

Figure 1: 1-dimensional continuous-time signals

## 1 Signals and systems

A **signal** is a function mapping an input variable to some output variable. For example

$$\sin(\pi t), \quad \frac{1}{2}t^3, \quad e^{-t^2}$$

all represent **signals** with input variable  $t \in \mathbb{R}$ , and they are plotted in Figure 1. If  $x$  is a signal and  $t$  an input variable we write  $x(t)$  for the output variable. Signals can be multidimensional. This page is an example of a 2-dimensional signal, the independent variables are the horizontal and vertical position on the page, and the signal maps this position to a colour, in this case either black or white. A moving image such as seen on your television or computer monitor is an example of a 3-dimensional signal, the three independent variables being vertical and horizontal screen position and time. The signal maps each position and time to a colour on the screen. In this course we focus exclusively on 1-dimensional signals such as those in Figure 1 and we will only consider signals that are real or complex valued.

### 1.1 Properties of signals

A signal  $x$  is **bounded** if there exists a real number  $M$  such that

$$|x(t)| \leq M \quad \text{for all } t \in \mathbb{R}$$

where  $|\cdot|$  denotes the complex magnitude. Both  $\sin(\pi t)$  and  $e^{-t^2}$  are examples of bounded signals because  $|\sin(\pi t)| \leq 1$  and  $|e^{-t^2}| \leq 1$  for all  $t \in \mathbb{R}$ . However,  $\frac{1}{2}t^3$  is not bounded because its magnitude grows indefinitely as  $t$  moves away from the origin.

A signal  $x$  is **periodic** if there exists a real number  $T$  such that

$$x(t) = x(t + kT) \quad \text{for all } k \in \mathbb{Z} \text{ and } t \in \mathbb{R}.$$

For example, the signal  $\sin(\pi t)$  is periodic with period  $T = 2$ . Neither  $\frac{1}{2}t^3$  or  $e^{-t^2}$  are periodic.

A function  $x : \mathbb{R} \rightarrow \mathbb{R}$  is called **locally integrable** if for all constants  $a$  and  $b$ ,

$$\int_a^b |x(t)| dt$$

exists. In this course we always assume that signals are locally integrable, that is, signals are locally integrable functions mapping  $\mathbb{R} \rightarrow \mathbb{R}$  or  $\mathbb{R} \rightarrow \mathbb{C}$ . Two signals  $x$  and  $y$  are equal, i.e.  $x = y$  if  $x(t) = y(t)$  for all  $t \in \mathbb{R}$ .

A signal  $x$  is called **absolutely integrable** if

$$\|x\|_1 = \int_{-\infty}^{\infty} |x(t)| dt \quad (1.1)$$

exists (evaluates to a finite number). Here we introduce the notation  $\|x\|_1$  called the  $\ell_1$ -**norm** of  $x$ . For example  $\sin(\pi t)$  and  $\frac{1}{2}t^3$  are not absolutely integrable, but  $e^{-t^2}$  is because

$$\int_{-\infty}^{\infty} |e^{-t^2}| dt = \int_{-\infty}^{\infty} e^{-t^2} dt = \sqrt{\pi}. \quad (1.2)$$

The signal  $x$  is called **square integrable** if

$$\|x\|_2 = \int_{-\infty}^{\infty} |x(t)|^2 dt$$

exists. Square integrable signals are also called **energy signals**, and the value of  $\|x\|_2$  is called the **energy** of  $x$  (it is also called the  $\ell_2$ -**norm** of  $x$ ). For example  $\sin(\pi t)$  and  $\frac{1}{2}t^3$  are not energy signals, but  $e^{-t^2}$  is.

A signal  $x$  is **right sided** if there exists a  $T \in \mathbb{R}$  such that  $x(t) = 0$  for all  $t < T$ . Correspondingly  $x$  is **left sided** if  $x(t) = 0$  for all  $T > t$ . For example, the signal

$$u(t) = \begin{cases} 1 & t > 0, \\ 0 & t \leq 0 \end{cases} \quad (1.3)$$

called the **step function** is right-sided. Its reflection in time  $u(-t)$  is left sided (Figure 2). A signal  $x$  is called **finite in time** if it is both left and right sided, that is, if there exists a  $T \in \mathbb{R}$  such that  $x(t) = x(-t) = 0$  for all  $t > T$ . A signal is called **unbounded in time** if it is neither left nor right sided. For example, the continuous time signals  $\sin(\pi t)$  and  $e^{-t^2}$  are unbounded in time, but the signal

$$\Pi(t) = \begin{cases} 1 & -\frac{1}{2} < t \leq \frac{1}{2} \\ 0 & \text{otherwise,} \end{cases} \quad (1.4)$$

called the rectangle function is finite in time.

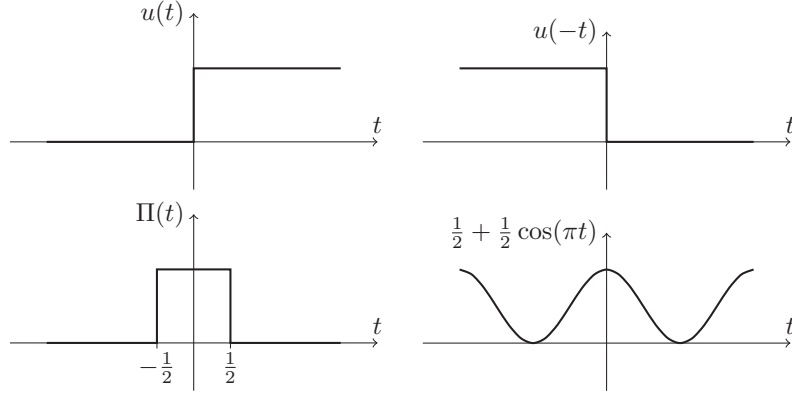


Figure 2: The right sided step function  $u(t)$ , its left sided reflection  $u(-t)$ , the finite in time rectangular pulse  $\Pi(t)$  and the unbounded in time signal  $\frac{1}{2} + \frac{1}{2} \cos(x)$ .

## 1.2 Systems (functions of signals)

A **system** (also known as an **operator** or **functional**) maps a signal to another signal. For example

$$x(t) + 3x(t-1), \quad \int_0^1 x(t-\tau) d\tau, \quad \frac{1}{x(t)}, \quad \frac{d}{dt}x(t)$$

represent systems, each mapping the signal  $x$  to another signal. Consider the electric circuit in Figure 3 called a **voltage divider**. If the voltage at time  $t$  is  $x(t)$  then, by Ohm's law, the current at time  $t$  satisfies

$$i(t) = \frac{1}{R_1 + R_2} x(t).$$

and the voltage over the resistor  $R_2$  is

$$y(t) = R_2 i(t) = \frac{R_2}{R_1 + R_2} x(t) \quad (1.5)$$

The circuit can be considered as a system mapping the signal  $x$  representing the voltage to the signal  $i = \frac{1}{R_1 + R_2} x$  representing the current, or a system mapping  $x$  to the signal  $y = \frac{R_2}{R_1 + R_2} x$  representing the voltage over resistor  $R_2$ .

We denote systems with capital letters such as  $H$  and  $G$ . A system  $H$  is a function that maps a signal  $x$  to another signal denoted  $H(x)$ . We call  $x$  the **input signal** and  $H(x)$  the **output signal** or the **response** of system  $H$  to signal  $x$ . If we want to include the independent variable  $t$  we will write  $H(x)(t)$  or  $H(x, t)$  and do not distinguish between these [Curry and Feys, 1968]. It is sometimes useful to depict systems with a block diagram. Figure 4 is simple block diagram showing the input signal and output signals of a system  $H$ .

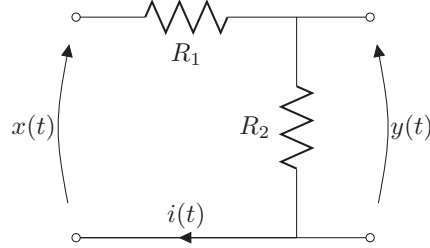


Figure 3: A **voltage divider** circuit.

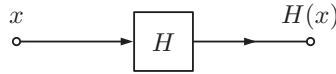


Figure 4: System block diagram with input signal  $x$  and output signal  $H(x)$ .

Using this notation the voltage divider circuit in Figure 3 corresponds with the system

$$H(x) = \frac{R_2}{R_1 + R_2} x = y$$

This system multiplies the input signal  $x$  by  $\frac{R_2}{R_1 + R_2}$ . This brings us to our first practical test.

**Test 1. (Voltage divider)** In this test we construct the voltage divider from Figure 3 on a breadboard with resistors  $R_1 = 100\Omega$  and  $R_2 = 470\Omega$ . Using a computer soundcard (an approximation of) the voltage signal

$$x(t) = \sin(2\pi f_1 t) \quad \text{with} \quad f_1 = 100$$

is passed through the circuit. The approximation is generated by sampling  $x(t)$  at rate  $F_s = \frac{1}{T_s} = 44100$  to generate samples

$$x_n = x(nT_s) \quad n = 0, \dots, 2F_s$$

corresponding to approximately 2 seconds of signal. These samples are passed to the soundcard which starts playback. The voltage over the resistor  $R_2$  is recorded (also using the soundcard) that returns a lists of samples  $y_1, \dots, y_L$  taken at rate  $F_s$ . The continuous-time voltage over the capacitor can be (approximately) reconstructed from these samples as

$$\tilde{y}(t) = \sum_{\ell=1}^L y_\ell \text{sinc}(t - F_s \ell) \quad (1.6)$$

where

$$\text{sinc}(t) = \frac{\sin(\pi t)}{\pi t} \quad (1.7)$$

is called the **sinc function**. We will justify this reconstruction in Section 7. Simultaneously the (stereo) soundcard is used to record the input voltage  $x(t)$  producing samples  $x_1, \dots, x_L$  taken at rate  $F_s$ . An approximation of the continuous-time input signal is

$$\tilde{x}(t) = \sum_{\ell=1}^L x_\ell \text{sinc}(t - F_s \ell). \quad (1.8)$$

In view of (1.5) we would expect the approximate relationship

$$\tilde{y} = \frac{R_2}{R_1 + R_2} \tilde{x} = \frac{47}{57} \tilde{x}.$$

However, the output resistance of the soundcard itself can not be ignored. Appendix A describes a method for estimating this resistance that, with this hardware used for this test, is approximately  $R_o = 250\Omega$ . This resistance is in series with  $R_1$  and so, we expect the approximate relationship

$$\tilde{y} \approx \frac{R_2}{R_1 + R_o + R_2} \tilde{x} = \frac{47}{82} \tilde{x}.$$

A plot of  $\tilde{y}$ ,  $\tilde{x}$  and  $\frac{47}{82} \tilde{x}$  over a 20ms period from 1s to 1.02s is given in Figure 5. Observe that  $\tilde{y}$  is indeed close to  $\frac{47}{82} \tilde{x}$ .

Not all systems apply to all signals. For example the system

$$H(x, t) = \frac{1}{x(t)}$$

is not defined at those  $t$  where  $x(t) = 0$  because we cannot divide by zero. Another example is the system

$$I_\infty(x, t) = \int_{-\infty}^t x(\tau) d\tau, \quad (1.9)$$

called an **integrator**, that is not defined for those signals where the integral above does not exist (is not finite). For example, the signal  $x(t) = 1$  cannot be applied with the integrator since the integral  $\int_{-\infty}^t dt$  does not exist.

Thus, when specifying a system it is necessary to also specify a set of signals to which the system can be applied. For example, the system  $H(x, t) = \frac{1}{x(t)}$  can be applied only to those signals  $x(t)$  which are not zero for any  $t$ . The integrator  $I_\infty(x, t)$  can be applied only to those  $x$  where the integral  $\int_{-\infty}^t x(\tau) d\tau$  exists. The set of signals associated with a given system is usually obvious from the specification of the system itself. For this reason we will not usually state the

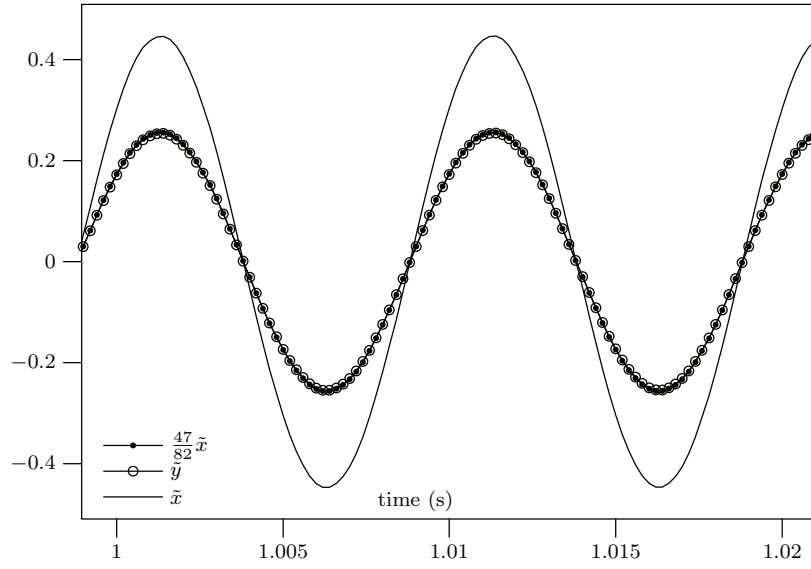


Figure 5: Plot of reconstructed input signal  $\tilde{x}$  (solid line), output signal  $\tilde{y}$  (solid line with circle) and hypothesised output signal  $\frac{47}{82}\tilde{x}$  (solid line with dot) over a 20ms time window for the voltage divider circuit in Figure 3. The input signal is  $x(t) = \sin(200\pi t)$ .

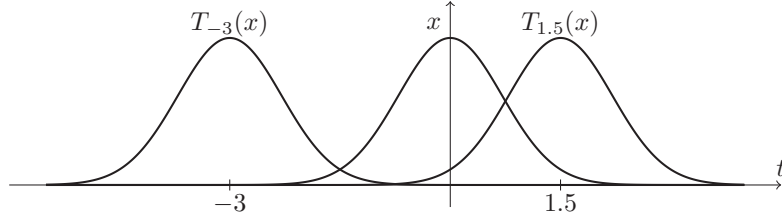


Figure 6: Time-shifter system  $T_{1.5}(x, t) = x(t - 1.5)$  and  $T_{-3}(x, t) = x(t + 3)$  acting on the signal  $x(t) = e^{-t^2}$ .

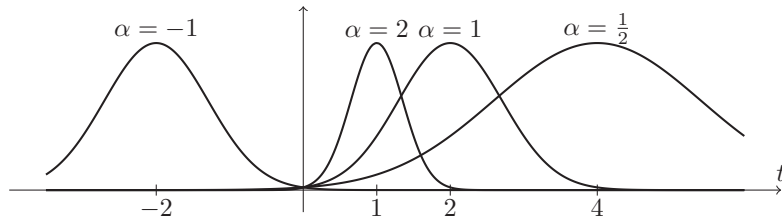


Figure 7: Time-scaler system  $H(x, t) = x(\alpha t)$  for  $\alpha = -1, \frac{1}{2}, 1$  and  $2$  acting on the signal  $x(t) = e^{-(t-2)^2}$ .

set of signals that apply to a given system. We will only do so if there is chance for confusion.

### 1.3 Some important systems

We call the system

$$T_\tau(x, t) = x(t - \tau)$$

a **time-shifter**. This system shifts the input signal along the  $t$  axis (‘time’ axis) by  $\tau$ . When  $\tau$  is positive  $T_\tau$  delays the input signal by  $\tau$ . The time-shifter will appear so regularly in this course that we use the special notation  $T_\tau$  to represent it. Figure 6 depicts the action of time-shifters  $T_{1.5}$  and  $T_{-3}$  on the signal  $x(t) = e^{-t^2}$ . When  $\tau = 0$  the time-shifter is the **identity system**

$$T_0(x) = x$$

that maps the signal  $x$  to itself.

Another important system is the **time-scaler** that has the form

$$H(x, t) = x(\alpha t)$$

for  $\alpha \in \mathbb{R}$ . Figure 7 depicts the action of a time-scaler with a number of values for  $\alpha$ . When  $\alpha = -1$  the time-scaler reflects the input signal in the time axis.

Another system we regularly encounter is the **differentiator**

$$D(x, t) = \frac{d}{dt}x(t).$$

the returns the derivative of the input signal. We also define a  $k$ th differentiator

$$D^k(x, t) = \frac{d^k}{dt^k}x(t)$$

that returns the  $k$ th derivative of the input signal.

Another important system is the **integrator**

$$I_a(x, t) = \int_{-a}^t x(\tau) d\tau.$$

The parameter  $a$  describes the lower bound of the integral. In this course it will often be that  $a = \infty$  or  $a = 0$ . The integrator can only be applied to those signals for which the integral above exists. For example, the integrator  $I_{-\infty}$  can be applied to the signal  $tu(t)$  where  $u(t)$  is the step function (1.3). The output signal is

$$\int_{-\infty}^t \tau u(\tau) d\tau = \int_0^t \tau d\tau = \frac{t^2}{2}.$$

However, the integrator cannot be applied to the signal  $x(t) = t$  because  $\int_{-\infty}^t \tau d\tau$  does not exist.

## 1.4 Properties of systems

A system  $H$  is called **memoryless** if the output signal  $H(x)$  at time  $t$  depends only on the input signal  $x$  at time  $t$ . For example  $\frac{1}{x(t)}$  and the identity system  $T_0$  are memoryless, but

$$x(t) + 3x(t-1) \quad \text{and} \quad \int_0^1 x(t-\tau) d\tau$$

are not. A time-shifter system  $T_\tau$  with  $\tau \neq 0$  is not memoryless.

A system  $H$  is **causal** if the output signal  $H(x)$  at time  $t$  depends on the input signal only at times less than or equal to  $t$ . Memoryless systems such as  $\frac{1}{x(t)}$  and  $T_0$  are also causal. Time-shifters  $T_\tau(x, t) = x(t-\tau)$  are causal when  $\tau \geq 0$ , but are not causal when  $\tau < 0$ . The systems

$$x(t) + 3x(t-1) \quad \text{and} \quad \int_0^1 x(t-\tau) d\tau$$

are causal, but the systems

$$x(t) + 3x(t+1) \quad \text{and} \quad \int_0^1 x(t+\tau) d\tau$$



are not causal.

A system  $H$  is called **bounded-input-bounded-output stable** or **stable** if the output signal  $H(x)$  is bounded whenever the input signal  $x$  is bounded. That is,  $H$  is stable if for every real number  $M$  there exists a real number  $K$  such that for all signals  $x$  satisfying

$$|x(t)| < M \quad \text{for all } t \in \mathbb{R},$$

it also holds that

$$|H(x, t)| < K \quad \text{for all } t \in \mathbb{R}.$$

For example, the system  $x(t) + 3x(t-1)$  is stable with  $K = 4M$  since if  $|x(t)| < M$  then

$$|x(t) + 3x(t-1)| \leq |x(t)| + 3|x(t-1)| < 4M = K.$$

The integrator  $I_a$  for any  $a \in \mathbb{R}$  and differentiator  $D$  are not stable (Exercises 4 and 5).

A system  $H$  is **linear** if

$$H(ax + by) = aH(x) + bH(y)$$

for all signals  $x$  and  $y$ , and for all complex numbers  $a$  and  $b$ . That is, a linear system has the property: If the input consists of a weighted sum of signals, then the output consists of the same weighted sum of the responses of the system to those signals. Figure 21 indicates the linearity property using a block diagram. For example, the differentiator is linear because

$$\begin{aligned} D(ax + by, t) &= \frac{d}{dt}(ax(t) + by(t)) \\ &= a \frac{d}{dt}x(t) + b \frac{d}{dt}y(t) \\ &= aD(x, t) + bD(y, t), \end{aligned}$$

but the system  $H(x, t) = \frac{1}{x(t)}$  is not linear because

$$H(ax + by, t) = \frac{1}{ax(t) + by(t)} \neq \frac{a}{x(t)} + \frac{b}{y(t)} = aH(x, t) + bH(y, t)$$

in general.

The property of linearity trivially generalises to more than two signals. For example if  $x_1, \dots, x_k$  are signals and  $a_1, \dots, a_k$  are complex numbers for some finite  $k$ , then

$$H(a_1x_1 + \dots + a_kx_k) = a_1H(x_1) + \dots + a_kH(x_k).$$

A system  $H$  is **time-invariant** if

$$H(T_\tau(x), t) = H(x, t - \tau)$$

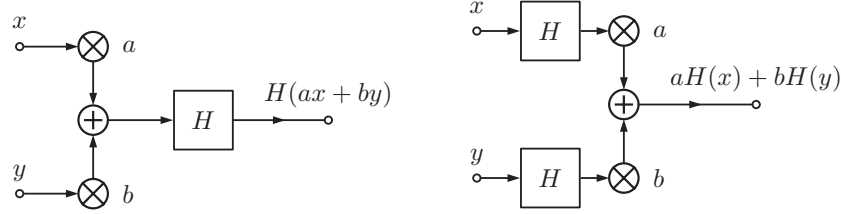


Figure 8: If  $H$  is a linear system the outputs of these two diagrams are the same signal, i.e.  $H(ax + by) = aH(x) + bH(y)$ .

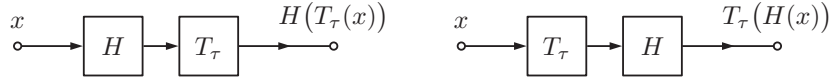


Figure 9: If  $H$  is a time-invariant system the outputs of these two diagrams are the same signal, i.e.  $H(T_\tau(x)) = T_\tau(H(x))$ .

for all signals  $x$  and all time-shifts  $\tau \in \mathbb{R}$ . That is, a system is time-invariant if time-shifting the input signal results in the same time-shift of the output signal. Equivalently,  $H$  is time-invariant if  $H$  commutes the time-shifter  $T_\tau$ , that is, if

$$H(T_\tau(x)) = T_\tau(H(x))$$

for all  $\tau \in \mathbb{R}$  and all signals  $x$ . Figure 9 represents the property of time-invariance with a block diagram.

Let  $S$  be a set of signals. A system  $H$  is said to be **invertible** on  $S$  if each signal  $x \in S$  is mapped to a unique signal  $H(x)$ . That is, for all signals  $x, y \in S$  then  $H(x) = H(y)$  if and only if  $x = y$ . If a system  $H$  is invertible on  $S$  then there exists an inverse system  $H^{-1}$  such that

$$x = H^{-1}(H(x)) \quad \text{for all } x \in S.$$

For example, let  $S$  be the set of differentiable signals. The differentiator system  $D$  is **not** invertible on  $S$  because if  $x \in S$  and if  $y(t) = x(t) + c$  for any constant  $c$  then  $D(y) = D(x)$ . However, if we restrict  $S$  to those differentiable signals for which  $x(0) = c$  is fixed, then  $D$  is invertible on  $S$ . The inverse system in this case is

$$D^{-1}(x, t) = I_0(x, t) + c = \int_0^\infty x(t) dt + c$$

because

$$D^{-1}(D(x), t) = \int_0^\infty D(x, t) dt + c = \int_0^\infty \frac{d}{dt} x(t) dt + x(0) = x(t)$$

by the fundamental theorem of calculus.

## 1.5 Exercises

1. State whether the step function  $u(t)$  is bounded, periodic, continuous, differentiable, absolutely summable, an energy signal.
2. Show that the function  $t^2$  is locally integrable, but that the function  $\frac{1}{t^2}$  is not.
3. Compute the energy of the signals  $e^{-t^2}$  and  $e^{-t^2/4}$  (Hint: use (1.2) and a change of variables).
4. Show that the integrator  $I_a$  for any  $a \in \mathbb{R}$  is not bibo stable.
5. Show that the differentiator system  $D$  is not bibo stable.
6. Show that the time-shifter is linear and time-invariant system, but that the time-scaler is neither linear or time invariant
7. Show that the  $k$ th differentiator  $D^k(x, t) = \frac{d^k}{dt^k}x(t)$  is linear and time-invariant
8. State whether the system  $H(x, t) = x(t) + 1$  is linear, time-invariant, bibo stable.
9. State whether the system  $H(x, t) = 0$  is linear, time-invariant, bibo stable.
10. State whether the system  $H(x, t) = 1$  is linear, time-invariant, bibo stable.