules or ASICs) (see Fig. 1.2). They consist of three differential pairs: two to transmit data from the front-end to the IpGBT (input eLinks) or from the IpGBT to the front-end (output eLinks) and a differential pair to transmit a clock to the front-end. Notice that, because of the asymmetric bandwidth of the up and downlinks, the number of input and output eLinks is not the same. The same is true for the bandwidth of the input/output eLinks. The number of **clock eLinks (eClocks)** available is the same as the number of input eLinks. The bandwidth of the eLinks is programmable and, in the case of the input eLinks, it also depends on the uplink bandwidth (5.12 or 10.24 Gb/s) and on the FEC code selected (FEC5 or FEC12).

Fig. 1.3 summarized the maximum number of output eLinks that can be used depending on the programmed data rate. While "Fig. 1.4" summarizes the maximum number of input eLinks that can be used. Notice that in this case this number depends not only on the programmed data rate but as well on the uplink bandwidth and the FEC code being used. The number of available eClocks is independent of the programmed clock frequency and is 29.

Output eLinks (down-link)			
Bandwidth [Mb/s]	80	160	320
Maximum number	16	8	4

Fig. 1.3: Number of output eLinks versus bandwidth

Input eLinks (up-link)												
Up-link bandwidth [Gb/s]	5.12			10.24								
FEC coding	FEC5		FEC12		FEC5		FEC12					
Bandwidth [Mb/s]	160	320	640	160	320	640	320	640	1280	320	640	1280
Maximum number	28	14	7	24	12	6	28	14	7	24	12	6

Fig. 1.4: Number of input eLinks versus bandwidth

1.2.3 ePorts

eLinks are associated with ePorts and ePorts are grouped in (**eGroups**). Each eGroup, is composed of 4 ePorts. The number of active ePorts (and consequently eLinks) is conditioned by the restrictions indicated in Fig. 1.3 and Fig. 1.4 plus the specificity of the user's configuration. Notice that the data rate of each group can be chosen independently and there is no relationship between the input and output ePort data rates. Many combinations are possible. One example, among the many possible, would be to operate the uplink at 10.24 Gb/s with FEC5; this means that one could have 7 active groups of input eLinks programmed as follows: 3 at 1.28 Gb/s, 4 at 640 Mb/s and 8 at 320 Mb/s.

1.2.4 ePorts Output/Input Phases

The IpGBT uses as a timing reference either an input clock (at the **LHC bunch crossing frequency (f_{LHC})**, when working as a simplex transmitter, or the the downlink data stream, when working as a simplex receiver or transceiver. In all cases, the clocks generated by the IpGBT will keep a fixed phase relationship with the timing reference. It is thus crucial, that either the reference clock or the data stream being fed to the IpGBT will have a well define phase relationship with the **LHC bunch crossing clock (CLK_{LHC})**. If this is the case, the IpGBT is guaranteed, after power on, RST and during operation, to always generate clocks and data outputs with the same phase in relation to the CLK_{LHC}. That is, output eLinks and eClocks will have a fixed phase in relation to the LHC bunch crossing clock. The phase of the eClocks and eLink data outputs is fixed and cannot be changed by the user. However, the IpGBT generates 4 special clock signals (see Section 10) that are phase programmable with a phase resolution of 50 ps. All clocks are frequency programmable with the frequency set being: 40, 80, 160, 320, 640 and 1280 MHz (please note that these are approximate numbers, the true frequencies are: f_{LHC}, 2*f_{LHC}, 4*f_{LHC}, 8*f_{LHC}, 16*f_{LHC} and 32*f_{LHC}).

Since the clocks generated by the IpGBT are, as explained above, phase-locked to CLK_{LHC}, and since the signal phase of the incoming data eLinks (input eLinks) can't be guarantied at system level (e.g. different routing distances,

spread of delays for frontend modules or ASICs) the IpGBT implements a phase aligning mechanism to ensure that the incoming data streams are sampled in the middle of the eye opening (see [Section 7.8](#)). The mode of operation of the circuit implementing this function is programmable ranging from fully automatic to user programmable, details are given in chapter [Section 7](#).

1.2.5 ePort Drivers and Receivers

ePort **Line Drivers (eTx)** and **Line Receivers (eRx)** are used to transmit and receive data over the eLinks and clocks over the eClocks. They use the "**CERN Low Power Signaling (CLPS)**" (see [Section 18](#)). Their main features are:

- Line Drivers (eTx):
 - Programmable output current: 1 to 4 mA;
 - 600 mV common mode voltage;
 - 100 Ohm receiving end termination;
 - Pre-emphasis.
- Line Receivers (eRx):
 - Internal 100 Ohm termination with enable/disable;
 - Auto bias for AC coupling with enable/disable;
 - Line equalization with three settings.

1.2.6 ASIC Control

Control of the IpGBT ASIC is done through a set of registers, a list of which is given in [Section 15](#). To access, read or write, these registers several options are offered to the user. These are however constrained by the operation mode of the ASIC and thus the user must check that the option selected is compatible with the transceiver mode in use. The most generic, that can be used with any of the transceiver modes, is the I2C interface slave. This interface allows to write and read all the registers that control the ASIC operation plus special status registers that are read only, please see [Section 3](#) for full details. Another possibility is to control the IpGBT is the **Internal Control (IC)** field of the down/uplinks. This field allows to read and write the internal registers as does the I2C slave interface but it is only operational when the ASIC operates as a Transceiver. When the ASIC operates as a simplex receiver or transmitter a third option is available. In this case the **External Control (EC)** ePort can be used to control the ASIC in the same way the IC channel is. For details please see [Section 3](#).

1.2.7 Experiment Control and Monitoring

The IpGBT implements several functions to facilitate experiment control and environment monitoring. All of the functions are accessed and controlled through internal registers, either through the I2C-slave, the IC-channel or the EC-port (with this last option being available only when operating the IpGBT as a simplex RX or TX). Please see [Section 12](#) and [Section 13](#) for details.

- Digital interfaces:
 - Three I2C masters;
 - A 16-bit Input/Output port;
 - A reset output pin;
- Environmental monitoring:

- 10-bit ADC to measure quasi-static signals within a range of 0 to 1 V. The ADC is preceded by a multiplexer allowing it to be switched between 8 chip inputs, the internal temperature sensor, and the internal supply power domains;
- On chip temperature measurement with a resolution of 0.5 degree centigrade;
- Environmental stimulus
 - 8-bit voltage DAC with an output range of 0 to 800 mV;
 - Each of the ASIC pins associated with the ADC contain a current generator that allows them to work as current generators. This allows, for example, a PT100 or a PT1000 device to be connected to one of the ADC input pins and the voltage across the device to be measured by the ADC thus allowing to effect an off-chip temperature measurement;

1.2.8 ASIC Package

The IpGBT is encapsulated in a 289-pin fine-pitch (0.5 mm) **Ball Grid Array (BGA)** package. The package dimensions are 9 mm x 9 mm x 1.24 mm. See [Section 17](#) for details.

1.3 Transceiver modes

The IpGBT supports both bidirectional and unidirectional data transmission. This imposes particular constraints on how the link can be configured and initialized at start-up. In all cases the IpGBT will be capable of establishing a working link connection by “itself”. To make sure that this can be accomplished the basic transceiver modes are selected via dedicated configuration pins that must be hardwired for a specific user application according to the specified transceiver mode. If enabled, a default configuration is loaded from the ASIC internal E-Fuse Bank (see [Section 3](#)) at power-up. The final configuration can, after basic link initialization, be modified either through the downlink (IC-Channel), if the ASIC is configured as a transceiver, or through the I2C slave interface. When working as a simplex Transceiver, the IpGBT needs an external clock reference (see **REFCLKP/REFCLKN** differential input). As a simplex receiver or transceiver the IpGBT recovers the clock from the downlink serial data stream, although the external reference can still be used (but not required) as a locking aid.

1.3.1 Simplex transmitter

In this mode the IpGBT works as a simple link transmitter for the uplink receiving the data to be transmitted from the front-end modules through the eLinks (differential signals **EDIN[6:0][3:0]P / EDIN[6:0][3:0]N**). The system reference clock must be driven to the IpGBT and the front-end modules must transmit data to the IpGBT synchronously with the reference clock. This can be achieved by either clocking the front-ends with:

- The system reference clock;
- One (or more) of the frequency programmable eLink clocks (differential signals **ECLK[28:0]P / ECLK[28:0]N**);
- One (or more) of the IpGBT phase/frequency programmable clocks (differential signals **PSCLK[3:0]P / PSCLK[3:0]N**)

Detailed configuration of the IpGBT must be done via the I2C configuration interface or the EC-port (differential signals **EDOUTECP / EDOUTECN** and **EDINECP / EDINECN**), please see [Section 3](#) for further details. In simplex transmitter mode, the downlink receiver functions of the ASIC are clock gated to minimize power consumption and, necessarily, the eLink ports operate as receivers only (exception made for the EC-port if selected as a control means to configure the ASIC).

1.3.2 Simplex receiver

In this mode the IpGBT works as a simple link receiver, receiving data and, implicitly, the clock reference from the counting room through the downlink. The received data are fed to the front-end modules through the eLinks (differential signals **EDOUT[3:0][3:0]P / EDOUT[3:0][3:0]N**). The eLinks data can be received by the frontend modules using one of the following strategies:

- Implementing **Clock and Data Recovery (CDR)** in the frontend;
- Clocking the frontend with one (or more) of the eLink clocks, probably set to the bit rate frequency (full data rate) or half that frequency (double data rate);
- The same as above but using the phase/frequency programmable clocks instead.

Detailed configuration of the IpGBT must be done at the start-up from the E-Fuses and can later be modified via the I2C configuration interface (or the EC-port).

In this mode, the link transmitter functions are clock gated to minimize power consumption and, consequently, the eLink ports operate as transmitters only (exception made for the EC-port if selected as a control means to configure the ASIC).

1.3.3 Transceiver

In this mode the IpGBT works as a full link transceiver with bidirectional data communication with the front-ends and the counting room. The IpGBT delivers the global system clock reference, coming from the counting room, to all front-ends. The detailed configuration (and monitoring) can be performed via the IC-channel or the I2C slave interface.

In this mode, the ePorts operate as transceivers with the ePort receivers feeding data to the serializer and the ePort transmitters receiving data from the CDR circuit.

QUICK START

The lpGBT is a highly flexible device, it has many setting related to transceiver modes, locking modes, uplink data rate, FEC coding, clock frequencies, clock phases, number of active eLinks, bit rate of eLinks, phase-aligner modes, pre-emphasis, equalization, driving strengths, and many many more ...

This chapter by any means is not trying to demonstrate how to use all this features. It is meant to guide the user through the configuration flow, at the same time giving references where more detailed information can be found.

2.1 Configuration pins

lpGBT chip has 13 configuration pins. Connect pins:

1. PORDIS to GND (details: [PORDIS](#) (page 67))
2. SC_I2C to GND (details: [SC_I2C](#) (page 18))
3. ADR3 to GND (details: [ADR3](#), [ADR2](#), [ADR1](#), [ADR0](#) (page 18))
4. ADR2 to GND (details: [ADR3](#), [ADR2](#), [ADR1](#), [ADR0](#) (page 18))
5. ADR1 to GND (details: [ADR3](#), [ADR2](#), [ADR1](#), [ADR0](#) (page 18))
6. ADR0 to GND (details: [ADR3](#), [ADR2](#), [ADR1](#), [ADR0](#) (page 18))
7. VCOBYPASS to GND (details: [VCOBYPASS](#) (page 76))
8. STATEOVRD to GND (details: [STATEOVRD](#) (page 67))

Decide on mode of operation based on the description [MODE3](#), [MODE2](#), [MODE1](#), [MODE0](#) (page 17).

9. Select MODE3 (details: [MODE3](#), [MODE2](#), [MODE1](#), [MODE0](#) (page 17))
10. Select MODE2 (details: [MODE3](#), [MODE2](#), [MODE1](#), [MODE0](#) (page 17))
11. Select MODE3 (details: [MODE3](#), [MODE2](#), [MODE1](#), [MODE0](#) (page 17))
12. Select MODE0 (details: [MODE3](#), [MODE2](#), [MODE1](#), [MODE0](#) (page 17))

Depending on mode of operation you may have to change lock mode for CDR/PLL.

13. Select locking mode LOCKMODE (details: [LOCKMODE](#) (page 75))

For simplicity, we can also start by

14. RSTB to VDD (details: [RSTB](#) (page 67))
15. SLSCL to VDD (details: [I2C slave interface](#) (page 22))
16. SLSDA to VDD (details: [I2C slave interface](#) (page 22))

One should notice, that while operating the real chip, most of the pins (besides MODE and LOCKMODE) can be left unconnected as all configuration pins have pull resistors which set default value. It should be noted, that the pull resistors are not included in the IpGBT model, and thus signals have to be explicitly driven.

Warning: Known issues: [Section 19.1.1](#), [Section 19.1.2](#), [Section 19.1.3](#).

2.2 Configuration registers

In the next step, user has to configure several dozens of registers. The IpGBT has several potential configuration flows, using [Serial control and monitoring interface](#) (page 20), [I2C slave interface](#) (page 22), [E-FUSES](#) (page 23). A detailed sequence is described in [Configuration flows](#) (page 25) chapter. Below we present only an example register-values pairs (standard Verilog convention is used: number in front specifies number of bits and letter specifies encoding format where *b* stands for binary, *d* for decimal and *h* for hexadecimal).

2.2.1 Clock generator registers

32. [\[0x01f\] EPRXLOCKFILTER](#) (page 131)
 - EPRXLockThreshold[3:0] = 4'd5;
 - EPRXReLockThreshold[3:0] = 4'd5
32. [\[0x020\] CLKGConfig0](#) (page 131)
 - CLKGCalibrationEndOfCount[3:0] = 4'd12
 - CLKGBiasGenConfig[3:0] = 4'd8
33. [\[0x021\] CLKGConfig1](#) (page 131)
 - CDRControlOverrideEnable = 1'b0
 - CLKGDisableFrameAlignerLockControl = 1'b0
 - CLKGCDRRes = 1'b1
 - CLKGVcoRailMode = 1'b0
 - CLKGVcoDAC[3:0] = 4'd4
34. [\[0x022\] CLKGPllRes](#) (page 131)
 - CLKGPllResWhenLocked[3:0] = 4'h4
 - CLKGPllRes[3:0] = 4'h4
35. [\[0x023\] CLKGPLLIntCur](#) (page 132)
 - CLKGPLLIntCurWhenLocked[3:0] = 4'h5
 - CLKGPLLIntCur[3:0] = 4'h5
36. [\[0x024\] CLKGPLLPropCur](#) (page 132)
 - CLKGPLLPropCurWhenLocked[3:0] = 4'h5
 - CLKGPLLPropCur[3:0] = 4'h5
37. [\[0x025\] CLKGCDRPropCur](#) (page 132)
 - CLKGCDRPropCurWhenLocked[3:0] = 4'h5

- CLKGCDRPropCur[3:0] = 4'h5
38. [\[0x026\] CLKGCDRIntCur](#) (page 132)
- CLKGCDRIntCurWhenLocked[3:0] = 4'h5
 - CLKGCDRIntCur[3:0] = 4'h5
39. [\[0x027\] CLKGCDRFFPropCur](#) (page 132)
- CLKGCDRFeedForwardPropCurWhenLocked[3:0] = 4'h5
 - CLKGCDRFeedForwardPropCur[3:0] = 4'h5
40. [\[0x028\] CLKGFLLIntCur](#) (page 132)
- CLKGFLLIntCurWhenLocked[3:0] = 4'h0
 - CLKGFLLIntCur[3:0] = 4'hF
41. [\[0x029\] CLKGFFCAP](#) (page 132)
- CDRCOConnectCDR = 1'h0
 - CLKGCapBankOverrideEnable = 1'h0
 - CLKGFeedForwardCapWhenLocked[2:0] = 3'h0
 - CLKGFeedForwardCap[2:0] = 3'h0
42. [\[0x02a\] CLKGCntOverride](#) (page 133)
- CLKGCOoverrideVc = 1'h0
 - CDRCORefClkSel = 1'h0
 - CDRCOEnablePLL = 1'h0
 - CDRCOEnableFD = 1'h0
 - CDRCOEnableCDR = 1'h0
 - CDRCODisDataCounterRef = 1'h0
 - CDRCODisDESvbiasGen = 1'h0
 - CDRCOConnectPLL = 1'h0
43. [\[0x02b\] CLKGOverrideCapBank](#) (page 133)
- CLKGCapBankSelect[7:0] = 8'h0
44. [\[0x02c\] CLKGWaitTime](#) (page 133)
- CLKGwaitCDRTime[3:0] = 4'h8
 - CLKGwaitPLLTime[3:0] = 4'h8
45. [\[0x02d\] CLKGLFConfig0](#) (page 133)
- CLKGLockFilterEnable = 1'h1
 - CLKGCapBankSelect[8] = 1'h0
 - CLKGLockFilterLockThrCounter[3:0] = 4'h9 (to be updated after SEU testing)
46. [\[0x02e\] CLKGLFConfig1](#) (page 134)
- CLKGLockFilterReLockThrCounter[3:0] = 4'h9 (to be updated after SEU testing)
 - CLKGLockFilterUnLockThrCounter[3:0] = 4'h9 (to be updated after SEU testing)

2.2.2 Uplink: ePort Inputs DLL's

52. [\[0x034\] EPRXDllConfig](#) (page 135)
- EPRXDllCurrent [1:0] = 2'h1
 - EPRXDLLConfirmCount [1:0] = 2'h1
 - EPRXDLLFSMClkAlwaysOn = 1'h0
 - EPRXDLLCoarseLockDetection = 1'h0
 - EPRXEnableReInit = 1'h0
 - EPRXDataGatingEnable = 1'h1

2.2.3 Uplink: Line driver settings (if high speed transmitter is used)

57. [\[0x039\] LDConfigH](#) (page 137)
- LDModulationCurrent [6:0] - 7'd32

2.2.4 Uplink: ePort Inputs Group 0 at 1.28 Gbps

Values before assume that the chip works in high speed mode (10 Gbps). One input is enabled : EDIN00.

196. [\[0x0c4\] EPRX0Control](#) (page 208)
- EPRX03Enable - 1'h0
 - EPRX02Enable - 1'h0
 - EPRX01Enable - 1'h0
 - EPRX00Enable - 1'h1
 - EPRX0DataRate [1:0] - 2'h3
 - EPRX0TrackMode [1:0] - 2'h2
204. [\[0x0cc\] EPRX00ChnCntr](#) (page 212)
- EPRX00Term - 1'h1

2.2.5 Uplink: ePort Inputs Group 1 at 640 Mbps

Values before assume that the chip works in high speed mode (10 Gbps). Two inputs are enabled : EDIN10 and EDIN12.

196. [\[0x0c5\] EPRX1Control](#) (page 209)
- EPRX13Enable - 1'h0
 - EPRX12Enable - 1'h1
 - EPRX11Enable - 1'h0
 - EPRX10Enable - 1'h1
 - EPRX0DataRate [1:0] - 2'h2
 - EPRX0TrackMode [1:0] - 2'h2

- 208. [\[0xd0\] EPRX10ChnCntr](#) (page 213)
 - EPRX10Term = 1'h1
- 209. [\[0xd2\] EPRX12ChnCntr](#) (page 214)
 - EPRX12Term = 1'h1

2.2.6 Downlink: Frame aligner settings (if high speed receiver is used)

- 47. [\[0x02f\] FAMaxHeaderFoundCount](#) (page 134)
 - FAMaxHeaderFoundCount [7:0] = 4'hA (Ask Jose)
- 48. [\[0x030\] FAMaxHeaderFoundCountAfterNF](#) (page 134)
 - FAMaxHeaderFoundCountAfterNF [7:0] = 4'hA (Ask Jose)
- 49. [\[0x031\] FAMaxHeaderNotFoundCount](#) (page 134)
 - FAMaxHeaderNotFoundCount [7:0] = 4'hA (Ask Jose)
- 50. [\[0x032\] FAFAMaxSkipCycleCountAfterNF](#) (page 134)
 - FAFAMaxSkipCycleCountAfterNF [7:0] = 4'hA (Ask Jose)

2.2.7 Downlink: ePort Outputs Group 0 at 320Mbps

Enable EDOUT00 (pins EDOUT00P and EDOUT00N) working at 320 Mbps.

- 167. [\[0xa7\] EPTXDataRate](#) (page 189)
 - EPTX0DataRate [1:0] = 2'h3
- 169. [\[0xa9\] EPTX10Enable](#) (page 190)
 - EPTX00Enable = 1'h1
- 172. [\[0xac\] EPTX00ChnCntr](#) (page 191)
 - EPTX00DriveStrength [2:0] = 3'h3

2.2.8 Downlink: ePort Outputs Group 3 at 80Mbps

Enable output group 3 (channels EDOUT30, EDOUT31, EDOUT32 EDOUT33) at 80 Mbps:

- 167. [\[0xa7\] EPTXDataRate](#) (page 189)
 - EPTX3DataRate [1:0] = 2'h1
- 170. [\[0xaa\] EPTX32Enable](#) (page 190)
 - EPTX30Enable = 1'h1
 - EPTX31Enable = 1'h1
 - EPTX32Enable = 1'h1
 - EPTX33Enable = 1'h1
- 184. [\[0xb8\] EPTX30ChnCntr](#) (page 200)
 - EPTX30DriveStrength [2:0] = 3'h3

185. [\[0x0b9\] EPTX31ChnCntr](#) (page 201)
 - EPTX31DriveStrength[2:0] = 3'h3
186. [\[0x0ba\] EPTX32ChnCntr](#) (page 202)
 - EPTX32DriveStrength[2:0] = 3'h3
187. [\[0x0bb\] EPTX33ChnCntr](#) (page 202)
 - EPTX33DriveStrength[2:0] = 3'h3

2.2.9 eLink clocks

Enable 40 MHz clock at ECLK00.

108. [\[0x06c\] EPCLK0ChnCntrH](#) (page 150)
 - EPCLK0Freq[2:0] = 3'h1
 - EPCLK0DriveStrength[2:0] = 3'h3

Enable 80 MHz clock at ECLK01.

109. [\[0x06e\] EPCLK1ChnCntrH](#) (page 151)
 - EPCLK1Freq[2:0] = 3'h2
 - EPCLK1DriveStrength[2:0] = 3'h3

Enable 160 MHz clock at ECLK02.

110. [\[0x070\] EPCLK2ChnCntrH](#) (page 152)
 - EPCLK2Freq[2:0] = 3'h3
 - EPCLK2DriveStrength[2:0] = 3'h3

2.2.10 Phase-shifter clocks

51. [\[0x033\] PSDllConfig](#) (page 134)
 - EPRXUnLockThreshold[3:0] = 4'h5
 - PSFSMClkAlwaysOn = 1'h0
 - PSDLLConfirmCount[1:0] = 2'h1
 - PSDllCurrentSel[1:0] = 2'h1

2.2.11 Finishing configuration

51. [\[0x0ef\] POWERUP2](#) (page 226)
 - dllConfigDone = 1'h1
 - pllConfigDone = 1'h1

CONFIGURATION

Warning:

Known issues: [Section 19.1.1](#), [Section 19.1.2](#), [Section 19.1.3](#). [Section 19.1.14](#).

The lpGBT chip can operate in one of several major modes: transmitter, receiver, transceiver. It has many application-specific settings, ePort data rates, clock speeds, driving strengths, and many more. **Out of the box, the lpGBT chip does not work until it is configured by the user to perform specific tasks.**

3.1 Configuration pins

The configuration is done by means of external configuration pins and internal configuration registers. All configuration pins should be tied up or down prior to chip power up. As all external pins have built in pull up/down resistors in particular cases they could be left unconnected. The logic level on the configuration pins should not change during chip operation.

3.1.1 MODE3, MODE2, MODE1, MODE0

MODE pins selects the basic operating mode as described in [Table 3.1](#). More verbose description can be found in [Section 1.3](#). All MODE pins have internal pull down resistors.

Table 3.1: MODE pins decoding.

MODE [3:0]	Tx Data Rate	Tx Encoding	IpGBT Mode
4'b0000	5 Gbps	FEC5	Off
4'b0001	5 Gbps	FEC5	Simplex TX
4'b0010	5 Gbps	FEC5	Simplex RX
4'b0011	5 Gbps	FEC5	Transceiver
4'b0100	5 Gbps	FEC12	Off
4'b0101	5 Gbps	FEC12	Simplex TX
4'b0110	5 Gbps	FEC12	Simplex RX
4'b0111	5 Gbps	FEC12	Transceiver
4'b1000	10 Gbps	FEC5	Off
4'b1001	10 Gbps	FEC5	Simplex TX
4'b1010	10 Gbps	FEC5	Simplex RX
4'b1011	10 Gbps	FEC5	Transceiver
4'b1100	10 Gbps	FEC12	Off
4'b1101	10 Gbps	FEC12	Simplex TX
4'b1110	10 Gbps	FEC12	Simplex RX
4'b1111	10 Gbps	FEC12	Transceiver

3.1.2 SC_I2C

SC_I2C pin selects I2C or serial interface (IC/EC) channel for ASIC configuration. The SC_I2C pin has an internal pull down resistor.

Table 3.2: SC_I2C pin function.

SC_I2C	Mode[1:0]	Description
1'b0	2'b00 (off)	Serial interface inactive.
1'b0	2'b01 (simplex TX)	EC Serial interface active.
1'b0	2'b10 (simplex RX)	EC Serial interface active.
1'b0	2'b11 (Transceiver)	IC Serial interface active.
1'b1	2'bxx	I2C interface active.

3.1.3 ADR3, ADR2, ADR1, ADR0

ADR pins set low significant bits of IpGBT chip address used in I2C and/or serial control interfaces. The whole chip address is derived according to the description in [Section 3.3](#).

3.2 Register access

The IpGBT contains a number of configuration registers. All the registers can be divided into three categories: Read/Write/Fuse, Read/Write, Read only. The first group of registers (Read/Write/Fuse) is special, as during the power-up sequence, these registers can be optionally loaded with the values of the e-fuses. The second group of registers (Read/Writa) can only be modified through the I2C interface or the serial (IC/EC) interface. There is also a number of read-only registers to allow monitoring of the status of certain logic blocks.

Figure [Fig. 3.1](#) illustrates the different options for accessing the IpGBT registers.

The IpGBT chip has 453 8-bit registers. First 320 registers can be written using the interfaces described above, while the last registers are read-only. Each register has a unique 16-bit address.

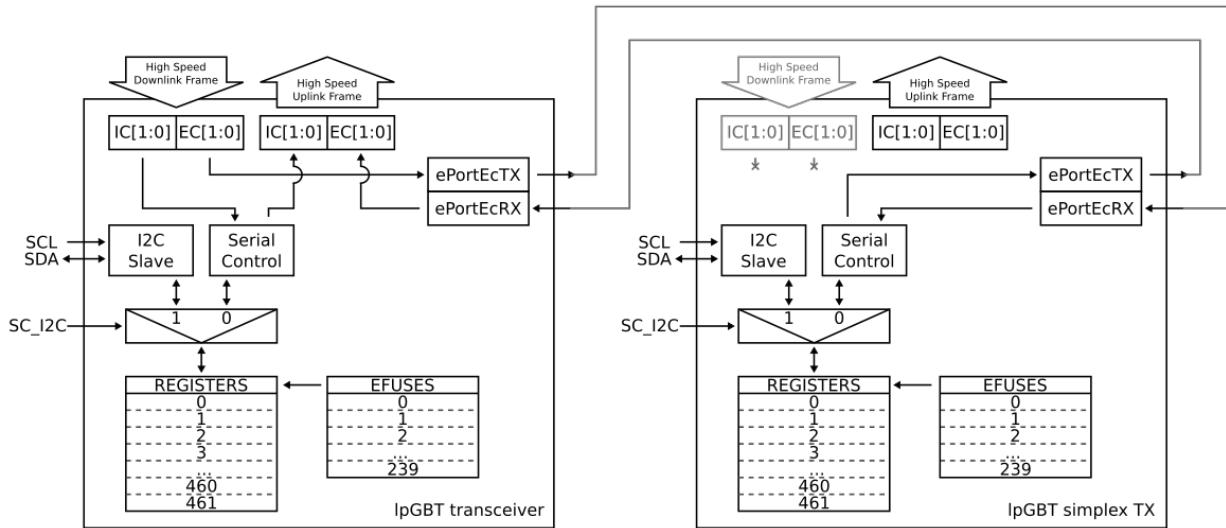


Fig. 3.1: IpGBT configuration

All writable registers are preset to zero after power up. Values of the first 240 registers can be loaded from e-fuses during chip initialization (see chapter [Section 8.1](#)). Most of the registers which have corresponding e-fuses are used to configure IpGBT operation.

Values of the registers can be set by:

1. Transfer from e-fuses. This is done only during the start-up sequence and if the `updateEnable` fuse has been blown (see [Section 8.1](#) for more details about power up). To read more about electrical fuses please refer to chapter [Section 3.6](#).
2. Through the serial interface. In the transceiver mode, the data for the serial interface is taken from the GBT frame and hence requires that the full duplex link is operational to allow write and read access (see left IpGBT in figure [Fig. 3.1](#)). In the simplex mode, the data is provided by **ePortRxEc/ePortTxEc** channel (see right IpGBT in figure [Fig. 3.1](#)). In order to use this interface the external pin **SC_I2C** has to be set to 0 (which is the default value when the pin is left unconnected). This interface is described in detail in chapter [Section 3.4](#).
3. Through the I2C interface. This is an asynchronous interface and does not require that the IpGBT link is working properly. This option is selected by setting the external pin **SC_I2C** to 1. This interface is described in detail in [Section 3.5](#).

All writable registers are protected against SEU by means of triplication. The IpGBT features a 16-bit counter which is incremented every time an upset is detected in the configuration memory. The value of the counter can be read by accessing [\[0x1cd\] ConfigErrorCounterH](#) (page 268) and [\[0x1ce\] ConfigErrorCounterL](#) (page 268) registers.

The IpGBT contains a number of read-only registers whose values can be read via the I2C interface or the serial interface. These registers contain the status of internal blocks of the IpGBT and can be used for diagnostic purposes.

The detailed list of all registers is summarized in chapter [Section 15](#).

3.3 Chip Address

The IpGBT I2C and IC/EC-channel addresses are identical and can be fully configured by the user. Four LSBs of the address are set by external **ADR3**, ..., **ADR0** pins (see [Section 3.1.3](#)), while the 3 MSBs are set by

ChipAddressBar[2:0] field in REG_CHIPADR register. Table 3.3 shows how the address is generated.

Table 3.3: IpGBT I2C/IC/EC address

ChipAddressBar[2:0]	ADR3	ADR2	ADR1	ADR0	Chip address[6:0]
3'b000	1'b0	1'b0	1'b0	1'b0	7'b1110000
3'b000	1'b0	1'b0	1'b0	1'b1	7'b1110001
3'b000	1'b0	1'b0	1'b1	1'b0	7'b1110010
3'b000	1'b0	1'b0	1'b1	1'b1	7'b1110011
3'b000	1'b0	1'b1	1'b0	1'b0	7'b1110100
3'b000	1'b0	1'b1	1'b0	1'b1	7'b1110101
3'b000	1'b0	1'b1	1'b1	1'b0	7'b1110110
3'b000	1'b0	1'b1	1'b1	1'b1	7'b1110111
3'b000	1'b1	1'b0	1'b0	1'b0	7'b1111000
3'b000	1'b1	1'b0	1'b0	1'b1	7'b1111001
3'b000	1'b1	1'b0	1'b1	1'b0	7'b1111010
3'b000	1'b1	1'b0	1'b1	1'b1	7'b1111011
3'b000	1'b1	1'b1	1'b0	1'b0	7'b1111100
3'b000	1'b1	1'b1	1'b0	1'b1	7'b1111101
3'b000	1'b1	1'b1	1'b1	1'b0	7'b1111110
3'b000	1'b1	1'b1	1'b1	1'b1	7'b1111111
...
3'b001	1'b0	1'b0	1'b0	1'b0	7'b1100000
3'b010	1'b0	1'b0	1'b0	1'b0	7'b1010000
3'b011	1'b0	1'b0	1'b0	1'b0	7'b1100000
3'b100	1'b0	1'b0	1'b0	1'b0	7'b0110000
3'b101	1'b0	1'b0	1'b0	1'b0	7'b0100000
3'b110	1'b0	1'b0	1'b0	1'b0	7'b0010000
3'b111	1'b0	1'b0	1'b0	1'b0	7'b0000000

The default chip address is $7'b1110000$ ($7'h70$) if field ChipAddressBar[2:0] is not changed and ADR3, ..., ADR0 pins are left floating. In most of the applications, the user is not expected to change ChipAddressBar[2:0] value. This feature is foreseen to be used only in systems where more than 16 IpGBT chips will operate on the same I2C bus or if the default IpGBT address collides with another device on the bus. When changing the ChipAddressBar[2:0] value, it is not recommended to set it to $3'b111$ while connecting ADR3, ..., ADR0 pins to zeros as it creates a collision with the I2C broadcast address.

The value of the currently set address can be read back from [\[0x141\] I2CSlaveAddress](#) (page 247) register.

3.4 Serial control and monitoring interface

The availability of IC or EC configuration interfaces is dictated by the mode of operation. If IpGBT is configured in transceiver mode, IC channel is used to control the chip. Otherwise, in simplex transmitter or simplex receiver modes, EC channel is used to control the chip (see Fig. 3.1).

As discussed in Section 4 four bits of the IpGBT frame are reserved for slow control applications. If the chip operates in transceiver mode two of these bits (IC[1:0]) are reserved for control and monitoring of the IpGBT operation. The other two (EC[1:0]) are made available externally to allow the implementation of a slow control link to another chip, however their actual use is not restricted to that type of applications. In simplex modes IC[1:0] and EC[1:0] fields in the IpGBT frame are not used.

When I2C mode is selected, the user should ensure that the bits IC[1:0] in the downlink IpGBT frame are set to 2'b11.

In the transceiver mode, data field bits $\text{IC}[1:0]$ are used for control and monitoring of the IpGBT operation. These 2 bits implement an 80 Mb/s serial channel that is used to read and write the IpGBT internal registers. This channel is used during normal operation to program and monitor the operation of the IpGBT.

The two bits $\text{IC}[1:0]$ from subsequent frames are demultiplexed to form 8-bit words which follow a frame-based protocol. The protocol for data sent to the IpGBT for a write-read operation is shown in Table [Table 3.4](#) and for a read-only operation in Table [Table 3.5](#).

Table 3.4: IC/EC channel frame structure sent to IpGBT for a write-read sequence

ID	Description	Parity check
A	Frame delimiter 8'b 01111110	No
B	Reserved	No
C	IpGBT address (7 bits) + R/W bit = 0	No
D	Command [7:0]	Yes
E	Number of data words n[7:0]	Yes
E	Number of data words n[15:8]	Yes
F	Memory address [7:0]	Yes
F	Memory address [15:8]	Yes
G	1st data (8 bits)	Yes
G	...	Yes
G	nth data (8 bits)	Yes
H	Parity word (8 bits)	Yes
A	Frame delimiter 8'b 01111110	No

Table 3.5: IC/EC channel frame structure sent to IpGBT for a read-only sequence

ID	Description	Parity check
A	Frame delimiter 8'b 01111110	No
B	Reserved	No
C	IpGBT address (7 bits) + R/W bit = 1	No
D	Command [7:0]	Yes
E	Number of data words n[7:0]	Yes
E	Number of data words n[15:8]	Yes
F	Memory address [7:0]	Yes
F	Memory address [15:8]	Yes
H	Parity word (8 bits)	Yes
A	Frame delimiter 8'b 01111110	No

When a write-read or read-only frame is received by the IpGBT and the addresses matches the chip address (see [Section 3.3](#)), the IpGBT will acknowledge receipt of the data by sending a similar frame back (on the uplink or EC channel). IpGBTs that are not addressed will not return any data. A broadcast address (7'b0000000) can be used to write the same data to a number of IpGBTs. In this case, the IpGBT will not send the acknowledge frame back. Note that the IpGBT will not carry out any subsequent operations until the read sequence is complete.

As shown in tables [Table 3.4](#) and [Table 3.5](#) the write-read and write-only operations follow the following structure:

1. The beginning and end of the frame are marked with the delimiter word (8'b 01111110). To ensure that a payload word is not misinterpreted as the delimiter, bit stuffing is used so that any sequence of five consecutive 1s in the payload is always followed by a 0. This bit-stuffing must be carried out by the corresponding transmitter and the de-stuffing by the receiver.
2. Reserved byte. Should be set to zero.

3. An address word is then transmitted and contains the 7-bit address of that particular IpGBT and a Read/Write (R/W) bit. If the address does not match, then the subsequent actions are not carried out and the IpGBT will not send the acknowledge frame back on the uplink. If the R/W bit is 1, then the configuration registers are not modified but their contents are read back in the transmitted GBT frame. If the R/W bit is 0, then the registers are over-written with the values transmitted within this frame. The new values are read back in the transmitted GBT frame.
4. A Command word is then transmitted. In version 0 of the IpGBT, the data in this word is ignored.
5. This is followed by two bytes to indicate the number of data words (n) in the packet, maximum 65k.
6. Then the internal address (2 bytes) of the first register to be accessed is transmitted.
7. The n data words then follow. This scheme allows access to a single register or a block of registers in consecutive memory addresses. In the frame for a read-only sequence, no data bytes are transmitted to the IpGBT.
8. Finally, a parity word is transmitted where each bit is the final parity of that bit through all bytes of the frame, except the frame delimiters and the first byte “IpGBT address + R/W”. The user should calculate a running parity and transmit it as this final word. The IpGBT constructs the same parity sum from the received data and compares it to the last word of the data packet. The result of this comparison is stored in a SCStatus register (see [Section 15.3.15](#)) and can be accessed by the user (logic 1 if the parity check was OK).

The structure of the frame returned by the IpGBT is the same as in [Table 3.4](#), with two exceptions: fields B and C are swapped in the response; the Command word consists of seven 0s concatenated with an LSB which is the status bit of the previous parity check (logic 1 if the parity check was OK). If the parity checks in a write-read operation, the data payload is written to the respective registers and the data bytes in the returned frame are the new values that have just been written into the registers. The parity word returned is calculated based on these values. The parity check can be disabled by asserting bit scParityCheckDisable in the SCCONFIG register (see [Section 15.1.6](#)). If the parity check is disabled, the data payload is always written to the respective registers independently of the result of the parity check.

The last parity bit check result can be read from bit SCParityValid of registers SCStatus (see [Section 15.3.15](#)). Note that the result of parity check for a read-only command is not stored in the status register so that the status bit always reflects the result of the parity check for the last write-read command.

3.5 I2C slave interface

Warning:

Known issues: [Section 19.1.2](#), [Section 19.1.3](#).

The I2C slave port allows the writing and reading of the IpGBT configuration registers. This can be used when the IpGBT is operated in any of its modes.

3.5.1 Write to Register

This configuration mode supports access to one individual register or a block of registers in consecutive addresses. To access registers, the I2C master must issue the correct slave-address, write the register address and then write/read the register data. The steps in the protocol are as follows:

1. Master transmits START command.
2. Master transmits the 7-bit IpGBT address followed by the 8th bit (R/W) set to zero.
3. Master transmits bits [7:0] of the register address.

4. Master transmits bits [15:8] of the register address.
5. Master transmits 8-bit register data word (can be repeated).
6. Master transmits STOP command.

After step 5, the register address is automatically incremented. This feature allows a block of consecutive registers to be written in one sequence. The address in steps 3/4 is the first register of the block and step 5 is repeated with the correct register data introduced each time.

3.5.2 Read from Register

1. Master transmits START command
2. Master transmits 7-bit IpGBT address followed by the 8th bit (R/W) set to zero.
3. Master transmits bits [7:0] of the register address.
4. Master transmits bits [15:8] of the register address.
5. Master transmits repeated START command
6. Master transmits 7-bit IpGBT address followed by the 8th bit (R/W) set to one.
7. Slave transmits 8-bit register data word (can be repeated).
8. Master transmits STOP command.

After step 7, the register address is automatically incremented. This feature allows a block of configuration registers to be read in one sequence starting with the register addressed by steps 3/4.

3.6 E-FUSES

The IpGBT is equipped with a number of e-fuses. Each of the bits of the first 240 write-able configuration registers has a corresponding e-fuse. The optimal configuration parameters of the IpGBT can then be written into the fuse array. The transfer from fuses to configuration registers is done during the automatic power-up sequence (see chapter [Section 8.1](#)). This then allows the IpGBT to self-configure into an operational state after power-up.

Each e-fuse connects to a node whose logical state can be sampled and stored in the corresponding register bit. By default, an un-programmed e-fuse will pull this node down to logic 0. If the e-fuse is programmed (blown), the node will be pulled up to logic 1.

3.6.1 E-fuse power

E-Fuse programming requires the pin VDDF2V5 to be powered at 2.5V. This power is only to be supplied during programming of the E-Fuses. For normal operation this voltage must be kept at 0V.

Warning: The pin VDDF2V5 should only be powered ($VDDF2V5 = 2.5V$) strictly during the duration of the E-Fuses programming. In all other circumstances, this pin should be grounded ($VDDF2V5 = 0V$). The users must observe this recommendation.

3.6.2 E-fuse addressing

The e-fuses are grouped into 32 fuses to correspond to the 4 subsequent configuration registers. Each group of fuses has a 16-bit address which is the same as the address of the corresponding register. All of the 32 fuses in one group are programmed together.

3.6.3 E-fuse programming

Warning: Known issues: [Section 19.1.14](#).

Each fuse can be programmed using the I²C or serial interface. The sequence is:

1. Load a magic number 0xA3 to the [Section 15.2.3](#) register to unlock fuse blowing.
2. Load 12d into FuseBlowPulseLength[3:0] field in the [Section 15.2.3](#) register.
3. Load fuse address into the [Section 15.2.3](#) and [Section 15.2.3](#) registers.
4. Load the 32-bit data pattern to be programmed into the fuses into the [Section 15.2.3](#), [Section 15.2.3](#), [Section 15.2.3](#), [Section 15.2.3](#) registers. Logic 1 will burn the fuse, logic 0 will not burn the fuse.
5. Switch on VDDF2V5 (2.5 V).
6. Initiate blowing sequence by writing one into FuseBlow bit in [Section 15.2.3](#) register.
7. Keep reading [Section 15.3.4](#) register until FuseBlowDone bit is set.
8. Now programming sequence is finished, VDDF2V5 could be switched off.
9. Deassert FuseBlow bit in [Section 15.2.3](#) register.

On completion of these steps, the corresponding fuses should have been burned. However, one should bear in mind, that the fuse values will only be transferred to the configuration registers during the next power up if the updateEnable bit is blown in the [Section 15.1.17](#) register. When this bit is set, the power-up state machine will automatically transfer all the fuse values into the corresponding configuration registers in the **UPDATE FROM FUSES** state, as described in chapter [Section 8.1](#). Note that all bits of the configuration registers will be loaded with the value of their corresponding fuse, logic 1 (burned) or logic 0 (not burned). The fuse burning is a not reversible process.

Reasons why the fuse blowing operation may be unsuccessful (asserting FuseBlowError bit in the FUSEStatus registers) are: magic number was not set correctly, the two LSB of the address were not zero,

3.6.4 E-fuse reading

After blowing sequence is finished, the user can read back the fuse values in order to confirm that the write operation was successful. The sequence is:

1. Initiate the readout sequence by writing one into FuseRead bit in [Section 15.2.3](#) register.
2. Keep reading [Section 15.3.4](#) register until FuseDataValid is set.
3. Now reading sequence is finished.
4. Load fuse address into the ref:`REG_FUSEBLOWADDH` and ref:`'REG_FUSEBLOWADDL` registers.
5. Read values of the currently selected 32-bits fuse values by reading [Section 15.3.4](#), [Section 15.3.4](#), [Section 15.3.4](#), [Section 15.3.4](#) registers.
6. Deassert FuseRead bit in [Section 15.2.3](#) register.

Steps 4 and 5 can be repeated more than one time to get values for more than one e-fuse group.

3.7 Configuration flows

There are three main configuration mechanisms foreseen: complete configuration stored in e-fuses, configuration written at power up via I2C interface, configuration written at power up via serial interface with minimum configuration stored in fuses. It should be noted, that the configuration process is very closely linked to IpGBT start-up sequence (see [Section 8](#)). Details of these configuration flows are given in the following chapters.

3.7.1 Complete configuration stored in fuses

In the majority of systems, the configuration of eLinks (number, data rate) is fixed by the system architecture. In this case, it is recommended to store the complete configuration in e-fuses. This mechanism provides the most robust operation, as no further action is required by the user after the system reset. Moreover, the user is not required to have an access to I2C nor IC/EC interface.

In order to use this configuration flow, one should blow the whole IpGBT configuration, making sure that bits:

- `updateEnable` - to copy values from fuses
- `pllConfigDone` - to start PLL initialization procedure immediately
- `dllConfigDone` - to start DLL initialization procedure immediately

are set (blown).

3.7.2 Configuration over I2C

This mode of configuration provides the most flexibility as fusing operation is not required. If `updateEnable` fuse is not blown, the IpGBT chip will pause and wait for configuration after power up. User can freely write required registers either using random memory access or providing the whole IpGBT configuration as a bit stream for addresses 0x01C-0xEF in one I2C transaction. Once the configuration process is finished, user should set bits `pllConfigDone` and `dllConfigDone` to proceed with the initialization.

3.7.3 Using serial control channel

In order to use serial control channel some prerequisites must be met. In particular, to establish reliable serial connection over IC/EC channel, the PLL inside the IpGBT has to be locked. This can only be achieved by blowing minimum configuration to e-fuses. The minimal configuration includes:

- PLL/CDR settings (blown during production testing)
- equalizer settings
- line driver settings (including an optional I2C master transaction to configure GBLD/GBLD10P/LDQ10/VLAD laser driver or TIA)
- EC/IC channel settings

In order to copy configuration from fuses to memory, the `updateEnable` bit has to be asserted. Moreover, the `pllConfigDone` has to be set to start PLL initialization procedure.

User should use IC/EC channel to read `PUSMState` field from [Section 15.3.14](#). Initially, during PLL looking phase, the IpGBT will not reply (or the response will be invalid). At some point, the valid frame should be returned and `PUSMState` should have value **WAIT_DLLS_CONFIG**. Now chip is ready to be configured by serial interface.

User can configure the chip using random memory access or providing the whole IpGBT configuration as a bit stream in one SC frame. Once the configuration process is finished, user should set bit `dllConfigDone` to proceed with the initialization.

It is worth mentioning that changing some register values may lead to the serial link rupture. Registers mentioned as prerequisites earlier in this section should not be changed after chip has been initialized.

HIGH SPEED LINKS

The lpGBT communicates with the counting room through optical links: the link from the counting room to the lpGBT is called "**downlink**" and its bandwidth is 2.56 Gb/s. The link from the lpGBT to the counting room is called "**uplink**" and its bandwidth can be set by the user to 5.12 Gb/s or 10.24 Gb/s. Both the up and downlinks use **Forward Error Correction (FEC)** to detect and correct transmission errors and **Scrambling** to constrain the DC unbalance of the transmitted data and to enable reliable **Clock and Data Recovery (CDR)**. Additionally, transmitted FEC codes are further interleaved to improve the efficiency of the FEC code. Details of the up and downlink frames are given below.

4.1 Downlink frame

The downlink frame is composed of 64-bits that are transmitted every 25 ns (the LHC bunch crossing period) resulting in a data rate of 2.56 Gb/s. The frame is organized as follows:

- **H-field:** Four bit Header ($\mathbf{H[3:0] = 4'b1001}$) that delimits the start of the frame. Four bits are used to guaranty the DC balance of the header code and to implement header redundancy allowing robust header detection in the presence of noise and/or single event upsets;
- **IC-field:** Composed of 2 bits, implements the downlink of the Internal Control (IC) channel used to control the lpGBT itself (only operational in transceiver mode). The data rate is 80 Mb/s;
- **EC-field:** Composed of 2 bits, implements the downlink of the External Control (EC) channel. These bits are made available on the differential pair pins **EDOUTECP** and **EDOUTECN**. The data rate is 80 Mb/s;
- **D-field:** Composed of 32 bits, carries the user data to be transmitted to the frontend by the eLinks. The associated eLinks can be configured to have bandwidths: 80, 160 or 320 Mb/s. The aggregated available bandwidth is 1.28 Gb/s;
- **FEC-field:** Composed of 24 bits carries the Forward Error Correction code to detect and correct transmission errors due to noise or Single Event Upsets (**SEU**).

4.1.1 Frame format

Clock and Data Recovery at the lpGBT requires the incoming serial data stream to have a high density of "0-to-1" and "1-to-0" transitions. This cannot be guaranteed on the outset for the IC, EC and D fields (although it is true for the Header field). To make sure that this is the case, those three fields are scrambled before they are inserted in the frame for transmission. The FEC codes to be transmitted, together with the data, are computed from the scrambled IC, EC and D fields. The assembled frame structure obtained after this chain of operations is represented in Fig. 4.1. (Please note that this figure actually represents the downlink frame structure either before it is interleaved at the transmitter or after it has been de-interleaved at the receiver.)

Several points are worth notice for the downlink frame:

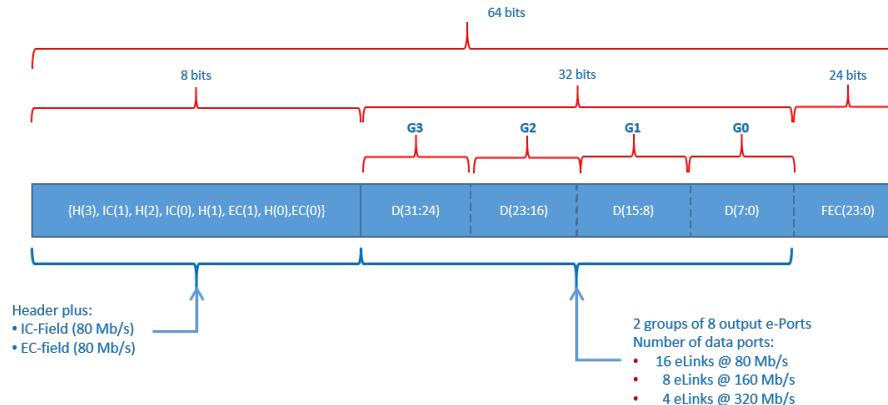


Fig. 4.1: Downlink frame structure before interleaving.

- The frame is transmitted using the convention of "MSB" first, being the first bit to be transmitted H(3) and the last FEC(0);
- The H, IC and EC fields are "split" and their bits interleaved. This is done to guarantee the statistical ("random") properties of the first 8 bits in the frame. Please keep in mind, that the H-field is a fixed pattern while the IC and EC fields are scrambled before they are inserted in the frame;
- The D-Field is "logically" split into four bytes with each byte associated with an ePort group;
- The FEC-field is computed, as will be detailed below, from the scrambled IC, EC and D fields.

How the downlink frame (**FRAMEDWN[63:0]**) is organized is given in Fig. 4.2.

Frame	Function	I/O Group
FRMDWN[23:0]	FEC[23:0]	
FRMDWN[31:24]	Data[7:0]	0
FRMDWN[39:32]	Data[15:8]	1
FRMDWN[47:40]	Data[23:16]	2
FRMDWN[55:48]	Data[31:24]	3
FRMDWN[56]	EC[0]	EC
FRMDWN[57]	H[0]	
FRMDWN[58]	EC[1]	EC
FRMDWN[59]	H[1]	
FRMDWN[60]	IC[0]	
FRMDWN[61]	H[2]	
FRMDWN[62]	IC[1]	
FRMDWN[63]	H[3]	H[3:0] = 4'b1001

Fig. 4.2: Downlink frame structure (before interleaving)

4.1.2 Forward error correction

Due to Noise, Intersymbol Interference or SEUs information might be corrupted during transmission, **Forward Error Correction (FEC)** allows to correct transmission errors without the need for the data to be re-transmitted. This is achieved by transmitting, together with the data, "parity" bits. This means that transmission robustness is achieved at the cost of data bandwidth. The FEC codes used in the IpGBT belong to a class of codes called **Reed-Solomon Codes [Reed-Solomon]** (page 325). These codes operate on "non-binary" symbols formed of **m** bits. A message composed of **k** symbols is encoded into an **n** symbol word with $n = 2^m - 1$. The number of parity symbols is then $n - k$, which allows to correct up to $t = (n - k)/2$ symbols (or equivalently $t = m(n - k)/2$ bits). The downlink uses a FEC code with **m = 3**, and thus the number of symbols in the work is $n = 2^m - 1 = 7$. The code was chosen to be able to correct one

symbol ($t = 1$), and thus the number of parity symbols is $n - k = 2$ $t = 2$ allowing for $k = n - 2t = 5$ data symbols. The code used is designated as $\text{RS}(n,k) = \text{RS}(7,5)$. In the IpGBT 4 code groups are interleaved allowing to correct up to 4 $t = 4$ symbols or, equivalently, 4 m $t = 12$ bits. Potentially the RS code can handle up to 60 user bits but only 40 are effectively used and transmitted due to the limited size of the frame (64 - bits): the IC, EC and D fields. For proposes of encoding, the 20 remaining bits must be padded and fed to the encoder (PADDWN[19:0]). The padding bits are not transmitted but assumed received error free by the receiver. At the receiver these "known" bits are used to feed the FEC decoder. However all the FEC code bits must be transmitted. The coding procedure at the transmitter (counting room) is as follows:

- Use frame bits EC, IC, D and PADDWN[19:0] to compute the FEC field FRMDWN[23:0];
- Interleave bits FRMDWN[63:0] to construct an interleaved frame IFRMDWN[63:0];
- Transmit the interleaved frame IFRMDWN[63:0];

At the receiver (IpGBT) the decoding procedure is as follows:

- De-interleave IFRMDWN[63:0] to obtain FRMDWN[63:0];
- Pad the received frame FRMDWN with the known pad bits PADDWN[19:0];
- Detect and correct errors (if needed)
- Recover the error free frame FRMDWN[63:24] to extract H[3:0] + IC[1:0] + EC[1:0] + Data[31:0];

For the uplink the operation sequence is the same except the encoding is made in the IpGBT and the decoding in the counting room. For both the up and downlinks the padding bits are all "0".

4.1.3 De-scrambling (and scrambling)

Clock and Data Recovery (**CDR**) circuits are used in the IpGBT system (the IpGBT ASIC and the IpGBT-FPGA) to generate a clock that is exactly at the same frequency and phase as the incoming serial bit stream (2.56 Gb/s for the downlink and 5.12 or 10.24 Gb/s for the uplink). This clock is used to re-sample the bit stream before it is further de-serialized. CDR circuits require the presence of "0-to-1" and "1-to-0" transitions in the bit-stream ("defining" the bit boundaries) to be able to extract the needed frequency and phase information. The higher the density of the transitions the easier it is for the CDR circuit to keep track of the phase/frequency information and the lower will be the jitter (phase noise) of the recovered clock. On the outset, there is no guaranty that the data to be transmitted by the IpGBT links will satisfy the condition of high density of transitions. Scrambling is thus used to make sure that the transmitted data has the characteristics of random data and thus that a high density of transitions will be present in the transmitted serial bit stream. To scramble the data the IpGBT systems use a scrambler circuit in the transmitter and a de-scrambler circuit in the receiver to recover the original data. It is important that the scrambler and the de-scrambler will be properly synchronized. In the IpGBT, the de-scrambler uses a self-synchronizing architecture [*scrambler*] (page 325) meaning that no synchronization pattern (or reset signal) needs to be transmitted (or activated) for the de-scrambler to synchronize. Given its architecture, the de-scrambler will synchronize in a single cycle which, in the case of the IpGBT, means that the reception of a single frame is enough (an necessary) to synchronize the de-scrambler. Scramblers / De-scramblers can be made to operate either on the serial bit stream or on the parallel data after de-serialization. "Serial scramblers/de-scramblers" require less circuitry but need the circuit to operate at the bit rate, while "parallel scramblers/de-scramblers", although slightly more complex can be made to operate at the parallel bus frequency and are thus less critical to implementation. In the case of the IpGBT the data path bus operates at 40.08 MHz and thus a scrambling / de-scrambling "cycle" takes approximately 25 ns (one LHC machine clock cycle).

For the downlink the number of bits to be scrambled/de-scrambled is 36: EC[1:0], IC[1:0] and D[31:0]. A 36-bit scramble/de-scrambler (order 36) is used that implements the following scrambling recursive equation: $S_i = D_i \text{ xnor } S_{i-25} \text{ xnor } S_{i-36}$. The architecture of the scrambler and de-scrambler are represented in [Fig. 4.3](#) and [Fig. 4.4](#) respectively.

For testing purposes, the de-scrambler can be bypassed as indicated in [Fig. 4.4](#). Please refer to register [[0x132](#)] [DataPath](#) (page 243) for details on how to bypass the de-scrambler.

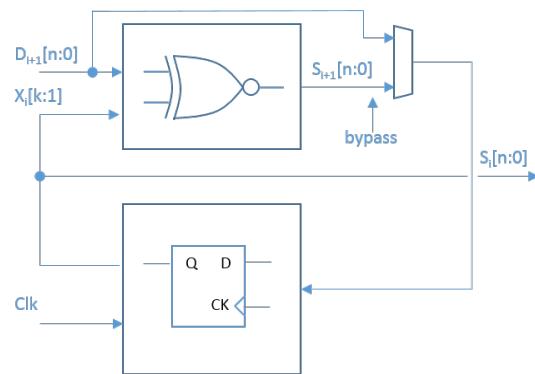


Fig. 4.3: Scrambler architecture

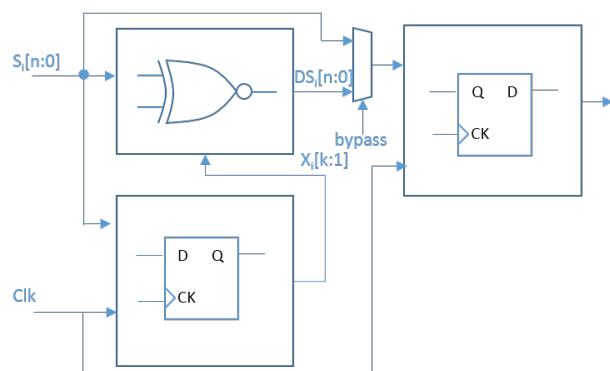


Fig. 4.4: De-scrambler architecture

4.1.4 De-interleaving (and interleaving)

Since the downlink FEC code is only able to correct single ($t=1$) symbol errors (3 bits) and since four encoders are used to cover the full frame (36 data bits), the codes for the four encoders are interleaved to give an error correction capability of up to four consecutive symbols or, which is the same, up to 12 consecutive wrong bits. The way the interleaving is done for the downlink frame is given in Fig. 4.5. In that table, IFRMDWN[63:0] represents the frame as transmitted over the optical fiber with IFRMDWN[63] being transmitted first and IFRMDWN[0] last. The Fig. 4.5 details how the frame bits are associated with the FEC codes, Data, EC, IC fields and Header. Notice that, as explained before, the header field is interleaved with the IC and EC fields. The table also specifies which code group a field belongs to.

Interleaved Frame		
Interleaved Frame	Assignment	Code group
IFRMDWN[2:0]	FEC[2:0]	0
IFRMDWN[5:3]	FEC[8:6]	1
IFRMDWN[8:6]	FEC[14:12]	2
IFRMDWN[11:9]	FEC[20:18]	3
IFRMDWN[14:12]	FEC[5:3]	0
IFRMDWN[17:15]	FEC[11:9]	1
IFRMDWN[20:18]	FEC[17:15]	2
IFRMDWN[23:21]	FEC[23:21]	3
IFRMDWN[26:24]	Data[2:0]	0
IFRMDWN[29:27]	Data[14:12]	1
IFRMDWN[32:30]	Data[26:24]	2
IFRMDWN[35:33]	Data[11:9]	3
IFRMDWN[38:36]	Data[5:3]	0
IFRMDWN[41:39]	Data[17:15]	1
IFRMDWN[44:42]	Data[29:27]	2
IFRMDWN[47:45]	Data[23:21]	3
IFRMDWN[50:48]	Data[8:6]	0
IFRMDWN[53:51]	Data[20:18]	1
IFRMDWN[56:54]	{EC[0], Data[31:30]}	2
IFRMDWN[59:57]	{H[1], EC[1], H[0]}	HEADER, 3
IFRMDWN[63:60]	{H[3], IC[1], H[2], IC[0]}	HEADER, 3

Fig. 4.5: Downlink interleaved frame structure

4.1.5 Data, groups and eLinks mapping

The way the data and EC fields map into eLinks is detailed in Section 7.

4.1.6 Frame alignment and fixed latency

The CDR circuit (see Section 9) recovers the clock and the serial data. The CDR circuit is followed by the Frame Aligner (FA) circuit whose function is to phase align the 40 MHz clock signals with the phase of the header of the incoming IpGBT frame.

The phase of the 40 MHz clocks can be rotated by pulsing the signal skip cycle. This is done automatically under the control of the Frame Aligner State Machine (FASM).

The Frame Aligner State Machine (FASM) controls the locking process of the IpGBT receiver. The states can be generically divided into search states and tracking states. The search states are entered at beginning of operation or when the FASM can't identify valid headers in the received frames. The tracking states keep track of frames with valid headers based on the locking phase. The tracking states implement a *robust* algorithm to avoid that either transmission errors or single event upsets will unduly lead the FASM to enter the search states.

When searching for the header, due to the random nature of the received data, it is likely that *valid frame headers* will be found even if the recovered clock phase (40 MHz) is not aligned with the frame header. To prevent false locks (frame lock), the FASM waits till it sees a certain number of consecutive frames detected with the same phase. Only if this condition is fulfilled does the FASM consider that frame lock has been achieved. The number of consecutive valid frame headers that have to be detected before frame lock is assumed is given by the contents of [\[0x02f\] FAMaxHeaderFoundCount](#) (page 134) register.

When in the tracking states the FASM will *resist* being lead back to search states, that is, it will avoid that either transmission errors or SEUs will lead to the search states since the search operation is time consuming and will result in a relatively large dead time for the link. How quick the FASM will return to the search states is controlled by [\[0x031\] FAMaxHeaderNotFoundCount](#) (page 134) and [\[0x030\] FAMaxHeaderFoundCountAfterNF](#) (page 134) registers.

The [\[0x031\] FAMaxHeaderNotFoundCount](#) (page 134) register sets represents the maximum number of invalid headers (consecutive or not) that can be received before the frame is considered misaligned. If the number of detected invalid headers does not exceed the number stored in [\[0x031\] FAMaxHeaderNotFoundCount](#) (page 134) and if the number of consecutive detected valid headers exceeds the number stored in [\[0x030\] FAMaxHeaderFoundCountAfterNF](#) (page 134), then the invalid frame count is reset.

4.2 Uplink frames

The lpGBT uplink frames have similar structure as the downlink frame however, given that two data rates (5.12 and 10.24 Gb/s) and two FEC codes (FEC5 and FEC12) can be used, there are four lpGBT uplink frame types. The rational and principles for scrambling, encoding and interleaving are similar to the ones used for the downlink and will not be revisited in what follows.

4.2.1 Frame formats

The uplink frame length depends on the data rate, being the frame 128-bit for 5.12 Gbps transmission and 256-bit for 10.24 Gbps. Additionally, since the length of the FEC field depends on the error correction strength (FEC5 or FEC12) the length of the fields differs among the four modes. The exceptions are the H, IC and EC fields that have the same length for the four modes of operation as detailed below:

- **H-field:** Two bit Header (**H[1:0] = 4'b10**) that delimits the start of the frame;
- **IC-field:** Composed of 2 bits, implements the uplink of the Internal Control (IC) channel used to control the lpGBT itself (only operational in transceiver mode). The data rate is 80 Mb/s;
- **EC-field:** Composed of 2 bits, implements the uplink of the External Control (EC) channel. These bits receive data from the input differential pair pins **EDINECP** and **EDINECN**. The data rate is 80 Mb/s;
- **D-field and FEC-field:** These fields have variable length (depending on data rate and FEC code used) with an impact on the user bandwidth and error correction strength. The length of those fields is given in [Fig. 4.6](#). This table also details the error correction strength and the number of eLink groups for each mode;
- **LM-field.** The "Latency Measurement" field is a special field that allows estimating the round-trip latency of the transceiver link (excluding eLinks). In this field the two bits of the **Downlink IC-field** are returned by the transmitter. Obviously, this field is only valid when operating the ASIC as a transceiver. Depending on the transmitter data rate and FEC encoding, this field is padded with a different number of leading "zeros". Note that this field is not available when operating at 5.12 Gb/s with FEC5 encoding, please see the tables below for details.

The details of the frame bit allocations are given in the following four tables.

Field	5.12 Gbps		10.24 Gbps	
	FEC5	FEC12	FEC5	FEC12
Frame [bits]	128		256	
Header [bits]	2		2	
IC [bits]	2		2	
EC [bits]	2		2	
D [bits]	112	96	224	192
FEC [bits]	10	24	20	48
LM [bits]	0	2	6	10
Correction [bits]	5	12	10	24
# of eLink groups	7	6	7	6

Fig. 4.6: Uplink frame field allocation summary

Frame	Function	I/O Group
FRMUP[9:0]	FEC[9:0]	
FRMUP[25:10]	Data[15:0]	0
FRMUP[41:26]	Data[31:16]	1
FRMUP[57:42]	Data[47:32]	2
FRMUP[73:58]	Data[63:48]	3
FRMUP[89:74]	Data[79:64]	4
FRMUP[105:90]	Data[95:80]	5
FRMUP[121:106]	Data[111:96]	6
FRMUP[123:122]	EC[1:0]	EC
FRMUP[125:124]	IC[1:0]	
FRMUP[127:126]	H[1:0] = 2'b10	HFH[1:0] = 2'b10

Fig. 4.7: 5.12 Gbps - FEC5 uplink frame structure (before interleaving)

Frame	Function	I/O Group
FRMUP[23:0]	FEC[23:0]	
FRMUP[39:24]	Data[15:0]	0
FRMUP[55:40]	Data[31:16]	1
FRMUP[71:56]	Data[47:32]	2
FRMUP[87:72]	Data[63:48]	3
FRMUP[103:88]	Data[79:64]	4
FRMUP[119:104]	Data[95:80]	5
FRMUP[121:120]	DownIC[1:0]	See text
FRMUP[123:122]	EC[1:0]	EC
FRMUP[125:124]	IC[1:0]	
FRMUP[127:126]	H[1:0]	HFH[1:0] = 2'b10

Fig. 4.8: 5.12 Gbps - FEC12 uplink frame structure (before interleaving)

Frame	Function	I/O Group
FRMUP[9:0]	FEC[9:0]	
FRMUP[19:10]	FEC[19:10]	
FRMUP[35:20]	Data[15:0]	0
FRMUP[51:36]	Data[31:16]	0
FRMUP[67:52]	Data[47:32]	1
FRMUP[83:68]	Data[63:48]	1
FRMUP[99:84]	Data[79:64]	2
FRMUP[115:100]	Data[95:80]	2
FRMUP[131:116]	Data[111:96]	3
FRMUP[147:132]	Data[127:112]	3
FRMUP[163:148]	Data[143:128]	4
FRMUP[179:164]	Data[159:144]	4
FRMUP[195:180]	Data[175:160]	5
FRMUP[211:196]	Data[191:176]	5
FRMUP[227:212]	Data[207:192]	6
FRMUP[243:228]	Data[223:208]	6
FRMUP[249:244]	{4'b0, DownIC[1:0]}	See text
FRMUP[251:250]	EC[1:0]	
FRMUP[253:252]	IC[1:0]	
FRMUP[255:254]	H[1:0]	HFH[1:0] = 2'b10

Fig. 4.9: 10.24 Gbps - FEC5 uplink frame structure (before interleaving)

Frame	Function	I/O Group
FRMUP[47:0]	FEC[47:0]	
FRMUP[79:48]	Data[31:0]	0
FRMUP[111:80]	Data[63:32]	1
FRMUP[143:112]	Data[95:64]	2
FRMUP[175:144]	Data[127:96]	3
FRMUP[207:176]	Data[159:128]	4
FRMUP[239:208]	Data[191:160]	5
FRMUP[249:240]	{8'b0, DownIC[1:0]}	See text
FRMUP[251:250]	EC[1:0]	
FRMUP[253:252]	IC[1:0]	
FRMUP[255:254]	H[1:0]	H[1:0] = 2'b 10

Fig. 4.10: 10.24 Gbps - FEC12 uplink frame structure (before interleaving)

4.2.2 Scrambling

For the uplink scrambling is a function of the data rate (number of scramblers used) and the FEC code (scrambling equation used). This is summarized in Fig. 4.11.

	5.12 Gb/s		10.24 Gb/s	
	FEC5	FEC12	FEC5	FEC12
Data [bits]	116	102	232	204
Scrambler width [bits]	58	51	58	51
Scrambler order	58	49	58	49
Number of scramblers		2		4
Recursive equation	eq 2	eq 3	eq 2	eq 3
eq 2:	$S_i = D_i \text{xnor } S_{i-39} \text{xnor } S_{i-58}$			
eq 3:	$S_i = D_i \text{xnor } S_{i-40} \text{xnor } S_{i-49}$			

Fig. 4.11: Uplink scrambling vs FEC and data rate

For testing purposes, the scrambler can be bypassed as indicated in figure Fig. 4.3. Please refer to register [\[0x132\] DataPath](#) (page 243) for details on how to bypass the scrambler.

4.2.3 Forward Error Correction

The user is free to choose between two FEC codes and two data rates (this is done by hard-wiring the configuration pins as detailed in [Section 3](#)). The ASIC configuration impacts the user bandwidth, the error correction strength, the maximum number of available uplink eLinks and their bandwidth. This is summarized in tables Fig. 4.6 and Fig. 1.4.

The FEC codes used are:

- **FEC12:** Consists of RS(15,3) interleaved 3 times for 5.12 Gbps or 6 times for 10.24 Gbps. Notice that although the code has the potential to protect up to 156 or 142 bits for 3 or 6 times interleaving, because of the IpGBT frame size the number of (user) bits protected is truncated to 102 or 204. FEC12 can correct up to 12 or 24 consecutive (error burst) wrong bits for 5.12 Gbps or 10.24 Gbps, respectively;
- **FEC5:** Consists of RS(31,2) with no interleaving for 5.12 Gbps or 2 times interleaved for 10.24 Gbps. Similar to the FEC12 case the code is truncated due to the frame size and the number of protected bits is 116 or 232. FEC5 can correct up to 5 or 10 consecutive wrong bits for 5.12 Gbps or 10.24 Gbps, respectively.

4.2.4 Interleaving

Since a single encoder is used for FEC5 at 5.12 Gbps, in this case the frame is not interleaved and simply transmitted as in Fig. 4.7. For the other cases interleaving is implemented as indicated in the tables below.

4.3 High speed links polarity

To facilitate optimum PCB layout with identical routing for the positive and negative signals (with no VIAs and layer changes) the polarity of high-speed links can be inverted for the uplink and downlink independently. To invert the signal polarity one should set bits **highSpeedDataOutInvert** and **highSpeedDataInInvert** in the register [\[0x036\] ChipConfig](#) (page 136) for serializer output (uplink) and deserializer input (downlink) respectively.

5G12 FEC12 Frame Interleaving		
Interleaved Frame	Assignment	Code group
Ifrmup[3:0]	FEC[3:0]	0
Ifrmup[7:4]	FEC[11:8]	1
Ifrmup[11:8]	FEC[19:16]	2
Ifrmup[15:12]	FEC[7:4]	0
Ifrmup[19:16]	FEC[15:12]	1
Ifrmup[23:20]	FEC[23:20]	2
Ifrmup[27:24]	FRMUP[27:24]	0
Ifrmup[31:28]	FRMUP[61:58]	1
Ifrmup[35:32]	FRMUP[95:92]	2
Ifrmup[39:36]	FRMUP[31:28]	0
Ifrmup[43:40]	FRMUP[65:62]	1
Ifrmup[47:44]	FRMUP[99:96]	2
Ifrmup[51:48]	FRMUP[35:32]	0
Ifrmup[55:52]	FRMUP[69:66]	1
Ifrmup[59:56]	FRMUP[103:100]	2
Ifrmup[63:60]	FRMUP[39:36]	0
Ifrmup[67:64]	FRMUP[73:70]	1
Ifrmup[71:68]	FRMUP[107:104]	2
Ifrmup[75:72]	FRMUP[43:40]	0
Ifrmup[79:76]	FRMUP[77:74]	1
Ifrmup[83:80]	FRMUP[111:108]	2
Ifrmup[87:84]	FRMUP[47:44]	0
Ifrmup[91:88]	FRMUP[81:78]	1
Ifrmup[95:92]	FRMUP[115:112]	2
Ifrmup[99:96]	FRMUP[51:48]	0
Ifrmup[103:100]	FRMUP[85:82]	1
Ifrmup[107:104]	FRMUP[119:116]	2
Ifrmup[111:108]	FRMUP[55:52]	0
Ifrmup[115:112]	FRMUP[89:86]	1
Ifrmup[119:116]	FRMUP[123:120]	2
Ifrmup[123:120]	{FRMUP[91:90], FRMUP[57:56]}	0
Ifrmup[125:124]	FRMUP[125:124]	1
Ifrmup[127:124]	HFH[1:0] = 2'b10	HEADER

Fig. 4.12: 5.12 Gbps - FEC12 uplink interleaved frame structure

10G24 FEC5 Frame Interleaving		
Interleaved Frame	Assignment	Code group
Ifrmup[4:0]	FEC[4:0]	0
Ifrmup[9:5]	FEC[14:10]	1
Ifrmup[14:10]	FEC[9:5]	0
Ifrmup[19:15]	FEC[19:15]	1
Ifrmup[24:20]	frmup[24:20]	0
Ifrmup[29:25]	frmup[141:137]	1
Ifrmup[34:30]	frmup[29:25]	0
Ifrmup[39:35]	frmup[146:142]	1
Ifrmup[44:40]	frmup[34:30]	0
Ifrmup[49:45]	frmup[151:147]	1
Ifrmup[54:50]	frmup[39:35]	0
Ifrmup[59:55]	frmup[156:152]	1
Ifrmup[64:60]	frmup[44:40]	0
Ifrmup[69:65]	frmup[161:157]	1
Ifrmup[74:70]	frmup[49:45]	0
Ifrmup[79:75]	frmup[166:162]	1
Ifrmup[84:80]	frmup[54:50]	0
Ifrmup[89:85]	frmup[171:167]	1
Ifrmup[94:90]	frmup[59:55]	0
Ifrmup[99:95]	frmup[176:172]	1
Ifrmup[104:100]	frmup[64:60]	0
Ifrmup[109:105]	frmup[181:177]	1
Ifrmup[114:110]	frmup[69:65]	0
Ifrmup[119:115]	frmup[186:182]	1
Ifrmup[124:120]	frmup[74:70]	0
Ifrmup[129:125]	frmup[191:187]	1
Ifrmup[134:130]	frmup[79:75]	0
Ifrmup[139:135]	frmup[196:192]	1
Ifrmup[144:140]	frmup[84:80]	0
Ifrmup[149:145]	frmup[201:197]	1
Ifrmup[154:150]	frmup[89:85]	0
Ifrmup[159:155]	frmup[206:202]	1
Ifrmup[164:160]	frmup[94:90]	0
Ifrmup[169:165]	frmup[211:207]	1
Ifrmup[174:170]	frmup[99:95]	0
Ifrmup[179:175]	frmup[216:212]	1
Ifrmup[184:180]	frmup[104:100]	0
Ifrmup[189:185]	frmup[221:217]	1
Ifrmup[194:190]	frmup[109:105]	0
Ifrmup[199:195]	frmup[226:222]	1
Ifrmup[204:200]	frmup[114:110]	0
Ifrmup[209:205]	frmup[231:227]	1
Ifrmup[214:210]	frmup[119:115]	0
Ifrmup[219:215]	frmup[236:232]	1
Ifrmup[224:220]	frmup[124:120]	0
Ifrmup[229:225]	frmup[241:237]	1
Ifrmup[234:230]	frmup[129:125]	0
Ifrmup[239:235]	frmup[246:242]	1
Ifrmup[244:240]	frmup[134:130]	0
Ifrmup[249:245]	frmup[251:247]	1
Ifrmup[253:250]	{frmup[253:252], frmup[136:135]}	0
Ifrmup[255:254]	H[1:0] = 2'b10	HEADER

Fig. 4.13: 10.24 Gbps FEC5 uplink interleaved frame structure

10G24 FEC12 Frame Interleaving		
Interleaved Frame	Assignment	Code group
Ifrmup[3:0]	FEC[3:0]	0
Ifrmup[7:4]	FEC[11:8]	1
Ifrmup[11:8]	FEC[19:16]	2
Ifrmup[15:12]	FEC[27:24]	3
Ifrmup[19:16]	FEC[35:32]	4
Ifrmup[23:20]	FEC[43:40]	5
Ifrmup[27:24]	FEC[7:4]	0
Ifrmup[31:28]	FEC[15:12]	1
Ifrmup[35:32]	FEC[23:20]	2
Ifrmup[39:36]	FEC[31:28]	3
Ifrmup[43:40]	FEC[39:36]	4
Ifrmup[47:44]	FEC[47:44]	5
Ifrmup[51:48]	FRMUP[51:48]	0
Ifrmup[55:52]	FRMUP[85:82]	1
Ifrmup[59:56]	FRMUP[119:116]	2
Ifrmup[63:60]	FRMUP[153:150]	3
Ifrmup[67:64]	FRMUP[187:184]	4
Ifrmup[71:68]	FRMUP[221:218]	5
Ifrmup[75:72]	FRMUP[55:52]	0
Ifrmup[79:76]	FRMUP[89:86]	1
Ifrmup[83:80]	FRMUP[123:120]	2
Ifrmup[87:84]	FRMUP[157:154]	3
Ifrmup[91:88]	FRMUP[191:188]	4
Ifrmup[95:92]	FRMUP[225:222]	5
Ifrmup[99:96]	FRMUP[59:56]	0
Ifrmup[103:100]	FRMUP[93:90]	1
Ifrmup[107:104]	FRMUP[127:124]	2
Ifrmup[111:108]	FRMUP[161:158]	3
Ifrmup[115:112]	FRMUP[195:192]	4
Ifrmup[119:116]	FRMUP[229:226]	5
Ifrmup[123:120]	FRMUP[63:60]	0
Ifrmup[127:124]	FRMUP[97:94]	1
Ifrmup[131:128]	FRMUP[131:128]	2
Ifrmup[135:132]	FRMUP[165:162]	3
Ifrmup[139:136]	FRMUP[199:196]	4
Ifrmup[143:140]	FRMUP[233:230]	5
Ifrmup[147:144]	FRMUP[67:64]	0
Ifrmup[151:148]	FRMUP[101:98]	1
Ifrmup[155:152]	FRMUP[135:132]	2
Ifrmup[159:156]	FRMUP[169:166]	3
Ifrmup[163:160]	FRMUP[203:200]	4
Ifrmup[167:164]	FRMUP[237:234]	5
Ifrmup[171:168]	FRMUP[71:68]	0
Ifrmup[175:172]	FRMUP[105:102]	1
Ifrmup[179:176]	FRMUP[139:136]	2
Ifrmup[183:180]	FRMUP[173:170]	3
Ifrmup[187:184]	FRMUP[207:204]	4
Ifrmup[191:188]	FRMUP[241:238]	5
Ifrmup[195:192]	FRMUP[75:72]	0
Ifrmup[199:196]	FRMUP[109:106]	1
Ifrmup[203:200]	FRMUP[143:140]	2
Ifrmup[207:204]	FRMUP[177:174]	3
Ifrmup[211:208]	FRMUP[211:208]	4
Ifrmup[215:212]	FRMUP[245:242]	5
Ifrmup[219:216]	FRMUP[79:76]	0
Ifrmup[223:220]	FRMUP[113:110]	1
Ifrmup[227:224]	FRMUP[147:144]	2
Ifrmup[231:228]	FRMUP[181:178]	3
Ifrmup[235:232]	FRMUP[215:212]	4
Ifrmup[239:236]	FRMUP[249:246]	5
Ifrmup[243:240]	{FRMUP[183:182], FRMUP[81:80]}	0
Ifrmup[247:244]	{FRMUP[217:216], FRMUP[115:114]}	1
Ifrmup[251:248]	{FRMUP[251:250], FRMUP[149:148]}	2
Ifrmup[253:252]	{PAD[50:49], FRMUP[253:252]}	3
Ifrmup[255:254]	HFH[1:0] = 2'b10	HEADER

Fig. 4.14: 10.24 Gbps - FEC12 uplink interleaved frame structure

HIGH-SPEED LINE DRIVER

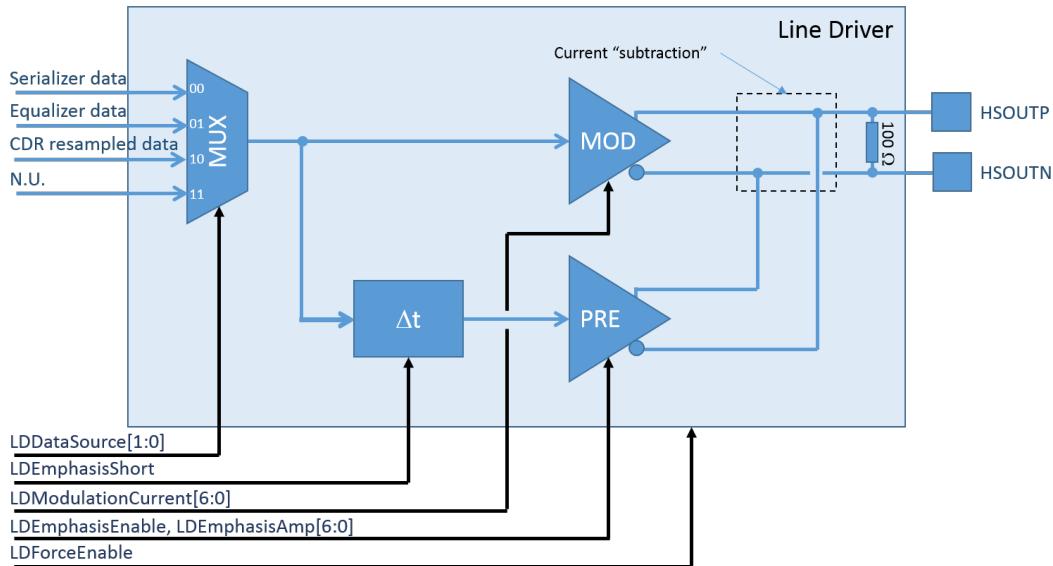


Fig. 5.1: High speed line driver block diagram

The block diagram of the high speed line driver is represented in Fig. 5.1. The purpose of this circuit is to drive the transmission line that connects the IGBT ASIC transmitter with the laser driver. As can be seen in that figure, the high speed output signal is differential and is available on pins **HSOUTP** and **HSOUTN**. The transmission line should present to the driver a 100 Ohm differential impedance that must be terminated by an AC coupled 100 Ohm termination resistor, this is schematically shown in Fig. 5.2. Please note that the AC coupling coupling capacitors together with the termination resistor form a high-pass filter. The low frequency corner of this filter is important since, depending on its value, DC wander will be generated and thus Inter Symbol Interference (ISI); 10 nF capacitors with good RF (or microwave) performance are recommended.

5.1 Line driver functionality

5.1.1 Input multiplexer

With reference to Fig. 5.1, it can be seen that the line driver includes a multiplexer allowing it to accept data from three different sources:

- **Serializer data:** This is the default input when the ASIC is working as a transceiver or a simplex transmitter. It receives the data from the serializer at either 5.12 Gbps or 10.24 Gbps depending on the ASIC operation mode;

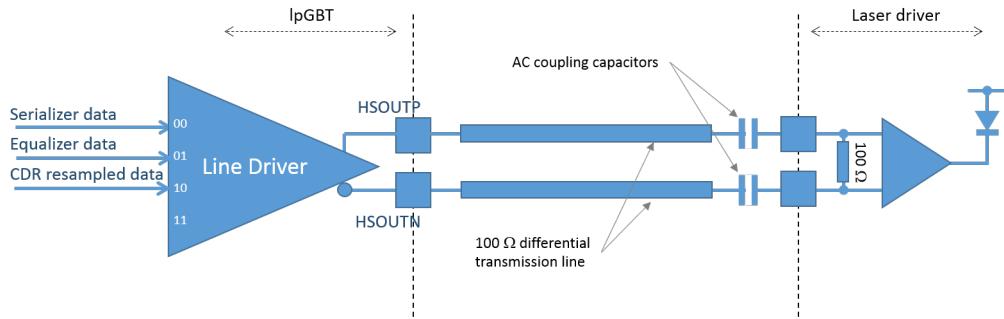


Fig. 5.2: Connecting the IpGBT with the laser driver

- **Equalizer data:** Used for testing purposes only, this multiplexer input is used to loop back the serial bit stream received by the IpGBT over the downlink and retransmit it on the uplink. The signal being observed is the output of the equalizer. Since the equalizer is tunable, from no equalization to several degrees of equalization, this allows to assess the quality of the downlink eye-diagram after and before equalization;
- **CDR resampled data:** Used for testing purposes only, similar to the 'Equalizer data' multiplexer input, it is used to loop back the serial bit stream received by the IpGBT however, in this case, the serial bit stream has been processed and re-timed by the clock and data recovery circuit.

The line driver input multiplexer is controlled by the signal **LDDatasource[1:0]**, for details please see register [\[0x119\] ULDataSource1](#) (page 235).

Depending on the operation mode of the ASIC the line driver is automatically enabled or disabled: selecting any of the Simplex TX or Transceiver modes, will automatically enable the line driver while the simplex RX modes will disable it. Additionally, the line driver is active whenever one of the loop-backs is active and the signal **LDForceEnable** set to '1', see registers: [\[0x035\] FORCEEnable](#) (page 135) and [\[0x119\] ULDataSource1](#) (page 235) for details.

5.1.2 Modulation and pre-emphasis

The line driver, as shown in Fig. 5.1, is composed of a modulator driver (**MOD**) and a pre-emphasis driver (**PRE**) working in "parallel". The modulator and pre-emphasis currents are controlled by the signals **LDModulationCurrent[6:0]** and **LDEmphasisAmp[6:0]**, respectively. Please note that, for the pre-emphasis driver to be active it is also necessary to enable it using the signal **LDEmphasisEnable**. Pre-emphasis has an additional control signal **LDEmphasisShort** that chooses between two pulse widths of the pre-emphasis pulse: 40 (short) or 60 ps. For further details on these signals please see registers [\[0x039\] LDConfigH](#) (page 137) and [\[0x03a\] LDConfigL](#) (page 137).

In the IpGBT, pre-emphasis is used to compensate for the (possible) bandwidth limitation of the transmission line connecting the line-driver and the laser-driver. The basic idea of pre-emphasis is to inject, in a band-limited transmission line, a signal with enhanced spectral contents at the frequencies most attenuated by the line. If this is judiciously done, at the end of the line (the laser-driver input) the spectral contents of the signal is such that no Inter Symbol Interference (**ISI**) will be present. Since a band-limited transmission line will attenuate mostly the high frequencies, the pre-emphasis circuit has to generate a signal with enhanced spectral contents at high frequencies. This is done by adding to the "usual" square wave current-signal (representing the bit to be transmitted) a high amplitude and narrow current pulse at the beginning of the bit period every-time there is a '0' to '1' or '1' to '0' signal transition (the sign of the current pulse depends on the signal transition direction). This technique is limited by two factors: the magnitude of the current pulses and the ability to generate current pulses that are a fraction of the bit period. The upper limit to the magnitude of the current pulses is imposed by the ASIC supply voltage (1.2 V) and the characteristic impedance of the transmission line (and its associated termination impedance, 100 Ohm). The generation of short pulses (shorter than the bit period) is limited by the semiconductor technology used that, in the case of the IpGBT, is a 65 nm CMOS technology. To circumvent that difficulty, the generation of very short pulses is avoided altogether by combining two signals delayed by the amount that corresponds to the desired pulse duration. This delay can be made arbitrarily small

by using passive circuit elements like transmission lines or "RC" delays (as is the case for the IpGBT). Combining the two signals can also be made arbitrarily fast since it can be also done passively by summing the two current signals in a circuit node (Kirchhoff's law). This is illustrated in Fig. 5.3. In that figure, the top current waveform represents a "0101" sequence (with the output current switching between I_m for a "1" and $-I_m$ for a "0"). A similar waveform is produced with inverted and scaled amplitude, as shown in the middle waveform. When the two currents are summed a wave shape similar to bottom waveform in the figure results. This current waveform has the desired characteristics: a "tall" and narrow pulse follows each transition with the current pulse returning to their normal value a fraction of the bit period latter.

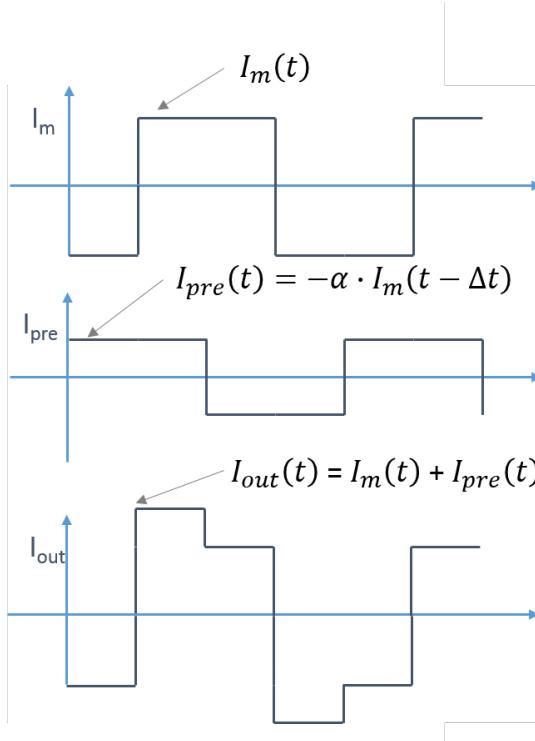


Fig. 5.3: Pre-emphasis signal generation principle

Fig. 5.1 can now be fully understood: two drivers the "MOD" and the "PRE" convert the serializer signal into a current. The "PRE" driver however works on a signal which is a delayed version of the serializer signal (with the delay being a fraction of the bit period). The two current signals from the two drivers are summed in the output node however their outputs are combined in such a way that the "PRE" driver signal current is subtracted from that of the "DRV" driver. Both drivers being differential, this is simply done by connecting the "minus" output of the "PRE" driver to the "plus" output of the "DRV" driver and, vice-versa, the "plus" output of the "PRE" driver to the "minus" output of the "DRV" driver. An "illustration" of the signal that would be obtained if the IpGBT would be connected to a "pure" (no capacitance) 100 Ohm termination is given in Fig. 5.4 (the figure shows the waveform for the two selectable delays, "short" and "long", at 5 Gbps). Notice that such waveform will never be observed in practice due to the limited bandwidth of the line. Nonetheless, when using such a signal to drive a low-bandwidth transmission line will help reduce ISI. Note the scheme is limited on the extent it can compensate for the bandwidth of the transmission line. If severe bandwidth limitation is encountered, equalization at the input of the laser driver will also be needed.

The pre-emphasis method used in the IpGBT has the potential for very high speed operation. However, as discussed above, this is at the cost of the subtraction of two current signals. This means that (when the pre-emphasis is operating) the two currents I_m and I_{pre} are permanently flowing in the output circuit (not just during the pre-emphasis pulse) increasing the power consumption of the line-driver. Moreover, the pre-emphasis is done, not by injecting additional current in the output, but by "stealing" current from the modulation current I_m during the periods when the pre-emphasis pulse is absent. In other words, it is done at the cost of reducing the modulation amplitude. It is thus

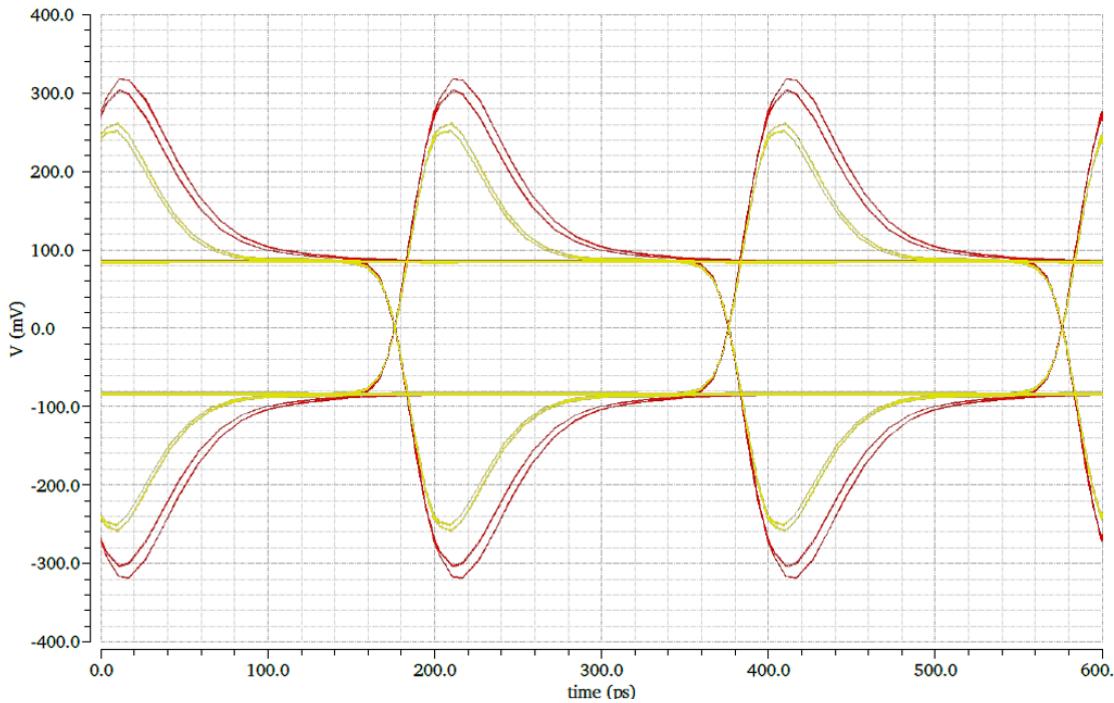


Fig. 5.4: Pre-emphasis waveform (see text for explanation)

recommended for pre-emphasis to be used when strictly necessary since it trades-off bandwidth for signal amplitude (and thus signal-to-noise ratio).

HIGH-SPEED EQUALIZER

The downlink signal (differential at 2.56 Gbps) is fed to the lpGBT through the pins **HSINP** and **HSINN**. The **Equalizer** processes this signal to restore it to the internal CML levels and/or to restore its spectral content (if needed) to minimize the amount of Inter-Symbol-Interference (ISI) and thus to reduce jitter and the Bit Error Rate before the signal is passed to the Clock and Data Recovery (CDR) circuit. A simulation example of what can be achieved by equalization is given in Fig. 6.2 where the top waveform represents the downlink eye-diagram after transmission over 75 cm of a band-limited cable and the bottom waveform the resulting eye-diagram after equalization.

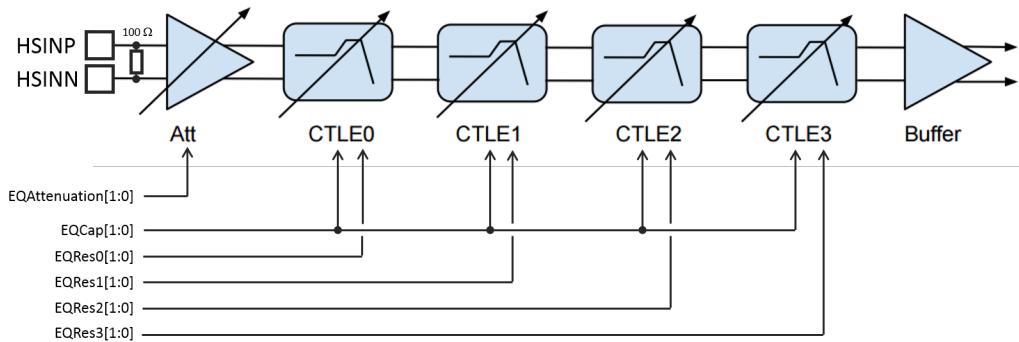


Fig. 6.1: High-speed equalizer block diagram

In most cases equalization will not be needed in systems using the lpGBT and thus the equalizer can be used to provide a flat transfer function thus acting as a simple buffer. Please note that equalization should only be used when needed and that, although it improves the bandwidth of the received signal, this is at the cost of adding noise to the signal since the frequencies where the SNR is worse (higher transmission line attenuation) are precisely the most amplified by the equalizer.

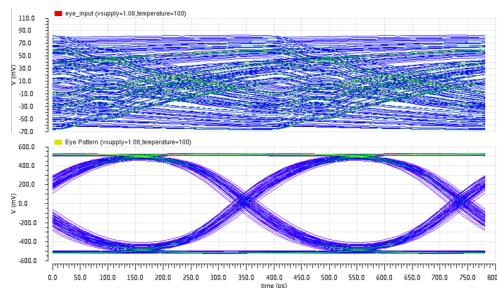


Fig. 6.2: Equalization example

The block diagram of the equalizer circuit is represented in Fig. 6.1. Following the figure from left to right, it is composed of:

- An 100 Ohm input termination;
- A programmable attenuator that allows the input circuit to handle differential signals as high as 1V, providing attenuations of 0, -3.5 and -9.5 dB;
- Four programmable Continuous Time Linear Equalizer stages (CTLE0, CTLE1, CTLE2, CTLE3) with programmable position of the zero in the transfer function (the gain of each stage also depends on the programmed zero position). The chain of four programmable equalizing stages allows to flexibly control the shape of the overall equalizer transfer function in order to compensate for the bandwidth of the transmission line preceding the IpGBT;
- A buffer to restore the equalizer signal to the internal CML levels.

To help understand how the frequency response is tuned Fig. 6.3 represents a single equalizer stage and its (ideal) transfer function. As represented, each stage transfer function contains one zero and two poles. The position of the poles is "roughly fixed" but the position of the zero can be moved by controlling the value of the resistor R_s and the capacitor C_s . Controlling the value of the capacitor moves the zero position (the higher the capacitor value the lower the frequency of the zero) and controlling the value of the resistor not only moves the position of the zero (the higher the resistor value the lower the frequency of the zero) but also changes the DC gain of the stage (the higher the resistor value the lower DC the gain of the stage).

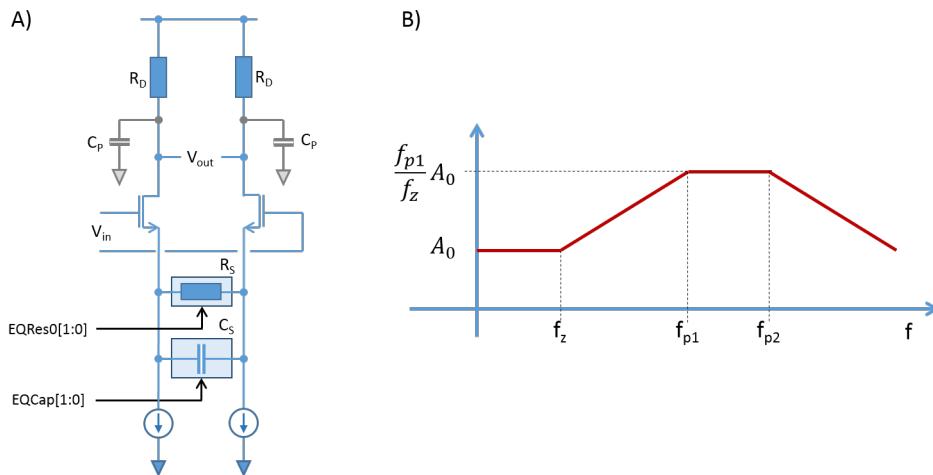


Fig. 6.3: CTLE stage and Transfer function

As depicted in Fig. 6.1, the selection of C_s is common for all the CTLE stages through the signal **EQCap[1:0]** and R_s is independently selected for each CTLE stage through the signals **EQRes0[1:0]**, **EQRes1[1:0]**, **EQRes2[1:0]** and **EQRes3[1:0]**. Please refer to registers [\[0x037\] EQConfig](#) (page 136) and [\[0x038\] EQRes](#) (page 136) for further details. Fig. 6.4 gives the zero positions for the four CTLE stages.

Choosing the right combination of zero positions among the four stages is likely to be a trial and error procedure. However, if the transfer function of the transmission line is known, the table below can be used to help synthesize its "inverse" transfer function.

Important note: The equalizer is a linear filter, as such it is important not to saturate any of the stages CTLE stages. For that the user should set the gain of the attenuator stage such that the signal amplitude at the equalizer input does not exceed 400 mV. This is only required when equalization is needed. If no equalization is necessary (flat response) the equalizer can be operated non-linearly.

To help optimizing the equalizer response, the IpGBT has several features that can be involved in the process of selecting the best combination of capacitance and resistance for each CTLE stage. In all cases, the optimization procedure will involve a parameter space search for the values of R_s and C_s . The procedures that can be used to guide the choice of the filter parameters are:

All Stages	Stage					
	1st and 2nd		3rd		4rd	
C _s [fF]	R _s [kΩ]	f _s [MHz]	R _s [kΩ]	f _s [MHz]	R _s [kΩ]	f _s [MHz]
70	3.0	758	0.6	3789	0.4	5684
	4.9	464	1.2	1895	1.0	2274
	7.0	325	2.4	947	1.6	1421
140	3.0	379	0.6	1895	0.4	2842
	4.9	232	1.2	947	1.0	1137
	7.0	162	2.4	474	1.6	711
280	3.0	189	0.6	947	0.4	1421
	4.9	116	1.2	474	1.0	568
	7.0	81	2.4	237	1.6	355

Fig. 6.4: CTLE stages zero positions

- Bit Error Rate based:** Such a procedure uses the FEC error count registers to guide the selection of the equalizer parameters. The user transmits properly encoded and formatted data to the IpGBT (see chapter [Section 4](#)). The FEC decoder will detect transmission errors and will accumulate their count in the **Down Link Data Path FEC error Correction Count {DLDPFecCorrectionCount[15:8], DLDPFecCorrectionCount[7:0]}** (see registers: [\[0x1b6\] DLDPFecCorrectionCountH](#) (page 264) and [\[0x1b7\] DLDPFecCorrectionCountL](#) (page 264)). By evaluating the error rate, that is, by periodically monitor the evolution of DLDPFecCorrectionCount[15:0], the user can decide on the effectiveness of the parameters selected and try a new set if needed. This procedure has the big advantage that the quality of the downlink channel can be monitored during normal operation, requiring only the regular read access of registers [\[0x1b6\] DLDPFecCorrectionCountH](#) (page 264) and [\[0x1b7\] DLDPFecCorrectionCountL](#) (page 264). Please note that the error correction count has to be enabled by setting **DLDPFECErrCntEna = 1'b1** in register [\[0x132\] DataPath](#) (page 243).
- Downlink Loopback based:** This procedure uses the high speed downlink loopback to monitor the quality of the downlink eye (please see [Section 14](#) and [Section 5](#) for more information on loopbacks). The downlink loopback supports retransmission of the equalizer output signal on the uplink allowing thus to monitor its quality. Again by searching the equalizer parameters it is possible to select the set that optimizes the eye opening both horizontally (jitter) and vertically (amplitude). Because the retransmission of the downlink data is made by the line driver, which is a non-linear circuit, the information on the input signal amplitude is lost. Instead, the vertically (amplitude) eye opening observed will be a combination of the signal amplitude itself and ISI (if any). An advantage of this method is that it allows almost "direct" observation of the eye-diagram quality. In particular, it allows to observe the eye-diagram before equalization ("flat" equalizer transfer function) and after equalization is applied. Its main disadvantage is that it requires instrumentation to be connected to the uplink to measure the eye-diagram. It is thus well suited for laboratory development and testing but not suitable, or difficult to implement, in the field.
- Eye Opening Monitor based:** This procedure uses the eye opening monitor circuit built in the IpGBT (see [Section 14.4](#) for detailed information on this circuit). It allows to measure both the horizontal and vertical eye opening. As for the previous case, if the equalizer is made to have a flat response, it is possible to observe the quality of the eye-diagram as received by the IpGBT prior to equalization. Given that the eye amplitude can be measured, it is important that the gain of the input attenuator is set to make the equalizer operate linearly. The main advantage of this method is that it allows to measure both the horizontal and vertical openings of the eye. The Eye Opening Monitor operation is similar to that of an "equivalent-time" scope. Since the scan of the eye-diagram is made by off-chip control and the eye-diagram reconstruction is also made off-chip this is a relatively slow and computing intensive process. Moreover, as is also the case for 1. above, the quality of the eye-diagram on the outset has to be good enough for the CDR circuit to lock to the incoming data stream.

ELECTRICAL LINKS

Warning:

Known issues: [Section 19.1.11](#), [Section 19.1.8](#), [Section 19.1.13](#), [Section 19.1.12](#).

The lpGBT can be electrically interfaced with the on detector electronics using different topologies. These depend on the ePort bandwidths as well as on the uplink data rate (5.12 or 10.24 Gbps). Depending on the configuration, the lpGBT can interface simultaneously with up to 28 frontend devices for uplink transmission and up to 16 devices for the downlink. The electrical connections between the lpGBT and the frontend devices are called **eLinks**. eLinks are used not only to transmit data between the lpGBT and the frontend devices but also for the clocks. eLinks use the CERN Low Power signalling (**CLPS**), please see [Section 7.5](#) for the further details.

More specifically, **eLinks** are used for:

- The differential clock lines (**ECLK[28:0]P / ECLK[28:0]N**): The clock lines are driven by lpGBT to the frontend modules;
- The differential downlink data outputs (**EDOUT[3:0][3:0]P / EDOUT[3:0][3:0]N** and **EDOUTECP / EDOUTECN**): The output data lines are driven by the lpGBT to the frontend devices;
- The differential uplink data inputs (**EDIN[6:0][3:0]P / EIN[6:0][3:0]N** and **EDINECP / EDINECN**): The input data lines are driven by the frontend devices to the lpGBT.

[Fig. 7.1](#) represents a generic interconnection topology between the lpGBT chip and the frontend electronics using eLinks.

To be noticed however that, since the number of input and output ePorts present in the lpGBT is different, the case depicted above, where each frontend device is served by an equal number of data input and output lines, is not always feasible. This is a consequence of the asymmetrical data bandwidth requirements of the detectors which is reflected in the lpGBT chip architecture: resulting on the eLink data rates and number of available eLinks being different for up and downlinks.

7.1 eLink Groups

The eLinks are subdivided in groups of 4 and each group contains 4 channels (or eLinks). The data rate of each group can be set independently as detailed in [Table *ePortRx \(uplink\) data rates*](#) (page 48) and [Table *ePortTx \(downlink\) data rates*](#) (page 48). As it can be seen in the table [ePortRx \(uplink\) data rates](#) (page 48), the uplink data speed depends on the lpGBT high speed link bit rate. Naturally, there is no relation between up and down eLink data rates since the High-Speed uplink and downlink bandwidths are different. The number of active eLinks within a group depends on the group data rate. For each case those tables indicates which eLink channels are active within a group. The bit shift in/out order for the eLink data inputs and outputs is MSB first.

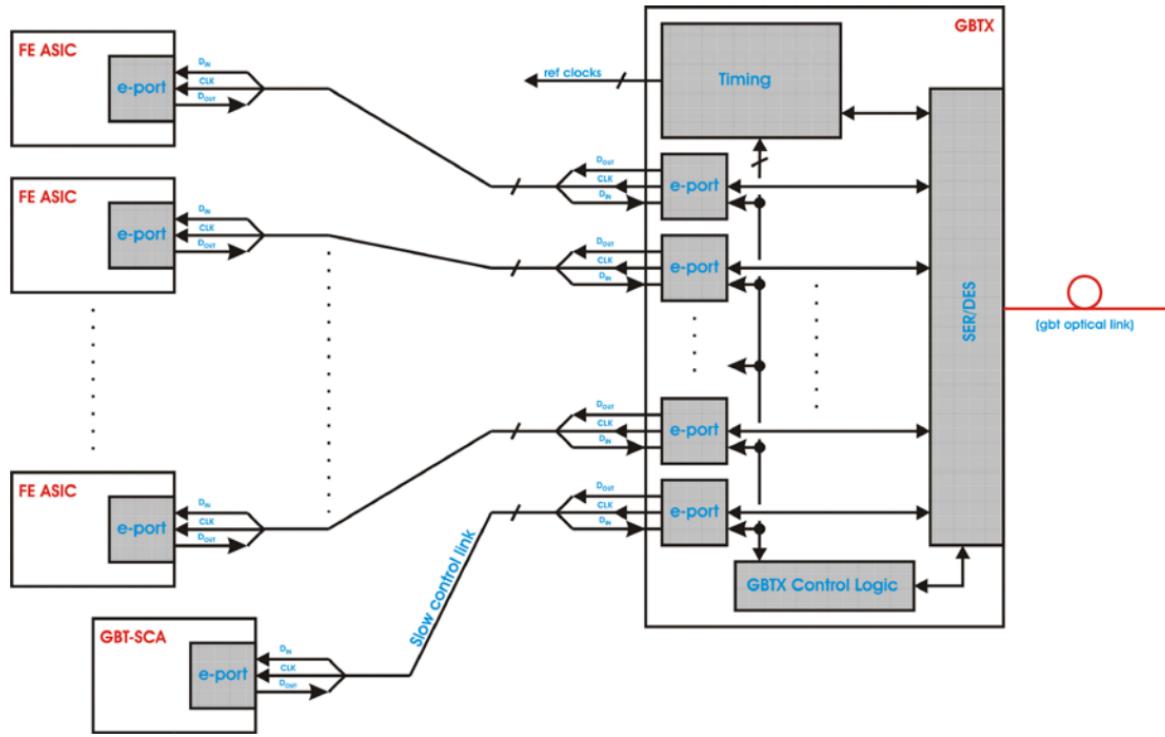


Fig. 7.1: eLink connection topology

Table 7.1: ePortRx (uplink) data rates

TxDataRate	Data Rate Select	Data Rate	Links per group	Active Channels
5.12 Gb/s	Off	0 Mb/s	0	none
5.12 Gb/s	x4	160 Mb/s	4	0, 1, 2 and 3
5.12 Gb/s	x8	320 Mb/s	2	0 and 2
5.12 Gb/s	x16	640 Mb/s	1	0
10.24 Gb/s	Off	0 Mb/s	0	none
10.24 Gb/s	x4	320 Mb/s	4	0, 1, 2 and 3
10.24 Gb/s	x8	640 Mb/s	2	0 and 2
10.24 Gb/s	x16	1.28 Gb/s	1	0

Table 7.2: ePortTx (downlink) data rates

Data Rate Select	Data Rate	Links per group	Active Channels
Off	0 Mb/s	0	none
x2	80 Mb/s	4	0, 1, 2 and 3
x4	160 Mb/s	2	0 and 2
x8	320 Mb/s	1	0

7.2 eLink pin naming conventions.

The naming conventions adopted for the eLink pins allows to easily identify to each group and "channel" a pin belongs to.

eLink inputs, **EDINGCP** (EDIN[6:0][3:0]P / EDIN[6:0][3:0]N)

- **E**: eLink;
- **D**: data;
- **IN**: uplink input;
- **G**: group number, 0 to 6 (there are 7 groups);
- **C**: channel number, 0 to 3 (there are 4 eLinks associated with every group);
- **P**: polarity, P for the positive polarity pin and N for the negative.

eLink outputs, **EDOUTCP** (EDOUT[3:0][3:0]P / EDOUT[3:0][3:0]N)

- **E**: eLink;
- **D**: data;
- **OUT**: downlink output;
- **G**: group number, 0 to 3 (there are 4 groups);
- **C**: channel number, 0 to 3 (there are 4 eLinks associated with every group);
- **P**: polarity, P for the positive polarity pin and N for the negative.

eClock **ECLKCP** (ECLK[28:0]P / ECLK[28:0]N)

- **E**: eLink;
- **CLK**: clock output;
- **C**: channel number, 0 to 28 (there are 29 eClocks)
- **P**: polarity, P for the positive polarity pin and N for the negative.

As an example, the pin EDIN32N is an eLink data input of group 3, channel 2 and it is the negative polarity pin.

7.3 eLink Tx Mirror function

The downlink eLinks implement a "mirror" function in which the data in one ePortTx channel, in a given group, is copied into one or more outputs in the same group. The mirror function can be useful to implement broadcast of data to the frontends, providing several copies of the same data in two or more outputs. It can also be used as a way to facilitate routing on the PCB allowing to have multiple choice of pins for the same data: 2 outputs at 160 Mbps and 4 at 320 Mbps. The availability of the function and how many channels can be used depends on the group data rate. The possibilities are no mirroring at 80 Mbps, two outputs per channel at 160 Mbps and four outputs per channel at 320 MHz. More specifically:

- **80 Mbps**: No mirroring;
- **160 Mbps**: Channel 3 repeats the data of channel 2 and channel 1 repeats the data of channel 0;
- **320 Mbps**: Channels 3, 2 and 1 repeat the data of channel 0;

The mirror function is controlled on a group basis by signals **EPTX[G]MirrorEnable** in register [\[0x0a8\] EPTXControl](#) (page 189) (Bits 3:0).

7.4 eLink Clocks

Besides the up and downlink data links, the IpGBT chip features up to 28 independent clock outputs. The output frequency is user programmable and it is decoupled from the up/downlink data rates. The available clock frequencies are presented in Table *ePortClock clock frequencies* (page 50). Each clock can be individually inverted (phase shifted by 180 degrees).

Table 7.3: ePortClock clock frequencies

Clk Freq Select	Clock Frequency
Off	0 MHz
x1	40 MHz
x2	80 MHz
x4	160 MHz
x8	320 MHz
x16	640 MHz
x32	1.28 GHz

Warning: Known issues: [Section 19.1.8](#), [Section 19.1.13](#).

7.5 CERN Low Power signalling (CLPS)

Interconnections between the IpGBT and the frontend devices (eLinks) is made through differential cables or transmission lines and the signalling adopted is defined by an ad-hoc "standard" called the **CERN Low Power signalling (CLPS)**. Its main characteristics are:

- Link types:
 - Point-to-point;
 - Multi-drop transmitter.
- Maximum data rate:
 - 1.28 Gbps (NRZ signalling).
- Maximum clock frequency:
 - 1.28 GHz.
- Programmable signalling level:
 - 100 mV to 400 mV (single-ended amplitude);
 - 200 mV to 800 mV (differential amplitude).
- Common mode voltage:
 - 600 mV (nominal, for 1.2 V supply voltage).
- Load impedance:
 - 100 Ohm differential.

[Fig. 7.2](#) clarifies the definitions of single and differential amplitudes. Please note that the amplitude range is specified for a 100 Ohm differential termination impedance. Reducing the termination value will reduce the transmitter signal swing while increasing it will increase the signal swing. Although the transmitter (eTx) termination can be in principle

different from 100 Ohm, the user must be aware that the termination impedance should match the impedance of the cable/transmission line being used. The IpGBT receivers (eRx) incorporate a 100 Ohm termination impedance. If the user intends to use a different line impedance he/she should disable the internal termination and provide an on PCB termination of appropriate value.

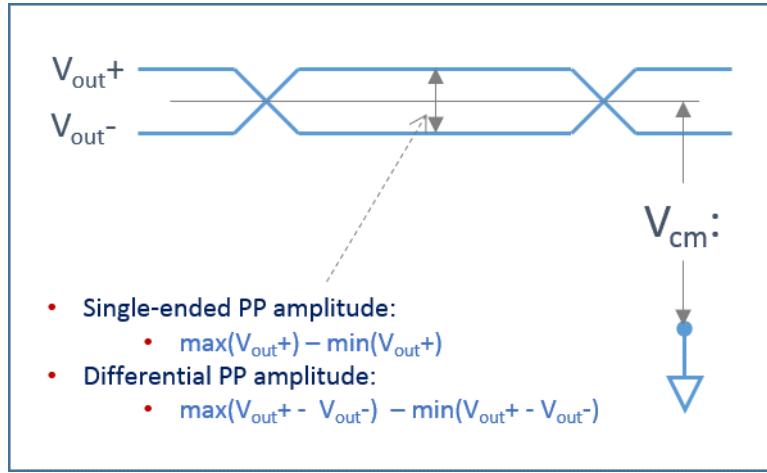


Fig. 7.2: CLPS single-ended and differential amplitude definitions

The IpGBT eLinks do not support multiple talkers driving the same line. However, for the downlinks it is possible to use a multiple drop bus configuration where multiple listeners are connected on the same line. This is illustrated in Fig. 7.3. Such configuration allows to transmit the same data from an IpGBT eLink port to several frontend devices (broadcasting). The IpGBT has no provision to address the listeners individually but it is perfectly feasible for the user protocol (transmitted over an eLink) to allow addressing of the listeners individually. If a multiple drop configuration is used the following should be observed:

- A termination impedance should be present at the end of the line;
- Only the last receiver (eRx) in the transmission line should have the termination impedance enabled.
- Star configurations are discouraged. They are however possible if a termination impedance is present at each branch. This will however reduce the signal amplitude since the equivalent impedance seen by the driver is reduced.

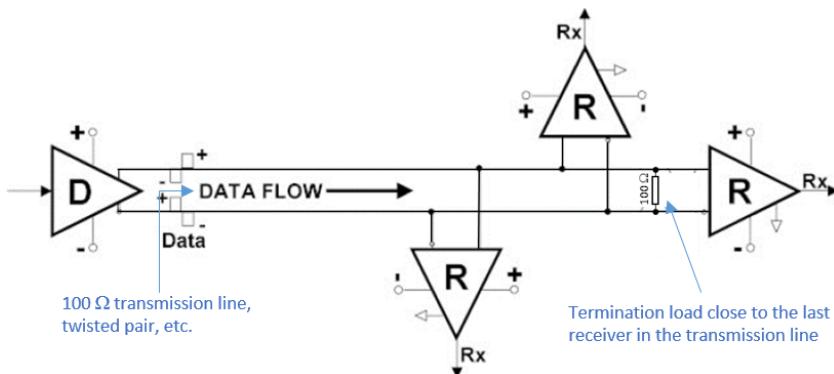


Fig. 7.3: eLink (downlink) multi-drop configuration

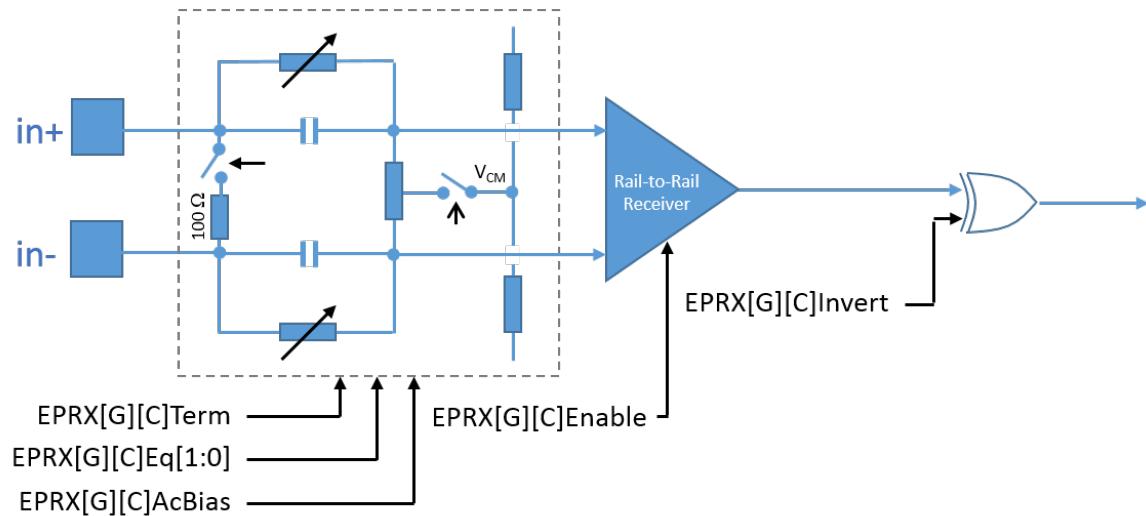


Fig. 7.4: eLink receiver

7.6 eLink Receivers (eRx)

Fig. 7.4 represents the architecture of the eLink Receiver (eRx). The eRx is designed to receive the CLPS signals described in this chapter. Its main features are:

- **Power OFF/ON** controlled by the signals **EPRX[G][C]Enable**;
- **Signal polarity: non-inverted/inverted** controlled by the signals **EPRX[G][C]Invert**;
- **Termination Disabled/Enabled** controlled by signals **EPRX[G][C]Term**;
- **AC bias OFF/ON** controlled by the signals **EPRX[G][C]AcBias**;
- **Programmable Equalization** Controlled by the signals **EPRX[G][C]Eq[1:0]**. This signals control the position of the zero and the amount of peaking produced by the eRx equalizer function as given in table *eRx equalizer zero position* (page 52);

Table 7.4: eRx equalizer zero position

EPRX[G][C]Eq[1:0]	Zero Frequency	Peaking
2'b00	No equalization	0 dB
2'b01	281 MHz	4.9 dB
2'b10	122 MHz	7.7 dB
2'b11	67 MHz	10.7 dB

The naming convention of the signals is such that [G] represents the **Group number** and [C] the **Channel number**.

Enabling of a receiver is also conditioned by the group data rate and thus the signals **EPRX[G]DataRate[1:0]**. Consequently only channels that are compatible with the data rate selected for a given group can be enabled.

All the registers controlling the eLink Receivers can be found in section *ePortRx* (page 208).

- Signals **EPRX[G][C]Enable** and **EPRX[G]DataRate[1:0]** are associated with registers [\[0x0c4\] EPRX0Control](#) (page 208) to [\[0x0ca\] EPRX6Control](#) (page 212) and, for the EC channel, [\[0x0e8\] EPRXEcChnCntr](#) (page 219). Note that the data rate is not programmable for the EC channel;
- Signals **EPRX[G][C]Invert**, **EPRX[G][C]AcBias**, **EPRX[G][C]Term** and **EPRX[G][C]Eq[1]** are associated with registers [\[0x0cc\] EPRX00ChnCntr](#) (page 212) to [\[0x0e7\] EPRX63ChnCntr](#) (page 219) and, for the EC

channel, [\[0x0e8\] EPRXEcChnCntr](#) (page 219);

- Signals **EPRX[G][C]Eq[0]** are associated with registers [\[0x0cf\] EPRX03ChnCntr](#) (page 213) to [\[0x0e7\] EPRX63ChnCntr](#) (page 219) and, for the EC channel, [\[0x0e8\] EPRXEcChnCntr](#) (page 219).

7.7 eLink Drivers (eTx)

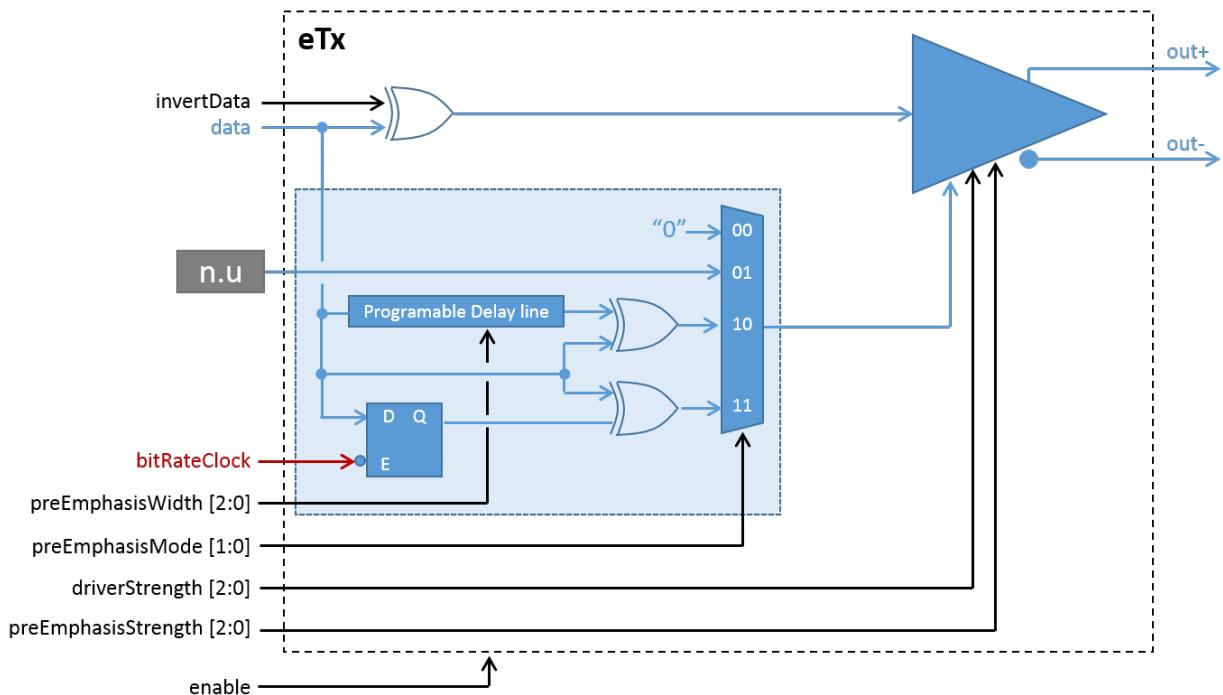


Fig. 7.5: eLink receiver

Fig. 7.5 illustrates the architecture of the IpGBT eLink driver. The same driver is used for the data outputs, up to 320 Mbps, and the clock outputs, up to 1.28 GHz. The driver has been designed to drive 100 Ohm loads with programmable strengths and controlled amounts of pre-emphasis. It has been designed to satisfy the CLPS standard previously described in this chapter. The driver offers many programmable features which include:

- **Power OFF/ON:** This feature allows to save power when a given channel is not in use even if the the group that it belongs to is active:
 - *Enabling a clock driver:* In this case the signal "enable" in Fig. 7.5 becomes: `enable = (EPCLK[G]Freq[2:0] !~ 2'b00)`. That is, every time a clock output is set to work at any of the frequencies between 40 MHz and 1.28 GHz the corresponding driver is enabled. The clock frequency is controlled by bits 2:0 of registers [\[0x06c\] EPCLK0ChnCntrH](#) (page 150) to [\[0x0a4\] EPCLK28ChnCntrH](#) (page 187);
 - *Enabling a data driver:* In this case it is possible that even if a given group is active that a specific channel within that group will not be used. Consequently it is necessary to explicitly enable the channels in use. Here, the signal "enable" in Fig. 7.5 becomes: `enable = EPTX[G][C]Enable`. These signals are controlled by registers [\[0x0a9\] EPTX10Enable](#) (page 190) and [\[0x0aa\] EPTX32Enable](#) (page 190); Notice that, if the mirror function is not enabled, for a given data rate only the channels indicated in table [ePortTx \(downlink\) data rates](#) (page 48) can be enabled. If the mirror function is enabled for a given group it is possible to enable all the channels in that group. The mirror function is controlled on a group basis by signals `EPTX[G]MirrorEnable` in register [\[0x0a8\] EPTXControl](#) (page 189) (Bits 3:0).

- **Driving current:** Programmable between 1 to 4 mA in steps of 0.5 mA. For an 100 Ohm termination this results in a differential amplitude of 200 mV to 800 mV Peak-to-Peak (signal "driverStrength[2:0]", see details below);
 - *Clock driver strength:* For clock drivers the driving strength is set by the signal "driverStrength[2:0]" in Fig. 7.5 with $\text{driverStrength}[2:0] = \text{EPCLK[C]DriveStrength}[2:0]$. These signals are controlled by bits 5:3 of registers [\[0x06c\] EPCLK0ChnCntrH](#) (page 150) to [\[0xa4\] EPCLK28ChnCntrH](#) (page 187);
 - *Data driver strength:* For data drivers the driving strength is set by the signal "driverStrength[2:0]" in Fig. 7.5 with $\text{driverStrength}[2:0] = \text{EPTX[G][0]DriveStrength}[2:0]$. These signals are controlled by bits 2:0 of registers [\[0x0ac\] EPTX00ChnCntr](#) (page 191) to [\[0xbb\] EPTX33ChnCntr](#) (page 202). For the EC channel $\text{driverStrength}[2:0] = \text{EPTXEcDriveStrength}[2:0]$, these signals are controlled by bits 2:0 of registers [\[0x0ab\] EPTXEcChnCntr](#) (page 190);
- **Programmable pre-emphasis:** To facilitate building systems that use relatively low bandwidth interconnects between the IpGBT and the frontend devices, the eTx provides a pre-emphasis function which is both programmable in driving strength and pulse width. The driving strength of the pre-emphasis pulse is programmable between 1 to 4 mA in steps of 0.5 mA (signal "preEmphasisStrength[2:0]", see details below) and its width can be programmed between 120 ps and 960 ps in steps of 120 ps or to be exactly half of the period of the selected bit rate (signals "preEmphasisMode[1:0]" and "preEmphasisWidth[2:0]", see details below).
 - *Clock driver pre-emphasis strength:* For clock drivers the driving strength is set by the signal "preEmphasisStrength[2:0]" in Fig. 7.5 with $\text{preEmphasisStrength}[2:0] = \text{EPCLK[C]PreEmphasisStrength}[2:0]$. These signals are controlled by bits 7:5 of registers [\[0x06d\] EPCLK0ChnCntrL](#) (page 150) to [\[0xa5\] EPCLK28ChnCntrL](#) (page 188).
 - *Clock driver pre-emphasis timing:* The pre-emphasis pulse width is controlled by two parameters (Fig. 7.5) the timing **Mode** and the **Pulse Length**. The "Mode" allows to select between no pre-emphasis, pre-emphasis with "clock timed" pulse duration (1/4 of the clock period) or "self timed" where the pulse width can be programmed from 120 ps up to 960 ps. Notice that the "clock timed" mode is not valid if the clock output is set to 1.28 GHz and that in the "self timed" mode the pulse width should never be programmed to exceed 1/2 clock period. The pre-emphasis mode is selected by the signals $\text{preEmphasisMode}[1:0] = \text{EPCLK[C]PreEmphasisMode}[2:0]$ that are controlled by bits 4:3 of registers [\[0x06d\] EPCLK0ChnCntrL](#) (page 150) to [\[0xa5\] EPCLK28ChnCntrL](#) (page 188). When in the "self timed" mode, the signals $\text{preEmphasisWidth}[2:0] = \text{EPCLK[C]PreEmphasisWidth}[2:0]$ set the pulse width and are controlled by bits 2:0 of registers [\[0x06d\] EPCLK0ChnCntrL](#) (page 150) to [\[0xa5\] EPCLK28ChnCntrL](#) (page 188).
 - *Data driver pre-emphasis strength:* For data drivers the driving strength is set by the signal "preEmphasisStrength[2:0]" in Fig. 7.5 with $\text{preEmphasisStrength}[2:0] = \text{EPTX[G][C]PreEmphasisStrength}[2:0]$. These signals are controlled by bits 7:5 of registers [\[0x0ac\] EPTX00ChnCntr](#) (page 191) to [\[0xbb\] EPTX33ChnCntr](#) (page 202). For the EC channel $\text{preEmphasisStrength}[2:0] = \text{EPTXEcDriveStrength}[2:0]$, these signals are controlled by bits 7:5 of registers [\[0x0ab\] EPTXEcChnCntr](#) (page 190);
 - *Data driver pre-emphasis timing:* The pre-emphasis pulse width is controlled by two parameters (Fig. 7.5) the timing **Mode** and the **Pulse Length**. The "Mode" allows to select between no pre-emphasis, pre-emphasis with "clock timed" pulse duration (1/2 of the clock period) or "self timed" where the pulse width can be programmed from 120 ps up to 960 ps. Notice in the "self timed" mode the pulse width should never be programmed to exceed the bit period. The pre-emphasis mode is selected by the signals $\text{preEmphasisMode}[1:0] = \text{EPTX[G][C]PreEmphasisMode}[1:0]$ that are controlled by bits 4:3 of registers [\[0x0ac\] EPTX00ChnCntr](#) (page 191) to [\[0xbb\] EPTX33ChnCntr](#) (page 202). When in the "self timed" mode, the signals $\text{preEmphasisWidth}[2:0] = \text{EPTX[G][C]PreEmphasisWidth}[2:0]$ set the pulse width and are controlled by bits 6:4 and 2:0 of registers [\[0x0bc\] EPTX01_00ChnCntr](#) (page 203) to [\[0xc3\] EPTX33_32ChnCntr](#) (page 208). For the EC channel the pre-emphasis mode is selected by the signals $\text{preEmphasisMode}[1:0] = \text{EPTXEcPreEmphasisMode}[1:0]$ that are controlled by bits 4:3 of register [\[0x0ab\] EPTXEcChnCntr](#) (page 190). When in the "self timed" mode, the signals $\text{preEmphasisWidth}[2:0] = \text{EPTXEcPreEmphasisWidth}[2:1]$ set the pulse width and are controlled by bits 6:4 of register :ref:REG_EPTXCONTROL.

- **Signal polarity: non-inverted/inverted:** the data or clock outputs polarity is programmable (signal "invertData", see details below).
 - *Clocks polarity* is controlled by the signals `invertData = EPCLK[C]Invert` in registers [\[0x06c\] EP-CLK0ChnCntrH](#) (page 150) to [\[0x0a4\] EPCLK28ChnCntrH](#) (page 187);
 - *Data polarity* is controlled by the signals `invertData = EPTX[G][C]Invert` in registers [\[0x0bc\] EPTX01_00ChnCntr](#) (page 203) to [\[0x0c3\] EPTX33_32ChnCntr](#) (page 208) and for the EC port `invertData = EPTXEcInvert` in register [\[0xa8\] EPTXControl](#) (page 189);

7.8 Phase alignment

Phase delays between the IpGBT and the frontend electronics will depend on the system configuration, local cable lengths and delays in the frontend circuits. It is thus necessary that the eLink input ports provide a means of adjusting the phases of the incoming data signals so that data is sampled reliably in the middle of the eye-opening. A phase adjustment/alignment mechanism is thus necessary in the IpGBT data input, and in some cases also in the ePorts of the frontend ASICs.

7.8.1 Downlink phase alignment

Two possibilities are foreseen for the eLinks in the down direction from the IpGBT to the ePort of the frontend ASICs. In one case both data and clock are simultaneously transmitted to the frontend ASIC and in the other only data is transmitted without a dedicated eLink clock. As the IpGBT is unaware of the code/frame/data structure carried by the eLinks, it does not contribute to the data down phase alignment and synchronization mechanism.

When both data (`EDOUT[G][C]P`/`EDOUT[G][C]N`) and clock (`ECLK[C]P`/`ECLK[C]N`) lines are routed to the frontend ASIC (Fig. 7.1) they must follow the same electrical route and have the same loading to assure that their relative phase is maintained at the arrival to the frontend ASIC.

In case no clock lines (`ECLK[C]P`/`ECLK[C]N`) are used and only the data lines (`EDOUT[G][C]P`/`EDOUT[G][C]N`) are routed to the frontend ASIC (to save PCB resources) the eLink clock needs to be recovered locally in the frontend ASIC with a Clock and Data Recovery (CDR) circuit. To assure that such a local CDR circuit can reliably re-generate the link clock, the data must be appropriately encoded with sufficient data transitions (and normally also DC balanced to enable AC coupling of the data signals). The IpGBT does not have built in dedicated logic for such a line encoding as it can be done in the counting room (the IpGBT is fully data transparent). Suggested line codes for this are scrambling, which incurs no bandwidth penalty and 7B/8B encoding with a code efficiency of 87.5%. The 7B/8B code has built-in comma characters that can be used for frame delimiting and synchronization. Other codes are possible but they must ensure a sufficiently high density of transitions for clock recovery.

7.8.2 Uplink phase alignment

Phase delays between the IpGBT and the frontend electronics will depend on the system configuration, local cable lengths and delays in the frontend circuits. It is thus necessary for the eLink ports to provide a means of adjusting the phases of the incoming data signals so that data is sampled reliably in the middle of the eye-opening. A phase adjustment/alignment mechanism is thus necessary in the IpGBT data inputs, and in some cases also in the ePorts of the frontend ASICs. For the uplinks the phase of the incoming data to the IpGBT is unknown. However, the data rate is known from the IpGBT and frontend modules configuration. The IpGBT clocks are synchronous with the frontend module clocks with a fixed and stable phase relationship. It is thus unnecessary to recover the clock from the data but it is necessary to phase align the incoming `EDIN[6:0][3:0]P` / `EDIN[6:0][3:0]N` data with the internal clocks in the IpGBT for each eLink. A dedicated phase-aligner circuit is responsible for this for each eLink.

The phase aligner circuit ensures that the eLink data received by the IpGBT is sampled by the IpGBT internal clock in the middle of the eye-diagram. The block diagram of the circuit is shown in Fig. 7.6.

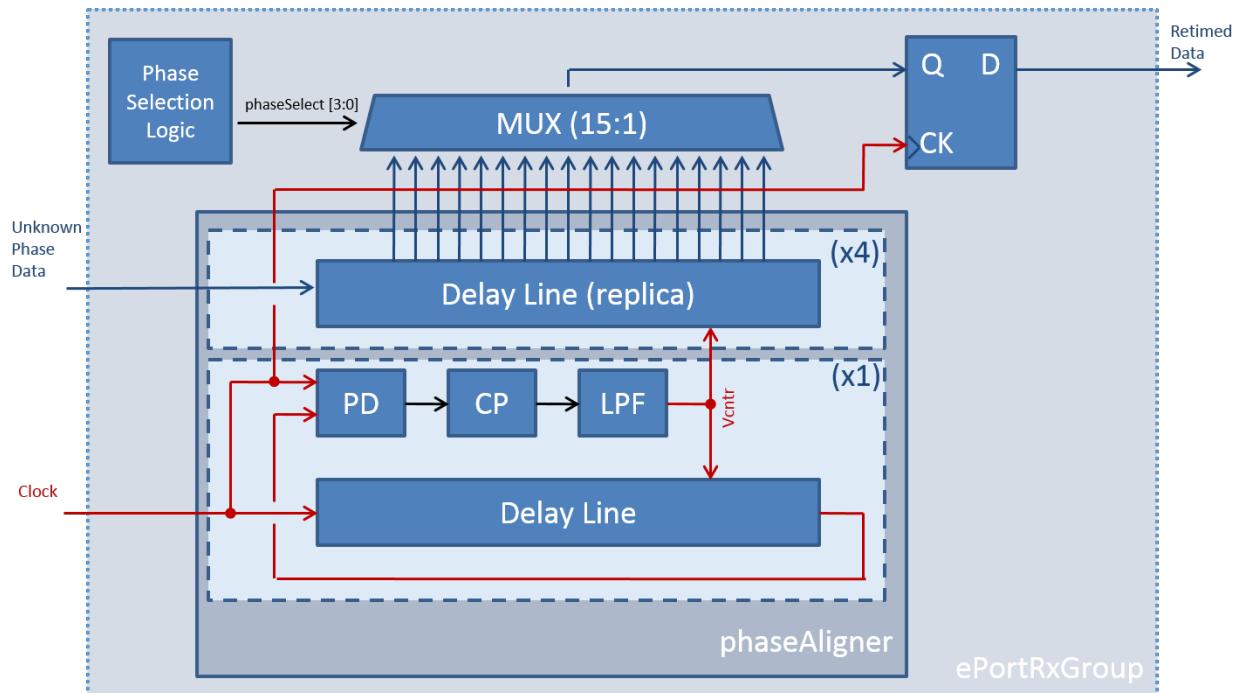


Fig. 7.6: ePortRx Phase Aligner

A phase-aligner per eGroup (ePortRx) is available with 4 phase adjustable channels as all eLinks in a group work at the same data rate, but may have different phases.

The phase aligner is composed of a master Delay Locked Loop (DLL) and four replica delay-lines with programmable phase taps (see Fig. 7.6). Each replica delay-line can adjust the phase of the incoming data via the delay taps along the delay line. The phase aligner channels can operate in four different modes: **Fixed phase**, **Initial training**, **Continuous phase tracking** and **Continuous phase tracking with initial phase**. The selection of the phase tracking mode is done on a group basis by signals EPRX[G]TrackMode[1:0]. These signals are controlled by bits 1:0 of registers [\[0x0c4\] EPRX0Control](#) (page 208) to [\[0x0ca\] EPRX6Control](#) (page 212) and [\[0x0cb\] EPRXEcControl](#) (page 212); The meaning and setup of the phase tracking modes is as follows:

- **Fixed phase:** (*Static phase selection*) In this mode, the channel phase is set by the user to a fixed value and remains static during operation (unless explicitly changed by the user). This allows the eLink ports to accept data without any DC balance restrictions. It is particularly useful when the data being transmitted to the IpGBT over the ePorts has relatively large amounts of jitter. It is the user responsibility however, to set the phase to a value that optimizes the sampling of the incoming data at the center of the eye diagram to minimize the bit error rate. The channel phase is set either at system initialization via the configuration stored in the eFuses or later via a dedicated ePort command through the I2C or IC channel or EC channel in Simplex Tx. The phase selection is made by signals EPRX[G][C]PhaseSelect[3:0] that are controlled by bits 7:4 of registers [\[0x0cc\] EPRX0ChnCntr](#) (page 212) to [\[0x0e7\] EPRX63ChnCntr](#) (page 219) and, in the case of the EC channel, by signals EPRXECPHaseSelect[3:0] controlled by bits 7:4 of register [\[0x0e8\] EPRXEcChnCntr](#) (page 219);
- **Initial training:** (*Initial training with learned static phase selection*). This phase tracking mode allows the IpGBT to determine what is the best phase selection for each channel and to switch to a static phase selection for the remaining of the operation. In some sense is very similar to the previous mode freeing the user from the task of finding the optimum phase value. Initially the user has to set the phase-aligner in the phase learning mode and ensure that DC balanced data is sent through the ePort channel being **trained**. After the phase-aligner has found the optimum phase value for the specified channel, the user sets the phase-aligner to hold the phase value. The user can then resume normal data transmission through the eLinks without concerns of DC balance on the data being transmitted. The phase settings found can be read through the I2C or IC channel or EC

channel in Simplex Tx. This mode requires however the intervention of the user to set the learning phase, to set the frontend module to send balanced data, to set the hold phase-settings operation and then to program the frontend modules to resume normal data transmission. The most likely use of this mode is during the initial-phases of system development to determine the best phase settings that can later be used regularly with the static phase selection mode.

- **Setting the training phase:** The user must explicitly set the channels in a group to be in the training phase. The signals that control phase training are EPRX[G]Train[3:0] where the n-th bit controls the n-th channel within the group and EPRXEcTrain for the EC channel. Channels not specifically enabled will be ignored during the training phase. The registers that control those signals are [\[0x105\] EPRXTrain10](#) (page 229) to [\[0x107\] EPRXTrain54](#) (page 230) and [\[0x108\] EPRXTrainEc6](#) (page 230);
- **Monitor the training process:** After enabling the training phase the user must monitor the status of the phase locking process. For that, the signals EPRX[G]ChnLocked[3:0] (with the n-th bit corresponding to the n-th channel) available in registers [\[0x142\] EPRX0Locked](#) (page 247) to [\[0x154\] EPRX6Locked](#) (page 250) (known issue: missing register for the EC channel!) must be read through one of the ASIC control interfaces. For each channel, when the phase-locking state-machine finds the optimum phase value it asserts the corresponding channel status bit to "one"; Furthermore, the state of the locking state machines for each channel can be read from the same registers.
- **Exiting the training phase:** Once all channels will be locked the user should disable the training mode by clearing bits EPRX[G]Train[3:0] and EPRXEcTrain in registers [\[0x105\] EPRXTrain10](#) (page 229) to [\[0x107\] EPRXTrain54](#) (page 230) and [\[0x108\] EPRXTrainEc6](#) (page 230); The user can read the phase values determined by the phase-locking state-machine from registers [\[0x143\] EPRX0CurrentPhase10](#) (page 247) to [\[0x156\] EPRX6CurrentPhase32](#) (page 251) and [\[0x157\] EPRXEcCurrentPhase](#) (page 251).
- **Continuous phase tracking:** (*Automatic phase tracking*) This phase tracking mode allows setting the optimum phase and dynamically adjusting it free of the user intervention. For systems where the incoming data phase is expected to slowly vary this mode allows to track the variations without introduction of bit errors during the process. The actual phase of the received data is estimated from the data bit transitions and used to dynamically adjust the phase alignment. For this to work relatively frequent data transitions are necessary. In this case a DC balanced code is desirable however it is strictly not necessary since the phase-aligner waits for sufficient data transitions in a given ePort channel before it decides to adjust the phase setting. It is possible to access the phase value selected for each channel by reading the registers [\[0x143\] EPRX0CurrentPhase10](#) (page 247) to [\[0x156\] EPRX6CurrentPhase32](#) (page 251) and [\[0x157\] EPRXEcCurrentPhase](#) (page 251); The lock state of the channels and the state of the locking state machines can be read from registers [\[0x142\] EPRX0Locked](#) (page 247) to [\[0x154\] EPRX6Locked](#) (page 250) (known issue: missing register for the EC channel!).
- **Continuous phase tracking with initial phase:** (*Automatic phase tracking with initial phase*) This phase tracking mode is virtually identical to the previous one however it allows to start the phase tracking process from a preset phase value for each channel. The motivation for this mode is the following: for channels where the phase selection has to be set at the beginning or at the end of the delay line, it is possible (due for example to jitter) that for each system initialization the phase is sometimes (randomly) selected at beginning of the line and sometimes at the end. Although this behavior is correct, due to the periodicity of the bit period, it will result as an apparent non-fixed latency for those channels. This mode avoids this ambiguity by forcing a starting phase but retains all the advantages and flexibility of the *continuous phase tracking* mode. The preset phase values needed are read from registers [\[0x105\] EPRXTrain10](#) (page 229) to [\[0x107\] EPRXTrain54](#) (page 230) and [\[0x108\] EPRXTrainEc6](#) (page 230) at beginning of operation. They are either set at system initialization via the configuration stored in the eFuses or later via a dedicated ePort command through the I2C or IC channel or EC channel in Simplex Tx. As for the *continuous phase tracking* mode, the lock state of the channels and the state of the locking state machines can be read from registers [\[0x142\] EPRX0Locked](#) (page 247) to [\[0x154\] EPRX6Locked](#) (page 250) (known issue: missing register for the EC channel!).

For all the tracking modes, except the *fixed phase*, for each ePort group the phase alignment is made in a **circular** fashion skipping channels that are not being used. It is very important that the user will disable the channels that are unused in order to save power. In the dynamic modes, the algorithm used to step the phase up or down is based on an average of 8 samples and phase-changes are done incrementally in steps of $\pm T/8$, where T is the bit period.

The phase-changes are done in such a way that no data transmission errors are introduced when that phase is being changed on the fly. The locking lock state machine counts the number of transitions that fall in the expected region. If 64 transitions are detected, then the channel is declared as locked. Conversely, the channel is considered unlocked if 64 transitions fall outside the expected range.

Warning: Known issues: [Section 19.1.11](#), [Section 19.1.12](#).

7.8.3 ePortRx group DLL programming

The operation of the phase-aligners depends on the delay lines being calibrated in relation to the bit period. This is the function of the DLL in each of the ePortRx groups. All the ePortRx DLL's share the same parameters. Programming of the phase-aligners' DLLs is made through register [\[0x034\] EPRXDllConfig](#) (page 135). In there signals:

- **EPRXDllCurrent[1:0]** (bits 7:6) control the charge-pump current;
- **EPRXDLLConfirmCount[1:0]** (bits 5:4). During DLL locking it is possible that, due to jitter, the DLL phase detector will give an inconsistent phase information. Since the starting point ("short" delay line) is forced at the beginning of operation. This count sets the minimum number of times the "late" information has to be reported by the phase-detector before the control is passed on to the DLL control loop;
- **EPRXDLLFSMClkAlwaysOn** (bit 3) This signal disables / enables clock gating of the DLL initialization state machine.
- **EPRXDLLCoarseLockDetection** (bit 2). This signal disables/enables coarse (less strict) detection of lock;
- **EPRXEnableReInit** (bit 1) disables /enables the re-initialization of an ePortRxGroup when the phase-aligner state machine finds the phase-selection to be out of range;
- **EPRXDataGatingEnable** (bit 0) disables/enables data gating.

The status of the phase-aligner DLLs' can be read from registers [\[0x158\] EPRX0DllStatus](#) (page 251) to [\[0x15e\] EPRX6DllStatus](#) (page 252). This registers report:

- The lock lock status of the **DLL EPRX[G]DllLocked** (bit 7);
- The state of lock filter state machine **EPRX[G]DllLFState[1:0]** (bits 1:0);
- The loss of Lock counter value **EPRX[G]DllLOLCnt[4:0]**.

7.8.4 Note about DC balancing and data/clock encoding for eLinks

For some applications it might be necessary to AC couple the eLink connections (e.g. serial powering of frontends). In this case a DC-Balanced code must be transmitted over each data line. If the eClocks are not used on the eLinks, Clock and Data Recovery (CDR) is required in the frontend ASIC. This encoding/decoding must take place at the optical link source in the counting room, where flexible FPGA based link interfaces are used, and in the frontend itself. The encoding overhead will depend on the type of encoding used. As the IpGBT is fully transparent to the user data being transferred it is not directly involved in any line coding being used on the local eLinks.

7.9 eClocks Wrap-up

eClocks are available for any of the operation modes of the IpGBT (Simplex RX, Simplex TX and Transceiver). To use the eClocks the user must:

- Enable the eClock drivers channels to be used: clock drivers are automatically enabled if the channel clock frequency is set to be non-zero;

- Select the channel clock frequency (see registers [\[0x06c\] EPCLK0ChnCntrH](#) (page 150) to [\[0x0a4\] EP-CLK28ChnCntrH](#) (page 187));
- Select the channel clock polarity (see registers [\[0x06c\] EPCLK0ChnCntrH](#) (page 150) to [\[0x0a4\] EP-CLK28ChnCntrH](#) (page 187));
- Select the channel driving strength (see registers [\[0x06c\] EPCLK0ChnCntrH](#) (page 150) to [\[0x0a4\] EPCLK28ChnCntrH](#) (page 187));
- Select the channel pre-emphasis mode (see registers [\[0x06d\] EPCLK0ChnCntrL](#) (page 150) to [\[0x0a5\] EP-CLK28ChnCntrL](#) (page 188));
- If pre-emphasis is used, select the channel pre-emphasis driving strength (see registers [\[0x06d\] EP-CLK0ChnCntrL](#) (page 150) to [\[0x0a5\] EPCLK28ChnCntrL](#) (page 188));
- If the self-timed pre-emphasis mode is used select channel the pre-emphasis pulse width (see registers [\[0x06d\] EPCLK0ChnCntrL](#) (page 150) to [\[0x0a5\] EPCLK28ChnCntrL](#) (page 188)).

All the above operations were detailed previously in this chapter. For setting up the "Phase programmable clocks" please refer to [Section 10](#).

7.10 Uplink eLinks (inputs) Wrap-up

For the uplink to be operational the IpGBT mode has to be set to either "Simplex TX" or "Transceiver". In these two modes the IpGBT can receive data through the input eLinks and forward it to the counting room via the HS link.

- Program the DLLs of the active input ePort groups. All the groups share the same configuration (see register [\[0x034\] EPRXDllConfig](#) (page 135)):
 - Set the DLL charge pump current;
 - Set the lock count number;
 - Disable / Enable clock gating of the DLL;
 - Disable / Enable DLL coarse lock detection.
- Set the general behavior of all active ePort groups (see register [\[0x034\] EPRXDllConfig](#) (page 135)):
 - Disable / Enable re-initialization of the ePort groups when the phase selection is detected out of range;
 - Disable / Enable data gating along the replica delay lines (gating reduces power consumption but the actual value might vary from initialization to initialization).
- Set the detailed behavior of each ePort group:
 - Disable / Enable the inactive / active ePort channels and corresponding receivers (see [\[0x0c4\] EPRX0Control](#) (page 208) to [\[0x0ca\] EPRX6Control](#) (page 212) and [\[0x0e8\] EPRXEcChnCntr](#) (page 219));
 - Set the data rate for the active groups and enable the used channels within each group (see registers [\[0x0c4\] EPRX0Control](#) (page 208) to [\[0x0ca\] EPRX6Control](#) (page 212)). Notice that the bit rate can be set independently for each group;
 - Set the group phase-aligner tracking mode (see registers [\[0x0c4\] EPRX0Control](#) (page 208) to [\[0x0ca\] EPRX6Control](#) (page 212));
- Set the ePort receivers (eRx) configuration:
 - Set the receiver signal polarity: Non-Invert / Invert (see [\[0x0cc\] EPRX00ChnCntr](#) (page 212) to [\[0x0e7\] EPRX63ChnCntr](#) (page 219), and [\[0x0e8\] EPRXEcChnCntr](#) (page 219));

- Disable / Enable the 100 Ohm termination (see [\[0x0cc\] EPRX00ChnCntr](#) (page 212) to [\[0x0e7\] EPRX63ChnCntr](#) (page 219), and [\[0x0e8\] EPRXEcChnCntr](#) (page 219));
- Disable / Enable AC biasing (see [\[0x0cc\] EPRX00ChnCntr](#) (page 212) to [\[0x0e7\] EPRX63ChnCntr](#) (page 219), and [\[0x0e8\] EPRXEcChnCntr](#) (page 219));
- Disable / Set the equalization (see [\[0x0cc\] EPRX00ChnCntr](#) (page 212) to [\[0x0e7\] EPRX63ChnCntr](#) (page 219), [\[0x0e8\] EPRXEcChnCntr](#) (page 219), [\[0x0cf\] EPRX03ChnCntr](#) (page 213) to [\[0x0e7\] EPRX63ChnCntr](#) (page 219) and [\[0x0e8\] EPRXEcChnCntr](#) (page 219));

7.11 Downlink eLinks (outputs) Wrap-up

For the downlink to be operational the IpGBT mode has to be set to either "Simplex RX" or "Transceiver". In these two modes the IpGBT can receive data from the counting room via the HS link and ship it to the frontend devices via the output eLinks.

- Set the bit rate for the active groups (see register [\[0x0a7\] EPTXDataRate](#) (page 189)). Notice that the bit rate can be set independently for each group;
- Disable / Enable the mirror function for specific (or all) groups (see register [\[0x0a8\] EPTXControl](#) (page 189));
- Enable the channels to be used (see registers [\[0x0a9\] EPTX10Enable](#) (page 190) and [\[0x0aa\] EPTX32Enable](#) (page 190));
- Set the driving strength for each driver (see registers [\[0x0ac\] EPTX00ChnCntr](#) (page 191) to [\[0x0bb\] EPTX33ChnCntr](#) (page 202) and [\[0x0ab\] EPTXEcChnCntr](#) (page 190));
- Disable / Enable the pre-emphasis for each driver (eTx). To enable the pre-emphasis:
 - Set the pre-emphasis driving strength (see registers [\[0x0ac\] EPTX00ChnCntr](#) (page 191) to [\[0x0bb\] EPTX33ChnCntr](#) (page 202) and [\[0x0ab\] EPTXEcChnCntr](#) (page 190));
 - Set the pre-emphasis timing mode (registers [\[0x0ac\] EPTX00ChnCntr](#) (page 191) to [\[0x0bb\] EPTX33ChnCntr](#) (page 202) and [\[0x0ab\] EPTXEcChnCntr](#) (page 190));
 - If the self timing mode is selected select the pre-emphasis pulse width (see registers [\[0x0bc\] EPTX01_00ChnCntr](#) (page 203) to [\[0x0c3\] EPTX33_32ChnCntr](#) (page 208) and [\[0x0a8\] EPTXControl](#) (page 189)).
- Set the driver polarity (see registers [\[0x0bc\] EPTX01_00ChnCntr](#) (page 203) to [\[0x0c3\] EPTX33_32ChnCntr](#) (page 208) and [\[0x0a8\] EPTXControl](#) (page 189)).

START-UP AND WATCHDOG

Warning: Known issues: [Section 19.1.1](#), [Section 19.1.9](#).

Features:

- Power-on reset
- Power good
- Brownout detector
- Timeout
- Watchdog
- Reset out
- I2C transaction during initialization

When the lpGBT is powered or the external reset is asserted, the chip will run an automatic configuration sequence. This is controlled by a finite-state machine (FSM), which issues resets to various blocks and monitors the state of the blocks until the complete chip is ready for operation. According to the choice of mode, the FSM will ignore unused blocks. When the chip is ready for operation, the FSM will assert the READY output.

The above mentioned FSM is in charge of coordinating several tasks: sampling of the e-fuses values, monitoring power supply level (power good, brownout detection), issuing an external reset signal or initiating I2C transaction. Once the chip is operational, the watchdog monitoring can also be used to automatically reset a particular block if it malfunctions.

8.1 Power-up state machine

The state diagram of the power-up FSM is shown in [Fig. 8.1](#).

The functions of each state are described below:

0. **ARESET** - the FSM stays in this state when power-on-reset or an external reset (RSTB) is asserted. When external signal PORdisable is asserted, the signal generated by the internal power-on-reset is ignored. All action flags are reset in this state.
1. **RESET** - synchronous reset state. In this state, the FSM produces synchronous reset signal for various circuits. All action flags are not reset in this state.
2. **WAIT_VDD_STABLE** - the FSM waits for VDD to raise. It has fixed duration of 4,000 clock cycles (~100us).

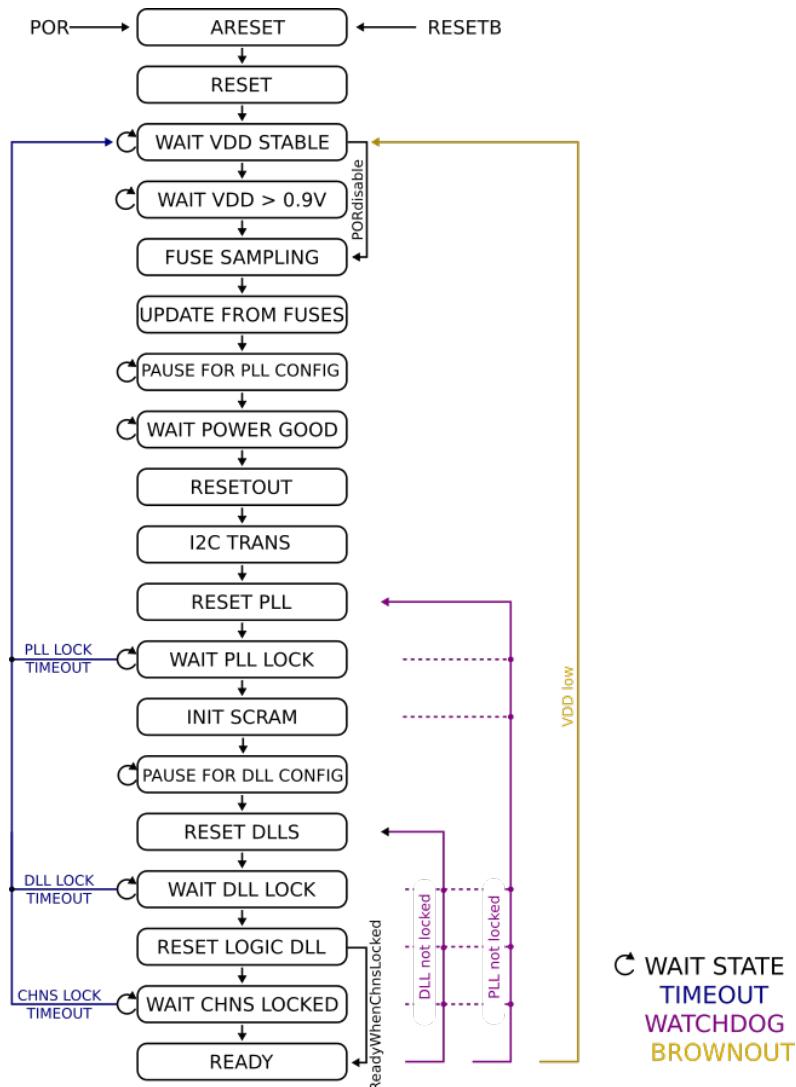


Fig. 8.1: The state diagram of power-up state machine.

3. **WAIT_VDD_HIGHER_THAN_0V90** - the FSM monitors the VDD voltage. It waits until VDD stays above 0.9V for a period longer than 1us. This state is bypassed if `PORdisable` is active.
4. **FUSE_SAMPLING** - initiate fuse sampling.
5. **UPDATE FROM FUSES** - transfer fuse values into registers. Transfer executed only if `updateEnable` fuse in `POWERUP2` register is blown.
6. **PAUSE_FOR_PLL_CONFIG** - this state is foreseen for initial testing of the chip when optimal registers settings are not yet known and the e-fuses have not been burned. The FSM will wait in this state until `pllConfigDone` bit is asserted. While in this state, the user can use the I2C interface to write values to the registers. For more details about intended use please refer to [Section 3.7](#).
7. **WAIT_POWER_GOOD** - this state is foreseen to make sure that the power supply voltage is stable before proceeding with further initialization. When `PGenable` bit is enabled the FSM will wait until VDD level stays above value configured by `PGLevel[2:0]` for longer than time configured by `PGDelay[4:0]`. If `PGenable` is not set, one can use `PGDelay[4:0]` as a fixed delay. The `PGLevel[2:0]` and `PGDelay[4:0]` are interpreted according to [Table 8.1](#) and [Table 8.2](#).

Table 8.1: Power good detection voltage levels.

<code>PGLevel[2:0]</code>	Voltage level [V]
0	0.70
1	0.75
2	0.80
3	0.85
4	0.90
5	0.95
6	1.00
7	1.05

Table 8.2: Power good wait times.

<code>PGDelay[3:0]</code>	Wait time
0	disabled
1	1 us
2	5 us
3	10 us
4	50 us
5	100 us
6	500 us
7	1 ms
8	5 ms
9	10 ms
10	20 ms
11	50 ms
12	100 ms
13	200 ms
14	500 ms
15	1 s

8. **RESETOUT** - in this state a reset signal is generated on the `resetout` pin. The reset signal is active low. The duration of the reset pulse is controlled by value of `ResetOutLength[1:0]` field according to [Table 8.3](#).

Table 8.3: Duration of reset pulse output.

ResetOutLength[1:0]	Reset pulse duration [s]
0	disabled
1	100n
2	1u
3	10u

9. **I2C_TRANS** - this state is foreseen to execute one I2C transaction. This feature can be used to configure a laser driver chip or any other component in the system. To enable transaction, the `I2CMTransEnable` bit has to be programmed and master channel has to be selected by `I2CMTransChannel[1:0]`. Remaining configuration like `I2CMTransAddressExt[2:0]`, `I2CMTransAddress[6:0]`, and `I2CMTransCtrl[127:0]` should be configured according to the description in the I2C slaves chapter.
10. **RESET_PLL** - reset PLL/CDR control logic.
11. **WAIT_PLL_LOCK** - waits for the PLL/CDR to lock. When IpGBT is configured in simplex RX or transceiver mode the lock signal comes from frame aligner. It means that the valid IpGBT frame has to be sent in the downlink. This state can be interrupted by timeout action (see the description below).
12. **INIT_SCRAM** - initializes scrambler in the uplink data path.
13. **PAUSE_FOR_DLL_CONFIG** - this state is foreseen for the case in which user wants to use serial interface (IC/EC) to configure the chip. The FSM will wait in this state until `d11ConfigDone` bit is asserted. While in this state, the user can use the serial interface (IC/EC) or I2C interface to write values to the registers. For more details about intended use please refer to [Section 3.7](#).
14. **RESET_DLLS** - reset DLLs in ePortRx groups and phase-shifter.
15. **WAIT_DLL_LOCK** - wait until all DLL report to be locked. This state can be interrupted by timeout action (see the description below).
16. **RESET_LOGIC_USING_DLL** - reset a logic using DLL circuitry. In case of ePortRx groups, this signal is used to initialize automatic phase training. This state has no impact on a phase-shifter operation.
17. **WAIT_CHNS_LOCKED** - in this state, FSM waits until automatic phase training is finished for all enabled ePortRx groups. One should keep in mind, that data transitions have to be present on the enabled channels to acquire lock. By default this state is bypassed, it can be enabled asserting `PUSMReadyWhenChnsLocked` bit in `POWERUP` register. This state can be interrupted by timeout action (see the description below).
18. **READY** - initialization is completed. Chip is operational. `READY` signal is asserted.

8.2 Power-on reset

The IpGBT has a built-in fully triplicated power-on reset circuit. It is composed of current source, integration capacitor and Schmitt trigger. The current source starts to operate once VDD exceeds ~0.7V. In typical corner, the duration of the reset pulse is around 10 μ s. The circuit can be disabled by asserting an external signal as described in [Section 8.7.2](#).

States of power-on resets circuits can be read back by accessing `PORC`, `PORB`, `PORA` fields in the `PORBOR` register.

8.3 Timeout feature

Some of the states of the FSM wait for a particular circuit to lock. In some cases, this locking may not occur (for example, if the downlink optical fibre is unplugged when the IpGBT is powered-on then the deserializer will not lock). To resolve problems like this, the wait-states have time-outs. If this time is exceeded the FSM moves back to the

WAIT_VDD_STABLE state and re-starts the full start-up procedure. In the example, the FSM will continue this time-out loop until the fibre is plugged in and the deserializer can lock. As the state machine does not go through the **RESET** state, the configuration is maintained.

The timeout is applied to the following states: **WAIT_PLL_LOCK**, **WAIT_DLL_LOCK**, **WAIT_CHNS_LOCKED**. For each state timeout feature can be enabled and its length can be controlled independently by accessing PUSMPllTimeoutConfig[3:0], PUSMDllTimeoutConfig[3:0], PUSMChannelsTimeoutConfig[3:0] fields in POWERUP registers. Timeout duration is derived according to Table 8.4.

Table 8.4: Timeout period.

Timeout[3:0]	Timeout period
0	disabled
1	10 us
2	20 us
3	50 us
4	100 us
5	200 us
6	500 us
7	1 ms
8	2 ms
9	5 ms
10	10 ms
11	20 ms
12	50 ms
13	100 ms
14	500 ms
15	1 s

When the IpGBT is already operating (FSM is in the **READY** state) and a problem occurs that lasts a long time (for example, the downlink fibre is unplugged) the combination of the watchdog and timeout features will ensure that the IpGBT automatically recovers when the problem is resolved (the reconnection of the fibre).

The occurrence of the timeout event is flagged on PUSMPllAction, PUSMDllAction, PUSMChannelsAction bits in PUSMActions register for **WAIT_PLL_LOCK**, **WAIT_DLL_LOCK**, **WAIT_CHNS_LOCKED** states respectively. The PUSMActions register is reset only by asynchronous reset originating from power-on reset block or external RSTB pin.

8.4 Watchdog operation

When enabled, this will monitor the state of each sub-block. If any sub-block stops operating correctly (for example a PLL loses lock), the watchdog will force the FSM to return to the reset state of that sub-block and the sequence will continue from that point until normal operating conditions are achieved.

For example, if the FSM is in the **READY** state and the PLL/CDR loses lock, the FSM will jump back to the **RESET_PLL** state and re-start the sequence from there. The intended use of the watchdog is to allow the chip to automatically recover from a functional interruption (such as a fibre disconnect) without the need of power-cycling or resetting. It does however have an impact on the number of data errors caused by a single-event-upset, as discussed below.

State transitions triggered by the watchdog are shown in the figure below. The watchdog can be disabled by asserting the PUSMPllWdogDisable and PUSMDllWdogDisable bits in POWERUP register for PLL and DLL locks accordingly.

The occurrence of the watchdog event is flagged on `PUSMPLLwatchdogAction`, `PUSMDLLwatchdogAction` bits in `PUSMActions` register for PLL, DLL watchdog actions respectively. The `PUSMActions` register is reset only by asynchronous reset originating from power-on reset block or external RSTB pin.

8.4.1 Notes on using the watchdog in an SEU environment

The IpGBT has been extensively tested with heavy ions provoking single-event upsets in the chip. In particular, the sensitivity of the serializer/deserializer circuits to SEUs and the resulting data errors have been measured. With the watchdog disabled, it has been observed that the chip always recovers to its functional state after an SEU. A certain number of bit and frame errors occur while the chip is recovering. With the watchdog enabled, the chip again recovers back to its functional state but the recovery process is longer and hence more bit and frame errors are observed. The user should therefore carefully assess the advantages and disadvantages of enabling the watchdog if they will operate in a significant radiation environment.

8.5 Brownout detection

Warning: Known issues: [Section 19.1.9](#).

The IpGBT chip features brownout detection circuit. This circuit can be activated by asserting `BODEnable` bit in the `RESETConfig` register. Level at which the detector activates can be controlled by `BODlevel[2:0]` field in the `RESETConfig` register according to [Table 8.5](#).

Table 8.5: Brownout detection voltage levels.

<code>BODlevel[2:0]</code>	Brownout voltage level [V]
0	0.70
1	0.75
2	0.80
3	0.85
4	0.90
5	0.95
6	1.00
7	1.05

When the brownout event is detected, the FSM will jump to **RESET** state and the complete chip initialization will be performed. Moreover, the occurrence of the brownout event is flagged on `PUSMbrownoutAction` bit in `PUSMActions` register. The `PUSMActions` register is reset only by asynchronous reset originating from power-on reset block or external RSTB pin.

States of brownout detector circuits can be read back by accessing `BORC`, `BORB`, `BORA` fields in the `PORBOR` register.

8.6 Disabling the power-up sequence

The FSM can be halted at any time by either asserting the `stateOverride` external signal or asserting `PUSMForceState` bit in the [\[0x12f\] POWERUP3](#) (page 242) register. The state is then selected by the value written to a `PUSMStateForced[3:0]` configuration field in `POWERUP3` register.

The register-based halting of power the FSM is disabled by default, in order to unlock user has to write `0xA3` value (magic number) to the [\[0x130\] POWERUP4](#) (page 243) register.

8.7 Configuration pins

The behavior of the power up state machine is also affected by external configuration pins.

8.7.1 STATEOVRD

STATEOVRD (state override) pin disables the automatic power-up state machine. The state of the power up state machine is then selected by the value written to a configuration PUSMStateForced field in the [\[0x12f\] POWERUP3](#) (page 242) register. The STATEOVRD pin has an internal pull down resistor.

Table 8.6: STATEOVRD pin function.

STATEOVRD	Description
1'b0	Normal operation
1'b1	Power Up state machine halted

8.7.2 PORDIS

PORDIS disables build in power on reset (POR) circuit. Moreover, when asserted, the **WAIT_VDD_HIGHER_THAN_0V90** state will be skipped by the power up state machine. When asserted the reset signal should be provided by user on the RSTB pin. The PORDIS pin has an internal pull down resistor.

Table 8.7: PORDIS pin function.

PORDIS	Description
1'b0	Normal operation
1'b1	Power on reset block disabled. WAIT_VDD_HIGHER_THAN_0V90 state skipped

8.7.3 RSTB

Warning: Known issues: [Section 19.1.1](#).

RSTB is active low reset signal. Asserting this signal starts the chip initialization procedure. The RST pin has an internal pull up resistor.

8.7.4 READY

READY pin is an output which signals that the initialization of the IpGBT chip has finished. This pin is high when the power-up state machine (described in [Section 8.1](#)) is in state **READY**.

8.7.5 RSTOUTB

RSTOUTB pin is an output which delivers an active low reset pulse. The pulse is generated by the power-up state machine in **RESETOUT** state. The duration of the pulse can be configured by ResetOutLength[1:0] field in [\[0x03d\] RESETConfig](#) (page 138) register. The driving strength of this pin is controlled by ResetOutDriveStrength bit in [\[0x03d\] RESETConfig](#) (page 138) register.

Note: Registers controlling RSTOUTB feature are triplicated and therefore no Single Event Upsets (SEU) are expected on this output. However, due to the fact that the output driver itself is not triplicated, occasional Single Event Transients (SET) cannot be eliminated.

CLOCK GENERATOR BLOCK

Warning: Known issues: [Section 19.1.10](#).

The Clock Generator Block is comprised of a CDR/PLL (lJCDR), fast CMS clock divider (1/2) and a CMOS clock divider (1/64) as shown in [Fig. 9.1](#). This block is responsible for the clock generation for the full chip (from VCO's 5.12 GHz clock down to 40 MHz clock).

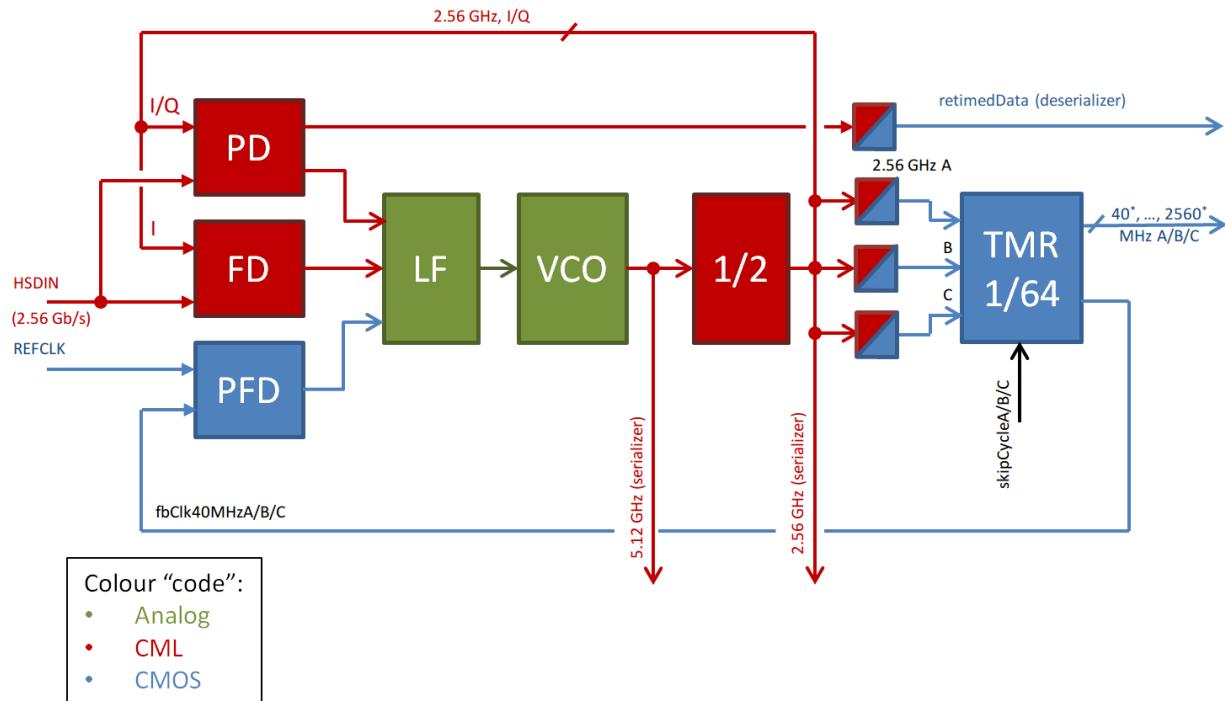


Fig. 9.1: Clock Generator Block top view

The lJGBT features a typical continuous-time charge pump PLL/CDR with an LC-tank VCO. The Phase Detector (PD) and Frequency Detector (FD) blocks are used when the lJGBT recovers the clock from data (simplex receiver or transceiver, for more details refer to [Section 1.3](#)) When the lJGBT operates in the simplex transmitter mode, the PLL operates as a frequency multiplier and the Phase and Frequency Detector (PFD) is used. In this mode the reference frequency has to be provided on *REFCLKP* and *REFCLKN* pins.

9.1 Initialization

The LC-tank VCO has low internal phase noise but due to its low gain the locking range is limited to a few MHz. In order to increase the locking range, extra discrete capacitors were added to the LC-tank VCO that will shift the locking range interval up or down in frequency. In order to find the correct VCO frequency band for the reference frequency, the IjCDR state machine will perform a binary search during calibration by comparing the reference frequency with the VCO frequency. A step that is necessary in both operation. The Fig. 9.2 depicts the state diagram of the initialization state machine.

The state machine reset is issued by the PUSM (see Section 8.1). During the reset and calibration state the VCO's control voltage is forced at the middle of the control voltage range. The VCO will start oscillating, providing the necessary clocks to the chip.

Once the reset is released by the PUSM, the VCO capacitor search begins. The reference clock is compared with the VCO clock by racing two counters - one clocked by the reference clock and another by the VCO clock. Once the race is over, if the VCO is faster, the state machine will increase the capacitor value, effectively decreasing the VCO's oscillating frequency. If it is slower, the capacitor value is decreased. Once the binary search is finished, the state machine will release the VCO's control voltage and close the loop. At this point the VCO's frequency should be only few kHz away from the reference frequency ensuring a smooth locking. Once the calibration procedure is finished, the selected value can be monitored by reading field `CLKG_vcoCapSelect [8 : 0]` in [\[0x1b1\] CLKGStatus5](#) (page 262) and [\[0x1b2\] CLKGStatus6](#) (page 262) registers. The depth of the two 16 bit calibration counters can be selected by the `CLKGCalibrationEndOfCount [3 : 0]` field in the [\[0x020\] CLKGConfig0](#) (page 131) register. Its value selects the VCO calibration race goal in number of clock cycles ($2^{(CLKGCalibrationEndOfCount[3:0]+1)}$).

The state of the IjCDR state machine can be monitored by reading the value of `CLKG_smState [3 : 0]` field in the [\[0x1b5\] CLKGStatus9](#) (page 263) register and decoded according to Table 9.1. If the state machine is in one of the locked states, the `CLKG_smLocked` bit is set in the [\[0x1b4\] CLKGStatus8](#) (page 263) register.

Table 9.1: IjCDR's state machine states

<code>CLKG_smState[3:0]</code>	Value	Description
<code>smResetState</code>	4'h0	reset state
<code>smInit</code>	4'h1	initialization state (1cycle)
<code>smCapSearchStart</code>	4'h2	start VCO calibration (jump to smPLLInit or smCDRInit when finished)
<code>smCapSearchClearCounters0</code>	4'h3	VCO calibration step; clear counters
<code>smCapSearchClearCounters1</code>	4'h4	VCO calibration step; clear counters
<code>smCapSearchEnableCounter</code>	4'h5	VCO calibration step; start counters
<code>smCapSearchWaitFreqDecision</code>	4'h6	VCO calibration step; wait for race end
<code>smCapSearchVCOFaster</code>	4'h7	VCO calibration step; VCO is faster than refClk, increase capBank
<code>smCapSearchRefClkFaster</code>	4'h8	VCO calibration step; refClk is faster than VCO, decrease capBank
<code>smPLLInit</code>	4'h9	PLL step; closing PLL loop and waiting for lock state
<code>smCDRInit</code>	4'hA	CDR step; closing CDR loop and waiting for lock state
<code>smPLLEnd</code>	4'hB	PLL step; PLL is locked
<code>smCDREnd</code>	4'hC	CDR step; CDR is locked

9.1.1 VCO calibration in PLL mode

The 40 MHz reference clock (pins *REFCLKP* and *REFCLKN*) is directly compared with the VCO's frequency divided by 128. In this mode pin *LOCKMODE* has to be low (see Section 9.4.3).

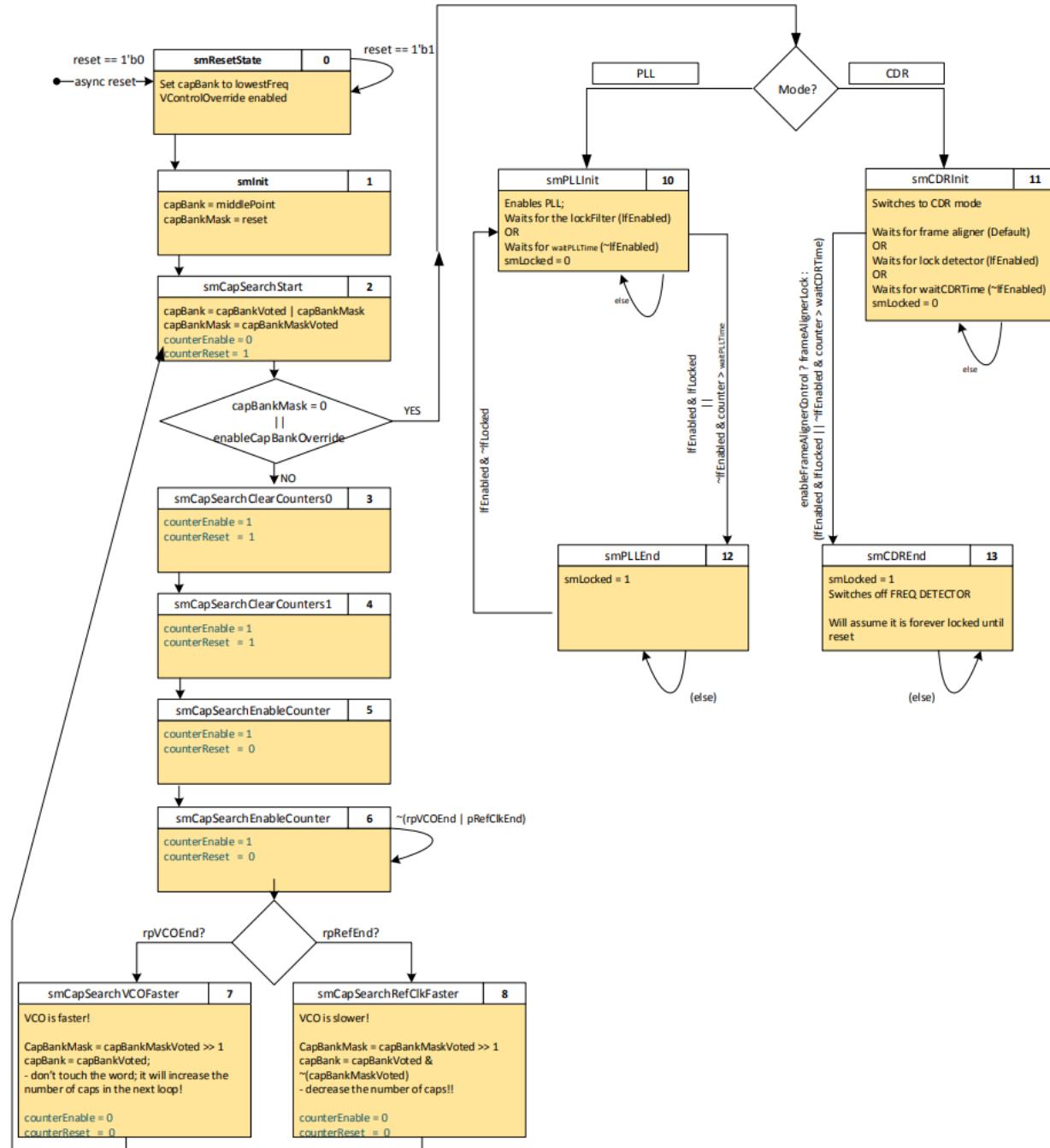


Fig. 9.2: IjCDR initialization state machine

9.1.2 VCO calibration in CDR mode

There are two possible modes of calibrating the VCO when in CDR mode, either reference-less mode or reference mode. This is controlled by the external pin *LOCKMODE* (see [Section 9.4.3](#)) with the possibility of overriding this (see [Section 9.6](#)). When in RX/TRX mode the reference-less mode is to be used by default (reference mode is only meant to be used during debug).

For reference-less mode, in order to avoid adding an external crystal (like in GBTX), the incoming 2.56 Gbps data stream is divided by 16 by means of a ripple counter. The output of this counter will be a jittery 40 MHz clock which effectively produces a sub-harmonic of the data content. This only works due to the fact that the data is non return to zero and DC balanced [reference-less] (page 325).

In reference mode the procedure is the same as in PLL mode, where it is necessary to provide a 40 MHz clock at the *REFCLKP* and *REFCLKN* pins. The reference clock is used to calibrate the VCO, instead of the data stream.

9.2 Lock detection

The locking mechanism depends on the mode of operation. For the PLL mode the default mode is the use of the Lock Filter, and for CDR mode the default is the use of the Frame Aligner (see [Section 4.1.6](#)). As a backup solution, both modes can be overridden to use a simple "wait" counter, where after a predefined number of cycles, the state machine will report locked state (this is however not recommended since it is intended for testing and debugging).

9.2.1 PLL mode - lock filter lock

The lock filter job is to filter the instant lock signal reported by the Phase and Frequency Detector. The lock filter uses a parameterizable state machine, depicted in Fig. 9.3.

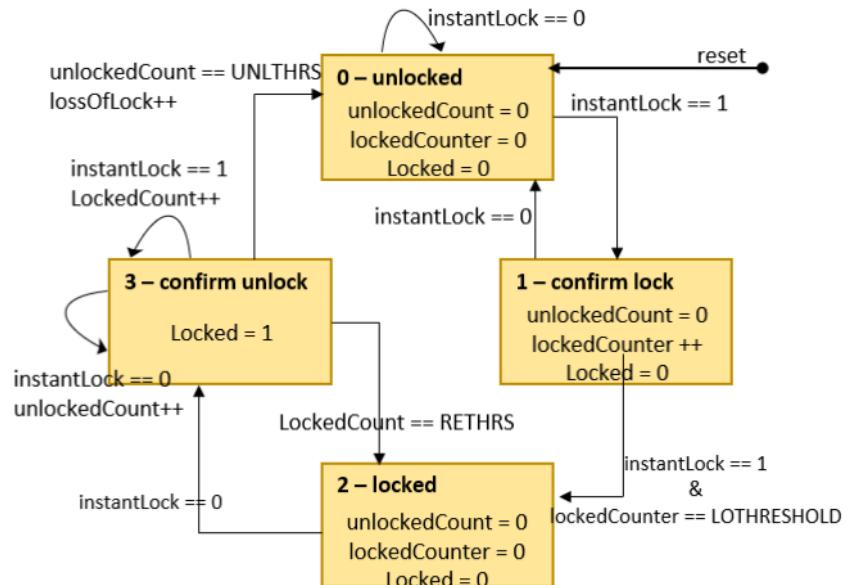


Fig. 9.3: Lock filter state machine

The lock filter can be enabled by CLKGLockFilterEnable bit in the [0x02d] *CLKGLFConfig0* (page 133) register. If the lock filter is disabled, the wait counter is used instead and the duration of the wait period can be set by CLKGwaitPLLTime[3:0] field in the [0x02c] *CLKGWaitTime* (page 133) register to $2^{CLKGwaitPLLTIme}$ clock cycles.

When the lock filter is enabled, the lock threshold value of the instant lock low pass filter can be set by CLKGLockFilterLockThrCounter[3:0] field. The number of clock cycles is set to $2^{CLKGlockFilterLockThrCounter}$. Similarly, the relock and unlock threshold values are set by CLKGLockFilterReLockThrCounter[3:0] and CLKGLockFilterUnLockThrCounter[3:0] fields respectively. Both fields can be found in the [0x02e] *CLKGLFConfig1* (page 134) register.

The current status of the lock filter can be monitored by reading value of the CLKG_lfState[1:0] field from the [0x1b5] *CLKGStatus9* (page 263) register and decoded according to the Table 9.2.

Table 9.2: Lock Filter States

CLKG_lfState[1:0]	Value	Description
lfUnlLockedState	2'b00	low-pass lock filter is unlocked
lfConfirmLockState	2'b01	low-pass lock filter is confirming lock
lfLockedState	2'b10	low-pass lock filter is locked
lfConfirmUnlockState	2'b11	low-pass lock filter is confirming unlock

Moreover, the lock filter counts the number of times when the PLL losses lock (state machine goes from lfConfirmUnlockState to lfUnlLockedState state). This information is available in the CLKG_lfLossOfLockCount[7:0] field in the [0x1af] *CLKGStatus3* (page 262) register. The state of the locked signal and instant lock signal can be monitored by reading CLKG_lfLocked and CLKG_lfInstLock bits (in the [0x1b4] *CLKGStatus8* (page 263) register) accordingly.

9.2.2 CDR mode - frame aligner lock

In the CDR mode, the lock signal comes from the frame aligner (see Section 4.1.6). This behavior can be altered by asserting CLKGDisableFrameAlignerLockControl bit in the [0x021] *CLKGConfig1* (page 131) register. If the frame aligner lock is disabled (bit CLKGDisableFrameAlignerLockControl set to 1'b1) and CLKGLockFilterEnable to 1'b0 the wait counter will be in use. In this case the state machine will wait $2^{CLKGwaitCDRTIme}$ clock cycles until declaring the lock. The CLKGwaitCDRTIme[3:0] field is located in the [0x02e] *CLKGLFConfig1* (page 134) register.

9.3 Loop control

9.3.1 PLL loop control

The PLL loop features a Phase and Frequency Detector which will drive two charge-pumps (integral and proportional path). The user has the possibility to choose different loop configurations during PLL locking and once the PLL is locked. This allows the use of a higher loop bandwidth during locking but a smaller loop bandwidth when locked (providing faster lock time and decreasing jitter).

The filter resistance is set by CLKGPllRes[3:0] and CLKGPllResWhenLocked[3:0] fields (in the [0x022] *CLKGPllRes* (page 131) register) for the locking and locked phases respectively. The value of the resistance is $39.9k/CLKGPllRes$. The selected value of the filter resistance can be monitored by reading CLKG_PLL_R_CONFIG[3:0] field in the [0x1ac] *CLKGStatus0* (page 262) register.

The integral charge pump current is set by CLKGPllIntCur[3:0] and CLKGPllIntCurWhenLocked[3:0] fields (in the [0x023] *CLKGPllIntCur* (page 132) register) for the locking and locked phases respectively. The

current value ranges from 0 to 8 μ A. The selected value of the integral charge pump current can be monitored by reading CLKG_CONFIG_I_PLL[3:0] field in the [0x1ac] *CLKGStatus0* (page 262) register.

The proportional charge pump current is set by CLKGPLLPropCur[3:0] and CLKGPLLPropCurWhenLocked[3:0] fields (in the [0x024] *CLKGPLLPropCur* (page 132) register) for the locking and locked phases respectively. The current value ranges from 0 to 82 μ A. The selected value of the proportional charge pump current can be monitored by reading CLKG_CONFIG_P_PLL[3:0] field in the [0x1b0] *CLKGStatus4* (page 262) register.

9.3.2 CDR loop control

The CDR architecture is based on a non-linear bang-bang phase detector PLL that provides early-late corrections to the loop through a proportional and integral path as is shown in Fig. 9.4 [cdr] (page 325).

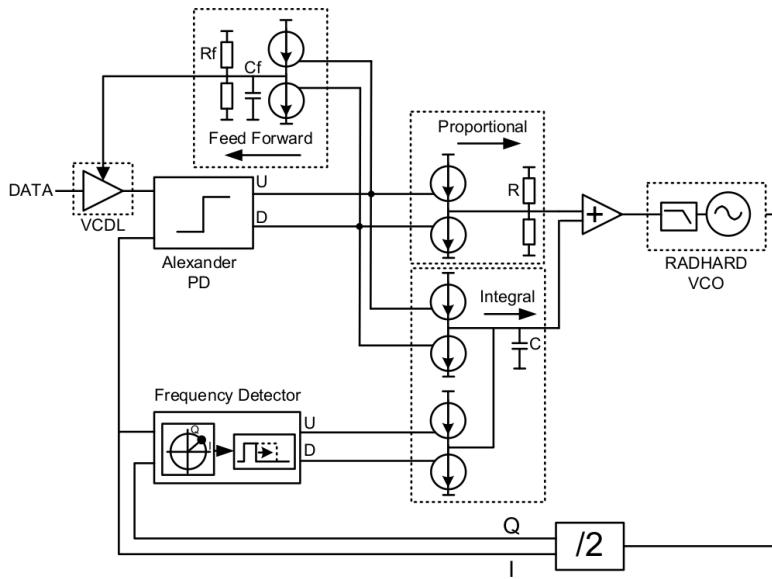


Fig. 9.4: Architecture of the CDR circuit

The proportional and integral path of the CDR is split to reduce the capacitor area of the loop filter while respecting the noise requirements of the loop resistor. The proportional path corrects the CDR instantaneously while the integral path tracks slowly varying offsets. The charge pump current for the proportional path can be set by CLKGCDRPropCur[3:0] and CLKGCDRPropCurWhenLocked[3:0] fields (in the [0x025] *CLKGCDRPropCur* (page 132) register) for the locking and locked phases respectively. The current can vary from 0 to 82 μ A. The currently selected value of the proportional charge pump current can be monitored by reading CLKG_CONFIG_P_CDR[3:0] field in the [0x1ae] *CLKGStatus2* (page 262) register.

The charge pump current for the integral path can be set by CLKGCDRIntCur[3:0] and CLKGCDRIntCurWhenLocked[3:0] fields (in the [0x026] *CLKGCDRIntCur* (page 132) register) for the locking and locked phases respectively. The current can vary from 0 to 82 μ A. The currently selected value of the integral charge pump current can be monitored by reading CLKG_CONFIG_I_CDR[3:0] field in the [0x1ad] *CLKGStatus1* (page 262) register.

An additional frequency detector is added to extend the pull-in range of the CDR during acquisition. This Frequency Detector (FD) is based on a 4-quadrant sampling method and is extended with a pulse time-amplifier to limit the required charge-pump current of the FD. The VCO runs at 5.12 GHz to create I and Q clocks at 2.56 GHz for the CDR clock recovery. The charge pump current for the frequency detector can be set by CLKGFLLIntCur[3:0] and CLKGFLLIntCurWhenLocked[3:0] fields (in the [0x028] *CLKGFLLIntCur* (page 132) register) for the locking

and locked phases respectively. The current can vary from 0 to 82 μ A. The currently selected value of the charge pump for the frequency detector current can be monitored by reading CLKG_CONFIG_I_FLL[3:0] field in the [\[0x1ad\] CLKGStatus1](#) (page 262) register.

An additional feed forward path is added. The core of this path is an additional charge-pump which acts as a integrator and controls a voltage controlled delay cell (VCDC) at the data input and compensates for fast adjustments that would be required by the proportional path but are filtered by the VCO. A VCO acts as an integrator and integrates the phase detector's control signals to phase. The fast feed forward compensation performs this integration internally and adds an additional delay using a voltage-controlled delay line to convert the integrated voltage to phase. To overcome the saturation, the integrator is made lossy with an average voltage equal to half the supply voltage. The charge pump current for the feed forward path can be set by CLKGCDRFeedForwardPropCur[3:0] and CLKGCDRFeedForwardPropCurWhenLocked[3:0] fields (in the [\[0x027\] CLKGCDRFFPropCur](#) (page 132) register) for the locking and locked phases respectively. The current can vary from 0 to 82 μ A. The currently selected value of the charge pump current for the feed forward current can be monitored by reading CLKG_CONFIG_P_FF_CDR[3:0] field in the [\[0x1ae\] CLKGStatus2](#) (page 262) register. The value of the feed forward capacitance (Cf) can be set by CLKGFeedForwardCap[2:0] and CLKGFeedForwardCapWhenLocked[2:0] fields (in the [\[0x029\] CLKGFFCAP](#) (page 132) register) for the locking and locked phases respectively. The capacitance value varies from 0 to 308 fF with step of 44 fF. The currently selected value of the feed forward capacitance can be monitored by reading CLKG_CONFIG_FF_CAP[3:0] field in the [\[0x1b4\] CLKGStatus8](#) (page 263) register.

Warning: Known issues: [Section 19.1.10](#).

9.4 Configuration and clock pins

9.4.1 REFCLKP and REFCLKN

REFCLKP and REFCLKN pins form a differential clock which is used as reference clock for the PLL. If the IpGBT operates in *simplex receiver* or *transceiver* mode (see [Section 1.3](#)) and the reference-less locking mode is used (see [Section 9.4.3](#)) these pins can be left unconnected.

9.4.2 TSTCLKINP and TSTCLKINN

TSTCLKINP and TSTCLKINN pins form a differential clock which can be used as a clock source instead of the VCO. This feature is implemented for testing purposes only and should not be used by users. If the [VCOBYPASS](#) (page 76) pins is set to zero (normal operation) TSTCLKINP and TSTCLKINN pins can be left unconnected.

9.4.3 LOCKMODE

LOCKMODE pin controls the lock mechanism for the CDR. For more details please refer to [Section 9.1.1](#). The LOCKMODE pin has an internal pull down resistor.

Table 9.3: LOCKMODE pin function.

LOCKMODE	Description
1'b0	Use external 40 MHz reference clock.
1'b1	Reference-less locking. Recover frequency from the data stream

9.5 VCOBYPASS

It allows to bypass PLL by providing 5.12 GHz clock to TSTCLKINP and TSTCLKINN pins. VCOBYPASS pin has an internal pull down.

Table 9.4: VCOBYPASS pin function.

VCOBYPASS	Description
1'b0	Normal operation. System clocks comes from PLL/CDR
1'b1	VCO Bypass mode. System clock come from TSTCLKINP / N (5.12 GHz)

9.6 Override mode and test outputs

It is possible to force the capacitor bank and to override the state machine. This is only meant to be used during debug and both overrides are independent.

VCO Capacitor Bank override

To force the VCO capacitor bank, it is necessary to enable the override switch for the capacitor bank override

- [\[0x029\] CLKGFFCAP \(page 132\)](#)
 - CLKGCapBankOverrideEnable Enables the override of the capacitor search during VCO calibration [0 - disable, 1 - enable];
- The capacitor values can be chosen using the registers
 - [\[0x02d\] CLKGLFConfig0 \(page 133\)](#)
 - CLKGCapBankSelect [8] Selects the capacitor bank value for the VCO (only when CLKGCapBankOverrideEnable is 1);
 - [\[0x02b\] CLKGOverrideCapBank \(page 133\)](#)
 - CLKGCapBankSelect [7 : 0] Selects the capacitor bank value for the VCO (only when CLKGCapBankOverrideEnable is 1);

State machine override

Every configuration signal applied to the I^jCDR analog core can be overridden. It is first necessary to enable the override switch located in the register

- [\[0x021\] CLKGConfig1 \(page 131\)](#)
 - CDRControlOverrideEnable Enables the control override of the state machine;
- The override signals are located in the registers
 - [\[0x029\] CLKGFFCAP \(page 132\)](#)
 - CDRCOConnectCDR Enables the connectCDR switch [0 - disable, 1 - enable] (only when CDRCControlOverrideEnable is 1)
 - [\[0x02a\] CLKGCntOverride \(page 133\)](#)
 - CLKGCOoverrideVc Forces the VCO's control voltage to be in mid range [0 - disable, 1 - enable] (only when CDRControlOverrideEnable is 1)
 - CDRCOREfClkSel Forces the reference clock selection for the VCO calibration [0 - data/4, 1 - external refClk] (only when CDRCControlOverrideEnable is 1)

- CDRCOEnablePLL Enables the enablePLL switch [0 - disable, 1 - enable] (only when CDRControlOverrideEnable is 1)
- CDRCOEnableFD Enables the frequency detector [0 - disable, 1 - enable] (only when CDRControlOverrideEnable is 1)
- CDRCOEnableCDR Enables the enableCDR switch [0 - disable, 1 - enable] (only when CDRControlOverrideEnable is 1)
- CDRCODisDataCounterRef Enables the data/4 ripple counter [1 - disable, 0 - enable] (only when CDRControlOverrideEnable is 1)
- CDRCODisDESvbiasGen Enables the vbias for the CDR [1 - disable, 0 - enable] (only when CDRControlOverrideEnable is 1)
- CDRCOConnectPLL Enables the connectPLL switch [0 - disable, 1 - enable] (only when CDRControlOverrideEnable is 1)

Please refer to section [Test outputs](#) (page 121) for valid test output output.

PHASE PROGRAMMABLE CLOCKS

Warning: Known issues: [Section 19.1.6](#), [Section 19.1.8](#).

The IpGBT ASIC provides 4 differential clock outputs that can be used as local timing references for the front-end modules. These clocks are generated by the phase-shifter circuit (see Fig. 10.1), and are fully synchronous with the IpGBT 40 MHz clock which is synchronous with LHC bunch crossing reference and maintain with it a stable phase relationship. The four reference clocks are programmable both in phase, 50 ps resolution and frequency 40, 80, 160, 320, 640, and 1280 MHz. These clocks are associated with differential drivers with programmable driving strength and pre-emphasis can be set independently from each other.

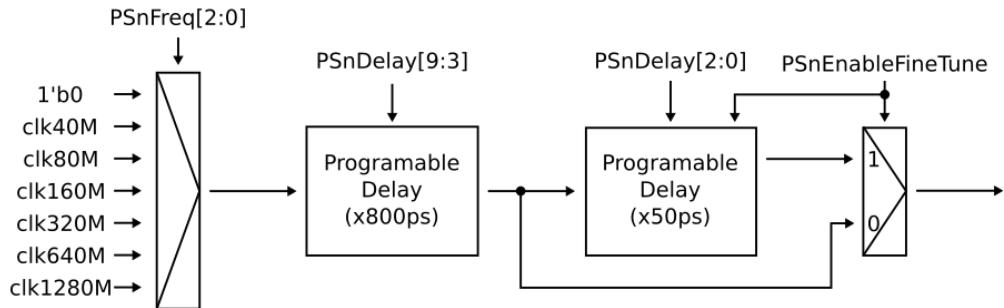


Fig. 10.1: Phase-shifter channel schematic diagram

10.1 Phase-shifter operation

Phase-shifter can deliver a 40, 80, 160, 320, 640, or 1280 MHz clock. The clock phase can be controlled with a resolution of 48.8 ps. Phase shifting is done in two stages: a coarse phase shifting stage where the phase is controlled with a resolution of 781.25 ps within the a full 25 ns clock period followed by a fine phase-shifting stage that further interpolates the clock phase with a resolution of 48.8 ps within the 781.25 ps. This results in a total phase span of 0 to 2π with a resolution of 48.8 ps. The fine phase-shifting stage is based on a DLL calibrated by the 1.28 GHz clock. There are thus 4 DLLs, one per phase-shifter channel. If a coarse phase shifting resolution is sufficient for the application, the fine phase shifting module with DLL can be disabled to save power.

10.2 Programming the phase-shifter channel

Programming the phase-shifter amounts to enable the desired number of active channels and for each channel to choose its frequency, phase, and configuration of the output driver. Circuit parameters for DLLs may also be set.

The channel frequency is set by programming registers PSnFreq[2:0] field in [\[0x05c\] PS0Config](#) (page 143), [\[0x05f\] PS1Config](#) (page 144), [\[0x062\] PS2Config](#) (page 146) or [\[0x065\] PS3Config](#) (page 147) register according to the table:

Table 10.1: PSnFreq

PSnFreq[2:0]	Frequency [MHz]
3'd0	disabled
3'd1	40
3'd2	80
3'd3	160
3'd4	320
3'd5	640
3'd6	1280
3'd7	reserved

Programming a phase-shifter channel delay involves setting two parameters: the coarse delay parameters PSnDelay[9:3] and the fine delay parameters PSnDelay[2:0].

If the PSnEnableFineTune bit is set, the channel phase-shift (or delay) is given by:

$$\text{ChannelDelay} = \text{PSnDelay}[9 : 0] \times 48.8\text{ps}$$

otherwise, it is can be calculated as:

$$\text{ChannelDelay} = \text{PSnDelay}[9 : 3] \times 781.25\text{ps}$$

For correct operation of the phase-shifter, the DLL parameters need to set at start-up. All settings are accessible in [\[0x033\] PSDllConfig](#) (page 134) register.

10.3 Output configuration

The phase-shifter has 4 differential outputs. eTX is used as an output driver. Please refer to [Section 7.7](#) for a description of the driver.

The driving strength of the driver can be controlled by PS0DriveStrength[2:0] field in the [\[0x05c\] PS0Config](#) (page 143) ([\[0x05f\] PS1Config](#) (page 144), [\[0x062\] PS2Config](#) (page 146), or [\[0x065\] PS3Config](#) (page 147)) register. Setting related to pre-emphasis PSnPreEmphasisStrength[2:0], PSnPreEmphasisMode[1:0], and PSnPreEmphasisWidth[2:0] can be found in [\[0x05c\] PS0Config](#) (page 143) ([\[0x05f\] PS1Config](#) (page 144), [\[0x062\] PS2Config](#) (page 146), or [\[0x065\] PS3Config](#) (page 147)) register.

GENERAL PURPOSE I/O

The IpGBT has 16 I/O pins logically divided in two ports: L(ow) and H(igh). One port consists of eight port pins. Each port pin can be configured as input or output with configurable driver and pull settings. All pins operations are synchronous with the internal system clock (40 MHz).

All functions are individually configurable per pin, but several pins can be configured in a single operation. The registers controlling behavior pins are mapped to R/W/F region, which means that some configuration (input or output, pull settings) can be loaded from fuses during power up. All registers used to control and monitor the behavior of the pins: [\[0x052\] PIODirH](#) (page 141), [\[0x053\] PIODirL](#) (page 142) [\[0x054\] PIOOutH](#) (page 142), [\[0x055\] PIOOutL](#) (page 142) [\[0x05a\] PIODriveStrengthH](#) (page 142), [\[0x05b\] PIODriveStrengthL](#) (page 143) [\[0x056\] PIOPullEnaH](#) (page 142), [\[0x057\] PIOPullEnaL](#) (page 142), [\[0x058\] PIOUpDownH](#) (page 142), [\[0x059\] PIOUpDownL](#) (page 142) [\[0x19f\] PIOInH](#) (page 260), [\[0x1a0\] PIOInL](#) (page 260).

Figure Fig. 11.1 shows a schematic diagram of one I/O port pin, here generically called GPIO.

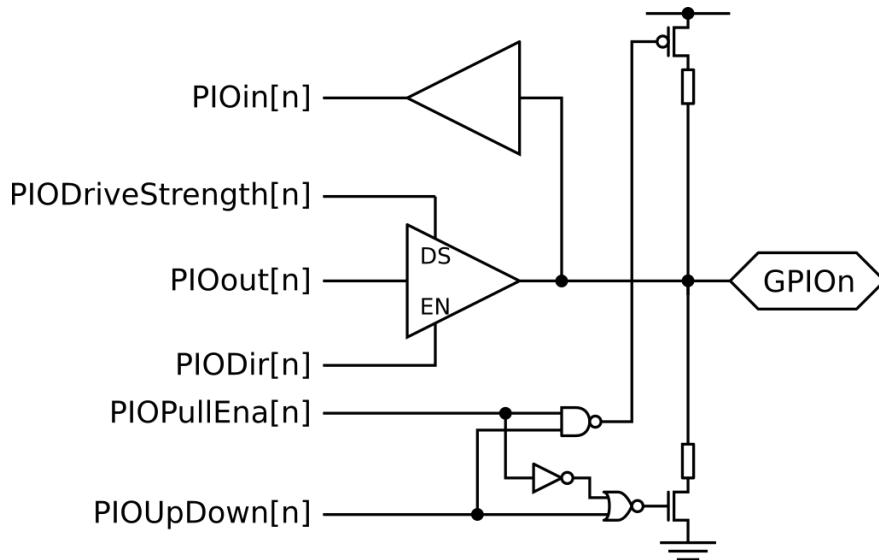


Fig. 11.1: Schematic diagram of IO pin.

The electrical characteristics of the driver are summarized in [Section 18.4](#).

Note: All registers controlling GPIO ports are triplicated and therefore no Single Event Upsets (SEU) are expected on GPIO outputs. However, due to the fact that the output driver itself is not triplicated, occasional Single Event Transients (SET) cannot be eliminated.

11.1 Configuring the Pin

Each pin (GPIOn) is controlled by five bits PIODir[n], PIOOut[n], PIOPullEna[n], PIOUpDown[n], PIODriverStrength[n], additionally there is a read only bit PIOin[n] which enables user to probe the state of the pin.

The PIODir[n] bit in the [\[0x052\] PIODirH](#) (page 141) or [\[0x053\] PIODirL](#) (page 142) register selects the direction of this pin. If PIODir[n] is logic one, GPIOn is configured as an output pin (output driver is enabled). If PIODir[n] is logic zero, GPIOn is configured as an input pin.

Table 11.1: PIODir bit

PIODir[n]	Function
1'b0	Pin configured as an input (output driver is disabled)
1'b1	Pin configured as an output (output driver is enabled)

If PIOOut [n] bit in the [\[0x054\] PIOOutH](#) (page 142) or [\[0x055\] PIOOutL](#) (page 142) register is logic one when the pin is configured as an output pin, the port pin is driven high (one). If PIOOut [n] is logic zero when the pin is configured as an output pin, the port pin is driven low (zero).

Table 11.2: PIOOut bit

PIOOut[n]	Function
1'b0	Pin driven low
1'b1	Pin driven high

All port pins have programmable output configuration which allows to limit output slew rate to reduce electromagnetic emission. The output driving capability is controlled by PIODriverStrength[n] bit in the [\[0x05a\] PIODriverStrengthH](#) (page 142) or [\[0x05b\] PIODriverStrengthL](#) (page 143) register.

Table 11.3: PIODriverStrength bit

PIODriverStrength[n]	Function
1'b0	Low driving strength
1'b1	High driving strength

Additionally, the pull-up or pull-down can be enabled for each pin. In order to enable pull resistor bit PIOPullEna[n] has to be set in [\[0x056\] PIOPullEnaH](#) (page 142) or [\[0x057\] PIOPullEnaL](#) (page 142) register. The pull direction is controlled by bits PIOUpDown[n] in [\[0x058\] PIOUpDownH](#) (page 142) or [\[0x059\] PIOUpDownL](#) (page 142) register. Pull-up resistor is enabled when the bit is set the pull and pull-down resistor is enabled when the bit is cleared. Pull up/down resistor control is independent on direction of the port selected in PIODir[n] which can lead to a direct path from power supply to ground if misconfigured.

Table 11.4: PIOPullEna and PIOUpDown bits

PIOPullEna[n]	PIOUpDown[n]	Function
1'b0	1'bx	Pull up / Pull down resistors disconnected
1'b1	1'b0	Pull up resistor connected
1'b1	1'b1	Pull down resistor connected

11.2 Reading the Pin Value

Independent of the setting of Data Direction bit `PIODir[n]`, the port pin can be read through the `PIOin[n]` bit in [\[0x19f\] `PIOInH`](#) (page 260) or [\[0x1a0\] `PIOInL`](#) (page 260) register. A logical value of one implies that the voltage at given pin is high.

Table 11.5: `PIOin` bit

<code>PIOin[n]</code>	Function
1'b0	Pin is low
1'b1	Pin is high

11.3 Unconnected pins

If some pins are unused, it is recommended to ensure that these pins have a defined level. The simplest method to ensure a defined level of an unused pin, is to enable the internal pull-up or pull-down resistor or to configure it as an output.

CHAPTER
TWELVE

I2C MASTERS

Warning: Known issues: [Section 19.1.4](#), [Section 19.1.5](#).

The lpGBT includes three independent I2C masters (M0, M1, M2) with the following features:

- Compliance with I2C standard, including 10-bit addressing;
- Concurrent operation of all three channels;
- Individually programmable data transfer rates: 100 KHz, 200 KHz, 400 KHz, 1 MHz;
- Supports single-byte and multi-byte (<=16) I2C read/write bus operations;
- Support read-modify-write bitwise operations with OR or XOR masks (7-bit addressing);
- The SCL outputs are configurable as open-drain or full CMOS.

Each of the three masters is connected to a pair of I/O pads driving/receiving data and clock on three independent I2C busses. These signals are on pins M0SCL, M0SDA, M1SCL, M1SDA, M2SCL, M2SDA.

Configuration and operation of the masters are through the lpGBT configuration interface. Each master has a number of control/data input and output signals. The input signals are driven from the lpGBT Read/Write configuration registers (see [Section 15.2.1](#)) and the output signals are read as lpGBT Read-Only registers (see [Section 15.3.3](#)). The functional blocks and interconnections are shown in [Fig. 12.1](#).

Each operation is a sequence of writing address/data words followed by writing a command to trigger the operation.

Initially, each master must be configured before I2C transactions can be made. This configuration can also be changed on-the-fly if necessary.

An I2C transaction can require one or more sequential commands. Single-byte write/read transactions are launched with a single command. Multi-byte-write transactions require data bytes to be written and stored locally within the master prior to launching the I2C transaction. Therefore such a transaction requires a sequence of commands.

Data read from an I2C slave (single-byte or multi-byte) are stored locally in the I2C master. These can then be read through the lpGBT configuration interface. The I2C master can only store the values from one I2C read transaction. They will be over-written when the next I2C read transaction is launched.

12.1 Input/Output signals of I2C masters

The inputs of each I2C master are address, data and command words.

1. Address words are used to address an I2C slave connected to one of the three busses.

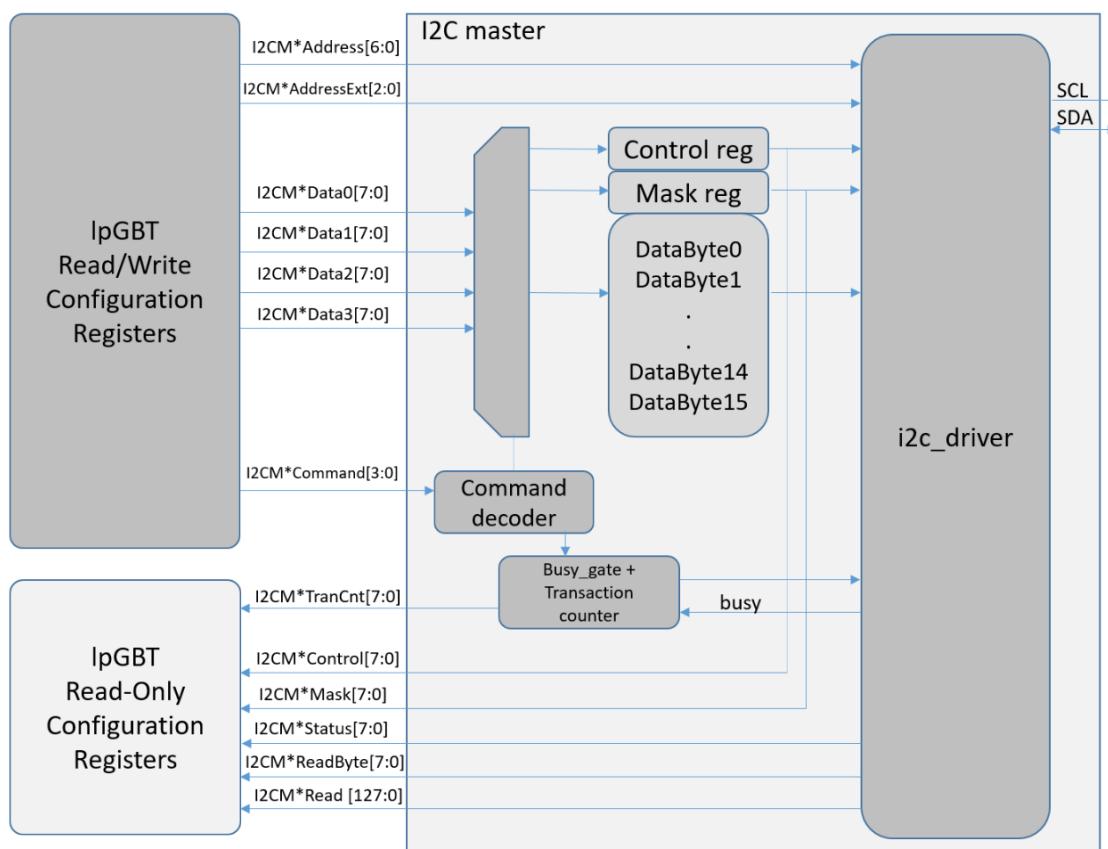


Fig. 12.1: Schematic diagram of I2C Masters block.

2. Data words are written into local registers within each I2C master. These data are either used to configure the master or are transferred to a slave during an I2C-write transaction, depending on the command.
3. The command word selects the action performed by an I2C master. The command can correspond to a configuration of the I2C master or an I2C transaction by that master. Note that the command should only be written after the address and data words are already written.

These inputs are summarized in [Table 12.1](#).

Table 12.1: I2C masters control registers

Master	IpGBT register	Mode	Description
M0	[0x0f1] I2CM0Address (page 227)	W	7 bits of address of slave to address in an I2C transaction
	I2CM0AddressExt field in [0x0f0] I2CM0Config (page 226)	W	3 additional bits of address of slave to address in an I2C transaction; only used in commands with 10-bit addressing
	[0x0f2] I2CM0Data0 (page 227)	W	Data input
	[0x0f3] I2CM0Data1 (page 227)	W	Data input
	[0x0f4] I2CM0Data2 (page 227)	W	Data input
	[0x0f5] I2CM0Data3 (page 227)	W	Data input
	[0x0f6] I2CM0Cmd (page 227)	W	Command word
M1	[0x0f8] I2CM1Address (page 228)	W	7 bits of address of slave to address in an I2C transaction
	I2CM1AddressExt field in [0x0f7] I2CM1Config (page 227)	W	3 additional bits of address of slave to address in an I2C transaction; only used in commands with 10-bit addressing
	[0x0f9] I2CM1Data0 (page 228)	W	Data input
	[0x0fa] I2CM1Data1 (page 228)	W	Data input
	[0x0fb] I2CM1Data2 (page 228)	W	Data input
	[0x0fc] I2CM1Data3 (page 228)	W	Data input
	[0x0fd] I2CM1Cmd (page 228)	W	Command word
M2	[0x0ff] I2CM2Address (page 229)	W	7 bits of address of slave to address in an I2C transaction
	I2CM2AddressExt field in [0x0fe] I2CM2Config (page 228)	W	3 additional bits of address of slave to address in an I2C transaction; only used in commands with 10-bit addressing
	[0x100] I2CM2Data0 (page 229)	W	Data input
	[0x101] I2CM2Data1 (page 229)	W	Data input
	[0x102] I2CM2Data2 (page 229)	W	Data input
	[0x103] I2CM2Data3 (page 229)	W	Data input
	[0x104] I2CM2Cmd (page 229)	W	Command word

The control outputs of each I2C master indicate the configuration of the master, the contents of a mask register in the master, the status of the master, the number of completed transactions, and data read from a slave with an I2C-read transaction. These control outputs are summarized in [Table 12.2](#).

Table 12.2: I2C masters control registers

Master	IpGBT register	Mode	Description
M0	[0x15f] I2CM0Ctrl (page 252)	R	Contents of control register written by user
	[0x160] I2CM0Mask (page 252)	R	Contents of mask register written by user
	[0x161] I2CM0Status (page 253)	R	Contents of status register
	[0x162] I2CM0TranCnt (page 253)	R	Contents of transaction counter
	[0x163] I2CM0ReadByte (page 253)	R	Data read from an I2C slave in a single-byte-read
	[0x164] I2CM0Read0 (page 253)	R	Data read from an I2C slave in a multi-byte-read (bits [7:0])
	[0x165] I2CM0Read1 (page 253)	R	Data read from an I2C slave in a multi-byte-read (bits [15:8])

Continued on next page

Table 12.2 – continued from previous page

Master	IpGBT register	Mode	Description
M1	[0x166] I2CM0Read2 (page 253)	R	Data read from an I2C slave in a multi-byte-read (bits [23:16])
	[0x167] I2CM0Read3 (page 253)	R	Data read from an I2C slave in a multi-byte-read (bits [31:24])
	[0x168] I2CM0Read4 (page 253)	R	Data read from an I2C slave in a multi-byte-read (bits [39:32])
	[0x169] I2CM0Read5 (page 253)	R	Data read from an I2C slave in a multi-byte-read (bits [47:40])
	[0x16a] I2CM0Read6 (page 254)	R	Data read from an I2C slave in a multi-byte-read (bits [55:48])
	[0x16b] I2CM0Read7 (page 254)	R	Data read from an I2C slave in a multi-byte-read (bits [63:56])
	[0x16c] I2CM0Read8 (page 254)	R	Data read from an I2C slave in a multi-byte-read (bits [71:64])
	[0x16d] I2CM0Read9 (page 254)	R	Data read from an I2C slave in a multi-byte-read (bits [79:72])
	[0x16e] I2CM0Read10 (page 254)	R	Data read from an I2C slave in a multi-byte-read (bits [87:80])
	[0x16f] I2CM0Read11 (page 254)	R	Data read from an I2C slave in a multi-byte-read (bits [95:88])
	[0x170] I2CM0Read12 (page 254)	R	Data read from an I2C slave in a multi-byte-read (bits [103:96])
	[0x171] I2CM0Read13 (page 254)	R	Data read from an I2C slave in a multi-byte-read (bits [111:104])
	[0x172] I2CM0Read14 (page 254)	R	Data read from an I2C slave in a multi-byte-read (bits [119:112])
	[0x173] I2CM0Read15 (page 255)	R	Data read from an I2C slave in a multi-byte-read (bits [127:120])
M2	[0x174] I2CM1Ctrl (page 255)	R	Contents of control register written by user
	[0x175] I2CM1Mask (page 255)	R	Contents of mask register written by user
	[0x176] I2CM1Status (page 255)	R	Contents of status register
	[0x177] I2CM1TranCnt (page 255)	R	Contents of transaction counter
	[0x178] I2CM1ReadByte (page 255)	R	Data read from an I2C slave in a single-byte-read
	[0x179] I2CM1Read0 (page 255)	R	Data read from an I2C slave in a multi-byte-read (bits [7:0])
	[0x17a] I2CM1Read1 (page 255)	R	Data read from an I2C slave in a multi-byte-read (bits [15:8])
	[0x17b] I2CM1Read2 (page 255)	R	Data read from an I2C slave in a multi-byte-read (bits [23:16])
	[0x17c] I2CM1Read3 (page 256)	R	Data read from an I2C slave in a multi-byte-read (bits [31:24])
	[0x17d] I2CM1Read4 (page 256)	R	Data read from an I2C slave in a multi-byte-read (bits [39:32])
	[0x17e] I2CM1Read5 (page 256)	R	Data read from an I2C slave in a multi-byte-read (bits [47:40])
	[0x17f] I2CM1Read6 (page 256)	R	Data read from an I2C slave in a multi-byte-read (bits [55:48])
	[0x180] I2CM1Read7 (page 256)	R	Data read from an I2C slave in a multi-byte-read (bits [63:56])
	[0x181] I2CM1Read8 (page 256)	R	Data read from an I2C slave in a multi-byte-read (bits [71:64])
	[0x182] I2CM1Read9 (page 256)	R	Data read from an I2C slave in a multi-byte-read (bits [79:72])
	[0x183] I2CM1Read10 (page 256)	R	Data read from an I2C slave in a multi-byte-read (bits [87:80])
	[0x184] I2CM1Read11 (page 256)	R	Data read from an I2C slave in a multi-byte-read (bits [95:88])
	[0x185] I2CM1Read12 (page 257)	R	Data read from an I2C slave in a multi-byte-read (bits [103:96])
	[0x186] I2CM1Read13 (page 257)	R	Data read from an I2C slave in a multi-byte-read (bits [111:104])
	[0x187] I2CM1Read14 (page 257)	R	Data read from an I2C slave in a multi-byte-read (bits [119:112])
	[0x188] I2CM1Read15 (page 257)	R	Data read from an I2C slave in a multi-byte-read (bits [127:120])
M2	[0x189] I2CM2Ctrl (page 257)	R	Contents of control register written by user
	[0x18a] I2CM2Mask (page 257)	R	Contents of mask register written by user
	[0x18b] I2CM2Status (page 257)	R	Contents of status register
	[0x18c] I2CM2TranCnt (page 257)	R	Contents of transaction counter
	[0x18d] I2CM2ReadByte (page 257)	R	Data read from an I2C slave in a single-byte-read
	[0x18e] I2CM2Read0 (page 258)	R	Data read from an I2C slave in a multi-byte-read (bits [7:0])
	[0x18f] I2CM2Read1 (page 258)	R	Data read from an I2C slave in a multi-byte-read (bits [15:8])
	[0x190] I2CM2Read2 (page 258)	R	Data read from an I2C slave in a multi-byte-read (bits [23:16])
	[0x191] I2CM2Read3 (page 258)	R	Data read from an I2C slave in a multi-byte-read (bits [31:24])
	[0x192] I2CM2Read4 (page 258)	R	Data read from an I2C slave in a multi-byte-read (bits [39:32])
	[0x193] I2CM2Read5 (page 258)	R	Data read from an I2C slave in a multi-byte-read (bits [47:40])
	[0x194] I2CM2Read6 (page 258)	R	Data read from an I2C slave in a multi-byte-read (bits [55:48])
	[0x195] I2CM2Read7 (page 258)	R	Data read from an I2C slave in a multi-byte-read (bits [63:56])
	[0x196] I2CM2Read8 (page 258)	R	Data read from an I2C slave in a multi-byte-read (bits [71:64])
	[0x197] I2CM2Read9 (page 259)	R	Data read from an I2C slave in a multi-byte-read (bits [79:72])
	[0x198] I2CM2Read10 (page 259)	R	Data read from an I2C slave in a multi-byte-read (bits [87:80])

Continued on next page

Table 12.2 – continued from previous page

Master	IpGBT register	Mode	Description
	[0x199] I2CM2Read11 (page 259)	R	Data read from an I2C slave in a multi-byte-read (bits [95:88])
	[0x19a] I2CM2Read12 (page 259)	R	Data read from an I2C slave in a multi-byte-read (bits [103:96])
	[0x19b] I2CM2Read13 (page 259)	R	Data read from an I2C slave in a multi-byte-read (bits [111:104])
	[0x19c] I2CM2Read14 (page 259)	R	Data read from an I2C slave in a multi-byte-read (bits [119:112])
	[0x19d] I2CM2Read15 (page 259)	R	Data read from an I2C slave in a multi-byte-read (bits [127:120])

When an I2C transaction is started by a master, the master will refuse any subsequent requests until that transaction is completed. During this time, it asserts a busy flag that vetoes further requests. The transaction counter is a means of checking if a request was executed or refused by the master. If the request was accepted and completed, the counter is incremented. Otherwise, the counter is not changed. The user can check by reading the value of the transaction counter before and after sending the request.

12.2 Internal registers of I2C masters

Each master has internal programming registers. The writable (W) register contents are for configuring the operation of the master, or storing data for a subsequent multi-byte I2C-write transaction. Values are written to the registers through the IpGBT configuration interface by firstly driving the correct data to the appropriate inputs Data0,1,2,3 and then writing the appropriate command. The readable (R) registers are read through the IpGBT configuration interface. The registers are described in [Table 12.3](#). The default value (after reset) of all bits in all registers is 8'h0.

Table 12.3: I2C masters internal registers

Register	Mode	Data input	Command to write to register	Description
Control Register	W/R	Data0[7:0]	I2C_WRITE_CR	Control register
Mask Register	W/R	Data0[7:0]	I2C_WRITE_MSK	Mask register
DataByte0	W	Data0[7:0]	I2C_W_MULTI_4BYTE0	1st byte
DataByte1	W	Data1[7:0]	I2C_W_MULTI_4BYTE0	2nd byte
DataByte2	W	Data2[7:0]	I2C_W_MULTI_4BYTE0	3rd byte
DataByte3	W	Data3[7:0]	I2C_W_MULTI_4BYTE0	4th byte
DataByte4	W	Data0[7:0]	I2C_W_MULTI_4BYTE1	5th byte
DataByte5	W	Data1[7:0]	I2C_W_MULTI_4BYTE1	6th byte
DataByte6	W	Data2[7:0]	I2C_W_MULTI_4BYTE1	7th byte
DataByte7	W	Data3[7:0]	I2C_W_MULTI_4BYTE1	8th byte
DataByte8	W	Data0[7:0]	I2C_W_MULTI_4BYTE2	9th byte
DataByte9	W	Data1[7:0]	I2C_W_MULTI_4BYTE2	10th byte
DataByte10	W	Data2[7:0]	I2C_W_MULTI_4BYTE2	11th byte
DataByte11	W	Data3[7:0]	I2C_W_MULTI_4BYTE2	12th byte
DataByte12	W	Data0[7:0]	I2C_W_MULTI_4BYTE3	13th byte
DataByte13	W	Data1[7:0]	I2C_W_MULTI_4BYTE3	14th byte
DataByte14	W	Data2[7:0]	I2C_W_MULTI_4BYTE3	15th byte
DataByte15	W	Data3[7:0]	I2C_W_MULTI_4BYTE3	16th byte
Status Register	R	•	•	Status register

12.2.1 Control register

- **Bit [7] - SCLDriveMode** - 1'b0 = SCL pad is open-drain, so it pulls down the line to VSS or is in high impedance. A pull-up resistor must be used. 1'b1 = SCL is driven by a CMOS buffer, so it pulls down the line to VSS or pulls up the line to VDD. No pull-up resistor is required.

- Bit [6:2] - NBYTE[4:0] - number of bytes in an I2C multi-byte write or read (maximum = d'16 = b'10000)
- Bit [1:0] - FREQ[1:0] - frequency of I2C bus transaction according to Table 12.4.

Table 12.4: I2C masters frequency of I2C bus transaction

FREQ[1:0]	frequency
2'b00	100 kHz
2'b01	200 kHz
2'b10	400 kHz
2'b11	1 MHz

12.2.2 Mask register

Used for bitwise operation of read-from-slave, apply mask, write-to-slave. Used with commands I2C_1BYTE_RMW_OR and I2C_1BYTE_RMW_XOR.

12.2.3 Status registers

- Bit [7] - unused.
- Bit [6] - set if the last transaction was not acknowledged by the I2C slave. Value is valid at the end of the I2C transaction. Reset if a slave acknowledges the next I2C transaction.
- Bit [5] - set if an invalid command was sent to the I2C channel. Only cleared by a reset.
- Bit [4] - unused.
- Bit [3] - set if the I2C master port finds that the SDA line is pulled low '0' before initiating a transaction. Indicates a problem with the I2C bus. Represents the status of the SDA line and cannot be reset.
- Bit [2] - set when the last I2C transaction was executed successfully. Reset by the start of the next I2C transaction.
- Bits [1:0] - unused.

12.3 I2C master commands

The Table 12.5 describes the commands for an I2C master. Each table contains the 4-bit command code and the usage of the address/data words for that command.

Table 12.5: I2C master commands

Command	Command code
I2C_WRITE_CR	0x0
I2C_WRITE_MSK	0x1
I2C_1BYTE_WRITE	0x2
I2C_1BYTE_READ	0x3
I2C_1BYTE_WRITE_EXT	0x4
I2C_1BYTE_READ_EXT	0x5
I2C_1BYTE_RMW_OR	0x6
I2C_1BYTE_RMW_XOR	0x7
I2C_W_MULTI_4BYTE0	0x8
I2C_W_MULTI_4BYTE1	0x9
I2C_W_MULTI_4BYTE2	0xA
I2C_W_MULTI_4BYTE3	0xB
I2C_WRITE_MULTI	0xC
I2C_READ_MULTI	0xD
I2C_WRITE_MULTI_EXT	0xE
I2C_READ_MULTI_EXT	0xF

12.3.1 I2C_WRITE_CR (0x0)

This command writes data to the configuration register of the master. The user must write the correct Data0 word BEFORE writing this command. Writing this command will NOT start a transaction on the I2C bus

Table 12.6: I2C_WRITE_CR command

AddressExt	unused
Address	unused
Data0	Data for control register
Data1	unused
Data2	unused
Data3	unused
Read	unused
ReadByte	unused

12.3.2 I2C_WRITE_MSK (0x1)

This command writes data to the mask register of the master. The user must write the correct Data0 word BEFORE writing this command. Writing this command will NOT start a transaction on the I2C bus

Table 12.7: I2C_WRITE_MSK command

AddressExt	unused
Address	unused
Data0	Data for mask register
Data1	unused
Data2	unused
Data3	unused
Read	unused
ReadByte	unused

12.3.3 I2C_1BYTE_WRITE (0x2)

Writing this command will immediately start a write transaction on the I2C bus. The user must write the correct Address and Data0 words BEFORE writing this command. On the I2C bus, the master first transmits the 7-bit slave address with R/W=0, and then the Data byte.

Table 12.8: I2C_1BYTE_WRITE command

AddressExt	unused
Address	7-bit I2C address of target slave
Data0	Data byte to write to target slave
Data1	unused
Data2	unused
Data3	unused
Read	unused
ReadByte	unused

12.3.4 I2C_1BYTE_READ (0x3)

Writing this command will immediately start a read transaction on the I2C bus. The user must write the correct Address word BEFORE writing this command. On the I2C bus, the master first transmits the 7-bit slave address with R/W=1. The slave then transmits the Data byte.

Table 12.9: I2C_1BYTE_READ command

AddressExt	unused
Address	7-bit I2C address of target slave
Data0	unused
Data1	unused
Data2	unused
Data3	unused
Read	unused
ReadByte	Data byte read from target slave

12.3.5 I2C_1BYTE_WRITE_EXT (0x4)

Writing this command will immediately start a write transaction on the I2C bus. The user must write the correct AddressExt, Address and Data0 words BEFORE writing this command. On the I2C bus, the master first transmits the 10-bit slave address with R/W=0, and then the Data byte.

Table 12.10: I2C_1BYTE_WRITE_EXT command

AddressExt	Bits[9:7] of 10-bit I2C address of target slave
Address	Bits[6:0] of 10-bit I2C address of target slave
Data0	Data byte to write to target slave
Data1	unused
Data2	unused
Data3	unused
Read	unused
ReadByte	unused

12.3.6 I2C_1BYTE_READ_EXT (0x5)

Writing this command will immediately start a read transaction on the I2C bus. The user must write the correct AddressExt and Address words BEFORE writing this command. On the I2C bus, the master first transmits the 10-bit slave address with R/W=1. The slave then transmits the Data byte.

Table 12.11: I2C_1BYTE_READ_EXT command

AddressExt	Bits[9:7] of 10-bit I2C address of target slave
Address	Bits[6:0] of 10-bit I2C address of target slave
Data0	unused
Data1	unused
Data2	unused
Data3	unused
Read	unused
ReadByte	Data byte read from target slave

12.3.7 I2C_1BYTE_RMW_OR (0x6)

Writing this command will immediately start transactions on the I2C bus. The user must write the correct Address word BEFORE writing this command. This is a READ-MODIFY-WRITE operation. On the I2C bus, the master first transmits the slave address with R/W=1. The slave then transmits the Data byte to the master. The master then makes a bitwise OR of the Data byte and the Mask register. The output byte from this operation is written back to the slave: the master transmits the slave address with R/W=0, and then the new Data byte.

Table 12.12: I2C_1BYTE_RMW_OR command

AddressExt	unused
Address	7-bit I2C address of target slave
Data0	unused
Data1	unused
Data2	unused
Data3	unused
Read	unused
ReadByte	unused

12.3.8 I2C_1BYTE_RMW_XOR (0x7)

Writing this command will immediately start transactions on the I2C bus. The user must write the correct Address word BEFORE writing this command. This is a READ-MODIFY-WRITE operation. On the I2C bus, the master first transmits the slave address with R/W=1. The slave then transmits the Data byte to the master. The master then makes a bitwise XOR of the Data byte and the Mask register. The output byte from this operation is written back to the slave: the master transmits the slave address with R/W=0, and then the new Data byte.

Table 12.13: I2C_1BYTE_RMW_XOR command

AddressExt	unused
Address	7-bit I2C address of target slave
Data0	unused
Data1	unused
Data2	unused
Data3	unused
Read	unused
ReadByte	unused

12.3.9 I2C_W_MULTI_4BYTE0 (0x8)

Writing this command will NOT start a transaction on the I2C bus. The 4 bytes of data are stored locally within the I2C master. These data can then be written to a slave by the I2C_WRITE_MULTI or I2C_WRITE_MULTI_EXT commands.

Table 12.14: I2C_W_MULTI_4BYTE0 command

AddressExt	unused
Address	unused
Data0	1st byte for I2C write
Data1	2nd byte for I2C write
Data2	3rd byte for I2C write
Data3	4th byte for I2C write
Read	unused
ReadByte	unused

12.3.10 I2C_W_MULTI_4BYTE1 (0x9)

Writing this command will NOT start a transaction on the I2C bus. The 4 bytes of data are stored locally within the I2C master. These data can then be written to a slave by the I2C_WRITE_MULTI or I2C_WRITE_MULTI_EXT commands.

Table 12.15: I2C_W_MULTI_4BYTE1 command

AddressExt	unused
Address	unused
Data0	5th byte for I2C write
Data1	6th byte for I2C write
Data2	7th byte for I2C write
Data3	8th byte for I2C write
Read	unused
ReadByte	unused

12.3.11 I2C_W_MULTI_4BYTE2 (0xA)

Writing this command will NOT start a transaction on the I2C bus. The 4 bytes of data are stored locally within the I2C master. These data can then be written to a slave by the I2C_WRITE_MULTI or I2C_WRITE_MULTI_EXT commands.

Table 12.16: I2C_W_MULTI_4BYTE2 command

AddressExt	unused
Address	unused
Data0	9th byte for I2C write
Data1	10th byte for I2C write
Data2	12th byte for I2C write
Data3	12th byte for I2C write
Read	unused
ReadByte	unused

12.3.12 I2C_W_MULTI_4BYTE3 (0xB)

Writing this command will NOT start a transaction on the I2C bus. The 4 bytes of data are stored locally within the I2C master. These data can then be written to a slave by the I2C_WRITE_MULTI or I2C_WRITE_MULTI_EXT commands.

Table 12.17: I2C_W_MULTI_4BYTE3 command

AddressExt	unused
Address	unused
Data0	13th byte for I2C write
Data1	14th byte for I2C write
Data2	15th byte for I2C write
Data3	16th byte for I2C write
Read	unused
ReadByte	unused

12.3.13 I2C_WRITE_MULTI (0xC)

Writing this command will immediately start a write transaction on the I2C bus. The user must write the correct Address word BEFORE writing this command. On the I2C bus, the master first transmits the 7-bit slave address with R/W=0, and then the Data bytes. The Data bytes are those previously written to the master using the commands I2C_W_MULTI_4BYTE3,2,1,0. The number of transmitted Data bytes is according to the value of bits [6:2] of the Control Register.

Table 12.18: I2C_WRITE_MULTI command

AddressExt	unused
Address	7-bit I2C address of target slave
Data0	unused
Data1	unused
Data2	unused
Data3	unused
Read	unused
ReadByte	unused

12.3.14 I2C_READ_MULTI (0xD)

Writing this command will immediately start a read transaction on the I2C bus. The user must write the correct Address word BEFORE writing this command. On the I2C bus, the master first transmits the 7-bit slave address with

R/W=1. The slave then transmits the Data bytes. The number of received Data bytes is according to the value of bits [6:2] of the Control Register. Note that the 1st byte received in time by the master is output as bits [127:120] of Read[127:0], the 2nd byte as bits [119:112], etc etc.

Table 12.19: I2C_READ_MULTI command

AddressExt	unused
Address	7-bit I2C address of target slave
Data0	unused
Data1	unused
Data2	unused
Data3	unused
Read	[127:120] = 1st byte received, [119:112] = 2nd byte, ..., [7:0] = 16th byte
ReadByte	unused

12.3.15 I2C_WRITE_MULTI_EXT (0xE)

Writing this command will immediately start a write transaction on the I2C bus. The user must write the correct Address word BEFORE writing this command. On the I2C bus, the master first transmits the 7-bit slave address with R/W=0, and then the Data bytes. The Data bytes are those previously written to the master using the commands I2C_W_MULTI_4BYTE3,2,1,0. The number of transmitted Data bytes is according to the value of bits [6:2] of the Control Register.

Table 12.20: I2C_WRITE_MULTI_EXT command

AddressExt	Bits[9:7] of 10-bit I2C address of target slave
Address	Bits[6:0] of 10-bit I2C address of target slave
Data0	unused
Data1	unused
Data2	unused
Data3	unused
Read	unused
ReadByte	unused

12.3.16 I2C_READ_MULTI_EXT (0xF)

Writing this command will immediately start a read transaction on the I2C bus. The user must write the correct AddressExt and Address words BEFORE writing this command. On the I2C bus, the master first transmits the 10-bit slave address with R/W=1. The slave then transmits the Data bytes. The number of received Data bytes is according to the value of bits [6:2] of the Control Register. Note that the 1st byte received in time by the master is output as bits [127:120] of Read[127:0], the 2nd byte as bits [119:112], etc etc.

Table 12.21: I2C_READ_MULTI_EXT command

AddressExt	Bits[9:7] of 10-bit I2C address of target slave
Address	Bits[6:0] of 10-bit I2C address of target slave
Data0	unused
Data1	unused
Data2	unused
Data3	unused
Read	[127:120] = 1st byte received, [119:112] = 2nd byte, ..., [7:0] = 16th byte
ReadByte	unused

12.4 Configuration of the I/O pins

Each of the SCL and SDA pins is configurable in terms of drive strength and enabling of an internal pull-up resistor. More details can be found in the section on ‘Electrical Characteristics: CMOS I/O Pin Characteristics’. Note that the pull-up resistors are connected to VDD (nominally 1.2V). The settings and the corresponding IpGBT configuration registers are listed in Table 12.22.

Table 12.22: I2C Masters configuration of the I/O pins

Master	Pin	Description	Configuration register
M0	M0SCL	SCL drive strength	I2CM0SCLDriveStrength in [0x0f0] I2CM0Config (page 226)
	M0SCL	SCL pull-up enable	I2CM0SCLPullUpEnable in [0x0f0] I2CM0Config (page 226)
	M0SDA	SDA drive strength	I2CM0SDADriveStrength in [0x0f0] I2CM0Config (page 226)
	M0SDA	SDA pull-up enable	I2CM0SDAPullUpEnable in [0x0f0] I2CM0Config (page 226)
M1	M1SCL	SCL drive strength	I2CM1SCLDriveStrength in [0x0f7] I2CM1Config (page 227)
	M1SCL	SCL pull-up enable	I2CM1SCLPullUpEnable in [0x0f7] I2CM1Config (page 227)
	M1SDA	SDA drive strength	I2CM1SDADriveStrength in [0x0f7] I2CM1Config (page 227)
	M1SDA	SDA pull-up enable	I2CM1SDAPullUpEnable in [0x0f7] I2CM1Config (page 227)
M2	M2SCL	SCL drive strength	I2CM2SCLDriveStrength in [0x0fe] I2CM2Config (page 228)
	M2SCL	SCL pull-up enable	I2CM2SCLPullUpEnable in [0x0fe] I2CM2Config (page 228)
	M2SDA	SDA drive strength	I2CM2SDADriveStrength in [0x0fe] I2CM2Config (page 228)
	M2SDA	SDA pull-up enable	I2CM2SDAPullUpEnable in [0x0fe] I2CM2Config (page 228)

12.5 I2C transaction during power-up

As mentioned in [Section 8.1](#), the IpGBT can execute one multi-byte write I2C transaction (I2C_WRITE_MULTI) during its power up. This feature can be used to configure a laser driver chip or any other component in the system. The transaction can be executed on any of the I2C master interfaces (M0, M1, or M2). All the registers controlling the behavior of this feature are placed in the Read/Write/Fuse region of the configuration memory (see [Section 15.1.9](#)), which means that this feature can be used even if no communication channel is available yet (e.g. serial control).

To enable transaction, the I2CMTransEnable bit in the [\[0x03f\] I2CMTransConfig](#) (page 140) bit has to be programmed and master channel has to be selected by I2CMTransChannel[1:0]. Remaining configuration like I2CMTransAddressExt[2:0], I2CMTransAddress[6:0], and I2CMTransCtrl[127:0] should be configured according to the description in [Section 12.2](#).

12.6 Examples

The following are examples of the procedure to launch different types of I2C transaction.

12.6.1 Example 1 : Single byte write

Write a single byte (8'hCA) to an I2C slave (address = 7'b1010101) using master M0 with clock speed 400kHz. SCL is open-drain.

```
// Configure master M0
writeReg(I2CM0DATA0, I2CM_CR_FREQ_400K<<I2CM_CR_FREQ_of);
writeReg(I2CM0CMD, I2CM_WRITE_CR);

// Launch a single-byte I2C write
```

```

writeReg(I2CM0ADDRESS, {1'b0,7'b1010101});
writeReg(I2CM0DATA0, 8'hCA);
writeReg(I2CM0CMD, I2C_1BYTE_WRITE);

```

12.6.2 Example 2 : Multi byte write

Write six bytes 8'hAA, BB, CC, DD, EE, FF (AA first, BB second etc) to an I2C slave (address = 7'b0001111) using master M1 with clock speed 1MHz. SCL is full CMOS.

```

// Configure master M1
writeReg(I2CM1DATA0, I2CM_CR_SCLDRIVE_bm | (5'd6<<I2CM_CR_NBYTES_of) | (I2CM_
↪CR_FREQ_400K<<I2CM_CR_FREQ_of));
writeReg(I2CM1CMD, I2C_WRITE_CR);

// Write first four data bytes to DataByte Registers 0,1,2,3 in M1
writeReg(I2CM1DATA0, 8'hAA);
writeReg(I2CM1DATA1, 8'hBB);
writeReg(I2CM1DATA2, 8'hCC);
writeReg(I2CM1DATA3, 8'hDD);
writeReg(I2CM1CMD, I2CM_W_MULTI_4BYTE0);
writeReg(I2CM1DATA0, 8'hEE);
writeReg(I2CM1DATA1, 8'hFF);
writeReg(I2CM1CMD, I2CM_W_MULTI_4BYTE1);

// Launch a multi-byte I2C write
writeReg(I2CM1ADDRESS, {1'b0,7'b0001111});
writeReg(I2CM1CMD, I2CM_WRITE_MULTI);

```

12.6.3 Example 3 : Multi byte read

Read fourteen bytes from an I2C slave (slaveAddress[9:0] = 10'b1010101111) using master M2 with clock speed 200kHz. SCL is open-drain.

```

// Configure master M2
writeReg(I2CM2DATA0, (5'd14<<I2CM_CR_NBYTES_of) | (I2CM_CR_FREQ_200K<<I2CM_
↪CR_FREQ_of));
writeReg(I2CM1CMD, I2C_WRITE_CR);

// Launch a multi-byte I2C write
writeReg(I2CM2ADDRESS, {1'b0,slaveAddress[6:0]});
writeReg(I2CM2CONFIG, slaveAddress[9:7]<<I2CM2CONFIG_I2CM2ADDRESSEXT_of);
writeReg(I2CM2CMD, I2CM_READ_MULTI_EXT);

// Read bytes from read-only registers
data = readReg(I2CM2READ15); // 1st byte received from slave
data = readReg(I2CM2READ14); // 2nd byte received from slave
data = readReg(I2CM2READ13); // 3rd byte received from slave
data = readReg(I2CM2READ12); // 4th byte received from slave
data = readReg(I2CM2READ11); // 5th byte received from slave
data = readReg(I2CM2READ10); // 6th byte received from slave
data = readReg(I2CM2READ9); // 7th byte received from slave
data = readReg(I2CM2READ8); // 8th byte received from slave
data = readReg(I2CM2READ7); // 9th byte received from slave

```

```
data = readReg(I2CM2READ6); // 10th byte received from slave
data = readReg(I2CM2READ5); // 11th byte received from slave
data = readReg(I2CM2READ4); // 12th byte received from slave
data = readReg(I2CM2READ3); // 13th byte received from slave
data = readReg(I2CM2READ2); // 14th byte received from slave
```


ANALOG PERIPHERALS

The following list summaries analog features of the lpGBT chip:

- 10-bit analog to digital converter
- 16 multiplexed inputs (out of which 8 are external)
- differential amplifier with configurable gain stage
- conversion time below $1\mu\text{s}$
- Embedded temperature sensor
- Embedded power supply monitors
- Programmable current source could be attached to any ADC input channel
- Internal or external voltage reference
- 12-bit voltage DAC
- Calibration constants for all analog blocks

Fig. 13.1 shows an organization and inter-connectivity of analog blocks inside lpGBT chip.

13.1 Analog to Digital Converter

The ADC implemented in lpGBT is a fully differential 10-bit SAR ADC. Its input dynamic range covers -VREF to VREF. A wide range of multiplexer (MUX) settings and integrated gain stage, make this a flexible module suitable for a wide range of applications, such as data acquisition, embedded control, and general signal processing.

13.1.1 Performing conversion

To enable ADC block, an `ADCEnable` bit has to be set in [\[0x113\] `ADCCConfig`](#) (page 233) register. To initiate a conversion, `ADCConvert` bit has to be set in [\[0x113\] `ADCCConfig`](#) (page 233) register. While the ADC is performing conversion, `ADCBusy` bit is set in [\[0x1b8\] `ADCStatusH`](#) (page 264). Once the conversion is finished, `ADCDone` bit is set in [\[0x1b8\] `ADCStatusH`](#) (page 264). The result of the conversion is available in `ADCValue` field spread across [\[0x1b9\] `ADCStatusL`](#) (page 264) and [\[0x1b8\] `ADCStatusH`](#) (page 264) registers. The conversion result is represented as 10-bit unsigned integer (0-1023).

13.1.2 Gain Stage

The ADC has an internal fully differential gain stage, which can be configured to amplify a voltage to allow measurement of smaller voltages in differential mode. The gain stage is inserted between the channel input selection MUX and

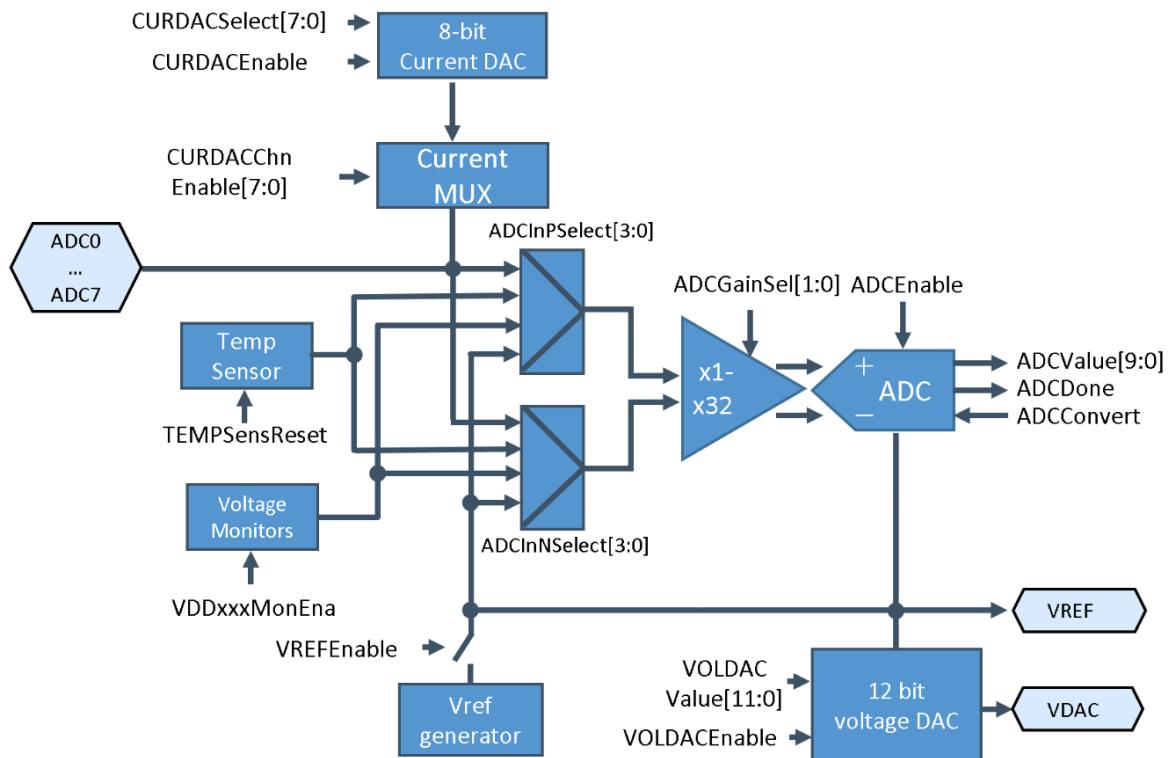


Fig. 13.1: Schematic diagram of analog peripherals.

the ADC core block. The available gain settings are 2x, 8x, 16x, and 32x and the gain is controlled by `ADCGainSel` field in [\[0x113\] ADCConfig](#) (page 233) register.

Table 13.1: ADC Gain settings

<code>ADCGainSel[1:0]</code>	Gain
2'd0	x2
2'd1	x8
2'd2	x16
2'd3	x32

13.1.3 Multiplexer Settings

The input MUXes are used to select input signal for the converter. The positive and negative inputs are selected using the MUX Positive Input (`ADCInPSelect`) and MUX Negative Input (`ADCInNSelect`) bit fields in the [\[0x111\] ADCSelect](#) (page 231) register according to the [Table 13.2](#).

Table 13.2: ADC Input MUX settings

<code>ADCInPSelect[3:0]</code>	Input
4'd0	ADC0 (external pin)
4'd1	ADC1 (external pin)
4'd2	ADC2 (external pin)
4'd3	ADC3 (external pin)
4'd4	ADC4 (external pin)
4'd5	ADC5 (external pin)
4'd6	ADC6 (external pin)
4'd7	ADC7 (external pin)
4'd8	EOM DAC (internal signal)
4'd9	VDDIO * 0.42 (internal signal)
4'd10	VDDTX * 0.42 (internal signal)
4'd11	VDDRX * 0.42 (internal signal)
4'd12	VDD * 0.42 (internal signal)
4'd13	VDDA * 0.42 (internal signal)
4'd14	Temperature sensor (internal signal)
4'd15	VREF/2 (internal signal)

13.1.4 Measurement modes

Several different measurement modes can be obtained by various combination of multiplexers and the gain stage:

- **Differential Input.** With this setting, the ADC measures the difference between any two signals. If higher resolution is required the gain of the differential amplifier can be adapted accordingly. Moreover, the ADC offset can be measured quite easily by setting up the positive and negative input to the same pin.
- **Single-Ended Input.** With this setting, the ADC measures the value of one input signal. The difference between this setting and differential measurement is that the negative input should be connected internally to a VREF/2 level.
- **Internal Input** With this setting, the ADC measures one of several internal signals. The negative should be connected internally to a VREF/2 level while the positive input can be connected to one of the following internal sources: Temperature sensor, one of the power supply monitors or EOM DAC monitor.

The configuration of the measurement mode should be done before starting the conversion.

13.1.5 Source Impedance

This is a very common problem when doing ADC designs. In order to avoid accuracy problems caused by input current of the ADC (see `Ibias` in [Section 18.8](#)) the source impedance should be relatively low (below 1 KΩ). In the case of source impedance being larger than recommended value, additional calibration may be applied in order to improve accuracy.

13.1.6 Calibration

In order to improve the DC accuracy of the ADC, the calibration procedure is performed during production testing process. Several input voltages are applied and an ADC conversion is performed in various measurement modes (differential/single ended, x2, x4, ...). Conversion codes are stored in electrical fuses and can be readout by the user to improve the measurement accuracy.

The calibration during production testing is performed at room temperature and may not be accurate across the whole temperature range. Please see [Section 18.8](#) for typical results.

13.1.7 Usage examples

A pseudo code attached below shows the sequence required to initiate the ADC block and perform single ended conversion of the signal connected to external ADC input ADC0.

```
// configure input multiplexers to measure ADC0 in single ended modePins
// ADCInPSelect = ADCCHN_EXT0 ; (4'd0)
// ADCInNSelect = ADCCHN_VREF2 ; (4'd15)
writeReg(ADCSELECT, ADCCHN_EXT0<<ADCSELECT_ADCINPSELECT_of | ADCCHN_VREF2 <<
          ↳ADCSELECT_ADCINNSELECT_of);

// enable ADC core and set gain of the differential amplifier
writeReg(ADCCONFIG, ADCCONFIG_ADCENABLE_bm | ADCGAIN_X2<<ADCCONFIG_-
          ↳ADCGAINSELECT_of);

// enable internal voltage reference
writeReg(VREFCNTR, VREFCNTR_VREFENABLE_bm);

// wait until voltage reference is stable
sleepms(10);

// start ADC conversion
writeReg(ADCCONFIG, ADCCONFIG_ADCCONVERT_bm | ADCCONFIG_ADCENABLE_bm | ↳
          ↳ADCGAIN_X2<<ADCCONFIG_ADCGAINSELECT_of);

do
    status=readReg(ADCSTATUSH)
  while !(status&ADCSTATUSH_ADCDONE_bm)

// read ADC value
adcValue[9:8]=readReg(ADCStatusH) [1:0]
adcValue[7:0]=readReg(ADCStatusL)

// if ADC is not longer needed you may power it down to save power
writeReg(ADCCONFIG, 8'b0);
writeReg(VREFCNTR, 8'b0);
```

In order to perform a conversion of a very small differential signal connected between external inputs ADC6 and ADC7, the configuration may look like:

```
// configure input multiplexers to measure ADC0 in single ended modePins
// ADCInPSelect = ADCCHN_EXT0 ; (4'd0)
// ADCInNSelect = ADCCHN_VREF2 ; (4'd15)
writeReg(ADCSELECT, ADCCHN_EXT6<<ADCSELECT_ADCINPSELECT_of | ADCCHN_EXT7 <<_
↪ADCSELECT_ADCINNSELECT_of);

// enable ADC core and set gain of the differential amplifier
writeReg(ADCCONFIG, ADCCONFIG_ADCENABLE_bm | ADCGAIN_X32<<ADCCONFIG_-
↪ADCGAINSELECT_of);
```

13.2 Reference voltage

The IpGBT has a built-in reference voltage generator which is shared between ADC and voltage DAC.

The internal reference is 1.00V and it is derived from an internal bandgap circuit. Bit VREFEnable has to be set in the [\[0x01c\] VREFCNTR](#) (page 130) register in order to enable internal voltage reference generator.

The accuracy of the reference is dependent on the bandgap circuit and on the multiplying amplifier. In order to improve the DC accuracy a tuning circuit is added. Tuning can be performed by adjusting VREFTune [5:0] field in the [\[0x01c\] VREFCNTR](#) (page 130) register. Value of VREFTune [5:0] register will be stored during production testing. It should be noted, that no external decoupling should be added to VREF pin if the internal reference generator is enabled.

When the internal reference generator is disabled (VREFEnable bit set to zero) one can provide reference voltage from outside using VREF pin.

Please refer Section 18.9 for electrical specifications.

13.3 Temperature sensor

The internal temperature sensor is linear, and is intended to give a rough approximation of the ambient temperature (not a PT100 sensor replacement). In order to measure temperature, the ADC block has to be used. ADC's positive input should be connected to the temperature sensor and the ADC's negative input should be connected to VREF / 2.

Information about calibration of the temperature sensor will be added once prototype is available.

13.4 Current Sources

All ADC inputs (ADC0, ..., ADC7) feature a switchable current source. This feature can be used to measure externally connected resistances (e.g. temperature sensors: PT100 or PT1000). The current DAC can be enabled by setting CURDACEnable bit in the [\[0x068\] DACConfigH](#) (page 149) register.

In order to maximize the flexibility of the block, the current can be programmed in wide range using 8 bit current DAC by setting CURDACSelect [7:0] field in [\[0x06a\] CURDACValue](#) (page 149) register. The output of the current DAC can be mirrored to any of the ADC inputs. In order to attach the current source to ADCn pin, one should set CURDACChnEnable [n] in the register [\[0x06b\] CURDACCHN](#) (page 150).

Information about calibration of the current source will be added once prototype is available.

13.4.1 Usage example

A pseudo code attached below shows how to generate constant current for pins ADC3 and ADC6.

```
// enable current DAC (access to this register affects also voltage DAC)
writeReg(DACCONFIGH, DACCONFIGH_CURDACENABLE_bm)

// set current value
writeReg(CURDACVALUE, 8'hxx)

// enable current source for pin ADC3 and ADC6
writeReg(DACCONFIGH, CURDAC_CHN3_bm | CURDAC_CHN6_bm )
```

13.5 Voltage Digital to Analog Converter

The IpGBT has a 12-bit R-2R voltage DAC. The DAC uses VREF voltage as a reference, and thus can produce voltages ranging from 0 to VREF. Output of the DAC is buffered to provide lower output impedance to facilitate driving resistive loads.

In order to save power, the DAC is disabled by default. In order to enable it, VOLDACEnable bit has to be set in the [\[0x068\] DACConfigH](#) (page 149) register. The output voltage is controlled by VOLDACValue field spread across [\[0x068\] DACConfigH](#) (page 149) and [\[0x069\] DACConfigL](#) (page 149) registers. The output voltage can be computed according to formula:

$$V_{out} = \frac{VOLDACValue}{4096} VREF$$

Information about calibration of the voltage DAC will be added once prototype is available.

13.5.1 Usage example

A pseudo code attached below shows how to generate 0.22 V at the output of the voltage DAC.

```
// calculate DAC value
vout=0.22
vref=1.00
value=vout/vref*4096

// write LSBs
writeReg(DACCONFIGL, value[7:0])

// write MSBs and enable DAC
writeReg(DACCONFIGH, DACCONFIGH_VOLDACENABLE_bm | (value[11:8]<<DACCONFIGH_
    ->VOLDACVALUE_of) )
```

CHAPTER
FOURTEEN

BUILT IN TEST FEATURES

The lpGBT has many test features. They can be divided in two broad groups: testing the internal operation of the chip (power up) and data path validation. The features that test the chip internal operation are described in [Section 8](#). This chapter addresses data path checking.

Built in link test features are essential for evaluation, production and in-system testing. Moreover, they allow standardized link tests procedures at the system level. The lpGBT offers a variety of link test features:

- loopbacks
- test pattern generators
- pattern checkers
- test outputs
- eye opening monitor

The schematic diagram of data path test features is presented in [Fig. 14.1](#).

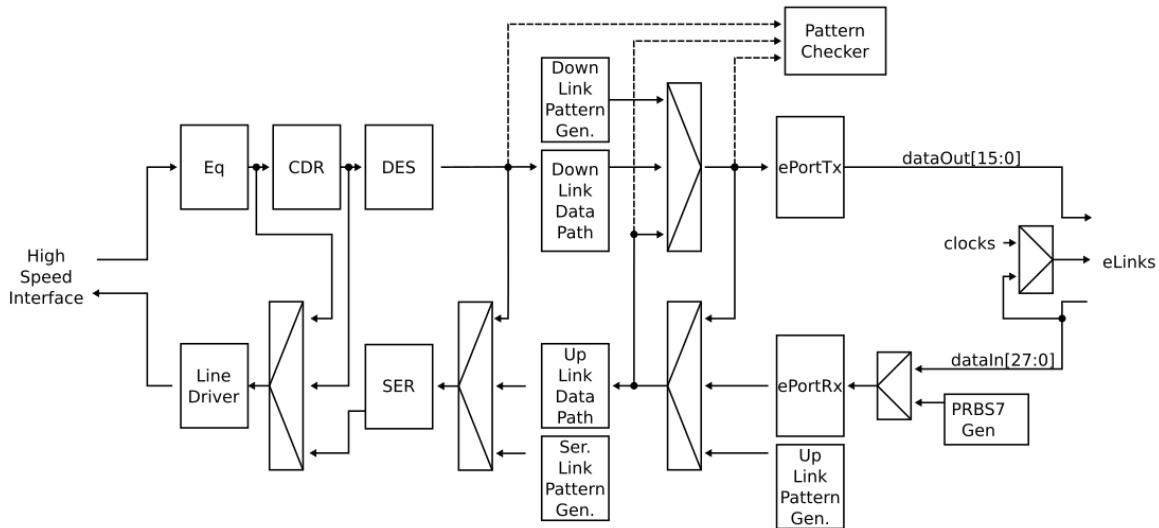


Fig. 14.1: lpGBT test features.

14.1 Test pattern generators

The IpGBT offers a possibility to generate patterns and inject them at various places in the data path in order to simplify chip/system debugging. Points at which data can be generated are highlighted on Fig. 14.2.

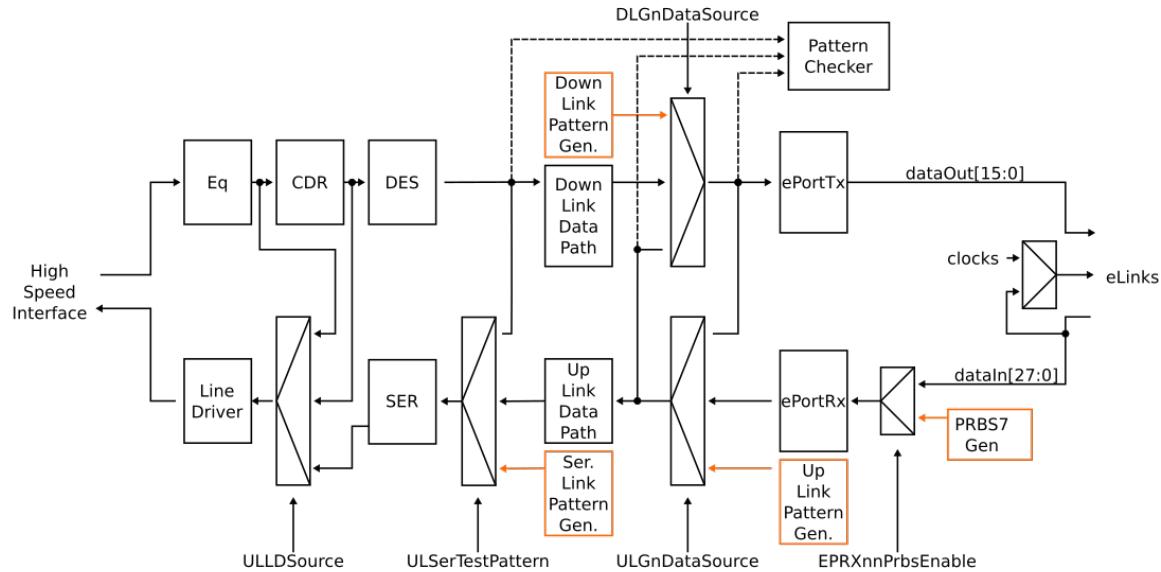


Fig. 14.2: IpGBT data path pattern generators.

One should note, that if any of the pattern generators is enabled, the user data (from ePortRx or from the downlink frame) are discarded. All generators are independent form each other and can be used at the same time.

14.1.1 Serializer data

The IpGBT transmitter can be programmed to transmit the following: a fixed pattern, PRBS sequence, clock sequence, or loop back incoming downlink frame. The data pattern to the serializer can be controlled by `ULSerTestPattern[3:0]` field in [\[0x118\] `ULDataSource0`](#) (page 234) register according to the table below.

Table 14.1: High speed serializer data source.

ULSerTestPattern[3:0]	Name	Description
4'd0	DATA	Normal mode of operation
4'd1	PRBS7	PRBS7 test pattern ($x_7 + x_6 + 1$)
4'd2	PRBS15	PRBS15 test pattern ($x_{15} + x_{14} + 1$)
4'd3	PRBS23	PRBS23 test pattern ($x_{23} + x_{18} + 1$)
4'd4	PRBS31	PRBS31 test pattern ($x_{31} + x_{28} + 1$)
4'd5	CLK5G12	5.12 GHz clock pattern (in 5Gbps mode it will produce only 2.56 GHz)
4'd6	CLK2G56	2.56 GHz clock pattern
4'd7	CLK1G28	1.28 GHz clock pattern
4'd8	CLK40M	40 MHz clock pattern
4'd9	DL-FRAME_10G24	Loop back, downlink frame repeated 4 times
4'd10	DL-FRAME_5G12	Loop back, downlink frame repeated 2 times, each bit repeated 2 times
4'd11	DL-FRAME_2G56	Loop back, downlink frame repeated 1 times, each bit repeated 4 times
4'd12	CONST_PATTERN	8 x DPDataPattern[31:0]
4'd13	Reserved	Reserved
4'd14	Reserved	Reserved
4'd15	Reserved	Reserved

If a constant pattern is transmitted, the user should ensure that the configuration word stored in the DPDataPattern register has equal number of ones and zeros, otherwise the output of the serializer will not be DC balanced.

14.1.2 Uplink data path test patterns

The data coming from each ePortRx group can be replaced with test patterns. The test pattern for each group can be controlled independently by field `ULGnDataSource[2:0]` ([0x118] `ULDDataSource0` (page 234), [0x119] `ULDDataSource1` (page 235), [0x11a] `ULDDataSource2` (page 236), [0x11b] `ULDDataSource3` (page 237), [0x11c] `ULDDataSource4` (page 237)), where n is group number, according to:

Table 14.2: Uplink group data source

ULGnDataSource[2:0]	Name	Description
3'd0	EPORTRX_DATA	Normal mode of operation, data from ePortRx
3'd1	PRBS7	PRBS7 test pattern
3'd2	BIN_CNTR_UP	Binary counter counting up
3'd3	BIN_CNTR_DOWN	Binary counter counting down
3'd4	CONST_PATTERN	Constant pattern (DPDataPattern[31:0])
3'd5	CONST_PATTERN_INV	Constant pattern inverted (~DPDataPattern[31:0])
3'd6	DLDATA_LOOPBACK	Loop back, downlink frame data
3'd7	Reserved	Reserved

The test pattern generator generates 16/32 bits in each clock cycle when the chip operates in low/high speed mode respectively. The pattern does not depend on the data rate selected for the corresponding ePortRx group.

It should be noticed that, due to the presence of the scrambler, when a fixed pattern is used to test the data transmission in fact a pseudo random sequence is actually transmitted over the high speed frame. However, since the scrambler can

be bypassed it is also possible to send the raw fixed pattern. Since the fixed pattern transmitted is DC balanced the receiver will have no problem locking to the incoming data stream.

14.1.3 Downlink data path test patterns

The data coming from the downlink frame to each ePortTx group can be replaced with test patterns. The test pattern for each group can be controlled independently by field `DLGnDataSource[1:0]` in the [\[0x11d\] ULDataSource5](#) (page 238) register, where n is group number, according to:

Table 14.3: Downlink group data source

<code>DLGnDataSource[1:0]</code>	Name	Description
2'd0	LINK_DATA	Normal mode of operation, data from ePortRx
2'd1	PRBS7	PRBS7 pattern on each channel
2'd2	BIN_CNTR_UP	Binary counter counting up on each channel
2'd3	CONST_PATTERN	Constant pattern

The test pattern generator generates 8 bits in each clock cycle. Contrary to the data generator for the uplink data path, the pattern generated by the downlink pattern generator depends on the data rate selected for the corresponding ePortTx group. It has several implications for various modes and data rates. When operating in PRBS7 mode, the user will see a valid PRBS7 sequence on all channels available in a given data rate. On the other hand, the user will have a variable number of bits available in the constant pattern mode (2 bits per channel for 80 Mbps data rate, ..., 8 bit per channel for 320 Mbps data rate).

14.1.4 PRBS generators for ePortRx group

There is a PRBS generator included at each ePortRx input. It is schematically depicted on [Fig. 14.3](#).

Each generator can be independently enabled by bits `EPRXnnPrbsEnable` in `EPRXPRBSx` registers. All PRBS generators are clocked with the same clock. The clock comes from channel 0 of phase-shifter (see [Section 10](#))

The clock frequency has to be configured by the user accordingly to the data rate of the ePortRx group being tested. Moreover, the phase of the clock can be adjusted to test the phaseAligner operation.

Example test case:

1. Configure **ePortRx** group g: - select a data rate (`EPRXgDataRate[1:0]`) - enable at least one channel c (`EPRXgcEnable`) - select automatic track mode (`EPRXgTrackMode[1:0]`)
2. Configure **phase-shifter** channel 0:
 - select frequency which matches data rate of the ePortRx group (`PS0Freq[2:0]`)
 - select delay (`PS0Delay[8:0]`)
3. Enable PRBS7 generator for the enabled channel (`EPRXgcPrbsEnable`)
4. Initialize phase training of enabled channel (`EPRXgTrain[c]`)
5. Wait until the training is completed (`EPRXgChnLocked[c]`)
6. Check if the data being sent in the uplink frame is valid (can be done in the FPGA or using built in BERT checker described in the next section)
7. Check if the selected phase is reasonable (`EPRXgcCurrentPhase[3:0]`)
8. Make a phase jump (+/-1) by updating delay of phase-shifter
9. Repeat checks described in point 6) and 7)

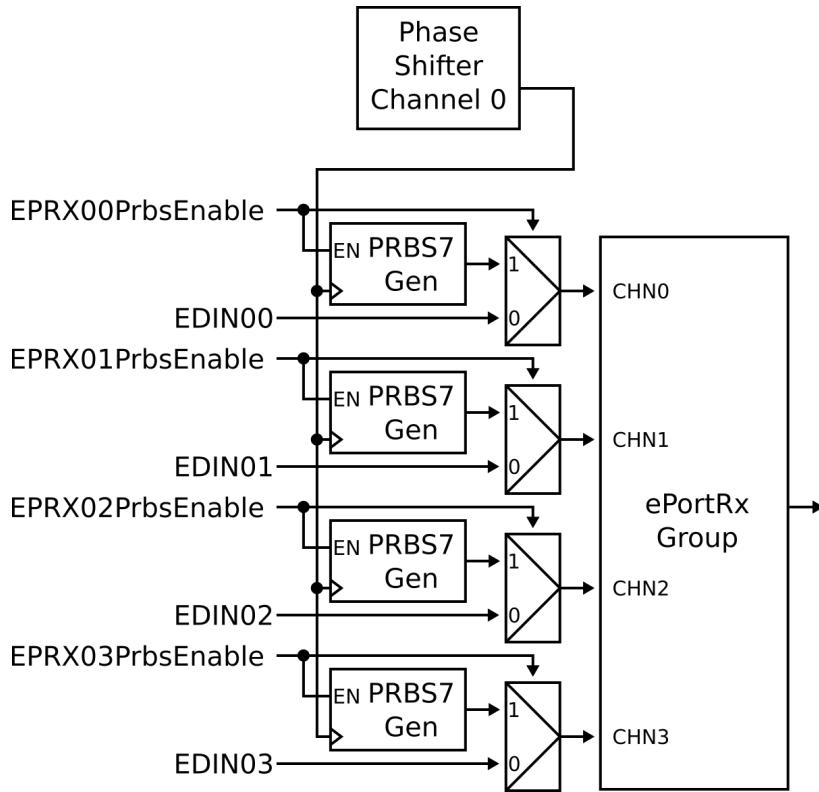


Fig. 14.3: PRBS generators in ePortRx

See registers:

- ePortRx: [\[0x0c4\] EPRX0Control](#) (page 208), [\[0x0c5\] EPRX1Control](#) (page 209), [\[0x0c6\] EPRX2Control](#) (page 209), [\[0x0c7\] EPRX3Control](#) (page 210), [\[0x0c8\] EPRX4Control](#) (page 211), [\[0x0c9\] EPRX5Control](#) (page 211), [\[0x0ca\] EPRX6Control](#) (page 212), [\[0x105\] EPRXTrain10](#) (page 229), [\[0x106\] EPRXTrain32](#) (page 230), [\[0x107\] EPRXTrain54](#) (page 230), [\[0x108\] EPRXTrainEc6](#) (page 230), [\[0x142\] EPRX0Locked](#) (page 247), [\[0x145\] EPRX1Locked](#) (page 247), [\[0x148\] EPRX2Locked](#) (page 248), [\[0x14b\] EPRX3Locked](#) (page 248), [\[0x14e\] EPRX4Locked](#) (page 249), [\[0x151\] EPRX5Locked](#) (page 250), [\[0x154\] EPRX6Locked](#) (page 250), [\[0x105\] EPRXTrain10](#) (page 229), [\[0x106\] EPRXTrain32](#) (page 230), [\[0x107\] EPRXTrain54](#) (page 230), [\[0x108\] EPRXTrainEc6](#) (page 230) .
- phase-shifter: [\[0x05c\] PS0Config](#) (page 143), [\[0x05d\] PS0Delay](#) (page 143).
- PRBS7 generator: [\[0x122\] EPRXPRBS3](#) (page 239), [\[0x123\] EPRXPRBS2](#) (page 239), [\[0x124\] EPRXPRBS1](#) (page 240), [\[0x125\] EPRXPRBS0](#) (page 240).

14.2 Test pattern checkers

The IpGBT has one pattern checker which can monitor various points along the data path as depicted on Fig. 14.4.

This architecture implies that only one data stream can be checked at any given time. The data stream to be checked is selected by [\[0x126\] BERTSource](#) (page 241) register. The register is divided into two parts: older bits (*BERTSource[7:4]*) allow coarse selection (particular up/downlink group) according to [Coarse BERT source](#) (page 112) while younger bits (*BERTSource[3:0]*) are used to select the channel.

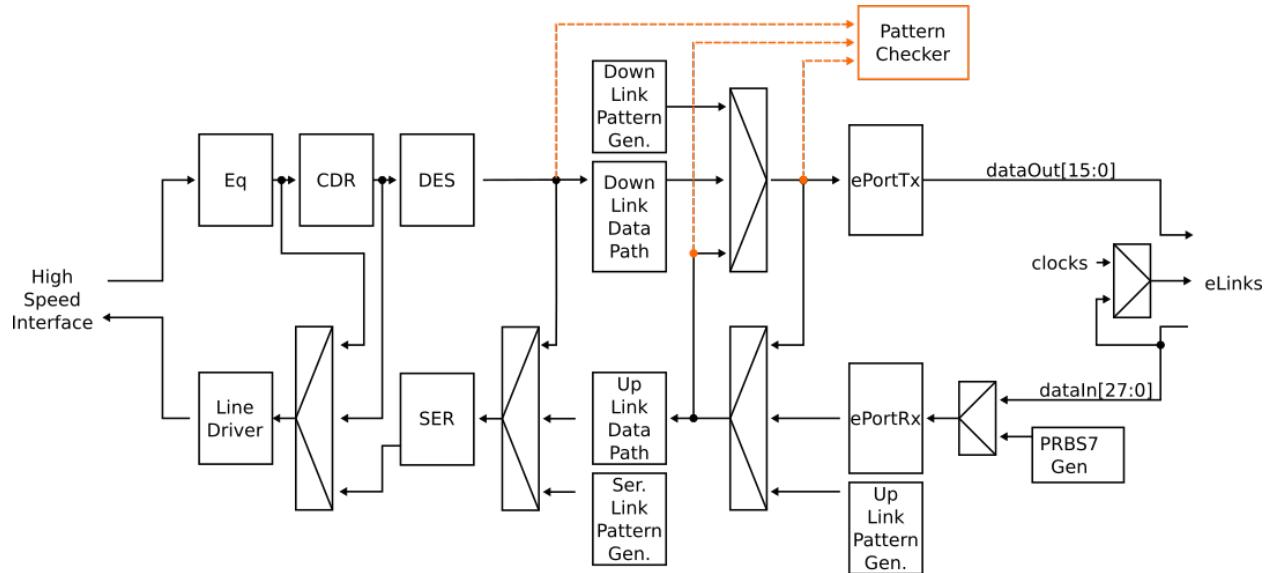


Fig. 14.4: IpGBT pattern checker

Table 14.4: Coarse BERT source

BERTSource[7:4]	Name	Description
4'd0	DISABLED	Checker disabled
4'd1	ULDG0	Uplink data group 0
4'd2	ULDG1	Uplink data group 1
4'd3	ULDG2	Uplink data group 2
4'd4	ULDG3	Uplink data group 3
4'd5	ULDG4	Uplink data group 4
4'd6	ULDG5	Uplink data group 5
4'd7	ULDG6	Uplink data group 6
4'd8	ULEC	Uplink data group EC
4'd9	DLDG0	Downlink data group 0
4'd10	DLDG1	Downlink data group 1
4'd11	DLDG2	Downlink data group 2
4'd12	DLDG3	Downlink data group 3
4'd13	DLEC	Downlink data group EC
4'd14	DLFRAME	Downlink deserializer frame

The duration of the measurement is expressed in the number of 40 MHz clock cycles and can be programmed according to Table 14.5. It should be noted, that accrual number of bits checked depends on the data source. For example, channel working at 80 Mbps produces only two bits per 40 MHz clock cycle, while channel working at 1.28 Gbps produces 32 bits per 40 MHz clock cycle.

Table 14.5: BER measurement time.

BERTMeasTime[7:4]	Name	Measurement time (clock cycles)
4'd0	BC_MT_2e5	2^5 (32)
4'd1	BC_MT_2e7	2^7 (128)
4'd2	BC_MT_2e9	2^9 (512)
4'd3	BC_MT_2e11	2^{11} (2k)
4'd4	BC_MT_2e13	2^{13} (8k)
4'd5	BC_MT_2e15	2^{15} (32k)
4'd6	BC_MT_2e17	2^{17} (128k)
4'd7	BC_MT_2e19	2^{19} (512k)
4'd8	BC_MT_2e21	2^{21} (2M)
4'd9	BC_MT_2e23	2^{23} (8M)
4'd10	BC_MT_2e25	2^{25} (32M)
4'd11	BC_MT_2e27	2^{27} (128M)
4'd12	BC_MT_2e29	2^{29} (512M)
4'd13	BC_MT_2e31	2^{31} (2G)
4'd14	BC_MT_2e33	2^{33} (8G)
4'd15	BC_MT_2e35	2^{35} (32G)

14.2.1 Uplink data checking

If one of uplink data groups is selected as a course data source, the fine data source should be set according to Table 14.6.

Table 14.6: BERT source.

BERTSource[3:0]	Name	Description
4'd0	UL_PRBS7_DR1_CHN0	Check PRBS7 sequence on channel 0 for data rate equal to 1
4'd1	UL_PRBS7_DR1_CHN1	Check PRBS7 sequence on channel 1 for data rate equal to 1
4'd2	UL_PRBS7_DR1_CHN2	Check PRBS7 sequence on channel 2 for data rate equal to 1
4'd3	UL_PRBS7_DR1_CHN3	Check PRBS7 sequence on channel 3 for data rate equal to 1
4'd4	UL_PRBS7_DR2_CHN0	Check PRBS7 sequence on channel 0 for data rate equal to 2
4'd5	UL_PRBS7_DR2_CHN2	Check PRBS7 sequence on channel 2 for data rate equal to 2
4'd6	UL_PRBS7_DR3_CHN0	Check PRBS7 sequence on channel 0 for data rate equal to 3
4'd7	UL_FIXED	Check the data against constant pattern

One should notice that the BERT checker is not aware of ePortRxGroup configuration and therefore the user has to select *BERTSource[3:0]* corresponding to the actual data rate configured for a selected group. Moreover, the overall number of bits checked will depend on the data rate as the measurement time is specified in term of number of 40 MHz clock cycles. When in *UL_FIXED* mode, the checker will compare deserialized bits coming from the ePortRxGroup (regardless the data rate) and compare them with a fixed pattern stored in *DPDataPattern[31:0]*. One should be aware, that the checker will not align the incoming data trying to find a match. It implies that the feature can be used the data transmission only in a fixed latency system.

Usage example

Lets consider a situation in which the IpGBT is operating in 10 Gbps mode and ePortRxGroup3 is working at 320 Mbps (data rate equal to 1, 4 channels available, each channel produces 8 bits per 40 MHz clock cycle). To check if channel 2 receives a valid PRBS7 sequence one should:

```

// select the data source for the measurement
writeReg(BERTSOURCE, {BC_ULDG3, UL_PRBS7_DR1_CHN2} << BERTSOURCE_BERTSOURCE_
↪of);

// set the measurement time to 2**7 clock cycles (128 * 25 ns = 3.2 μs)
config = BC_MT_2e7 << Lpgbt.BERTCONFIG_BERTMEASTIME_of

// Channel working at 320 Mbps produces 8 bits per 40 MHz clock cycle
bits_per_clock_cycle = 8

bits_checked = 2**7 * bits_per_clock_cycle

// start the measurement
writeReg(BERTCONFIG, config | BERTCONFIG_BERTSTART_bm)

do
    status=readReg(BERTSTATUS)
    while !(status & BERTSTATUS_BERTDONE_bm)

    if status & BERTSTATUS_BERTPRBSERRORFLAG_bm:
        # stop the measurement by deasserting the start bit
        write_reg(BERTCONFIG, 8'b0)
        raise Exception("BERT error flag (there was not data on the input)")

    // read the result
    bertResult[7:0] = readReg(BERTRESULT0)
    bertResult[15:8] = readReg(BERTRESULT1)
    bertResult[23:16] = readReg(BERTRESULT2)
    bertResult[31:24] = readReg(BERTRESULT3)
    bertResult[39:32] = readReg(BERTRESULT4)

    // stop the measurement by deasserting the start bit
    write_reg(BERTCONFIG, 8'b0)

    // calculate Bit Error Rate
    ber = bertResult / bits_checked

```

14.2.2 Downlink data checking

If one of downlink data groups is selected as a course data source, the fine data source should be set according to Table 14.7.

Table 14.7: BERT source.

BERTSource[3:0]	Name	Description
4'd0	DL_PRBS7_DR1_CHN0	Check PRBS7 sequence on channel 0 for data rate equal to 1
4'd1	DL_PRBS7_DR1_CHN1	Check PRBS7 sequence on channel 1 for data rate equal to 1
4'd2	DL_PRBS7_DR1_CHN2	Check PRBS7 sequence on channel 2 for data rate equal to 1
4'd3	DL_PRBS7_DR1_CHN3	Check PRBS7 sequence on channel 3 for data rate equal to 1
4'd4	DL_PRBS7_DR2_CHN0	Check PRBS7 sequence on channel 0 for data rate equal to 2
4'd5	DL_PRBS7_DR2_CHN2	Check PRBS7 sequence on channel 2 for data rate equal to 2
4'd6	DL_PRBS7_DR3_CHN0	Check PRBS7 sequence on channel 0 for data rate equal to 3
4'd7	DL_FIXED	Check the data against constant pattern

One should notice that the BERT checker is not aware of ePortTxGroup configuration and therefore the user has to

select *BERTSource[3:0]* corresponding to the actual data rate configured for a selected group. Moreover, the overall number of bits checked will depend on the data rate as the measurement time is specified in term of number of 40 MHz clock cycles. When in *UL_FIXED* mode, the checker will compare deserialized bits coming from the ePortTxGroup (regardless the data rate) and compare them with a fixed pattern stored in *DPPattern[7:0]*. One should be aware, that the checker will not align the incoming data trying to find a match. It implies that this feature can only be used with a fixed latency data transmission system.

Usage example

Lets consider a situation in which ePortTxGroup3 is working at 160 Mbps (data rate equal to 2, 2 channels available, each channel produces 4 bits per 40 MHz clock cycle). To check if channel 2 receives a valid PRBS7 sequence one should:

```
// select the data source for the measurement
writeReg(BERTSOURCE, {BC_DLDG3, DL_PRBS7_DR2_CHN2} << BERTSOURCE_BERTSOURCE_
↪of);

// set the measurement time to 2**31 clock cycles (2.1G * 25 ns = 53.6s)
config = BC_MT_2e13 << Lpgbt.BERTCONFIG_BERTMEASTIME_of

// Channel working at 160 Mbps produces 4 bits per 40 MHz clock cycle
bits_per_clock_cycle = 4

bits_checked = 2**31 * bits_per_clock_cycle

// start the measurement

writeReg(BERTCONFIG, config | BERTCONFIG_BERTSTART_bm)

do
    status=readReg(BERTSTATUS)
    while !(status & BERTSTATUS_BERTDONE_bm)

    if status & BERTSTATUS_BERTPRBSERRORFLAG_bm:
        // stop the measurement by deasserting the start bit
        write_reg(BERTCONFIG, 8'b0)
        raise Exception("BERT error flag (there was not data on the input)")

    // read the result
    bertResult[7:0]    = readReg(BERTRESULT0)
    bertResult[15:8]   = readReg(BERTRESULT1)
    bertResult[23:16]  = readReg(BERTRESULT2)
    bertResult[31:24]  = readReg(BERTRESULT3)
    bertResult[39:32]  = readReg(BERTRESULT4)

    // stop the measurement by deasserting the start bit
    write_reg(BERTCONFIG, 8'b0)

    // calculate Bit Error Rate
    ber = bertResult / bits_checked
```

14.2.3 Deserializer data checking

The last group of data sources is related to the high-speed downlink frame. If the *DLF* is selected in *BERT-Source[7:4]* the fine data source should be set according to [Table 14.8](#).

Table 14.8: BERT source.

BERTSource[3:0]	Name	Description
4'd0	DLDATA_PRBS	•
4'd1	DLDATA_FIXED	Checks the group data in the down-link frame
4'd2	DLFRAME_PRBS7	PRBS7 (no header)
4'd3	DLFRAME_PRBS15	PRBS15 (no header)
4'd4	DLFRAME_PRBS23	PRBS23 (no header)
4'd5	DLFRAME_PRBS31	PRBS31 (no header)
4'd6	•	Reserved
4'd7	DLFRAME_FIXED	Check the data against constant pattern

If one wants to check PRBS sequence (with no IpGBT header in it) the frame aligner has to be disabled to prevent bit slips. This can be achieved by asserting bit *SKIPDisable* in [0x127] *BERTConfig* (page 241) register.

Usage example

Lets consider a situation in user sends a raw PRBS7 sequence on the downlink (not encapsulated in IpGBT frame). In order to perform BER measurement one could follow the example below:

```
# select the data source for the measurement
writeReg(BERTSOURCE, {BC_DLFRAME, DLFRAME_PRBS7} << BERTSOURCE_BERTSOURCE_
˓˓of);

# set the measurement time to 2**31 clock cycles (2.1G * 25 ns = 53.6s)
config = BC_MT_2e13 << Lpgbt.BERTCONFIG_BERTMEASTIME_of | BERTCONFIG_
˓˓SKIPDISABLE_bm

# Downlink frame contains 64 bits (2.56 Gbps)
bits_per_clock_cycle = 64

bits_checked = 2**31 * bits_per_clock_cycle

# start the measurement

writeReg(BERTCONFIG, config | BERTCONFIG_BERTSTART_bm)

do
    status=readReg(BERTSTATUS)
    while !(status & BERTSTATUS_BERTDONE_bm)

    if status & BERTSTATUS_BERTPRSERRORFLAG_bm:
        # stop the measurement by deasserting the start bit
        write_reg(BERTCONFIG, 8'b0)
        raise Exception("BERT error flag (there was not data on the input)")

// read the result
bertResult[7:0] = readReg(BERTRESULT0)
bertResult[15:8] = readReg(BERTRESULT1)
bertResult[23:16] = readReg(BERTRESULT2)
bertResult[31:24] = readReg(BERTRESULT3)
```

```

bertResult[39:32] = readReg(BERTRESULT4)

# stop the measurement by deasserting the start bit
write_reg(BERTCONFIG, 8'b0)

# calculate Bit Error Rate
ber = bertResult / bits_checked

```

14.3 Data loopbacks

When the IpGBT operates as a transceiver it is possible to implement extensive loopback tests.

14.3.1 High speed link loopbacks

As described in [High-Speed Line Driver](#) (page 39), the line driver input multiplexer is controlled by the signal **LD-DataSource[1:0]**, for details please see register [\[0x119\] ULDatasource1](#) (page 235).

14.3.2 Data loopbacks

Data fields for any of the uplink groups can be overwritten with loop back data as indicated in [Table 14.2](#) (DL-DATA_LOOPBACK). This feature allows to resend on the uplink data received on the downlink. As the uplink and downlink have various bandwidths, the the loopback is not one to one (as it was for GBTX chip). The data from all downlink groups form 32 bit long vector which can be looped back. Depending on the high speed serializer mode (5 or 10 Gbps) 16 or 32 bits from are transmitted.

14.3.3 ePorts loopbacks

The IpGBT offers also a possibility to realize a eLink loopback closer to the physical layer. As the IpGBT has the same number of data inputs and clock outputs (29), it is possible to route signal from data input (EDI) to clock output (ECLK).

This functionality is controlled by *EPCLKnFreq[2:0]* field. If *EPCLKnFreq[2:0]* is set to 3'd7 than output of the eRX for *EDIn* is connected to the input of eTX of *ECLKn*.

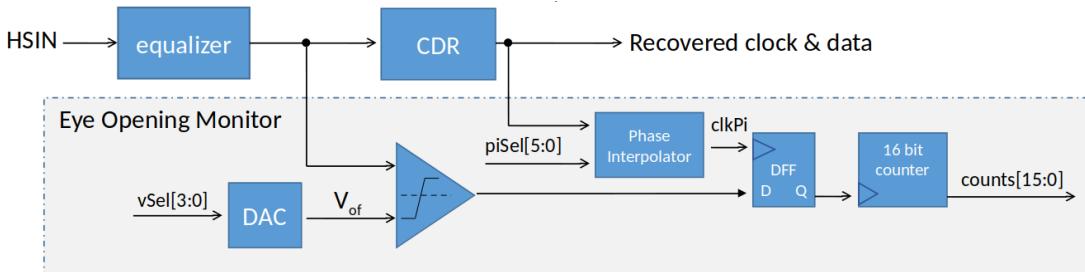
14.4 Eye Opening Monitor

The Eye Opening Monitor (EOM) block allows the user to make an on-chip Eye-Diagram measurement of the incoming 2.56 Gb/s high speed data. The EOM block monitors the signal at the output of the equalized block as depicted in [Fig. 14.5](#).

The eye scan does not affect the data transition and can be performed in the final system.

The eye-diagram opening will mostly depend on power supply voltage, temperate and the transmission medium. With the aid of the EOM, the user can optimize the equalizer's parameters, achieving the best eye-diagram in their specific operating condition. By default the EOM is disabled and thus does not consume any power, when in operation the extra current is consumed from the RX power supply (see [Table 18.14](#)).

The EOM block is not triplicated but was designed to endure a TID of 200 Mrad. The main purpose of the block is to understand the quality of the receiver's eye diagram over time, making this exercise only when the beam is off. The

Fig. 14.5: EOM architecture (inspired by [\[eom\]](#) (page 325))

EOM can only be exercised in RX or TRX mode once data transmission is valid as it is necessary for the CDR to be locked.

14.4.1 Working principle

The 2.56 Gbps data (*HSIN*) is compared with a static voltage (*V_{of}*) using a differential comparator. The comparator is sampled with a phase interpolated clock (*clkPi*). If the static voltage is within the *HSIN* voltage limits, the output of the comparator will toggle. If it is below, the output will have a static '0' and if it is above, the output will be a static '1'. The output of the comparator is fed to a 16-bit counter which can be read.

By scanning in the y-axis direction (voltage) and in the x-axis direction (time) it is possible to get an image of the eye diagram. The user has to read one sampling point at a time. The Fig. 14.6 depicts the working principle and Fig. 14.7 an example of the output that can be produced with the EOM.

The key block in the x-axis is the phase interpolation block. A 2.56 GHz phase interpolated clock is generated from the VCO's 5.12 GHz output clock with a step of ~6.1 ps in nominal conditions ($[0:1/(fvco*64):63/(fvco*64)]$).

The y-axis uses a differential comparator where $V_{comp} = Av \cdot (HSINP - HSINN) - (V_{ofp} - V_{ofn})$, V_{comp} being the output of the comparator, Av the gain and V_{of} the static voltage that is generated for the comparison. The static voltage ($V_{of} = V_{ofp} - V_{ofn}$) is generated by the means of a resistive divider. The step is 40 mV from -VDDRX up to VDDRX ($[-VDDRX/2:VDDRX/30:VDDRX/2]$).

The output of the comparator is sampled by the *clkPi* and this signal is fed to a counter.

14.4.2 Measurement flow

The user has to make a single (x,y) point measurement at a time. Before starting the measurement, the IgGBT has to be locked to the incoming data with valid data transmission. The read/write EOM registers are described in Section 15.2.5 and the read only registers in Section 15.3.10.

To setup the EOM it is necessary to configure the *EOMendOfCountSel[3:0]* field in the [\[0x114\] EOMConfigH](#) (page 233) register. This register controls the gating time for the measurement according to the formula:

$$GatingTime = 2^{selEndOfCount+1} \times 25\text{ ns}$$

For example, *EOMendOfCountSel[3:0] = 4'd3* sets the gating time to $16 \times 25\text{ ns} = 400\text{ ns}$.

The *EOMenable* bit in the [\[0x114\] EOMConfigH](#) (page 233) register powers up the EOM circuit and the *EOMstart* bit starts the measurement. Few milliseconds should be given between both signals to ensure all bias voltages have stabilized.

The x-axis is controlled by the *EOMphaseSel[5:0]* field in the [\[0x115\] EOMConfigL](#) (page 233) register and the y-axis by the *EOMvofSel[4:0]* field in the [\[0x116\] EOMvofSel](#) (page 233).

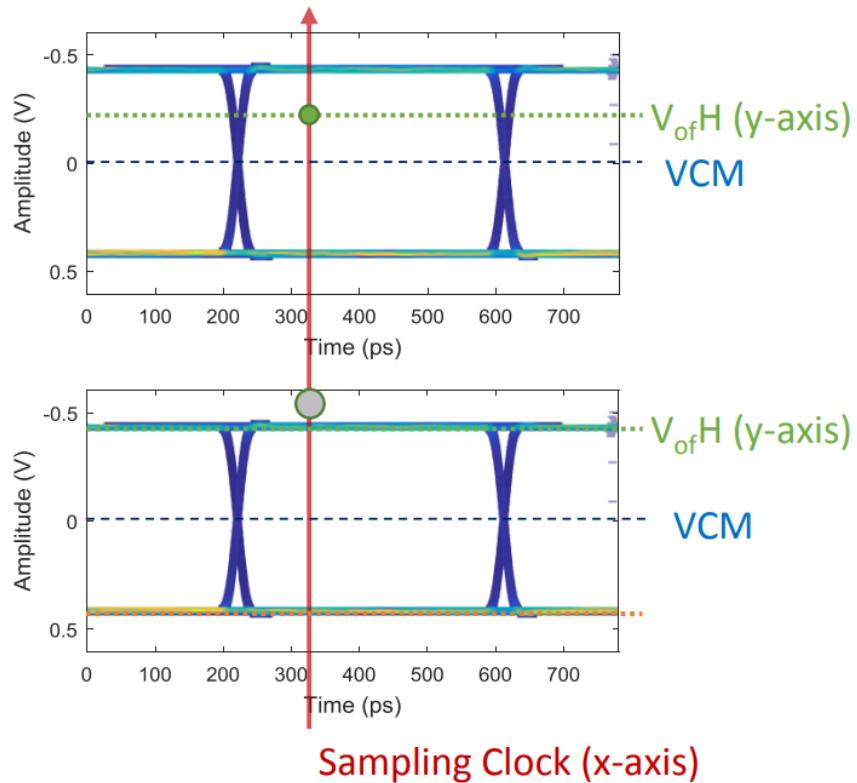


Fig. 14.6: EOM working principle

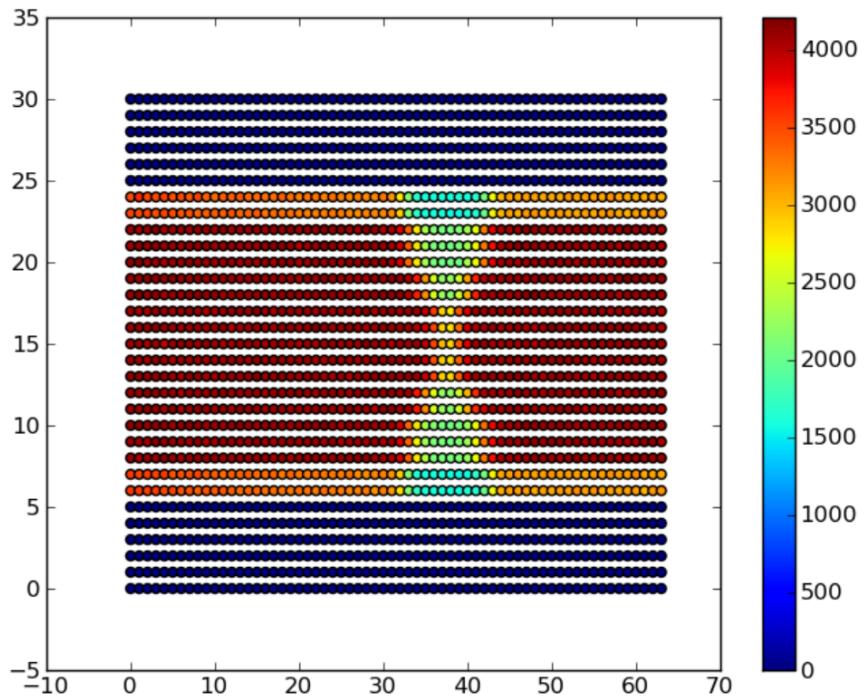


Fig. 14.7: EOM output example of the eye diagram

After the EOMStart goes high, the status of the EOM can be monitored by the read only registers. The EOMBusy goes high when the measurement starts, and EOMEnd goes high when the measurement finishes. This is to be used as an "handshake" with the user's measurement algorithm. The EOMsmState[1:0] field holds the EOM's state machine state which can be used for debugging.

The other two registers of interest are EOMcounterValue[15:0] which holds how many times the counter has ticked during the measurement and EOMCounter40M[15:0] which holds how many 40 MHz clock cycles have occurred during the gating time (debug).

The following pseudo-code should give a clear idea of the measurement flow.

```
// start lpGBT and wait for idle state
[...]

// EOM configuration (256 cycles = 6.4 µs)
config = EOMCONFIGH_EOMENABLE_bm | 4'd7 << EOMCONFIGH_EOMENDOFCOUNTSEL_of
writeReg(EOMCONFIGH, config)

array[64][31] eyeImage;

for(y_axis = 0; y_axis < 5'd31; y_axis++)
    // update yaxis
    writeReg(EOMVOFSEL, y_axis)
    for(x_axis = 0; x_axis < 6'd64; x_axis++)
        // update xaxis
        writeReg(EOMCONFIGL, x_axis)
        // wait few milliseconds
        sleep(5ms)

    // start measurement
    writeReg(EOMCONFIGH, config | EOMCONFIGH_EOMSTART_bm)

    // wait until measurement is finished
    do
        status=readReg(EOMSTATUS)
    while (status & EOMSTATUS_EOMBUSY_bm && !(status & EOMSTATUS_EOMEND_
        ↵bm) )

    counterValue[15:8]=readReg(EOMCOUNTER40MH);
    counterValue[7:0]=readReg(EOMCOUNTER40ML);

    // store measurement result
    eyeImage[x_axis][y_axis] = counterValue;

    // deassert EOMstart bit
    writeReg(EOMCONFIGH, config)
```

14.4.3 Testability

Bypassing the phase interpolated clock

In case of failure of the phase interpolated block, this block can be bypassed by selecting EOMBypassPhaseInterpolator = 1'b1. In order to work properly, the user will need a clock generator that allows phase deskewing and to follow the procedure below:

- Set the IpGBT in TRX or RX mode;

- Provide a 40 MHz clock to the IpGBT's refClk pin that **shares the same timebase** as the ref clock supplied to the back-end FPGA (ie, it is locked to the back-end clock);
- Ensure the refClk block is enabled and properly configured (REFCLKForceEnable = 1'b1 in [\[0x03b\] REFCLK](#) (page 138));
- Ensure IpGBT is locked and you have valid data transmission;
- Make the measurement.

Measuring the static voltage ramp using the built-in ADC

The INL and DNL of the static voltage can be measured using the internal ADC. It is necessary to keep the following items in mind when performing the measurement:

- Due to the architecture and common-centric layout the resistive divider is expected to have very good linearity. As the ADC cannot convert values over 1 V, the user has to extrapolate these values, if desired;
- Only the Vofp branch of the static voltage is measured ($V_{of} = V_{ofp} - V_{ofn}$); the expected range for this measurement is $[V_{DDRX}/2 : V_{DDRX}/30 : V_{DDRX}]$;
- This is just to ensure nothing has gone wrong during manufacturing and that the voltage steps are as expected.

14.5 Test outputs

The IpGBT chip has 6 test outputs, four of which are CMOS and two are differential. Each signal can be connected to one of a number of different internal signals. These are selected by the TOnSel configuration register (e.g. [\[0x133\] TO0Sel](#) (page 243)), where n is an index of the test output.

Table 14.9: TOnSelect

TOnSelect[7:0]	Signal
8'd0	1'b0
8'd1	1'b1
8'd2	clk40MA
8'd3	clk40MB
8'd4	clk40MA
8'd5	clk40MB
8'd6	clk40MC
8'd7	clk80MA
8'd8	clk80MB
8'd9	clk80MC
8'd10	clk160MA
8'd11	clk160MB
8'd12	clk160MC
8'd13	clk320MA
8'd14	clk320MB
8'd15	clk320MC
8'd16	clk640MA
8'd17	clk640MB
8'd18	clk640MC
8'd19	clk1G28A
8'd20	clk1G28B
8'd21	clk1G28C
8'd22	seuEvent

Continued on next page

Table 14.9 – continued from previous page

TOnSelect[7:0]	Signal
8'd23	PMClkOut
8'd24	endCounterRefClkV (voted)
8'd25	fromcdr_clkRef
8'd26	fromcdr_instlockPLL
8'd27	endCounterVCOV (voted)
8'd28	fusesRampEnable
8'd29	fusesPowerShort
8'd30	fusesPowerEnable
8'd31	fusesSclk
8'd32	fusesPgm
8'd33	fusesDin
8'd34	fusesCsb
8'd35	PS0DIILockedV (voted)
8'd36	PS1DIILockedV (voted)
8'd37	PS2DIILockedV (voted)
8'd38	PS3DIILockedV (voted)
8'd39	PS0dllLateV (voted)
8'd40	PS1dllLateV (voted)
8'd41	PS2dllLateV (voted)
8'd42	PS3dllLateV (voted)
8'd43	PS0dllOutRef
8'd44	PS1dllOutRef
8'd45	PS2dllOutRef
8'd46	PS3dllOutRef
8'd47	ePortRxDllInstantLock[0]
8'd48	ePortRxDllInstantLock[1]
8'd49	ePortRxDllInstantLock[2]
8'd50	ePortRxDllInstantLock[3]
8'd51	ePortRxDllInstantLock[4]
8'd52	ePortRxDllInstantLock[5]
8'd53	ePortRxDllInstantLock[6]
8'd54	ePortRxDllOutRef[0]
8'd55	ePortRxDllOutRef[1]
8'd56	ePortRxDllOutRef[2]
8'd57	ePortRxDllOutRef[3]
8'd58	ePortRxDllOutRef[4]
8'd59	ePortRxDllOutRef[5]
8'd60	ePortRxDllOutRef[6]
8'd61	downLinkFrame[60]
8'd62	downLinkFrame[61]
8'd63	downLinkFrame[62]
8'd64	downLinkFrame[63]
8'd65	ePortRxDataIn[0]
8'd66	ePortRxDataIn[1]
8'd67	ePortRxDataIn[2]
8'd68	ePortRxDataIn[3]
8'd69	ePortRxDataIn[4]
8'd70	ePortRxDataIn[5]
8'd71	ePortRxDataIn[6]

Continued on next page

Table 14.9 – continued from previous page

TOnSelect[7:0]	Signal
8'd72	ePortRxDataIn[7]
8'd73	ePortRxDataIn[8]
8'd74	ePortRxDataIn[9]
8'd75	ePortRxDataIn[10]
8'd76	ePortRxDataIn[11]
8'd77	ePortRxDataIn[12]
8'd78	ePortRxDataIn[13]
8'd79	ePortRxDataIn[14]
8'd80	ePortRxDataIn[15]
8'd81	ePortRxDataIn[16]
8'd82	ePortRxDataIn[17]
8'd83	ePortRxDataIn[18]
8'd84	ePortRxDataIn[19]
8'd85	ePortRxDataIn[20]
8'd86	ePortRxDataIn[21]
8'd87	ePortRxDataIn[22]
8'd88	ePortRxDataIn[23]
8'd89	ePortRxDataIn[24]
8'd90	ePortRxDataIn[25]
8'd91	ePortRxDataIn[26]
8'd92	ePortRxDataIn[27]
8'd93	PORA
8'd94	PORB
8'd95	PORC
8'd96	ePortRxDataInEc
8'd97	BODA
8'd98	BODB
8'd99	BODC
8'd100	i2cTransactionStartV (voted)
8'd101	i2cTransactionDoneV (voted)
8'd102	i2cmaster_go0_V (voted)
8'd103	i2cmaster_go1_V (voted)
8'd104	i2cmaster_go2_V (voted)
8'd105	skipCycleRaw
8'd106	skipCycleV (voted)
8'd107	rxReady
8'd108	txReadyV (voted)
8'd109	pllStateMachineLockedV (voted)
8'd110	EPortRxDllLockedV[0]
8'd111	EPortRxDllLockedV[1]
8'd112	EPortRxDllLockedV[2]
8'd113	EPortRxDllLockedV[3]
8'd114	EPortRxDllLockedV[4]
8'd115	EPortRxDllLockedV[5]
8'd116	EPortRxDllLockedV[6]
8'd117	ePortRxDllLateV[0]
8'd118	ePortRxDllLateV[1]
8'd119	ePortRxDllLateV[2]
8'd120	ePortRxDllLateV[3]

Continued on next page

Table 14.9 – continued from previous page

TOnSelect[7:0]	Signal
8'd121	ePortRxDIIILateV[4]
8'd122	ePortRxDIIILateV[5]
8'd123	ePortRxDIIILateV[6]
8'd124	frameAlignerReadyV (voted)

Test outputs use the same CMOS drivers as in the one used in the PIO block (described in Section 11). The drive strength can be controlled individually for each output by corresponding *TOnDS* bit in [\[0x139\] TODrivingStrength](#) (page 244) register. The eRX is used for the differential test outputs. The configuration of differential drivers can be changed in [\[0x139\] TODrivingStrength](#) (page 244), [\[0x13a\] TO4Driver](#) (page 244), [\[0x13b\] TO5Driver](#) (page 245), [\[0x13c\] TOPreEmp](#) (page 246) registers.

The current value of any of the test outputs can be readout by accessing register [\[0x1c9\] TOValue](#) (page 267). One should mention, that this feature should not be used for any systematic measurements. The timing for this register is not constrained, implying that it varies with PVT. It is recommended to use this feature to check if a given signal is zero, one, or is toggling. Moreover, one should be aware that the sampling of the signals is synchronous with the internal 40 MHz system clock, which means that the clock itself should always return the same value.

See registers: [\[0x133\] TO0Sel](#) (page 243), [\[0x134\] TO1Sel](#) (page 243), [\[0x135\] TO2Sel](#) (page 243), [\[0x136\] TO3Sel](#) (page 244), [\[0x137\] TO4Sel](#) (page 244), [\[0x138\] TO5Sel](#) (page 244), [\[0x139\] TODrivingStrength](#) (page 244), [\[0x13a\] TO4Driver](#) (page 244), [\[0x13b\] TO5Driver](#) (page 245), [\[0x13c\] TOPreEmp](#) (page 246).

14.6 TMR testing

The IpGBT logic uses Triple Modular Redundancy (TMR). When testing a TMR circuit there is always a risk that one of the triplicated sections is not working but the TMR logic makes it appear that all is performing well. It is only at the time when an SEU hits one of the functional circuits that the fault is revealed. To test successfully a TMR circuit requires that all the instances of the triplicated circuit be tested individually. To avoid this lengthy procedure, in the IpGBT is possible to stop individually any of the triplicated clocks. Stopping one of these clocks is like injecting a fault in one of the triplicated sections of the circuit and will result in malfunction if any of the other two circuits (fed by the other two clocks) is defective. To cover fully the TMR logic it is thus necessary to run the logic tests three times with a different triplicated section whose clock is stopped.

14.7 Single Event Upset monitoring

The IpGBT has Single Event Upset monitor.

- internal counter
- can be connected to the output

Typical use case:

```
// enable SEU monitor counter
writeReg(PROCESSANDSEUMONITOR, PROCESSANDSEUMONITOR_SEUENABLE_bm);

while True
    seuCounter[15:8] = readReg(SEUCOUNTH);
    seuCounter[ 7:0] = readReg(SEUCOUNTL);
    display("SEU Monitor Counter : %d", seuCounter);

// finish the measurement by deasserting enable bit
writeReg(PROCESSANDSEUMONITOR, 0);
```

Asynchronous output from the SEU monitor can be also connected to any of the test outputs:

```
// Output single event monitor to test output 0
writeReg(TO0SEL, TO_SEUEVENT);
```

14.8 Process monitors

The IpGBT has 4 built in ring oscillators placed in corners of the chip. The frequency of these oscillators can be measured precisely providing that the main PLL is locked (as it is used for the time reference).

This feature can be used to study process variation (lot-to-lot) or processing gradients present on the same chip. If the operating condition of chip (power supply and temperature) are constant, the variation of the frequency can be correlated with TID absorbed by the chip.

Typical use case:

```
// select data source and start the measurement by asserting PMENABLE bit
chn=2
writeReg(PROCESSANDSEUMONITOR, chn<<PROCESSANDSEUMONITOR_PMCHANNEL_of | ↴
          PROCESSANDSEUMONITOR_PMENABLE_bm);

// wait until the measurement is done
do
    status = readReg(PROCESSMONITORSTATUS);
    while ( not status&PROCESSMONITORSTATUS_PMDONE_bm )

// read measurement result
freq[23:16]=readReg(PMFREQA);
freq[15: 8]=readReg(PMFREQB);
freq[ 7: 0]=readReg(PMFREQC);

display("PM frequency : %d", freq);

// finish the measurement by deserting enable bit
writeReg(PROCESSANDSEUMONITOR, 0);
```

Frequency of any ring oscillator can be also monitored using test output pin:

```
// Output process monitor frequency to test output 0
writeReg(TO0SEL, TO_PMCLKOUT);
```


REGISTER MAP

15.1 Read/Write/Fuse

15.1.1 CHIPID

[0x000] CHIPID0

Stores bits 31:24 of the CHIPID

- **Bit 7:0 - ChipID[31:24]** - Bits 31:24 of the CHIPID

See also: [\[0x001\] CHIPID1](#) (page 127), [\[0x002\] CHIPID2](#) (page 127), [\[0x003\] CHIPID3](#) (page 127)

[0x001] CHIPID1

Stores bits 23:16 of the CHIPID

- **Bit 7:0 - ChipID[23:16]** - Bits 23:16 of the CHIPID

See also: [\[0x000\] CHIPID0](#) (page 127), [\[0x002\] CHIPID2](#) (page 127), [\[0x003\] CHIPID3](#) (page 127)

[0x002] CHIPID2

Stores bits 15:8 of the CHIPID

- **Bit 7:0 - ChipID[15:8]** - Bits 15:8 of the CHIPID

See also: [\[0x000\] CHIPID0](#) (page 127), [\[0x001\] CHIPID1](#) (page 127), [\[0x003\] CHIPID3](#) (page 127)

[0x003] CHIPID3

Stores bits 7:0 of the CHIPID

- **Bit 7:0 - ChipID[7:0]** - Bits 7:0 of the CHIPID

See also: [\[0x000\] CHIPID0](#) (page 127), [\[0x001\] CHIPID1](#) (page 127), [\[0x002\] CHIPID2](#) (page 127)

[0x004] USERID0

Stores bits 31:24 of the USERID

- **Bit 7:0 - UserID[31:24]** - Bits 31:24 of the USERID

[0x005] USERID1

Stores bits 23:16 of the USERID

- **Bit 7:0 - UserID[23:16]** - Bits 23:16 of the USERID

[0x006] USERID2

Stores bits 15:8 of the USERID

- **Bit 7:0 - UserID[15:8]** - Bits 15:8 of the USERID

[0x007] USERID3

Stores bits 7:0 of the USERID

- **Bit 7:0 - UserID[7:0]** - Bits 7:0 of the USERID

15.1.2 Calibration Data

[0x008] DACCal0

Calibration data for the voltage DAC. Usage is TBD.

- **Bit 7:0 - DACCalMinCode[7:0]** - Calibration data for the voltage DAC. Usage is TBD.

[0x009] DACCal1

Calibration data for the voltage DAC. Usage is TBD.

- **Bit 7:0 - DACCalMaxCode[7:0]** - Calibration data for the voltage DAC. Usage is TBD.

[0x00a] DACCal2

Calibration data for the voltage DAC. Usage is TBD.

- **Bit 7:4 - DACCalMinCode[11:8]** - Calibration data for the voltage DAC. Usage is TBD.
- **Bit 3:0 - DACCalMaxCode[11:8]** - Calibration data for the voltage DAC. Usage is TBD.

[0x00b] ADCCal0

Calibration data for the ADC. Usage is TBD.

- **Bit 7:0 - ADCCalGain2SeHigh[7:0]** - Calibration data for the ADC. Usage is TBD.

[0x00c] ADCCal1

Calibration data for the ADC. Usage is TBD.

- **Bit 7:0 - ADCCalGain2SeLow[7:0]** - Calibration data for the ADC. Usage is TBD.

[0x00d] ADCCal2

Calibration data for the ADC. Usage is TBD.

- Bit 7:4 - ADCCalGain2SeHigh[12:8] - Calibration data for the ADC. Usage is TBD.
- Bit 3:0 - ADCCalGain2SeLow[12:8] - Calibration data for the ADC. Usage is TBD.

[0x00e] ADCCal3

Calibration data for the ADC. Usage is TBD.

- Bit 7:0 - ADCCalGain2DifHigh[7:0] - Calibration data for the ADC. Usage is TBD.

[0x00f] ADCCal4

Calibration data for the ADC. Usage is TBD.

- Bit 7:0 - ADCCalGain2DifLow[7:0] - Calibration data for the ADC. Usage is TBD.

[0x010] ADCCal5

Calibration data for the ADC. Usage is TBD.

- Bit 7:4 - ADCCalGain2DifHigh[12:8] - Calibration data for the ADC. Usage is TBD.
- Bit 3:0 - ADCCalGain2DifLow[12:8] - Calibration data for the ADC. Usage is TBD.

[0x011] ADCCal6

Calibration data for the ADC. Usage is TBD.

- Bit 7:0 - ADCCalGain4DifHigh[7:0] - Calibration data for the ADC. Usage is TBD.

[0x012] ADCCal7

Calibration data for the ADC. Usage is TBD.

- Bit 7:0 - ADCCalGain4DifLow[7:0] - Calibration data for the ADC. Usage is TBD.

[0x013] ADCCal8

Calibration data for the ADC. Usage is TBD.

- Bit 7:4 - ADCCalGain4DifHigh[12:8] - Calibration data for the ADC. Usage is TBD.
- Bit 3:0 - ADCCalGain4DifLow[12:8] - Calibration data for the ADC. Usage is TBD.

[0x014] ADCCal9

Calibration data for the ADC. Usage is TBD.

- Bit 7:0 - ADCCalGain8DifHigh[7:0] - Calibration data for the ADC. Usage is TBD.

[0x015] ADCCal10

Calibration data for the ADC. Usage is TBD.

- **Bit 7:0 - ADCCalGain8DifLow[7:0]** - Calibration data for the ADC. Usage is TBD.

[0x016] ADCCal11

Calibration data for the ADC. Usage is TBD.

- **Bit 7:4 - ADCCalGain8DifHigh[12:8]** - Calibration data for the ADC. Usage is TBD.
- **Bit 3:0 - ADCCalGain8DirfLow[12:8]** - Calibration data for the ADC. Usage is TBD.

[0x017] ADCCal12

Calibration data for the ADC. Usage is TBD.

- **Bit 7:0 - ADCCalGain16DifHigh[7:0]** - Calibration data for the ADC. Usage is TBD.

[0x018] ADCCal13

Calibration data for the ADC. Usage is TBD.

- **Bit 7:0 - ADCCalGain16DifLow[7:0]** - Calibration data for the ADC. Usage is TBD.

[0x019] ADCCal14

Calibration data for the ADC. Usage is TBD.

- **Bit 7:4 - ADCCalGain16DifHigh[12:8]** - Calibration data for the ADC. Usage is TBD.
- **Bit 3:0 - ADCCalGain16DirfLow[12:8]** - Calibration data for the ADC. Usage is TBD.

[0x01a] TEMPCalH

Calibration data for the temperature sensor.

- **Bit 7:0 - TEMPCal[15:8]** - Calibration data for the temperature sensor. Usage is TBD.

[0x01b] TEMPCalL

Calibration data for the temperature sensor.

- **Bit 7:0 - TEMPCal[7:0]** - Calibration data for the temperature sensor. Usage is TBD.

[0x01c] VREFCNTR

Voltage reference control.

- **Bit 7 - VREFEnable** - Enable internal voltage reference.
- **Bit 5:0 - VREFTune[5:0]** - Tuning word for internal voltage reference.

[0x01d] CURDACCALH

Calibration data for current DAC. Usage is TBD.

- **Bit 7:0 - CURDACCAL[15:8]** - Calibration data for current DAC. Usage is TBD.

[0x01e] CURDACCALL

Calibration data for current DAC. Usage is TBD.

- **Bit 7:0 - CURDACCAL[7:0]** - Calibration data for current DAC. Usage is TBD.

[0x01f] EPRXLOCKFILTER

Lock filter settings for DLL in ePortRxGroups.

- **Bit 7:4 - EPRXLockThreshold[3:0]** - Sets the lock threshold value of the instant lock low pass filter for ePortRx DLL's (default: 5)
- **Bit 3:0 - EPRXReLockThreshold[3:0]** - Sets the relock threshold value of the instant lock low pass filter for ePortRx DLL's (default: 5)

[0x020] CLKGConfig

- **Bit 7:4 - CLKGCalibrationEndOfCount[3:0]** - Selects the VCO calibration race goal in number of clock cycles between refClk (refClkCounter) and vco_40MHz (vcoClkCounter) ($2^{\text{CLKGCalibrationEndOfCount}[3:0]} + 1$); default: 12
- **Bit 3:0 - CLKGBiasGenConfig[3:0]** - Bias DAC for the charge pumps [0 : 8 : 120] uA; default: 8

[0x021] CLKGConfig1

- **Bit 7 - CDRControlOverrideEnable** - Enables the control override of the state machine; default: 0
- **Bit 6 - CLKGDisableFrameAlignerLockControl** - Disables the use of the frame aligner's lock status; default: 0
- **Bit 5 - CLKGCDRRes** - CDR's filter resistance; default: 1 when in RX/TRX mode, 0 when in TX mode
- **Bit 4 - CLKGVcoRailMode** - VCO rail mode; [0: voltage mode, fixed to VDDRX; 1: current mode, value selectable using CLKGVcoDAC (default)]
- **Bit 3:0 - CLKGVcoDAC[3:0]** - Current DAC for the VCO [0: 0.470 : 7.1] mA; default: 8

15.1.3 Clock Generator**[0x022] CLKGPILRes**

- **Bit 7:4 - CLKGPILResWhenLocked[3:0]** - PLL's filter resistance when PLL is locked [$R = 1/2 * 79.8k / \text{CONFIG}$] Ohm; default: 4; set to 0 if RX or TRX mode
- **Bit 3:0 - CLKGPILRes[3:0]** - PLL's filter resistance when PLL is locking [$R = 1/2 * 79.8k / \text{CONFIG}$] Ohm; default: 4; set to 0 if RX or TRX mode

[0x023] CLKGPLLIntCur

- **Bit 7:4 - CLKGPLLIntCurWhenLocked[3:0]** - PLL's integral current path when in locked state [0 : 1.1 : 8] uA; default: 5
- **Bit 3:0 - CLKGPLLIntCur[3:0]** - PLL's integral current path when in locking state [0 : 1.1 : 8] uA; default: 5

[0x024] CLKGPLLPropCur

- **Bit 7:4 - CLKGPLLPropCurWhenLocked[3:0]** - PLL's proportional current path when in locked state [0 : 5.46 : 82] uA; default: 5
- **Bit 3:0 - CLKGPLLPropCur[3:0]** - PLL's proportional current path when in locking state [0 : 5.46 : 82] uA; default: 5

[0x025] CLKGCDRPropCur

- **Bit 7:4 - CLKGCDRPropCurWhenLocked[3:0]** - CDR's Alexander phase detector proportional current path when in locked state [0 : 5.46 : 82] uA; default: 5
- **Bit 3:0 - CLKGCDRPropCur[3:0]** - CDR's Alexander phase detector proportional current path when in locking state [0 : 5.46 : 82] uA; default: 5

[0x026] CLKGCDRIntCur

- **Bit 7:4 - CLKGCDRIntCurWhenLocked[3:0]** - CDR's Alexander phase detector integral current path when in locked state [0 : 5.46 : 82] uA; default: 5
- **Bit 3:0 - CLKGCDRIntCur[3:0]** - CDR's Alexander phase detector integral integral current path when in locking state [0 : 5.46 : 82] uA; default: 5

[0x027] CLKGCDRFFPropCur

- **Bit 7:4 - CLKGCDRFeedForwardPropCurWhenLocked[3:0]** - CDR's proportional feed forward current path when in locked state [0 : 5.46 : 82] uA; default: 5
- **Bit 3:0 - CLKGCDRFeedForwardPropCur[3:0]** - CDR's proportional feed forward current path when in locking state [0 : 5.46 : 82] uA; default: 5

[0x028] CLKGFLLIntCur

- **Bit 7:4 - CLKGFLLIntCurWhenLocked[3:0]** - CDR's frequency detector charge pump when in locked state [0 : 5.46 : 82] uA; default: 0
- **Bit 3:0 - CLKGFLLIntCur[3:0]** - CDR's frequency detector charge pump when in locking state [0 : 5.46 : 82] uA; default: 5

[0x029] CLKGFFCAP

- **Bit 7 - CDRCOConnectCDR** - Enables the connectCDR switch [0 - disable, 1 - enable] (only when CDRCControlOverrideEnable is 1)

- **Bit 6 - CLKGCapBankOverrideEnable** - Enables the override of the capacitor search during VCO calibration [0 - disable, 1 - enable]; default: 0
- **Bit 5:3 - CLKGFeedForwardCapWhenLocked[2:0]** - CDR's feed forward filter's capacitance when in locked state [0: 44 : 308] fF; default: 3
- **Bit 2:0 - CLKGFeedForwardCap[2:0]** - CDR's feed forward filter's capacitance when in locking state [0: 44 : 308] fF; default: 3

[0x02a] CLKGCntOverride

- **Bit 7 - CLKGCOoverrideVc** - Forces the VCO's control voltage to be in mid range [0 - disable, 1 - enable] (only when CDRCControlOverrideEnable is 1)
- **Bit 6 - CDRCORefClkSel** - Forces the reference clock selection for the VCO calibration [0 - data/4, 1 - external refClk] (only when CDRCControlOverrideEnable is 1)
- **Bit 5 - CDRCOEnablePLL** - Enables the enablePLL switch [0 - disable, 1 - enable] (only when CDRCControlOverrideEnable is 1)
- **Bit 4 - CDRCOEnableFD** - Enables the frequency detector [0 - disable, 1 - enable] (only when CDRCControlOverrideEnable is 1)
- **Bit 3 - CDRCOEnableCDR** - Enables the enableCDR switch [0 - disable, 1 - enable] (only when CDRCControlOverrideEnable is 1)
- **Bit 2 - CDRCODisDataCounterRef** - Enables the data/4 ripple counter [1 - disable, 0 - enable] (only when CDRCControlOverrideEnable is 1)
- **Bit 1 - CDRCODisDESvbiasGen** - Enables the vbias for the CDR [1 - disable, 0 - enable] (only when CDRCControlOverrideEnable is 1)
- **Bit 0 - CDRCOConnectPLL** - Enables the connectPLL switch [0 - disable, 1 - enable] (only when CDRCControlOverrideEnable is 1)

[0x02b] CLKGOVERRIDEcapBank

- **Bit 7:0 - CLKGCapBankSelect[7:0]** - Selects the capacitor bank value for the VCO [check attached table] (only when CLKGCapBankOverrideEnable is 1); default: n/a

[0x02c] CLKGWaitTime

- **Bit 7:4 - CLKGwaitCDRTIME[3:0]** - ljCDR state machine waiting for lock, RX/TRX mode, ($16'h0001 << \text{waitCDRTIME}$); (only when CLKGDisableFrameAlignerLockControl == 1); default: 8
- **Bit 3:0 - CLKGwaitPLLTime[3:0]** - ljCDR state machine waiting for lock, TX mode, ($16'h0001 << \text{waitPLLTime}$); (only when lfEnable == 0); default: 8

[0x02d] CLKGLFConfig0

- **Bit 7 - CLKGLockFilterEnable** - Enables the lock filter on the instant lock signal; only in TX mode [0 - disable, 1 enable]; default: 1
- **Bit 4 - CLKGCapBankSelect[8]** - Selects the capacitor bank value for the VCO [check attached table] (only when CLKGCapBankOverrideEnable is 1); default: n/a

- **Bit 3:0 - CLKGLockFilterLockThrCounter[3:0]** - Sets the lock threshold value of the instant lock low pass filter ($16'h0001 \ll \text{lfLockThrCounter}$); only in TX mode; default: after SEU

[0x02e] CLKGLFConfig1

- **Bit 7:4 - CLKGLockFilterReLockThrCounter[3:0]** - Sets the relock threshold value of the instant lock low pass filter ($16'h0001 \ll \text{lfReLockThrCounter}$); only in TX mode; default: after SEU
- **Bit 3:0 - CLKGLockFilterUnLockThrCounter[3:0]** - Sets the unlock threshold value of the instant lock low pass filter ($16'h0001 \ll \text{lfUnLockThrCounter}$); only in TX mode; default: after SEU

[0x02f] FAMaxHeaderFoundCount

Frame aligner configuration register.

- **Bit 7:0 - FAMaxHeaderFoundCount[7:0]** - The number of consecutive valid frame headers that have to be detected before frame lock is assumed.

[0x030] FAMaxHeaderFoundCountAfterNF

Frame aligner configuration register.

- **Bit 7:0 - FAMaxHeaderFoundCountAfterNF[7:0]** - The number of consecutive detected valid headers to be found to clear the invalid frame count. The reset can happen only if the number of detected invalid headers does not exceed the FAMaxHeaderNotFoundCount[7:0].

[0x031] FAMaxHeaderNotFoundCount

Frame aligner configuration register.

- **Bit 7:0 - FAMaxHeaderNotFoundCount[7:0]** - The maximum number of invalid headers (consecutive or not) that can be received before the frame is considered misaligned.

[0x032] FAFAMaxSkipCycleCountAfterNF

Frame aligner configuration register.

- **Bit 7:0 - FAMaxSkipCycleCountAfterNF[7:0]** -

[0x033] PSDIIConfig

Configuration for phase-shifter DLL

- **Bit 7:4 - EPRXUnLockThreshold[3:0]** - Sets the unlock threshold value of the instant lock low pass filter for ePortRx DLL's (default: 5)
- **Bit 3:2 - PSDLLConfirmCount[1:0]** - Number of clock cycles (in the 40 MHz clock domain) to confirm locked state.

PSDLLConfirmCount[1:0]	Number of clock cycles
2'd0	1
2'd1	4
2'd2	16
2'd3	31

- **Bit 1:0 - PSDllCurrentSel[1:0]** - Current for the DLL charge pump

PSDIICurrentSel[1:0]	Current [uA]
2'd0	1
2'd1	2
2'd2	4
2'd3	8

[0x034] EPRXDIIConfig

DLL configuration register

- **Bit 7:6 - EPRXDIICurrent[1:0]** - Current for the DLL charge pump

EPRXDIICurrent[1:0]	Current [uA]
2'd0	1
2'd1	2
2'd2	4
2'd3	8

- **Bit 5:4 - EPRXDLLConfirmCount[1:0]** - Number of clock cycles (in the 40 MHz clock domain) to confirm locked state.

EPRXDLLConfirmCount[1:0]	Number of clock cycles
2'd0	1
2'd1	4
2'd2	16
2'd3	31

- **Bit 3 - EPRXDLLFSMClkAlwaysOn** - Force clock of ePortRx DLL state machine to be always enabled (disables clock gating).
- **Bit 2 - EPRXDLLCoarseLockDetection** - Use coarse detector for the DLL lock condition.
- **Bit 1 - EPRXEnableReInit** - Allow re-initialization in ePortRxGroup when the tuning is out of range.
- **Bit 0 - EPRXDataGatingEnable** - Enable data gating.

[0x035] FORCEEnable

Enables user to enable specific sub-circuits regardless of the operation mode

- **Bit 7 - ForceTxEnable** - Enable the TX logic regardless of the operation mode
- **Bit 6 - ForceRxEnable** - Enable the RX logic regardless of the operation mode

- **Bit 5 - LDForceEnable** - Enables the Line Driver, regardless of the operation mode, when one of the loop-backs is selected.
- **Bit 4 - TESTCLKForceEnable** - Enable the test clock input pad regardless of the operation mode.
- **Bit 3 - I2CMclkAlwaysEnable** - I2C masters clock always enable (disable clock gating).
- **Bit 2 - PSFSMClkAlwaysOn** - Forces an initialization state machine clock to be always active (disables clock gating)

15.1.4 CHIP Config

[0x036] ChipConfig

- **Bit 7 - highSpeedDataOutInvert** - Inverts high speed data output lines (equivalent to swapping HSOUTP and HSOUTN on the PCB)
- **Bit 6 - highSpeedDataInInvert** - Inverts high speed data input lines (equivalent to swapping HSINP and HSINN on the PCB)
- **Bit 2:0 - ChipAddressBar[2:0]** - Sets most significant bits of the chip address (see Section 3.3).

15.1.5 Equalizer

[0x037] EQConfig

Main equalizer configuration register

- **Bit 4:3 - EQAttenuation[1:0]** - Attenuation of the equalizer

EQAttenuation[1:0]	Gain [V/V]
2'd0	1/3
2'd1	2/3
2'd2	2/3
2'd3	1/1

- **Bit 1:0 - EQCap[1:0]** - Capacitance select for the equalizer

EQCap[1:0]	Capacitance [fF]
2'd0	0
2'd1	70
2'd2	70
2'd3	140

[0x038] EQRes

Resistance configuration for the equalizer

- **Bit 7:6 - EQRes3[1:0]** - Resistance to be used in the fourth stage of the data input equalizer

EQRes3[1:0]	Resistance [kOhm]
2'd0	0
2'd1	3.0
2'd2	4.9
2'd3	7.1

- **Bit 5:4 - EQRes2[1:0]** - Resistance to be used in the third stage of the data input equalizer

EQRes2[1:0]	Resistance [kOhm]
2'd0	0
2'd1	3.0
2'd2	4.9
2'd3	7.1

- **Bit 3:2 - EQRes1[1:0]** - Resistance to be used in the second stage of the data input equalizer

EQRes1[1:0]	Resistance [kOhm]
2'd0	0
2'd1	3.0
2'd2	4.9
2'd3	7.1

- **Bit 1:0 - EQRes0[1:0]** - Resistance to be used in the first stage of the data input equalizer

EQRes0[1:0]	Resistance [kOhm]
2'd0	0
2'd1	3.0
2'd2	4.9
2'd3	7.1

15.1.6 Line Driver

[0x039] LDConfigH

Line driver configuration register

- **Bit 7 - LDEmphasisEnable** - Enable pre-emphasis in the line driver. The amplitude of the pre-emphasis is controlled by LDEmphasisAmp[6:0] and the duration by LDEmphasisShort.
- **Bit 6:0 - LDModulationCurrent[6:0]** - Sets high-speed line driver modulation current: $I_m = 137 \text{ uA} * \text{LD-ModulationCurrent}[6:0]$

[0x03a] LDConfigL

Line driver configuration register

- **Bit 7 - LDEmphasisShort** - Sets the duration of the pre-emphasis pulse. Please note that pre-emphasis has to be enabled (LDEmphasisEnable) for this field to have any impact.

- **Bit 6:0 - LDEmphasisAmp[6:0]** - Sets high-speed line driver pre-emphasis current: $I_{pre} = 137 \text{ uA} * \text{LDEmphasisAmp}[6:0]$. Please note that pre-emphasis has to be enabled (LDEmphasisEnable) for these registers bits to be active.

-Note for the LDConfigH and LDConfigL registers: since the high-speed line driver contains an internal 100 Ohm "termination", the currents set by LDModulationCurrent[6:0] and LDEmphasisAmp[6:0] bits are shared between the internal and external load impedances. This needs to be taken into account when computing the output signal amplitude. To calculate the modulation amplitude the user should thus use the equivalent resistor value of 50 Ohm, that is, the internal 100 Ohm resistor in parallel with the external 100 Ohm termination impedance.

[0x03b] REFCLK

Configuration for the reference clock pad

- **Bit 4 - TESTCLKsetCM** - Enable common mode generation for TSTCLKN/P input pads.
- **Bit 2 - REFCLKForceEnable** - Enable the reference clock pad regardless of the operation mode.
- **Bit 1 - REFCLKAcBias** - Enables the common mode generation for the REFCLK.
- **Bit 0 - REFCLKTerm** - Enables the 100 Ohm termination for the REFCLK input.

[0x03c] SCCONFIG

Serial interface (IC/EC) configuration register.

- **Bit 0 - scParityCheckDisable** - Disable parity check for incoming frames. If asserted, the data will be copied to registers regardless of parity check.

15.1.7 Reset

[0x03d] RESETConfig

Reset configuration.

- **Bit 7 - ResetOutDriveStrength** - Reset out pin driving strength.
- **Bit 6:5 - ResetOutLength[1:0]** -

ResetOutLength[1:0]	Reset pulse duration [s]
0	disabled
1	100n
2	1u
3	10u

- **Bit 3 - BODEnable** - Enables brownout detector.
- **Bit 2:0 - BODlevel[2:0]** -

BODlevel[2:0]	Brownout voltage level [V]
0	0.70
1	0.75
2	0.80
3	0.85
4	0.90
5	0.95
6	1.00
7	1.05

15.1.8 Power Good

[0x03e] PGConfig

Power Good configuration.

- **Bit 7 - PGEnable** - Enable Power Good feature. For more details see *Power-up state machine* (page 61).
- **Bit 6:4 - PGLevel[2:0]** - Enable Power Good feature. For more details see *Power-up state machine* (page 61).

PGLevel[2:0]	Voltage level [V]
0	0.70
1	0.75
2	0.80
3	0.85
4	0.90
5	0.95
6	1.00
7	1.05

- **Bit 3:0 - PGDelay[4:0]** -

PGDelay[3:0]	Wait time
0	disabled
1	1 us
2	5 us
3	10 us
4	50 us
5	100 us
6	500 us
7	1 ms
8	5 ms
9	10 ms
10	20 ms
11	50 ms
12	100 ms
13	200 ms
14	500 ms
15	1 s

15.1.9 I2C Masters

[0x03f] I2CMTransConfig

- Bit 7 - I2CMTransEnable -
- Bit 6:5 - I2CMTransChannel[1:0] -
- Bit 4:2 - I2CMTransAddressExt[2:0] -

[0x040] I2CMTransAddress

- Bit 6:0 - I2CMTransAddress[6:0] -

[0x041] I2CMTransCtrl

- Bit 7:0 - I2CMTransCtrl[7:0] -

[0x042] I2CMTransData0

- Bit 7:0 - I2CMTransData[7:0] -

[0x043] I2CMTransData1

- Bit 7:0 - I2CMTransData[15:8] -

[0x044] I2CMTransData2

- Bit 7:0 - I2CMTransData[23:16] -

[0x045] I2CMTransData3

- Bit 7:0 - I2CMTransData[31:24] -

[0x046] I2CMTransData4

- Bit 7:0 - I2CMTransData[39:32] -

[0x047] I2CMTransData5

- Bit 7:0 - I2CMTransData[47:40] -

[0x048] I2CMTransData6

- Bit 7:0 - I2CMTransData[55:48] -

[0x049] I2CMTransData7

- Bit 7:0 - I2CMTransData[63:56] -

[0x04a] I2CMTransData8

- Bit 7:0 - I2CMTransData[71:64] -

[0x04b] I2CMTransData9

- Bit 7:0 - I2CMTransData[79:72] -

[0x04c] I2CMTransData10

- Bit 7:0 - I2CMTransData[87:80] -

[0x04d] I2CMTransData11

- Bit 7:0 - I2CMTransData[95:88] -

[0x04e] I2CMTransData12

- Bit 7:0 - I2CMTransData[103:96] -

[0x04f] I2CMTransData13

- Bit 7:0 - I2CMTransData[111:104] -

[0x050] I2CMTransData14

- Bit 7:0 - I2CMTransData[119:112] -

[0x051] I2CMTransData15

- Bit 7:0 - I2CMTransData[127:120] -

15.1.10 Parallel IO**[0x052] PIODirH**

Direction control for Parallel IO port

- Bit 7:0 - PIODir[15:8] -

PIODir[n]	Function
1'b0	Pin configured as an input
1'b1	Pin configured as an output

[0x053] PIODirL

Direction control for Parallel IO port

- Bit 7:0 - **PIODir[7:0]** -

PIODir[n]	Function
1'b0	Pin configured as an input
1'b1	Pin configured as an output

[0x054] PIOOutH

Output control for Parallel IO port

- Bit 7:0 - **PIOOut[15:8]** -

[0x055] PIOOutL

Output control for Parallel IO port

- Bit 7:0 - **PIOOut[7:0]** -

[0x056] PIOPullEnaH

Pull-up/pull-down control for Parallel IO port

- Bit 7:0 - **PIOPullEnable[15:8]** -

[0x057] PIOPullEnaL

Pull-up/pull-down control for Parallel IO port

- Bit 7:0 - **PIOPullEnable[7:0]** -

[0x058] PIOUpDownH

Selects pull up or pull down for Parallel IO port

- Bit 7:0 - **PIOUpDown[15:8]** -

[0x059] PIOUpDownL

Selects pull up or pull down for Parallel IO port

- Bit 7:0 - **PIOUpDown[7:0]** -

[0x05a] PIODriveStrengthH

Selects driving strength for Parallel IO port when configured as an output

- Bit 7:0 - **PIODriveStrength[15:8]** -

[0x05b] PIODriveStrengthL

Selects driving strength for Parallel IO port when configured as an output

- **Bit 7:0 - PIODriveStrength[7:0]** -

15.1.11 Phase Shifter

[0x05c] PS0Config

Main configuration of the phase-shifter clock 0

- **Bit 7 - PS0Delay[8]** - MSB of the delay select for clock 0. For more information check [\[0x05d\] PS0Delay](#) (page 143) register.
- **Bit 6 - PS0EnableFineTune** - Enable fine deskewing for clock 0.
- **Bit 5:3 - PS0DriveStrength[2:0]** - Sets the driving strength for 0 clock output.

PS0DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 2:0 - PS0Freq[2:0]** - Sets the frequency for 0 clock output.

PS0Freq[2:0]	Frequency [MHz]
3'd0	disabled
3'd1	40
3'd2	80
3'd3	160
3'd4	320
3'd5	640
3'd6	1280
3'd7	

[0x05d] PS0Delay

Delay of the phase-shifter clock 0

- **Bit 7:0 - PS0Delay[7:0]** - Delay select for clock 0. Please note that that most significant bit of the PS0Delay field is stored in the [\[0x05c\] PS0Config](#) (page 143) register.

[0x05e] PS0OutDriver

Output driver configuration for the phase-shifter clock 0

- **Bit 7:5 - PS0PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for phase-shifter 0 clock output.

PS0PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - PS0PreEmphasisMode[1:0]** - Sets the pre-emphasis mode for clock output 0.

PS0PreEmphasisMode[2:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - PS0PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for 0 clock output in self timed mode.

PS0PreEmphasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

[0x05f] PS1Config

Main configuration of the phase-shifter clock 1

- **Bit 7 - PS1Delay[8]** - MSB of the delay select for clock 1. For more information check [\[0x060\] PS1Delay](#) (page 145) register.
- **Bit 6 - PS1EnableFineTune** - Enable fine deskewing for clock 1.
- **Bit 5:3 - PS1DriveStrength[2:0]** - Sets the driving strength for 1 clock output.

PS1DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 2:0 - PS1Freq[2:0]** - Sets the frequency for 1 clock output.

PS1Freq[2:0]	Frequency [MHz]
3'd0	disabled
3'd1	40
3'd2	80
3'd3	160
3'd4	320
3'd5	640
3'd6	1280
3'd7	

[0x060] PS1Delay

Delay of the phase-shifter clock 1

- **Bit 7:0 - PS1Delay[7:0]** - Delay select for clock 1. Please note that that most significant bit of the PS1Delay field is stored in the [\[0x05f\] PS1Config](#) (page 144) register.

[0x061] PS1OutDriver

Output driver configuration for the phase-shifter clock 1

- **Bit 7:5 - PS1PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for phase-shifter 1 clock output.

PS1PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - PS1PreEmphasisMode[1:0]** - Sets the pre-emphasis mode for clock output 1.

PS1PreEmphasisMode[2:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - PS1PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for 1 clock output in self timed mode.

PS1PreEmpahasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

[0x062] PS2Config

Main configuration of the phase-shifter clock 2

- **Bit 7 - PS2Delay[8]** - MSB of the delay select for clock 2. For more information check [\[0x063\] PS2Delay](#) (page 146) register.
- **Bit 6 - PS2EnableFineTune** - Enable fine deskewing for clock 2.
- **Bit 5:3 - PS2DriveStrength[2:0]** - Sets the driving strength for 2 clock output.

PS2DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 2:0 - PS2Freq[2:0]** - Sets the frequency for 2 clock output.

PS2Freq[2:0]	Frequency [MHz]
3'd0	disabled
3'd1	40
3'd2	80
3'd3	160
3'd4	320
3'd5	640
3'd6	1280
3'd7	

[0x063] PS2Delay

Delay of the phase-shifter clock 2

- **Bit 7:0 - PS2Delay[7:0]** - Delay select for clock 2. Please note that that most significant bit of the PS2Delay field is stored in the [\[0x062\] PS2Config](#) (page 146) register.

[0x064] PS2OutDriver

Output driver configuration for the phase-shifter clock 2

- **Bit 7:5 - PS2PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for phase-shifter 2 clock output.

PS2PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - PS2PreEmphasisMode[1:0]** - Sets the pre-emphasis mode for clock output 2.

PS2PreEmphasisMode[2:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - PS2PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for 2 clock output in self timed mode.

PS2PreEmphasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

[0x065] PS3Config

Main configuration of the phase-shifter clock 3

- **Bit 7 - PS3Delay[8]** - MSB of the delay select for clock 3. For more information check [\[0x066\] PS3Delay](#) (page 148) register.
- **Bit 6 - PS3EnableFineTune** - Enable fine deskewing for clock 3.
- **Bit 5:3 - PS3DriveStrength[2:0]** - Sets the driving strength for 3 clock output.

PS3DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 2:0 - PS3Freq[2:0]** - Sets the frequency for 3 clock output.

PS3Freq[2:0]	Frequency [MHz]
3'd0	disabled
3'd1	40
3'd2	80
3'd3	160
3'd4	320
3'd5	640
3'd6	1280
3'd7	

[0x066] PS3Delay

Delay of the phase-shifter clock 3

- **Bit 7:0 - PS3Delay[7:0]** - Delay select for clock 3. Please note that that most significant bit of the PS3Delay field is stored in the [\[0x065\] PS3Config](#) (page 147) register.

[0x067] PS3OutDriver

Output driver configuration for the phase-shifter clock 3

- **Bit 7:5 - PS3PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for phase-shifter 3 clock output.

PS3PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - PS3PreEmphasisMode[1:0]** - Sets the pre-emphasis mode for clock output 3.

PS3PreEmphasisMode[2:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - PS3PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for 3 clock output in self timed mode.

PS3PreEmphasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

15.1.12 Voltage DAC

[0x068] DACConfigH

DACs configuration register.

- **Bit 7 - VOLDACEnable** - Enable voltage DAC.
- **Bit 6 - CURDACEnable** - Enables current DAC.
- **Bit 3:0 - VOLDACValue[11:8]** - Sets output voltage for the Voltage DAC.

See also: [\[0x069\] DACConfigL](#) (page 149), [\[0x06a\] CURDACValue](#) (page 149)

[0x069] DACConfigL

DACs configuration register.

- **Bit 7:0 - VOLDACValue[7:0]** - Sets output voltage for the Voltage DAC.

See also: [\[0x068\] DACConfigH](#) (page 149)

15.1.13 CURDAC

[0x06a] CURDACValue

Output current

- **Bit 7:0 - CURDACSelect[7:0]** - Sets output current for the current DAC. Current = CURDACSelect * XX uA.

See also: [\[0x06b\] CURDACCHN](#) (page 150), [\[0x068\] DACConfigH](#) (page 149)

[0x06b] CURDACCHN

Current DAC output multiplexer.

- **Bit 7:0 - CURDACChnEnable[7:0]** - Setting Nth bit in this register attaches current DAC to ADCN pin. Current source can be attached to any number of channels.

See also: [\[0x06a\] CURDACPValue](#) (page 149), [\[0x068\] DACConfigH](#) (page 149)

15.1.14 ePortClk**[0x06c] EPCLK0ChnCntrH**

Configuration of clock output 0

- **Bit 6 - EPCLK0Invert** - Inverts 0 clock output.
- **Bit 5:3 - EPCLK0DriveStrength[2:0]** - Sets the driving strength for 0 clock output.

EPCLK0DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 2:0 - EPCLK0Freq[2:0]** - Sets the frequency for 0 clock output.

EPCLK0Freq[2:0]	Frequency [MHz]
3'd0	disabled
3'd1	40
3'd2	80
3'd3	160
3'd4	320
3'd5	640
3'd6	1280
3'd7	ePortRxDataIn[0]

[0x06d] EPCLK0ChnCntrL

Configuration of clock output 0

- **Bit 7:5 - EPCLK0PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for 0 clock output.

EPCLK0PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPCLK0PreEmphasisMode[1:0]** - Sets the pre-emphasis mode for clock output 0.

EPCLK0PreEmphasisMode[2:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPCLK0PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for 0 clock output in self timed mode.

EPCLK0PreEmphasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

[0x06e] EPCLK1ChnCntrH

Configuration of clock output 1

- **Bit 6 - EPCLK1Invert** - Inverts 1 clock output.
- **Bit 5:3 - EPCLK1DriveStrength[2:0]** - Sets the driving strength for 1 clock output.

EPCLK1DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 2:0 - EPCLK1Freq[2:0]** - Sets the frequency for 1 clock output.

EPCLK1Freq[2:0]	Frequency [MHz]
3'd0	disabled
3'd1	40
3'd2	80
3'd3	160
3'd4	320
3'd5	640
3'd6	1280
3'd7	ePortRxDataIn[1]

[0x06f] EPCLK1ChnCntrL

Configuration of clock output 1

- **Bit 7:5 - EPCLK1PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for 1 clock output.

EPCLK1PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPCLK1PreEmphasisMode[1:0]** - Sets the pre-emphasis mode for clock output 1.

EPCLK1PreEmphasisMode[2:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPCLK1PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for 1 clock output in self timed mode.

EPCLK1PreEmphasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

[0x070] EPCLK2ChnCntrH

Configuration of clock output 2

- **Bit 6 - EPCLK2Invert** - Inverts 2 clock output.
- **Bit 5:3 - EPCLK2DriveStrength[2:0]** - Sets the driving strength for 2 clock output.

EPCLK2DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 2:0 - EPCLK2Freq[2:0]** - Sets the frequency for 2 clock output.

EPCLK2Freq[2:0]	Frequency [MHz]
3'd0	disabled
3'd1	40
3'd2	80
3'd3	160
3'd4	320
3'd5	640
3'd6	1280
3'd7	ePortRxDataIn[2]

[0x071] EPCLK2ChnCntrL

Configuration of clock output 2

- **Bit 7:5 - EPCLK2PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for 2 clock output.

EPCLK2PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPCLK2PreEmphasisMode[1:0]** - Sets the pre-emphasis mode for clock output 2.

EPCLK2PreEmphasisMode[2:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPCLK2PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for 2 clock output in self timed mode.

EPCLK2PreEmpahasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

[0x072] EPCLK3ChnCntrH

Configuration of clock output 3

- **Bit 6 - EPCLK3Invert** - Inverts 3 clock output.
- **Bit 5:3 - EPCLK3DriveStrength[2:0]** - Sets the driving strength for 3 clock output.

EPCLK3DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 2:0 - EPCLK3Freq[2:0]** - Sets the frequency for 3 clock output.

EPCLK3Freq[2:0]	Frequency [MHz]
3'd0	disabled
3'd1	40
3'd2	80
3'd3	160
3'd4	320
3'd5	640
3'd6	1280
3'd7	ePortRxDataIn[3]

[0x073] EPCLK3ChnCntrL

Configuration of clock output 3

- **Bit 7:5 - EPCLK3PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for 3 clock output.

EPCLK3PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPCLK3PreEmphasisMode[1:0]** - Sets the pre-emphasis mode for clock output 3.

EPCLK3PreEmphasisMode[2:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPCLK3PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for 3 clock output in self timed mode.

EPCLK3PreEmphasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

[0x074] EPCLK4ChnCntrH

Configuration of clock output 4

- **Bit 6 - EPCLK4Invert** - Inverts 4 clock output.
- **Bit 5:3 - EPCLK4DriveStrength[2:0]** - Sets the driving strength for 4 clock output.

EPCLK4DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 2:0 - EPCLK4Freq[2:0]** - Sets the frequency for 4 clock output.

EPCLK4Freq[2:0]	Frequency [MHz]
3'd0	disabled
3'd1	40
3'd2	80
3'd3	160
3'd4	320
3'd5	640
3'd6	1280
3'd7	ePortRxDataIn[4]

[0x075] EPCLK4ChnCntrL

Configuration of clock output 4

- **Bit 7:5 - EPCLK4PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for 4 clock output.

EPCLK4PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPCLK4PreEmphasisMode[1:0]** - Sets the pre-emphasis mode for clock output 4.

EPCLK4PreEmphasisMode[2:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPCLK4PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for 4 clock output in self timed mode.

EPCLK4PreEmphasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

[0x076] EPCLK5ChnCntrH

Configuration of clock output 5

- **Bit 6 - EPCLK5Invert** - Inverts 5 clock output.
- **Bit 5:3 - EPCLK5DriveStrength[2:0]** - Sets the driving strength for 5 clock output.

EPCLK5DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 2:0 - EPCLK5Freq[2:0]** - Sets the frequency for 5 clock output.

EPCLK5Freq[2:0]	Frequency [MHz]
3'd0	disabled
3'd1	40
3'd2	80
3'd3	160
3'd4	320
3'd5	640
3'd6	1280
3'd7	ePortRxDataIn[5]

[0x077] EPCLK5ChnCntrL

Configuration of clock output 5

- **Bit 7:5 - EPCLK5PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for 5 clock output.

EPCLK5PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPCLK5PreEmphasisMode[1:0]** - Sets the pre-emphasis mode for clock output 5.

EPCLK5PreEmphasisMode[2:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPCLK5PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for 5 clock output in self timed mode.

EPCLK5PreEmpahasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

[0x078] EPCLK6ChnCntrH

Configuration of clock output 6

- **Bit 6 - EPCLK6Invert** - Inverts 6 clock output.
- **Bit 5:3 - EPCLK6DriveStrength[2:0]** - Sets the driving strength for 6 clock output.

EPCLK6DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 2:0 - EPCLK6Freq[2:0]** - Sets the frequency for 6 clock output.

EPCLK6Freq[2:0]	Frequency [MHz]
3'd0	disabled
3'd1	40
3'd2	80
3'd3	160
3'd4	320
3'd5	640
3'd6	1280
3'd7	ePortRxDataIn[6]

[0x079] EPCLK6ChnCntrL

Configuration of clock output 6

- **Bit 7:5 - EPCLK6PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for 6 clock output.

EPCLK6PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPCLK6PreEmphasisMode[1:0]** - Sets the pre-emphasis mode for clock output 6.

EPCLK6PreEmphasisMode[2:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPCLK6PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for 6 clock output in self timed mode.

EPCLK6PreEmphasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

[0x07a] EPCLK7ChnCntrH

Configuration of clock output 7

- **Bit 6 - EPCLK7Invert** - Inverts 7 clock output.
- **Bit 5:3 - EPCLK7DriveStrength[2:0]** - Sets the driving strength for 7 clock output.

EPCLK7DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 2:0 - EPCLK7Freq[2:0]** - Sets the frequency for 7 clock output.

EPCLK7Freq[2:0]	Frequency [MHz]
3'd0	disabled
3'd1	40
3'd2	80
3'd3	160
3'd4	320
3'd5	640
3'd6	1280
3'd7	ePortRxDataIn[7]

[0x07b] EPCLK7ChnCntrL

Configuration of clock output 7

- **Bit 7:5 - EPCLK7PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for 7 clock output.

EPCLK7PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPCLK7PreEmphasisMode[1:0]** - Sets the pre-emphasis mode for clock output 7.

EPCLK7PreEmphasisMode[2:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPCLK7PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for 7 clock output in self timed mode.

EPCLK7PreEmphasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

[0x07c] EPCLK8ChnCntrH

Configuration of clock output 8

- **Bit 6 - EPCLK8Invert** - Inverts 8 clock output.
- **Bit 5:3 - EPCLK8DriveStrength[2:0]** - Sets the driving strength for 8 clock output.

EPCLK8DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 2:0 - EPCLK8Freq[2:0]** - Sets the frequency for 8 clock output.

EPCLK8Freq[2:0]	Frequency [MHz]
3'd0	disabled
3'd1	40
3'd2	80
3'd3	160
3'd4	320
3'd5	640
3'd6	1280
3'd7	ePortRxDataIn[8]

[0x07d] EPCLK8ChnCntrL

Configuration of clock output 8

- **Bit 7:5 - EPCLK8PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for 8 clock output.

EPCLK8PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPCLK8PreEmphasisMode[1:0]** - Sets the pre-emphasis mode for clock output 8.

EPCLK8PreEmphasisMode[2:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPCLK8PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for 8 clock output in self timed mode.

EPCLK8PreEmpahasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

[0x07e] EPCLK9ChnCntrH

Configuration of clock output 9

- **Bit 6 - EPCLK9Invert** - Inverts 9 clock output.
- **Bit 5:3 - EPCLK9DriveStrength[2:0]** - Sets the driving strength for 9 clock output.

EPCLK9DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 2:0 - EPCLK9Freq[2:0]** - Sets the frequency for 9 clock output.

EPCLK9Freq[2:0]	Frequency [MHz]
3'd0	disabled
3'd1	40
3'd2	80
3'd3	160
3'd4	320
3'd5	640
3'd6	1280
3'd7	ePortRxDataIn[9]

[0x07f] EPCLK9ChnCntrL

Configuration of clock output 9

- **Bit 7:5 - EPCLK9PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for 9 clock output.

EPCLK9PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPCLK9PreEmphasisMode[1:0]** - Sets the pre-emphasis mode for clock output 9.

EPCLK9PreEmphasisMode[2:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPCLK9PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for 9 clock output in self timed mode.

EPCLK9PreEmphasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

[0x080] EPCLK10ChnCntrH

Configuration of clock output 10

- **Bit 6 - EPCLK10Invert** - Inverts 10 clock output.
- **Bit 5:3 - EPCLK10DriveStrength[2:0]** - Sets the driving strength for 10 clock output.

EPCLK10DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 2:0 - EPCLK10Freq[2:0]** - Sets the frequency for 10 clock output.

EPCLK10Freq[2:0]	Frequency [MHz]
3'd0	disabled
3'd1	40
3'd2	80
3'd3	160
3'd4	320
3'd5	640
3'd6	1280
3'd7	ePortRxDataIn[10]

[0x081] EPCLK10ChnCntrL

Configuration of clock output 10

- **Bit 7:5 - EPCLK10PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for 10 clock output.

EPCLK10PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPCLK10PreEmphasisMode[1:0]** - Sets the pre-emphasis mode for clock output 10.

EPCLK10PreEmphasisMode[2:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPCLK10PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for 10 clock output in self timed mode.

EPCLK10PreEmphasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

[0x082] EPCLK11ChnCntrH

Configuration of clock output 11

- **Bit 6 - EPCLK11Invert** - Inverts 11 clock output.
- **Bit 5:3 - EPCLK11DriveStrength[2:0]** - Sets the driving strength for 11 clock output.

EPCLK11DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 2:0 - EPCLK11Freq[2:0]** - Sets the frequency for 11 clock output.

EPCLK11Freq[2:0]	Frequency [MHz]
3'd0	disabled
3'd1	40
3'd2	80
3'd3	160
3'd4	320
3'd5	640
3'd6	1280
3'd7	ePortRxDataIn[11]

[0x083] EPCLK11ChnCntrL

Configuration of clock output 11

- **Bit 7:5 - EPCLK11PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for 11 clock output.

EPCLK11PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPCLK11PreEmphasisMode[1:0]** - Sets the pre-emphasis mode for clock output 11.

EPCLK11PreEmphasisMode[2:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPCLK11PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for 11 clock output in self timed mode.

EPCLK11PreEmpahasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

[0x084] EPCLK12ChnCntrH

Configuration of clock output 12

- **Bit 6 - EPCLK12Invert** - Inverts 12 clock output.
- **Bit 5:3 - EPCLK12DriveStrength[2:0]** - Sets the driving strength for 12 clock output.

EPCLK12DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 2:0 - EPCLK12Freq[2:0]** - Sets the frequency for 12 clock output.

EPCLK12Freq[2:0]	Frequency [MHz]
3'd0	disabled
3'd1	40
3'd2	80
3'd3	160
3'd4	320
3'd5	640
3'd6	1280
3'd7	ePortRxDataIn[12]

[0x085] EPCLK12ChnCntrL

Configuration of clock output 12

- **Bit 7:5 - EPCLK12PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for 12 clock output.

EPCLK12PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPCLK12PreEmphasisMode[1:0]** - Sets the pre-emphasis mode for clock output 12.

EPCLK12PreEmphasisMode[2:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPCLK12PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for 12 clock output in self timed mode.

EPCLK12PreEmphasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

[0x086] EPCLK13ChnCntrH

Configuration of clock output 13

- **Bit 6 - EPCLK13Invert** - Inverts 13 clock output.
- **Bit 5:3 - EPCLK13DriveStrength[2:0]** - Sets the driving strength for 13 clock output.

EPCLK13DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 2:0 - EPCLK13Freq[2:0]** - Sets the frequency for 13 clock output.

EPCLK13Freq[2:0]	Frequency [MHz]
3'd0	disabled
3'd1	40
3'd2	80
3'd3	160
3'd4	320
3'd5	640
3'd6	1280
3'd7	ePortRxDataIn[13]

[0x087] EPCLK13ChnCntrL

Configuration of clock output 13

- **Bit 7:5 - EPCLK13PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for 13 clock output.

EPCLK13PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPCLK13PreEmphasisMode[1:0]** - Sets the pre-emphasis mode for clock output 13.

EPCLK13PreEmphasisMode[2:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPCLK13PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for 13 clock output in self timed mode.

EPCLK13PreEmphasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

[0x088] EPCLK14ChnCntrH

Configuration of clock output 14

- **Bit 6 - EPCLK14Invert** - Inverts 14 clock output.
- **Bit 5:3 - EPCLK14DriveStrength[2:0]** - Sets the driving strength for 14 clock output.

EPCLK14DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 2:0 - EPCLK14Freq[2:0]** - Sets the frequency for 14 clock output.

EPCLK14Freq[2:0]	Frequency [MHz]
3'd0	disabled
3'd1	40
3'd2	80
3'd3	160
3'd4	320
3'd5	640
3'd6	1280
3'd7	ePortRxDataIn[14]

[0x089] EPCLK14ChnCntrL

Configuration of clock output 14

- **Bit 7:5 - EPCLK14PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for 14 clock output.

EPCLK14PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPCLK14PreEmphasisMode[1:0]** - Sets the pre-emphasis mode for clock output 14.

EPCLK14PreEmphasisMode[2:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPCLK14PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for 14 clock output in self timed mode.

EPCLK14PreEmpahasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

[0x08a] EPCLK15ChnCntrH

Configuration of clock output 15

- **Bit 6 - EPCLK15Invert** - Inverts 15 clock output.
- **Bit 5:3 - EPCLK15DriveStrength[2:0]** - Sets the driving strength for 15 clock output.

EPCLK15DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 2:0 - EPCLK15Freq[2:0]** - Sets the frequency for 15 clock output.

EPCLK15Freq[2:0]	Frequency [MHz]
3'd0	disabled
3'd1	40
3'd2	80
3'd3	160
3'd4	320
3'd5	640
3'd6	1280
3'd7	ePortRxDataIn[15]

[0x08b] EPCLK15ChnCntrL

Configuration of clock output 15

- **Bit 7:5 - EPCLK15PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for 15 clock output.

EPCLK15PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPCLK15PreEmphasisMode[1:0]** - Sets the pre-emphasis mode for clock output 15.

EPCLK15PreEmphasisMode[2:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPCLK15PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for 15 clock output in self timed mode.

EPCLK15PreEmphasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

[0x08c] EPCLK16ChnCntrH

Configuration of clock output 16

- **Bit 6 - EPCLK16Invert** - Inverts 16 clock output.
- **Bit 5:3 - EPCLK16DriveStrength[2:0]** - Sets the driving strength for 16 clock output.

EPCLK16DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 2:0 - EPCLK16Freq[2:0]** - Sets the frequency for 16 clock output.

EPCLK16Freq[2:0]	Frequency [MHz]
3'd0	disabled
3'd1	40
3'd2	80
3'd3	160
3'd4	320
3'd5	640
3'd6	1280
3'd7	ePortRxDataIn[16]

[0x08d] EPCLK16ChnCntrL

Configuration of clock output 16

- **Bit 7:5 - EPCLK16PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for 16 clock output.

EPCLK16PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPCLK16PreEmphasisMode[1:0]** - Sets the pre-emphasis mode for clock output 16.

EPCLK16PreEmphasisMode[2:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPCLK16PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for 16 clock output in self timed mode.

EPCLK16PreEmphasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

[0x08e] EPCLK17ChnCntrH

Configuration of clock output 17

- **Bit 6 - EPCLK17Invert** - Inverts 17 clock output.
- **Bit 5:3 - EPCLK17DriveStrength[2:0]** - Sets the driving strength for 17 clock output.

EPCLK17DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 2:0 - EPCLK17Freq[2:0]** - Sets the frequency for 17 clock output.

EPCLK17Freq[2:0]	Frequency [MHz]
3'd0	disabled
3'd1	40
3'd2	80
3'd3	160
3'd4	320
3'd5	640
3'd6	1280
3'd7	ePortRxDataIn[17]

[0x08f] EPCLK17ChnCntrL

Configuration of clock output 17

- **Bit 7:5 - EPCLK17PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for 17 clock output.

EPCLK17PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPCLK17PreEmphasisMode[1:0]** - Sets the pre-emphasis mode for clock output 17.

EPCLK17PreEmphasisMode[2:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPCLK17PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for 17 clock output in self timed mode.

EPCLK17PreEmpahasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

[0x090] EPCLK18ChnCntrH

Configuration of clock output 18

- **Bit 6 - EPCLK18Invert** - Inverts 18 clock output.
- **Bit 5:3 - EPCLK18DriveStrength[2:0]** - Sets the driving strength for 18 clock output.

EPCLK18DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 2:0 - EPCLK18Freq[2:0]** - Sets the frequency for 18 clock output.

EPCLK18Freq[2:0]	Frequency [MHz]
3'd0	disabled
3'd1	40
3'd2	80
3'd3	160
3'd4	320
3'd5	640
3'd6	1280
3'd7	ePortRxDataIn[18]

[0x091] EPCLK18ChnCntrL

Configuration of clock output 18

- **Bit 7:5 - EPCLK18PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for 18 clock output.

EPCLK18PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPCLK18PreEmphasisMode[1:0]** - Sets the pre-emphasis mode for clock output 18.

EPCLK18PreEmphasisMode[2:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPCLK18PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for 18 clock output in self timed mode.

EPCLK18PreEmphasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

[0x092] EPCLK19ChnCntrH

Configuration of clock output 19

- **Bit 6 - EPCLK19Invert** - Inverts 19 clock output.
- **Bit 5:3 - EPCLK19DriveStrength[2:0]** - Sets the driving strength for 19 clock output.

EPCLK19DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 2:0 - EPCLK19Freq[2:0]** - Sets the frequency for 19 clock output.

EPCLK19Freq[2:0]	Frequency [MHz]
3'd0	disabled
3'd1	40
3'd2	80
3'd3	160
3'd4	320
3'd5	640
3'd6	1280
3'd7	ePortRxDataIn[19]

[0x093] EPCLK19ChnCntrL

Configuration of clock output 19

- **Bit 7:5 - EPCLK19PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for 19 clock output.

EPCLK19PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPCLK19PreEmphasisMode[1:0]** - Sets the pre-emphasis mode for clock output 19.

EPCLK19PreEmphasisMode[2:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPCLK19PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for 19 clock output in self timed mode.

EPCLK19PreEmphasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

[0x094] EPCLK20ChnCntrH

Configuration of clock output 20

- **Bit 6 - EPCLK20Invert** - Inverts 20 clock output.
- **Bit 5:3 - EPCLK20DriveStrength[2:0]** - Sets the driving strength for 20 clock output.

EPCLK20DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 2:0 - EPCLK20Freq[2:0]** - Sets the frequency for 20 clock output.

EPCLK20Freq[2:0]	Frequency [MHz]
3'd0	disabled
3'd1	40
3'd2	80
3'd3	160
3'd4	320
3'd5	640
3'd6	1280
3'd7	ePortRxDataIn[20]

[0x095] EPCLK20ChnCntrL

Configuration of clock output 20

- **Bit 7:5 - EPCLK20PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for 20 clock output.

EPCLK20PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPCLK20PreEmphasisMode[1:0]** - Sets the pre-emphasis mode for clock output 20.

EPCLK20PreEmphasisMode[2:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPCLK20PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for 20 clock output in self timed mode.

EPCLK20PreEmpahasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

[0x096] EPCLK21ChnCntrH

Configuration of clock output 21

- **Bit 6 - EPCLK21Invert** - Inverts 21 clock output.
- **Bit 5:3 - EPCLK21DriveStrength[2:0]** - Sets the driving strength for 21 clock output.

EPCLK21DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 2:0 - EPCLK21Freq[2:0]** - Sets the frequency for 21 clock output.

EPCLK21Freq[2:0]	Frequency [MHz]
3'd0	disabled
3'd1	40
3'd2	80
3'd3	160
3'd4	320
3'd5	640
3'd6	1280
3'd7	ePortRxDataIn[21]

[0x097] EPCLK21ChnCntrL

Configuration of clock output 21

- **Bit 7:5 - EPCLK21PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for 21 clock output.

EPCLK21PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPCLK21PreEmphasisMode[1:0]** - Sets the pre-emphasis mode for clock output 21.

EPCLK21PreEmphasisMode[2:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPCLK21PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for 21 clock output in self timed mode.

EPCLK21PreEmphasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

[0x098] EPCLK22ChnCntrH

Configuration of clock output 22

- **Bit 6 - EPCLK22Invert** - Inverts 22 clock output.
- **Bit 5:3 - EPCLK22DriveStrength[2:0]** - Sets the driving strength for 22 clock output.

EPCLK22DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 2:0 - EPCLK22Freq[2:0]** - Sets the frequency for 22 clock output.

EPCLK22Freq[2:0]	Frequency [MHz]
3'd0	disabled
3'd1	40
3'd2	80
3'd3	160
3'd4	320
3'd5	640
3'd6	1280
3'd7	ePortRxDataIn[22]

[0x099] EPCLK22ChnCntrL

Configuration of clock output 22

- **Bit 7:5 - EPCLK22PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for 22 clock output.

EPCLK22PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPCLK22PreEmphasisMode[1:0]** - Sets the pre-emphasis mode for clock output 22.

EPCLK22PreEmphasisMode[2:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPCLK22PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for 22 clock output in self timed mode.

EPCLK22PreEmphasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

[0x09a] EPCLK23ChnCntrH

Configuration of clock output 23

- **Bit 6 - EPCLK23Invert** - Inverts 23 clock output.
- **Bit 5:3 - EPCLK23DriveStrength[2:0]** - Sets the driving strength for 23 clock output.

EPCLK23DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 2:0 - EPCLK23Freq[2:0]** - Sets the frequency for 23 clock output.

EPCLK23Freq[2:0]	Frequency [MHz]
3'd0	disabled
3'd1	40
3'd2	80
3'd3	160
3'd4	320
3'd5	640
3'd6	1280
3'd7	ePortRxDataIn[23]

[0x09b] EPCLK23ChnCntrL

Configuration of clock output 23

- **Bit 7:5 - EPCLK23PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for 23 clock output.

EPCLK23PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPCLK23PreEmphasisMode[1:0]** - Sets the pre-emphasis mode for clock output 23.

EPCLK23PreEmphasisMode[2:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPCLK23PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for 23 clock output in self timed mode.

EPCLK23PreEmpahasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

[0x09c] EPCLK24ChnCntrH

Configuration of clock output 24

- **Bit 6 - EPCLK24Invert** - Inverts 24 clock output.
- **Bit 5:3 - EPCLK24DriveStrength[2:0]** - Sets the driving strength for 24 clock output.

EPCLK24DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 2:0 - EPCLK24Freq[2:0]** - Sets the frequency for 24 clock output.

EPCLK24Freq[2:0]	Frequency [MHz]
3'd0	disabled
3'd1	40
3'd2	80
3'd3	160
3'd4	320
3'd5	640
3'd6	1280
3'd7	ePortRxDataIn[24]

[0x09d] EPCLK24ChnCntrL

Configuration of clock output 24

- **Bit 7:5 - EPCLK24PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for 24 clock output.

EPCLK24PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPCLK24PreEmphasisMode[1:0]** - Sets the pre-emphasis mode for clock output 24.

EPCLK24PreEmphasisMode[2:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPCLK24PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for 24 clock output in self timed mode.

EPCLK24PreEmphasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

[0x09e] EPCLK25ChnCntrH

Configuration of clock output 25

- **Bit 6 - EPCLK25Invert** - Inverts 25 clock output.
- **Bit 5:3 - EPCLK25DriveStrength[2:0]** - Sets the driving strength for 25 clock output.

EPCLK25DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 2:0 - EPCLK25Freq[2:0]** - Sets the frequency for 25 clock output.

EPCLK25Freq[2:0]	Frequency [MHz]
3'd0	disabled
3'd1	40
3'd2	80
3'd3	160
3'd4	320
3'd5	640
3'd6	1280
3'd7	ePortRxDataIn[25]

[0x09f] EPCLK25ChnCntrL

Configuration of clock output 25

- **Bit 7:5 - EPCLK25PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for 25 clock output.

EPCLK25PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPCLK25PreEmphasisMode[1:0]** - Sets the pre-emphasis mode for clock output 25.

EPCLK25PreEmphasisMode[2:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPCLK25PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for 25 clock output in self timed mode.

EPCLK25PreEmphasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

[0xa0] EPCLK26ChnCntrH

Configuration of clock output 26

- **Bit 6 - EPCLK26Invert** - Inverts 26 clock output.
- **Bit 5:3 - EPCLK26DriveStrength[2:0]** - Sets the driving strength for 26 clock output.

EPCLK26DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 2:0 - EPCLK26Freq[2:0]** - Sets the frequency for 26 clock output.

EPCLK26Freq[2:0]	Frequency [MHz]
3'd0	disabled
3'd1	40
3'd2	80
3'd3	160
3'd4	320
3'd5	640
3'd6	1280
3'd7	ePortRxDataIn[26]

[0x0a1] EPCLK26ChnCntrL

Configuration of clock output 26

- **Bit 7:5 - EPCLK26PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for 26 clock output.

EPCLK26PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPCLK26PreEmphasisMode[1:0]** - Sets the pre-emphasis mode for clock output 26.

EPCLK26PreEmphasisMode[2:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPCLK26PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for 26 clock output in self timed mode.

EPCLK26PreEmpahasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

[0x0a2] EPCLK27ChnCntrH

Configuration of clock output 27

- **Bit 6 - EPCLK27Invert** - Inverts 27 clock output.
- **Bit 5:3 - EPCLK27DriveStrength[2:0]** - Sets the driving strength for 27 clock output.

EPCLK27DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 2:0 - EPCLK27Freq[2:0]** - Sets the frequency for 27 clock output.

EPCLK27Freq[2:0]	Frequency [MHz]
3'd0	disabled
3'd1	40
3'd2	80
3'd3	160
3'd4	320
3'd5	640
3'd6	1280
3'd7	ePortRxDataIn[27]

[0x0a3] EPCLK27ChnCntrL

Configuration of clock output 27

- **Bit 7:5 - EPCLK27PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for 27 clock output.

EPCLK27PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPCLK27PreEmphasisMode[1:0]** - Sets the pre-emphasis mode for clock output 27.

EPCLK27PreEmphasisMode[2:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPCLK27PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for 27 clock output in self timed mode.

EPCLK27PreEmphasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

[0x0a4] EPCLK28ChnCntrH

Configuration of clock output 28

- **Bit 6 - EPCLK28Invert** - Inverts 28 clock output.
- **Bit 5:3 - EPCLK28DriveStrength[2:0]** - Sets the driving strength for 28 clock output.

EPCLK28DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 2:0 - EPCLK28Freq[2:0]** - Sets the frequency for 28 clock output.

EPCLK28Freq[2:0]	Frequency [MHz]
3'd0	disabled
3'd1	40
3'd2	80
3'd3	160
3'd4	320
3'd5	640
3'd6	1280
3'd7	ePortRxDataIn[28]

[0x0a5] EPCLK28ChnCntrL

Configuration of clock output 28

- **Bit 7:5 - EPCLK28PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for 28 clock output.

EPCLK28PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPCLK28PreEmphasisMode[1:0]** - Sets the pre-emphasis mode for clock output 28.

EPCLK28PreEmphasisMode[2:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPCLK28PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for 28 clock output in self timed mode.

EPCLK28PreEmphasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

[0x0a6] Reserved1

Reserved

15.1.15 ePortTx

[0x0a7] EPTXDataRate

Data rate control for ePortTx

- Bit 7:6 - **EPTX3DataRate[1:0]** - Data rate for ePortTx group 3

EPTX3DataRate[1:0]	Group 3 data date
2'd0	disabled
2'd1	80M bps
2'd2	160 Mbps
2'd3	320 Mbps

- Bit 5:4 - **EPTX2DataRate[1:0]** - Data rate for ePortTx group 2

EPTX2DataRate[1:0]	Group 2 data date
2'd0	disabled
2'd1	80M bps
2'd2	160 Mbps
2'd3	320 Mbps

- Bit 3:2 - **EPTX1DataRate[1:0]** - Data rate for ePortTx group 1

EPTX1DataRate[1:0]	Group 1 data date
2'd0	disabled
2'd1	80M bps
2'd2	160 Mbps
2'd3	320 Mbps

- Bit 1:0 - **EPTX0DataRate[1:0]** - Data rate for ePortTx group 0

EPTX0DataRate[1:0]	Group 0 data date
2'd0	disabled
2'd1	80M bps
2'd2	160 Mbps
2'd3	320 Mbps

[0x0a8] EPTXControl

EportTx configuration register.

- Bit 7:6 - **EPTXEcPreEmphasisWidth[2:1]** - Sets the width of pre-emphasis pulse for EC channel output

EPTXECPREEMPHASISWIDTH[2:1]	Pulse length [ps]
2'd0	120
2'd1	360
2'd2	600
2'd3	840

- Bit 5 - **EPTXEcInvert** - Invert data for EC channel output

- **Bit 4 - EPTXEcEnable** - Enable EC channel output
- **Bit 3 - EPTX3MirrorEnable** - Enables mirror feature for group 3
- **Bit 2 - EPTX2MirrorEnable** - Enables mirror feature for group 2
- **Bit 1 - EPTX1MirrorEnable** - Enables mirror feature for group 1
- **Bit 0 - EPTX0MirrorEnable** - Enables mirror feature for group 0

[0x0a9] EPTX10Enable

Channel enable control for EPTX0 and EPTX1.

- **Bit 7 - EPTX13Enable** - Enable channel 3 in group 1
- **Bit 6 - EPTX12Enable** - Enable channel 2 in group 1
- **Bit 5 - EPTX11Enable** - Enable channel 1 in group 1
- **Bit 4 - EPTX10Enable** - Enable channel 0 in group 1
- **Bit 3 - EPTX03Enable** - Enable channel 3 in group 0
- **Bit 2 - EPTX02Enable** - Enable channel 2 in group 0
- **Bit 1 - EPTX01Enable** - Enable channel 1 in group 0
- **Bit 0 - EPTX00Enable** - Enable channel 0 in group 0

[0x0aa] EPTX32Enable

Channel enable control for EPTX2 and EPTX3.

- **Bit 7 - EPTX33Enable** - Enable channel 3 in group 3
- **Bit 6 - EPTX32Enable** - Enable channel 2 in group 3
- **Bit 5 - EPTX31Enable** - Enable channel 1 in group 3
- **Bit 4 - EPTX30Enable** - Enable channel 0 in group 3
- **Bit 3 - EPTX23Enable** - Enable channel 3 in group 2
- **Bit 2 - EPTX22Enable** - Enable channel 2 in group 2
- **Bit 1 - EPTX21Enable** - Enable channel 1 in group 2
- **Bit 0 - EPTX20Enable** - Enable channel 0 in group 2

[0x0ab] EPTXEcChnCntr

EC channel driver configuration.

- **Bit 7:5 - EPTXEcPreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for the EC channel.

EPTXEcPreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- Bit 4:3 - **EPTXEcPreEmphasisMode[1:0]** - Sets the pre-emphasis mode for the EC channel.

EPCLK28PreEmphasisMode[2:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- Bit 2:0 - **EPTXEcDriveStrength[2:0]** - Sets the pre-emphasis strength for the EC channel.

EPTXEcDriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

[0x0ac] EPTX00ChnCntr

Control register for output driver of channel 0 in group 0

- Bit 7:5 - **EPTX00PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for channel 0 in group 0.

EPTX00PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- Bit 4:3 - **EPTX00PreEmphasisMode[1:0]** - Selects the pre-emphasis mode for channel 0 in group 0.

EPTX00PreEmphasisMode[1:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPTX00DriveStrength[2:0]** - Sets the driving strength for channel 0 in group 0.

EPTX00DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

[0x0ad] EPTX01ChnCntr

Control register for output driver of channel 1 in group 0

- **Bit 7:5 - EPTX01PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for channel 0 in group 1.

EPTX01PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPTX01PreEmphasisMode[1:0]** - Selects the pre-emphasis mode for channel 0 in group 1.

EPTX01PreEmphasisMode[1:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPTX01DriveStrength[2:0]** - Sets the driving strength for channel 0 in group 1.

EPTX01DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

[0x0ae] EPTX02ChnCntr

Control register for output driver of channel 2 in group 0

- **Bit 7:5 - EPTX02PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for channel 0 in group 2.

EPTX02PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPTX02PreEmphasisMode[1:0]** - Selects the pre-emphasis mode for channel 0 in group 2.

EPTX02PreEmphasisMode[1:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPTX02DriveStrength[2:0]** - Sets the driving strength for channel 0 in group 2.

EPTX02DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

[0x0af] EPTX03ChnCntr

Control register for output driver of channel 3 in group 0

- Bit 7:5 - **EPTX03PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for channel 0 in group 3.

EPTX03PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- Bit 4:3 - **EPTX03PreEmphasisMode[1:0]** - Selects the pre-emphasis mode for channel 0 in group 3.

EPTX03PreEmphasisMode[1:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- Bit 2:0 - **EPTX03DriveStrength[2:0]** - Sets the driving strength for channel 0 in group 3.

EPTX03DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

[0x0b0] EPTX10ChnCntr

Control register for output driver of channel 0 in group 1

- Bit 7:5 - **EPTX10PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for channel 1 in group 0.

EPTX10PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- Bit 4:3 - **EPTX10PreEmphasisMode[1:0]** - Selects the pre-emphasis mode for channel 1 in group 0.

EPTX10PreEmphasisMode[1:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPTX10DriveStrength[2:0]** - Sets the driving strength for channel 1 in group 0.

EPTX10DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

[0x0b1] EPTX11ChnCntr

Control register for output driver of channel 1 in group 1

- **Bit 7:5 - EPTX11PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for channel 1 in group 1.

EPTX11PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPTX11PreEmphasisMode[1:0]** - Selects the pre-emphasis mode for channel 1 in group 1.

EPTX11PreEmphasisMode[1:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPTX11DriveStrength[2:0]** - Sets the driving strength for channel 1 in group 1.

EPTX11DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

[0x0b2] EPTX12ChnCntr

Control register for output driver of channel 2 in group 1

- **Bit 7:5 - EPTX12PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for channel 1 in group 2.

EPTX12PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPTX12PreEmphasisMode[1:0]** - Selects the pre-emphasis mode for channel 1 in group 2.

EPTX12PreEmphasisMode[1:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPTX12DriveStrength[2:0]** - Sets the driving strength for channel 1 in group 2.

EPTX12DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

[0x0b3] EPTX13ChnCntr

Control register for output driver of channel 3 in group 1

- Bit 7:5 - **EPTX13PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for channel 1 in group 3.

EPTX13PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- Bit 4:3 - **EPTX13PreEmphasisMode[1:0]** - Selects the pre-emphasis mode for channel 1 in group 3.

EPTX13PreEmphasisMode[1:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- Bit 2:0 - **EPTX13DriveStrength[2:0]** - Sets the driving strength for channel 1 in group 3.

EPTX13DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

[0x0b4] EPTX20ChnCntr

Control register for output driver of channel 0 in group 2

- Bit 7:5 - **EPTX20PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for channel 2 in group 0.

EPTX20PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- Bit 4:3 - **EPTX20PreEmphasisMode[1:0]** - Selects the pre-emphasis mode for channel 2 in group 0.

EPTX20PreEmphasisMode[1:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPTX20DriveStrength[2:0]** - Sets the driving strength for channel 2 in group 0.

EPTX20DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

[0x0b5] EPTX21ChnCntr

Control register for output driver of channel 1 in group 2

- **Bit 7:5 - EPTX21PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for channel 2 in group 1.

EPTX21PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPTX21PreEmphasisMode[1:0]** - Selects the pre-emphasis mode for channel 2 in group 1.

EPTX21PreEmphasisMode[1:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPTX21DriveStrength[2:0]** - Sets the driving strength for channel 2 in group 1.

EPTX21DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

[0x0b6] EPTX22ChnCntr

Control register for output driver of channel 2 in group 2

- **Bit 7:5 - EPTX22PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for channel 2 in group 2.

EPTX22PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPTX22PreEmphasisMode[1:0]** - Selects the pre-emphasis mode for channel 2 in group 2.

EPTX22PreEmphasisMode[1:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPTX22DriveStrength[2:0]** - Sets the driving strength for channel 2 in group 2.

EPTX22DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

[0x0b7] EPTX23ChnCntr

Control register for output driver of channel 3 in group 2

- Bit 7:5 - **EPTX23PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for channel 2 in group 3.

EPTX23PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- Bit 4:3 - **EPTX23PreEmphasisMode[1:0]** - Selects the pre-emphasis mode for channel 2 in group 3.

EPTX23PreEmphasisMode[1:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- Bit 2:0 - **EPTX23DriveStrength[2:0]** - Sets the driving strength for channel 2 in group 3.

EPTX23DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

[0x0b8] EPTX30ChnCntr

Control register for output driver of channel 0 in group 3

- Bit 7:5 - **EPTX30PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for channel 3 in group 0.

EPTX30PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- Bit 4:3 - **EPTX30PreEmphasisMode[1:0]** - Selects the pre-emphasis mode for channel 3 in group 0.

EPTX30PreEmphasisMode[1:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPTX30DriveStrength[2:0]** - Sets the driving strength for channel 3 in group 0.

EPTX30DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

[0x0b9] EPTX31ChnCntr

Control register for output driver of channel 1 in group 3

- **Bit 7:5 - EPTX31PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for channel 3 in group 1.

EPTX31PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPTX31PreEmphasisMode[1:0]** - Selects the pre-emphasis mode for channel 3 in group 1.

EPTX31PreEmphasisMode[1:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPTX31DriveStrength[2:0]** - Sets the driving strength for channel 3 in group 1.

EPTX31DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

[0x0ba] EPTX32ChnCntr

Control register for output driver of channel 2 in group 3

- **Bit 7:5 - EPTX32PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for channel 3 in group 2.

EPTX32PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - EPTX32PreEmphasisMode[1:0]** - Selects the pre-emphasis mode for channel 3 in group 2.

EPTX32PreEmphasisMode[1:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- **Bit 2:0 - EPTX32DriveStrength[2:0]** - Sets the driving strength for channel 3 in group 2.

EPTX32DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

[0x0bb] EPTX33ChnCntr

Control register for output driver of channel 3 in group 3

- Bit 7:5 - **EPTX33PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for channel 3 in group 3.

EPTX33PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- Bit 4:3 - **EPTX33PreEmphasisMode[1:0]** - Selects the pre-emphasis mode for channel 3 in group 3.

EPTX33PreEmphasisMode[1:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- Bit 2:0 - **EPTX33DriveStrength[2:0]** - Sets the driving strength for channel 3 in group 3.

EPTX33DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

[0x0bc] EPTX01_00ChnCntr

Output driver settings for ePortTx group 0

- Bit 7 - **EPTX01Invert** - Invert data for channel 1 in ePortTx Group 0
- Bit 6:4 - **EPTX01PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for channel 1 in ePortTx Group 0.

EPTX01PreEmphasisWidth[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 3 - EPTX00Invert** - Invert data for channel 0 in ePortTx Group 0
- **Bit 2:0 - EPTX00PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for channel 0 in ePortTx Group 0.

EPTX00PreEmphasisWidth[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

[0x0bd] EPTX03_02ChnCntr

Output driver settings for ePortTx group 0

- **Bit 7 - EPTX03Invert** - Invert data for channel 3 in ePortTx Group 0
- **Bit 6:4 - EPTX03PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for channel 3 in ePortTx Group 0.

EPTX03PreEmphasisWidth[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 3 - EPTX02Invert** - Invert data for channel 2 in ePortTx Group 0
- **Bit 2:0 - EPTX02PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for channel 2 in ePortTx Group 0.

EPTX02PreEmphasisWidth[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

[0x0be] EPTX11_10ChnCntr

Output driver settings for ePortTx group 1

- **Bit 7 - EPTX11Invert** - Invert data for channel 1 in ePortTx Group 1
- **Bit 6:4 - EPTX11PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for channel 1 in ePortTx Group 1.

EPTX11PreEmphasisWidth[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 3 - EPTX10Invert** - Invert data for channel 0 in ePortTx Group 1
- **Bit 2:0 - EPTX10PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for channel 0 in ePortTx Group 1.

EPTX10PreEmphasisWidth[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

[0x0bf] EPTX13_12ChnCntr

Output driver settings for ePortTx group 1

- **Bit 7 - EPTX13Invert** - Invert data for channel 3 in ePortTx Group 1
- **Bit 6:4 - EPTX13PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for channel 3 in ePortTx Group 1.

EPTX13PreEmphasisWidth[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 3 - EPTX12Invert** - Invert data for channel 2 in ePortTx Group 1
- **Bit 2:0 - EPTX12PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for channel 2 in ePortTx Group 1.

EPTX12PreEmpahasisWidth[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

[0x0c0] EPTX21_20ChnCntr

Output driver settings for ePortTx group 2

- **Bit 7 - EPTX21Invert** - Invert data for channel 1 in ePortTx Group 2
- **Bit 6:4 - EPTX21PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for channel 1 in ePortTx Group 2.

EPTX21PreEmphasisWidth[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 3 - EPTX20Invert** - Invert data for channel 0 in ePortTx Group 2
- **Bit 2:0 - EPTX20PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for channel 0 in ePortTx Group 2.

EPTX20PreEmphasisWidth[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

[0x0c1] EPTX23_22ChnCntr

Output driver settings for ePortTx group 2

- **Bit 7 - EPTX23Invert** - Invert data for channel 3 in ePortTx Group 2
- **Bit 6:4 - EPTX23PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for channel 3 in ePortTx Group 2.

EPTX23PreEmpahasisWidth[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 3 - EPTX22Invert** - Invert data for channel 2 in ePortTx Group 2
- **Bit 2:0 - EPTX22PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for channel 2 in ePortTx Group 2.

EPTX22PreEmphasisWidth[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

[0x0c2] EPTX31_30ChnCntr

Output driver settings for ePortTx group 3

- **Bit 7 - EPTX31Invert** - Invert data for channel 1 in ePortTx Group 3
- **Bit 6:4 - EPTX31PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for channel 1 in ePortTx Group 3.

EPTX31PreEmphasisWidth[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 3 - EPTX30Invert** - Invert data for channel 0 in ePortTx Group 3
- **Bit 2:0 - EPTX30PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for channel 0 in ePortTx Group 3.

EPTX30PreEmpahasisWidth[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

[0x0c3] EPTX33_32ChnCntr

Output driver settings for ePortTx group 3

- **Bit 7 - EPTX33Invert** - Invert data for channel 3 in ePortTx Group 3
- **Bit 6:4 - EPTX33PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for channel 3 in ePortTx Group 3.

EPTX33PreEmphasisWidth[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 3 - EPTX32Invert** - Invert data for channel 2 in ePortTx Group 3
- **Bit 2:0 - EPTX32PreEmphasisWidth[2:0]** - Sets the width of pre-emphasis pulse for channel 2 in ePortTx Group 3.

EPTX32PreEmphasisWidth[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

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[0x0c4] EPRX0Control

Configuration of ePortRx Group 0

- **Bit 7 - EPRX03Enable** - Enables channel 3 in group 0.

- **Bit 6 - EPRX02Enable** - Enables channel 2 in group 0.
- **Bit 5 - EPRX01Enable** - Enables channel 1 in group 0.
- **Bit 4 - EPRX00Enable** - Enables channel 0 in group 0.
- **Bit 3:2 - EPRX0DataRate[1:0]** - Sets the data rate for group 0.

EPRX0DataRate[1:0]	5 Gbps	10 Gbps
2'd0	disabled	disabled
2'd1	160M bps	320 Mbps
2'd2	320 Mbps	640 Mbps
2'd3	640 Mbps	1280 Mbps

- **Bit 1:0 - EPRX0TrackMode[1:0]** - Sets the phase tracking mode for group 0.

EPRX0TrackMode[1:0]	Mode
2'd0	Fixed phase
2'd1	Initial training
2'd2	Continuous phase tracking
2'd3	Continuous phase tracking with initial phase

[0x0c5] EPRX1Control

Configuration of ePortRx Group 1

- **Bit 7 - EPRX13Enable** - Enables channel 3 in group 1.
- **Bit 6 - EPRX12Enable** - Enables channel 2 in group 1.
- **Bit 5 - EPRX11Enable** - Enables channel 1 in group 1.
- **Bit 4 - EPRX10Enable** - Enables channel 0 in group 1.
- **Bit 3:2 - EPRX1DataRate[1:0]** - Sets the data rate for group 1.

EPRX1DataRate[1:0]	5 Gbps	10 Gbps
2'd0	disabled	disabled
2'd1	160M bps	320 Mbps
2'd2	320 Mbps	640 Mbps
2'd3	640 Mbps	1280 Mbps

- **Bit 1:0 - EPRX1TrackMode[1:0]** - Sets the phase tracking mode for group 1.

EPRX1TrackMode[1:0]	Mode
2'd0	Fixed phase
2'd1	Initial training
2'd2	Continuous phase tracking
2'd3	Continuous phase tracking with initial phase

[0x0c6] EPRX2Control

Configuration of ePortRx Group 2

- **Bit 7 - EPRX23Enable** - Enables channel 3 in group 2.
- **Bit 6 - EPRX22Enable** - Enables channel 2 in group 2.
- **Bit 5 - EPRX21Enable** - Enables channel 1 in group 2.
- **Bit 4 - EPRX20Enable** - Enables channel 0 in group 2.
- **Bit 3:2 - EPRX2DataRate[1:0]** - Sets the data rate for group 2.

EPRX2DataRate[1:0]	5 Gbps	10 Gbps
2'd0	disabled	disabled
2'd1	160M bps	320 Mbps
2'd2	320 Mbps	640 Mbps
2'd3	640 Mbps	1280 Mbps

- **Bit 1:0 - EPRX2TrackMode[1:0]** - Sets the phase tracking mode for group 2.

EPRX2TrackMode[1:0]	Mode
2'd0	Fixed phase
2'd1	Initial training
2'd2	Continuous phase tracking
2'd3	Continuous phase tracking with initial phase

[0x0c7] EPRX3Control

Configuration of ePortRx Group 3

- **Bit 7 - EPRX33Enable** - Enables channel 3 in group 3.
- **Bit 6 - EPRX32Enable** - Enables channel 2 in group 3.
- **Bit 5 - EPRX31Enable** - Enables channel 1 in group 3.
- **Bit 4 - EPRX30Enable** - Enables channel 0 in group 3.
- **Bit 3:2 - EPRX3DataRate[1:0]** - Sets the data rate for group 3.

EPRX3DataRate[1:0]	5 Gbps	10 Gbps
2'd0	disabled	disabled
2'd1	160M bps	320 Mbps
2'd2	320 Mbps	640 Mbps
2'd3	640 Mbps	1280 Mbps

- **Bit 1:0 - EPRX3TrackMode[1:0]** - Sets the phase tracking mode for group 3.

EPRX3TrackMode[1:0]	Mode
2'd0	Fixed phase
2'd1	Initial training
2'd2	Continuous phase tracking
2'd3	Continuous phase tracking with initial phase

[0x0c8] EPRX4Control

Configuration of ePortRx Group 4

- **Bit 7 - EPRX43Enable** - Enables channel 3 in group 4.
- **Bit 6 - EPRX42Enable** - Enables channel 2 in group 4.
- **Bit 5 - EPRX41Enable** - Enables channel 1 in group 4.
- **Bit 4 - EPRX40Enable** - Enables channel 0 in group 4.
- **Bit 3:2 - EPRX4DataRate[1:0]** - Sets the data rate for group 4.

EPRX4DataRate[1:0]	5 Gbps	10 Gbps
2'd0	disabled	disabled
2'd1	160M bps	320 Mbps
2'd2	320 Mbps	640 Mbps
2'd3	640 Mbps	1280 Mbps

- **Bit 1:0 - EPRX4TrackMode[1:0]** - Sets the phase tracking mode for group 4.

EPRX4TrackMode[1:0]	Mode
2'd0	Fixed phase
2'd1	Initial training
2'd2	Continuous phase tracking
2'd3	Continuous phase tracking with initial phase

[0x0c9] EPRX5Control

Configuration of ePortRx Group 5

- **Bit 7 - EPRX53Enable** - Enables channel 3 in group 5.
- **Bit 6 - EPRX52Enable** - Enables channel 2 in group 5.
- **Bit 5 - EPRX51Enable** - Enables channel 1 in group 5.
- **Bit 4 - EPRX50Enable** - Enables channel 0 in group 5.
- **Bit 3:2 - EPRX5DataRate[1:0]** - Sets the data rate for group 5.

EPRX5DataRate[1:0]	5 Gbps	10 Gbps
2'd0	disabled	disabled
2'd1	160M bps	320 Mbps
2'd2	320 Mbps	640 Mbps
2'd3	640 Mbps	1280 Mbps

- **Bit 1:0 - EPRX5TrackMode[1:0]** - Sets the phase tracking mode for group 5.

EPRX5TrackMode[1:0]	Mode
2'd0	Fixed phase
2'd1	Initial training
2'd2	Continuous phase tracking
2'd3	Continuous phase tracking with initial phase

[0x0ca] EPRX6Control

Configuration of ePortRx Group 6

- **Bit 7 - EPRX63Enable** - Enables channel 3 in group 6.
- **Bit 6 - EPRX62Enable** - Enables channel 2 in group 6.
- **Bit 5 - EPRX61Enable** - Enables channel 1 in group 6.
- **Bit 4 - EPRX60Enable** - Enables channel 0 in group 6.
- **Bit 3:2 - EPRX6DataRate[1:0]** - Sets the data rate for group 6.

EPRX6DataRate[1:0]	5 Gbps	10 Gbps
2'd0	disabled	disabled
2'd1	160M bps	320 Mbps
2'd2	320 Mbps	640 Mbps
2'd3	640 Mbps	1280 Mbps

- **Bit 1:0 - EPRX6TrackMode[1:0]** - Sets the phase tracking mode for group 6.

EPRX6TrackMode[1:0]	Mode
2'd0	Fixed phase
2'd1	Initial training
2'd2	Continuous phase tracking
2'd3	Continuous phase tracking with initial phase

[0x0cb] EPRXEcControl

Configuration of ePortRx EC channel

- **Bit 1:0 - EPRXEcTrackMode[1:0]** - Sets the phase tracking mode for EC channel

EPRXEcTrackMode[1:0]	Mode
2'd0	Fixed phase
2'd1	Initial training
2'd2	Continuous phase tracking
2'd3	Continuous phase tracking with initial phase

[0x0cc] EPRX00ChnCntr

Configuration of the channel 0 in group 0

- **Bit 7:4 - EPRX00PhaseSelect[3:0]** - Selects the phase for channel 0 in group 0.
- **Bit 3 - EPRX00Invert** - Inverts the channel 0 in group 0.
- **Bit 2 - EPRX00AcBias** - Enables the common mode generation for channel 0 in group 0.
- **Bit 1 - EPRX00Term** - Enables the 100 Ohm termination for channel 0 in group 0.
- **Bit 0 - EPRX00Eq[1]** - Equalization control for channel 0 in group 0.

[0xcd] EPRX01ChnCntr

Configuration of the channel 1 in group 0

- **Bit 7:4 - EPRX01PhaseSelect[3:0]** - Selects the phase for channel 1 in group 0.
- **Bit 3 - EPRX01Invert** - Inverts the channel 1 in group 0.
- **Bit 2 - EPRX01AcBias** - Enables the common mode generation for channel 1 in group 0.
- **Bit 1 - EPRX01Term** - Enables the 100 Ohm termination for channel 1 in group 0.
- **Bit 0 - EPRX01Eq[1]** - Equalization control for channel 1 in group 0.

[0xce] EPRX02ChnCntr

Configuration of the channel 2 in group 0

- **Bit 7:4 - EPRX02PhaseSelect[3:0]** - Selects the phase for channel 2 in group 0.
- **Bit 3 - EPRX02Invert** - Inverts the channel 2 in group 0.
- **Bit 2 - EPRX02AcBias** - Enables the common mode generation for channel 2 in group 0.
- **Bit 1 - EPRX02Term** - Enables the 100 Ohm termination for channel 2 in group 0.
- **Bit 0 - EPRX02Eq[1]** - Equalization control for channel 2 in group 0.

[0xcf] EPRX03ChnCntr

Configuration of the channel 3 in group 0

- **Bit 7:4 - EPRX03PhaseSelect[3:0]** - Selects the phase for channel 3 in group 0.
- **Bit 3 - EPRX03Invert** - Inverts the channel 3 in group 0.
- **Bit 2 - EPRX03AcBias** - Enables the common mode generation for channel 3 in group 0.
- **Bit 1 - EPRX03Term** - Enables the 100 Ohm termination for channel 3 in group 0.
- **Bit 0 - EPRX03Eq[1]** - Equalization control for channel 3 in group 0.

[0xd0] EPRX10ChnCntr

Configuration of the channel 0 in group 1

- **Bit 7:4 - EPRX10PhaseSelect[3:0]** - Selects the phase for channel 0 in group 1.
- **Bit 3 - EPRX10Invert** - Inverts the channel 0 in group 1.
- **Bit 2 - EPRX10AcBias** - Enables the common mode generation for channel 0 in group 1.
- **Bit 1 - EPRX10Term** - Enables the 100 Ohm termination for channel 0 in group 1.
- **Bit 0 - EPRX10Eq[1]** - Equalization control for channel 0 in group 1.

[0xd1] EPRX11ChnCntr

Configuration of the channel 1 in group 1

- **Bit 7:4 - EPRX11PhaseSelect[3:0]** - Selects the phase for channel 1 in group 1.
- **Bit 3 - EPRX11Invert** - Inverts the channel 1 in group 1.
- **Bit 2 - EPRX11AcBias** - Enables the common mode generation for channel 1 in group 1.
- **Bit 1 - EPRX11Term** - Enables the 100 Ohm termination for channel 1 in group 1.
- **Bit 0 - EPRX11Eq[1]** - Equalization control for channel 1 in group 1.

[0xd2] EPRX12ChnCntr

Configuration of the channel 2 in group 1

- **Bit 7:4 - EPRX12PhaseSelect[3:0]** - Selects the phase for channel 2 in group 1.
- **Bit 3 - EPRX12Invert** - Inverts the channel 2 in group 1.
- **Bit 2 - EPRX12AcBias** - Enables the common mode generation for channel 2 in group 1.
- **Bit 1 - EPRX12Term** - Enables the 100 Ohm termination for channel 2 in group 1.
- **Bit 0 - EPRX12Eq[1]** - Equalization control for channel 2 in group 1.

[0xd3] EPRX13ChnCntr

Configuration of the channel 3 in group 1

- **Bit 7:4 - EPRX13PhaseSelect[3:0]** - Selects the phase for channel 3 in group 1.
- **Bit 3 - EPRX13Invert** - Inverts the channel 3 in group 1.
- **Bit 2 - EPRX13AcBias** - Enables the common mode generation for channel 3 in group 1.
- **Bit 1 - EPRX13Term** - Enables the 100 Ohm termination for channel 3 in group 1.
- **Bit 0 - EPRX13Eq[1]** - Equalization control for channel 3 in group 1.

[0xd4] EPRX20ChnCntr

Configuration of the channel 0 in group 2

- **Bit 7:4 - EPRX20PhaseSelect[3:0]** - Selects the phase for channel 0 in group 2.
- **Bit 3 - EPRX20Invert** - Inverts the channel 0 in group 2.
- **Bit 2 - EPRX20AcBias** - Enables the common mode generation for channel 0 in group 2.
- **Bit 1 - EPRX20Term** - Enables the 100 Ohm termination for channel 0 in group 2.
- **Bit 0 - EPRX20Eq[1]** - Equalization control for channel 0 in group 2.

[0xd5] EPRX21ChnCntr

Configuration of the channel 1 in group 2

- **Bit 7:4 - EPRX21PhaseSelect[3:0]** - Selects the phase for channel 1 in group 2.
- **Bit 3 - EPRX21Invert** - Inverts the channel 1 in group 2.
- **Bit 2 - EPRX21AcBias** - Enables the common mode generation for channel 1 in group 2.
- **Bit 1 - EPRX21Term** - Enables the 100 Ohm termination for channel 1 in group 2.
- **Bit 0 - EPRX21Eq[1]** - Equalization control for channel 1 in group 2.

[0xd6] EPRX22ChnCntr

Configuration of the channel 2 in group 2

- **Bit 7:4 - EPRX22PhaseSelect[3:0]** - Selects the phase for channel 2 in group 2.
- **Bit 3 - EPRX22Invert** - Inverts the channel 2 in group 2.
- **Bit 2 - EPRX22AcBias** - Enables the common mode generation for channel 2 in group 2.
- **Bit 1 - EPRX22Term** - Enables the 100 Ohm termination for channel 2 in group 2.
- **Bit 0 - EPRX22Eq[1]** - Equalization control for channel 2 in group 2.

[0xd7] EPRX23ChnCntr

Configuration of the channel 3 in group 2

- **Bit 7:4 - EPRX23PhaseSelect[3:0]** - Selects the phase for channel 3 in group 2.
- **Bit 3 - EPRX23Invert** - Inverts the channel 3 in group 2.
- **Bit 2 - EPRX23AcBias** - Enables the common mode generation for channel 3 in group 2.
- **Bit 1 - EPRX23Term** - Enables the 100 Ohm termination for channel 3 in group 2.
- **Bit 0 - EPRX23Eq[1]** - Equalization control for channel 3 in group 2.

[0xd8] EPRX30ChnCntr

Configuration of the channel 0 in group 3

- **Bit 7:4 - EPRX30PhaseSelect[3:0]** - Selects the phase for channel 0 in group 3.
- **Bit 3 - EPRX30Invert** - Inverts the channel 0 in group 3.
- **Bit 2 - EPRX30AcBias** - Enables the common mode generation for channel 0 in group 3.
- **Bit 1 - EPRX30Term** - Enables the 100 Ohm termination for channel 0 in group 3.
- **Bit 0 - EPRX30Eq[1]** - Equalization control for channel 0 in group 3.

[0xd9] EPRX31ChnCntr

Configuration of the channel 1 in group 3

- **Bit 7:4 - EPRX31PhaseSelect[3:0]** - Selects the phase for channel 1 in group 3.
- **Bit 3 - EPRX31Invert** - Inverts the channel 1 in group 3.
- **Bit 2 - EPRX31AcBias** - Enables the common mode generation for channel 1 in group 3.
- **Bit 1 - EPRX31Term** - Enables the 100 Ohm termination for channel 1 in group 3.
- **Bit 0 - EPRX31Eq[1]** - Equalization control for channel 1 in group 3.

[0xda] EPRX32ChnCntr

Configuration of the channel 2 in group 3

- **Bit 7:4 - EPRX32PhaseSelect[3:0]** - Selects the phase for channel 2 in group 3.
- **Bit 3 - EPRX32Invert** - Inverts the channel 2 in group 3.
- **Bit 2 - EPRX32AcBias** - Enables the common mode generation for channel 2 in group 3.
- **Bit 1 - EPRX32Term** - Enables the 100 Ohm termination for channel 2 in group 3.
- **Bit 0 - EPRX32Eq[1]** - Equalization control for channel 2 in group 3.

[0xdb] EPRX33ChnCntr

Configuration of the channel 3 in group 3

- **Bit 7:4 - EPRX33PhaseSelect[3:0]** - Selects the phase for channel 3 in group 3.
- **Bit 3 - EPRX33Invert** - Inverts the channel 3 in group 3.
- **Bit 2 - EPRX33AcBias** - Enables the common mode generation for channel 3 in group 3.
- **Bit 1 - EPRX33Term** - Enables the 100 Ohm termination for channel 3 in group 3.
- **Bit 0 - EPRX33Eq[1]** - Equalization control for channel 3 in group 3.

[0xdc] EPRX40ChnCntr

Configuration of the channel 0 in group 4

- **Bit 7:4 - EPRX40PhaseSelect[3:0]** - Selects the phase for channel 0 in group 4.
- **Bit 3 - EPRX40Invert** - Inverts the channel 0 in group 4.
- **Bit 2 - EPRX40AcBias** - Enables the common mode generation for channel 0 in group 4.
- **Bit 1 - EPRX40Term** - Enables the 100 Ohm termination for channel 0 in group 4.
- **Bit 0 - EPRX40Eq[1]** - Equalization control for channel 0 in group 4.

[0xdd] EPRX41ChnCntr

Configuration of the channel 1 in group 4

- **Bit 7:4 - EPRX41PhaseSelect[3:0]** - Selects the phase for channel 1 in group 4.
- **Bit 3 - EPRX41Invert** - Inverts the channel 1 in group 4.
- **Bit 2 - EPRX41AcBias** - Enables the common mode generation for channel 1 in group 4.
- **Bit 1 - EPRX41Term** - Enables the 100 Ohm termination for channel 1 in group 4.
- **Bit 0 - EPRX41Eq[1]** - Equalization control for channel 1 in group 4.

[0xde] EPRX42ChnCntr

Configuration of the channel 2 in group 4

- **Bit 7:4 - EPRX42PhaseSelect[3:0]** - Selects the phase for channel 2 in group 4.
- **Bit 3 - EPRX42Invert** - Inverts the channel 2 in group 4.
- **Bit 2 - EPRX42AcBias** - Enables the common mode generation for channel 2 in group 4.
- **Bit 1 - EPRX42Term** - Enables the 100 Ohm termination for channel 2 in group 4.
- **Bit 0 - EPRX42Eq[1]** - Equalization control for channel 2 in group 4.

[0xdf] EPRX43ChnCntr

Configuration of the channel 3 in group 4

- **Bit 7:4 - EPRX43PhaseSelect[3:0]** - Selects the phase for channel 3 in group 4.
- **Bit 3 - EPRX43Invert** - Inverts the channel 3 in group 4.
- **Bit 2 - EPRX43AcBias** - Enables the common mode generation for channel 3 in group 4.
- **Bit 1 - EPRX43Term** - Enables the 100 Ohm termination for channel 3 in group 4.
- **Bit 0 - EPRX43Eq[1]** - Equalization control for channel 3 in group 4.

[0xe0] EPRX50ChnCntr

Configuration of the channel 0 in group 5

- **Bit 7:4 - EPRX50PhaseSelect[3:0]** - Selects the phase for channel 0 in group 5.
- **Bit 3 - EPRX50Invert** - Inverts the channel 0 in group 5.
- **Bit 2 - EPRX50AcBias** - Enables the common mode generation for channel 0 in group 5.
- **Bit 1 - EPRX50Term** - Enables the 100 Ohm termination for channel 0 in group 5.
- **Bit 0 - EPRX50Eq[1]** - Equalization control for channel 0 in group 5.

[0x0e1] EPRX51ChnCntr

Configuration of the channel 1 in group 5

- **Bit 7:4 - EPRX51PhaseSelect[3:0]** - Selects the phase for channel 1 in group 5.
- **Bit 3 - EPRX51Invert** - Inverts the channel 1 in group 5.
- **Bit 2 - EPRX51AcBias** - Enables the common mode generation for channel 1 in group 5.
- **Bit 1 - EPRX51Term** - Enables the 100 Ohm termination for channel 1 in group 5.
- **Bit 0 - EPRX51Eq[1]** - Equalization control for channel 1 in group 5.

[0x0e2] EPRX52ChnCntr

Configuration of the channel 2 in group 5

- **Bit 7:4 - EPRX52PhaseSelect[3:0]** - Selects the phase for channel 2 in group 5.
- **Bit 3 - EPRX52Invert** - Inverts the channel 2 in group 5.
- **Bit 2 - EPRX52AcBias** - Enables the common mode generation for channel 2 in group 5.
- **Bit 1 - EPRX52Term** - Enables the 100 Ohm termination for channel 2 in group 5.
- **Bit 0 - EPRX52Eq[1]** - Equalization control for channel 2 in group 5.

[0x0e3] EPRX53ChnCntr

Configuration of the channel 3 in group 5

- **Bit 7:4 - EPRX53PhaseSelect[3:0]** - Selects the phase for channel 3 in group 5.
- **Bit 3 - EPRX53Invert** - Inverts the channel 3 in group 5.
- **Bit 2 - EPRX53AcBias** - Enables the common mode generation for channel 3 in group 5.
- **Bit 1 - EPRX53Term** - Enables the 100 Ohm termination for channel 3 in group 5.
- **Bit 0 - EPRX53Eq[1]** - Equalization control for channel 3 in group 5.

[0x0e4] EPRX60ChnCntr

Configuration of the channel 0 in group 6

- **Bit 7:4 - EPRX60PhaseSelect[3:0]** - Selects the phase for channel 0 in group 6.
- **Bit 3 - EPRX60Invert** - Inverts the channel 0 in group 6.
- **Bit 2 - EPRX60AcBias** - Enables the common mode generation for channel 0 in group 6.
- **Bit 1 - EPRX60Term** - Enables the 100 Ohm termination for channel 0 in group 6.
- **Bit 0 - EPRX60Eq[1]** - Equalization control for channel 0 in group 6.

[0xe5] EPRX61ChnCntr

Configuration of the channel 1 in group 6

- **Bit 7:4 - EPRX61PhaseSelect[3:0]** - Selects the phase for channel 1 in group 6.
- **Bit 3 - EPRX61Invert** - Inverts the channel 1 in group 6.
- **Bit 2 - EPRX61AcBias** - Enables the common mode generation for channel 1 in group 6.
- **Bit 1 - EPRX61Term** - Enables the 100 Ohm termination for channel 1 in group 6.
- **Bit 0 - EPRX61Eq[1]** - Equalization control for channel 1 in group 6.

[0xe6] EPRX62ChnCntr

Configuration of the channel 2 in group 6

- **Bit 7:4 - EPRX62PhaseSelect[3:0]** - Selects the phase for channel 2 in group 6.
- **Bit 3 - EPRX62Invert** - Inverts the channel 2 in group 6.
- **Bit 2 - EPRX62AcBias** - Enables the common mode generation for channel 2 in group 6.
- **Bit 1 - EPRX62Term** - Enables the 100 Ohm termination for channel 2 in group 6.
- **Bit 0 - EPRX62Eq[1]** - Equalization control for channel 2 in group 6.

[0xe7] EPRX63ChnCntr

Configuration of the channel 3 in group 6

- **Bit 7:4 - EPRX63PhaseSelect[3:0]** - Selects the phase for channel 3 in group 6.
- **Bit 3 - EPRX63Invert** - Inverts the channel 3 in group 6.
- **Bit 2 - EPRX63AcBias** - Enables the common mode generation for channel 3 in group 6.
- **Bit 1 - EPRX63Term** - Enables the 100 Ohm termination for channel 3 in group 6.
- **Bit 0 - EPRX63Eq[1]** - Equalization control for channel 3 in group 6.

[0xe8] EPRXEcChnCntr

Configuration of the EC channel in ePortRx

- **Bit 7:4 - EPRXECPhaseSelect[3:0]** - Static phase for the EC channel.
- **Bit 3 - EPRXECInvert** - Inverts the EC channel data input.
- **Bit 2 - EPRXECAcBias** - Enables the common mode generation for the EC channel.
- **Bit 1 - EPRXECTerm** - Enables the 100 Ohm termination for EC channel.
- **Bit 0 - EPRXECEnable** - Enables the EC channel.

[0x0e9] EPRXEq10Control

Eport Rx equalization control for groups 1 and 0

- **Bit 7 - EPRX13Eq[0]** - Equalization control for channel 3 in group 1.

EPRX13Eq[1:0]	Zero location [MHz]	Peaking [dB]
2'd0	N/A	N/A
2'd1	300	4.9
2'd2	125	7.8
2'd3	70	10.7

- **Bit 6 - EPRX12Eq[0]** - Equalization control for channel 2 in group 1.

EPRX12Eq[1:0]	Zero location [MHz]	Peaking [dB]
2'd0	N/A	N/A
2'd1	300	4.9
2'd2	125	7.8
2'd3	70	10.7

- **Bit 5 - EPRX11Eq[0]** - Equalization control for channel 1 in group 1.

EPRX11Eq[1:0]	Zero location [MHz]	Peaking [dB]
2'd0	N/A	N/A
2'd1	300	4.9
2'd2	125	7.8
2'd3	70	10.7

- **Bit 4 - EPRX10Eq[0]** - Equalization control for channel 0 in group 1.

EPRX10Eq[1:0]	Zero location [MHz]	Peaking [dB]
2'd0	N/A	N/A
2'd1	300	4.9
2'd2	125	7.8
2'd3	70	10.7

- **Bit 3 - EPRX03Eq[0]** - Equalization control for channel 3 in group 0.

EPRX03Eq[1:0]	Zero location [MHz]	Peaking [dB]
2'd0	N/A	N/A
2'd1	300	4.9
2'd2	125	7.8
2'd3	70	10.7

- **Bit 2 - EPRX02Eq[0]** - Equalization control for channel 2 in group 0.

EPRX02Eq[1:0]	Zero location [MHz]	Peaking [dB]
2'd0	N/A	N/A
2'd1	300	4.9
2'd2	125	7.8
2'd3	70	10.7

- **Bit 1 - EPRX01Eq[0]** - Equalization control for channel 1 in group 0.

EPRX01Eq[1:0]	Zero location [MHz]	Peaking [dB]
2'd0	N/A	N/A
2'd1	300	4.9
2'd2	125	7.8
2'd3	70	10.7

- **Bit 0 - EPRX00Eq[0]** - Equalization control for channel 0 in group 0.

EPRX00Eq[1:0]	Zero location [MHz]	Peaking [dB]
2'd0	N/A	N/A
2'd1	300	4.9
2'd2	125	7.8
2'd3	70	10.7

[0x0ea] EPRXEq32Control

Eport Rx equalization control for groups 3 and 2

- **Bit 7 - EPRX33Eq[0]** - Equalization control for channel 3 in group 3.

EPRX33Eq[1:0]	Zero location [MHz]	Peaking [dB]
2'd0	N/A	N/A
2'd1	300	4.9
2'd2	125	7.8
2'd3	70	10.7

- **Bit 6 - EPRX32Eq[0]** - Equalization control for channel 2 in group 3.

EPRX32Eq[1:0]	Zero location [MHz]	Peaking [dB]
2'd0	N/A	N/A
2'd1	300	4.9
2'd2	125	7.8
2'd3	70	10.7

- **Bit 5 - EPRX31Eq[0]** - Equalization control for channel 1 in group 3.

EPRX31Eq[1:0]	Zero location [MHz]	Peaking [dB]
2'd0	N/A	N/A
2'd1	300	4.9
2'd2	125	7.8
2'd3	70	10.7

- **Bit 4 - EPRX30Eq[0]** - Equalization control for channel 0 in group 3.

EPRX30Eq[1:0]	Zero location [MHz]	Peaking [dB]
2'd0	N/A	N/A
2'd1	300	4.9
2'd2	125	7.8
2'd3	70	10.7

- **Bit 3 - EPRX23Eq[0]** - Equalization control for channel 3 in group 2.

EPRX23Eq[1:0]	Zero location [MHz]	Peaking [dB]
2'd0	N/A	N/A
2'd1	300	4.9
2'd2	125	7.8
2'd3	70	10.7

- **Bit 2 - EPRX22Eq[0]** - Equalization control for channel 2 in group 2.

EPRX22Eq[1:0]	Zero location [MHz]	Peaking [dB]
2'd0	N/A	N/A
2'd1	300	4.9
2'd2	125	7.8
2'd3	70	10.7

- **Bit 1 - EPRX21Eq[0]** - Equalization control for channel 1 in group 2.

EPRX21Eq[1:0]	Zero location [MHz]	Peaking [dB]
2'd0	N/A	N/A
2'd1	300	4.9
2'd2	125	7.8
2'd3	70	10.7

- **Bit 0 - EPRX20Eq[0]** - Equalization control for channel 0 in group 2.

EPRX20Eq[1:0]	Zero location [MHz]	Peaking [dB]
2'd0	N/A	N/A
2'd1	300	4.9
2'd2	125	7.8
2'd3	70	10.7

[0x0eb] EPRXEq54Control

Eport Rx equalization control for groups 5 and 4

- **Bit 7 - EPRX53Eq[0]** - Equalization control for channel 3 in group 5.

EPRX53Eq[1:0]	Zero location [MHz]	Peaking [dB]
2'd0	N/A	N/A
2'd1	300	4.9
2'd2	125	7.8
2'd3	70	10.7

- **Bit 6 - EPRX52Eq[0]** - Equalization control for channel 2 in group 5.

EPRX52Eq[1:0]	Zero location [MHz]	Peaking [dB]
2'd0	N/A	N/A
2'd1	300	4.9
2'd2	125	7.8
2'd3	70	10.7

- **Bit 5 - EPRX51Eq[0]** - Equalization control for channel 1 in group 5.

EPRX51Eq[1:0]	Zero location [MHz]	Peaking [dB]
2'd0	N/A	N/A
2'd1	300	4.9
2'd2	125	7.8
2'd3	70	10.7

- **Bit 4 - EPRX50Eq[0]** - Equalization control for channel 0 in group 5.

EPRX50Eq[1:0]	Zero location [MHz]	Peaking [dB]
2'd0	N/A	N/A
2'd1	300	4.9
2'd2	125	7.8
2'd3	70	10.7

- **Bit 3 - EPRX43Eq[0]** - Equalization control for channel 3 in group 4.

EPRX43Eq[1:0]	Zero location [MHz]	Peaking [dB]
2'd0	N/A	N/A
2'd1	300	4.9
2'd2	125	7.8
2'd3	70	10.7

- **Bit 2 - EPRX42Eq[0]** - Equalization control for channel 2 in group 4.

EPRX42Eq[1:0]	Zero location [MHz]	Peaking [dB]
2'd0	N/A	N/A
2'd1	300	4.9
2'd2	125	7.8
2'd3	70	10.7

- **Bit 1 - EPRX41Eq[0]** - Equalization control for channel 1 in group 4.

EPRX41Eq[1:0]	Zero location [MHz]	Peaking [dB]
2'd0	N/A	N/A
2'd1	300	4.9
2'd2	125	7.8
2'd3	70	10.7

- **Bit 0 - EPRX40Eq[0]** - Equalization control for channel 0 in group 4.

EPRX40Eq[1:0]	Zero location [MHz]	Peaking [dB]
2'd0	N/A	N/A
2'd1	300	4.9
2'd2	125	7.8
2'd3	70	10.7

[0x0ec] EPRXEq6Control

Eport Rx equalization control for group6

- **Bit 3 - EPRX63Eq[0]** - Equalization control for channel 3 in group 6.

EPRX63Eq[1:0]	Zero location [MHz]	Peaking [dB]
2'd0	N/A	N/A
2'd1	300	4.9
2'd2	125	7.8
2'd3	70	10.7

- **Bit 2 - EPRX62Eq[0]** - Equalization control for channel 2 in group 6.

EPRX62Eq[1:0]	Zero location [MHz]	Peaking [dB]
2'd0	N/A	N/A
2'd1	300	4.9
2'd2	125	7.8
2'd3	70	10.7

- **Bit 1 - EPRX61Eq[0]** - Equalization control for channel 1 in group 6.

EPRX61Eq[1:0]	Zero location [MHz]	Peaking [dB]
2'd0	N/A	N/A
2'd1	300	4.9
2'd2	125	7.8
2'd3	70	10.7

- **Bit 0 - EPRX60Eq[0]** - Equalization control for channel 0 in group 6.

EPRX60Eq[1:0]	Zero location [MHz]	Peaking [dB]
2'd0	N/A	N/A
2'd1	300	4.9
2'd2	125	7.8
2'd3	70	10.7

[0x0ed] POWERUP0

Controls behavior of the power up state machine (for more details refer to [Power-up state machine](#) (page 61))

- **Bit 6 - PUSMpllWdogDisable** - Disables watch dog monitoring PLL locked signal
- **Bit 5 - PUSMdllWdogDisable** - Disables watch dog monitoring DLL locked signal
- **Bit 4 - PUSMReadyWhenChnsLocked** - When selected, ready signal is reported only after all enabled channels in all ePortRx groups are locked
- **Bit 3:0 - PUSMPLLTimeoutConfig[3:0]** - Determines the timeout duration in the **WAIT_PLL_LOCK** state in the power up state machine. For more details see [Power-up state machine](#) (page 61)

PUSMPIITimeoutConfig[3:0]	Wait time
4'd0	disabled
4'd1	10 us
4'd2	20 us
4'd3	50 us
4'd4	100 us
4'd5	200 us
4'd6	500 us
4'd7	1 ms
4'd8	2 ms
4'd9	5 ms
4'd10	10 ms
4'd11	20 ms
4'd12	50 ms
4'd13	100 ms
4'd14	500 ms
4'd15	1 s

[0xee] POWERUP1

Controls behavior of the power up state machine (for more details refer to *Power-up state machine* (page 61))

- **Bit 7:4 - PUSMDllTimeoutConfig[3:0]** - Determines the timeout duration in the **WAIT_DLL_LOCK** state in the power up state machine. For more details see *Power-up state machine* (page 61)

PUSMDllTimeoutConfig[3:0]	Wait time
4'd0	disabled
4'd1	10 us
4'd2	20 us
4'd3	50 us
4'd4	100 us
4'd5	200 us
4'd6	500 us
4'd7	1 ms
4'd8	2 ms
4'd9	5 ms
4'd10	10 ms
4'd11	20 ms
4'd12	50 ms
4'd13	100 ms
4'd14	500 ms
4'd15	1 s

- **Bit 3:0 - PUSMChannelsTimeoutConfig[3:0]** - Determines the timeout duration in the **WAIT_CHNS_LOCKED** state in the power up state machine. For more details see *Power-up state machine* (page 61).

PUSMChannelsTimeoutConfig[3:0]	Wait time
4'd0	disabled
4'd1	10 us
4'd2	20 us
4'd3	50 us
4'd4	100 us
4'd5	200 us
4'd6	500 us
4'd7	1 ms
4'd8	2 ms
4'd9	5 ms
4'd10	10 ms
4'd11	20 ms
4'd12	50 ms
4'd13	100 ms
4'd14	500 ms
4'd15	1 s

15.1.17 Power Up State Machine

[0x0ef] POWERUP2

Controls behavior of the power up state machine (for more details refer to [Power-up state machine](#) (page 61))

- **Bit 2 - dllConfigDone** - When asserted, the power up state machine is allowed to proceed to PLL initialization. Please refer [Configuration](#) (page 17) for more details.
- **Bit 1 - pllConfigDone** - When asserted, the power up state machine is allowed to proceed to initialization of components containing DLLs (ePortRx, phase-shifter). Please refer [Configuration](#) (page 17) for more details.
- **Bit 0 - updateEnable** - When asserted, the power up state machine copies the values from fuses to configuration registers during power. Please refer [Configuration](#) (page 17) for more details.

15.2 Read/Write

15.2.1 I2C Masters

[0x0f0] I2CM0Config

General configuration register for I2C Master 0

- **Bit 6 - I2CM0SCLPullUpEnable** - Enable pull up for M0SCL pin.
- **Bit 5 - I2CM0SCLDriveStrength** - M0SCL drive strength ([CMOS I/O Pin Characteristics](#) (page 310))
- **Bit 4 - I2CM0SDAPullUpEnable** - Enable pull up for M0SDA pin.
- **Bit 3 - I2CM0SDADriveStrength** - M0SDA drive strength ([CMOS I/O Pin Characteristics](#) (page 310)).
- **Bit 2:0 - I2CM0AddressExt[2:0]** - 3 additional bits of address of slave to address in an I2C transaction for I2C Master 0; only used in commands with 10-bit addressing

[0x0f1] I2CM0Address

7 bits of address of slave to address in an I2C transaction for I2C Master 0

- **Bit 6:0 - I2CM0Address[6:0]** - 7 bits of address of slave to address in an I2C transaction

[0x0f2] I2CM0Data0

Data input for I2C Master 0

- **Bit 7:0 - I2CM0Data[7:0]** - Bits 7:0 of the data input

[0x0f3] I2CM0Data1

Data input for I2C Master 0

- **Bit 7:0 - I2CM0Data[15:8]** - Bits 15:8 of the data input

[0x0f4] I2CM0Data2

Data input for I2C Master 0

- **Bit 7:0 - I2CM0Data[23:16]** - Bits 23:16 of the data input

[0x0f5] I2CM0Data3

Data input for I2C Master 0

- **Bit 7:0 - I2CM0Data[31:24]** - Bits 31:24 of the data input

[0x0f6] I2CM0Cmd

Command word register for I2C Master 0

- **Bit 3:0 - I2CM0Cmd[3:0]** - Command word.

[0x0f7] I2CM1Config

General configuration register for I2C Master 1

- **Bit 6 - I2CM1SCLPullUpEnable** - Enable pull up for M1SCL pin.
- **Bit 5 - I2CM1SCLDriveStrength** - M1SCL drive strength (*CMOS I/O Pin Characteristics* (page 310))
- **Bit 4 - I2CM1SDAPullUpEnable** - Enable pull up for M1SDA pin.
- **Bit 3 - I2CM1SDADriveStrength** - M1SDA drive strength (*CMOS I/O Pin Characteristics* (page 310)).
- **Bit 2:0 - I2CM1AddressExt[2:0]** - 3 additional bits of address of slave to address in an I2C transaction for I2C Master 1; only used in commands with 10-bit addressing

[0x0f8] I2CM1Address

7 bits of address of slave to address in an I2C transaction for I2C Master 1

- **Bit 6:0 - I2CM1Address[6:0]** - 7 bits of address of slave to address in an I2C transaction

[0x0f9] I2CM1Data0

Data input for I2C Master 1

- **Bit 7:0 - I2CM1Data[7:0]** - Bits 7:0 of the data input

[0x0fa] I2CM1Data1

Data input for I2C Master 1

- **Bit 7:0 - I2CM1Data[15:8]** - Bits 15:8 of the data input

[0x0fb] I2CM1Data2

Data input for I2C Master 1

- **Bit 7:0 - I2CM1Data[23:16]** - Bits 23:16 of the data input

[0x0fc] I2CM1Data3

Data input for I2C Master 1

- **Bit 7:0 - I2CM1Data[31:24]** - Bits 31:24 of the data input

[0x0fd] I2CM1Cmd

Command word register for I2C Master 1

- **Bit 3:0 - I2CM1Cmd[3:0]** - Command word.

[0x0fe] I2CM2Config

General configuration register for I2C Master 2

- **Bit 6 - I2CM2SCLPullUpEnable** - Enable pull up for M2SCL pin.
- **Bit 5 - I2CM2SCLDriveStrength** - M2SCL drive strength (*CMOS I/O Pin Characteristics* (page 310))
- **Bit 4 - I2CM2SDAPullUpEnable** - Enable pull up for M2SDA pin.
- **Bit 3 - I2CM2SDADriveStrength** - M2SDA drive strength (*CMOS I/O Pin Characteristics* (page 310)).
- **Bit 2:0 - I2CM2AddressExt[2:0]** - 3 additional bits of address of slave to address in an I2C transaction for I2C Master 2; only used in commands with 10-bit addressing

[0x0ff] I2CM2Address

7 bits of address of slave to address in an I2C transaction for I2C Master 2

- **Bit 6:0 - I2CM2Address[6:0]** - 7 bits of address of slave to address in an I2C transaction

[0x100] I2CM2Data0

Data input for I2C Master 2

- **Bit 7:0 - I2CM2Data[7:0]** - Bits 7:0 of the data input

[0x101] I2CM2Data1

Data input for I2C Master 2

- **Bit 7:0 - I2CM2Data[15:8]** - Bits 15:8 of the data input

[0x102] I2CM2Data2

Data input for I2C Master 2

- **Bit 7:0 - I2CM2Data[23:16]** - Bits 23:16 of the data input

[0x103] I2CM2Data3

Data input for I2C Master 2

- **Bit 7:0 - I2CM2Data[31:24]** - Bits 31:24 of the data input

[0x104] I2CM2Cmd

Command word register for I2C Master 2

- **Bit 3:0 - I2CM2Cmd[3:0]** - Command word.

15.2.2 ePortRx

[0x105] EPRXTrain10

Channel phase training request for groups 1 and 0

- **Bit 7:4 - EPRX1Train[3:0]** - **Initialize phase training. N-th bit control N-th channel training.** One should assert bits corresponding to channels to be trained and dessert them.
- **Bit 3:0 - EPRX0Train[3:0]** - **Initialize phase training. N-th bit control N-th channel training.** One should assert bits corresponding to channels to be trained and dessert them.

[0x106] EPRXTrain32

Channel phase training request for groups 3 and 2

- **Bit 7:4 - EPRX3Train[3:0] - Initialize phase training. N-th bit control N-th channel training.** One should assert bits corresponding to channels to be trained and dessert them.
- **Bit 3:0 - EPRX2Train[3:0] - Initialize phase training. N-th bit control N-th channel training.** One should assert bits corresponding to channels to be trained and dessert them.

[0x107] EPRXTrain54

Channel phase training request for groups 5 and 4

- **Bit 7:4 - EPRX5Train[3:0] - Initialize phase training. N-th bit control N-th channel training.** One should assert bits corresponding to channels to be trained and dessert them.
- **Bit 3:0 - EPRX4Train[3:0] - Initialize phase training. N-th bit control N-th channel training.** One should assert bits corresponding to channels to be trained and dessert them.

[0x108] EPRXTrainEc6

Channel phase training request for group 6 and EC channel

- **Bit 4 - EPRXEcTrain** - Initialize phase training for EC channel.
- **Bit 3:0 - EPRX6Train[3:0] - Initialize phase training. N-th bit control N-th channel training.** One should assert bits corresponding to channels to be trained and dessert them.

15.2.3 E-FUSES

[0x109] FUSEControl

Fuse control register.

- **Bit 7:4 - FuseBlowPulseLength[3:0]** - Duration of fuse blowing pulse (default:12).
- **Bit 1 - FuseRead** - Execute fuse readout sequence.
- **Bit 0 - FuseBlow** - Execute fuse blowing sequence.

[0x10a] FUSEBlowDataA

Data to be programmed to the fuses.

- **Bit 7:0 - FuseBlowData[7:0]** - Bits 7:0 of the data word.

[0x10b] FUSEBlowDataB

Data to be programmed to the fuses.

- **Bit 7:0 - FuseBlowData[15:8]** - Bits 15:8 of the data word.

[0x10c] FUSEBlowDataC

Data to be programmed to the fuses.

- **Bit 7:0 - FuseBlowData[23:16]** - Bits 23:16 of the data word.

[0x10d] FUSEBlowDataD

Data to be programmed to the fuses.

- **Bit 7:0 - FuseBlowData[31:24]** - Bits 31:24 of the data word.

[0x10e] FUSEBlowAddH

Address of the fuse block to be programmed.

- **Bit 7:0 - FuseBlowAddress[15:8]** - Bits 15:8 of the address.

[0x10f] FUSEBlowAddL

Address of the fuse block to be programmed.

- **Bit 7:0 - FuseBlowAddress[7:0]** - Bits 7:0 of the address.

[0x110] FuseMagic

Registers containing magic number for the fuse block.

- **Bit 7:0 - FuseMagicNumber[7:0]** - One has to write a magic number 0xA3 to this register in order to unlock fuse blowing.

15.2.4 ADC

[0x111] ADCSelect

ADC MUXes control.

- **Bit 7:4 - ADCInPSelect[3:0]** - Controls MUX for ADC Positive Input

ADCInPSelect[3:0]	Input
4'd0	ADC0 (external pin)
4'd1	ADC1 (external pin)
4'd2	ADC2 (external pin)
4'd3	ADC3 (external pin)
4'd4	ADC4 (external pin)
4'd5	ADC5 (external pin)
4'd6	ADC6 (external pin)
4'd7	ADC7 (external pin)
4'd8	EOM DAC (internal signal)
4'd9	VDDIO * 0.42 (internal signal)
4'd10	VDDTX * 0.42 (internal signal)
4'd11	VDDRX * 0.42 (internal signal)
4'd12	VDD * 0.42 (internal signal)
4'd13	VDDA * 0.42 (internal signal)
4'd14	Temperature sensor (internal signal)
4'd15	VREF / 2 (internal signal)

- **Bit 3:0 - ADCInNSelect[3:0]** - Controls MUX for ADC Negative Input

ADCInNSelect[3:0]	Input
4'd0	ADC0 (external pin)
4'd1	ADC1 (external pin)
4'd2	ADC2 (external pin)
4'd3	ADC3 (external pin)
4'd4	ADC4 (external pin)
4'd5	ADC5 (external pin)
4'd6	ADC6 (external pin)
4'd7	ADC7 (external pin)
4'd8	EOM DAC (internal signal)
4'd9	VDDIO * 0.42 (internal signal)
4'd10	VDDTX * 0.42 (internal signal)
4'd11	VDDRX * 0.42 (internal signal)
4'd12	VDD * 0.42 (internal signal)
4'd13	VDDA * 0.42 (internal signal)
4'd14	Temperature sensor (internal signal)
4'd15	VREF / 2 (internal signal)

See also: [\[0x113\] ADCCConfig](#) (page 233), [\[0x112\] ADCMon](#) (page 232)

[0x112] ADCMon

Control ADC's internal signals

- **Bit 5 - TEMPSENSReset** - Resets temperature sensor.
- **Bit 4 - VDDmonEna** - Enable resistive divider for VDD probing.
- **Bit 3 - VDDTXmonEna** - Enable resistive divider for VDDTX probing.
- **Bit 2 - VDDRXmonEna** - Enable resistive divider for VDDRX probing.
- **Bit 1 - VDDPSTmonEna** - Enable resistive divider for VDDPST probing.

- **Bit 0 - VDDANmonEna** - Enable resistive divider for VDDAB probing.

See also: [\[0x113\] ADCConfig](#) (page 233), [\[0x111\] ADCSelect](#) (page 231)

[0x113] ADCConfig

ADC configuration register

- **Bit 7 - ADCConvert** - Start ADC conversion.
- **Bit 2 - ADCEnable** - Enables ADC core and differential amplifier.
- **Bit 1:0 - ADCGainSelect[1:0]** - Selects gain for the differential amplifier

ADCGainSel[1:0]	Gain
2'd0	x2
2'd1	x8
2'd2	x16
2'd3	x32

See also: [\[0x112\] ADCMon](#) (page 232), [\[0x111\] ADCSelect](#) (page 231)

15.2.5 Eye Opening Monitor

[0x114] EOMConfigH

- **Bit 7:4 - EOMendOfCountSel[3:0]** - Amount of refClk clock cycles (40 MHz cycles) the EOM counter is gated ($2^{(selEndOfCount+1)}$)
- **Bit 2 - EOMBypassPhaseInterpolator** - Bypass the VCO 5.12 GHz clock (uses the refClk as the phase interpolated clock; the phase interpolation has to be done off-chip) [0 - vco, 1 - refClk]
- **Bit 1 - EOMStart** - Starts EOM acquisition
- **Bit 0 - EOMEnable** - Enables the EOM; wait few ms for all bias voltages to stabilize before starting EOM measurement

[0x115] EOMConfigL

- **Bit 5:0 - EOMphaseSel[5:0]** - Selects the sampling phase from the phase interpolation block; steps $[0:1/(fvco*64):63/(fvco*64)]$ s

[0x116] EOMvofSel

- **Bit 4:0 - EOMvofSel[4:0]** - Selects the comparison voltage; the comparator is differential; steps [-VDDRX/2:VDDRX/30:VDDRX/2] V; value 5'd32 is invalid

15.2.6 Process Monitor

[0x117] ProcessAndSeuMonitor

Process Monitor block configuration register

- **Bit 4 - DLDPFecCounterEnable** - Enable downlink FEC counter.
- **Bit 3 - SEUEnable** - Enable SEU counter.
- **Bit 2:1 - PMChannel[1:0]** - Select process monitor channel to be measured.
- **Bit 0 - PMEnable** - Enable process monitor block.

15.2.7 Testing

[0x118] ULDataSource0

Uplink data path test patterns.

- **Bit 7:5 - ULECDatasource[2:0]** - Data source for uplink EC channel

ULGECData-Source[2:0]	Name	Description
3'd0	EPORTRX_DATA	Normal mode of operation, data from ePortRx
3'd1	PRBS7	PRBS7 test pattern
3'd2	BIN_CNTR_UP	Binary counter counting up
3'd3	BIN_CNTR_DOWN	Binary counter counting down
3'd4	CONST_PATTERN	Constant pattern (DPDataPattern[1:0])
3'd5	CONST_PATTERN_IN	Constant pattern inverted (~DPDataPattern[1:0])
3'd6	DL-DATA_LOOPBACK	Loop back, downlink frame data
3'd7	Reserved	Reserved

- **Bit 3:0 - ULSerTestPattern[3:0]** - Controls the serializer data source.

ULSerTestPattern[3:0]	Name	Description
4'd0	DATA	Normal mode of operation
4'd1	PRBS7	PRBS7 test pattern
4'd2	PRBS15	PRBS15 test pattern
4'd3	PRBS23	PRBS23 test pattern
4'd4	PRBS31	PRBS31 test pattern
4'd5	CLK5G12	5.12 GHz clock pattern (in 5Gbps mode it will produce only 2.56 GHz)
4'd6	CLK2G56	2.56 GHz clock pattern
4'd7	CLK1G28	1.28 GHz clock pattern
4'd8	CLK40M	40 MHz clock pattern
4'd9	DLFRAME_10G24	Loopback, downlink frame repeated 4 times
4'd10	DLFRAME_5G12	Loopback, downlink frame repeated 2 times, each bit repeated 2 times
4'd11	DLFRAME_2G56	Loopback, downlink frame repeated 1 times, each bit repeated 4 times
4'd12	CONST PATTERN	8 x DPDataPattern[31:0]
4'd13	-	Reserved
4'd14	-	Reserved
4'd15	-	Reserved

[0x119] ULDataSource1

Uplink data path test patterns.

- **Bit 7:6 - LDDDataSource[1:0]** - Data source for the line driver.

LDDDataSource[1:0]	Description
2'd0	Data from serializer (normal mode of operation)
2'd1	Equalizer output data loopback
2'd2	Data resampled by CDR loopback
2'd3	reserved

- **Bit 5:3 - ULG1DataSource[2:0]** - Data source for uplink data group 1

ULG1DataSource[2:0]Name	Description
3'd0	EPORTRX_DATA
3'd1	PRBS7
3'd2	BIN_CNTR_UP
3'd3	BIN_CNTR_DOWN
3'd4	CONST_PATTERN
3'd5	CONST_PATTERN_IN
3'd6	DL-DATA_LOOPBACK
3'd7	Reserved

- **Bit 2:0 - ULG0DataSource[2:0]** - Data source for uplink data group 0

ULG0DataSource[2:0]Name	Description
3'd0	EPORTRX_DATA
3'd1	PRBS7
3'd2	BIN_CNTR_UP
3'd3	BIN_CNTR_DOWN
3'd4	CONST_PATTERN
3'd5	CONST_PATTERN_IN
3'd6	DL-DATA_LOOPBACK
3'd7	Reserved

[0x11a] ULDataSource2

Uplink data path test patterns.

- **Bit 5:3 - ULG3DataSource[2:0]** - Data source for uplink data group 3

ULG3DataSource[2:0]Name	Description
3'd0	EPORTRX_DATA
3'd1	PRBS7
3'd2	BIN_CNTR_UP
3'd3	BIN_CNTR_DOWN
3'd4	CONST_PATTERN
3'd5	CONST_PATTERN_IN
3'd6	DL-DATA_LOOPBACK
3'd7	Reserved

- **Bit 2:0 - ULG2DataSource[2:0]** - Data source for uplink data group 2

ULG2DataSource[2:0]Name	Description
3'd0	EPORTRX_DATA
3'd1	PRBS7
3'd2	BIN_CNTR_UP
3'd3	BIN_CNTR_DOWN
3'd4	CONST_PATTERN
3'd5	CONST_PATTERN_IN
3'd6	DL-DATA_LOOPBACK
3'd7	Reserved

[0x11b] ULDataSource3

Uplink data path test patterns.

- Bit 5:3 - **ULG5DataSource[2:0]** - Data source for uplink data group 5

ULG5DataSource[2:0]Name	Description
3'd0	EPORTRX_DATA
3'd1	PRBS7
3'd2	BIN_CNTR_UP
3'd3	BIN_CNTR_DOWN
3'd4	CONST_PATTERN
3'd5	CONST_PATTERN_IN
3'd6	DL-DATA_LOOPBACK
3'd7	Reserved

- Bit 2:0 - **ULG4DataSource[2:0]** - Data source for uplink data group 4

ULG4DataSource[2:0]Name	Description
3'd0	EPORTRX_DATA
3'd1	PRBS7
3'd2	BIN_CNTR_UP
3'd3	BIN_CNTR_DOWN
3'd4	CONST_PATTERN
3'd5	CONST_PATTERN_IN
3'd6	DL-DATA_LOOPBACK
3'd7	Reserved

[0x11c] ULDataSource4

Uplink data path test patterns.

- Bit 7:6 - **DLECDDataSource[1:0]** -
- Bit 5:3 - **ULICDataSource[2:0]** -

- Bit 2:0 - **ULG6DataSource[2:0]** - Data source for uplink data group 6

ULG6DataSource[2:0]	Name	Description
3'd0	EPORTRX_DATA	Normal mode of operation, data from ePortRx
3'd1	PRBS7	PRBS7 test pattern
3'd2	BIN_CNTR_UP	Binary counter counting up
3'd3	BIN_CNTR_DOWN	Binary counter counting down
3'd4	CONST_PATTERN	Constant pattern (DPDataPattern [31:0])
3'd5	CONST_PATTERN_INVERTED	Constant pattern inverted (~DPDataPattern [31:0])
3'd6	DL-DATA_LOOPBACK	Loop back, downlink frame data
3'd7	Reserved	Reserved

[0x11d] ULDataSource5

- Bit 7:6 - **DLG3DataSource[1:0]** - Controls the ePortTx group 3 data source

DLG3DataSource[1:0]	Name	Description
2'd0	LINK_DATA	Normal mode of operation, data from ePortRx
2'd1	PRBS7	PRBS7 patten on each channel
2'd2	BIN_CNTR_UP	Binary counter counting up on each channel
2'd3	CONST_PATTERN	Constant pattern

- Bit 5:4 - **DLG2DataSource[1:0]** - Controls the ePortTx group 2 data source

DLG2DataSource[1:0]	Name	Description
2'd0	LINK_DATA	Normal mode of operation, data from ePortRx
2'd1	PRBS7	PRBS7 patten on each channel
2'd2	BIN_CNTR_UP	Binary counter counting up on each channel
2'd3	CONST_PATTERN	Constant pattern

- Bit 3:2 - **DLG1DataSource[1:0]** - Controls the ePortTx group 1 data source

DLG1DataSource[1:0]	Name	Description
2'd0	LINK_DATA	Normal mode of operation, data from ePortRx
2'd1	PRBS7	PRBS7 patten on each channel
2'd2	BIN_CNTR_UP	Binary counter counting up on each channel
2'd3	CONST_PATTERN	Constant pattern

- Bit 1:0 - **DLG0DataSource[1:0]** - Controls the ePortTx group 0 data source

DLG0DataSource[1:0]	Name	Description
2'd0	LINK_DATA	Normal mode of operation, data from ePortRx
2'd1	PRBS7	PRBS7 patten on each channel
2'd2	BIN_CNTR_UP	Binary counter counting up on each channel
2'd3	CONST_PATTERN	Constant pattern

[0x11e] DPDataPattern3

Constant patter to be used in test pattern generation/checking

- **Bit 7:0 - DPDataPattern[31:24]** - Bits 31:24 of the canst pattern.

[0x11f] DPDataPattern2

Constant patter to be used in test pattern generation/checking

- **Bit 7:0 - DPDataPattern[23:16]** - Bits 23:16 of the canst pattern.

[0x120] DPDataPattern1

Constant patter to be used in test pattern generation/checking

- **Bit 7:0 - DPDataPattern[15:8]** - Bits 15:8 of the canst pattern.

[0x121] DPDataPattern0

Constant patter to be used in test pattern generation/checking

- **Bit 7:0 - DPDataPattern[7:0]** - Bits 7:0 of the canst pattern.

[0x122] EPRXPRBS3

Control registers for build-in PRBS7 generators in ePortRx.

- **Bit 4 - EPRXECPrbsEnable** -
- **Bit 3 - EPRX63PrbsEnable** - Enables PRBS7 generator for channel 3 in group 6. If enabled, the data from the input pin are discarded.
- **Bit 2 - EPRX62PrbsEnable** - Enables PRBS7 generator for channel 2 in group 6. If enabled, the data from the input pin are discarded.
- **Bit 1 - EPRX61PrbsEnable** - Enables PRBS7 generator for channel 1 in group 6. If enabled, the data from the input pin are discarded.
- **Bit 0 - EPRX60PrbsEnable** - Enables PRBS7 generator for channel 0 in group 6. If enabled, the data from the input pin are discarded.

[0x123] EPRXPRBS2

Control registers for build-in PRBS7 generators in ePortRx.

- **Bit 7 - EPRX53PrbsEnable** - Enables PRBS7 generator for channel 3 in group 5. If enabled, the data from the input pin are discarded.
- **Bit 6 - EPRX52PrbsEnable** - Enables PRBS7 generator for channel 2 in group 5. If enabled, the data from the input pin are discarded.
- **Bit 5 - EPRX51PrbsEnable** - Enables PRBS7 generator for channel 1 in group 5. If enabled, the data from the input pin are discarded.
- **Bit 4 - EPRX50PrbsEnable** - Enables PRBS7 generator for channel 0 in group 5. If enabled, the data from the input pin are discarded.

- **Bit 3 - EPRX43PrbsEnable** - Enables PRBS7 generator for channel 3 in group 4. If enabled, the data from the input pin are discarded.
- **Bit 2 - EPRX42PrbsEnable** - Enables PRBS7 generator for channel 2 in group 4. If enabled, the data from the input pin are discarded.
- **Bit 1 - EPRX41PrbsEnable** - Enables PRBS7 generator for channel 1 in group 4. If enabled, the data from the input pin are discarded.
- **Bit 0 - EPRX40PrbsEnable** - Enables PRBS7 generator for channel 0 in group 4. If enabled, the data from the input pin are discarded.

[0x124] EPRXPRBS1

Control registers for build-in PRBS7 generators in ePortRx.

- **Bit 7 - EPRX33PrbsEnable** - Enables PRBS7 generator for channel 3 in group 3. If enabled, the data from the input pin are discarded.
- **Bit 6 - EPRX32PrbsEnable** - Enables PRBS7 generator for channel 2 in group 3. If enabled, the data from the input pin are discarded.
- **Bit 5 - EPRX31PrbsEnable** - Enables PRBS7 generator for channel 1 in group 3. If enabled, the data from the input pin are discarded.
- **Bit 4 - EPRX30PrbsEnable** - Enables PRBS7 generator for channel 0 in group 3. If enabled, the data from the input pin are discarded.
- **Bit 3 - EPRX23PrbsEnable** - Enables PRBS7 generator for channel 3 in group 2. If enabled, the data from the input pin are discarded.
- **Bit 2 - EPRX22PrbsEnable** - Enables PRBS7 generator for channel 2 in group 2. If enabled, the data from the input pin are discarded.
- **Bit 1 - EPRX21PrbsEnable** - Enables PRBS7 generator for channel 1 in group 2. If enabled, the data from the input pin are discarded.
- **Bit 0 - EPRX20PrbsEnable** - Enables PRBS7 generator for channel 0 in group 2. If enabled, the data from the input pin are discarded.

[0x125] EPRXPRBS0

Control registers for build-in PRBS7 generators in ePortRx.

- **Bit 7 - EPRX13PrbsEnable** - Enables PRBS7 generator for channel 3 in group 1. If enabled, the data from the input pin are discarded.
- **Bit 6 - EPRX12PrbsEnable** - Enables PRBS7 generator for channel 2 in group 1. If enabled, the data from the input pin are discarded.
- **Bit 5 - EPRX11PrbsEnable** - Enables PRBS7 generator for channel 1 in group 1. If enabled, the data from the input pin are discarded.
- **Bit 4 - EPRX10PrbsEnable** - Enables PRBS7 generator for channel 0 in group 1. If enabled, the data from the input pin are discarded.
- **Bit 3 - EPRX03PrbsEnable** - Enables PRBS7 generator for channel 3 in group 0. If enabled, the data from the input pin are discarded.
- **Bit 2 - EPRX02PrbsEnable** - Enables PRBS7 generator for channel 2 in group 0. If enabled, the data from the input pin are discarded.

- **Bit 1 - EPRX01PrbsEnable** - Enables PRBS7 generator for channel 1 in group 0. If enabled, the data from the input pin are discarded.
- **Bit 0 - EPRX00PrbsEnable** - Enables PRBS7 generator for channel 0 in group 0. If enabled, the data from the input pin are discarded.

[0x126] BERTSource

Data source for the built-in BER checker.

- **Bit 7:0 - BERTSource[7:0]** - Please refer to [Section 14.2](#) for more details.

[0x127] BERTConfig

Configuration for the Bit Error Rate Test.

- **Bit 7:4 - BERTMeasTime[3:0]** - Test time. For more details please refer to [Table 14.5](#)
- **Bit 1 - SKIPDisable** - Disable the skip cycle signal originating from the frame aligner. It is used when testing raw PRBS sequences on the downlink.
- **Bit 0 - BERTStart** - Asserting this bit start the BERT measurement.

[0x128] BERTDataPattern3

Fixed data pattern used by the BERT checker.

- **Bit 7:0 - BERTDataPattern[31:24]** - Bits 31:24 of the fixed pattern for BERT.

[0x129] BERTDataPattern2

Fixed data pattern used by the BERT checker.

- **Bit 7:0 - BERTDataPattern[23:16]** - Bits 23:16 of the fixed pattern for BERT.

[0x12a] BERTDataPattern1

Fixed data pattern used by the BERT checker.

- **Bit 7:0 - BERTDataPattern[15:8]** - Bits 15:8 of the fixed pattern for BERT.

[0x12b] BERTDataPattern0

Fixed data pattern used by the BERT checker.

- **Bit 7:0 - BERTDataPattern[7:0]** - Bits 7:0 of the fixed pattern for BERT.

15.2.8 Reset Manager

[0x12c] RST0

Reset related register. Enables resetting several components.

- **Bit 7 - RSTpllDigital** - Resets the PLL control logic.

- **Bit 6 - RSTfuses** - Resets the e-fuses control logic.
- **Bit 5 - RSTconfig** - Resets the configuration block.
- **Bit 4 - RSTrxLogic** - Resets the RXphy of serial configuration interface.
- **Bit 3 - RSTtxLogic** - Resets the TXphy of serial configuration interface.
- **Bit 2 - RSTi2cm0** - Resets channel 0 I2C master. One should generate a pulse on this bit (0->1->0).
- **Bit 1 - RSTi2cm1** - Resets channel 1 I2C master. One should generate a pulse on this bit (0->1->0).
- **Bit 0 - RSTi2cm2** - Resets channel 2 I2C master. One should generate a pulse on this bit (0->1->0).

[0x12d] RST1

Reset related register. Enables resetting several components.

- **Bit 7 - RSTframeAligner** - Resets the frame aligner.
- **Bit 6 - RSTprx6DII** - Resets the master DLL in ePortRx group 6.
- **Bit 5 - RSTprx5DII** - Resets the master DLL in ePortRx group 5.
- **Bit 4 - RSTprx4DII** - Resets the master DLL in ePortRx group 4.
- **Bit 3 - RSTprx3DII** - Resets the master DLL in ePortRx group 3.
- **Bit 2 - RSTprx2DII** - Resets the master DLL in ePortRx group 2.
- **Bit 1 - RSTprx1DII** - Resets the master DLL in ePortRx group 1.
- **Bit 0 - RSTprx0DII** - Resets the master DLL in ePortRx group 0.

[0x12e] RST2

Reset related register. Enables resetting several components.

- **Bit 7 - SKIPforce** -
- **Bit 6 - ResetOutForceActive** - As long as this bit is set low, the `resetOut` signal is active (low level).
- **Bit 3 - RSTps3DII** - Resets DLL in 3 phase aligner channel.
- **Bit 2 - RSTps2DII** - Resets DLL in 2 phase aligner channel.
- **Bit 1 - RSTps1DII** - Resets DLL in 1 phase aligner channel.
- **Bit 0 - RSTps0DII** - Resets DLL in 0 phase aligner channel.

15.2.9 Power Up

[0x12f] POWERUP3

Controls behavior of the power up state machine (for more details refer to [Power-up state machine](#) (page 61))

- **Bit 7 - PUSMForceState** - Allows to override the state of power up state machine. To enable this feature, register `PUSMForceMagic` has to be set to 0xA3 (magic number)
- **Bit 4:0 - PUSMStateForced[4:0]** - Selects state of the power up state machine when `STATEOVRD` pin is asserted or `PUSMForceState` bit is asserted. For more details refer to [Power-up state machine](#) (page 61) and [STATEOVRD](#) (page 67))

[0x130] POWERUP4

Controls behavior of the power up state machine (for more details refer to [Power-up state machine](#) (page 61))

- **Bit 7:0 - PUSMForceMagic[7:0]** - Has to be set to 0xA3 (magic number) in order to enable PUSMForceState feature.

15.2.10 Debug

[0x131] CLKTree

Clock tree disable feature. Could be used for TMR testing.

- **Bit 7:3 - ClkTreeMagicNumber[4:0]** - Has to be set to 5'h15 in order for clock masking (disabling) to be active
- **Bit 2 - clkTreeCDDisable** - If asserted and ClkTreeMagicNumber set to 5'h15, branch C of clock tree is disabled
- **Bit 1 - clkTreeBDDisable** - If asserted and ClkTreeMagicNumber set to 5'h15, branch B of clock tree is disabled
- **Bit 0 - clkTreeADDisable** - If asserted and ClkTreeMagicNumber set to 5'h15, branch A of clock tree is disabled

[0x132] DataPath

Data path configuration.

- **Bit 7 - DLDPBypassDeInterleaver** - Bypass de-interleaver in the downlink data path.
- **Bit 6 - DLDPBypassFECDecoder** - Bypass FEC decoder in the downlink data path.
- **Bit 5 - DLDPBypassDeScrambler** - Bypass de-scrambler in the downlink data path.
- **Bit 4 - DLDPFECErrCntEna** - Enable FEC error counter for the downlink data path.
- **Bit 2 - ULDPBypassInterleaver** - Bypass interleaver in the uplink data path.
- **Bit 1 - ULDPBypassScrambler** - Bypass scrambler in the uplink data path.
- **Bit 0 - ULDPBypassFECCoder** - Bypass FEC coder in the uplink data path.

[0x133] TO0Sel

Control register for test output 0

- **Bit 7:0 - TO0Select[7:0]** - Selects a signal to be outputted on TSTOUT0 according to [TOnSelect](#) (page 121)

[0x134] TO1Sel

Control register for test output 1

- **Bit 7:0 - TO1Select[7:0]** - Selects a signal to be outputted on TSTOUT1 according to [TOnSelect](#) (page 121)

[0x135] TO2Sel

Control register for test output 2

- **Bit 7:0 - TO2Select[7:0]** - Selects a signal to be outputted on TSTOUT2 according to [TOnSelect](#) (page 121)

[0x136] TO3Sel

Control register for test output 3

- **Bit 7:0 - TO3Select[7:0]** - Selects a signal to be outputted on TSTOUT3 according to [TOSelect](#) (page 121)

[0x137] TO4Sel

Control register for test output 4

- **Bit 7:0 - TO4Select[7:0]** - Selects a signal to be outputted on TSTOUT4 according to [TOSelect](#) (page 121)

[0x138] TO5Sel

Control register for test output 5

- **Bit 7:0 - TO5Select[7:0]** - Selects a signal to be outputted on TSTOUT5 according to [TOSelect](#) (page 121)

[0x139] TODrivingStrength

Driving strength control for CMOS test outputs

- **Bit 3 - TO3DS** - Drive strength for TSTOUT3 ([CMOS I/O Pin Characteristics](#) (page 310))
- **Bit 2 - TO2DS** - Drive strength for TSTOUT2 ([CMOS I/O Pin Characteristics](#) (page 310))
- **Bit 1 - TO1DS** - Drive strength for TSTOUT1 ([CMOS I/O Pin Characteristics](#) (page 310))
- **Bit 0 - TO0DS** - Drive strength for TSTOUT0 ([CMOS I/O Pin Characteristics](#) (page 310))

[0x13a] TO4Driver

Output driver control for test output 4

- **Bit 7:5 - TO4PreEmphasisStrength[2:0]** - Sets the pre-emphasis strength for TO4.

TO4PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- **Bit 4:3 - TO4PreEmphasisMode[1:0]** - Selects the pre-emphasis mode for TO4.

TO4PreEmphasisMode[1:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- Bit 2:0 - TO4DriveStrength[2:0] - Sets the pre-emphasis strength for TO4.

TO4DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

[0x13b] TO5Driver

Output driver control for test output 5

- Bit 7:5 - TO5PreEmphasisStrength[2:0] - Sets the pre-emphasis strength for TO5.

TO5PreEmphasisStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

- Bit 4:3 - TO5PreEmphasisMode[1:0] - Selects the pre-emphasis mode for TO5.

TO5PreEmphasisMode[1:0]	Mode
3'd0	disabled
3'd1	disabled
3'd2	Self timed
3'd3	Clock timed

- Bit 2:0 - TO5DriveStrength[2:0] - Sets the pre-emphasis strength for TO5.

TO5DriveStrength[2:0]	Strength [mA]
3'd0	0
3'd1	1.0
3'd2	1.5
3'd3	2.0
3'd4	2.5
3'd5	3.0
3'd6	3.5
3'd7	4.0

[0x13c] TOPreEmp

Pre-emphasis control for differential test outputs

- **Bit 7 - TO5Invert** -
- **Bit 6:4 - TO5PreEmphasisWidth[2:0]** - Sets the pre-emphasis strength for TO5.

TO5PreEmphasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

- **Bit 3 - TO4Invert** -
- **Bit 2:0 - TO4PreEmphasisWidth[2:0]** - Sets the pre-emphasis strength for TO4.

TO4PreEmphasisWidth[2:0]	Pulse length [ps]
3'd0	120
3'd1	240
3'd2	360
3'd3	480
3'd4	600
3'd5	720
3'd6	840
3'd7	960

15.3 Read Only

15.3.1 LPGBTSettings

[0x140] ConfigPins

Status of the IpGBT external configuration pins

- **Bit 7:4 - LPGBTMode[3:0]** - State of MODE pins. The function of the pin is described in *MODE3*, *MODE2*, *MODE1*, *MODE0* (page 17).
- **Bit 3 - ConfigSelect** - State of SC_I2C pin. The function of the pin is described in *SC_I2C* (page 18).
- **Bit 2 - VCOBypass** - State of the VCOBYPASS pin. The function of the pin is described in *VCOBYPASS* (page 76).
- **Bit 1 - lockMode** - State of the LOCKMODE pin. The function of the pin is described in *LOCKMODE* (page 75).
- **Bit 0 - stateOverride** - State of the STATEOVRD. The function of the pin is described in *STATEOVRD* (page 67).

[0x141] I2CSlaveAddress

Chip address.

- **Bit 6:0 - AsicControlAdr[6:0]** - Address of the chip used in slow control protocols (I2C, IC, EC).

15.3.2 ePortRx

[0x142] EPRX0Locked

ePortRx group 0 status register

- **Bit 7:4 - EPRX0ChnLocked[3:0]** - Status of phase selection logic for channels in group 0. Bit 0 corresponds to channel 0, bit 1 to channel 1 and so on. Logic value of 1 means that the channel is locked.
- **Bit 1:0 - EPRX0State[1:0]** - State of initialization state machine for ePortRx group 0. State can be according to the table:

EPRX0State[1:0]	State
2'd0	Reset
2'd1	Force down
2'd2	Confirm early state
2'd3	Free running state

[0x143] EPRX0CurrentPhase10

Currently selected phases for channels 0 and 1 in ePortRx group 0

- **Bit 7:4 - EPRX0CurrentPhase1[3:0]** - Currently selected phases for channels 1 in ePortRx group 0
- **Bit 3:0 - EPRX0CurrentPhase0[3:0]** - Currently selected phases for channels 0 in ePortRx group 0

[0x144] EPRX0CurrentPhase32

Currently selected phases for channels 2 and 3 in ePortRx group 0

- **Bit 7:4 - EPRX0CurrentPhase3[3:0]** - Currently selected phases for channels 3 in ePortRx group 0
- **Bit 3:0 - EPRX0CurrentPhase2[3:0]** - Currently selected phases for channels 2 in ePortRx group 0

[0x145] EPRX1Locked

ePortRx group 1 status register

- **Bit 7:4 - EPRX1ChnLocked[3:0]** - Status of phase selection logic for channels in group 1. Bit 0 corresponds to channel 0, bit 1 to channel 1 and so on. Logic value of 1 means that the channel is locked.
- **Bit 1:0 - EPRX1State[1:0]** - State of initialization state machine for ePortRx group 1. State can be according to the table:

EPRX1State[1:0]	State
2'd0	Reset
2'd1	Force down
2'd2	Confirm early state
2'd3	Free running state

[0x146] EPRX1CurrentPhase10

Currently selected phases for channels 0 and 1 in ePortRx group 1

- **Bit 7:4 - EPRX1CurrentPhase1[3:0]** - Currently selected phases for channels 1 in ePortRx group 1
- **Bit 3:0 - EPRX1CurrentPhase0[3:0]** - Currently selected phases for channels 0 in ePortRx group 1

[0x147] EPRX1CurrentPhase32

Currently selected phases for channels 2 and 3 in ePortRx group 1

- **Bit 7:4 - EPRX1CurrentPhase3[3:0]** - Currently selected phases for channels 3 in ePortRx group 1
- **Bit 3:0 - EPRX1CurrentPhase2[3:0]** - Currently selected phases for channels 2 in ePortRx group 1

[0x148] EPRX2Locked

ePortRx group 2 status register

- **Bit 7:4 - EPRX2ChnLocked[3:0]** - Status of phase selection logic for channels in group 2. Bit 0 corresponds to channel 0, bit 1 to channel 1 and so on. Logic value of 1 means that the channel is locked.
- **Bit 1:0 - EPRX2State[1:0]** - State of initialization state machine for ePortRx group 2. State can be according to the table:

EPRX2State[1:0]	State
2'd0	Reset
2'd1	Force down
2'd2	Confirm early state
2'd3	Free running state

[0x149] EPRX2CurrentPhase10

Currently selected phases for channels 0 and 1 in ePortRx group 2

- **Bit 7:4 - EPRX2CurrentPhase1[3:0]** - Currently selected phases for channels 1 in ePortRx group 2
- **Bit 3:0 - EPRX2CurrentPhase0[3:0]** - Currently selected phases for channels 0 in ePortRx group 2

[0x14a] EPRX2CurrentPhase32

Currently selected phases for channels 2 and 3 in ePortRx group 2

- **Bit 7:4 - EPRX2CurrentPhase3[3:0]** - Currently selected phases for channels 3 in ePortRx group 2
- **Bit 3:0 - EPRX2CurrentPhase2[3:0]** - Currently selected phases for channels 2 in ePortRx group 2

[0x14b] EPRX3Locked

ePortRx group 3 status register

- **Bit 7:4 - EPRX3ChnLocked[3:0]** - Status of phase selection logic for channels in group 3. Bit 0 corresponds to channel 0, bit 1 to channel 1 and so on. Logic value of 1 means that the channel is locked.

- **Bit 1:0 - EPRX3State[1:0]** - State of initialization state machine for ePortRx group 3. State can be according to the table:

EPRX3State[1:0]	State
2'd0	Reset
2'd1	Force down
2'd2	Confirm early state
2'd3	Free running state

[0x14c] EPRX3CurrentPhase10

Currently selected phases for channels 0 and 1 in ePortRx group 3

- **Bit 7:4 - EPRX3CurrentPhase1[3:0]** - Currently selected phases for channels 1 in ePortRx group 3
- **Bit 3:0 - EPRX3CurrentPhase0[3:0]** - Currently selected phases for channels 0 in ePortRx group 3

[0x14d] EPRX3CurrentPhase32

Currently selected phases for channels 2 and 3 in ePortRx group 3

- **Bit 7:4 - EPRX3CurrentPhase3[3:0]** - Currently selected phases for channels 3 in ePortRx group 3
- **Bit 3:0 - EPRX3CurrentPhase2[3:0]** - Currently selected phases for channels 2 in ePortRx group 3

[0x14e] EPRX4Locked

ePortRx group 4 status register

- **Bit 7:4 - EPRX4ChnLocked[3:0]** - Status of phase selection logic for channels in group 4. Bit 0 corresponds to channel 0, bit 1 to channel 1 and so on. Logic value of 1 means that the channel is locked.
- **Bit 1:0 - EPRX4State[1:0]** - State of initialization state machine for ePortRx group 4. State can be according to the table:

EPRX4State[1:0]	State
2'd0	Reset
2'd1	Force down
2'd2	Confirm early state
2'd3	Free running state

[0x14f] EPRX4CurrentPhase10

Currently selected phases for channels 0 and 1 in ePortRx group 4

- **Bit 7:4 - EPRX4CurrentPhase1[3:0]** - Currently selected phases for channels 1 in ePortRx group 4
- **Bit 3:0 - EPRX4CurrentPhase0[3:0]** - Currently selected phases for channels 0 in ePortRx group 4

[0x150] EPRX4CurrentPhase32

Currently selected phases for channels 2 and 3 in ePortRx group 4

- **Bit 7:4 - EPRX4CurrentPhase3[3:0]** - Currently selected phases for channels 3 in ePortRx group 4
- **Bit 3:0 - EPRX4CurrentPhase2[3:0]** - Currently selected phases for channels 2 in ePortRx group 4

[0x151] EPRX5Locked

ePortRx group 5 status register

- **Bit 7:4 - EPRX5ChnLocked[3:0]** - Status of phase selection logic for channels in group 5. Bit 0 corresponds to channel 0, bit 1 to channel 1 and so on. Logic value of 1 means that the channel is locked.
- **Bit 1:0 - EPRX5State[1:0]** - State of initialization state machine for ePortRx group 5. State can be according to the table:

EPRX5State[1:0]	State
2'd0	Reset
2'd1	Force down
2'd2	Confirm early state
2'd3	Free running state

[0x152] EPRX5CurrentPhase10

Currently selected phases for channels 0 and 1 in ePortRx group 5

- **Bit 7:4 - EPRX5CurrentPhase1[3:0]** - Currently selected phases for channels 1 in ePortRx group 5
- **Bit 3:0 - EPRX5CurrentPhase0[3:0]** - Currently selected phases for channels 0 in ePortRx group 5

[0x153] EPRX5CurrentPhase32

Currently selected phases for channels 2 and 3 in ePortRx group 5

- **Bit 7:4 - EPRX5CurrentPhase3[3:0]** - Currently selected phases for channels 3 in ePortRx group 5
- **Bit 3:0 - EPRX5CurrentPhase2[3:0]** - Currently selected phases for channels 2 in ePortRx group 5

[0x154] EPRX6Locked

ePortRx group 6 status register

- **Bit 7:4 - EPRX6ChnLocked[3:0]** - Status of phase selection logic for channels in group 6. Bit 0 corresponds to channel 0, bit 1 to channel 1 and so on. Logic value of 1 means that the channel is locked.
- **Bit 1:0 - EPRX6State[1:0]** - State of initialization state machine for ePortRx group 6. State can be according to the table:

EPRX6State[1:0]	State
2'd0	Reset
2'd1	Force down
2'd2	Confirm early state
2'd3	Free running state

[0x155] EPRX6CurrentPhase10

Currently selected phases for channels 0 and 1 in ePortRx group 6

- **Bit 7:4 - EPRX6CurrentPhase1[3:0]** - Currently selected phases for channels 1 in ePortRx group 6
- **Bit 3:0 - EPRX6CurrentPhase0[3:0]** - Currently selected phases for channels 0 in ePortRx group 6

[0x156] EPRX6CurrentPhase32

Currently selected phases for channels 2 and 3 in ePortRx group 6

- **Bit 7:4 - EPRX6CurrentPhase3[3:0]** - Currently selected phases for channels 3 in ePortRx group 6
- **Bit 3:0 - EPRX6CurrentPhase2[3:0]** - Currently selected phases for channels 2 in ePortRx group 6

[0x157] EPRXEcCurrentPhase

Status register for ePortRxEc

- **Bit 3:0 - EPRXEcCurrentPhase[3:0]** - Currently selected phase for EC channel phase aligner.

[0x158] EPRX0DIIStatus

Status register of lock filter for ePortRxGroup0

- **Bit 7 - EPRX0DIILocked** - Lock status of the master DLL
- **Bit 6:5 - EPRX0DIILFState[1:0]** - State of lock filter state machine
- **Bit 4:0 - EPRX0DIILOLCn[4:0]** - Loss of Lock counter value

[0x159] EPRX1DIIStatus

Status register of lock filter for ePortRxGroup1

- **Bit 7 - EPRX1DIILocked** - Lock status of the master DLL
- **Bit 6:5 - EPRX1DIILFState[1:0]** - State of lock filter state machine
- **Bit 4:0 - EPRX1DIILOLCn[4:0]** - Loss of Lock counter value

[0x15a] EPRX2DIIStatus

Status register of lock filter for ePortRxGroup2

- **Bit 7 - EPRX2DIILocked** - Lock status of the master DLL
- **Bit 6:5 - EPRX2DIILFState[1:0]** - State of lock filter state machine
- **Bit 4:0 - EPRX2DIILOLCn[4:0]** - Loss of Lock counter value

[0x15b] EPRX3DIIStatus

Status register of lock filter for ePortRxGroup3

- **Bit 7 - EPRX3DIILocked** - Lock status of the master DLL
- **Bit 6:5 - EPRX3DII LFState[1:0]** - State of lock filter state machine
- **Bit 4:0 - EPRX3DII LOLCnt[4:0]** - Loss of Lock counter value

[0x15c] EPRX4DIIStatus

Status register of lock filter for ePortRxGroup4

- **Bit 7 - EPRX4DIILocked** - Lock status of the master DLL
- **Bit 6:5 - EPRX4DII LFState[1:0]** - State of lock filter state machine
- **Bit 4:0 - EPRX4DII LOLCnt[4:0]** - Loss of Lock counter value

[0x15d] EPRX5DIIStatus

Status register of lock filter for ePortRxGroup5

- **Bit 7 - EPRX5DIILocked** - Lock status of the master DLL
- **Bit 6:5 - EPRX5DII LFState[1:0]** - State of lock filter state machine
- **Bit 4:0 - EPRX5DII LOLCnt[4:0]** - Loss of Lock counter value

[0x15e] EPRX6DIIStatus

Status register of lock filter for ePortRxGroup6

- **Bit 7 - EPRX6DIILocked** - Lock status of the master DLL
- **Bit 6:5 - EPRX6DII LFState[1:0]** - State of lock filter state machine
- **Bit 4:0 - EPRX6DII LOLCnt[4:0]** - Loss of Lock counter value

15.3.3 I2C Masters

[0x15f] I2CM0Ctrl

Contents of control register written by user for I2C Master 0

- **Bit 7:0 - I2CM0Ctrl[7:0]** - Contents of control register written by user

[0x160] I2CM0Mask

Contents of mask register written by user for I2C Master 0

- **Bit 7:0 - I2CM0Mask[7:0]** - Content of the status register.

[0x161] I2CM0Status

Contents of status register for I2C Master 0

- **Bit 7:0 - I2CM0Status[7:0]** - Content of the status register.

[0x162] I2CM0TranCnt

Contents of transaction counter for I2C Master 0

- **Bit 7:0 - I2CM0TranCnt[7:0]** - Content of transaction counter.

[0x163] I2CM0ReadByte

Data read from an I2C slave in a single-byte-read for I2C Master 0

- **Bit 7:0 - I2CM0ReadByte[7:0]** - Data read from an I2C slave in a single-byte-read

[0x164] I2CM0Read0

Data read from an I2C slave in a multi-byte-read by I2C Master 0

- **Bit 7:0 - I2CM0Read[7:0]** - Data read from an I2C slave in a multi-byte-read

[0x165] I2CM0Read1

Data read from an I2C slave in a multi-byte-read by I2C Master 0

- **Bit 7:0 - I2CM0Read[15:8]** - Data read from an I2C slave in a multi-byte-read

[0x166] I2CM0Read2

Data read from an I2C slave in a multi-byte-read by I2C Master 0

- **Bit 7:0 - I2CM0Read[23:16]** - Data read from an I2C slave in a multi-byte-read

[0x167] I2CM0Read3

Data read from an I2C slave in a multi-byte-read by I2C Master 0

- **Bit 7:0 - I2CM0Read[31:24]** - Data read from an I2C slave in a multi-byte-read

[0x168] I2CM0Read4

Data read from an I2C slave in a multi-byte-read by I2C Master 0

- **Bit 7:0 - I2CM0Read[39:32]** - Data read from an I2C slave in a multi-byte-read

[0x169] I2CM0Read5

Data read from an I2C slave in a multi-byte-read by I2C Master 0

- **Bit 7:0 - I2CM0Read[47:40]** - Data read from an I2C slave in a multi-byte-read

[0x16a] I2CM0Read6

Data read from an I2C slave in a multi-byte-read by I2C Master 0

- **Bit 7:0 - I2CM0Read[55:48]** - Data read from an I2C slave in a multi-byte-read

[0x16b] I2CM0Read7

Data read from an I2C slave in a multi-byte-read by I2C Master 0

- **Bit 7:0 - I2CM0Read[63:56]** - Data read from an I2C slave in a multi-byte-read

[0x16c] I2CM0Read8

Data read from an I2C slave in a multi-byte-read by I2C Master 0

- **Bit 7:0 - I2CM0Read[71:64]** - Data read from an I2C slave in a multi-byte-read

[0x16d] I2CM0Read9

Data read from an I2C slave in a multi-byte-read by I2C Master 0

- **Bit 7:0 - I2CM0Read[79:72]** - Data read from an I2C slave in a multi-byte-read

[0x16e] I2CM0Read10

Data read from an I2C slave in a multi-byte-read by I2C Master 0

- **Bit 7:0 - I2CM0Read[87:80]** - Data read from an I2C slave in a multi-byte-read

[0x16f] I2CM0Read11

Data read from an I2C slave in a multi-byte-read by I2C Master 0

- **Bit 7:0 - I2CM0Read[95:88]** - Data read from an I2C slave in a multi-byte-read

[0x170] I2CM0Read12

Data read from an I2C slave in a multi-byte-read by I2C Master 0

- **Bit 7:0 - I2CM0Read[103:96]** - Data read from an I2C slave in a multi-byte-read

[0x171] I2CM0Read13

Data read from an I2C slave in a multi-byte-read by I2C Master 0

- **Bit 7:0 - I2CM0Read[111:104]** - Data read from an I2C slave in a multi-byte-read

[0x172] I2CM0Read14

Data read from an I2C slave in a multi-byte-read by I2C Master 0

- **Bit 7:0 - I2CM0Read[119:112]** - Data read from an I2C slave in a multi-byte-read

[0x173] I2CM0Read15

Data read from an I2C slave in a multi-byte-read by I2C Master 0

- **Bit 7:0 - I2CM0Read[127:120]** - Data read from an I2C slave in a multi-byte-read

[0x174] I2CM1Ctrl

Contents of control register written by user for I2C Master 1

- **Bit 7:0 - I2CM1Ctrl[7:0]** - Contents of control register written by user

[0x175] I2CM1Mask

Contents of mask register written by user for I2C Master 1

- **Bit 7:0 - I2CM1Mask[7:0]** - Content of the status register.

[0x176] I2CM1Status

Contents of status register for I2C Master 1

- **Bit 7:0 - I2CM1Status[7:0]** - Content of the status register.

[0x177] I2CM1TranCnt

Contents of transaction counter for I2C Master 1

- **Bit 7:0 - I2CM1TranCnt[7:0]** - Content of transaction counter.

[0x178] I2CM1ReadByte

Data read from an I2C slave in a single-byte-read for I2C Master 1

- **Bit 7:0 - I2CM1ReadByte[7:0]** - Data read from an I2C slave in a single-byte-read

[0x179] I2CM1Read0

Data read from an I2C slave in a multi-byte-read by I2C Master 1

- **Bit 7:0 - I2CM1Read[7:0]** - Data read from an I2C slave in a multi-byte-read

[0x17a] I2CM1Read1

Data read from an I2C slave in a multi-byte-read by I2C Master 1

- **Bit 7:0 - I2CM1Read[15:8]** - Data read from an I2C slave in a multi-byte-read

[0x17b] I2CM1Read2

Data read from an I2C slave in a multi-byte-read by I2C Master 1

- **Bit 7:0 - I2CM1Read[23:16]** - Data read from an I2C slave in a multi-byte-read

[0x17c] I2CM1Read3

Data read from an I2C slave in a multi-byte-read by I2C Master 1

- **Bit 7:0 - I2CM1Read[31:24]** - Data read from an I2C slave in a multi-byte-read

[0x17d] I2CM1Read4

Data read from an I2C slave in a multi-byte-read by I2C Master 1

- **Bit 7:0 - I2CM1Read[39:32]** - Data read from an I2C slave in a multi-byte-read

[0x17e] I2CM1Read5

Data read from an I2C slave in a multi-byte-read by I2C Master 1

- **Bit 7:0 - I2CM1Read[47:40]** - Data read from an I2C slave in a multi-byte-read

[0x17f] I2CM1Read6

Data read from an I2C slave in a multi-byte-read by I2C Master 1

- **Bit 7:0 - I2CM1Read[55:48]** - Data read from an I2C slave in a multi-byte-read

[0x180] I2CM1Read7

Data read from an I2C slave in a multi-byte-read by I2C Master 1

- **Bit 7:0 - I2CM1Read[63:56]** - Data read from an I2C slave in a multi-byte-read

[0x181] I2CM1Read8

Data read from an I2C slave in a multi-byte-read by I2C Master 1

- **Bit 7:0 - I2CM1Read[71:64]** - Data read from an I2C slave in a multi-byte-read

[0x182] I2CM1Read9

Data read from an I2C slave in a multi-byte-read by I2C Master 1

- **Bit 7:0 - I2CM1Read[79:72]** - Data read from an I2C slave in a multi-byte-read

[0x183] I2CM1Read10

Data read from an I2C slave in a multi-byte-read by I2C Master 1

- **Bit 7:0 - I2CM1Read[87:80]** - Data read from an I2C slave in a multi-byte-read

[0x184] I2CM1Read11

Data read from an I2C slave in a multi-byte-read by I2C Master 1

- **Bit 7:0 - I2CM1Read[95:88]** - Data read from an I2C slave in a multi-byte-read

[0x185] I2CM1Read12

Data read from an I2C slave in a multi-byte-read by I2C Master 1

- **Bit 7:0 - I2CM1Read[103:96]** - Data read from an I2C slave in a multi-byte-read

[0x186] I2CM1Read13

Data read from an I2C slave in a multi-byte-read by I2C Master 1

- **Bit 7:0 - I2CM1Read[111:104]** - Data read from an I2C slave in a multi-byte-read

[0x187] I2CM1Read14

Data read from an I2C slave in a multi-byte-read by I2C Master 1

- **Bit 7:0 - I2CM1Read[119:112]** - Data read from an I2C slave in a multi-byte-read

[0x188] I2CM1Read15

Data read from an I2C slave in a multi-byte-read by I2C Master 1

- **Bit 7:0 - I2CM1Read[127:120]** - Data read from an I2C slave in a multi-byte-read

[0x189] I2CM2Ctrl

Contents of control register written by user for I2C Master 2

- **Bit 7:0 - I2CM2Ctrl[7:0]** - Contents of control register written by user

[0x18a] I2CM2Mask

Contents of mask register written by user for I2C Master 2

- **Bit 7:0 - I2CM2Mask[7:0]** - Content of the status register.

[0x18b] I2CM2Status

Contents of status register for I2C Master 2

- **Bit 7:0 - I2CM2Status[7:0]** - Content of the status register.

[0x18c] I2CM2TranCnt

Contents of transaction counter for I2C Master 2

- **Bit 7:0 - I2CM2TranCnt[7:0]** - Content of transaction counter.

[0x18d] I2CM2ReadByte

Data read from an I2C slave in a single-byte-read for I2C Master 2

- **Bit 7:0 - I2CM2ReadByte[7:0]** - Data read from an I2C slave in a single-byte-read

[0x18e] I2CM2Read0

Data read from an I2C slave in a multi-byte-read by I2C Master 2

- **Bit 7:0 - I2CM2Read[7:0]** - Data read from an I2C slave in a multi-byte-read

[0x18f] I2CM2Read1

Data read from an I2C slave in a multi-byte-read by I2C Master 2

- **Bit 7:0 - I2CM2Read[15:8]** - Data read from an I2C slave in a multi-byte-read

[0x190] I2CM2Read2

Data read from an I2C slave in a multi-byte-read by I2C Master 2

- **Bit 7:0 - I2CM2Read[23:16]** - Data read from an I2C slave in a multi-byte-read

[0x191] I2CM2Read3

Data read from an I2C slave in a multi-byte-read by I2C Master 2

- **Bit 7:0 - I2CM2Read[31:24]** - Data read from an I2C slave in a multi-byte-read

[0x192] I2CM2Read4

Data read from an I2C slave in a multi-byte-read by I2C Master 2

- **Bit 7:0 - I2CM2Read[39:32]** - Data read from an I2C slave in a multi-byte-read

[0x193] I2CM2Read5

Data read from an I2C slave in a multi-byte-read by I2C Master 2

- **Bit 7:0 - I2CM2Read[47:40]** - Data read from an I2C slave in a multi-byte-read

[0x194] I2CM2Read6

Data read from an I2C slave in a multi-byte-read by I2C Master 2

- **Bit 7:0 - I2CM2Read[55:48]** - Data read from an I2C slave in a multi-byte-read

[0x195] I2CM2Read7

Data read from an I2C slave in a multi-byte-read by I2C Master 2

- **Bit 7:0 - I2CM2Read[63:56]** - Data read from an I2C slave in a multi-byte-read

[0x196] I2CM2Read8

Data read from an I2C slave in a multi-byte-read by I2C Master 2

- **Bit 7:0 - I2CM2Read[71:64]** - Data read from an I2C slave in a multi-byte-read

[0x197] I2CM2Read9

Data read from an I2C slave in a multi-byte-read by I2C Master 2

- **Bit 7:0 - I2CM2Read[79:72]** - Data read from an I2C slave in a multi-byte-read

[0x198] I2CM2Read10

Data read from an I2C slave in a multi-byte-read by I2C Master 2

- **Bit 7:0 - I2CM2Read[87:80]** - Data read from an I2C slave in a multi-byte-read

[0x199] I2CM2Read11

Data read from an I2C slave in a multi-byte-read by I2C Master 2

- **Bit 7:0 - I2CM2Read[95:88]** - Data read from an I2C slave in a multi-byte-read

[0x19a] I2CM2Read12

Data read from an I2C slave in a multi-byte-read by I2C Master 2

- **Bit 7:0 - I2CM2Read[103:96]** - Data read from an I2C slave in a multi-byte-read

[0x19b] I2CM2Read13

Data read from an I2C slave in a multi-byte-read by I2C Master 2

- **Bit 7:0 - I2CM2Read[111:104]** - Data read from an I2C slave in a multi-byte-read

[0x19c] I2CM2Read14

Data read from an I2C slave in a multi-byte-read by I2C Master 2

- **Bit 7:0 - I2CM2Read[119:112]** - Data read from an I2C slave in a multi-byte-read

[0x19d] I2CM2Read15

Data read from an I2C slave in a multi-byte-read by I2C Master 2

- **Bit 7:0 - I2CM2Read[127:120]** - Data read from an I2C slave in a multi-byte-read

15.3.4 ECLK

[0x19e] PSStatus

Status of phase-shifter DLL initialization state machines

- **Bit 7:6 - PS3DllInitState[1:0]** - Status of the DLL initialization state machine for phase-shifter channel 3
- **Bit 5:4 - PS2DllInitState[1:0]** - Status of the DLL initialization state machine for phase-shifter channel 2
- **Bit 3:2 - PS1DllInitState[1:0]** - Status of the DLL initialization state machine for phase-shifter channel 1

- **Bit 1:0 - PS0DllInitState[1:0]** - Status of the DLL initialization state machine for phase-shifter channel 0

[0x19f] PIOInH

Allows read back of the PIO state

- **Bit 7:0 - PIOIn[15:8]** -

PIOIn[n]	Input signal level
1'b0	Low
1'b1	High

[0x1a0] PIOInL

Allows read back of the PIO state

- **Bit 7:0 - PIOIn[7:0]** -

PIOIn[n]	Input signal level
1'b0	Low
1'b1	High

[0x1a1] FUSEStatus

Status of fuse block.

- **Bit 3 - FuseBlowError** - Error flag (attempt to blow fuses without magic number).
- **Bit 2 - FuseDataValid** - Fuse read sequence was successful, SelectedFuseValues [31:0] is valid.
- **Bit 1 - FuseBlowDone** - Fuse blowing sequence was successful.
- **Bit 0 - FuseBlowBusy** - Fuse block is busy (either read or blowing sequence).

[0x1a2] FUSEValuesA

Value of selected FUSE block. (should be accessed only if FuseDataValid bit is set in [\[0x1a1\] FUSEStatus](#) (page 260) register).

- **Bit 7:0 - SelectedFuseValues[7:0]** - Bits 7:0 of the data word.

[0x1a3] FUSEValuesB

Value of selected FUSE block. (should be accessed only if FuseDataValid bit is set in [\[0x1a1\] FUSEStatus](#) (page 260) register).

- **Bit 7:0 - SelectedFuseValues[15:8]** - Bits 15:8 of the data word.

[0x1a4] FUSEValuesC

Value of selected FUSE block. (should be accessed only if FuseDataValid bit is set in [\[0x1a1\] FUSEStatus](#) (page 260) register).

- **Bit 7:0 - SelectedFuseValues[23:16]** - Bits 23:16 of the data word.

[0x1a5] FUSEValuesD

Value of selected FUSE block. (should be accessed only if FuseDataValid bit is set in [\[0x1a1\] FUSEStatus](#) (page 260) register).

- **Bit 7:0 - SelectedFuseValues[31:24]** - Bits 31:24 of the data word.

15.3.5 Process Monitor

[0x1a6] ProcessMonitorStatus

Process Monitor block status register.

- **Bit 1 - PMDone** - Measurement done.
- **Bit 0 - PMBusy** - Measurement in progress.

[0x1a7] PMFreqA

Process Monitor frequency measurement result.

- **Bit 7:0 - PMFreq[23:16]** - Bits 23:16 of frequency measurement result.

[0x1a8] PMFreqB

Process Monitor frequency measurement result.

- **Bit 7:0 - PMFreq[15:8]** - Bits 15:8 of frequency measurement result.

[0x1a9] PMFreqC

Process Monitor frequency measurement result.

- **Bit 7:0 - PMFreq[7:0]** - Bits 7:0 of frequency measurement result.

15.3.6 SEU

[0x1aa] SEUCountH

Value of SEU counter.

- **Bit 7:0 - SEUCount[15:8]** - Bits 15:8 of SEU counter.

[0x1ab] SEUCountL

Value of SEU counter.

- **Bit 7:0 - SEUCount[7:0]** - Bits 7:0 of SEU counter.

15.3.7 Clock Generator

[0x1ac] CLKGStatus0

- **Bit 7:4 - CLKG_PLL_R_CONFIG[3:0]** - Selected PLL's filter resistance value [min:step:max] Ohm
- **Bit 3:0 - CLKG_CONFIG_I_PLL[3:0]** - Selected PLL's integral current [min:step:max] uA

[0x1ad] CLKGStatus1

- **Bit 7:4 - CLKG_CONFIG_I_FLL[3:0]** - Selected CDR's FLL current [min:step:max] uA
- **Bit 3:0 - CLKG_CONFIG_I_CDR[3:0]** - Selected CDR's integral current [min:step:max] uA

[0x1ae] CLKGStatus2

- **Bit 7:4 - CLKG_CONFIG_P_FF_CDR[3:0]** - Selected CDR's proportional feedforward current [min:step:max] uA
- **Bit 3:0 - CLKG_CONFIG_P_CDR[3:0]** - Selected CDR's phase detector proportional current [min:step:max] uA

[0x1af] CLKGStatus3

- **Bit 7:0 - CLKG_lfLossOfLockCount[7:0]** - Lock filter loss of lock (increases when lock filter's state goes lfConfirmUnlockState -> lfUnlfLockedState); only in TX mode

[0x1b0] CLKGStatus4

- **Bit 7:4 - CLKG_CONFIG_P_PLL[3:0]** - Selected PLL's proportional current [min:step:max] uA
- **Bit 3:0 - CLKG_BIASGEN_CONFIG[3:0]** - Selected bias DAC for the charge pumps [min:step:max] uA

[0x1b1] CLKGStatus5

- **Bit 7:0 - CLKG_vcoCapSelect[8:1]** - Selected vco capacitor bank (thermistor value)

[0x1b2] CLKGStatus6

- **Bit 7 - CLKG_vcoCapSelect[0]** - Selected vco capacitor bank (thermistor value)

- Bit 6:5 - **CLKG_dataMuxCfg[1:0]** - Selected data MUX loopback (test only)

CLKG_dataMuxCfg[1:0]	Description
2'd0	invalid state
2'd1	Equalizer output data loopback
2'd2	Data resampled by CDR loopback
2'd3	disabled

- Bit 3:0 - **CLKG_vcoDAC[3:0]** - Selected current DAC for the VCO [min:step:max] uA

[0x1b3] CLKGStatus

- Bit 7 - **CLKG_connectCDR** - 0: CDR loop is disconnected from VCO; 1: CDR loop is connected to VCO;
- Bit 6 - **CLKG_connectPLL** - 0: PLL loop is disconnected from VCO; 1: PLL loop is connected to VCO;
- Bit 5 - **CLKG_disDataCounterRef** - 0: data/4 ripple counter is enabled; 1: disabled
- Bit 4 - **CLKG_enableCDR** - 0: Alexander PD UP/DOWN buffers + Alexander PD are disabled; 1: enabled
- Bit 3 - **CLKG_enableFD** - 0: PLL's FD + FD up/down signals are disabled; 1: enabled
- Bit 2 - **CLKG_enablePLL** - 0: PLL's PFD up/down signals are disabled; 1: enabled
- Bit 1 - **CLKG_overrideVc** - 0: The VCO's control voltage override is disabled; 1: enabled vcoControlVoltage is mid range
- Bit 0 - **CLKG_refClkSel** - 0: clkRef-> data counter; 1: clkRef->40MHz ref

[0x1b4] CLKGStatus8

- Bit 7 - **CLKG_vcoRailMode** - 0: voltage mode, 1: current mode
- Bit 6 - **CLKG_ENABLE_CDR_R** - 0: CDR's resistor is disconnected; 1: connected
- Bit 5 - **CLKG_smLocked** - ljCDR state machine locked flag
- Bit 4 - **CLKG_lfInstLock** - lock filter instant lock signal (only in TX mode)
- Bit 3 - **CLKG_lfLocked** - lock filter locked signal (only in TX mode)
- Bit 2:0 - **CLKG_CONFIG_FF_CAP[2:0]** - CDR's feed forward filter's capacitance

[0x1b5] CLKGStatus9

- Bit 5:4 - **CLKG_lfState[1:0]** - ljCDR's lock filter state machine

CLKG_lfState[1:0]	Value	Description
lfUnlfLockedState	2'b00	low-pass lock filter is unlocked
lfConfirmLockState	2'b01	low-pass lock filter is confirming lock
lfLockedState	2'b10	low-pass lock filter is locked
lfConfirmUnlockState	2'b11	low-pass lock filter is confirming unlock

- Bit 3:0 - **CLKG_smState[3:0]** - ljCDR's state machine

CLKG_smState[3:0]	Value	Description
smResetState	4'h0	reset state
smInit	4'h1	initialization state (1cycle)
smCapSearchStart	4'h2	start VCO calibration (jump to smPLLInit or smCDRInit when finished)
smCapSearchClearCounters0	4'h3	VCO calibration step; clear counters
smCapSearchClearCounters1	4'h4	VCO calibration step; clear counters
smCapSearchEnableCounter	4'h5	VCO calibration step; start counters
smCapSearchWaitFreqDecision	4'h6	VCO calibration step; wait for race end
smCapSearchVCOFaster	4'h7	VCO calibration step; VCO is faster than refClk, increase capBank
smCapSearchRefClkFaster	4'h8	VCO calibration step; refClk is faster than VCO, decrease capBank
smPLLInit	4'h9	PLL step; closing PLL loop and waiting for lock state
smCDRInit	4'hA	CDR step; closing CDR loop and waiting for lock state
smPLLEnd	4'hB	PLL step; PLL is locked
smCDREnd	4'hC	CDR step; CDR is locked

15.3.8 FEC

[0x1b6] DLDPFecCorrectionCountH

Number of error reported by the FEC decoder in the downlink data path.

- **Bit 7:0 - DLDPFecCorrectionCount[15:8]** - Bits 15:8 of the FEC correction counter.

[0x1b7] DLDPFecCorrectionCountL

Number of error reported by the FEC decoder in the downlink data path.

- **Bit 7:0 - DLDPFecCorrectionCount[7:0]** - Bits 7:0 of the FEC correction counter.

15.3.9 ADC

[0x1b8] ADCStatusH

ADC status register

- **Bit 7 - ADCBusy** - ADC core is performing conversion.
- **Bit 6 - ADCDone** - ADC conversion is done. Result of conversion can be accessed in ADCValue
- **Bit 1:0 - ADCValue[9:8]** - Result of the last conversion.

See also: [\[0x1b9\] ADCStatusL](#) (page 264)

[0x1b9] ADCStatusL

ADC status register

- **Bit 7:0 - ADCValue[7:0]** - Result of the last conversion.

See also: [\[0x1b8\] ADCStatusH](#) (page 264)

15.3.10 Eye Opening Monitor

[0x1ba] EOMStatus

- **Bit 3:2 - EOMsmState[1:0]** - EOM state machine

EOMsmState[1:0]	Value	Description
smIdle	2'b00	idle state
smResetCounters	2'b01	resets the EOM ripple counter
smCount	2'b10	EOM ripple counter is counting
smEndOfCount	2'b11	finished state; waiting for EOMStart to go down

- **Bit 1 - EOMBusy** - Its hold high by the state machine when the ripple counter is in use
- **Bit 0 - EOMEnd** - Its hold high when the counting is done. It is kept high until (EOMStart | EOMEnable) goes low

[0x1bb] EOMCouterValueH

- **Bit 7:0 - EOMcounterValue[15:8]** - MSB word of EOM ripple counter (bigger the value, more the eye is open in this (x,y) position)

[0x1bc] EOMCouterValueL

- **Bit 7:0 - EOMcounterValue[7:0]** - LSB word of EOM ripple counter (bigger the value, more the eye is open in this (x,y) position)

[0x1bd] EOMCounter40MH

- **Bit 7:0 - EOMCounter40M[15:8]** - MSB word of EOM gating counter (toggles at 40 MHz); used to estimate number of data transitions

[0x1be] EOMCounter40ML

- **Bit 7:0 - EOMCounter40M[7:0]** - LSB word of EOM gating counter (toggles at 40 MHz); used to estimate number of data transitions

15.3.11 BERT Tester

[0x1bf] BERTStatus

Status register of BERT checker.

- **Bit 2 - BERTPrbsErrorFlag** - This flag is set when data input was always zero during the test.
- **Bit 1 - BERTBusy** - Measurement is ongoing when this bit is asserted.
- **Bit 0 - BERTDone** - Measurement is down when this bit is asserted.

[0x1c0] BERTResult4

BERT result.

- **Bit 7:0 - BERTErrorCount[39:32]** - Bits 39:32 of BERT result.

[0x1c1] BERTResult3

BERT result.

- **Bit 7:0 - BERTErrorCount[31:24]** - Bits 31:24 of BERT result.

[0x1c2] BERTResult2

BERT result.

- **Bit 7:0 - BERTErrorCount[23:16]** - Bits 23:16 of BERT result.

[0x1c3] BERTResult1

BERT result.

- **Bit 7:0 - BERTErrorCount[15:8]** - Bits 15:8 of BERT result.

[0x1c4] BERTResult0

BERT result.

- **Bit 7:0 - BERTErrorCount[7:0]** - Bits 7:0 of BERT result.

15.3.12 ROM

[0x1c5] ROM

Register with fixed (non zero value). Can be used for testing purposes.

- **Bit 7:0 - ROMREG[7:0]** - All read requests for this register should yield value *0xA5*.

15.3.13 POR

[0x1c6] PORBOR

Status of POR and BOR instances

- **Bit 6 - PORC** - State of PORC output
- **Bit 5 - PORB** - State of PORB output
- **Bit 4 - PORA** - State of PORA output
- **Bit 2 - BODC** - State of BODC output
- **Bit 1 - BODB** - State of BODB output
- **Bit 0 - BODA** - State of BODA output

15.3.14 Power Up State Machine

[0x1c7] PUSMStatus

Status of power up state machine

- **Bit 4:0 - PUSMState[4:0]** - Current state of the power up state machine

[0x1c8] PUSMActions

Status of power up state machine actions

- **Bit 5 - PUSMPllTimeoutAction** - This flag is set if the PLL timeout action has been executed since the last chip reset.
- **Bit 4 - PUSMDllTimeoutAction** - This flag is set if the DLL timeout action has been executed since the last chip reset.
- **Bit 3 - PUSMChannelsTimeoutAction** - This flag is set if the wait for channels locked timeout action has been executed since the last chip reset.
- **Bit 2 - PUSMbrownoutAction** - This flag is set if the brownout action has been executed since the last chip reset.
- **Bit 1 - PUSMPLLwatchdogAction** - This flag is set if the PLL watchdog action has been executed since the last chip reset.
- **Bit 0 - PUSMDLLwatchdogAction** - This flag is set if the DLL watchdog action has been executed since the last chip reset.

15.3.15 Debug

[0x1c9] TValue

Test output read back

- **Bit 5:0 - TOVal[5:0]** - Current value of all test outputs

[0x1ca] SCStatus

Serial interface (IC/EC) status register.

- **Bit 0 - SCParityValid** - The last parity bit check result.

[0x1cb] FAState

State of the frame aligner state machine

- **Bit 3:0 - FAState[3:0]** - State of the frame aligner state machine. TODO add the description.

[0x1cc] FACounter

Value of the counter in frame aligner.

- **Bit 7:0 - FACounter[7:0]** - Interpretation of this field depends on the current state.

[0x1cd] ConfigErrorCounterH

Counter of SEU events in configuration memory.

- **Bit 7:0 - ConfigErrorCounter[15:8]** - Bits 15:8 of the configuration memory SEU counter.

See also: [\[0x1ce\] ConfigErrorCounterL](#) (page 268)

[0x1ce] ConfigErrorCounterL

Counter of SEU events in configuration memory.

- **Bit 7:0 - ConfigErrorCounter[7:0]** - Bits 7:0 of the configuration memory SEU counter.

See also: [\[0x1cd\] ConfigErrorCounterH](#) (page 268)

**CHAPTER
SIXTEEN**

MODEL

The lpGBT model is available for users in order to simplify the design and verification of front-end (ASIC) and back-end (FPGA) systems. The model contains all essential features related to the data transition and slow control interfaces. As the model is meant to be used for digital simulations, some of the chip's analog features are not modeled. Among the others, features related to pre-emphasis, equalization, analog I/O (ADC,DAC), pull up / pull down resistors are omitted. The detailed list of blocks is presented [Table 16.1](#).

Table 16.1: IpGBT model features breakdown

Feature	Model	Remarks
High Speed Serializer	full	
High Speed Deserializer	full	
Clock generation and distribution system	full	Includes CDR and PLL
ePortRx	full	
ePortTx	full	
Phase shifted clocks	full	
Configuration interfaces	full	Includes EC/IC channels, fuses, I2C slave
I2C Masters	full	
Data path	full	Includes scramblers, interleaves, FEC codecs, pattern generators and pattern checkers
Process monitor and SEU monitor	full	
Power up sequence	partial	Brown-out detection circuit is not modeled. Power up counters are shorter (to decrease simulation time). Power on reset pulse is shorter (to decrease simulation time).
General purpose IO	partial	Pull up/ pull down features are not modeled
eRX (differential receiver)	partial	Equalization, common mode bias, termination are not modeled
eTX (differential driver)	partial	Output amplitude and pre-emphasis are not modeled
Line driver	partial	Output amplitude and pre-emphasis are not modeled
Eye opening monitor	partial	Analog blocks are not modeled
Analog peripherals	not modeled	ADC, DAC, temperature sensor features are not modeled

The model is hosted in GITLAB repository which can be found here: https://gitlab.cern.ch/lpgbt/lpgbt_model.

16.1 Top module connectivity

The IpGBT model comes as one encrypted System Verilog (SVP) file in which the IpGBT module is declared. All **233 ports** of the module correspond to pins described in [Section 17](#). Moreover, the module offers **240 8-bit parameters** which correspond to individual electrical fuses. Definition of the top level module is shown below:

```
module IpGBT #(
    parameter [7:0] FUSE0x00_CHIPID0 = 8'h0,
    parameter [7:0] FUSE0x01_CHIPID1 = 8'h0,
    parameter [7:0] FUSE0x02_CHIPID2 = 8'h0,
    parameter [7:0] FUSE0x03_CHIPID3 = 8'h0,
    parameter [7:0] FUSE0x04_USERID0 = 8'h0,
    parameter [7:0] FUSE0x05_USERID1 = 8'h0,
```

```

parameter [7:0] FUSE0x06_USERID2 = 8'h0,
parameter [7:0] FUSE0x07_USERID3 = 8'h0,
parameter [7:0] FUSE0x08_DACCAL0 = 8'h0,
parameter [7:0] FUSE0x09_DACCAL1 = 8'h0,
parameter [7:0] FUSE0x0A_DACCAL2 = 8'h0,
parameter [7:0] FUSE0x0B_DACCAL0 = 8'h0,
parameter [7:0] FUSE0x0C_DACCAL1 = 8'h0,
parameter [7:0] FUSE0x0D_DACCAL2 = 8'h0,
parameter [7:0] FUSE0x0E_DACCAL3 = 8'h0,
parameter [7:0] FUSE0x0F_DACCAL4 = 8'h0,
parameter [7:0] FUSE0x10_DACCAL5 = 8'h0,
parameter [7:0] FUSE0x11_DACCAL6 = 8'h0,
parameter [7:0] FUSE0x12_DACCAL7 = 8'h0,
parameter [7:0] FUSE0x13_DACCAL8 = 8'h0,
parameter [7:0] FUSE0x14_DACCAL9 = 8'h0,
parameter [7:0] FUSE0x15_DACCAL10 = 8'h0,
parameter [7:0] FUSE0x16_DACCAL11 = 8'h0,
parameter [7:0] FUSE0x17_DACCAL12 = 8'h0,
parameter [7:0] FUSE0x18_DACCAL13 = 8'h0,
parameter [7:0] FUSE0x19_DACCAL14 = 8'h0,
parameter [7:0] FUSE0x1A_TEMPCALH = 8'h0,
parameter [7:0] FUSE0x1B_TEMPCALL = 8'h0,
parameter [7:0] FUSE0x1C_VREFCNTR = 8'h0,
parameter [7:0] FUSE0x1D_CURDACCALH = 8'h0,
parameter [7:0] FUSE0x1E_CURDACCAL = 8'h0,
parameter [7:0] FUSE0x1F_RESERVED0 = 8'h0,
parameter [7:0] FUSE0x20_CLKGCONFIG0 = 8'h0,
parameter [7:0] FUSE0x21_CLKGCONFIG1 = 8'h0,
parameter [7:0] FUSE0x22_CLKGPLLRES = 8'h0,
parameter [7:0] FUSE0x23_CLKGPLLINTCUR = 8'h0,
parameter [7:0] FUSE0x24_CLKGPLLPROPCUR = 8'h0,
parameter [7:0] FUSE0x25_CLKGCDRPROPCUR = 8'h0,
parameter [7:0] FUSE0x26_CLKGCDRINTCUR = 8'h0,
parameter [7:0] FUSE0x27_CLKGCDRFFPROPCUR = 8'h0,
parameter [7:0] FUSE0x28_CLKGFLLINTCUR = 8'h0,
parameter [7:0] FUSE0x29_CLKGFFCAP = 8'h0,
parameter [7:0] FUSE0x2A_CLKGCNTOVERRIDE = 8'h0,
parameter [7:0] FUSE0x2B_CLKGOVERRIDECAPBANK = 8'h0,
parameter [7:0] FUSE0x2C_CLKGWAITTIME = 8'h0,
parameter [7:0] FUSE0x2D_CLKGLFCONFIG0 = 8'h0,
parameter [7:0] FUSE0x2E_CLKGLFCONFIG1 = 8'h0,
parameter [7:0] FUSE0x2F_FAMAXHEADERFOUNDCOUNT = 8'h0,
parameter [7:0] FUSE0x30_FAMAXHEADERFOUNDCOUNTAFTERNF = 8'h0,
parameter [7:0] FUSE0x31_FAMAXHEADERNOTFOUNDCOUNT = 8'h0,
parameter [7:0] FUSE0x32_FAFAAXSKIPCYCLECOUNTAFTERNF = 8'h0,
parameter [7:0] FUSE0x33_PSDLLCONFIG = 8'h0,
parameter [7:0] FUSE0x34_EPRXDLLCONFIG = 8'h0,
parameter [7:0] FUSE0x35_FORCEENABLE = 8'h0,
parameter [7:0] FUSE0x36_CHIPADR = 8'h0,
parameter [7:0] FUSE0x37_EQCONFIG = 8'h0,
parameter [7:0] FUSE0x38_EQRES = 8'h0,
parameter [7:0] FUSE0x39_LDCCONFIGH = 8'h0,
parameter [7:0] FUSE0x3A_LDCCONFIGL = 8'h0,
parameter [7:0] FUSE0x3B_REFCLK = 8'h0,
parameter [7:0] FUSE0x3C_SCCONFIG = 8'h0,
parameter [7:0] FUSE0x3D_RESETCONFIG = 8'h0,
parameter [7:0] FUSE0x3E_PGCConfig = 8'h0,
parameter [7:0] FUSE0x3F_I2CMTRANS CONFIG = 8'h0,

```

```

parameter [7:0] FUSE0x40_I2CMTRANSADDRESS = 8'h0,
parameter [7:0] FUSE0x41_I2CMTRANSCTRL = 8'h0,
parameter [7:0] FUSE0x42_I2CMTRANSADATA0 = 8'h0,
parameter [7:0] FUSE0x43_I2CMTRANSADATA1 = 8'h0,
parameter [7:0] FUSE0x44_I2CMTRANSADATA2 = 8'h0,
parameter [7:0] FUSE0x45_I2CMTRANSADATA3 = 8'h0,
parameter [7:0] FUSE0x46_I2CMTRANSADATA4 = 8'h0,
parameter [7:0] FUSE0x47_I2CMTRANSADATA5 = 8'h0,
parameter [7:0] FUSE0x48_I2CMTRANSADATA6 = 8'h0,
parameter [7:0] FUSE0x49_I2CMTRANSADATA7 = 8'h0,
parameter [7:0] FUSE0x4A_I2CMTRANSADATA8 = 8'h0,
parameter [7:0] FUSE0x4B_I2CMTRANSADATA9 = 8'h0,
parameter [7:0] FUSE0x4C_I2CMTRANSADATA10 = 8'h0,
parameter [7:0] FUSE0x4D_I2CMTRANSADATA11 = 8'h0,
parameter [7:0] FUSE0x4E_I2CMTRANSADATA12 = 8'h0,
parameter [7:0] FUSE0x4F_I2CMTRANSADATA13 = 8'h0,
parameter [7:0] FUSE0x50_I2CMTRANSADATA14 = 8'h0,
parameter [7:0] FUSE0x51_I2CMTRANSADATA15 = 8'h0,
parameter [7:0] FUSE0x52_PIODIRH = 8'h0,
parameter [7:0] FUSE0x53_PIODIRL = 8'h0,
parameter [7:0] FUSE0x54_PIOOUTH = 8'h0,
parameter [7:0] FUSE0x55_PIOOUTL = 8'h0,
parameter [7:0] FUSE0x56_PIOFULLENAH = 8'h0,
parameter [7:0] FUSE0x57_PIOFULLENAL = 8'h0,
parameter [7:0] FUSE0x58_PIOUPDOWNH = 8'h0,
parameter [7:0] FUSE0x59_PIOUPDOWNL = 8'h0,
parameter [7:0] FUSE0x5A_PIODRIVESTRENGTHH = 8'h0,
parameter [7:0] FUSE0x5B_PIODRIVESTRENGTHL = 8'h0,
parameter [7:0] FUSE0x5C_PS0CONFIG = 8'h0,
parameter [7:0] FUSE0x5D_PS0DELAY = 8'h0,
parameter [7:0] FUSE0x5E_PS0OUTDRIVER = 8'h0,
parameter [7:0] FUSE0x5F_PS1CONFIG = 8'h0,
parameter [7:0] FUSE0x60_PS1DELAY = 8'h0,
parameter [7:0] FUSE0x61_PS1OUTDRIVER = 8'h0,
parameter [7:0] FUSE0x62_PS2CONFIG = 8'h0,
parameter [7:0] FUSE0x63_PS2DELAY = 8'h0,
parameter [7:0] FUSE0x64_PS2OUTDRIVER = 8'h0,
parameter [7:0] FUSE0x65_PS3CONFIG = 8'h0,
parameter [7:0] FUSE0x66_PS3DELAY = 8'h0,
parameter [7:0] FUSE0x67_PS3OUTDRIVER = 8'h0,
parameter [7:0] FUSE0x68_DACCONFIGH = 8'h0,
parameter [7:0] FUSE0x69_DACCONFIGL = 8'h0,
parameter [7:0] FUSE0x6A_CURDACVALUE = 8'h0,
parameter [7:0] FUSE0x6B_CURDACCHN = 8'h0,
parameter [7:0] FUSE0x6C_EPCLK0CHNCNTRH = 8'h0,
parameter [7:0] FUSE0x6D_EPCLK0CHNCNTRL = 8'h0,
parameter [7:0] FUSE0x6E_EPCLK1CHNCNTRH = 8'h0,
parameter [7:0] FUSE0x6F_EPCLK1CHNCNTRL = 8'h0,
parameter [7:0] FUSE0x70_EPCLK2CHNCNTRH = 8'h0,
parameter [7:0] FUSE0x71_EPCLK2CHNCNTRL = 8'h0,
parameter [7:0] FUSE0x72_EPCLK3CHNCNTRH = 8'h0,
parameter [7:0] FUSE0x73_EPCLK3CHNCNTRL = 8'h0,
parameter [7:0] FUSE0x74_EPCLK4CHNCNTRH = 8'h0,
parameter [7:0] FUSE0x75_EPCLK4CHNCNTRL = 8'h0,
parameter [7:0] FUSE0x76_EPCLK5CHNCNTRH = 8'h0,
parameter [7:0] FUSE0x77_EPCLK5CHNCNTRL = 8'h0,
parameter [7:0] FUSE0x78_EPCLK6CHNCNTRH = 8'h0,
parameter [7:0] FUSE0x79_EPCLK6CHNCNTRL = 8'h0,

```

```

parameter [7:0] FUSE0x7A_EPCLK7CHNCNTRH = 8'h0,
parameter [7:0] FUSE0x7B_EPCLK7CHNCNTRL = 8'h0,
parameter [7:0] FUSE0x7C_EPCLK8CHNCNTRH = 8'h0,
parameter [7:0] FUSE0x7D_EPCLK8CHNCNTRL = 8'h0,
parameter [7:0] FUSE0x7E_EPCLK9CHNCNTRH = 8'h0,
parameter [7:0] FUSE0x7F_EPCLK9CHNCNTRL = 8'h0,
parameter [7:0] FUSE0x80_EPCLK10CHNCNTRH = 8'h0,
parameter [7:0] FUSE0x81_EPCLK10CHNCNTRL = 8'h0,
parameter [7:0] FUSE0x82_EPCLK11CHNCNTRH = 8'h0,
parameter [7:0] FUSE0x83_EPCLK11CHNCNTRL = 8'h0,
parameter [7:0] FUSE0x84_EPCLK12CHNCNTRH = 8'h0,
parameter [7:0] FUSE0x85_EPCLK12CHNCNTRL = 8'h0,
parameter [7:0] FUSE0x86_EPCLK13CHNCNTRH = 8'h0,
parameter [7:0] FUSE0x87_EPCLK13CHNCNTRL = 8'h0,
parameter [7:0] FUSE0x88_EPCLK14CHNCNTRH = 8'h0,
parameter [7:0] FUSE0x89_EPCLK14CHNCNTRL = 8'h0,
parameter [7:0] FUSE0x8A_EPCLK15CHNCNTRH = 8'h0,
parameter [7:0] FUSE0x8B_EPCLK15CHNCNTRL = 8'h0,
parameter [7:0] FUSE0x8C_EPCLK16CHNCNTRH = 8'h0,
parameter [7:0] FUSE0x8D_EPCLK16CHNCNTRL = 8'h0,
parameter [7:0] FUSE0x8E_EPCLK17CHNCNTRH = 8'h0,
parameter [7:0] FUSE0x8F_EPCLK17CHNCNTRL = 8'h0,
parameter [7:0] FUSE0x90_EPCLK18CHNCNTRH = 8'h0,
parameter [7:0] FUSE0x91_EPCLK18CHNCNTRL = 8'h0,
parameter [7:0] FUSE0x92_EPCLK19CHNCNTRH = 8'h0,
parameter [7:0] FUSE0x93_EPCLK19CHNCNTRL = 8'h0,
parameter [7:0] FUSE0x94_EPCLK20CHNCNTRH = 8'h0,
parameter [7:0] FUSE0x95_EPCLK20CHNCNTRL = 8'h0,
parameter [7:0] FUSE0x96_EPCLK21CHNCNTRH = 8'h0,
parameter [7:0] FUSE0x97_EPCLK21CHNCNTRL = 8'h0,
parameter [7:0] FUSE0x98_EPCLK22CHNCNTRH = 8'h0,
parameter [7:0] FUSE0x99_EPCLK22CHNCNTRL = 8'h0,
parameter [7:0] FUSE0x9A_EPCLK23CHNCNTRH = 8'h0,
parameter [7:0] FUSE0x9B_EPCLK23CHNCNTRL = 8'h0,
parameter [7:0] FUSE0x9C_EPCLK24CHNCNTRH = 8'h0,
parameter [7:0] FUSE0x9D_EPCLK24CHNCNTRL = 8'h0,
parameter [7:0] FUSE0x9E_EPCLK25CHNCNTRH = 8'h0,
parameter [7:0] FUSE0x9F_EPCLK25CHNCNTRL = 8'h0,
parameter [7:0] FUSE0xA0_EPCLK26CHNCNTRH = 8'h0,
parameter [7:0] FUSE0xA1_EPCLK26CHNCNTRL = 8'h0,
parameter [7:0] FUSE0xA2_EPCLK27CHNCNTRH = 8'h0,
parameter [7:0] FUSE0xA3_EPCLK27CHNCNTRL = 8'h0,
parameter [7:0] FUSE0xA4_EPCLK28CHNCNTRH = 8'h0,
parameter [7:0] FUSE0xA5_EPCLK28CHNCNTRL = 8'h0,
parameter [7:0] FUSE0xA6_RESERVED1 = 8'h0,
parameter [7:0] FUSE0xA7_EPTXDATARATE = 8'h0,
parameter [7:0] FUSE0xA8_EPTXCONTROL = 8'h0,
parameter [7:0] FUSE0xA9_EPTX10ENABLE = 8'h0,
parameter [7:0] FUSE0xAA_EPTX32ENABLE = 8'h0,
parameter [7:0] FUSE0xAB_EPTXECCHNCNTR = 8'h0,
parameter [7:0] FUSE0xAC_EPTX00CHNCNTR = 8'h0,
parameter [7:0] FUSE0xAD_EPTX01CHNCNTR = 8'h0,
parameter [7:0] FUSE0xAE_EPTX02CHNCNTR = 8'h0,
parameter [7:0] FUSE0xAF_EPTX03CHNCNTR = 8'h0,
parameter [7:0] FUSE0xB0_EPTX10CHNCNTR = 8'h0,
parameter [7:0] FUSE0xB1_EPTX11CHNCNTR = 8'h0,
parameter [7:0] FUSE0xB2_EPTX12CHNCNTR = 8'h0,
parameter [7:0] FUSE0xB3_EPTX13CHNCNTR = 8'h0,

```

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parameter [7:0] FUSE0xB4_EPTX20CHNCNTR = 8'h0,
parameter [7:0] FUSE0xB5_EPTX21CHNCNTR = 8'h0,
parameter [7:0] FUSE0xB6_EPTX22CHNCNTR = 8'h0,
parameter [7:0] FUSE0xB7_EPTX23CHNCNTR = 8'h0,
parameter [7:0] FUSE0xB8_EPTX30CHNCNTR = 8'h0,
parameter [7:0] FUSE0xB9_EPTX31CHNCNTR = 8'h0,
parameter [7:0] FUSE0xBA_EPTX32CHNCNTR = 8'h0,
parameter [7:0] FUSE0xBB_EPTX33CHNCNTR = 8'h0,
parameter [7:0] FUSE0xBC_EPTX01_00CHNCNTR = 8'h0,
parameter [7:0] FUSE0xBD_EPTX03_02CHNCNTR = 8'h0,
parameter [7:0] FUSE0xBE_EPTX11_10CHNCNTR = 8'h0,
parameter [7:0] FUSE0xBF_EPTX13_12CHNCNTR = 8'h0,
parameter [7:0] FUSE0xC0_EPTX21_20CHNCNTR = 8'h0,
parameter [7:0] FUSE0xC1_EPTX23_22CHNCNTR = 8'h0,
parameter [7:0] FUSE0xC2_EPTX31_30CHNCNTR = 8'h0,
parameter [7:0] FUSE0xC3_EPTX33_32CHNCNTR = 8'h0,
parameter [7:0] FUSE0xC4_EPRX0CONTROL = 8'h0,
parameter [7:0] FUSE0xC5_EPRX1CONTROL = 8'h0,
parameter [7:0] FUSE0xC6_EPRX2CONTROL = 8'h0,
parameter [7:0] FUSE0xC7_EPRX3CONTROL = 8'h0,
parameter [7:0] FUSE0xC8_EPRX4CONTROL = 8'h0,
parameter [7:0] FUSE0xC9_EPRX5CONTROL = 8'h0,
parameter [7:0] FUSE0xCA_EPRX6CONTROL = 8'h0,
parameter [7:0] FUSE0xCB_EPRXECCONTROL = 8'h0,
parameter [7:0] FUSE0xCC_EPRX00CHNCNTR = 8'h0,
parameter [7:0] FUSE0xCD_EPRX01CHNCNTR = 8'h0,
parameter [7:0] FUSE0xCE_EPRX02CHNCNTR = 8'h0,
parameter [7:0] FUSE0xCF_EPRX03CHNCNTR = 8'h0,
parameter [7:0] FUSE0xD0_EPRX10CHNCNTR = 8'h0,
parameter [7:0] FUSE0xD1_EPRX11CHNCNTR = 8'h0,
parameter [7:0] FUSE0xD2_EPRX12CHNCNTR = 8'h0,
parameter [7:0] FUSE0xD3_EPRX13CHNCNTR = 8'h0,
parameter [7:0] FUSE0xD4_EPRX20CHNCNTR = 8'h0,
parameter [7:0] FUSE0xD5_EPRX21CHNCNTR = 8'h0,
parameter [7:0] FUSE0xD6_EPRX22CHNCNTR = 8'h0,
parameter [7:0] FUSE0xD7_EPRX23CHNCNTR = 8'h0,
parameter [7:0] FUSE0xD8_EPRX30CHNCNTR = 8'h0,
parameter [7:0] FUSE0xD9_EPRX31CHNCNTR = 8'h0,
parameter [7:0] FUSE0xDA_EPRX32CHNCNTR = 8'h0,
parameter [7:0] FUSE0xDB_EPRX33CHNCNTR = 8'h0,
parameter [7:0] FUSE0xDC_EPRX40CHNCNTR = 8'h0,
parameter [7:0] FUSE0xDD_EPRX41CHNCNTR = 8'h0,
parameter [7:0] FUSE0xDE_EPRX42CHNCNTR = 8'h0,
parameter [7:0] FUSE0xDF_EPRX43CHNCNTR = 8'h0,
parameter [7:0] FUSE0xE0_EPRX50CHNCNTR = 8'h0,
parameter [7:0] FUSE0xE1_EPRX51CHNCNTR = 8'h0,
parameter [7:0] FUSE0xE2_EPRX52CHNCNTR = 8'h0,
parameter [7:0] FUSE0xE3_EPRX53CHNCNTR = 8'h0,
parameter [7:0] FUSE0xE4_EPRX60CHNCNTR = 8'h0,
parameter [7:0] FUSE0xE5_EPRX61CHNCNTR = 8'h0,
parameter [7:0] FUSE0xE6_EPRX62CHNCNTR = 8'h0,
parameter [7:0] FUSE0xE7_EPRX63CHNCNTR = 8'h0,
parameter [7:0] FUSE0xE8_EPRXECCCHNCNTR = 8'h0,
parameter [7:0] FUSE0xE9_EPRXEQ10CONTROL = 8'h0,
parameter [7:0] FUSE0xEA_EPRXEQ32CONTROL = 8'h0,
parameter [7:0] FUSE0xEB_EPRXEQ54CONTROL = 8'h0,
parameter [7:0] FUSE0xEC_EPRXEQ6CONTROL = 8'h0,
parameter [7:0] FUSE0xED_POWERUP0 = 8'h0,

```

```

parameter [7:0] FUSE0xEE_POWERUP1 = 8'h0,
parameter [7:0] FUSE0xEF_POWERUP2 = 8'h0
) (
    // CORE and IO power supply
    input GND,
    input VDD1V2,

    // Transmitter power supply
    input GNDTX,
    input VDDTX1V2,

    // Receiver power supply
    input GNDRX,
    input VDDRX1V2,

    // Analog power supply
    input GNDA,
    input VDDA1V2,

    // Fuses power supply (uses GND for return currents)
    input VDDF2V5,

    // High speed serializer outputs
    output HSOUTP,
    output HSOUTN,

    // High speed deserializer inputs
    input HSINP,
    input HSINN,

    // ePort clock differential outputs
    output ECLK0P,
    output ECLK0N,

    output ECLK1P,
    output ECLK1N,

    output ECLK2P,
    output ECLK2N,

    output ECLK3P,
    output ECLK3N,

    output ECLK4P,
    output ECLK4N,

    output ECLK5P,
    output ECLK5N,

    output ECLK6P,
    output ECLK6N,

    output ECLK7P,
    output ECLK7N,

    output ECLK8P,
    output ECLK8N,

```

```
    output ECLK9P,  
    output ECLK9N,  
  
    output ECLK10P,  
    output ECLK10N,  
  
    output ECLK11P,  
    output ECLK11N,  
  
    output ECLK12P,  
    output ECLK12N,  
  
    output ECLK13P,  
    output ECLK13N,  
  
    output ECLK14P,  
    output ECLK14N,  
  
    output ECLK15P,  
    output ECLK15N,  
  
    output ECLK16P,  
    output ECLK16N,  
  
    output ECLK17P,  
    output ECLK17N,  
  
    output ECLK18P,  
    output ECLK18N,  
  
    output ECLK19P,  
    output ECLK19N,  
  
    output ECLK20P,  
    output ECLK20N,  
  
    output ECLK21P,  
    output ECLK21N,  
  
    output ECLK22P,  
    output ECLK22N,  
  
    output ECLK23P,  
    output ECLK23N,  
  
    output ECLK24P,  
    output ECLK24N,  
  
    output ECLK25P,  
    output ECLK25N,  
  
    output ECLK26P,  
    output ECLK26N,  
  
    output ECLK27P,  
    output ECLK27N,  
  
    output ECLK28P,
```

```

output ECLK28N,  

// ePortTX group 0 differential data outputs (downlink)  

output EDOUT00P,  

output EDOUT00N,  

output EDOUT01P,  

output EDOUT01N,  

output EDOUT02P,  

output EDOUT02N,  

output EDOUT03P,  

output EDOUT03N,  

// ePortTX group 1 differential data outputs (downlink)  

output EDOUT10P,  

output EDOUT10N,  

output EDOUT11P,  

output EDOUT11N,  

output EDOUT12P,  

output EDOUT12N,  

output EDOUT13P,  

output EDOUT13N,  

// ePortTX group 2 differential data outputs (downlink)  

output EDOUT20P,  

output EDOUT20N,  

output EDOUT21P,  

output EDOUT21N,  

output EDOUT22P,  

output EDOUT22N,  

output EDOUT23P,  

output EDOUT23N,  

// ePortTX group 3 differential data outputs (downlink)  

output EDOUT30P,  

output EDOUT30N,  

output EDOUT31P,  

output EDOUT31N,  

output EDOUT32P,  

output EDOUT32N,  

output EDOUT33P,  

output EDOUT33N,  

// ePortTX EC differential data outputs  

output EDOUTECP,  

output EDOUTECN,  

// ePortRX group 0 differential data inputs (uplink)  

input EDIN00P,  

input EDIN00N,  

input EDIN01P,  

input EDIN01N,  

input EDIN02P,  

input EDIN02N,  

input EDIN03P,  

input EDIN03N,  

// ePortRX group 1 differential data inputs (uplink)  

input EDIN10P,

```

```
    input EDIN10N,  
    input EDIN11P,  
    input EDIN11N,  
    input EDIN12P,  
    input EDIN12N,  
    input EDIN13P,  
    input EDIN13N,  
  
    // ePortRX group 2 differential data inputs (uplink)  
    input EDIN20P,  
    input EDIN20N,  
    input EDIN21P,  
    input EDIN21N,  
    input EDIN22P,  
    input EDIN22N,  
    input EDIN23P,  
    input EDIN23N,  
  
    // ePortRX group 3 differential data inputs (uplink)  
    input EDIN30P,  
    input EDIN30N,  
    input EDIN31P,  
    input EDIN31N,  
    input EDIN32P,  
    input EDIN32N,  
    input EDIN33P,  
    input EDIN33N,  
  
    // ePortRX group 4 differential data inputs (uplink)  
    input EDIN40P,  
    input EDIN40N,  
    input EDIN41P,  
    input EDIN41N,  
    input EDIN42P,  
    input EDIN42N,  
    input EDIN43P,  
    input EDIN43N,  
  
    // ePortRX group 5 differential data inputs (uplink)  
    input EDIN50P,  
    input EDIN50N,  
    input EDIN51P,  
    input EDIN51N,  
    input EDIN52P,  
    input EDIN52N,  
    input EDIN53P,  
    input EDIN53N,  
  
    // ePortRX group 6 differential data inputs (uplink)  
    input EDIN60P,  
    input EDIN60N,  
    input EDIN61P,  
    input EDIN61N,  
    input EDIN62P,  
    input EDIN62N,  
    input EDIN63P,  
    input EDIN63N,
```

```

// ePortRX EC differential data inputs
 EDINECP,
 EDINECN,

// Phase shifted clocks
 PSCLK0P,
 PSCLK0N,
 PSCLK1P,
 PSCLK1N,
 PSCLK2P,
 PSCLK2N,
 PSCLK3P,
 PSCLK3N,

// I2C slave for ASIC control
 SLSDA,
 SLSCL,

// Address
 ADR0,
 ADR1,
 ADR2,
 ADR3,

// lock mode
 LOCKMODE,

// reset input (active low)
 RSTB,

// reset output signal
 RSTOUTB,

// mode of operation
 MODE0,
 MODE1,
 MODE2,
 MODE3,

// lpGBT Ready signal
 READY,

// Power On Reset Disable
 PORDIS,

// reference clock
 REFCLKP,
 REFCLKN,

// Test clock input (used only for debugging)
 TSTCLKINP,
 TSTCLKINN,

// Bypass VCO and use test clock input (TSTCLKIN) instead
 VCOBYPASS,

// State override for the power up state machine
 STATEOVRD,

```

```
// Selects configuration interface (0=SerialControl1, 1=Slave I2C)
 SC_I2C,  
  
// Test outputs (0-3 CMOS, 4-5 differential)
 TSTOUT0,
 TSTOUT1,
 TSTOUT2,
 TSTOUT3,
 TSTOUT4P,
 TSTOUT4N,
 TSTOUT5P,
 TSTOUT5N,  
  
// I2C Master 0 signals
 M0SDA,
 M0SCL,  
  
// I2C Master 1 signals
 M1SDA,
 M1SCL,  
  
// I2C Master 2 signals
 M2SDA,
 M2SCL,  
  
// Parallel I/O
 GPIO0,
 GPIO1,
 GPIO2,
 GPIO3,
 GPIO4,
 GPIO5,
 GPIO6,
 GPIO7,
 GPIO8,
 GPIO9,
 GPIO10,
 GPIO11,
 GPIO12,
 GPIO13,
 GPIO14,
 GPIO15,  
  
// ADC input (and current source output)
 ADC0,
 ADC1,
 ADC2,
 ADC3,
 ADC4,
 ADC5,
 ADC6,
 ADC7,  
  
// Voltage DAC output
 VDAC,  
  
// reference voltage
```

```

inout VREF,
// debug signals (not present on the chip package)
output [463*8-1:0] debug_registers,
output [127:0] debug_testOutputs
);
endmodule

```

Besides *real* pins, the models provides two additional outputs: `debug_registers` and `debug_testOutputs` which allow to spy on internal register values and test outputs respectively.

An example instigation of the lpGBT model is presented in the code snippet below:

```

`include "lpGBTDefines.v"

localparam [31:0] MYID = 32'h76543210;
wire [3:0] MODE = LPGBT_5G_FEC12_TX;

[...]

lpGBT #( 
    // assign chip ID
    .FUSE0x00_CHIPIDO      (MYID[ 7: 0]),
    .FUSE0x01_CHIPID1       (MYID[15: 8]),
    .FUSE0x02_CHIPID2       (MYID[23:16]),
    .FUSE0x03_CHIPID3       (MYID[31:24]),
    // configure clock generator block
    .FUSE0x20_CLKGCONFIG0   ( ('h8 << CLKGCONFIG0_CLKGCALIBRATIONENDOFCOUNT_of) ,
    ↪ | ('h8 << CLKGCONFIG0_CLKGBIASGENCONFIG_of) ),
    .FUSE0x21_CLKGCONFIG1   ( (CLKGCONFIG1_CLKGCDRRES_bm
    ↪ | ('h4 << CLKGCONFIG1_CLKGVCODAC_of) ),
    .FUSE0x23_CLKGPLLINTCUR ( ('h5 << CLKGPLLINTCUR_CLKGPLLINTCURWHENLOCKED_
    ↪ of) | ('h5 << CLKGPLLINTCUR_CLKGPLLINTCUR_of) ),
    .FUSE0x24_CLKGPLLPROPCUR ( ('h5 << CLKGPLLPROPCUR_
    ↪ CLKGPLLPROPCURWHENLOCKED_of) | ('h5 << CLKGPLLPROPCUR_
    ↪ CLKGPLLPROPCUR_of) ),
    .FUSE0x22_CLKGPLLRES    ( ('h4 << CLKGPLLRES_CLKGPLLRESWHENLOCKED_of) |
    ↪ | ('h4 << CLKGPLLRES_CLKGPLLRES_of) ),
    .FUSE0x29_CLKGFFCAP     ( ('h6 << CLKGFFCAP_CLKGFEEDFORWARDCAPWHENLOCKED_
    ↪ of) | ('h6 << CLKGFFCAP_CLKGFEEDFORWARDCAP_of) ),
    .FUSE0x26_CLKGCDRINTCUR ( ('h5 << CLKGCDRINTCUR_CLKGCDRINTCURWHENLOCKED_
    ↪ of) | ('h5 << CLKGCDRINTCUR_CLKGCDRINTCUR_of) ),
    .FUSE0x28_CLKGFLLINTCUR ( ('h0 << CLKGFLLINTCUR_CLKGFLLINTCURWHENLOCKED_
    ↪ of) | ('hF << CLKGCDRINTCUR_CLKGCDRINTCUR_of) ),
    .FUSE0x25_CLKGCDRPROPCUR ( ('h5 << CLKGCDRPROPCUR_
    ↪ CLKGCDRPROPCURWHENLOCKED_of) | ('h5 << CLKGCDRPROPCUR_
    ↪ CLKGCDRPROPCUR_of) ),
    .FUSE0x27_CLKGCDRFFPROPCUR ( ('h5 << CLKGCDRFFPROPCUR_
    ↪ CLKGCDRFFPROPCURWHENLOCKED_of) | ('h5 << CLKGCDRFFPROPCUR_
    ↪ CLKGCDRFFPROPCUR_of) ),
    .FUSE0x2D_CLKGLFCONFIG0  ( (CLKGLFCONFIG0_CLKGLOCKFILTERENABLE_bm
    ↪ | ('d9<<CLKGLFCONFIG0_CLKGLOCKFILTERLOCKTHRCOUNTER_
    ↪ of)), 
    .FUSE0x2E_CLKGLFCONFIG1  ( ('d9<<CLKGLFCONFIG1_
    ↪ CLKGLOCKFILTERRELOCKTHRCOUNTER_of) | ('d9<<CLKGLFCONFIG1_
    ↪ CLKGLOCKFILTERUNLOCKTHRCOUNTER_of)),
    .FUSE0x2C_CLKGWAITTIME   ( ('ha<<CLKGWAITTIME_CLKGWAITCDRTIME_of) |
    ↪ | ('ha<<CLKGWAITTIME_CLKGWAITPLLTIME_of)), 

```

```

// configure ePortClocks
.FUSE0x6C_EPCLK0CHNCNTRH  ( (1<<EPCLK0CHNCNTRH_EPCLK0DRIVESTRENGTH_of)
← | (EPORTCLOCKS_CLK160M << EPCLK0CHNCNTRH_EPCLK0FREQ_
→_of)),
.FUSE0x6E_EPCLK1CHNCNTRH  ( (1<<EPCLK1CHNCNTRH_EPCLK1DRIVESTRENGTH_of)
← | (EPORTCLOCKS_CLK160M << EPCLK1CHNCNTRH_EPCLK1FREQ_
→_of)),
.FUSE0x70_EPCLK2CHNCNTRH  ( (1<<EPCLK2CHNCNTRH_EPCLK2DRIVESTRENGTH_of)
← | (EPORTCLOCKS_CLK160M << EPCLK2CHNCNTRH_EPCLK2FREQ_
→_of)),
.FUSE0x72_EPCLK3CHNCNTRH  ( (1<<EPCLK3CHNCNTRH_EPCLK3DRIVESTRENGTH_of)
← | (EPORTCLOCKS_CLK160M << EPCLK3CHNCNTRH_EPCLK3FREQ_
→_of)),
.FUSE0x74_EPCLK4CHNCNTRH  ( (1<<EPCLK4CHNCNTRH_EPCLK4DRIVESTRENGTH_of)
← | (EPORTCLOCKS_CLK160M << EPCLK4CHNCNTRH_EPCLK4FREQ_
→_of)),
.FUSE0x76_EPCLK5CHNCNTRH  ( (1<<EPCLK5CHNCNTRH_EPCLK5DRIVESTRENGTH_of)
← | (EPORTCLOCKS_CLK160M << EPCLK5CHNCNTRH_EPCLK5FREQ_
→_of)),
.FUSE0x78_EPCLK6CHNCNTRH  ( (1<<EPCLK6CHNCNTRH_EPCLK6DRIVESTRENGTH_of)
← | (EPORTCLOCKS_CLK160M << EPCLK6CHNCNTRH_EPCLK6FREQ_
→_of)),
.FUSE0x7A_EPCLK7CHNCNTRH  ( (1<<EPCLK7CHNCNTRH_EPCLK7DRIVESTRENGTH_of)
← | (EPORTCLOCKS_CLK160M << EPCLK7CHNCNTRH_EPCLK7FREQ_
→_of)),
.FUSE0x7C_EPCLK8CHNCNTRH  ( (1<<EPCLK8CHNCNTRH_EPCLK8DRIVESTRENGTH_of)
← | (EPORTCLOCKS_CLK160M << EPCLK8CHNCNTRH_EPCLK8FREQ_
→_of)),
.FUSE0x7E_EPCLK9CHNCNTRH  ( (1<<EPCLK9CHNCNTRH_EPCLK9DRIVESTRENGTH_of)
← | (EPORTCLOCKS_CLK160M << EPCLK9CHNCNTRH_EPCLK9FREQ_
→_of)),
.FUSE0x80_EPCLK10CHNCNTRH ( (1<<EPCLK10CHNCNTRH_EPCLK10DRIVESTRENGTH_of)
← | (EPORTCLOCKS_CLK160M << EPCLK10CHNCNTRH_EPCLK10FREQ_
→_of)),
.FUSE0x82_EPCLK11CHNCNTRH ( (1<<EPCLK11CHNCNTRH_EPCLK11DRIVESTRENGTH_of)
← | (EPORTCLOCKS_CLK160M << EPCLK11CHNCNTRH_EPCLK11FREQ_
→_of)),
.FUSE0x84_EPCLK12CHNCNTRH ( (1<<EPCLK12CHNCNTRH_EPCLK12DRIVESTRENGTH_of)
← | (EPORTCLOCKS_CLK160M << EPCLK12CHNCNTRH_EPCLK12FREQ_
→_of)),
.FUSE0x86_EPCLK13CHNCNTRH ( (1<<EPCLK13CHNCNTRH_EPCLK13DRIVESTRENGTH_of)
← | (EPORTCLOCKS_CLK160M << EPCLK13CHNCNTRH_EPCLK13FREQ_
→_of)),
.FUSE0x88_EPCLK14CHNCNTRH ( (1<<EPCLK14CHNCNTRH_EPCLK14DRIVESTRENGTH_of)
← | (EPORTCLOCKS_CLK160M << EPCLK14CHNCNTRH_EPCLK14FREQ_
→_of)),
.FUSE0x8A_EPCLK15CHNCNTRH ( (1<<EPCLK15CHNCNTRH_EPCLK15DRIVESTRENGTH_of)
← | (EPORTCLOCKS_CLK160M << EPCLK15CHNCNTRH_EPCLK15FREQ_
→_of)),
.FUSE0x8C_EPCLK16CHNCNTRH ( (1<<EPCLK16CHNCNTRH_EPCLK16DRIVESTRENGTH_of)
← | (EPORTCLOCKS_CLK160M << EPCLK16CHNCNTRH_EPCLK16FREQ_
→_of)),
.FUSE0x8E_EPCLK17CHNCNTRH ( (1<<EPCLK17CHNCNTRH_EPCLK17DRIVESTRENGTH_of)
← | (EPORTCLOCKS_CLK160M << EPCLK17CHNCNTRH_EPCLK17FREQ_
→_of)),
.FUSE0x90_EPCLK18CHNCNTRH ( (1<<EPCLK18CHNCNTRH_EPCLK18DRIVESTRENGTH_of)
← | (EPORTCLOCKS_CLK160M << EPCLK18CHNCNTRH_EPCLK18FREQ_
→_of)),

```

```

.FUSE0x92_EPCLK19CHNCNTRH ( (1<<EPCLK19CHNCNTRH_EPCLK19DRIVESTRENGTH_of) |
→ | (EPORTCLOCKS_CLK160M << EPCLK19CHNCNTRH_EPCLK19FREQ_
→ of)),
.FUSE0x94_EPCLK20CHNCNTRH ( (1<<EPCLK20CHNCNTRH_EPCLK20DRIVESTRENGTH_of) |
→ | (EPORTCLOCKS_CLK160M << EPCLK20CHNCNTRH_EPCLK20FREQ_
→ of)),
.FUSE0x96_EPCLK21CHNCNTRH ( (1<<EPCLK21CHNCNTRH_EPCLK21DRIVESTRENGTH_of) |
→ | (EPORTCLOCKS_CLK160M << EPCLK21CHNCNTRH_EPCLK21FREQ_
→ of)),
.FUSE0x98_EPCLK22CHNCNTRH ( (1<<EPCLK22CHNCNTRH_EPCLK22DRIVESTRENGTH_of) |
→ | (EPORTCLOCKS_CLK160M << EPCLK22CHNCNTRH_EPCLK22FREQ_
→ of)),
.FUSE0x9A_EPCLK23CHNCNTRH ( (1<<EPCLK23CHNCNTRH_EPCLK23DRIVESTRENGTH_of) |
→ | (EPORTCLOCKS_CLK160M << EPCLK23CHNCNTRH_EPCLK23FREQ_
→ of)),
.FUSE0x9C_EPCLK24CHNCNTRH ( (1<<EPCLK24CHNCNTRH_EPCLK24DRIVESTRENGTH_of) |
→ | (EPORTCLOCKS_CLK160M << EPCLK24CHNCNTRH_EPCLK24FREQ_
→ of)),
.FUSE0x9E_EPCLK25CHNCNTRH ( (1<<EPCLK25CHNCNTRH_EPCLK25DRIVESTRENGTH_of) |
→ | (EPORTCLOCKS_CLK160M << EPCLK25CHNCNTRH_EPCLK25FREQ_
→ of)),
.FUSE0xA0_EPCLK26CHNCNTRH ( (1<<EPCLK26CHNCNTRH_EPCLK26DRIVESTRENGTH_of) |
→ | (EPORTCLOCKS_CLK160M << EPCLK26CHNCNTRH_EPCLK26FREQ_
→ of)),
.FUSE0xA2_EPCLK27CHNCNTRH ( (1<<EPCLK27CHNCNTRH_EPCLK27DRIVESTRENGTH_of) |
→ | (EPORTCLOCKS_CLK160M << EPCLK27CHNCNTRH_EPCLK27FREQ_
→ of)),
.FUSE0xA4_EPCLK28CHNCNTRH ( (1<<EPCLK28CHNCNTRH_EPCLK28DRIVESTRENGTH_of) |
→ | (EPORTCLOCKS_CLK80M << EPCLK28CHNCNTRH_EPCLK28FREQ_
→ of)),
// configure ePortRx
.FUSE0xC4_EPRX0CONTROL(EPRX0CONTROL_EPRX03ENABLE_bm | EPRX0CONTROL_
→ EPRX02ENABLE_bm | EPRX0CONTROL_EPRX01ENABLE_bm | EPRX0CONTROL_
→ EPRX00ENABLE_bm |
→ EPORTRXLS_X4<<EPRX0CONTROL_EPRX0DATARATE_of | |
→ EPORTRXMODE_CONTINUOUSTRACKING << EPRX0CONTROL_EPRX0TRACKMODE_of),

.FUSE0xC5_EPRX1CONTROL(EPRX1CONTROL_EPRX13ENABLE_bm | EPRX1CONTROL_
→ EPRX12ENABLE_bm | EPRX1CONTROL_EPRX11ENABLE_bm | EPRX1CONTROL_
→ EPRX10ENABLE_bm |
→ EPORTRXLS_X4<<EPRX1CONTROL_EPRX1DATARATE_of | |
→ EPORTRXMODE_CONTINUOUSTRACKING << EPRX1CONTROL_EPRX1TRACKMODE_of),

.FUSE0xC6_EPRX2CONTROL(EPRX2CONTROL_EPRX23ENABLE_bm | EPRX2CONTROL_
→ EPRX22ENABLE_bm | EPRX2CONTROL_EPRX21ENABLE_bm | EPRX2CONTROL_
→ EPRX20ENABLE_bm |
→ EPORTRXLS_X4<<EPRX2CONTROL_EPRX2DATARATE_of | |
→ EPORTRXMODE_CONTINUOUSTRACKING << EPRX2CONTROL_EPRX2TRACKMODE_of),

.FUSE0xC7_EPRX3CONTROL(EPRX3CONTROL_EPRX33ENABLE_bm | EPRX3CONTROL_
→ EPRX32ENABLE_bm | EPRX3CONTROL_EPRX31ENABLE_bm | EPRX3CONTROL_
→ EPRX30ENABLE_bm |
→ EPORTRXLS_X4<<EPRX3CONTROL_EPRX3DATARATE_of | |
→ EPORTRXMODE_CONTINUOUSTRACKING << EPRX3CONTROL_EPRX3TRACKMODE_of),

.FUSE0xC8_EPRX4CONTROL(EPRX4CONTROL_EPRX43ENABLE_bm | EPRX4CONTROL_
→ EPRX42ENABLE_bm | EPRX4CONTROL_EPRX41ENABLE_bm | EPRX4CONTROL_
→ EPRX40ENABLE_bm |

```

```

    EPORTRXLS_X4<<EPRX4CONTROL_EPRX4DATARATE_of | ↴
    ↵EPORTRXMODE_CONTINUOUSTRACKING << EPRX4CONTROL_EPRX4TRACKMODE_of) ,
    .FUSE0xC9_EPRX5CONTROL (EPRX5CONTROL_EPRX53ENABLE_bm | EPRX5CONTROL_
    ↵EPRX52ENABLE_bm | EPRX5CONTROL_EPRX51ENABLE_bm | EPRX5CONTROL_
    ↵EPRX50ENABLE_bm | EPORTRXLS_X4<<EPRX5CONTROL_EPRX5DATARATE_of | ↴
    ↵EPORTRXMODE_CONTINUOUSTRACKING << EPRX5CONTROL_EPRX5TRACKMODE_of) ,
    .FUSE0xCA_EPRX6CONTROL (EPRX6CONTROL_EPRX63ENABLE_bm | EPRX6CONTROL_
    ↵EPRX62ENABLE_bm | EPRX6CONTROL_EPRX61ENABLE_bm | EPRX6CONTROL_
    ↵EPRX60ENABLE_bm | EPORTRXLS_X4<<EPRX6CONTROL_EPRX6DATARATE_of | ↴
    ↵EPORTRXMODE_CONTINUOUSTRACKING << EPRX6CONTROL_EPRX6TRACKMODE_of) ,
    // configure line driver
    .FUSE0x39_LDCONFIGH ( 1<<LDCONFIGH_LDMODULATIONCURRENT_of) ,
    // configure power up state machine
    .FUSE0xED_POWERUP0 ( POWERUP0_PUSMREADYWHENCHNSLOCKED_bm) ,
    .FUSE0xEF_POWERUP2 ( POWERUP2_DLLCONFIGDONE_bm | POWERUP2_
    ↵PLLCONFIGDONE_bm | POWERUP2_UPDATEENABLE_bm)
) lpGBT (
    // High speed link section
    .HSINN(HSINN),
    .HSINP(HSINP),
    .HSOUTN(HSOUTN),
    .HSOUTP(HSOUTP),
    // ePort Clocks
    .ECLK0N(ECLK0N),
    .ECLK0P(ECLK0P),
    .ECLK1N(ECLK1N),
    .ECLK1P(ECLK1P),
    .ECLK2N(ECLK2N),
    .ECLK2P(ECLK2P),
    .ECLK3N(ECLK3N),
    .ECLK3P(ECLK3P),
    .ECLK4N(ECLK4N),
    .ECLK4P(ECLK4P),
    .ECLK5N(ECLK5N),
    .ECLK5P(ECLK5P),
    .ECLK6N(ECLK6N),
    .ECLK6P(ECLK6P),
    .ECLK7N(ECLK7N),
    .ECLK7P(ECLK7P),
    .ECLK8N(ECLK8N),
    .ECLK8P(ECLK8P),
    .ECLK9N(ECLK9N),
    .ECLK9P(ECLK9P),
    .ECLK10N(ECLK10N),
    .ECLK10P(ECLK10P),
    .ECLK11N(ECLK11N),
    .ECLK11P(ECLK11P),
    .ECLK12N(ECLK12N),
    .ECLK12P(ECLK12P),
    .ECLK13N(ECLK13N),
    .ECLK13P(ECLK13P),

```

```

.ECLK14N (ECLK14N),
.ECLK14P (ECLK14P),
.ECLK15N (ECLK15N),
.ECLK15P (ECLK15P),
.ECLK16N (ECLK16N),
.ECLK16P (ECLK16P),
.ECLK17N (ECLK17N),
.ECLK17P (ECLK17P),
.ECLK18N (ECLK18N),
.ECLK18P (ECLK18P),
.ECLK19N (ECLK19N),
.ECLK19P (ECLK19P),
.ECLK20N (ECLK20N),
.ECLK20P (ECLK20P),
.ECLK21N (ECLK21N),
.ECLK21P (ECLK21P),
.ECLK22N (ECLK22N),
.ECLK22P (ECLK22P),
.ECLK23N (ECLK23N),
.ECLK23P (ECLK23P),
.ECLK24N (ECLK24N),
.ECLK24P (ECLK24P),
.ECLK25N (ECLK25N),
.ECLK25P (ECLK25P),
.ECLK26N (ECLK26N),
.ECLK26P (ECLK26P),
.ECLK27N (ECLK27N),
.ECLK27P (ECLK27P),
.ECLK28N (ECLK28N),
.ECLK28P (ECLK28P),

// phase shifted clocks
.PSCLK0N (PSCLK0N),
.PSCLK0P (PSCLK0P),
.PSCLK1N (PSCLK1N),
.PSCLK1P (PSCLK1P),
.PSCLK2N (PSCLK2N),
.PSCLK2P (PSCLK2P),
.PSCLK3N (PSCLK3N),
.PSCLK3P (PSCLK3P),

// ePort data inputs
.EDIN00N (EDIN00N),
.EDIN00P (EDIN00P),
.EDIN01N (EDIN01N),
.EDIN01P (EDIN01P),
.EDIN02N (EDIN02N),
.EDIN02P (EDIN02P),
.EDIN03N (EDIN03N),
.EDIN03P (EDIN03P),
.EDIN10N (EDIN10N),
.EDIN10P (EDIN10P),
.EDIN11N (EDIN11N),
.EDIN11P (EDIN11P),
.EDIN12N (EDIN12N),
.EDIN12P (EDIN12P),
.EDIN13N (EDIN13N),
.EDIN13P (EDIN13P),

```

```
.EDIN20N(EDIN20N),
.EDIN20P(EDIN20P),
.EDIN21N(EDIN21N),
.EDIN21P(EDIN21P),
.EDIN22N(EDIN22N),
.EDIN22P(EDIN22P),
.EDIN23N(EDIN23N),
.EDIN23P(EDIN23P),
.EDIN30N(EDIN30N),
.EDIN30P(EDIN30P),
.EDIN31N(EDIN31N),
.EDIN31P(EDIN31P),
.EDIN32N(EDIN32N),
.EDIN32P(EDIN32P),
.EDIN33N(EDIN33N),
.EDIN33P(EDIN33P),
.EDIN40N(EDIN40N),
.EDIN40P(EDIN40P),
.EDIN41N(EDIN41N),
.EDIN41P(EDIN41P),
.EDIN42N(EDIN42N),
.EDIN42P(EDIN42P),
.EDIN43N(EDIN43N),
.EDIN43P(EDIN43P),
.EDIN50N(EDIN50N),
.EDIN50P(EDIN50P),
.EDIN51N(EDIN51N),
.EDIN51P(EDIN51P),
.EDIN52N(EDIN52N),
.EDIN52P(EDIN52P),
.EDIN53N(EDIN53N),
.EDIN53P(EDIN53P),
.EDIN60N(EDIN60N),
.EDIN60P(EDIN60P),
.EDIN61N(EDIN61N),
.EDIN61P(EDIN61P),
.EDIN62N(EDIN62N),
.EDIN62P(EDIN62P),
.EDIN63N(EDIN63N),
.EDIN63P(EDIN63P),
.EDINECN(EDINECN),
.EDINECP(EDINECP),  
  
//ePort data outputs
.EDOUT00N(EDOUT00N),
.EDOUT00P(EDOUT00P),
.EDOUT01N(EDOUT01N),
.EDOUT01P(EDOUT01P),
.EDOUT02N(EDOUT02N),
.EDOUT02P(EDOUT02P),
.EDOUT03N(EDOUT03N),
.EDOUT03P(EDOUT03P),
.EDOUT10N(EDOUT10N),
.EDOUT10P(EDOUT10P),
.EDOUT11N(EDOUT11N),
.EDOUT11P(EDOUT11P),
.EDOUT12N(EDOUT12N),
.EDOUT12P(EDOUT12P),
```

```

.EDOUT13N(EDOUT13N),
.EDOUT13P(EDOUT13P),
.EDOUT20N(EDOUT20N),
.EDOUT20P(EDOUT20P),
.EDOUT21N(EDOUT21N),
.EDOUT21P(EDOUT21P),
.EDOUT22N(EDOUT22N),
.EDOUT22P(EDOUT22P),
.EDOUT23N(EDOUT23N),
.EDOUT23P(EDOUT23P),
.EDOUT30N(EDOUT30N),
.EDOUT30P(EDOUT30P),
.EDOUT31N(EDOUT31N),
.EDOUT31P(EDOUT31P),
.EDOUT32N(EDOUT32N),
.EDOUT32P(EDOUT32P),
.EDOUT33N(EDOUT33N),
.EDOUT33P(EDOUT33P),
.EDOUTECN(EDOUTECN),
.EDOUTECP(EDOUTECP),


// power supply section
.GND(GND),
.GNDA(GNDA),
.GNDRX(GNDRX),
.GNDTX(GNDTX),
.VDD1V2(VDD1V2),
.VDDA1V2(VDDA1V2),
.VDDF2V5(VDDF2V5),
.VDDRX1V2(VDDRX1V2),
.VDDTX1V2(VDDTX1V2),


// GPIO section
.GPIO0(GPIO0),
.GPIO1(GPIO1),
.GPIO2(GPIO2),
.GPIO3(GPIO3),
.GPIO4(GPIO4),
.GPIO5(GPIO5),
.GPIO6(GPIO6),
.GPIO7(GPIO7),
.GPIO8(GPIO8),
.GPIO9(GPIO9),
.GPIO10(GPIO10),
.GPIO11(GPIO11),
.GPIO12(GPIO12),
.GPIO13(GPIO13),
.GPIO14(GPIO14),
.GPIO15(GPIO15),


// Configuration and slow control
.LOCKMODE(LOCKMODE),
.MODE0(MODE[0]),
.MODE1(MODE[1]),
.MODE2(MODE[2]),
.MODE3(MODE[3]),
.PORDIS(PORDIS),
.ADR0(ADR[0]),

```

```

    .ADR1(ADR[1]),
    .ADR2(ADR[2]),
    .ADR3(ADR[3]),
    .REFCLKN(REFCLKN),
    .REFCLKP(REFCLKP),
    .READY(READY),
    .RSTB(RSTB),
    .RSTOUTB(RSTOUTB),
    .SC_I2C(SC_I2C),
    .SLSCL(SLSCL),
    .SLSDA(SLSDA),
    .STATEOVRD(STATEOVRD),
    .VCOBYPASS(VCOBYPASS),

    // I2C masters
    .M0SCL(M0SCL),
    .M0SDA(M0SDA),
    .M1SCL(M1SCL),
    .M1SDA(M1SDA),
    .M2SCL(M2SCL),
    .M2SDA(M2SDA),

    // ADC inputs
    .ADC7(ADC7),
    .ADC6(ADC6),
    .ADC5(ADC5),
    .ADC4(ADC4),
    .ADC3(ADC3),
    .ADC2(ADC2),
    .ADC1(ADC1),
    .ADC0(ADC0),
    .VDAC(VDAC),
    .VREF(VREF),

    // test signals
    .TSTCLKINN(TSTCLKINN),
    .TSTCLKINP(TSTCLKINP),
    .TSTOUT0(TSTOUT0),
    .TSTOUT1(TSTOUT1),
    .TSTOUT2(TSTOUT2),
    .TSTOUT3(TSTOUT3),
    .TSTOUT4N(TSTOUT4N),
    .TSTOUT4P(TSTOUT4P),
    .TSTOUT5N(TSTOUT5N),
    .TSTOUT5P(TSTOUT5P),

    .debug_registers(debug_registers),
    .debug_testOutputs(debug_testOutputs)
);

```

There are several points worth pointing out in the presented code. A file `rtl/lpGBTDefines.v` is distributed along with the model files. It contains definitions of all the constants used internally in the chip. It is recommended to use this parameters in order to increase readability and re-usability of the code. A typical convention was adapted for describing register bit fields, where `_bm` implies bit mask and `_of` is used for offset.

As visible in the example above, all parameters corresponding to electrical fuses can be omitted leaving corresponding fuses in default (not blown) state.

16.2 How to get the IpGBT model

The IpGBT is exclusively available in git repository and it is not distributed as an archive. In order to clone the model please use one of the commands below (depending on your authentication method):

```
# for KRB5
$ git clone https://gitlab.cern.ch/lpgbt/lpgbt_model
# for HTTPS
$ git clone https://gitlab.cern.ch/lpgbt/lpgbt_model.git
# for SSH
$ git clone ssh://git@gitlab.cern.ch:7999/lpgbt/lpgbt_model.git
```

The distribution contains only several files:

<pre> └── README.md └── rtl ├── incisive │ └── lpGBT.svp ├── questa │ └── lpGBT.svp └── vcs └── lpGBT.svp └── lpGBTDefines.v └── constants └── vrf └── lpGBT_sim.v └── work └── Makefile </pre>	<ul style="list-style-type: none"> - most up-to-date information about the model - encrypted model for incisive (Cadence) simulator - encrypted model for questa (Mentor) simulator - encrypted model for vcs (Synopsys) simulator - file containing defines of all lpGBT-related - simple simulation test bench - make file showing how to launch various simulators
---	--

16.3 Example how to use IpGBT model

A simple simulation test bench is distributed together with the model. As a prerequisites, user has to download the repository (as outlined in [Section 16.2](#)) and setup simulation tools (irun,vrun, or vcs).

Once both prerequisites are met, users can launch his favorite simulator following simulator-specific instructions below.

16.3.1 Cadence Incisive

The model was generated and tested with **INCISIVE 15.20.038**.

```
$ cd work
$ make incisive-sim
[...]
[ 110615.0] lpGBT is ready
[ 111615.0] Simulation finished
[...]
$ make incisive-sim-gui
```

(some timing violations during initialization are expected)

16.3.2 Mentor Questa

The model was generated and tested with **QUESTA 10.6c-1**.

```
$ cd work
$ make questa-sim
[...]
[ 111540.0] lpGBT is ready
[ 112540.0] Simulation finished
[...]
$ make questa-sim-gui
```

(some timing violations during initialization are expected)

16.3.3 Synopsys VCS

Model was generated and tested with **VCSMX 2017.12-1**.

```
$ cd work
$ make vcs-sim
[...]
[ 110615.0] lpGBT is ready
[ 111615.0] Simulation finished
[...]
```

(some timing violations during initialization are expected)

CHAPTER
SEVENTEEN

PACKAGE

17.1 Mechanical characteristics

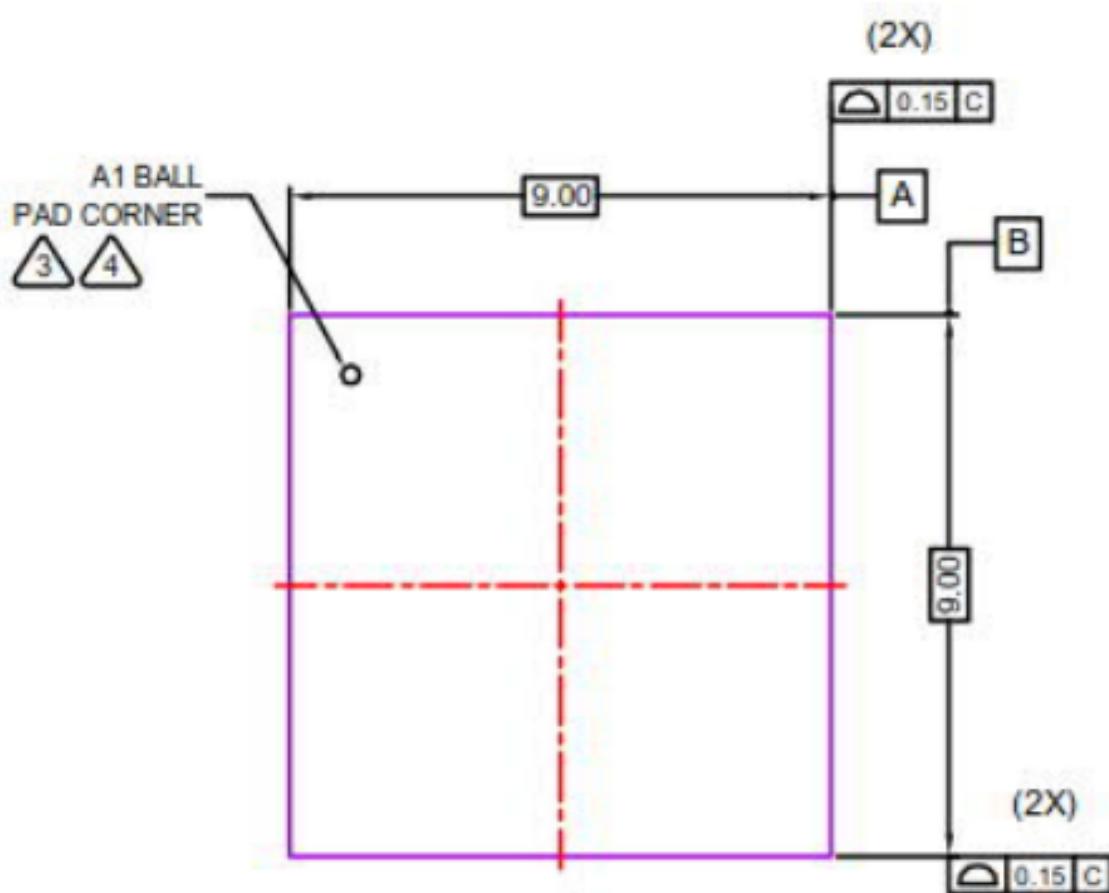


Fig. 17.1: lpGBT package top view

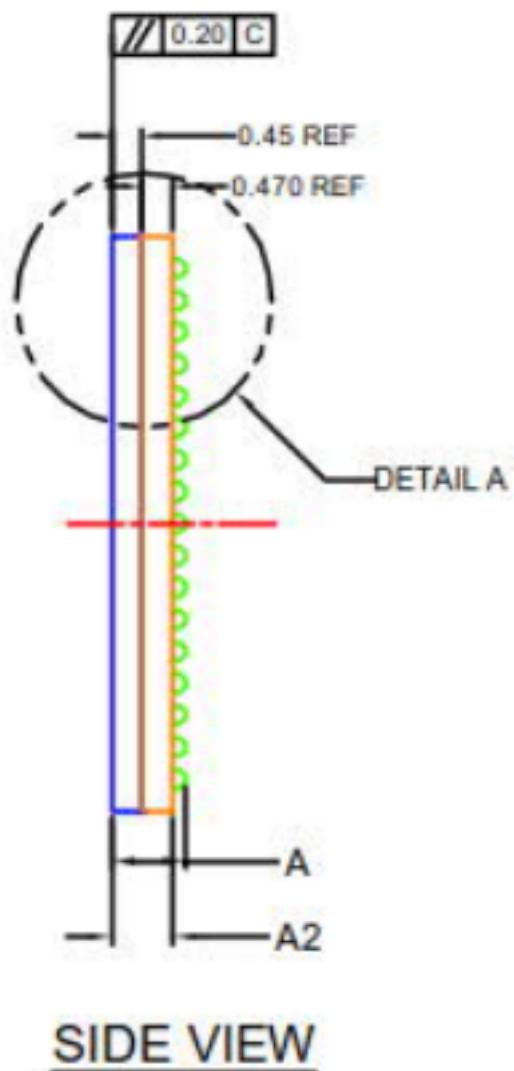


Fig. 17.2: IpGBT package side view

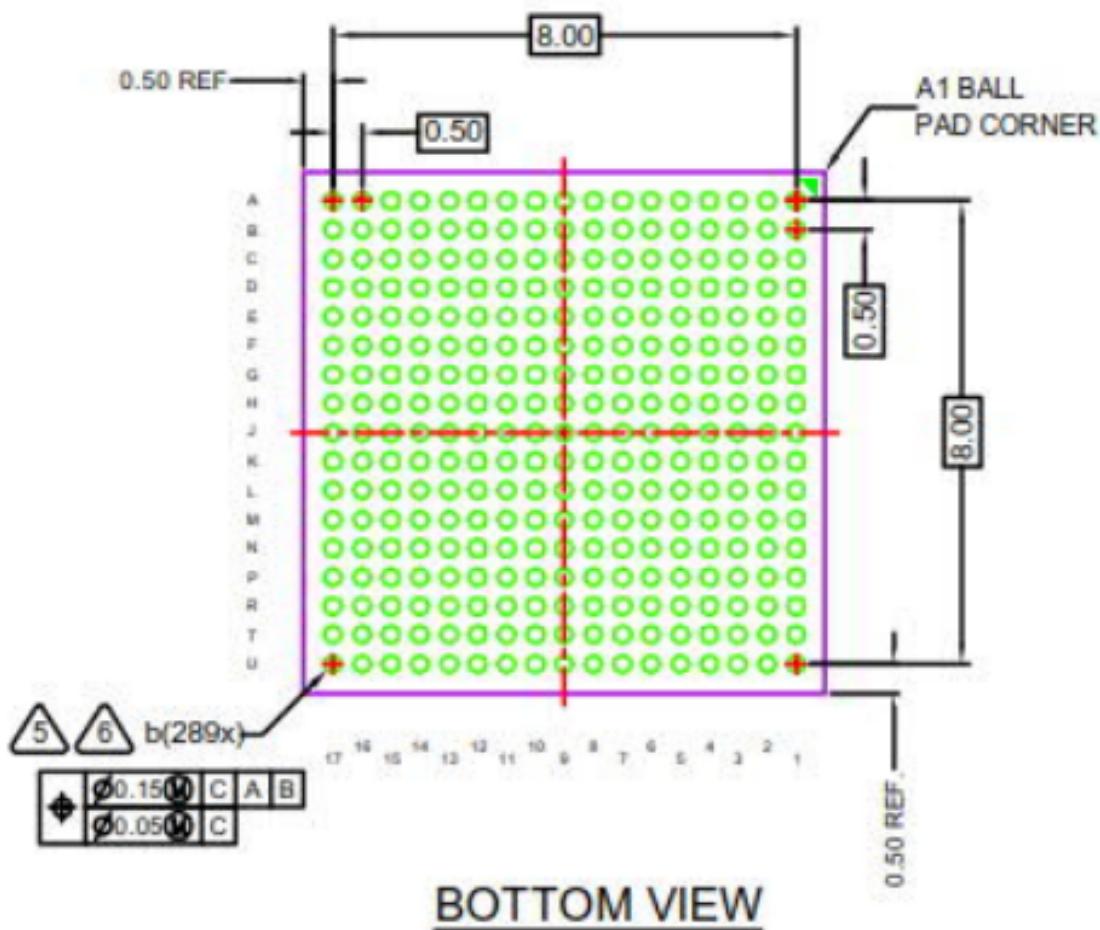


Fig. 17.3: IpGBT package bottom view

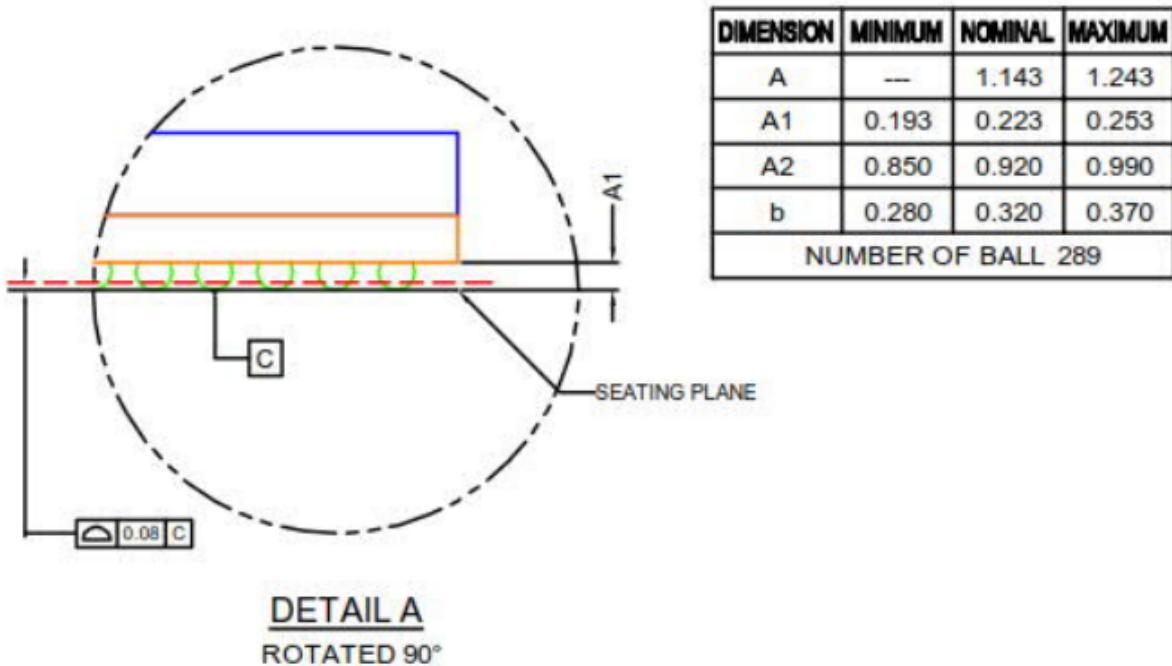


Fig. 17.4: IpGBT package details

17.2 Pinout (top view, balls down)

17.3 Pinout (bottom view, balls up)

17.4 Pin list (by pin designator)

Name	Pin Designator	Electrical Type	More information
ADC1	A1	Passive	<i>Analog to Digital Converter</i> (page 101)
ADC0	A2	Passive	<i>Analog to Digital Converter</i> (page 101)
EDIN63N	A3	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN62N	A4	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN61N	A5	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN60N	A6	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN53N	A7	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN52N	A8	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN51N	A9	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN50N	A10	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN43N	A11	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN42N	A12	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN41N	A13	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN40N	A14	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
ECLK25N	A15	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)

Continued on next page

Table 17.1 – continued from previous page

Name	Pin Designator	Electrical Type	More information
ECLK23P	A16	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK23N	A17	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ADC3	B1	Passive	<i>Analog to Digital Converter</i> (page 101)
ADC2	B2	Passive	<i>Analog to Digital Converter</i> (page 101)
EDIN63P	B3	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN62P	B4	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN61P	B5	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN60P	B6	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN53P	B7	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN52P	B8	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN51P	B9	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN50P	B10	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN43P	B11	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN42P	B12	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN41P	B13	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN40P	B14	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
ECLK25P	B15	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK22P	B16	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK22N	B17	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ADC6	C1	Passive	<i>Analog to Digital Converter</i> (page 101)
ADC5	C2	Passive	<i>Analog to Digital Converter</i> (page 101)
ADC4	C3	Passive	<i>Analog to Digital Converter</i> (page 101)
ECLK27N	C4	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK26N	C5	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT33N	C6	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT32N	C7	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT31N	C8	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT30N	C9	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT23N	C10	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT22N	C11	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT21N	C12	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT20N	C13	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK24P	C14	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK24N	C15	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK21P	C16	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK21N	C17	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ADC7	D1	Passive	<i>Analog to Digital Converter</i> (page 101)
VDAC	D2	Output	<i>Voltage Digital to Analog Converter</i> (page 106)
RSTB	D3	Input	<i>RSTB</i> (page 67), <i>CMOS I/O Pin Characteristics</i> (page 310)
ECLK27P	D4	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK26P	D5	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT33P	D6	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT32P	D7	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT31P	D8	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT30P	D9	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT23P	D10	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT22P	D11	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT21P	D12	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT20P	D13	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)

Continued on next page

Table 17.1 – continued from previous page

Name	Pin Designator	Electrical Type	More information
ECLK20P	D14	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK20N	D15	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK19P	D16	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK19N	D17	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
LOCKMODE	E1	Input	<i>LOCKMODE</i> (page 75)
STATEOVRD	E2	Input	<i>STATEOVRD</i> (page 67), <i>CMOS I/O Pin Characteristics</i> (page 310)
VDDA1V2	E3	Power	<i>General Operating Ratings</i> (page 309)
VDDA1V2	E4	Power	<i>General Operating Ratings</i> (page 309)
M2SCL	E5	Output	<i>I₂C Masters</i> (page 85)
M1SCL	E6	Output	<i>I₂C Masters</i> (page 85)
M0SCL	E7	Output	<i>I₂C Masters</i> (page 85)
ADR2	E8	Input	
ADR0	E9	Input	
TSTOUT5N	E10	Output	<i>Test outputs</i> (page 121)
TSTOUT5P	E11	Output	<i>Test outputs</i> (page 121)
TSTOUT4N	E12	Output	<i>Test outputs</i> (page 121)
TSTOUT4P	E13	Output	<i>Test outputs</i> (page 121)
ECLK18P	E14	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK18N	E15	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDIN33P	E16	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN33N	E17	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
GNDTX	F1	Power	<i>General Operating Ratings</i> (page 309)
GNDTX	F2	Power	<i>General Operating Ratings</i> (page 309)
VDDTX1V2	F3	Power	<i>General Operating Ratings</i> (page 309)
GNDA	F4	Power	<i>General Operating Ratings</i> (page 309)
M2SDA	F5	I/O	<i>I₂C Masters</i> (page 85)
M1SDA	F6	I/O	<i>I₂C Masters</i> (page 85)
M0SDA	F7	I/O	<i>I₂C Masters</i> (page 85)
ADR3	F8	Input	
ADR1	F9	Input	
TSTOUT3	F10	Output	<i>Test outputs</i> (page 121)
TSTOUT2	F11	Output	<i>Test outputs</i> (page 121)
TSTOUT1	F12	Output	<i>Test outputs</i> (page 121)
TSTOUT0	F13	Output	<i>Test outputs</i> (page 121)
ECLK17P	F14	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK17N	F15	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDIN32P	F16	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN32N	F17	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
HSOUTN	G1	Output	<i>High-Speed Line Driver</i> (page 39)
GNDTX	G2	Power	<i>General Operating Ratings</i> (page 309)
VDDTX1V2	G3	Power	<i>General Operating Ratings</i> (page 309)
GNDA	G4	Power	<i>General Operating Ratings</i> (page 309)
MODE0	G5	Input	<i>MODE3, MODE2, MODE1, MODE0</i> (page 17)
GND	G6	Power	<i>General Operating Ratings</i> (page 309)
GND	G7	Power	<i>General Operating Ratings</i> (page 309)
GND	G8	Power	<i>General Operating Ratings</i> (page 309)
GND	G9	Power	<i>General Operating Ratings</i> (page 309)
GND	G10	Power	<i>General Operating Ratings</i> (page 309)
GND	G11	Power	<i>General Operating Ratings</i> (page 309)

Continued on next page

Table 17.1 – continued from previous page

Name	Pin Designator	Electrical Type	More information
GND	G12	Power	<i>General Operating Ratings</i> (page 309)
GND	G13	Power	<i>General Operating Ratings</i> (page 309)
ECLK16P	G14	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK16N	G15	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDIN31P	G16	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN31N	G17	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
HSOUTP	H1	Output	<i>High-Speed Line Driver</i> (page 39)
GNDTX	H2	Power	<i>General Operating Ratings</i> (page 309)
VDDTX1V2	H3	Power	<i>General Operating Ratings</i> (page 309)
VREF	H4	Power	<i>Reference voltage</i> (page 105)
GND	H5	Power	<i>General Operating Ratings</i> (page 309)
GND	H6	Power	<i>General Operating Ratings</i> (page 309)
GND	H7	Power	<i>General Operating Ratings</i> (page 309)
GND	H8	Power	<i>General Operating Ratings</i> (page 309)
GND	H9	Power	<i>General Operating Ratings</i> (page 309)
GND	H10	Power	<i>General Operating Ratings</i> (page 309)
GND	H11	Power	<i>General Operating Ratings</i> (page 309)
GND	H12	Power	<i>General Operating Ratings</i> (page 309)
GND	H13	Power	<i>General Operating Ratings</i> (page 309)
ECLK15P	H14	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK15N	H15	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDIN30P	H16	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN30N	H17	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
GNDTX	J1	Power	<i>General Operating Ratings</i> (page 309)
GNDTX	J2	Power	<i>General Operating Ratings</i> (page 309)
VDDTX1V2	J3	Power	<i>General Operating Ratings</i> (page 309)
VREF	J4	Power	<i>Reference voltage</i> (page 105)
VDD1V2	J5	Power	<i>General Operating Ratings</i> (page 309)
VDD1V2	J6	Power	<i>General Operating Ratings</i> (page 309)
VDD1V2	J7	Power	<i>General Operating Ratings</i> (page 309)
VDD1V2	J8	Power	<i>General Operating Ratings</i> (page 309)
VDD1V2	J9	Power	<i>General Operating Ratings</i> (page 309)
VDDF2V5	J10	Power	<i>General Operating Ratings</i> (page 309)
VDD1V2	J11	Power	<i>General Operating Ratings</i> (page 309)
GND	J12	Power	<i>General Operating Ratings</i> (page 309)
GND	J13	Power	<i>General Operating Ratings</i> (page 309)
ECLK14P	J14	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK14N	J15	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK12P	J16	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK12N	J17	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
GNDRX	K1	Power	<i>General Operating Ratings</i> (page 309)
GNDRX	K2	Power	<i>General Operating Ratings</i> (page 309)
VDDRX1V2	K3	Power	<i>General Operating Ratings</i> (page 309)
TSTCLKINP	K4	Input	<i>TSTCLKINP</i> and <i>TSTCLKINN</i> (page 75)
MODE1	K5	Input	<i>MODE3, MODE2, MODE1, MODE0</i> (page 17)
VDD1V2	K6	Power	<i>General Operating Ratings</i> (page 309)
VDD1V2	K7	Power	<i>General Operating Ratings</i> (page 309)
VDD1V2	K8	Power	<i>General Operating Ratings</i> (page 309)
VDD1V2	K9	Power	<i>General Operating Ratings</i> (page 309)

Continued on next page

Table 17.1 – continued from previous page

Name	Pin Designator	Electrical Type	More information
VDDF2V5	K10	Power	<i>General Operating Ratings</i> (page 309)
VDD1V2	K11	Power	<i>General Operating Ratings</i> (page 309)
VDD1V2	K12	Power	<i>General Operating Ratings</i> (page 309)
VDD1V2	K13	Power	<i>General Operating Ratings</i> (page 309)
ECLK13P	K14	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK13N	K15	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK11P	K16	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK11N	K17	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
HSINP	L1	Input	<i>High-Speed Equalizer</i> (page 43)
GNDRX	L2	Power	<i>General Operating Ratings</i> (page 309)
VDDRX1V2	L3	Power	<i>General Operating Ratings</i> (page 309)
TSTCLKINN	L4	Input	<i>TSTCLKINP and TSTCLKINN</i> (page 75)
GPIO14	L5	I/O	<i>General Purpose I/O</i> (page 81), <i>CMOS I/O Pin Characteristics</i> (page 310)
GPIO13	L6	I/O	<i>General Purpose I/O</i> (page 81), <i>CMOS I/O Pin Characteristics</i> (page 310)
GPIO12	L7	I/O	<i>General Purpose I/O</i> (page 81), <i>CMOS I/O Pin Characteristics</i> (page 310)
GPIO9	L8	I/O	<i>General Purpose I/O</i> (page 81), <i>CMOS I/O Pin Characteristics</i> (page 310)
GPIO6	L9	I/O	<i>General Purpose I/O</i> (page 81), <i>CMOS I/O Pin Characteristics</i> (page 310)
VDD1V2	L10	Power	<i>General Operating Ratings</i> (page 309)
VDD1V2	L11	Power	<i>General Operating Ratings</i> (page 309)
VDD1V2	L12	Power	<i>General Operating Ratings</i> (page 309)
VDD1V2	L13	Power	<i>General Operating Ratings</i> (page 309)
ECLK10P	L14	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK10N	L15	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDIN23P	L16	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN23N	L17	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
HSINN	M1	Input	<i>High-Speed Equalizer</i> (page 43)
GNDRX	M2	Power	<i>General Operating Ratings</i> (page 309)
VDDRX1V2	M3	Power	<i>General Operating Ratings</i> (page 309)
PSCLK0P	M4	Output	<i>Phase programmable clocks</i> (page 79)
PSCLK1P	M5	Output	<i>Phase programmable clocks</i> (page 79)
PSCLK2P	M6	Output	<i>Phase programmable clocks</i> (page 79)
PSCLK3P	M7	Output	<i>Phase programmable clocks</i> (page 79)
GPIO10	M8	I/O	<i>General Purpose I/O</i> (page 81), <i>CMOS I/O Pin Characteristics</i> (page 310)
GPIO7	M9	I/O	<i>General Purpose I/O</i> (page 81), <i>CMOS I/O Pin Characteristics</i> (page 310)
GPIO4	M10	I/O	<i>General Purpose I/O</i> (page 81), <i>CMOS I/O Pin Characteristics</i> (page 310)
GPIO2	M11	I/O	<i>General Purpose I/O</i> (page 81), <i>CMOS I/O Pin Characteristics</i> (page 310)
VDD1V2	M12	Power	<i>General Operating Ratings</i> (page 309)
VDD1V2	M13	Power	<i>General Operating Ratings</i> (page 309)
ECLK9P	M14	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK9N	M15	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDIN22P	M16	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN22N	M17	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
GNDRX	N1	Power	<i>General Operating Ratings</i> (page 309)
GNDRX	N2	Power	<i>General Operating Ratings</i> (page 309)
VDDRX1V2	N3	Power	<i>General Operating Ratings</i> (page 309)
PSCLK0N	N4	Output	<i>Phase programmable clocks</i> (page 79)
PSCLK1N	N5	Output	<i>Phase programmable clocks</i> (page 79)
PSCLK2N	N6	Output	<i>Phase programmable clocks</i> (page 79)
PSCLK3N	N7	Output	<i>Phase programmable clocks</i> (page 79)

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Table 17.1 – continued from previous page

Name	Pin Designator	Electrical Type	More information
GPIO11	N8	I/O	<i>General Purpose I/O</i> (page 81), <i>CMOS I/O Pin Characteristics</i> (page 310)
GPIO8	N9	I/O	<i>General Purpose I/O</i> (page 81), <i>CMOS I/O Pin Characteristics</i> (page 310)
GPIO5	N10	I/O	<i>General Purpose I/O</i> (page 81), <i>CMOS I/O Pin Characteristics</i> (page 310)
GPIO3	N11	I/O	<i>General Purpose I/O</i> (page 81), <i>CMOS I/O Pin Characteristics</i> (page 310)
GPIO1	N12	I/O	<i>General Purpose I/O</i> (page 81), <i>CMOS I/O Pin Characteristics</i> (page 310)
GPIO0	N13	I/O	<i>General Purpose I/O</i> (page 81), <i>CMOS I/O Pin Characteristics</i> (page 310)
ECLK8P	N14	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK8N	N15	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDIN21P	N16	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN21N	N17	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
REFCLKN	P1	Input	<i>REFCLKP and REFCLKN</i> (page 75)
MODE3	P2	Input	<i>MODE3, MODE2, MODE1, MODE0</i> (page 17)
MODE2	P3	Input	<i>MODE3, MODE2, MODE1, MODE0</i> (page 17)
ECLK1P	P4	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT00P	P5	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT01P	P6	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT02P	P7	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT03P	P8	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT10P	P9	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT11P	P10	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT12P	P11	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT13P	P12	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK4P	P13	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK5P	P14	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK7P	P15	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDIN20P	P16	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN20N	P17	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
REFCLKP	R1	Input	<i>REFCLKP and REFCLKN</i> (page 75)
READY	R2	Output	<i>READY</i> (page 67), <i>CMOS I/O Pin Characteristics</i> (page 310)
GPIO15	R3	I/O	<i>General Purpose I/O</i> (page 81), <i>CMOS I/O Pin Characteristics</i> (page 310)
ECLK1N	R4	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT00N	R5	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT01N	R6	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT02N	R7	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT03N	R8	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT10N	R9	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT11N	R10	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT12N	R11	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT13N	R12	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK4N	R13	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK5N	R14	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK7N	R15	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUTECP	R16	Output	<i>Electrical links</i> (page 47)
EDOUTECN	R17	Output	<i>Electrical links</i> (page 47)
PORDIS	T1	Input	<i>PORDIS</i> (page 67)
RSTOUTB	T2	Output	<i>RSTOUTB</i> (page 67), <i>CMOS I/O Pin Characteristics</i> (page 310)
SC_I2C	T3	Input	<i>SC_I2C</i> (page 18)
ECLK0P	T4	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDIN00P	T5	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)

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Table 17.1 – continued from previous page

Name	Pin Designator	Electrical Type	More information
EDIN01P	T6	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN02P	T7	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN03P	T8	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
ECLK2P	T9	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK3P	T10	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDIN10P	T11	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN11P	T12	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN12P	T13	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN13P	T14	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
ECLK6P	T15	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK28P	T16	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDINECP	T17	Input	<i>Electrical links</i> (page 47)
VCOBYPASS	U1	Input	<i>VCOBYPASS</i> (page 76)
SLSCL	U2	Input	<i>I₂C slave interface</i> (page 22)
SLSDA	U3	I/O	<i>I₂C slave interface</i> (page 22)
ECLK0N	U4	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDIN00N	U5	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN01N	U6	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN02N	U7	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN03N	U8	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
ECLK2N	U9	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK3N	U10	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDIN10N	U11	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN11N	U12	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN12N	U13	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN13N	U14	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
ECLK6N	U15	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK28N	U16	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDINECN	U17	Input	<i>Electrical links</i> (page 47)

17.5 Pin list (by pin name)

Name	Pin Designator	Electrical Type	More information
ADC0	A2	Passive	<i>Analog to Digital Converter</i> (page 101)
ADC1	A1	Passive	<i>Analog to Digital Converter</i> (page 101)
ADC2	B2	Passive	<i>Analog to Digital Converter</i> (page 101)
ADC3	B1	Passive	<i>Analog to Digital Converter</i> (page 101)
ADC4	C3	Passive	<i>Analog to Digital Converter</i> (page 101)
ADC5	C2	Passive	<i>Analog to Digital Converter</i> (page 101)
ADC6	C1	Passive	<i>Analog to Digital Converter</i> (page 101)
ADC7	D1	Passive	<i>Analog to Digital Converter</i> (page 101)
ADR0	E9	Input	
ADR1	F9	Input	
ADR2	E8	Input	
ADR3	F8	Input	

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Table 17.2 – continued from previous page

Name	Pin Designator	Electrical Type	More information
ECLK0N	U4	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK0P	T4	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK10N	L15	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK10P	L14	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK11N	K17	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK11P	K16	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK12N	J17	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK12P	J16	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK13N	K15	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK13P	K14	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK14N	J15	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK14P	J14	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK15N	H15	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK15P	H14	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK16N	G15	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK16P	G14	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK17N	F15	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK17P	F14	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK18N	E15	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK18P	E14	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK19N	D17	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK19P	D16	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK1N	R4	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK1P	P4	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK20N	D15	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK20P	D14	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK21N	C17	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK21P	C16	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK22N	B17	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK22P	B16	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK23N	A17	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK23P	A16	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK24N	C15	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK24P	C14	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK25N	A15	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK25P	B15	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK26N	C5	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK26P	D5	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK27N	C4	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK27P	D4	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK28N	U16	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK28P	T16	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK2N	U9	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK2P	T9	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK3N	U10	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK3P	T10	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK4N	R13	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK4P	P13	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK5N	R14	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)

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Table 17.2 – continued from previous page

Name	Pin Designator	Electrical Type	More information
ECLK5P	P14	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK6N	U15	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK6P	T15	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK7N	R15	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK7P	P15	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK8N	N15	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK8P	N14	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK9N	M15	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
ECLK9P	M14	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDIN00N	U5	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN00P	T5	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN01N	U6	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN01P	T6	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN02N	U7	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN02P	T7	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN03N	U8	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN03P	T8	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN10N	U11	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN10P	T11	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN11N	U12	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN11P	T12	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN12N	U13	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN12P	T13	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN13N	U14	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN13P	T14	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN20N	P17	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN20P	P16	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN21N	N17	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN21P	N16	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN22N	M17	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN22P	M16	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN23N	L17	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN23P	L16	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN30N	H17	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN30P	H16	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN31N	G17	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN31P	G16	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN32N	F17	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN32P	F16	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN33N	E17	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN33P	E16	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN40N	A14	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN40P	B14	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN41N	A13	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN41P	B13	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN42N	A12	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN42P	B12	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN43N	A11	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN43P	B11	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)

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Table 17.2 – continued from previous page

Name	Pin Designator	Electrical Type	More information
EDIN50N	A10	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN50P	B10	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN51N	A9	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN51P	B9	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN52N	A8	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN52P	B8	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN53N	A7	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN53P	B7	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN60N	A6	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN60P	B6	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN61N	A5	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN61P	B5	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN62N	A4	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN62P	B4	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN63N	A3	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDIN63P	B3	Input	<i>Electrical links</i> (page 47), <i>eRX differential receiver</i> (page 310)
EDINECN	U17	Input	<i>Electrical links</i> (page 47)
EDINECP	T17	Input	<i>Electrical links</i> (page 47)
EDOUT00N	R5	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT00P	P5	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT01N	R6	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT01P	P6	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT02N	R7	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT02P	P7	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT03N	R8	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT03P	P8	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT10N	R9	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT10P	P9	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT11N	R10	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT11P	P10	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT12N	R11	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT12P	P11	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT13N	R12	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT13P	P12	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT20N	C13	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT20P	D13	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT21N	C12	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT21P	D12	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT22N	C11	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT22P	D11	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT23N	C10	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT23P	D10	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT30N	C9	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT30P	D9	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT31N	C8	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT31P	D8	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT32N	C7	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT32P	D7	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUT33N	C6	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)

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Table 17.2 – continued from previous page

Name	Pin Designator	Electrical Type	More information
EDOUT33P	D6	Output	<i>Electrical links</i> (page 47), <i>eTX differential driver</i> (page 311)
EDOUTECN	R17	Output	<i>Electrical links</i> (page 47)
EDOUTECP	R16	Output	<i>Electrical links</i> (page 47)
GND	G6	Power	<i>General Operating Ratings</i> (page 309)
GND	G7	Power	<i>General Operating Ratings</i> (page 309)
GND	G8	Power	<i>General Operating Ratings</i> (page 309)
GND	G9	Power	<i>General Operating Ratings</i> (page 309)
GND	G10	Power	<i>General Operating Ratings</i> (page 309)
GND	G11	Power	<i>General Operating Ratings</i> (page 309)
GND	G12	Power	<i>General Operating Ratings</i> (page 309)
GND	G13	Power	<i>General Operating Ratings</i> (page 309)
GND	H5	Power	<i>General Operating Ratings</i> (page 309)
GND	H6	Power	<i>General Operating Ratings</i> (page 309)
GND	H7	Power	<i>General Operating Ratings</i> (page 309)
GND	H8	Power	<i>General Operating Ratings</i> (page 309)
GND	H9	Power	<i>General Operating Ratings</i> (page 309)
GND	H10	Power	<i>General Operating Ratings</i> (page 309)
GND	H11	Power	<i>General Operating Ratings</i> (page 309)
GND	H12	Power	<i>General Operating Ratings</i> (page 309)
GND	H13	Power	<i>General Operating Ratings</i> (page 309)
GND	J12	Power	<i>General Operating Ratings</i> (page 309)
GND	J13	Power	<i>General Operating Ratings</i> (page 309)
GNDA	F4	Power	<i>General Operating Ratings</i> (page 309)
GNDA	G4	Power	<i>General Operating Ratings</i> (page 309)
GNDRX	K1	Power	<i>General Operating Ratings</i> (page 309)
GNDRX	K2	Power	<i>General Operating Ratings</i> (page 309)
GNDRX	L2	Power	<i>General Operating Ratings</i> (page 309)
GNDRX	M2	Power	<i>General Operating Ratings</i> (page 309)
GNDRX	N1	Power	<i>General Operating Ratings</i> (page 309)
GNDRX	N2	Power	<i>General Operating Ratings</i> (page 309)
GNDTX	F1	Power	<i>General Operating Ratings</i> (page 309)
GNDTX	F2	Power	<i>General Operating Ratings</i> (page 309)
GNDTX	G2	Power	<i>General Operating Ratings</i> (page 309)
GNDTX	H2	Power	<i>General Operating Ratings</i> (page 309)
GNDTX	J1	Power	<i>General Operating Ratings</i> (page 309)
GNDTX	J2	Power	<i>General Operating Ratings</i> (page 309)
GPIO0	N13	I/O	<i>General Purpose I/O</i> (page 81), <i>CMOS I/O Pin Characteristics</i> (page 310)
GPIO1	N12	I/O	<i>General Purpose I/O</i> (page 81), <i>CMOS I/O Pin Characteristics</i> (page 310)
GPIO10	M8	I/O	<i>General Purpose I/O</i> (page 81), <i>CMOS I/O Pin Characteristics</i> (page 310)
GPIO11	N8	I/O	<i>General Purpose I/O</i> (page 81), <i>CMOS I/O Pin Characteristics</i> (page 310)
GPIO12	L7	I/O	<i>General Purpose I/O</i> (page 81), <i>CMOS I/O Pin Characteristics</i> (page 310)
GPIO13	L6	I/O	<i>General Purpose I/O</i> (page 81), <i>CMOS I/O Pin Characteristics</i> (page 310)
GPIO14	L5	I/O	<i>General Purpose I/O</i> (page 81), <i>CMOS I/O Pin Characteristics</i> (page 310)
GPIO15	R3	I/O	<i>General Purpose I/O</i> (page 81), <i>CMOS I/O Pin Characteristics</i> (page 310)
GPIO2	M11	I/O	<i>General Purpose I/O</i> (page 81), <i>CMOS I/O Pin Characteristics</i> (page 310)
GPIO3	N11	I/O	<i>General Purpose I/O</i> (page 81), <i>CMOS I/O Pin Characteristics</i> (page 310)
GPIO4	M10	I/O	<i>General Purpose I/O</i> (page 81), <i>CMOS I/O Pin Characteristics</i> (page 310)
GPIO5	N10	I/O	<i>General Purpose I/O</i> (page 81), <i>CMOS I/O Pin Characteristics</i> (page 310)
GPIO6	L9	I/O	<i>General Purpose I/O</i> (page 81), <i>CMOS I/O Pin Characteristics</i> (page 310)

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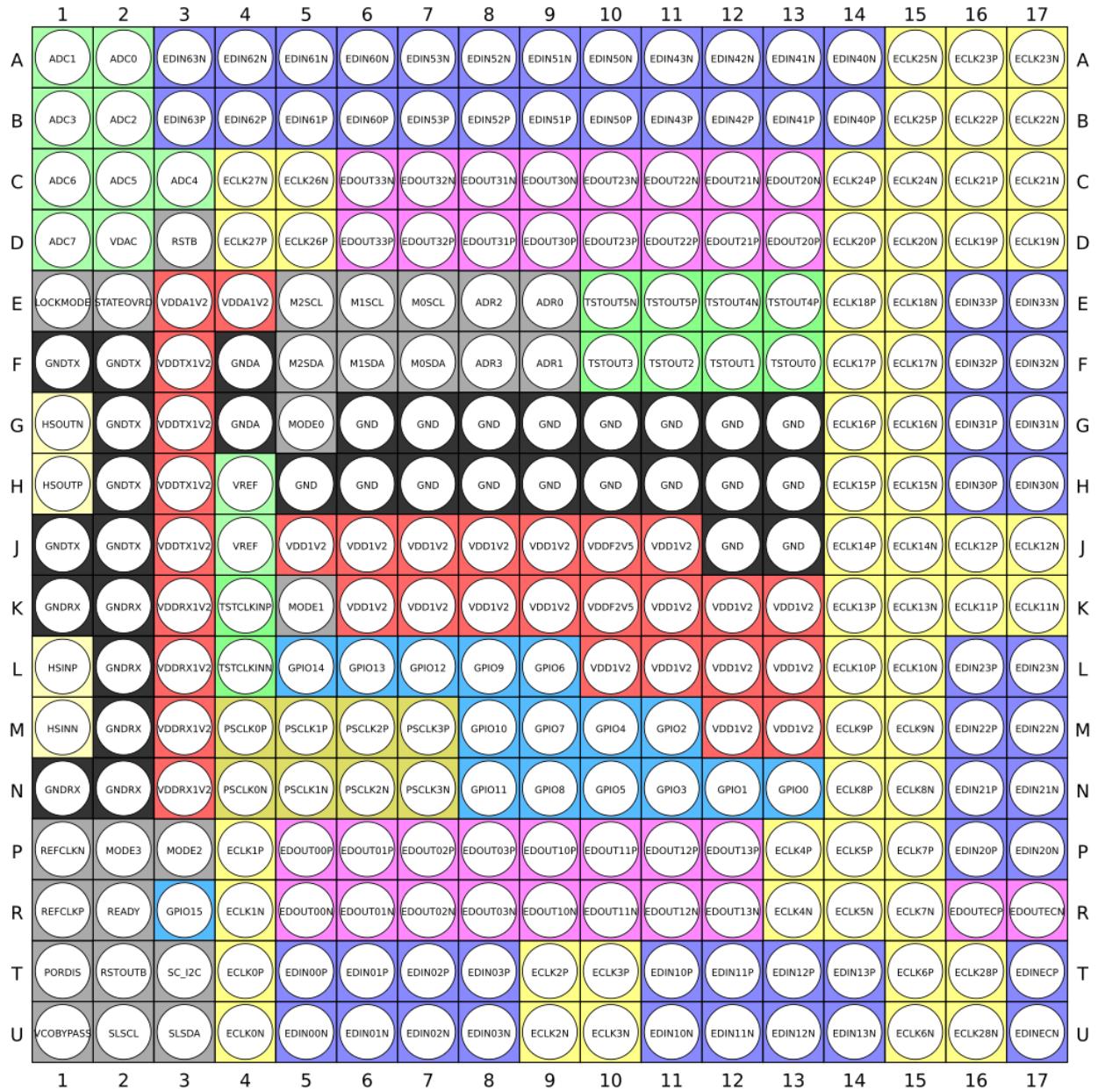
Table 17.2 – continued from previous page

Name	Pin Designator	Electrical Type	More information
GPIO7	M9	I/O	<i>General Purpose I/O</i> (page 81), <i>CMOS I/O Pin Characteristics</i> (page 310)
GPIO8	N9	I/O	<i>General Purpose I/O</i> (page 81), <i>CMOS I/O Pin Characteristics</i> (page 310)
GPIO9	L8	I/O	<i>General Purpose I/O</i> (page 81), <i>CMOS I/O Pin Characteristics</i> (page 310)
HSINN	M1	Input	<i>High-Speed Equalizer</i> (page 43)
HSINP	L1	Input	<i>High-Speed Equalizer</i> (page 43)
HSOUTN	G1	Output	<i>High-Speed Line Driver</i> (page 39)
HSOUTP	H1	Output	<i>High-Speed Line Driver</i> (page 39)
LOCKMODE	E1	Input	<i>LOCKMODE</i> (page 75)
M0SCL	E7	Output	<i>I2C Masters</i> (page 85)
M0SDA	F7	I/O	<i>I2C Masters</i> (page 85)
M1SCL	E6	Output	<i>I2C Masters</i> (page 85)
M1SDA	F6	I/O	<i>I2C Masters</i> (page 85)
M2SCL	E5	Output	<i>I2C Masters</i> (page 85)
M2SDA	F5	I/O	<i>I2C Masters</i> (page 85)
MODE0	G5	Input	<i>MODE3, MODE2, MODE1, MODE0</i> (page 17)
MODE1	K5	Input	<i>MODE3, MODE2, MODE1, MODE0</i> (page 17)
MODE2	P3	Input	<i>MODE3, MODE2, MODE1, MODE0</i> (page 17)
MODE3	P2	Input	<i>MODE3, MODE2, MODE1, MODE0</i> (page 17)
PORDIS	T1	Input	<i>PORDIS</i> (page 67)
PSCLK0N	N4	Output	<i>Phase programmable clocks</i> (page 79)
PSCLK0P	M4	Output	<i>Phase programmable clocks</i> (page 79)
PSCLK1N	N5	Output	<i>Phase programmable clocks</i> (page 79)
PSCLK1P	M5	Output	<i>Phase programmable clocks</i> (page 79)
PSCLK2N	N6	Output	<i>Phase programmable clocks</i> (page 79)
PSCLK2P	M6	Output	<i>Phase programmable clocks</i> (page 79)
PSCLK3N	N7	Output	<i>Phase programmable clocks</i> (page 79)
PSCLK3P	M7	Output	<i>Phase programmable clocks</i> (page 79)
READY	R2	Output	<i>READY</i> (page 67), <i>CMOS I/O Pin Characteristics</i> (page 310)
REFCLKN	P1	Input	<i>REFCLKP and REFCLKN</i> (page 75)
REFCLKP	R1	Input	<i>REFCLKP and REFCLKN</i> (page 75)
RSTB	D3	Input	<i>RSTB</i> (page 67), <i>CMOS I/O Pin Characteristics</i> (page 310)
RSTOUTB	T2	Output	<i>RSTOUTB</i> (page 67), <i>CMOS I/O Pin Characteristics</i> (page 310)
SC_I2C	T3	Input	<i>SC_I2C</i> (page 18)
SLSCL	U2	Input	<i>I2C slave interface</i> (page 22)
SLSDA	U3	I/O	<i>I2C slave interface</i> (page 22)
STATEOVRD	E2	Input	<i>STATEOVRD</i> (page 67), <i>CMOS I/O Pin Characteristics</i> (page 310)
TSTCLKINN	L4	Input	<i>TSTCLKINP and TSTCLKINN</i> (page 75)
TSTCLKINP	K4	Input	<i>TSTCLKINP and TSTCLKINN</i> (page 75)
TSTOUT0	F13	Output	<i>Test outputs</i> (page 121)
TSTOUT1	F12	Output	<i>Test outputs</i> (page 121)
TSTOUT2	F11	Output	<i>Test outputs</i> (page 121)
TSTOUT3	F10	Output	<i>Test outputs</i> (page 121)
TSTOUT4N	E12	Output	<i>Test outputs</i> (page 121)
TSTOUT4P	E13	Output	<i>Test outputs</i> (page 121)
TSTOUT5N	E10	Output	<i>Test outputs</i> (page 121)
TSTOUT5P	E11	Output	<i>Test outputs</i> (page 121)
VCOBYPASS	U1	Input	<i>VCOBYPASS</i> (page 76)
VDAC	D2	Output	<i>Voltage Digital to Analog Converter</i> (page 106)
VDD1V2	J5	Power	<i>General Operating Ratings</i> (page 309)

Continued on next page

Table 17.2 – continued from previous page

Name	Pin Designator	Electrical Type	More information
VDD1V2	J6	Power	<i>General Operating Ratings</i> (page 309)
VDD1V2	J7	Power	<i>General Operating Ratings</i> (page 309)
VDD1V2	J8	Power	<i>General Operating Ratings</i> (page 309)
VDD1V2	J9	Power	<i>General Operating Ratings</i> (page 309)
VDD1V2	J11	Power	<i>General Operating Ratings</i> (page 309)
VDD1V2	K6	Power	<i>General Operating Ratings</i> (page 309)
VDD1V2	K7	Power	<i>General Operating Ratings</i> (page 309)
VDD1V2	K8	Power	<i>General Operating Ratings</i> (page 309)
VDD1V2	K9	Power	<i>General Operating Ratings</i> (page 309)
VDD1V2	K11	Power	<i>General Operating Ratings</i> (page 309)
VDD1V2	K12	Power	<i>General Operating Ratings</i> (page 309)
VDD1V2	K13	Power	<i>General Operating Ratings</i> (page 309)
VDD1V2	L10	Power	<i>General Operating Ratings</i> (page 309)
VDD1V2	L11	Power	<i>General Operating Ratings</i> (page 309)
VDD1V2	L12	Power	<i>General Operating Ratings</i> (page 309)
VDD1V2	L13	Power	<i>General Operating Ratings</i> (page 309)
VDD1V2	M12	Power	<i>General Operating Ratings</i> (page 309)
VDD1V2	M13	Power	<i>General Operating Ratings</i> (page 309)
VDDA1V2	E3	Power	<i>General Operating Ratings</i> (page 309)
VDDA1V2	E4	Power	<i>General Operating Ratings</i> (page 309)
VDDF2V5	J10	Power	<i>General Operating Ratings</i> (page 309)
VDDF2V5	K10	Power	<i>General Operating Ratings</i> (page 309)
VDDRX1V2	K3	Power	<i>General Operating Ratings</i> (page 309)
VDDRX1V2	L3	Power	<i>General Operating Ratings</i> (page 309)
VDDRX1V2	M3	Power	<i>General Operating Ratings</i> (page 309)
VDDRX1V2	N3	Power	<i>General Operating Ratings</i> (page 309)
VDDTX1V2	F3	Power	<i>General Operating Ratings</i> (page 309)
VDDTX1V2	G3	Power	<i>General Operating Ratings</i> (page 309)
VDDTX1V2	H3	Power	<i>General Operating Ratings</i> (page 309)
VDDTX1V2	J3	Power	<i>General Operating Ratings</i> (page 309)
VREF	H4	Power	<i>Reference voltage</i> (page 105)
VREF	J4	Power	<i>Reference voltage</i> (page 105)



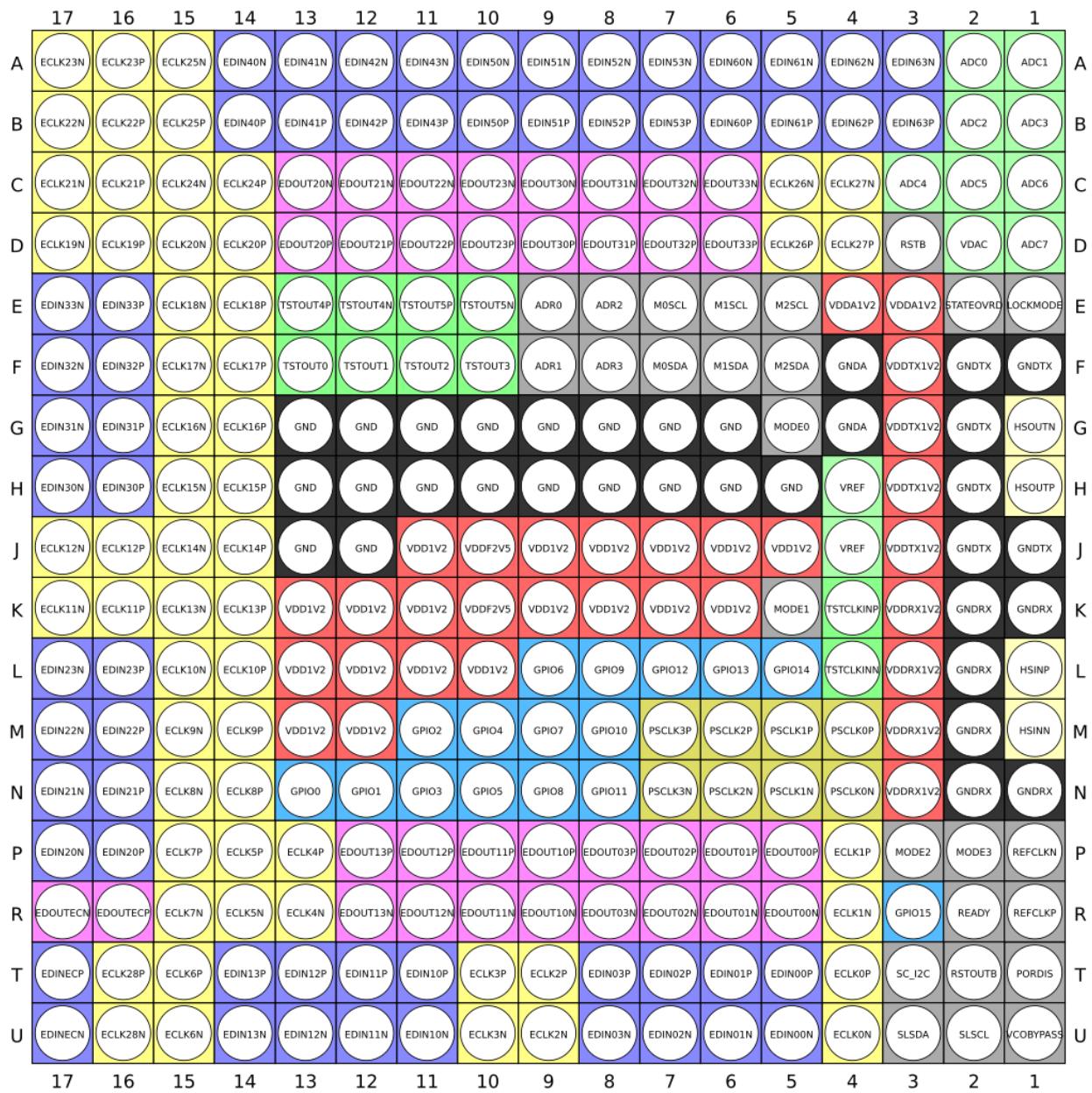


Fig. 17.6: IpGBT pinout (bottom view, balls up).

CHAPTER
EIGHTEEN

ELECTRICAL CHARACTERISTICS

All typical values are measured at $T = 25^\circ\text{C}$ unless other temperature condition is given. All minimum and maximum values are valid across operating temperature and voltage unless other conditions are given.

18.1 Absolute Maximum Ratings

Stresses beyond those listed in Table *Absolute maximum ratings* (page 309) may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or other conditions beyond those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Table 18.1: Absolute maximum ratings

Symbol	Parameter	Min	Max	Units
VDD	Power supply	-0.3	1.32	V
VDDA	Power supply	-0.3	1.32	V
VDDTX	Power supply	-0.3	1.32	V
VDDRX	Power supply	-0.3	1.32	V
VDDF2V5	Power supply	-0.3	2.75	V
TA	Storage temperature	??	??	°C
Tj	Junction temperature	-20	100	°C

18.2 General Operating Ratings

The device must operate within the ratings listed in Table *General Operating Ratings* (page 309) in order for all other electrical characteristics and typical characteristics of the device to be valid.

Table 18.2: General operating conditions.

Symbol	Parameter	Min	Typ	Max	Units
VDD	Power supply	1.08	1.2	1.32	V
VDDA	Power supply	1.08	1.2	1.32	V
VDDTX	Power supply	1.08	1.2	1.32	V
VDDRX	Power supply	1.08	1.2	1.32	V
VDDF2V5	Power supply	2.25	2.5	2.75	V
Tj	Junction temperature	-20		100	°C

18.3 Current consumption

18.4 CMOS I/O Pin Characteristics

Table 18.3: CMOS I/O pin characteristics.

Symbol	Parameter	Condition	Min.	Typ.	Max.	Units
IOH/IOL	I/O pin source/sink current	DS=1	-10		10	mA
		DS=0	-3		3	mA
VIH	High Level Input Voltage		0.3		VDD	V
VIL	Low Level Input Voltage		0		VDD-0.3	V
VOH	High Level Output Voltage	DS=1, Ioh = -10mA	VDD-0.13			V
		DS=1, Ioh = -1mA	VDD-0.02			V
		DS=0, Ioh = -3mA	VDD-0.13			V
		DS=0, Ioh = -1mA	VDD-0.04			V
VOL	Low Level Output Voltage	DS=1, Iol = 10mA			0.13	V
		DS=1, Iol = 1mA			0.02	V
		DS=0, Iol = 3mA			0.13	V
		DS=0, Iol = 1mA			0.04	V
Ileak	Input Leakage Current		-20		20	µA
Rpu	Pull Up Resistor			40		kΩ
Rpd	Pull Down Resistor			40		kΩ
Tr	Rise time (10-90%)	DS=1, Cload = 0 pF		0.7		ns
		DS=1, Cload = 100 pF		3.1		ns
		DS=0, Cload = 0 pF		0.7		ns
		DS=0, Cload = 100 pF		9.1		ns
Tf	Fall time (90-10%)	DS=1, Cload = 0 pF		0.7		ns
		DS=1, Cload = 100 pF		3.0		ns
		DS=0, Cload = 0 pF		0.7		ns
		DS=0, Cload = 100 pF		9.2		ns

18.5 eRX differential receiver

Symbol	Parameter	Condition	Min.	Typ.	Max.	Units
Vdin	Differential input voltage		140		450	mV
Vcm	Input Common Mode Voltage		0.07		1.13	V
Fmax	Data Rate				1.28	Gbps
Rin	Termination impedance			100		Ω
Idd	Current consumption	DC input	x?		mA	
Vcmbias	Bias generator voltage			0.6		V
Rbias	Bias generator output impedance			x?		kΩ

18.6 eTX differential driver

Table 18.4: eTX differential driver characteristics.

Symbol	Parameter	Condition	Min.	Typ.	Max.	Units
VAMP	Voltage amplitude	In 100 Ω load	100		400	mV
VCM	Common Mode Voltage			0.6		V
Tr	Rise time (10-90%)					ns
Tf	Fall time (90-10%)					ns

18.7 Clock and Oscillator Characteristics

Table 18.5: Clock conditions.

Symbol	Parameter/Conditions	Min.	Typ.	Max.	Units
Fref	System frequency	39	40.079	41	MHz

18.8 ADC characteristics

Table 18.6: ADC specifications.

Symbol	Parameter/Conditions	Min.	Typ.	Max.	Units
ENOB	Effective number of bits	9			Bits
VDDA	Analog power supply voltage	1.08	1.2	1.32	V
Idd	Average current consumption			1	mA
Temp	Operating temperature	-20	25	100	C
Ibias	Input bias current	-250		250	nA
INL	Integral non-linearity			1	LSB
DNL	Differential non-linearity			1	LSB
Fsmp	Conversion rate			100	kHz
Tinsel	Input selection time			100	μ s
Vref	Reference voltage	0.9	1.0	1.1	V

18.8.1 Single-ended mode (Gain 2x)

Table 18.7: ADC specifications for single-ended measurements (Gain x2, VREF = 1V).

Symbol	Parameter/Conditions	Min.	Typ.	Max.	Units
Vmin_g2	Minimum Input Voltage			0	V
Vmax_g2	Maximum Input Voltage	1.0			V

18.8.2 Differential mode (Gain 8x)

Table 18.8: ADC specifications for single-ended measurements (Gain x8, VREF = 1V).

Symbol	Parameter/Conditions	Min.	Typ.	Max.	Units
Vmin_g8	Minimum Input Voltage			-125	mV
Vmax_g8	Maximum Input Voltage	125			mV
Vcm	Maximum Input Voltage	0.5		0.7	V

18.8.3 Differential mode (Gain 16x)

Table 18.9: ADC specifications for single-ended measurements (Gain x16, VREF = 1V).

Symbol	Parameter/Conditions	Min.	Typ.	Max.	Units
Vmin_g16	Minimum Input Voltage			-62.5	mV
Vmax_g16	Maximum Input Voltage	62.5			mV
Vcm	Maximum Input Voltage	0.5		0.7	V

18.8.4 Differential mode (Gain 32x)

Table 18.10: ADC specifications for single-ended measurements (Gain x32, VREF = 1V).

Symbol	Parameter/Conditions	Min.	Typ.	Max.	Units
Vmin_g32	Minimum Input Voltage			-32.25	mV
Vmax_g32	Maximum Input Voltage	32.25			mV
Vcm	Maximum Input Voltage	0.5		0.7	V

18.9 VREF

18.9.1 Internal VREF generator (VREFEnable=1)

Table 18.11: Internal VREF generator (VREFEnable=1).

Symbol	Parameter/Conditions	Min.	Typ.	Max.	Units
VREF	Output voltage		1.0		V
LinReg	Line regulation		xx		μ V/V
LoadReg	Load regulation		xx		μ V/ μ A
Iout	Drive current capability	-YY		YY	μ A
Cmax	Load capacitance			10 (?)	pF
Ton	Turn on time			XX	ms
Vrms	Output noise voltage			XX	mV (RMS)
Isup	Supply current (No Load)		XX		μ A

18.9.2 External VREF voltage source (VREFEnable=0)

Table 18.12: External VREF voltage source (VREFEnable=0).

Symbol	Parameter/Conditions	Min.	Typ.	Max.	Units
VREF	Input voltage	0.9 (?)	1.0	1.1 (?)	V
Idc	Current consumption		XX		µA

18.10 Voltage DAC Specifications

convolute voltage follower

Table 18.13: Voltage DAC specifications.

Symbol	Parameter/Conditions	Min.	Typ.	Max.	Units
ENOB	Resolution	11 (?)	12		BITS
Idc	Current consumption			40 + XX	µA
INL	Integral non-linearity (No Load)	-2		2	LSB
DNL	Differential non-linearity (No Load)	-1		1	LSB

18.11 Brownout Detection Characteristics

18.12 Power-on Reset Characteristics

18.13 External Reset Characteristics

18.14 Eye Opening Monitor

Table 18.14: EOM specifications.

Symbol	Parameter/Conditions	Min.	Typ.	Max.	Units
Peom	Power consumption		10		mW
INLpi	INL of phase-interpolator		0.5	2	LSB
DNLpi	DNL of phase-interpolator		0.5	2	LSB

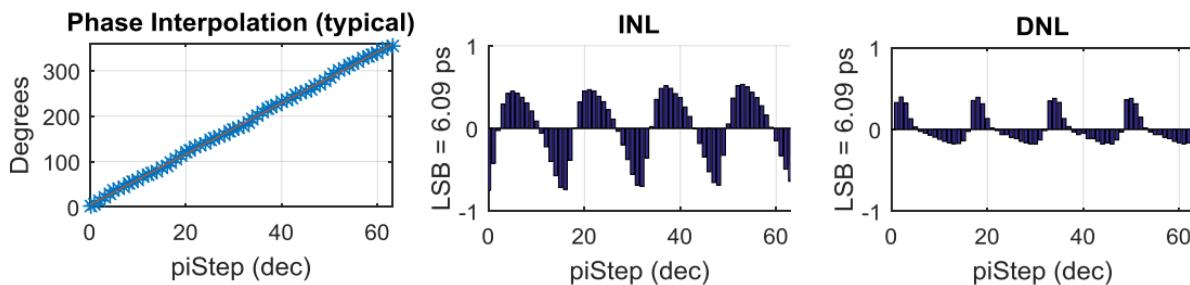


Fig. 18.1: INL and DNL of phase interpolated clock (simulation results)

KNOWN ISSUES

19.1 lpGBTv0

All issues listed below **will be** fixed in the next revision of the lpGBT chip.

19.1.1 Reset pin

Issue: Pin RSTN has an internal **pull-down** instead of pull-up.

Workarounds:

- Use external pull-up resistor (preferably below 4 k Ω). As the pad does not have Schmidt trigger at its input, it is not recommended to generate reset pulse with RC time constant.
- Actively drive the pin using CMOS buffer

19.1.2 I2C slave pins

Issue: Pins SLSDA, SLSCL have internal **pull-downs** instead of pull-ups.

Workaround: The value of the internal pull-down is relatively high (40 k Ω), which means than in most of the systems (where pull up should be of the order of k Ω) this will not create any problem. This issue is not a problem for systems where SC_I2C pin configures the chip to use serial configuration interface (see [Section 3.1.2](#)).

19.1.3 I2C slave signals transition times

Issue: Slow transition times on SLSDA and SLSCL pins can lead to unreliable operation of the I2C slave interface. The problems were observed for rise times of 300 ns. No studies have been done to estimate maximum transition times resulting in reliable operation so far.

Work around: It is recommended to use the lowest possible pull-up resistor values or to actively drive (high and low) both SLSDA and SLSCL lines.

19.1.4 I2C Masters pins

Issue: Pins M2SDA, M1SDA, M0SDA, M0SCL, M1SCL, M2SCL have **pull-downs** instead of pull-ups.

Workaround: Use external pull-up resistor with value significantly smaller than the internal 40 k Ω .

19.1.5 I2C Master 0 yield

Issue: An yield problem has been observed for I2CMaster0 in the first batch of ASICs.

Work around: It is recommended to user I2CMaster1 and I2CMaster2 whenever possible.

19.1.6 Phase-shifter

Issue: Phase-shifter channels 0 and 3 (PSCLK0N/P, PSCLK3N/P) are **under-performing**. Clock 0 and 3 have distorted duty cycle for all output frequencies. Moreover, for the highest clock frequency (1.28 GHz) some clock pulses may be missing for a specific settings of the fine delay.

Workaround: Use channels 1 and 2 if possible.

19.1.7 PRBS generator at the ePortRx Input

Issue: When using the PRBS generator included at the ePortRx input (see ref) with the highest clock frequency of 1.28 GHz, errors are present in the data stream. This issue affects only test mode and does not affect normal operation. The issue most likely is linked to the phase-shifter problem described in [Section 19.1.6](#).

Workaround: None.

19.1.8 Deterministic jitter

Issue: Deterministic jitter on all clock outputs (ECLK and Phase-shifted clocks).

Workaround: None.

19.1.9 Brownout detector mismatch

Issue: Brownout (power good) detector has a relatively large mismatch between triplicated instances.

Workaround: None.

19.1.10 IjCDR registers

Issue: The feed forward current value for the IjCDR is set to CLKGCDRFeedForwardPropCur regardless if the loop is locked or not (in locked state it should be set to CLKGCDRFeedForwardPropCurWhenLocked).

Workaround: None.

19.1.11 ePortRxEc registers

Issue: It is not possible to readout the state of the phase alignment process of the EC channel (EDINEC).

Workaround: None.

19.1.12 ePortRx phases swap

Issue: Phases 2 and 3 from the data delay line are swapped (see Fig. 7.6) for all ePortRx groups.

Workaround: When static phase selection is used the bug does not affect the operation of the ePortRx group in any way. Only for cases when the ePortRx continuously tracks the data phase and the automatically selected phase is close to the problematic codes (2-4) and the incoming signal jitter is very high (exceeding 0.5 UI) one may see data transmission errors. As a mitigation one can use **Continuous phase tracking with initial phase** with the initial phase value of 12.

19.1.13 ePorts time alignment

Issue: eLink clocks are not very precisely aligned. The skew can exceed XX ps.

Workaround: None.

19.1.14 Fusing updateEnable bit

Issue: Once updateEnable bit in the *[0x0ef] POWERUP2* (page 226) register is blown no other fuse can be blown.

Workaround: Burn the updateEnable bit only after all other registers are burned as required.

FREQUENTLY ASKED QUESTIONS

20.1 Can I use eLinks receivers at 80 Mbps?

The only eLink receiver natively capable of receiving data stream at 80 Mbps is EDINEC. If more than one 80 Mbps link is required, one can oversample the data stream with the eLink operating at 160 Mbps (or even higher). Unfortunately, this solution has several disadvantages:

- As every bit is sampled twice, the data stream occupies twice the bandwidth and an additional processing (de-duplication) is required on the FPGA side.
- The tuning range of the phase aligner is not sufficient to offer an efficient automatic phase selection it is recommended to use fix the phase aligner to static phase selection mode.

20.2 Does IpGBT have a master SPI interface?

In general no. However, one can use PIO (see [Section 11](#)) bit banging to emulate SPI interface. Of course, this implementation offers limited clock frequency.

20.3 Does IpGBT have a master JTAG interface?

In general no. However, one can use PIO (see [Section 11](#)) bit banging to emulate JTAG interface. Of course, this implementation offers limited clock frequency.

20.4 I need more I2C Masters, what can I do?

One can use PIO (see [Section 11](#)) bit banging to emulate I2C master interface. Of course, this implementation offers limited clock frequency.

20.5 Does IpGBT have a slave JTAG interface (scan chain or boundary scan)?

No.

CHAPTER
TWENTYONE

VERSION HISTORY

Date	Author	Description
11 Jan 2019 (09:31:52)		Version 0.1 released

CHAPTER
TWENTYTWO

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